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LMH6672

Dual, High Output Current, High Speed Op Amp

General Description

The LMH6672 is a low cost, dual high speed op amp capable of driving signals to within 1V of the power supply rails. It features the high output drive with low distortion required for the demanding application of a single supply xDSL line driver.

When connected as a differential output driver, the LMH6672 can drive a 50Ω load to 16.8 V_{PP} swing with only -98 dBc distortion, fully supporting the peak upstream power levels for upstream full-rate ADSL. The LMH6672 is fully specified for operation with 5V and 12V supplies. Ideal for PCI modem cards and xDSL modems.

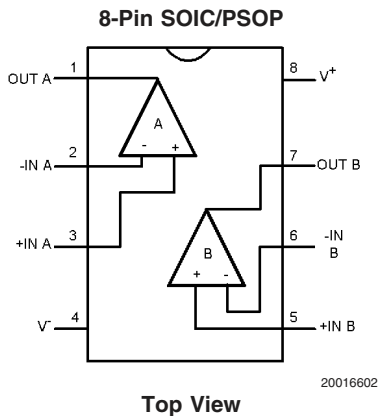
Applications

- ADSL PCI modem cards
- xDSL external modems
- Line drivers

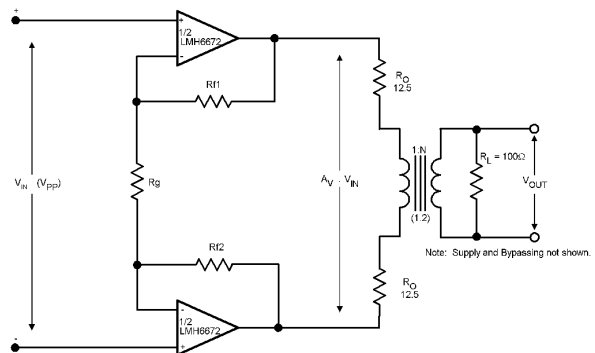
Features

- **High Output Drive**
19.2 V_{PP} differential output voltage, R_L = 50Ω
9.6 V_{PP} single-ended output voltage, R_L = 25Ω
- **High Output Current**
±200 mA @ V_O = 9 V_{PP}, V_S = 12V
- **Low Distortion**
105 dB SFDR @ 100 kHz, V_O = 8.4 V_{PP}, R_L = 25Ω
98 dB SFDR @ 1MHz, V_O = 2 V_{PP}, R_L = 100Ω
- **High Speed**
90 MHz 3 dB bandwidth (G = 2)
135 V/μs slew rate
- **Low Noise**
3.1 nV/√Hz : input noise voltage
1.8 pA/√Hz : input noise current
- Low supply current: 7.2mA/amp
- Single-supply operation: 5V to 12V
- Available in 8-pin SOIC and PSOP

Connection Diagram



Typical Application



Ordering Information

Package	Part Number	Package Marking	Transport Media	NSC Drawing
8-Pin SOIC	LMH6672MA	LMH6672MA	Rails	M08A
	LMH6672MAX	LMH6672MA	2.5k Units Tape and Reel	
8-Pin PSOP	LMH6672MR	LMH6672MR	Rails	MRA08A
	LMH6672MRX	LMH6672MR	2.5k Units Tape and Reel	

Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

ESD Tolerance	(Note 2)
Human Body Model	2kV
Machine Model	200V
V_{IN} Differential	$\pm 1.2V$
Output Short Circuit Duration	(Note 3)
Supply Voltage ($V^+ - V^-$)	13.2V
Voltage at Input/Output pins	$V^+ +0.8V, V^- -0.8V$
Storage Temperature Range	$-65^\circ C$ to $+150^\circ C$

Junction Temperature	$+150^\circ C$ (Note 4)
Soldering Information	
Infrared or Convection (20 sec)	$235^\circ C$
Wave Soldering (10 sec)	$260^\circ C$

Operating Ratings (Note 1)

Supply Voltage ($V^+ - V^-$)	$\pm 2.5V$ to $\pm 6.5V$
Junction Temperature Range	$-40^\circ C$ to $150^\circ C$
Package Thermal Resistance (θ_{JA})	
8-pin SOIC	$172^\circ C/W$
8-pin PSOP	$58.6^\circ C/W$

Electrical Characteristics

$T_J = 25^\circ C, G = +2, V_S = \pm 2.5$ to $\pm 6V, R_F = R_{IN} = 470\Omega, R_L = 100\Omega$; Unless otherwise specified.

Symbol	Parameter	Conditions	Min (Note 6)	Typ (Note 5)	Max (Note 6)	Units
Dynamic Performance						
	-3dB Bandwidth			90		MHz
	0.1dB Bandwidth	$V_S = \pm 6V$		12		MHz
	Slew Rate	$V_S = \pm 6V, 4V$ Step, 10-90%		135		V/ μs
	Rise and Fall Time	$V_S = 6V, 4V$ Step, 10-90%		23.5		ns
Distortion and Noise Response						
	2 nd Harmonic Distortion	$V_O = 8.4 V_{PP}, f = 100$ kHz, $R_L = 25\Omega$		-105		dBc
		$V_O = 8.4 V_{PP}, f = 1$ MHz, $R_L = 100\Omega$		-90		dBc
	3 rd Harmonic Distortion	$V_O = 8.4 V_{PP}, f = 100$ kHz, $R_L = 25\Omega$		-110		dBc
		$V_O = 8.4 V_{PP}, f = 1$ MHz, $R_L = 100\Omega$		-87		dBc
	Input Noise Voltage	$f = 100$ kHz		3.1		nV \sqrt{Hz}
	Input Noise Current	$f = 100$ kHz		1.8		pA \sqrt{Hz}
Input Characteristics						
V_{OS}	Input Offset Voltage	$T_J = -40^\circ C$ to $125^\circ C$	-5.5	0.1	5.5	mV
			-4	-0.2	4	
I_B	Input Bias Current	$T_J = -40^\circ C$ to $125^\circ C$		8	16	μA
I_{OS}	Input Offset Current	$T_J = -40^\circ C$ to $125^\circ C$	-2.1	0	2.1	μA
CMVR	Common Voltage Range	$V_S = \pm 6V$	-6.0	-5.7 to 4.5	4.5	V
CMRR	Common-Mode Rejection Ratio	$V_S = \pm 6V, T_J = -40^\circ C$ to $125^\circ C$	150	7.5		$\mu V/V$
Transfer Characteristics						
A_{VOL}	Voltage Gain	$R_L = 1k, T_J = -40^\circ C$ to $125^\circ C$	1.0	5		V/mV
		$R_L = 25\Omega, T_J = -40^\circ C$ to $125^\circ C$	0.67	3.4		V/mV
V_O	Output Swing	$R_L = 25\Omega, V_S = \pm 6V$	-4.5	± 4.8	4.5	V
		$R_L = 25\Omega, T_J = -40^\circ C$ to $125^\circ C, V_S = \pm 6V$	-4.4	± 4.8	4.4	
V_O	Output Swing	$R_L = 1k, V_S = \pm 6V$	-4.8	± 4.8	4.8	V
		$R_L = 1k, T_J = -40^\circ C$ to $125^\circ C, V_S = \pm 6V$	-4.7	± 4.8	4.7	

Electrical Characteristics (Continued)

$T_J = 25^\circ\text{C}$, $G = +2$, $V_S = \pm 2.5$ to $\pm 6\text{V}$, $R_F = R_{IN} = 470\Omega$, $R_L = 100\Omega$; Unless otherwise specified.

Symbol	Parameter	Conditions	Min (Note 6)	Typ (Note 5)	Max (Note 6)	Units
I_{SC}	Output Current (Note 3)	$V_O = 0$, $V_S = \pm 6\text{V}$	350	525		mA
		$V_O = 0$, $V_S = \pm 6\text{V}$, $T_J = -40^\circ\text{C}$ to 125°C	260	600		mA
Power Supply						
I_S	Supply Current/Amp	$V_S = \pm 6\text{V}$			8	mA
		$V_S = \pm 6\text{V}$, $T_J = -40^\circ\text{C}$ to 125°C		7.2	9	
PSRR	Power Supply Rejection Ratio	$V_S = \pm 2.5\text{V}$ to $\pm 6\text{V}$, $T_J = -40^\circ\text{C}$ to 125°C	72	88.5		dB

 $\pm 2.5\text{V}$ Electrical Characteristics

$T_J = 25^\circ\text{C}$, $G = +2$, $V_S = \pm 2.5$ to $\pm 6\text{V}$, $R_F = R_{IN} = 470\Omega$, $R_L = 100\Omega$; Unless otherwise specified.

Symbol	Parameter	Conditions	Min (Note 6)	Typ (Note 5)	Max (Note 6)	Units
Dynamic Performance						
	-3 dB Bandwidth			80		MHz
	0.1 dB Bandwidth			12		MHz
	Slew Rate	2V Step, 10-90%		15		V/ μs
	Rise and Fall Time	2V Step, 10-90%		14		ns
Distortion and Noise Response						
	2 nd Harmonic Distortion	$V_O = 2 V_{PP}$, $f = 100\text{ kHz}$, $R_L = 25\Omega$		-96		dBc
		$V_O = 2 V_{PP}$, $f = 1\text{ MHz}$, $R_L = 100\Omega$		-85		dBc
	3 rd Harmonic Distortion	$V_O = 2 V_{PP}$, $f = 100\text{ kHz}$, $R_L = 25\Omega$		-98		dBc
		$V_O = 2 V_{PP}$, $f = 1\text{ MHz}$, $R_L = 100\Omega$		-87		dBc

Input Characteristics						
V_{OS}	Input Offset Voltage	$T_J = -40^\circ\text{C}$ to 125°C	-5.5		5.5	mV
			-4.0	0.02	4.0	
I_B	Input Bias Current	$T_J = -40^\circ\text{C}$ to 125°C		8.0	16	μA
CMVR	Common-Mode Voltage Range		-2.5		1.0	V
CMRR	Common-Mode Rejection Ratio	$T_J = -40^\circ\text{C}$ to 125°C	150	8		$\mu\text{V/V}$

Transfer Characteristics						
A_{VOL}	Voltage Gain	$R_L = 25\Omega$, $T_J = -40^\circ\text{C}$ to 125°C	0.67	3		V/mV
		$R_L = 1\text{k}$, $T_J = -40^\circ\text{C}$ to 125°C	1.0	4		

Output Characteristics						
V_O	Output Voltage Swing	$R_L = 25\Omega$	1.20	1.45		V
		$R_L = 25\Omega$, $T_J = -40^\circ\text{C}$ to 125°C	1.10	1.35		
		$R_L = 1\text{k}$	1.30	1.60		
		$R_L = 1\text{k}$, $T_J = -40^\circ\text{C}$ to 125°C	1.25	1.50		

Power Supply						
I_S	Supply Current/Amp				8.0	mA
		$T_J = -40^\circ\text{C}$ to 125°C		6.7	9.0	

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.

Note 2: Human body model, 1.5k Ω in series with 100pF. Machine model, 200 Ω in series with 100pF.

Note 3: Shorting the output to either supply or ground will exceed the absolute maximum T_J and can result in failure.

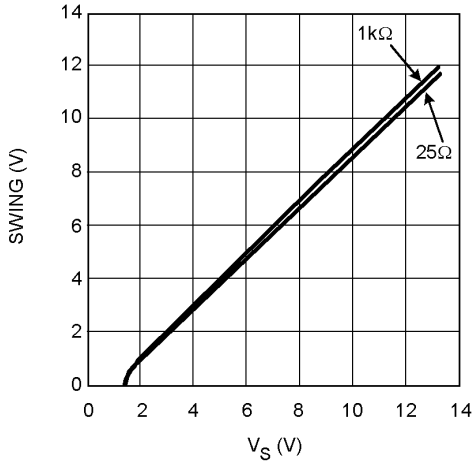
Note 4: The maximum power dissipation is a function of $T_{J(MAX)}$, θ_{JA} and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} - T_A)/\theta_{JA}$. All numbers apply for packages soldered directly onto a PC board.

Note 5: Typical values represent the most likely parametric norm.

Note 6: All limits are guaranteed by testing, characterization or statistical analysis.

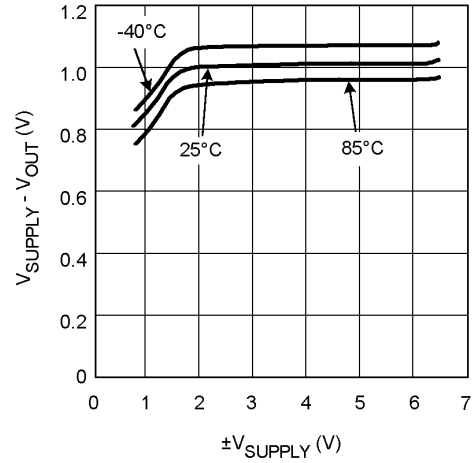
Typical Performance Characteristics

Output Swing $R_L = 25\Omega, 1\text{ k}\Omega$ @ $-40^\circ\text{C}, 25^\circ\text{C}, 85^\circ\text{C}$



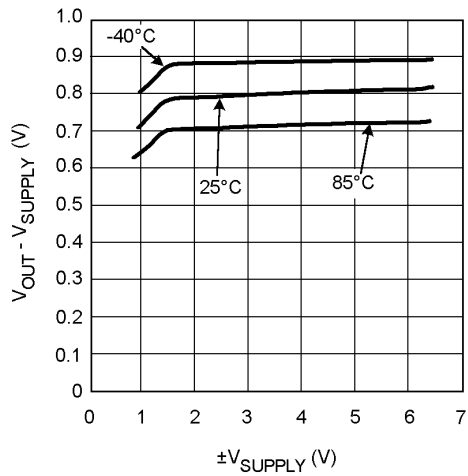
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Positive Output Swing into $1\text{ k}\Omega$



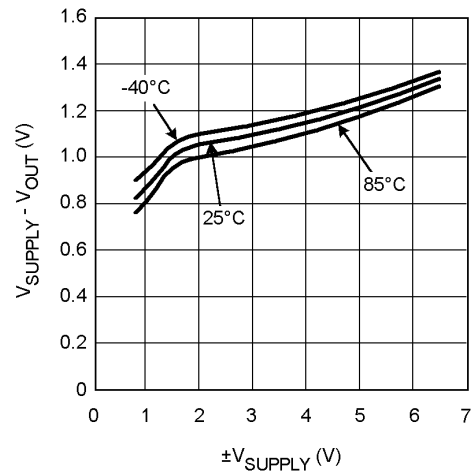
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Negative Output Swing into $1\text{ k}\Omega$



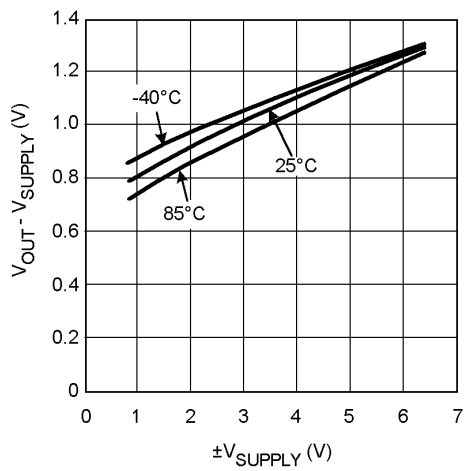
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Positive Output Swing into 25Ω



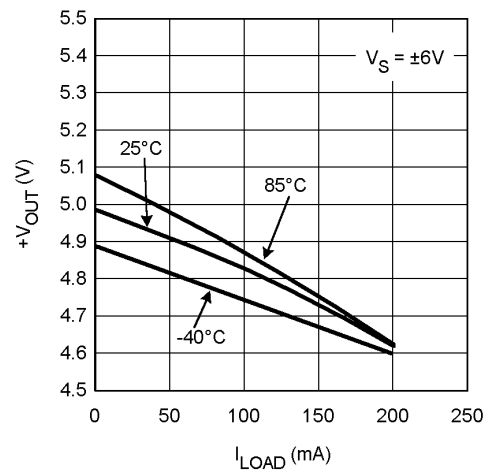
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Negative Output Swing into 25Ω



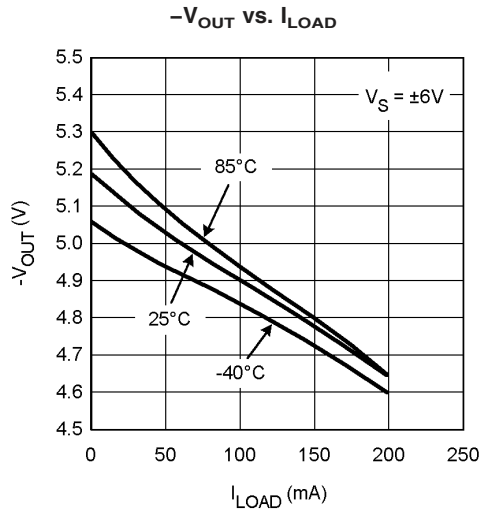
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$+V_{\text{OUT}}$ vs. I_{LOAD}

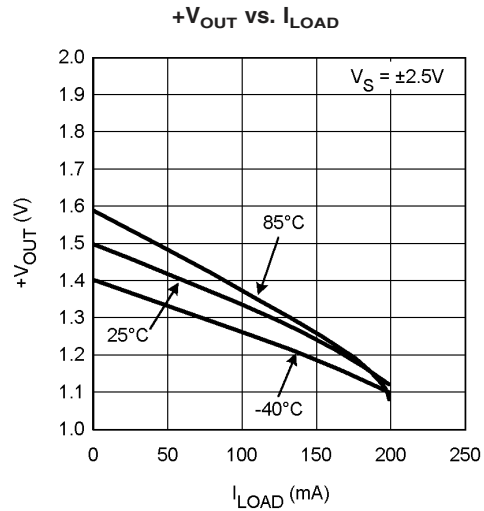


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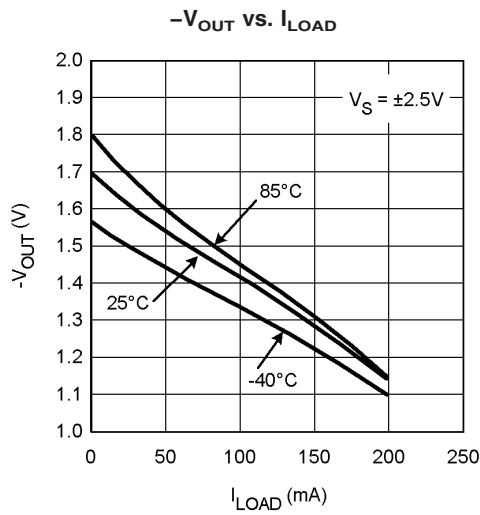
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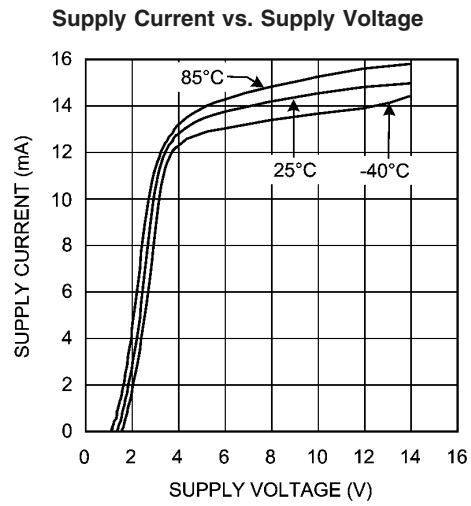
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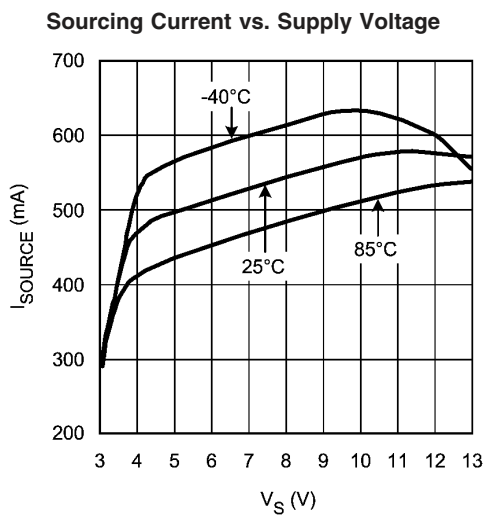
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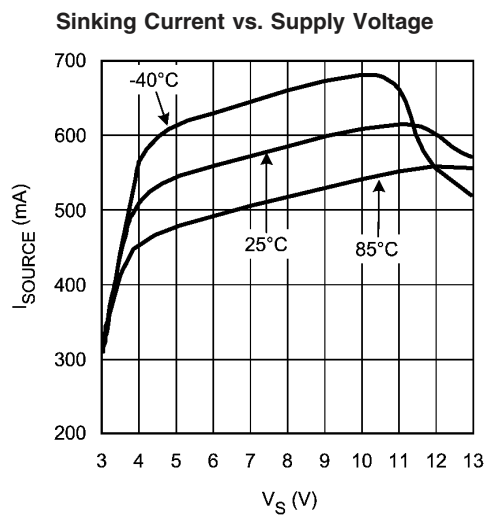
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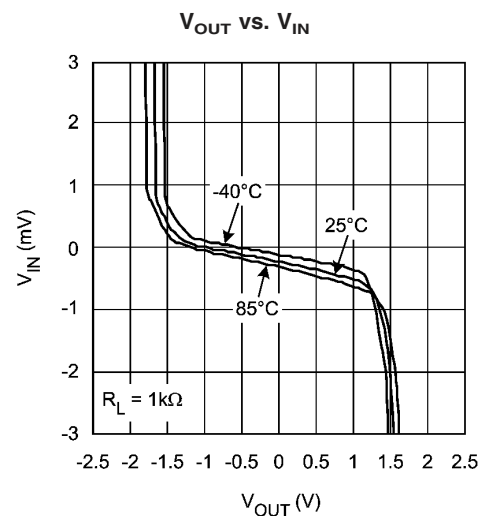
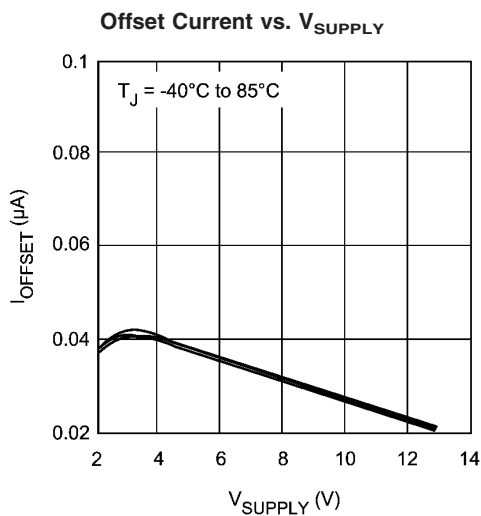
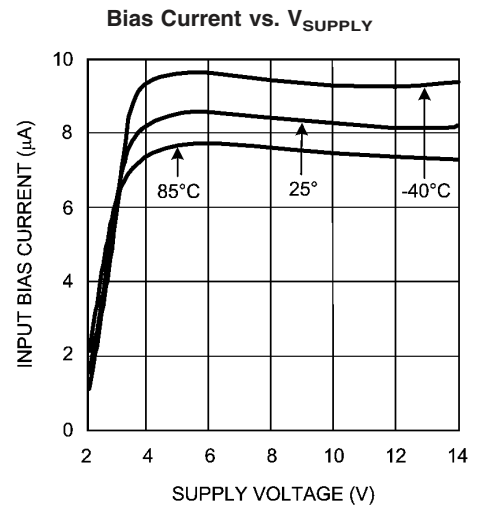
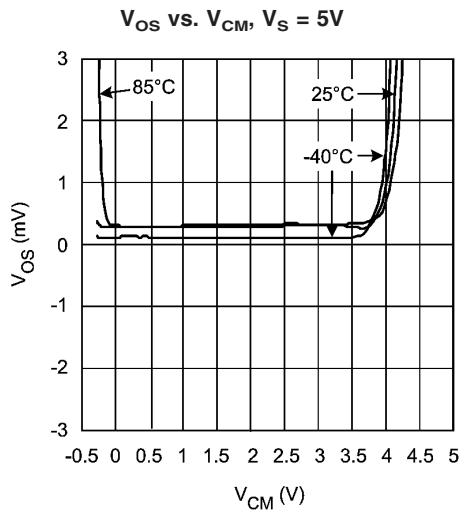
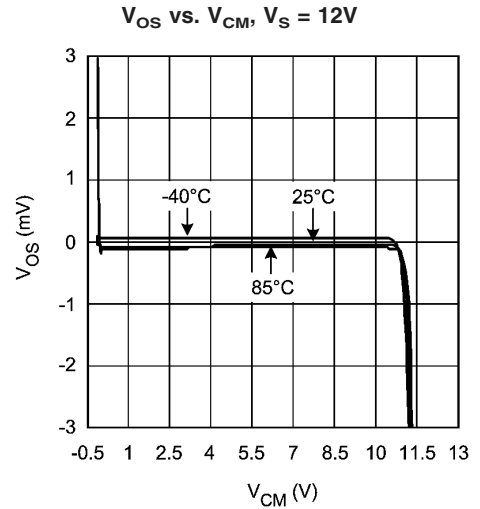
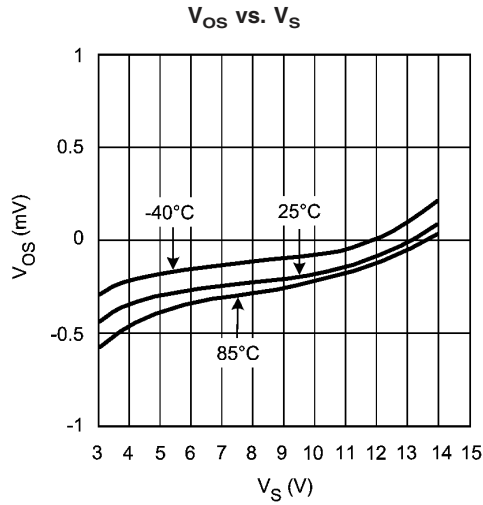


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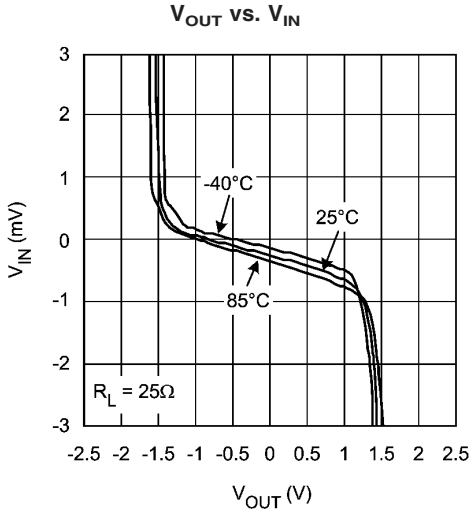


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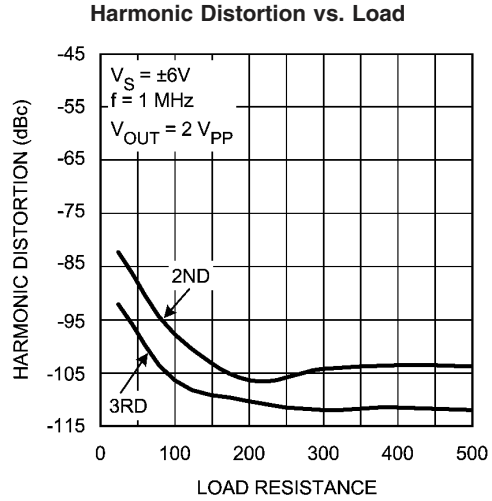
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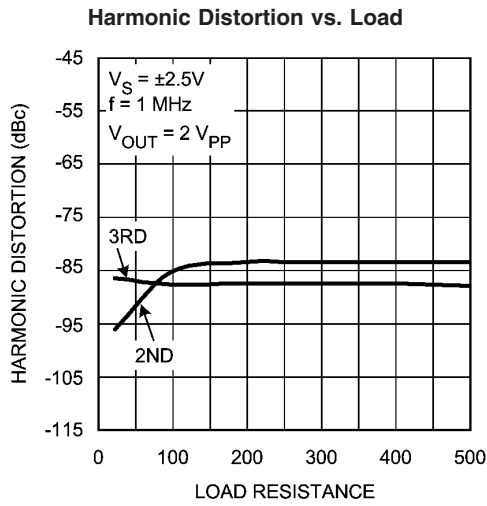
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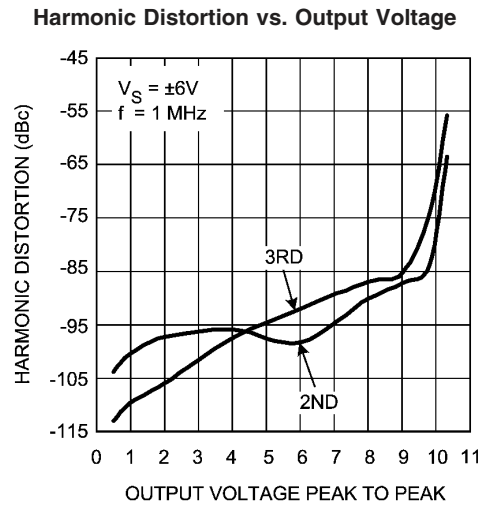
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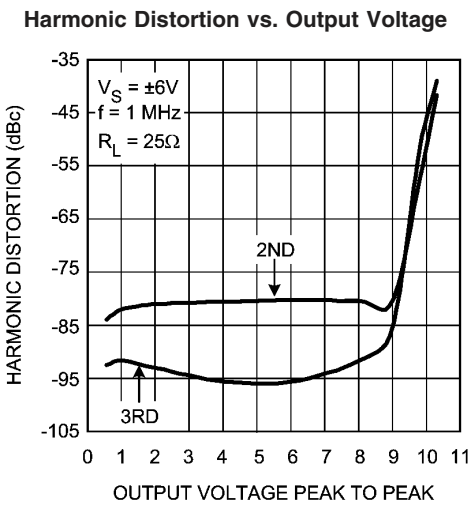
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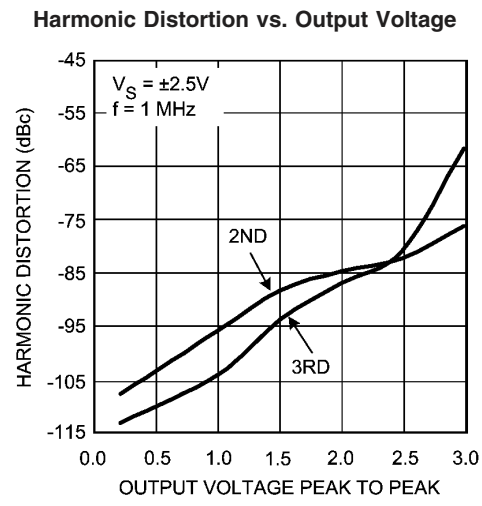
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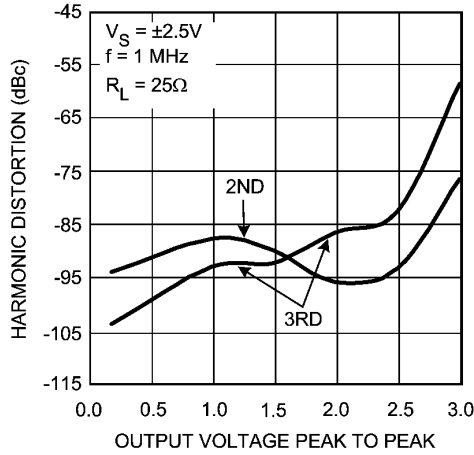
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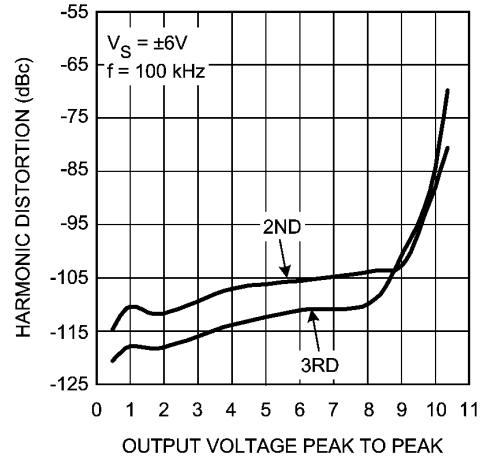
Typical Performance Characteristics (Continued)

Harmonic Distortion vs. Output Voltage



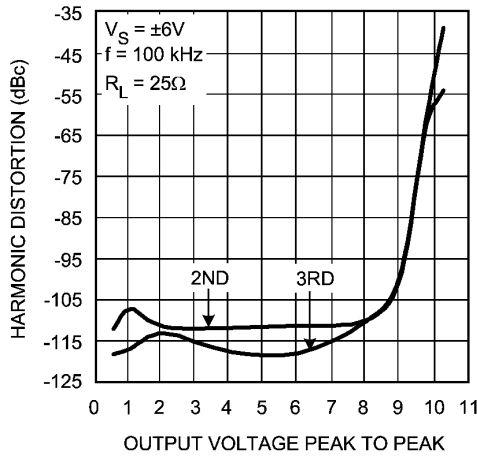
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Harmonic Distortion vs. Output Voltage



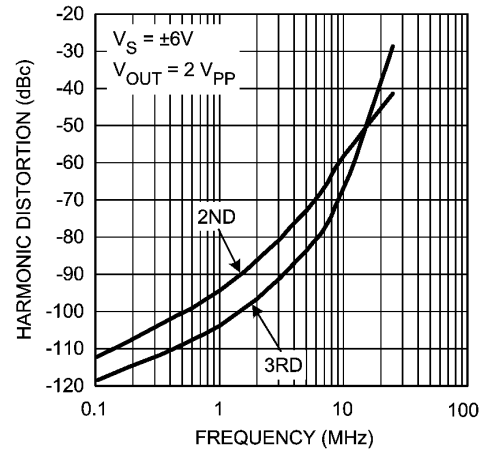
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Harmonic Distortion vs. Output Voltage



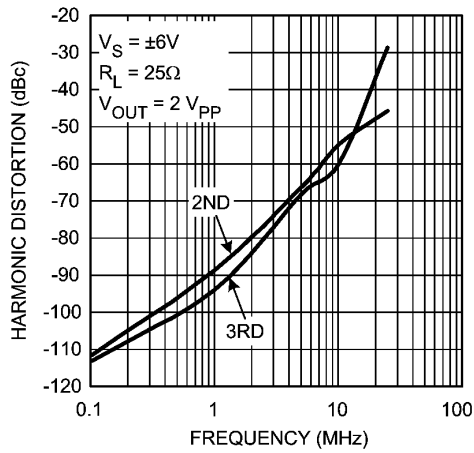
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Harmonic Distortion vs. Frequency



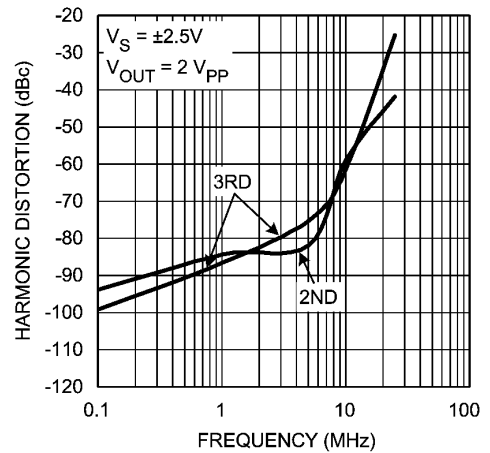
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Harmonic Distortion vs. Frequency



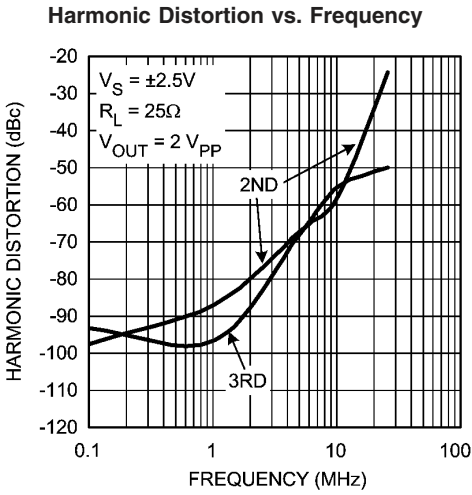
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Harmonic Distortion vs. Frequency

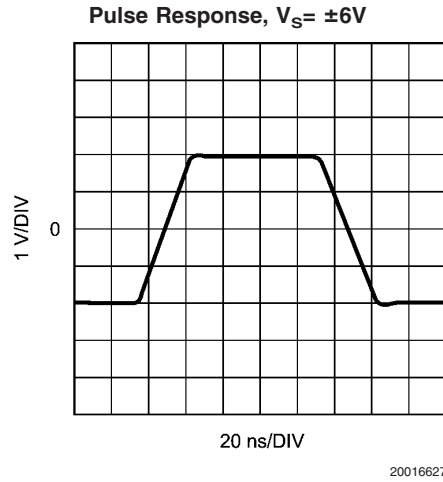


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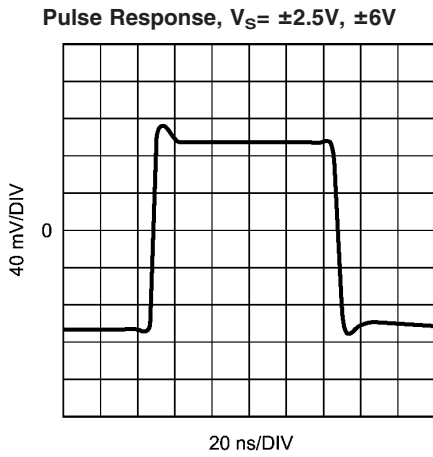
Typical Performance Characteristics (Continued)



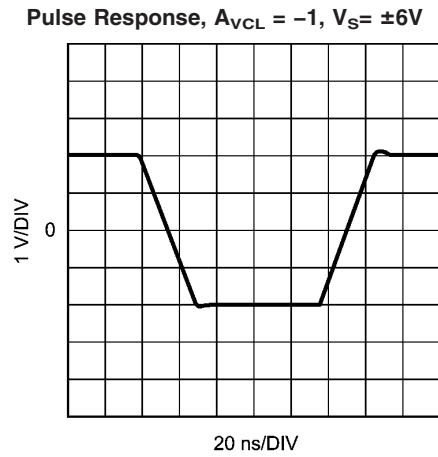
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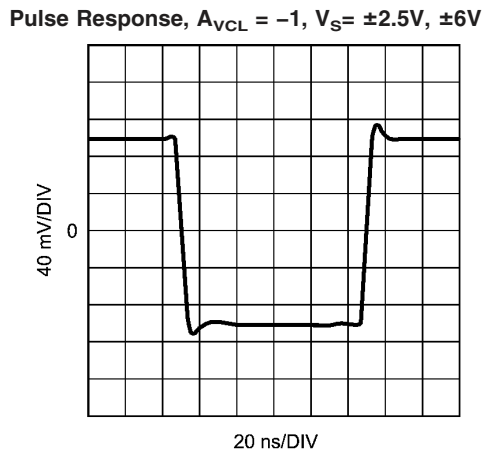
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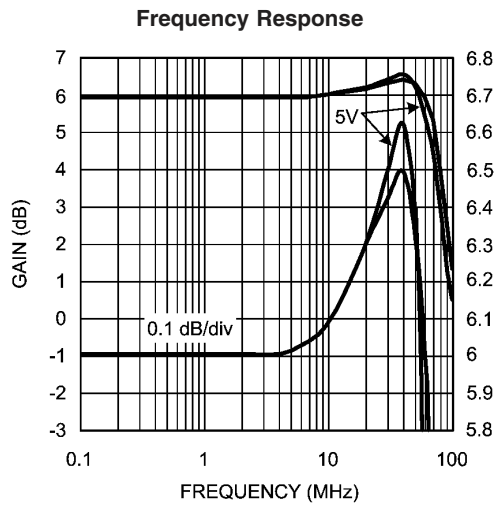
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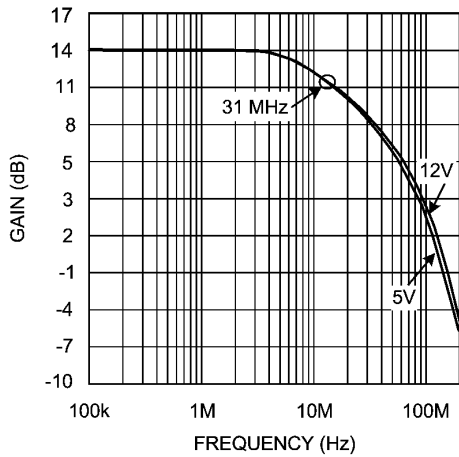
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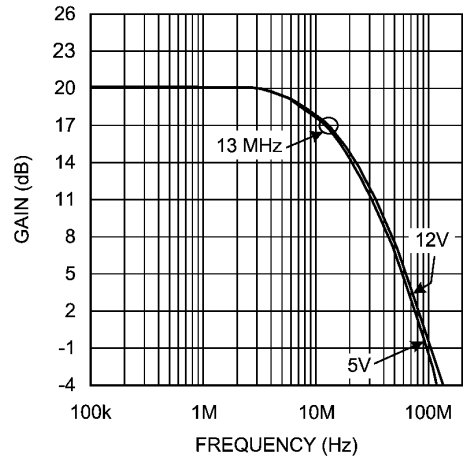
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Typical Performance Characteristics (Continued)

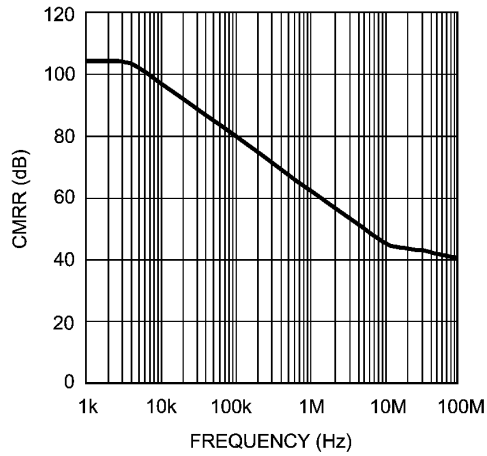
Frequency Response, $A_{VCL} = +5V$



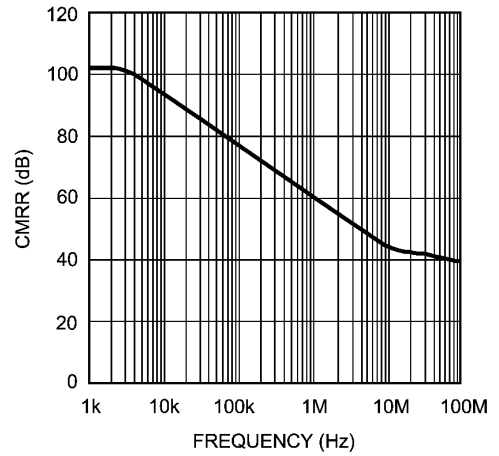
Frequency Response, $A_{VCL} = +10V$



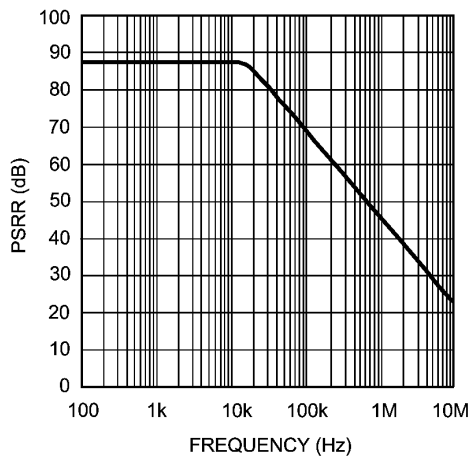
CMRR vs. Frequency @ 12V



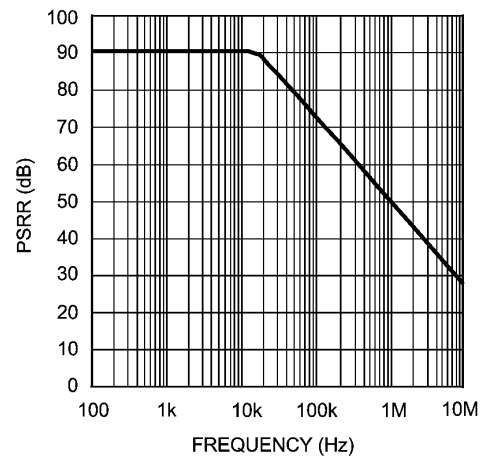
CMRR vs. Frequency @ 5V

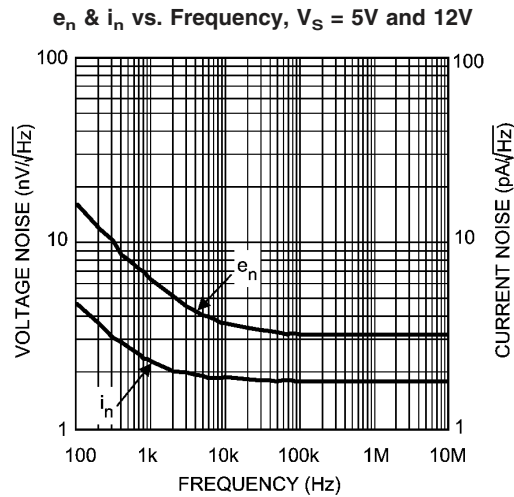


PSRR+ vs. Frequency, $V_S = 5V$ and 12V



PSRR- vs. Frequency $V_S = 5V$ and 12V





Application Notes

THERMAL MANAGEMENT

The LMH6672 is a high-speed, high power, dual operational amplifier with a very high slew rate and very low distortion. For ease of use, it uses conventional voltage feedback. These characteristics make the LMH6672 ideal for applications where driving low impedances of 25-100 Ω such as xDSL and active filters.

A class AB output stage allows the LMH6672 to deliver high currents to low impedance loads with low distortion while consuming low quiescent supply current. For most op-amps, class AB topology means that internal power dissipation is rarely an issue, even with the trend to smaller surface mount packages. However, the LMH6672 has been designed for applications where high levels of power dissipation may be encountered.

Several factors contribute to power dissipation and consequently higher junction temperatures. These factors need to be well understood if the LMH6672 is to perform to specifications in all applications. This section will examine the typical application that is shown on the front page of this data sheet as an example. (Figure 1) Because both amplifiers are in a single package, the calculations will be for the total power dissipated by both amplifiers.

There are two separate contributors to the internal power dissipation:

1. The product of the supply voltage and the quiescent current when no signal is being delivered to the external load.
2. The additional power dissipated while delivering power to the external load.

The first of these components appears easy to calculate simply by inspecting the data sheet. The typical quiescent supply current for this part is 7.2 mA per amplifier, therefore, with a ± 6 volt supply, the total power dissipation is:

$$P_D = V_S \times 2 \times I_Q = 12 \times (14.4 \times 10^{-3}) = 173 \text{ mW}$$

$$(V_S = V_{CC} + V_{EE})$$

With a thermal resistance of 172 $^{\circ}\text{C}/\text{W}$ for the SOIC package, this level of internal power dissipation will result in a junction temperature (T_J) of 30 $^{\circ}\text{C}$ above ambient.

Using the worst-case maximum supply current of 18 mA and an ambient of 85 $^{\circ}\text{C}$, a similar calculation results in a power dissipation of 216 mW, or a T_J of 122 $^{\circ}\text{C}$.

This is approaching the maximum allowed T_J of 150 $^{\circ}\text{C}$ before a signal is applied. Fortunately, in normal operation, this term is reduced, for reasons that will soon be explained.

The second contributor to high T_J is the power dissipated internally when power is delivered to the external load. This cause of temperature rise is more difficult to calculate, even when the actual operating conditions are known.

To maintain low distortion, in a Class AB output stage, an idle current, I_Q , is maintained through the output transistors when there is little or no output signal. In the LMH6672, about 4.8 mA of the total quiescent supply current of 14.4 mA flows through the output stages.

Under normal large signal conditions, as the output voltage swings positive, one transistor of the output pair will conduct the load current, while the other transistor shuts off, and dissipates no power. During the negative signal swing this situation is reversed, with the lower transistor sinking the load current while the upper transistor is cut off. The current in each transistor will approximate a half wave rectified version of the total load current.

Because the output stage idle current is now routed into the load, 4.8 mA can be subtracted from the quiescent supply current when calculating the quiescent power when the output is driving a load.

The power dissipation caused by driving a load in a DSL application, using a 1:2 turns ratio transformer driving 20 mW into the subscriber line and 20 mW into the back termination resistors, can be calculated as follows:

$$P_{\text{DRIVER}} = P_{\text{TOT}} - (P_{\text{TERM}} + P_{\text{LINE}}) \text{ where}$$

P_{DRIVER} is the LMH6672 power dissipation

P_{TOT} is the total power drawn from the power supply

P_{TERM} is the power dissipated in the back termination resistors

P_{LINE} is the power sent into the subscriber line

At full specified power, $P_{\text{TERM}} = P_{\text{LINE}} = 20 \text{ mW}$, $P_{\text{TOT}} = V_S \times I_S$.

In this application, $V_S = 12V$.

$$I_S = I_Q + A_{VG} |I_{OUT}|.$$

I_Q = the LMH6672 quiescent current minus the output stage idle current.

Application Notes (Continued)

$$I_Q = 14.4 - 4.8 = 9.6 \text{ mA}$$

$A_{VG} |I_{OUT}|$ for a full-rate ADSL CPE application, using a 1:2 turns ratio transformer, is $\sqrt{(40 \text{ mW}/50\Omega)} = 28.28 \text{ mA RMS}$.

For a Gaussian signal, which the DMT ADSL signal approximates, $A_{VG} |I_{OUT}| = \sqrt{2/\pi} \times I_{RMS} = 22.6 \text{ mA}$. Therefore, $P_{TOT} = (22.6 \text{ mA} + 9.6 \text{ mA}) \times 12\text{V} = 386 \text{ mW}$ and P_{DRIVER} is $40 = 346 \text{ mW}$.

In the SOIC package, with a θ_{JA} of 172°C/W , this causes a temperature rise of 60°C . With an ambient temperature at the maximum recommended 85°C , the T_J is at 145°C , well below the specified 150°C maximum.

Even if we assume the absolute maximum I_S over temperature of 18 mA , when we scale up the I_Q proportionally to 7 mA , the P_{DRIVER} only goes up by 41 mW causing a 62°C rise to 147°C .

Although very few CPE applications will ever operate in an environment as hot as 85°C , if a lower T_J is desired or the LMH6672 is to be used in an application where the power dissipation is higher, the PSOP package provides a much lower θ_{JA} of only 58.6°C/W .

Using the same P_{DRIVER} as above, we find that the temperature rise is only 19° and 21°C , resulting in T_J 's in an 85°C ambient of 104°C and 106°C respectively.

CIRCUIT LAYOUT CONSIDERATIONS

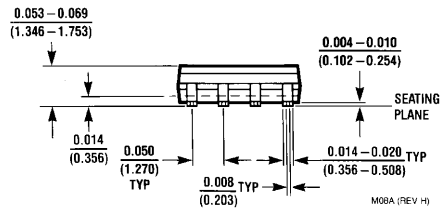
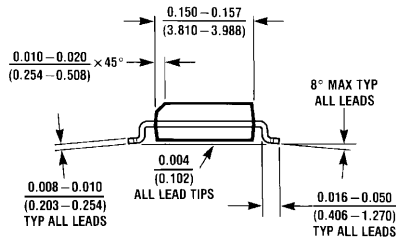
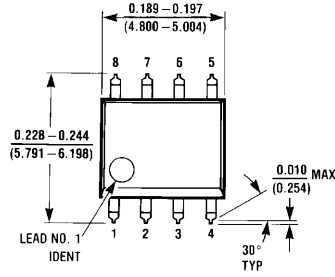
National Semiconductor suggests the following evaluation boards as a guide for high frequency layout and as an aid in device testing and characterization. Since the exposed PAD (or DAP) of the PSOP package is internally floating, the footprint for DAP could be connected to ground plane in PCB for better heat dissipation.

Device	Package	Evaluation Board PN
LMH6672MA	8-Pin SOIC	CLC730036
LMH6672MR	8-Pin PSOP	CLC730121

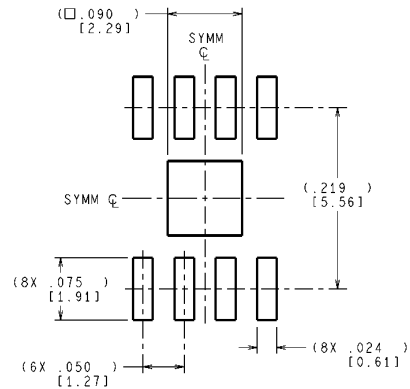
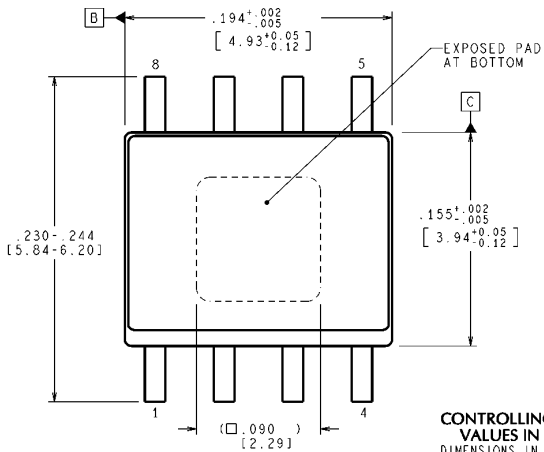
These free evaluation boards are shipped when a device sample request is placed with National Semiconductor.

Physical Dimensions inches (millimeters)

unless otherwise noted

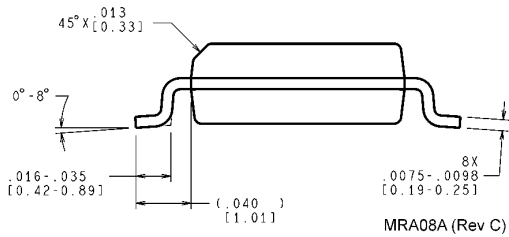
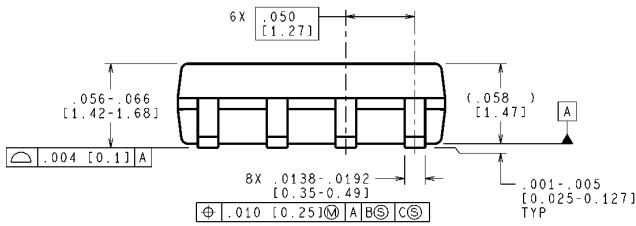


8-Pin SOIC
NS Package Number M08A



CONTROLLING DIMENSION IS INCH
VALUES IN [] ARE MILLIMETERS
DIMENSIONS IN () FOR REFERENCE ONLY

RECOMMENDED LAND PATTERN



8-Pin PSOP
NS Package Number MRA08A

MRA08A (Rev C)

Notes

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National Semiconductor
Americas Customer
Support Center
 Email: new.feedback@nsc.com
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www.national.com

National Semiconductor
Europe Customer Support Center
 Fax: +49 (0) 180-530 85 86
 Email: europe.support@nsc.com
 Deutsch Tel: +49 (0) 69 9508 6208
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National Semiconductor
Asia Pacific Customer
Support Center
 Email: ap.support@nsc.com

National Semiconductor
Japan Customer Support Center
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 Email: jpn.feedback@nsc.com
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