



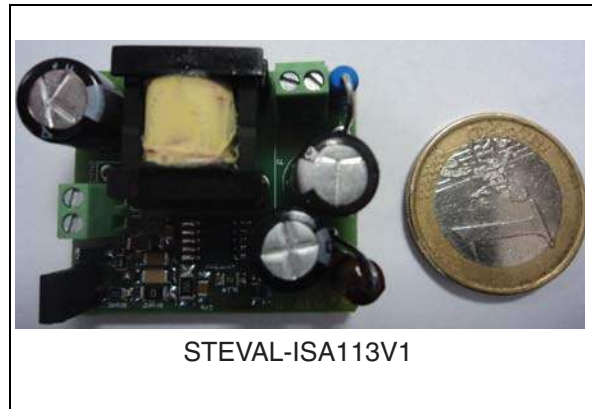
# STEVAL-ISA113V1

## Wide range single-output demonstration board based on the VIPER06HS

Data brief

### Features

- Universal input mains range:
  - input voltage 90 - 265 V<sub>AC</sub>
  - frequency 45 - 65 Hz
- Single-output voltage: 12 V at 0.35 A continuous operation
- Standby mains consumption: < 30 mW at 230 V<sub>AC</sub>
- Average efficiency: > 74%
- Fully protected against faults (overload, feedback disconnection and overheating)
- EMI: according to EN55022-Class-B
- RoHS compliant



### Description

The STEVAL-ISA113V1 demonstration board is a 12 V-0.35 A power supply set in non-isolated flyback topology using the new VIPER06HS offline high-voltage converter by STMicroelectronics.

The features of the device include an 800 V avalanche rugged power section, PWM operation at 115 kHz with frequency jittering for lower EMI, current limiting with adjustable set point, onboard soft-start, a safe auto-restart after a fault condition and a low standby power.

The protection features available include a thermal shutdown with hysteresis, delayed overload protection, and open loop failure protection.

# 1 Adapter features

The electrical specifications are given in [Table 1](#), the schematic in [Figure 1](#), and the bill of material in [Table 2](#).

**Table 1. Electrical specifications**

Parameter	Symbol	Value
Input voltage range	$V_{IN}$	[90 V <sub>AC</sub> ; 265 V <sub>AC</sub> ]
Output voltage	$V_{OUT}$	12 V
Max. output current	$I_{OUT}$	0.35 A
Precision of output regulation	$\Delta V_{OUT\_LF}$	±5%
High frequency output voltage ripple	$\Delta V_{OUT\_HF}$	50 mV
Max. ambient operating temperature	$T_{AMB}$	60 °C

**Table 2. Bill of material**

Ref.	Part	Description	Package	Manufacturer
Cin1		2.2 µF, 400 V NHG series electrolytic capacitor		
Cin2		4.7 µF, 400 V AX series electrolytic capacitor		Saxon
CVDD		1 µF, 50 V electrolytic capacitor	1206	Murata
Cfilt1		100 nF, 50 V ceramic capacitor	0805	
Cfilt2	Not mounted			
Cc		10 nF, 50 V ceramic capacitor	1206	
Cp		1 nF, 50 V ceramic capacitor	1206	
Cfb		1 nF, 50 V ceramic capacitor	0805	
Cout		330 µF, 16 V ZL series ultra-low ESR electrolytic cap.		Rubycon
D0	MB6S	600 V, 1 A diode bridge	TO-269AA	Vishay
D2	STPS2H100	100 V, 2 A power Schottky rectifier	SMA	ST
Daux	1N4148W	Surface mount fast switching diode	SOD-123	Zetex
R0		4.7 Ω 3/4 W resistor		
RLIM		15 kΩ 5% 1/4 W resistor	0805	
Rc		47 kΩ 5% 1/4 W resistor	0805	
RfbH1		33 kΩ 1% 1/4 W resistor	0805	
RfbH2		0 Ω	1206	
RfbL1		12 kΩ 1% 1/4 W resistor	1206	
RfbL2		0.47 kΩ 1% 1/4 W resistor	0805	

**Table 2. Bill of material (continued)**

Ref.	Part	Description	Package	Manufacturer
IC1	VIPer06HS	Offline high-voltage PWM controller	SSO-10	ST
T1	1921.0040	Transformer		Magnetica
Lin	B82144A2105J	1 mH inductor LBC series		Epcos

The transformer core is a standard E13. The output voltage value is set in a simple way through the RfbH-RfbL voltage divider between the output terminal and the FB pin, according to the following formula:

**Equation 1**

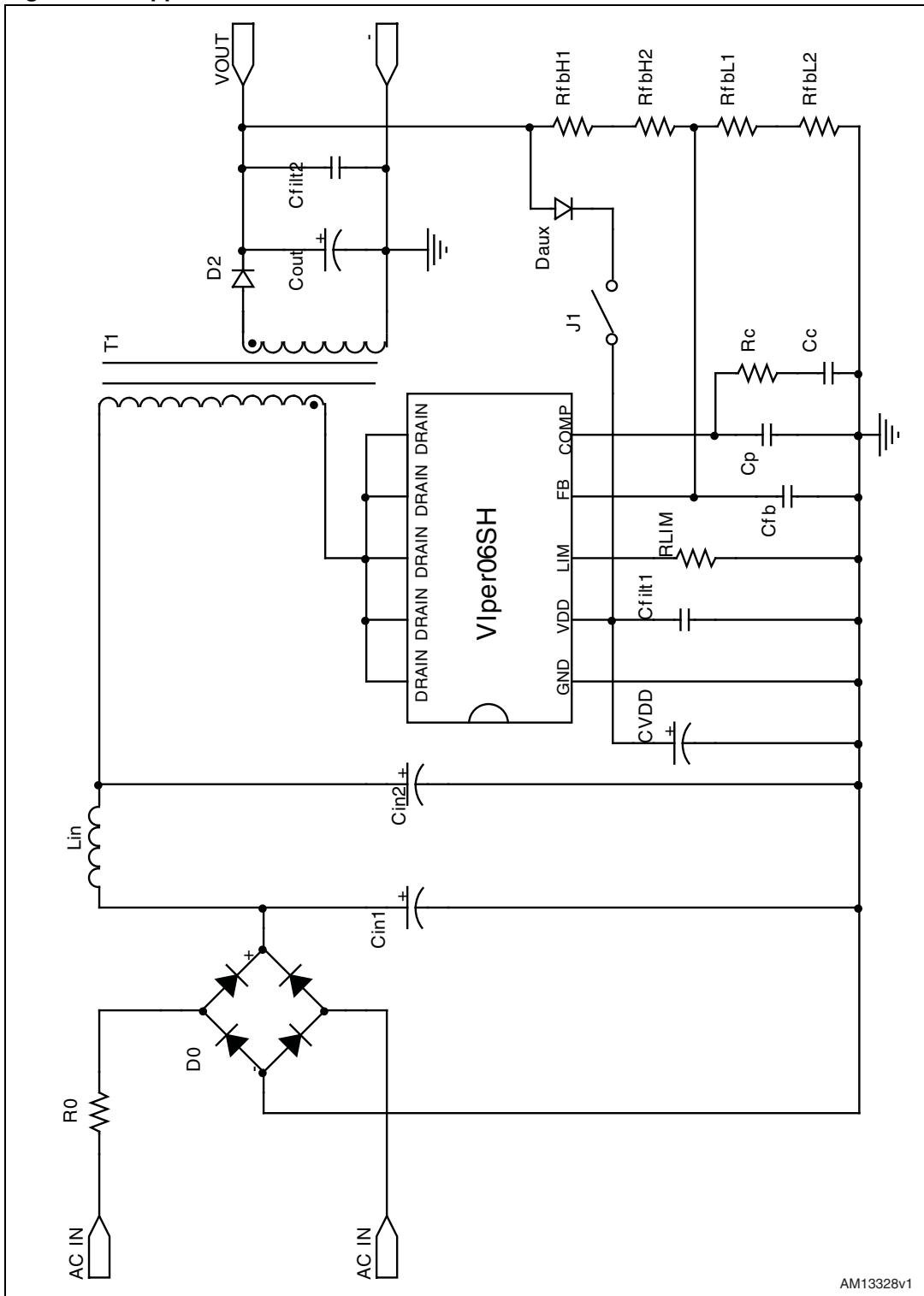
$$V_{OUT} = 3.3V \cdot \left( 1 + \frac{R_{fbH}}{R_{fbL}} \right)$$

In the schematic, RfbH has been split into RfbH1 and RfbH2; and RfbL into RfbL1 and RfbL2 in order to allow a better tuning of the output voltage value.

If the jumper J1 is not selected, the IC is biased through the internal HV-startup current generator (“self-biasing”).

If low standby consumption and good efficiency performance are required, the HV-startup current generator must be excluded. This can be done selecting the jumper J1, which connects the output terminal to the V<sub>DD</sub> pin through a small signal diode. The IC biasing through the output is referred to as “external biasing”.

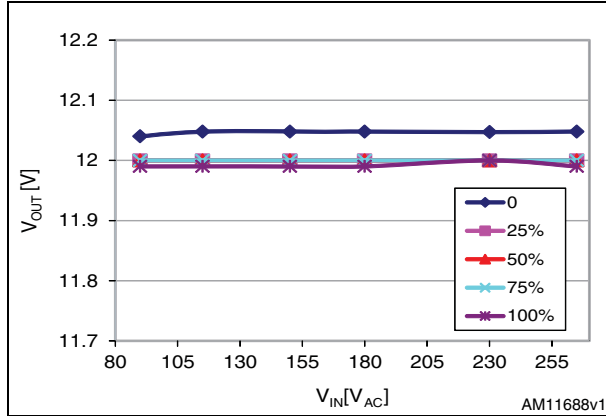
Figure 1. Application schematic



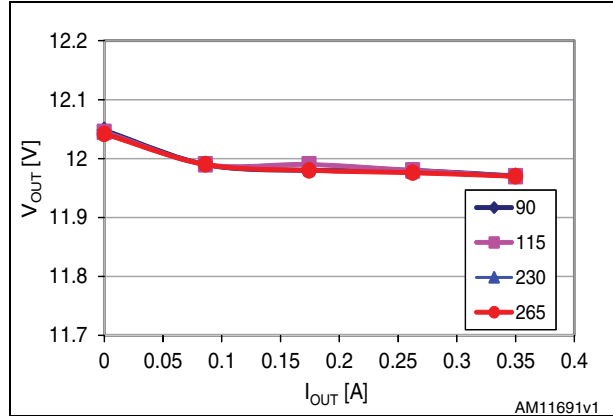
AM13328v1

## 2 Measurements

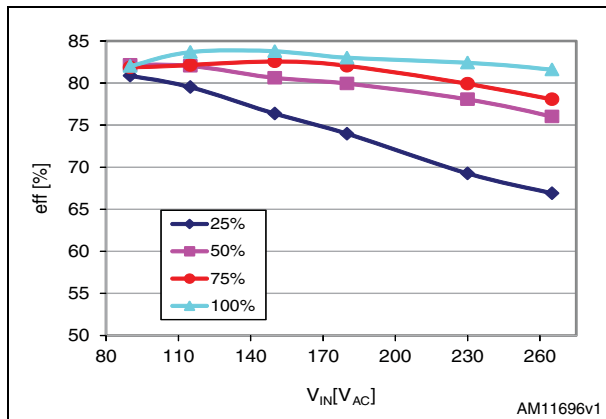
**Figure 2. Line regulation at different loads: IC externally biased (J1 selected)**



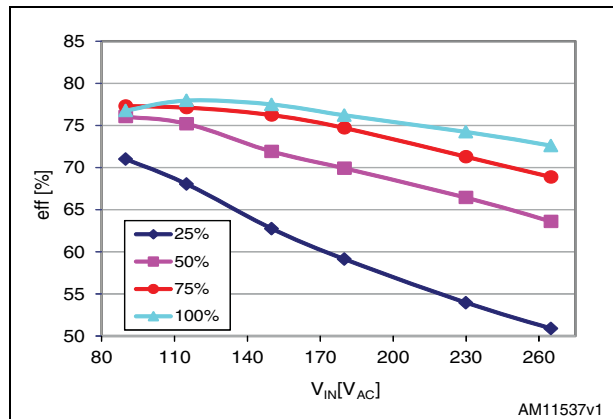
**Figure 3. Line regulation at different loads: IC self-biased (J1 not selected)**



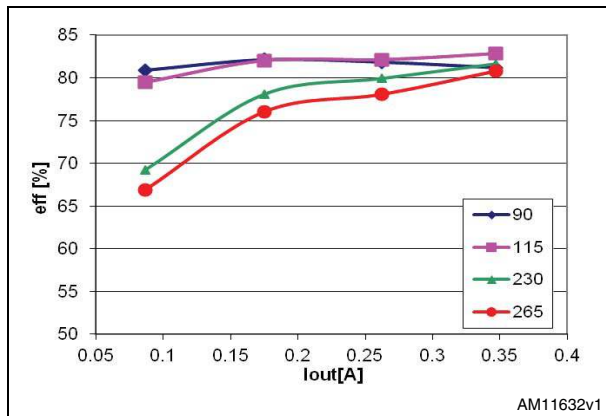
**Figure 4. Efficiency vs. Vin IC externally biased (J1 selected)**



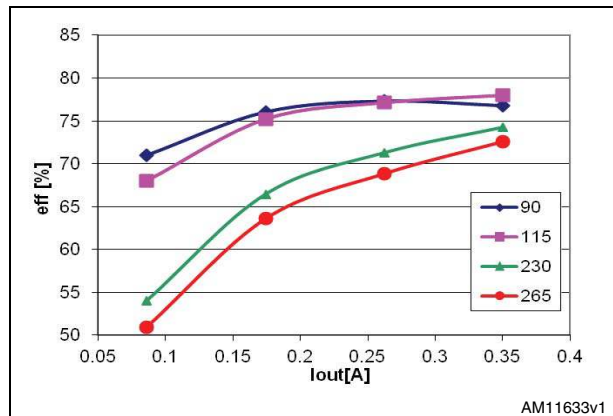
**Figure 5. Efficiency vs. Vin IC self-biased (J1 not selected)**



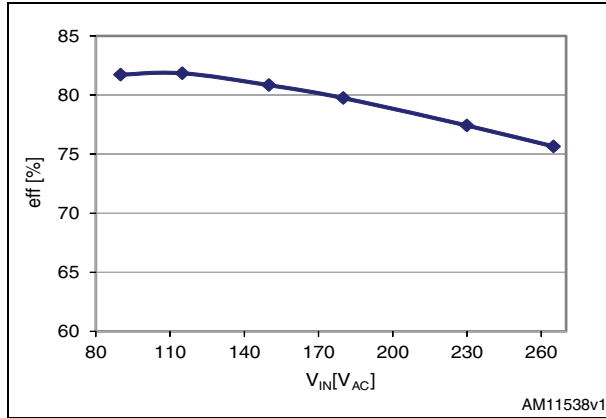
**Figure 6. Efficiency at different input voltages: IC externally biased (J1 selected)**



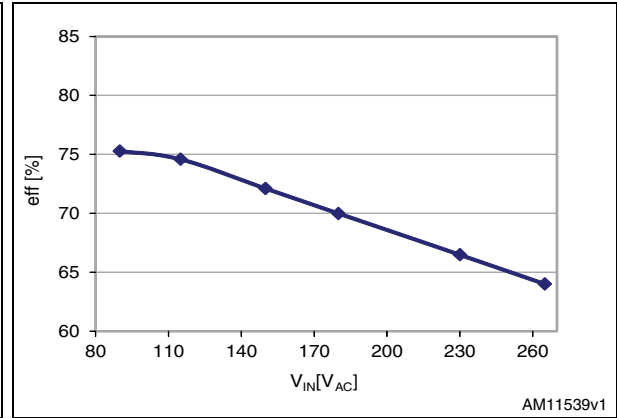
**Figure 7. Efficiency at different input voltages: IC self-biased (J1 not selected)**



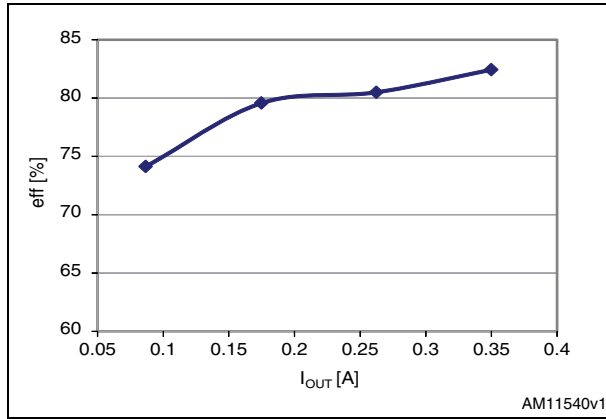
**Figure 8. Active mode efficiency vs.  $V_{IN}$  IC externally biased (J1 selected)**



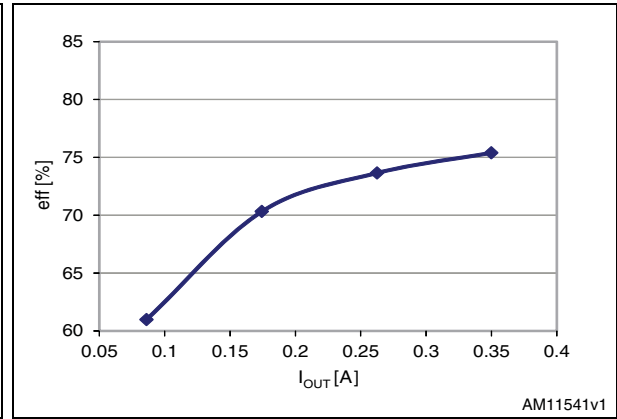
**Figure 9. Active mode efficiency vs.  $V_{IN}$  IC self-biased (J1 not selected)**



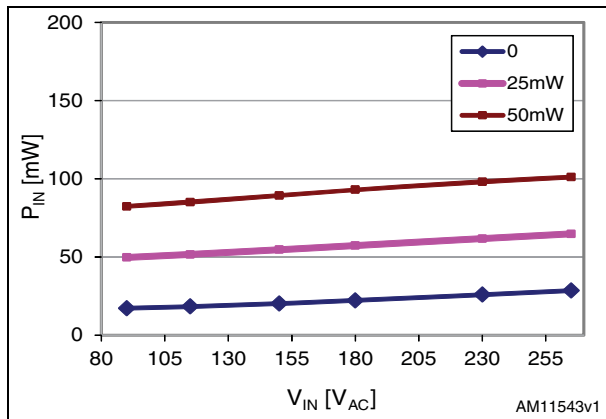
**Figure 10. Input voltage averaged efficiency vs. load IC externally biased (J1 selected)**



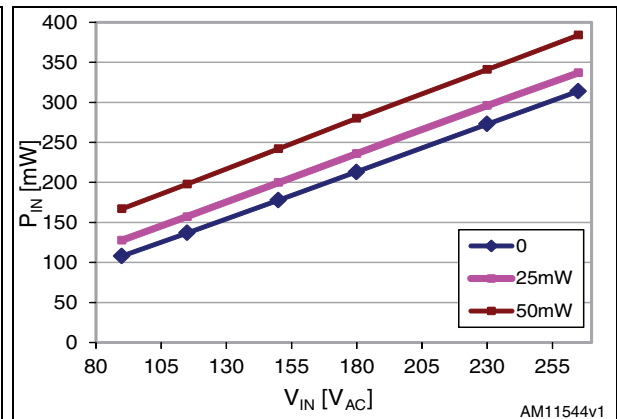
**Figure 11. Input voltage averaged efficiency vs. load IC self-biased (J1 not selected)**



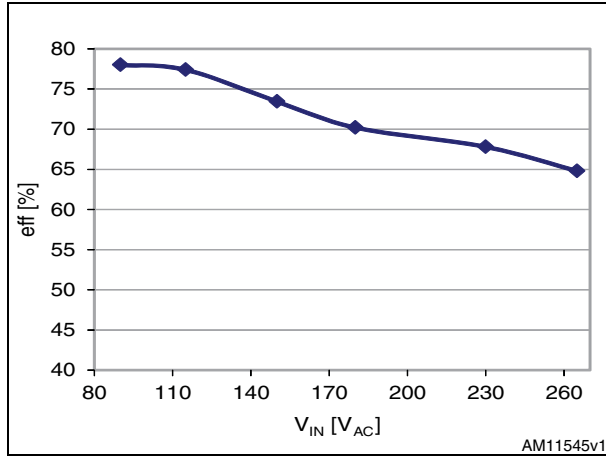
**Figure 12.  $P_{IN}$  vs.  $V_{IN}$  at no load and light load: IC externally biased (J1 selected)**



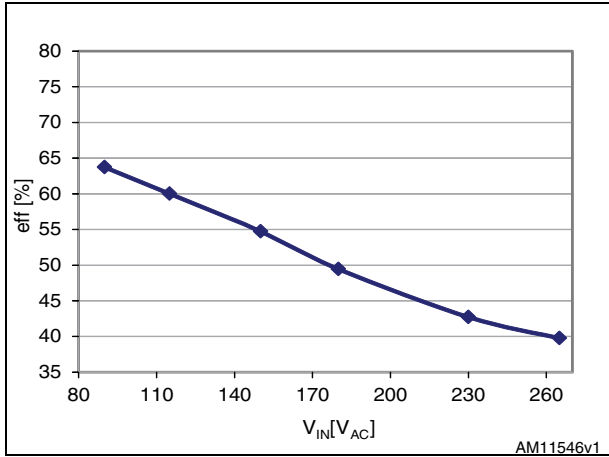
**Figure 13.  $P_{IN}$  vs.  $V_{IN}$  at no load and light load: IC self-biased (J1 not selected)**



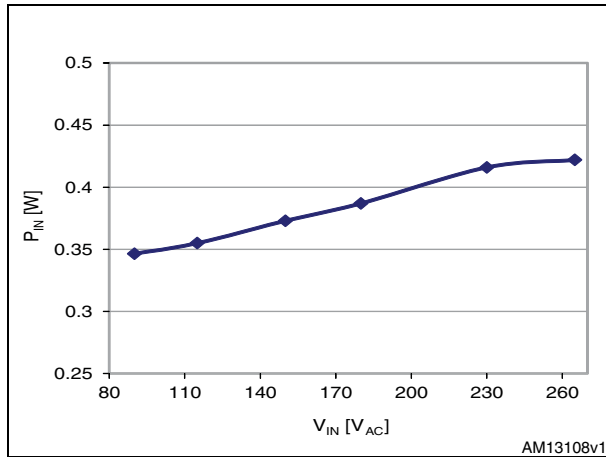
**Figure 14. Efficiency at  $P_{IN} = 1$  W: IC externally biased (J1 selected)**



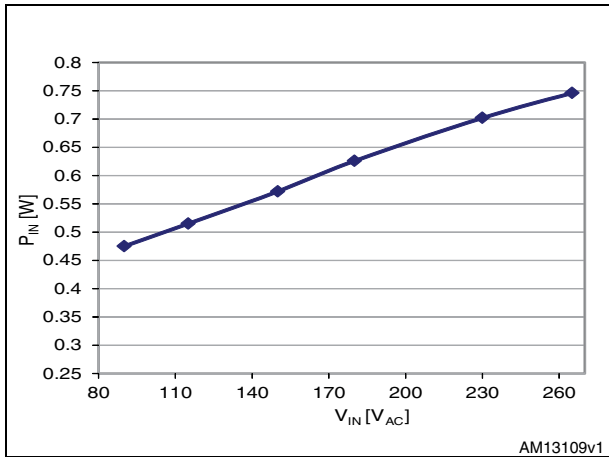
**Figure 15. Efficiency at  $P_{IN} = 1$  W: IC self-biased (J1 not selected)**



**Figure 16.  $P_{IN}$  at  $P_{OUT} = 250$  mW: IC externally biased (J1 selected)**



**Figure 17.  $P_{IN}$  at  $P_{OUT} = 250$  mW: IC self-biased (J1 not selected)**









## 4 Revision history

**Table 3. Document revision history**

Date	Revision	Changes
10-Jan-2013	1	Initial release.

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