### **General Description**

The MAX4040-MAX4044 family of micropower op amps operates from a single +2.4V to +5.5V supply or dual ±1.2V to ±2.75V supplies and have rail-to-rail input and output capabilities. These amplifiers provide a 90kHz gain-bandwidth product while using only 10µA of supply current per amplifier. The MAX4041/MAX4043 have a low-power shutdown mode that reduces supply current to less than 1µA and forces the output into a high-impedance state. The combination of low-voltage operation, rail-to-rail inputs and outputs, and ultra-low power consumption makes these devices ideal for any portable/battery-powered system.

These amplifiers have outputs that typically swing to within 10mV of the rails with a 100k $\Omega$  load. Rail-to-rail input and output characteristics allow the full powersupply voltage to be used for signal range. The combination of low input offset voltage, low input bias current, and high open-loop gain makes them suitable for lowpower/low-voltage precision applications.

The MAX4040 is offered in a space-saving 5-pin SOT23 package. All specifications are guaranteed over the -40°C to +85°C extended temperature range.

	Applications
Battery-Powered	Strain Gauges
Systems	Sensor Amplifiers
Portable/Battery-Powered	Cellular Phones
Electronic Equipment	Notebook Computers
Digital Scales	PDAs

#### NO. OF PART SHUTDOWN **PIN-PACKAGE** AMPS 5-pin SOT23, MAX4040 1 8-pin µMAX/SO MAX4041 Yes 8-pin µMAX/SO 1 MAX4042 2 8-pin uMAX/SO 10-pin µMAX/ MAX4043 2 Yes 14-pin SO MAX4044 14-pin SO 4

µMAX is a registered trademark of Maxim Integrated Products, Inc.

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Maxim Integrated Products 1

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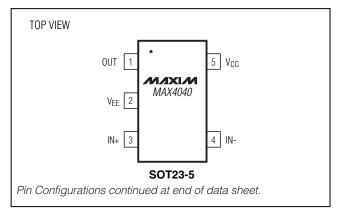
#### **Features**

- Single-Supply Operation Down to +2.4V
- Ultra-Low Power Consumption: 10µA Supply Current per Amplifier 1µA Shutdown Mode (MAX4041/MAX4043)
- Rail-to-Rail Input Common-Mode Range
- Outputs Swing Rail-to-Rail
- No Phase Reversal for Overdriven Inputs
- ♦ 200µV Input Offset Voltage
- Unity-Gain Stable for Capacitive Loads up to 200pF
- 90kHz Gain-Bandwidth Product
- Available in Space-Saving 5-Pin SOT23 and 8-Pin µMAX<sup>®</sup> Packages

### **Ordering Information**

PART	TEMP RANGE	PIN- PACKAGE	SOT TOP MARK
MAX4040EUK-T	-40°C to +85°C	5 SOT23-5	ACGF
MAX4040EUA	-40°C to +85°C	8 µMAX	_
MAX4040ESA	-40°C to +85°C	8 SO	
MAX4041ESA	-40°C to +85°C	8 SO	_
MAX4041EUA	-40°C to +85°C	8 µMAX	
MAX4042EUA	-40°C to +85°C	8 µMAX	
MAX4042ESA	-40°C to +85°C	8 SO	_
MAX4043EUB	-40°C to +85°C	10 µMAX	_
MAX4043ESD	-40°C to +85°C	14 SO	_
MAX4044ESD	-40°C to +85°C	14 SO	_

### **Pin Configurations**



For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

Selector Guide

### **ABSOLUTE MAXIMUM RATINGS**

Supply Voltage (V <sub>CC</sub> to V <sub>EE</sub> )+6V All Other Pins(V <sub>CC</sub> + 0.3V) to (V <sub>EE</sub> - 0.3V) Output Short-Circuit Duration to V <sub>CC</sub> or V <sub>EE</sub> Continuous Continuous Power Dissipation ( $T_A$ = +70°C)
5-Pin SOT23 (derate 7.1mW/°C above +70°C)571mW 8-Pin μMAX (derate 4.1mW/°C above +70°C)330mW 8-Pin SO (derate 5.88mW/°C above +70°C)471mW

10-Pin µMAX (derate 5.6mW/°C above +	70°C)444mW
14-Pin SO (derate 8.33mW/°C above +70	)°C)667mW
Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### ELECTRICAL CHARACTERISTICS—TA = +25°C

 $(V_{CC} = +5.0V, V_{EE} = 0V, V_{CM} = 0V, V_{OUT} = V_{CC} / 2, \overline{SHDN} = V_{CC}, R_L = 100k\Omega$  tied to  $V_{CC} / 2$ , unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS				ТҮР	MAX	UNITS
Supply-Voltage Range	Vcc	Inferred from PSRR test			2.4		5.5	V
Supply Current	100	$V_{CC} = 2.4 V$				10		
per Amplifier	Icc	$V_{CC} = 5.0V$				14	20	μΑ
Shutdown Supply		$\overline{\text{SHDN}} = V_{\text{EE}}, \text{MAX4041}$	$V_{CC} = 2.4$	/		1.0		
Current per Amplifier	ICC(SHDN)	and MAX4043 only	$V_{CC} = 5.0$	/		2.0	5.0	μΑ
			MAX4044E	SD		±0.20	±2.0	mV
Input Offset Voltage	Vos	$V_{EE} \le V_{CM} \le V_{CC}$	MAX404_E	U_		±0.25	±2.5	1110
			All other pa	ackages		±0.20	±1.50	mV
Input Bias Current	IB	(Note 1)				±2	±10	nA
Input Offset Current	los	(Note 1)				±0.5	±3.0	nA
Differential Input	RIN(DIFF)	V <sub>IN+</sub> - V <sub>IN−</sub>   < 1.0V				45		MΩ
Resistance		V <sub>IN+</sub> - V <sub>IN-</sub>   > 2.5V				4.4		kΩ
Input Common-Mode Voltage Range	Vсм	Inferred from the CMRR test					Vcc	V
Common-Mode	CMRR		MAX404_E	U_	65	94		dB
Rejection Ratio	Civinn	$V_{EE} \le V_{CM} \le V_{CC}$	All other pa	ackages	70	94		uв
Power-Supply Rejection Ratio	PSRR	$2.4V \le V_{CC} \le 5.5V$			75	85		dB
Large-Signal	A		- 0.0\/\	$R_L = 100 k\Omega$		94		dB
Voltage Gain	Avol	$(V_{EE} + 0.2V) \le V_{OUT} \le (V_{C})$	( - U.ZV)	$R_L = 25k\Omega$	74	85		uв
Output Voltage	Vон	Specified as V <sub>CC</sub> - V <sub>OH</sub>		$R_L = 100 k\Omega$		10		mV
Swing High	VOH			$R_L = 25k\Omega$		60	90	IIIV
Output Voltage	VOL	$ \begin{array}{ c c c c c } \mbox{Specified as }  V_{EE} - V_{OL}  & \hline & R_L = 100 k \Omega \\ \hline & R_L = 25 k \Omega \\ \end{array} $		$R_L = 100 k\Omega$		10		mV
Swing Low	VOL			$R_L = 25k\Omega$	$R_L = 25k\Omega$		40	60
Output Short-Circuit		Sourcing Sinking			0.7		mA	
Current	IOUT(SC)					2.5		111/5
Channel-to-Channel Isolation		Specified at DC, MAX4042	2/MAX4043/M	IAX4044 only		80		dB

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### ELECTRICAL CHARACTERISTICS—T<sub>A</sub> = +25°C (continued)

 $(V_{CC} = +5.0V, V_{EE} = 0V, V_{CM} = 0V, V_{OUT} = V_{CC} / 2, \overline{SHDN} = V_{CC}, R_L = 100k\Omega$  tied to  $V_{CC} / 2$ , unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
Output Leakage Current in Shutdown	IOUT(SHDN)	SHDN=VEE=0, MAX4041/MAX4043 only(Note 2)		20	100	nA
SHDN Logic Low	VIL	MAX4041/MAX4043 only			0.3 x Vcc	V
SHDN Logic High	VIH	MAX4041/MAX4043 only	0.7 x V <sub>CC</sub>			V
SHDN Input Bias Current	Iih, Iil	MAX4041/MAX4043 only		40	120	nA
Gain Bandwidth Product	GBW			90		kHz
Phase Margin	$\Phi_{M}$			68		degrees
Gain Margin	Gm			18		dB
Slew Rate	SR			40		V/ms
Input Voltage Noise Density	en	f = 1kHz		70		nV/√Hz
Input Current Noise Density	in	f = 1kHz		0.05		pA/√Hz
Capacitive-Load Stability		$A_{VCL} = +1V/V$ , no sustained oscillations		200		pF
Power-Up Time	ton			200		μs
Shutdown Time	<b>t</b> SHDN	MAX4041 and MAX4043 only		50		μs
Enable Time from Shutdown	ten	MAX4041 and MAX4043 only		150		μs
Input Capacitance	CIN			3		pF
Total Harmonic Distortion	THD	$f_{IN} = 1 \text{kHz}, V_{OUT} = 2 \text{Vp-p}, A_V = +1 \text{V/V}$		0.05		%
Settling Time to 0.01%	ts	$A_V = +1V/V, V_{OUT} = 2V_{STEP}$		50		μs

### ELECTRICAL CHARACTERISTICS—TA = TMIN to TMAX

 $(V_{CC} = +5.0V, V_{EE} = 0V, V_{CM} = 0V, V_{OUT} = V_{CC} / 2, \overline{SHDN} = V_{CC}, R_L = 100k\Omega$  tied to  $V_{CC} / 2$ , unless otherwise noted.) (Note 3)

PARAMETER	SYMBOL	C	ONDITIONS	MIN	ТҮР	МАХ	UNITS
Supply-Voltage Range	Vcc	Inferred from PSRR tes	st	2.4		5.5	V
Supply Current per Amplifier	Icc					28	μA
Shutdown Supply Current per Amplifier	ICC(SHDN)	$\overline{\text{SHDN}} = V_{\text{EE}}$ , MAX4041 and MAX4043 only				6.0	μA
			MAX4044ESA			±4.5	
Input Offset Voltage	Vos	$V_{EE} \le V_{CM} \le V_{CC}$	MAX404_EU_			±5.0	mV
			All other packages			±3.5	
Input Offset Voltage Drift	TCvos				2		µV/°C
Input Bias Current	Ι <sub>Β</sub>	(Note 1)			±20	nA	
Input Offset Current	los	(Note 1)			±8	nA	

### ELECTRICAL CHARACTERISTICS—TA = TMIN to TMAX (continued)

 $(V_{CC} = +5.0V, V_{EE} = 0V, V_{CM} = 0V, V_{OUT} = V_{CC} / 2, \overline{SHDN} = V_{CC}, R_L = 100k\Omega$  tied to  $V_{CC} / 2$ , unless otherwise noted.) (Note 3)

PARAMETER	SYMBOL	COND	ITIONS	MIN	TYP	MAX	UNITS
Input Common-Mode Voltage Range	V <sub>CM</sub>	Inferred from the CMRR test		VEE		V <sub>CC</sub>	V
Common-Mode	CMRR	VFF ≤ VCM ≤ VCC	MAX404_EU_	60			dD
Rejection Ratio	Civinn	$V \in E \leq V \subset M \leq V \subset C$	All other packages	65			dB
Power-Supply Rejection Ratio	PSRR	$2.4V \le V_{CC} \le 5.5V$					dB
Large-Signal Voltage Gain	Avol	$(V_{EE} + 0.2V) \le V_{OUT} \le (V_{CC} - 0.2V), R_{L} = 25k\Omega$					dB
Output Voltage Swing High	V <sub>OH</sub>	Specified as V <sub>CC</sub> - V <sub>OH</sub>  ,			125	mV	
Output Voltage Swing Low	V <sub>OL</sub>	Specified as VEE - VOL , F			75	mV	

Note 1: Input bias current and input offset current are tested with V<sub>CC</sub> = +5.0V and +0.5V  $\leq$  V<sub>CM</sub>  $\leq$  +4.5V.

 $(V_{CC} = +5.0V, V_{EE} = 0, V_{CM} = V_{CC}/2, \overline{SHDN} = V_{CC}, R_L = 100k\Omega$  to  $V_{CC}/2, T_A = +25^{\circ}C$ , unless otherwise noted.)

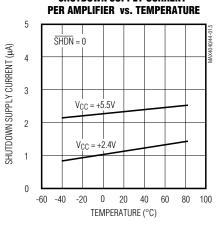
**Note 2:** Tested for  $V_{EE} \le V_{OUT} \le V_{CC}$ . Does not include current through external feedback network.

**Note 3:** All devices are 100% tested at  $T_A = +25^{\circ}C$ . All temperature limits are guaranteed by design.

#### **SUPPLY CURRENT PER AMPLIFIER** vs. TEMPERATURE 20 18 16 V<sub>CC</sub> = +5.5V SUPPLY CURRENT (µA) 14 12 10 $V_{CC} = +2.4V$ 8 6 4 2 0 -40 -20 40 60 -60 0 20 80 100 TEMPERATURE (°C)



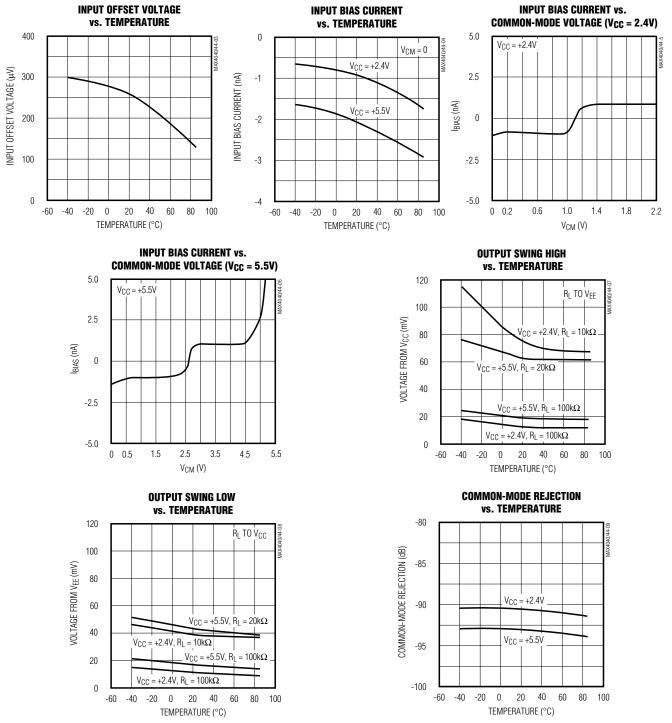
**Typical Operating Characteristics** 



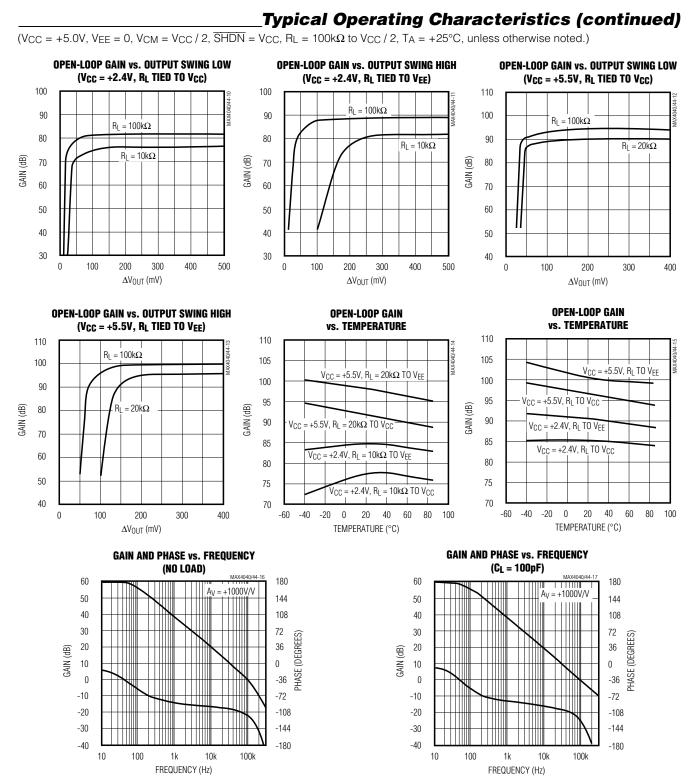
M/IXI/M

### **Typical Operating Characteristics (continued)**

(V<sub>CC</sub> = +5.0V, V<sub>EE</sub> = 0, V<sub>CM</sub> = V<sub>CC</sub> / 2, SHDN = V<sub>CC</sub>, R<sub>L</sub> = 100kΩ to V<sub>CC</sub> / 2, T<sub>A</sub> = +25°C, unless otherwise noted.)







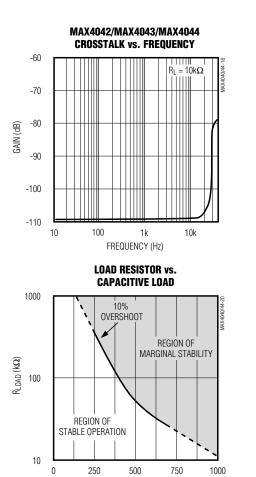
M/IXI/M

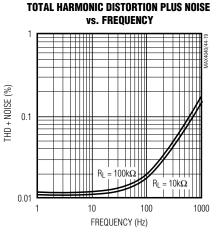
MAX4040-MAX4044

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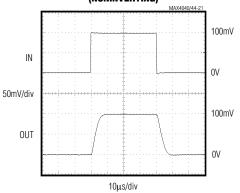
### **Typical Operating Characteristics (continued)**

 $(V_{CC} = +5.0V, V_{EE} = 0, V_{CM} = V_{CC}/2, \overline{SHDN} = V_{CC}, R_L = 100k\Omega$  to  $V_{CC}/2, T_A = +25^{\circ}C$ , unless otherwise noted.)





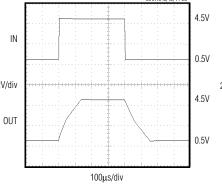
#### SMALL-SIGNAL TRANSIENT RESPONSE (NONINVERTING)



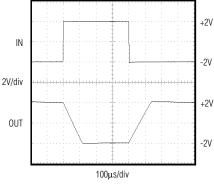
**SMALL-SIGNAL TRANSIENT RESPONSE** LARGE-SIGNAL TRANSIENT RESPONSE (NONINVERTING) (INVERTING) MAX4040/44-22 100mV IN 0V 2V/div 100mV OUT

0V

CLOAD (pF)







//IXI///

10µs/div

IN

50mV/div

OUT

### **Pin Description**

		PIN						
MAX	(4040	MAX4041	MAX4042	MAX	4043	MAX4044	NAME	FUNCTION
SOT23-5	SO/µMAX	INIAA4041	WAA4042	μΜΑΧ	MAX SO			
1	6	6		_	_	_	OUT	Amplifier Output. High impedance when in shutdown mode.
2	4	4	4	4	4	11	V <sub>EE</sub>	Negative Supply. Tie to ground for single-supply operation.
3	3	3	_	—	—	—	IN+	Noninverting Input
4	2	2	—	_		—	IN-	Inverting Input
5	7	7	8	10	14	4	Vcc	Positive Supply
_	1, 5, 8	1, 5		_	5, 7, 8, 10	_	N.C.	No Connection. Not internally con- nected.
_		8					SHDN	Shutdown Input. Drive high, or tie to V <sub>CC</sub> for normal operation. Drive to V <sub>EE</sub> to place device in shutdown mode.
_		_	1, 7	1, 9	1, 13	1, 7	OUTA, OUTB	Outputs for Amplifiers A and B. High impedance when in shutdown mode.
_			2, 6	2, 8	2, 12	2, 6	INA-, INB-	Inverting Inputs to Amplifiers A and B
_		_	3, 5	3, 7	3, 11	3, 5	INA+, INB+	Noninverting Inputs to Amplifiers A and B
_				5, 6	6, 9		SHDNA, SHDNB	Shutdown Inputs for Amplifiers A and B. Drive high, or tie to V <sub>CC</sub> for normal operation. Drive to V <sub>EE</sub> to place device in shutdown mode.
_	_	_	_	_		8, 14	OUTC, OUTD	Outputs for Amplifiers C and D
	_	_	_	_	_	9, 13	INC-, IND-	Inverting Inputs to Amplifiers C and D
_	_					10, 12	INC+, IND+	Noninverting Inputs to Amplifiers C and D

### **Detailed Description**

#### **Rail-to-Rail Input Stage**

The MAX4040–MAX4044 have rail-to-rail inputs and rail-to-rail output stages that are specifically designed for low-voltage, single-supply operation. The input stage consists of separate NPN and PNP differential stages, which operate together to provide a common-mode range extending to both supply rails. The crossover region of these two pairs occurs halfway between VCC and VEE. The input offset voltage is typically 200µV. Low operating supply voltage, low supply current, rail-to-rail common-mode input range, and rail-to-rail outputs make this family of operational amplifiers

an excellent choice for precision or general-purpose, low-voltage battery-powered systems.

Since the input stage consists of NPN and PNP pairs, the input bias current changes polarity as the commonmode voltage passes through the crossover region. Match the effective impedance seen by each input to reduce the offset error caused by input bias currents flowing through external source impedances (Figures 1a and 1b). The combination of high source impedance plus input capacitance (amplifier input capacitance plus stray capacitance) creates a parasitic pole that produces an underdamped signal response. Reducing input capacitance or placing a small capacitor across the feedback resistor improves response in this case.



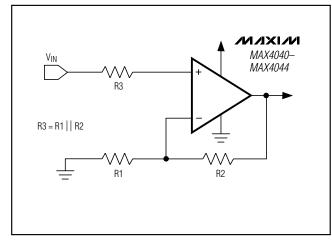


Figure 1a. Minimizing Offset Error Due to Input Bias Current (Noninverting)

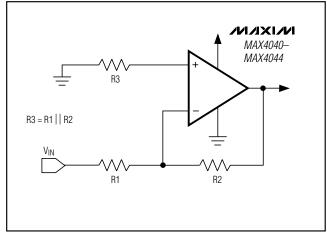


Figure 1b. Minimizing Offset Error Due to Input Bias Current (Inverting)

The MAX4040–MAX4044 family's inputs are protected from large differential input voltages by internal  $2.2k\Omega$ series resistors and back-to-back triple-diode stacks across the inputs (Figure 2). For differential input voltages (much less than 1.8V), input resistance is typically 45M $\Omega$ . For differential input voltages greater than 1.8V, input resistance is around  $4.4k\Omega$ , and the input bias current can be approximated by the following equation:

$$\mathsf{I}_{\mathsf{BIAS}} = (\mathsf{V}_{\mathsf{DIFF}} - 1.8\mathsf{V}) / 4.4\mathsf{k}\Omega$$

In the region where the differential input voltage approaches 1.8V, the input resistance decreases exponentially from  $45M\Omega$  to  $4.4k\Omega$  as the diode block begins conducting. Conversely, the bias current increases with the same curve.

#### **Rail-to-Rail Output Stage**

The MAX4040–MAX4044 output stage can drive up to a 25k $\Omega$  load and still swing to within 60mV of the rails. Figure 3 shows the output voltage swing of a MAX4040 configured as a unity-gain buffer, powered from a single +4.0V supply voltage. The output for this setup typically swings from (VEE + 10mV) to (VCC - 10mV) with a 100k $\Omega$  load.

### **Applications Information**

#### **Power-Supply Considerations**

The MAX4040–MAX4044 operate from a single +2.4V to +5.5V supply (or dual  $\pm$ 1.2V to  $\pm$ 2.75V supplies) and consume only 10µA of supply current per amplifier. A high power-supply rejection ratio of 85dB allows the amplifiers to be powered directly off a decaying battery voltage, simplifying design and extending battery life.

#### **Power-Up Settling Time**

The MAX4040–MAX4044 typically require 200 $\mu$ s to power up after V<sub>CC</sub> is stable. During this start-up time, the output is indeterminant. The application circuit should allow for this initial delay.

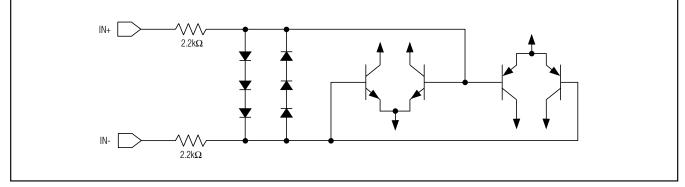


Figure 2. Input Protection Circuit



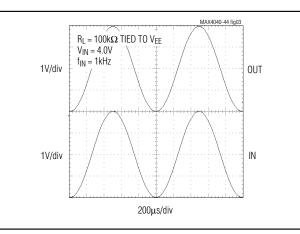


Figure 3. Rail-to-Rail Input/Output Voltage Range

#### Shutdown Mode

The MAX4041 (single) and MAX4043 (dual) feature a low-power shutdown mode. When the shutdown pin (SHDN) is pulled low, the supply current drops to 1 $\mu$ A per amplifier, the amplifier is disabled, and the outputs enter a high-impedance state. Pulling SHDN high or leaving it floating enables the amplifier. Take care to ensure that parasitic leakage current at the SHDN pin does not inadvertently place the part into shutdown mode when SHDN is left floating. Figure 4 shows the output voltage response to a shutdown pulse. The logic threshold for SHDN is always referred to V<sub>CC</sub> / 2 (not to GND). When using dual supplies, pull SHDN to V<sub>EE</sub> to enter shutdown mode.

#### **Load-Driving Capability**

The MAX4040–MAX4044 are fully guaranteed over temperature and supply voltage to drive a maximum resistive load of  $25k\Omega$  to V<sub>CC</sub> / 2, although heavier loads can be driven in many applications. The rail-to-rail output stage of the amplifier can be modeled as a current source when driving the load toward V<sub>CC</sub>, and as a current sink when driving the load toward V<sub>EE</sub>. The magnitude of this current source/sink varies with supply voltage, ambient temperature, and lot-to-lot variations of the units.

Figures 5a and 5b show the typical current source and sink capability of the MAX4040–MAX4044 family as a function of supply voltage and ambient temperature. The contours on the graph depict the output current value, based on driving the output voltage to within 50mV, 100mV, and 200mV of either power-supply rail.

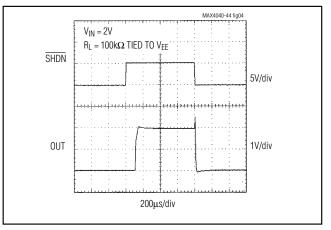


Figure 4. Shutdown Enable/Disable Output Voltage

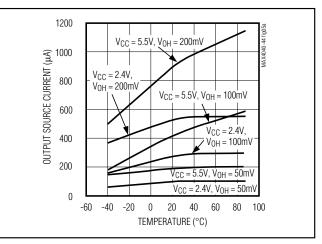


Figure 5a. Output Source Current vs. Temperature

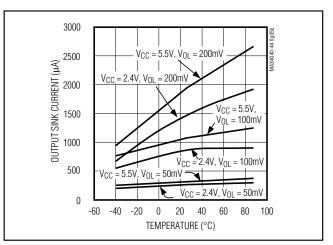


Figure 5b. Output Sink Current vs. Temperature

For example, a MAX4040 running from a single +2.4V supply, operating at  $T_A$  = +25°C, can source 240µA to within 100mV of V<sub>CC</sub> and is capable of driving a 9.6k $\Omega$  load resistor to V<sub>EE</sub>:

$$R_L = \frac{2.4V - 0.1V}{240\mu A} = 9.6k\Omega$$
 to  $V_{EE}$ 

The same application can drive a 4.6k  $\Omega$  load resistor when terminated in V\_CC / 2 (+1.2V in this case).

#### **Driving Capacitive Loads**

The MAX4040–MAX4044 are unity-gain stable for loads up to 200pF (see Load Resistor vs. Capacitive Load graph in *Typical Operating Characteristics*). Applications that require greater capacitive drive capability should use an isolation resistor between the output and the capacitive load (Figures 6a–6c). Note that this alternative results in a loss of gain accuracy because RISO forms a voltage divider with the load resistor.

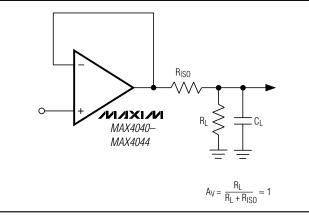
#### **Power-Supply Bypassing and Layout**

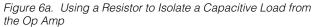
The MAX4040–MAX4044 family operates from either a single +2.4V to +5.5V supply or dual  $\pm$ 1.2V to  $\pm$ 2.75V supplies. For single-supply operation, bypass the power supply with a 100nF capacitor to V<sub>EE</sub> (in this case GND). For dual-supply operation, both the V<sub>CC</sub> and V<sub>EE</sub> supplies should be bypassed to ground with separate 100nF capacitors.

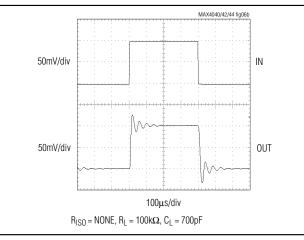
Good PC board layout techniques optimize performance by decreasing the amount of stray capacitance at the op amp's inputs and output. To decrease stray capacitance, minimize trace lengths by placing external components as close as possible to the op amp. Surface-mount components are an excellent choice.

#### Using the MAX4040–MAX4044 as Comparators

Although optimized for use as operational amplifiers, the MAX4040–MAX4044 can also be used as rail-to-rail I/O comparators. Typical propagation delay depends on the input overdrive voltage, as shown in Figure 7. External hysteresis can be used to minimize the risk of output oscillation. The positive feedback circuit, shown in Figure 8, causes the input threshold to change when the output voltage changes state. The two thresholds create a hysteresis band that can be calculated by the following equations:









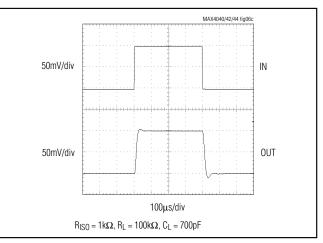


Figure 6c. Pulse Response with Isolating Resistor

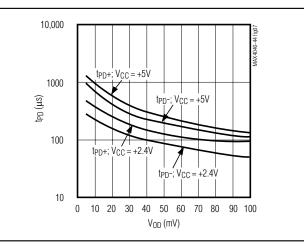


Figure 7. Propagation Delay vs. Input Overdrive

The MAX4040–MAX4044 contain special circuitry to boost internal drive currents to the amplifier output stage. This maximizes the output voltage range over which the amplifiers are linear. In an open-loop comparator application, the excursion of the output voltage is so close to the supply rails that the output stage transistors will saturate, causing the quiescent current to increase from the normal 10µA. Typical quiescent currents increase to  $35\mu$ A for the output saturating at V<sub>CC</sub> and 28µA for the output at V<sub>EE</sub>.

#### Using the MAX4040-MAX4044 as Ultra-Low-Power Current Monitors

The MAX4040–MAX4044 are ideal for applications powered from a battery stack. Figure 9 shows an application circuit in which the MAX4040 is used for monitoring the current of a battery stack. In this circuit, a current load is applied, and the voltage drop at the battery terminal is sensed.

The voltage on the load side of the battery stack is equal to the voltage at the emitter of Q1, due to the feedback loop containing the op amp. As the load current increases, the voltage drop across R1 and R2 increases. Thus, R2 provides a fraction of the load current (set by the ratio of R1 and R2) that flows into the emitter of the PNP transistor. Neglecting PNP base current, this current flows into R3, producing a ground-referenced voltage proportional to the load current. Scale R1 to give a voltage drop large enough in comparison to Vos of the op amp, in order to minimize errors.

The output voltage of the application can be calculated using the following equation:

 $V_{OUT} = [I_{LOAD} \times (R1 / R2)] \times R3$ 

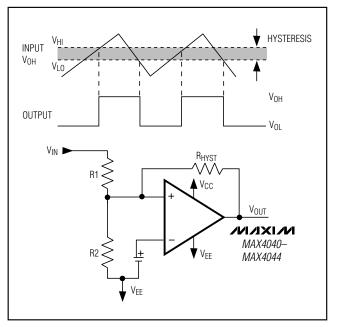


Figure 8. Hysteresis Comparator Circuit

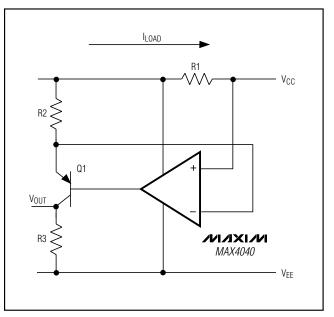
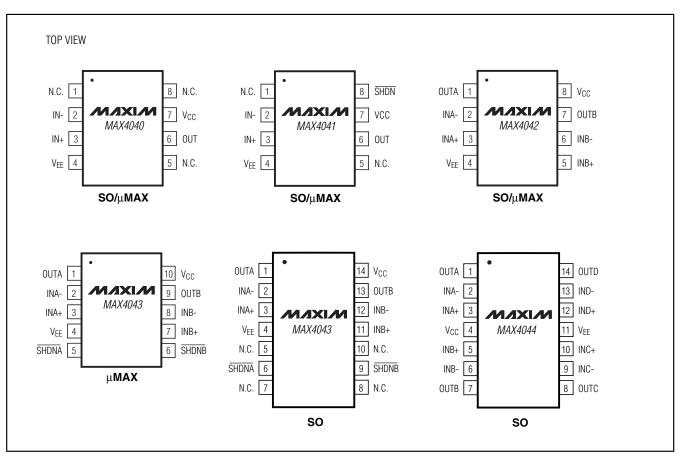


Figure 9. Current Monitor for a Battery Stack

For a 1V output and a current load of 50mA, the choice of resistors can be R1 =  $2\Omega$ , R2 =  $100k\Omega$ , R3 =  $1M\Omega$ . The circuit consumes less power (but is more susceptible to noise) with higher values of R1, R2, and R3.





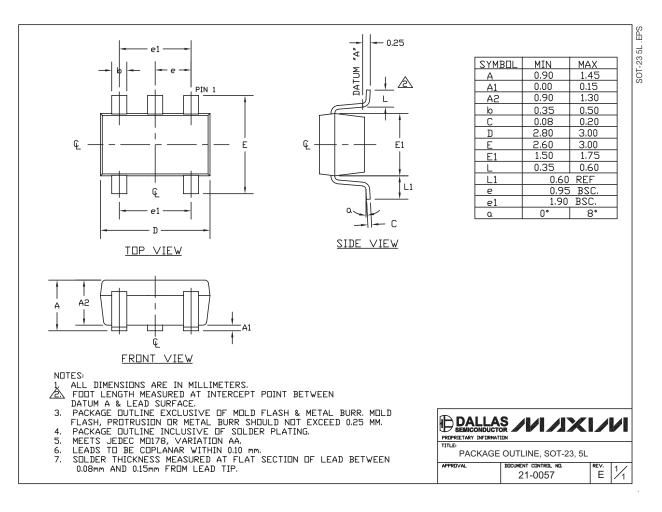
### Pin Configurations (continued)

### Chip Information

MAX4040/MAX4041 TRANSISTOR COUNT: 234 MAX4042/MAX4043 TRANSISTOR COUNT: 466 MAX4044 TRANSISTOR COUNT: 932 SUBSTRATE CONNECTED TO VEE MAX4040-MAX4044

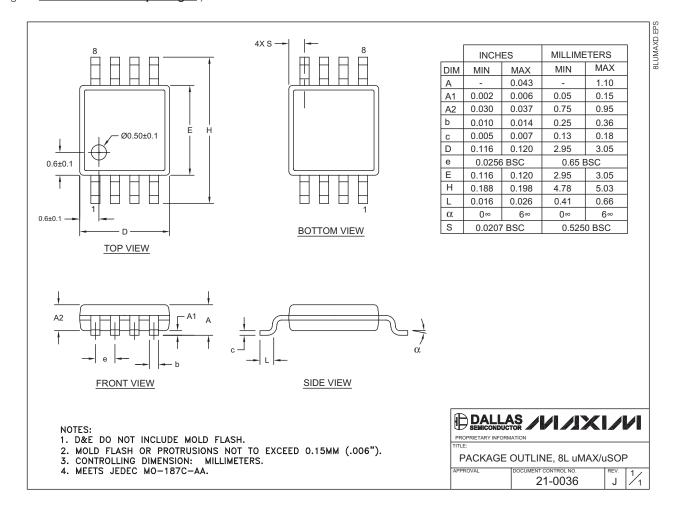
### **Package Information**

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to **www.maxim-ic.com/packages**.)



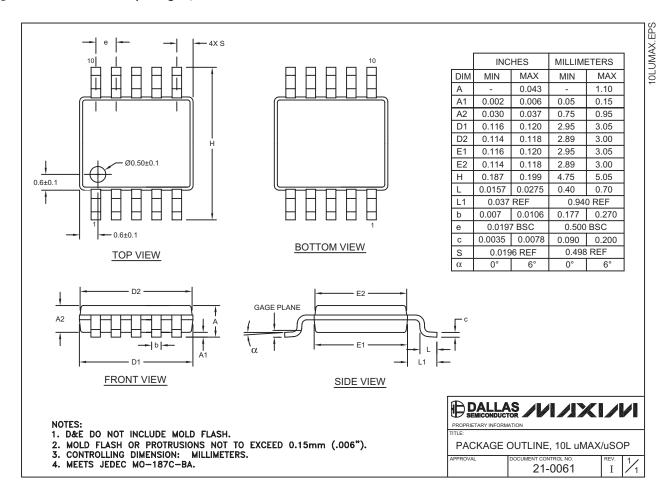
### Package Information (continued)

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### **Package Information (continued)**

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