DATASHEET

GENERAL DESCRIPTION

The 841654 is an optimized PCIe and sRIO clock generator. The device uses a 25MHz parallel crystal to generate 100MHz and 125MHz clock signals, replacing solutions requiring multiple oscillator and fanout buffer solutions. The device has excellent phase jitter (< 1ps rms) suitable to clock components requiring precise and low-jitter PCIe or sRIO or both clock signals. Designed for telecom, networking and industrial applications, the 841654 can also drive the high-speed sRIO and PCIe SerDes clock inputs of communication processors, DSPs, switches and bridges.

FEATURES

- Four differential HCSL clock outputs: configurable for PCIe (100MHz) and sRIO (100MHz or 125MHz) clock signals One REF_OUT LVCMOS/LVTTL clock output
- Selectable crystal oscillator interface, 25MHz, 18pF parallel resonant crystal or LVCMOS/LVTTL single-ended reference clock input
- Supports the following output frequencies: 100MHz or 125MHz
- VCO: 500MHz
- PLL bypass and output enable
- RMS phase jitter at 100MHz, using a 25MHz crystal (1.875MHz - 20MHz): 0.44ps (typical)
- Full 3.3V power supply mode
- -40°C to 85°C ambient operating temperature
- Available in lead-free (RoHS 6) package

28-Lead TSSOP 6.1mm x 9.7mm x 0.925mm package body **G Package** Top View

TABLE 1. PIN DESCRIPTIONS

NOTE: Pullup and Pulldown refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

TABLE 2. PIN CHARACTERISTICS

TABLE 3A. FSELX FUNCTION TABLE (f ref = 25MHZ)

TABLE 3B. REF_SEL FUNCTION TABLE

TABLE 3C. BYPASS FUNCTION TABLE

NOTE 1: Asynchronous function.

TABLE 3D. MR/nOE FUNCTION TABLE

NOTE 1: Asynchronous function.

TABLE 3E. nREF_OE FUNCTION TABLE

NOTE 1: Asynchronous function.

ABSOLUTE MAXIMUM RATINGS

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the DC Characteristics or AC Characteristics is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

$\bf{Table 4A.}$ \bf{Power} \bf{Supply} $\bf{ChARACTERISTICS, V_{_{DD}}=V_{_{DDOA}}=V_{_{DDOB}}=3.3V\pm5\%, T\text{}A=\text{}-40^{\circ}\bf{C}$ to $\bf{85^{\circ}C}$

$\mathbf{T}_{\mathbf{A}}$ ble <code>4B. LVCMOS</code> / <code>LVTTL DC Characteristics, V</sup> $_{\text{DD}}$ = 3.3V±5%, Ta = -40°C to 85°C $\,$ </code>

NOTE 1: Outputs terminated with 50 Ω to V_{$_{\text{\tiny{D}}}$}/2. See Parameter Measurement Information Section, Output Load Test Circuit diagram.

TABLE 5. CRYSTAL CHARACTERISTICS

NOTE: Characterized using an 18pF parallel resonant crystal.

$\bf{Table 6A.}$ LVCMOS AC Characteristics, $\rm V_{_{\rm DD}}$ = $\rm 3.3V\pm5\%$, Ta = -40°C to 85°C

$\bf{Table 6B. \; HCSL \; AC \; ChARACTERISTICS, V}_{_{DDOA}}=V_{_{DDOA}}=3.3V\pm5\%, T{\rm A}=-40^{\circ}{\rm C}$ to 85°C

NOTE: All specifications are taken at 100MHz and 125MHz.

NOTE 1: Please refer to the Phase Noise Plot.

NOTE 2: Defined as skew between outputs at the same supply voltage and with equal load conditions.

Measured at the output differential cross points.

NOTE 3: This parameter is defined in accordance with JEDEC Standard 65.

RENESAS 841654 DATA SHEET

FEMTOCLOCKS™ CRYSTAL-TO-HCSL CLOCK GENERATOR

PARAMETER MEASUREMENT INFORMATION

PARAMETER MEASUREMENT INFORMATION, CONTINUED

FEMTOCLOCKS™ CRYSTAL-TO-HCSL CLOCK GENERATOR

PARAMETER MEASUREMENT INFORMATION, CONTINUED

SE MEASUREMENT POINTS FOR ABSOLUTE CROSS POINT/SWING

APPLICATION INFORMATION

POWER SUPPLY FILTERING TECHNIQUES

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. The 841654 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL. $\mathsf{V}_{_{\sf DD}}$ $\mathsf{V}_{_{\sf DDA}}$ $\mathsf{V}_{_{\sf DDAA}}$ and $\bm{{\mathsf{V}}}_{_{\texttt{DDOB}}}$ should be individually connected to the power supply plane through vias, and 0.01µF bypass capacitors should be used for each pin. Figure 1 illustrates this for a generic V_{D} pin and also shows that V $_{_{\rm{DDA}}}$ requires that an additional10 Ω resistor along with a 10µF bypass capacitor be connected to the $V_{_{DDA}}$ pin.

RECOMMENDATIONS FOR UNUSED INPUT AND OUTPUT PINS

INPUTS:

CRYSTAL INPUTS

For applications not requiring the use of the crystal oscillator input, both XTAL_IN and XTAL_OUT can be left floating. Though not required, but for additional protection, a 1kΩ resistor can be tied from XTAL_IN to ground.

REF_IN INPUT

For applications not requiring the use of the reference clock, it can be left floating. Though not required, but for additional protection, a 1k Ω resistor can be tied from the REF_IN to ground.

LVCMOS CONTROL PINS

All control pins have internal pull-ups or pull-downs; additional resistance is not required but can be added for additional protection. A 1kΩ resistor can be used.

OUTPUTS:

HCSL OUTPUTS

All unused HCSL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

LVCMOS OUTPUT

The unused LVCMOS output can be left floating. We recommend that there is no trace attached.

CRYSTAL INPUT INTERFACE

The 841654 has been characterized with 18pF parallel resonant crystals. The capacitor values shown in Figure 2 below were determined using a 25MHz, 18pF parallel resonant crystal and were chosen to minimize the ppm error.

FIGURE 2. CRYSTAL INPUt INTERFACE

LVCMOS TO XTAL INTERFACE

The XTAL_IN input can accept a single-ended LVCMOS signal through an AC couple capacitor. A general interface diagram is shown in Figure 3. The XTAL_OUT pin can be left floating. The input edge rate can be as slow as 10ns. For LVCMOS inputs, it is recommended that the amplitude be reduced from full swing to half swing in order to prevent signal interference with the power rail and to reduce noise. This configuration requires that the output impedance of the driver (Ro) plus the series resistance (Rs) equals the transmission line impedance. In addition, matched termination at the crystal input will attenuate the signal in half. This can be done in one of two ways. First, R1 and R2 in parallel should equal the transmission line impedance. For most 50Ω applications, R1 and R2 can be 100Ω. This can also be accomplished by removing R1 and making R2 50Ω.

FIGURE 3. GENERAL DIAGRAM FOR LVCMOS DRIVER TO XTAL INPUT INTERFACE

SCHEMATIC LAYOUT

Figure 4 shows an example of 841654 application schematic. In this example, the device is operated at $V_{\rm cc}$ = 3.3V. The 18pF parallel resonant 25MHz crystal is used. The C1 = 27pF and C2 = 27pF are recommended for frequency accuracy. For different board layout, the C1 and C2 may be slightly adjusted

for optimizing frequency accuracy. One example of HCSL and one example of LVCMOS terminations are shown in this schematic. The decoupling capacitors should be located as close as possible to the power pin.

FIGURE 4. 841654 SCHEMATIC LAYOUT

POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the 841654. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the 841654 is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for V_{DD} = 3.3V + 5% = 3.465V, which gives worst case results. **NOTE:** Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)_{MAX} = V_{DD_MAX} * I_{DD_MAX} = 3.465V * 85mA = **294.5mW**
- Power (outputs)_{MAX} = 50.06mW/Loaded Output pair If all outputs are loaded, the total power is 4 * 50.06mW = **200.24mW**

Total Power _MAX (3.465V, with all outputs switching) = 294.5mW + 200.24mW = 494.74mW

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockSTM devices is 125 $^{\circ}$ C.

The equation for Tj is as follows: $Tj = \theta_{JA} * Pd_total + T_A$

 $Ti =$ Junction Temperature

 θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in Section 1 above)

 T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance $\theta_{\mathbb{A}}$ must be used. Assuming no air flow and a multi-layer board, the appropriate value is 64.5°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is: 85°C + 0.495W $*$ 64.5°C/W = 116.9°C. This is below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

TABLE 7. THERMAL RESISTANCE θ**JA FOR 28-LEAD TSSOP, FORCED CONVECTION**

3. Calculations and Equations.

The purpose of this section is to calculate power dissipation on the IC per HCSL output pair.

HCSL output driver circuit and termination are shown in Figure 4.

FIGURE 4. HCSL DRIVER CIRCUIT AND TERMINATION

HCSL is a current steering output which sources a maximum of 17mA of current per output. To calculate worst case on-chip power dissipation, use the following equations which assume a 50Ω load to ground.

The highest power dissipation occurs when V_{DD} is HIGH.

Power = $(V_{\text{DD_HIGH}} - V_{\text{OUT}})$ * I_{OUT} since $V_{\text{OUT}} = I_{\text{OUT}}$ * R_L

- $=(V_{\text{DD_HIGH}} I_{\text{OUT}}^* R_L)^* I_{\text{OUT}}$
- $= (3.465V 17mA * 50Ω) * 17mA$

Total Power Dissipation per output pair = **50.06mW**

RECOMMENDED TERMINATION

Figure 5A is the recommended termination for applications which require the receiver and driver to be on a separate PCB. All traces should be 50Ω impedance.

FIGURE 5A. RECOMMENDED TERMINATION

Figure 5B is the recommended termination for applications which require a point to point connection and contain the driver and receiver on the same PCB. All traces should all be 50Ω impedance.

FIGURE 5B. RECOMMENDED TERMINATION

RELIABILITY INFORMATION

\mathbf{T}_{A} Ble $\mathbf{8}. \ \theta_{\mathsf{A}}$ vs. Air Flow Table for 28 Lead TSSOP

TRANSISTOR COUNT

The transistor count for 841654 is: 2954

PACKAGE OUTLINE AND PACKAGE DIMENSIONS

PACKAGE OUTLINE - G SUFFIX FOR 28 LEAD TSSOP TABLE 9. PACKAGE DIMENSIONS

Reference Document: JEDEC Publication 95, MO-153

TABLE 10. ORDERING INFORMATION

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

REVISION HISTORY SHEET

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