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# SG5841J — Highly Integrated Green-Mode PWM Controller

#### **Features**

- Green-Mode PWM Controller
- Low Startup Current : 14µA
- Low Operating Current: 4mA
- Programmable PWM Frequency with Hopping
- Peak-Current-Mode Control
- Cycle-by-Cycle Current Limiting
- Synchronized Slope Compensation
- Leading-Edge Blanking (LEB)
- Constant Output Power Limit
- Totem Pole Output with Soft Driving
- V<sub>DD</sub> Over-Voltage Clamping
- Programmable Over-Temperature Protection (OTP)
- Internal Open-Loop Protection
- V<sub>DD</sub> Under-Voltage Lockout (UVLO)
- GATE Output Maximum Voltage Clamp:18V

#### **Applications**

General-purpose, switch-mode, power supplies and flyback power converters, including:

- Power Adapters
- Open-Frame SMPS

#### Description

The highly integrated SG5841/J series of PWM controllers provides several features to enhance the performance of flyback converters.

To minimize standby power consumption, a proprietary green-mode function provides off-time modulation to continuously decrease the switching frequency at light-load conditions. This green-mode function enables the power supply to meet international power conservation requirements. To further reduce power consumption, SG5841/J is manufactured using the BiCMOS process. This allows a low startup current, around 14 $\mu$ A, and an operating current of only 4mA. As a result, a large startup resistance can be used.

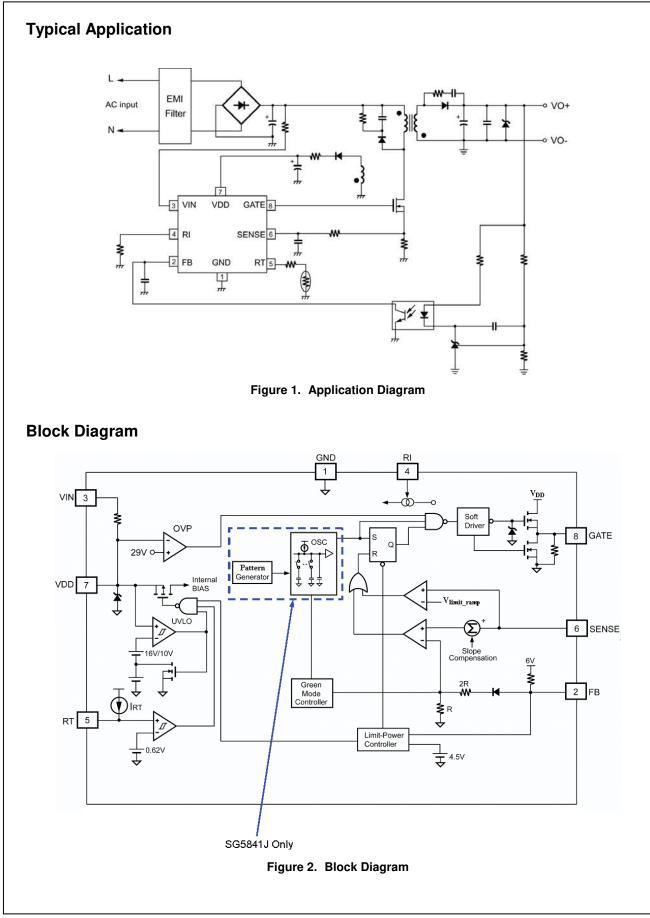
The built-in synchronized slope compensation achieves stable peak-current-mode control. The proprietary internal sawtooth power-limiter ensures a constant output power limit over a wide range of AC input voltages, from  $90V_{AC}$  to  $264V_{AC}$ .

SG5841/J provides many protections. In addition to cycle-by-cycle current limiting, the internal open-loop protection circuit ensures safety should an open-loop or output-short-circuit failure occur. PWM output is disabled until  $V_{DD}$  drops below the UVLO lower limit, then the controller restarts. An external NTC thermistor can be applied for over-temperature protection.

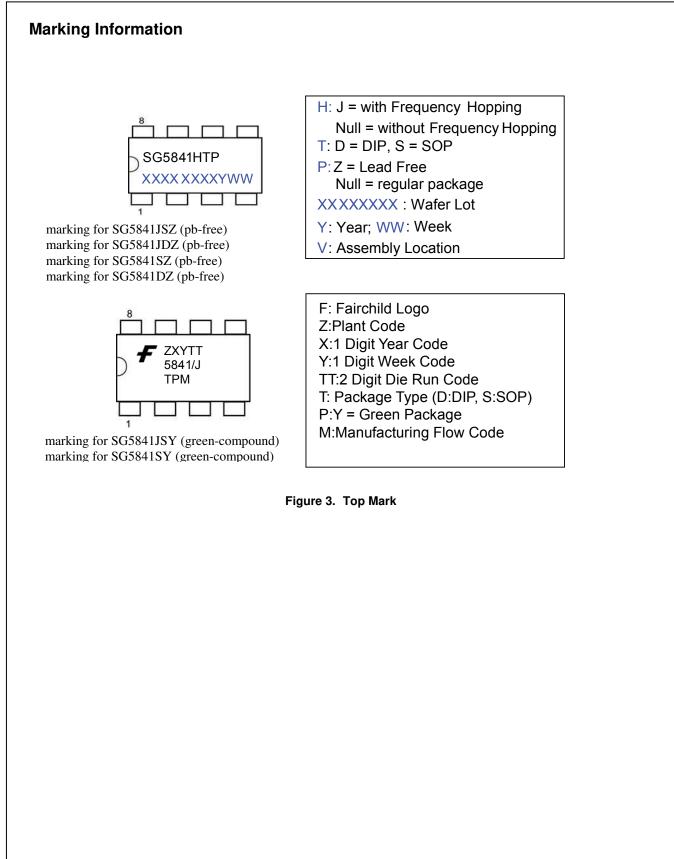
SG5841/J is available in an 8-pin DIP or SOP package.

Part Number	Ambient Operating Temperature Range		Package
SG5841JSZ	-20 to +85°C	Yes	8-Pin Small Outline Package (SOP)
SG5841JSY	-20 to +85°C	Yes	8-Pin Small Outline Package (SOP)
SG5841SZ	-20 to +85°C	No	8-Pin Small Outline Package (SOP)
SG5841SY	-20 to +85°C	No	8-Pin Small Outline Package (SOP)

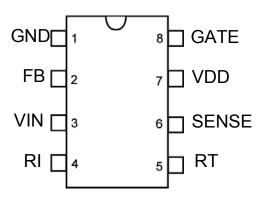
#### **Ordering Information**

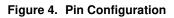


SG5841J — Highly Integrated Green-Mode PWM Controller



#### **Pin Configuration**





#### **Pin Definitions**

Pin #	Name	Function	Description
1	GND	Ground	Ground.
2	FB	Feedback	The signal from the external compensation circuit is fed into this pin. The PWM duty cycle is determined in response to the signal from this pin and the current-sense signal from pin 6. If FB voltage exceeds the threshold, the internal protection circuit disables PWM output after a predetermined delay time.
3	VIN	Startup Input	For startup, this pin is pulled HIGH to the rectified line input via a resistor. Since the startup current requirement is very small, a large startup resistance is used to minimize power loss.
4	RI	Reference Setting	A resistor connected from the RI to GND provides a constant current source. This determines the center PWM frequency. Increasing the resistance reduces PWM frequency. Using a $26K\Omega$ resistor results in a $65KHz$ center PWM frequency.
5	RT	Temperature Detection	For over-temperature protection. An external NTC thermistor is connected from this pin to the GND pin. The impedance of the NTC decreases at high temperatures. Once the voltage of the RT pin drops below a fixed limit, PWM output is disabled.
6	SENSE	Current Sense	Current sense. The sensed voltage is used for peak-current-mode control and cycle-by-cycle current limiting.
7	VDD	Power Supply	Power supply. If $V_{\text{DD}}$ exceeds a threshold, the internal protection circuit disables PWM output.
8	GATE	Driver Output	The totem-pole output driver for the power MOSFET, which is internally clamped below 18V.

#### **Absolute Maximum Ratings**

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only. All voltage values, except differential voltages, are given with respect to GND pin.

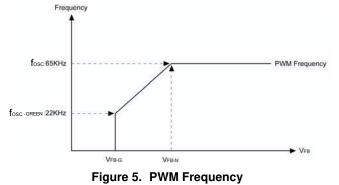
Symbol	Parameter		Min.	Max.	Unit
$V_{\text{DD}}$	Supply Voltage			30	V
V <sub>IN</sub>	Input Terminal			30	V
$V_{FB}$	Input Voltage to FB Pin		-0.3	7.0	V
VSENSE	Input Voltage to SENSE Pin		-0.3	7.0	V
V <sub>RT</sub>	Input Voltage to RT Pin		-0.3	7.0	V
V <sub>RI</sub>	Input Voltage to RI Pin		-0.3	7.0	V
D				800	
PD	Power Dissipation ( $T_A < 50^{\circ}C$ )	SOP		400	mW
0		DIP		82.5	°C/M
ΑLΘ	Thermal Resistance (Junction-to-Air)	SOP		141	°C/W
0				59.7	°C/M
ΟLC	Thermal Resistance (Junction-to-Case)	SOP		80.8	°C/W
TJ	Operating Junction Temperature		-40	+125	°C
T <sub>STG</sub>	Storage Temperature Range		-55	+150	°C
TL	Lead Temperature (Wave Soldering or Infrared, 10 Seconds)			260	°C
FOD	Human Body Model, JESD22-A114			3	kV
E9D	ESD Charged Device Model, JESD22-C101			250	V

#### **Recommended Operating Conditions**

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings.

Symbol	Parameter	Min.	Max.	Unit
T <sub>A</sub>	Operating Ambient Temperatures	-20	+85	°C

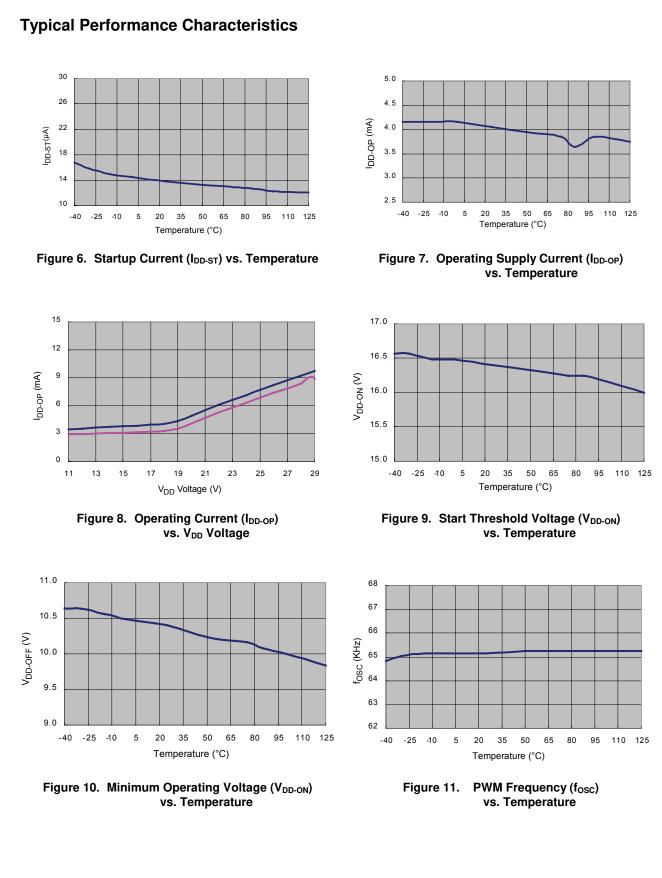
Symbol	Pa	arameter	Conditions	Min.	Тур.	Max.	Units
V <sub>DD</sub> Section	on						
V <sub>DD-OP</sub>	Continuously	Operating Voltage				24.7	V
V <sub>DD-ON</sub>	Start Threshold Voltage			15	16	17	V
V <sub>DD-OFF</sub>	Minimum Ope	erating Voltage		9	10	11	V
I <sub>DD-ST</sub>	Startup Curre	nt	V <sub>DD</sub> =V <sub>DD-ON</sub> -0.16V		14	30	μA
I <sub>DD-OP</sub>	Operating Su	oply Current	$V_{DD}$ =15V, R <sub>I</sub> =26K $\Omega$ , GATE=OPEN		4	5	mA
V <sub>DD-CLAMP</sub>	V <sub>DD</sub> Over-Volt	age-Clamping Level		28	29		V
t <sub>D-VDDCLAMP</sub>	V <sub>DD</sub> Over-Voltage-Clamping Debounce Time		R <sub>i</sub> =26KΩ	50	100	200	μs
R <sub>I</sub> Sectior	1						
RI <sub>NOR</sub>	R <sub>I</sub> Operating	Range		15.5		36.0	KΩ
RI <sub>MAX</sub>	Maximum R <sub>I</sub> Value for Protection				230		KΩ
RI <sub>MIN</sub>	Minimum $R_I V$	alue for Protection			10		KΩ
Oscillator	Section						
	Normal	Center Frequency	R <sub>I</sub> =26KΩ	62	65	68	
fosc	PWM Frequency	Hopping Range	R <sub>I</sub> =26KΩ (SG5841J only)	±3.7	±4.2	±4.7	KHz
t <sub>HOP</sub>	Hopping Period		Rι=26KΩ (SG5841J only)	3.9	4.4	4.9	ms
f <sub>OSC-G</sub>	Green-Mode	Frequency	R <sub>I</sub> =26KΩ	18	22	25	KHz
$\mathbf{f}_{DV}$	Frequency Variation vs. V <sub>DD</sub> Deviation		V <sub>DD</sub> =11.5V to 24.7V			5	%
f <sub>DT</sub>	Frequency Va Temperature		T <sub>A</sub> =-20 to +85°C			5	%
Feedback	Input Section	n					
A <sub>V</sub>	FB Input to Current Comparator Attenuation			1/3.75	1/3.20	1/2.75	V/V
Z <sub>FB</sub>	Input Impedance			4		7	KΩ
$V_{\text{FB-OPEN}}$	FB Output High Voltage		FB pin open	5	6		V
$V_{\text{FB-OLP}}$	FB Open-Loop Trigger Level			4.2	4.5	4.8	V
t <sub>D-OLP</sub>	Delay Time of FB Pin Open-Loop Protection		R <sub>i</sub> =26KΩ	26	29	32	ms
$V_{\text{FB-N}}$	Green-Mode	Entry FB Voltage	R <sub>I</sub> =26KΩ	1.9	2.1	2.3	V
$V_{\text{FB-G}}$	Green-Mode	Ending FB Voltage	R <sub>I</sub> =26KΩ		V <sub>FB-N</sub> -0.5		V

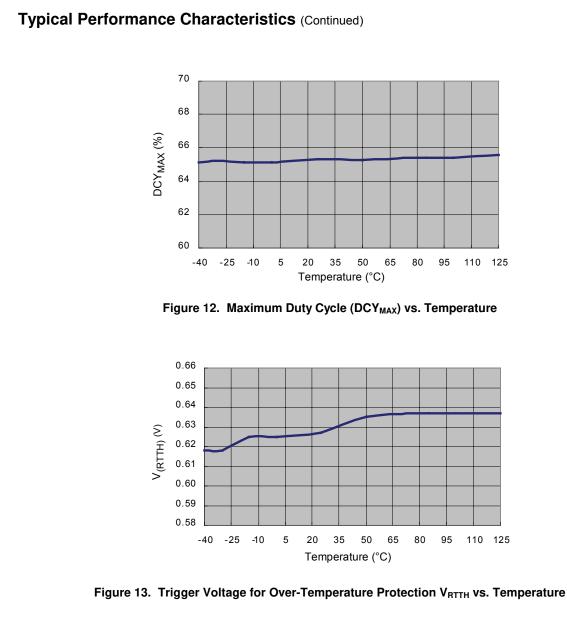


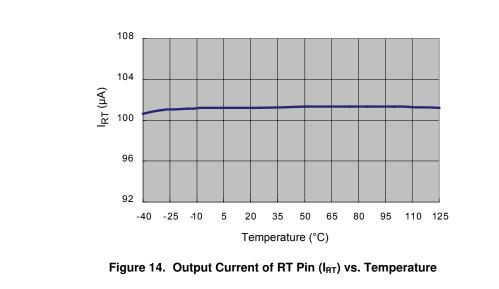
#### Electrical Characteristics (Continued)

 $V_{DD}$  = 15V,  $T_A$  = 25°C, unless otherwise noted.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
Current-	Sense Section			1		
ZSENSE	Input Impedance			12		KΩ
VSTHFL	Current Limit Flatten Threshold Voltage		0.85	0.90	0.95	V
V <sub>STHVA</sub>	Current Limit Valley Threshold Voltage	V <sub>STHFL</sub> –V <sub>STHVA</sub>		0.22		V
t <sub>PD</sub>	Propagation Delay to GATE Output	R <sub>I</sub> =26KΩ		150	200	ns
t <sub>LEB</sub>	Leading-Edge Blanking Time	R <sub>I</sub> =26KΩ	200	270	350	ns
GATE Se	ection					
$DCY_MAX$	Maximum Duty Cycle		60	65	70	%
$V_{\text{GATE-L}}$	Output Voltage Low	V <sub>DD</sub> =15V, I <sub>O</sub> =50mA			1.5	V
$V_{\text{GATE-H}}$	Output Voltage High	$V_{DD}$ =12.5V, $I_O$ =50mA	7.5			V
tr	Rising Time	$V_{DD}$ =15V, C <sub>L</sub> =1nF	150	250	350	ns
t <sub>f</sub>	Falling Time	$V_{DD}$ =15V, C <sub>L</sub> =1nF	30	50	90	ns
I <sub>0</sub>	Peak Output Current	V <sub>DD</sub> =15V, GATE=6V	230			mA
V <sub>GATE-</sub> CLAMP	Gate Output Clamping Voltage	V <sub>DD</sub> =24.7V		18	19	V
RT Secti	on					
I <sub>RT</sub>	Output Current of RT Pin	R <sub>I</sub> =26KΩ	92	100	108	μA
V <sub>RTTH</sub>	Trigger Voltage for Over- Temperature Protection		0.585	0.620	0.655	V
$V_{\text{RT-RLS}}$	OTP Release Voltage			V <sub>RTTH</sub> +0.03		V
t <sub>D-OTP</sub>	Over-Temperature Debounce	R <sub>I</sub> =26KΩ	60	100	140	μs







#### **Functional Description**

#### **Startup Current**

Typical startup current is only 14µA, which allows a high-resistance and low-wattage startup resistor to minimize power loss. For an AC/DC adapter with universal input range, a 1.5M $\Omega$ , 0.25W startup resistor and a 10µF/25V V<sub>DD</sub> hold-up capacitor are enough for this application.

#### **Operating Current**

Operating current is around 4mA. The low operating current enables better efficiency and reduces the requirement of  $V_{DD}$  hold-up capacitance.

#### **Green-Mode Operation**

The proprietary green-mode function provides off-time modulation to continuously decrease the PWM frequency under light-load conditions. To avoid acoustic noise problems, the minimum PWM frequency is set above 22KHz. Green mode dramatically reduces power consumption under light-load and zero-load conditions. Power supplies using a SG5841/J controller can meet restrictive international regulations regarding standby power consumption.

#### **Oscillator Operation**

A resistor connected from the RI pin to the GND pin generates a constant current source for the SG5841/J controller. This current is used to determine the center PWM frequency. Increasing the resistance reduces PWM frequency. Using a  $26K\Omega$  resistor, R<sub>I</sub>, results in a corresponding 65KHz PWM frequency. The relationship between R<sub>I</sub> and the switching frequency is:

$$f_{PWM} = \frac{1690}{\mathsf{R}_{\mathsf{I}}(\mathsf{K}\Omega)}(\mathsf{KHz}) \tag{1}$$

The range of the PWM oscillation frequency is designed as 47KHz  $\sim 109$ KHz.

SG5841J also integrates a frequency hopping function internally. The frequency variation ranges from around 62KHz to 68KHz for a center frequency of 65KHz. The frequency-hopping function helps reduce EMI emission of a power supply with minimum line filters.

#### **Current Sensing / PWM Current Limiting**

Peak-current-mode control is utilized in to regulate output voltage and provide pulse-by-pulse current limiting. The switch current is detected by a sense resistor into the SENSE pin. The PWM duty cycle is determined by this current-sense signal and the feedback voltage. When the voltage on the SENSE pin reaches around  $V_{COMP} = (V_{FB}-1.0)/3.2$ , a switch cycle is terminated immediately.  $V_{COMP}$  is internally clamped to a variable voltage around 0.85V for output power limit.

#### Leading-Edge Blanking (LEB)

Each time the power MOSFET is switched on, a turn-on spike occurs at the sense-resistor. To avoid premature termination of the switching pulse, a leading-edge blanking time is built in. During this blanking period, the current-limit comparator is disabled and cannot switch off the gate drive.

#### **Under-Voltage Lockout (UVLO)**

The turn-on and turn-off thresholds are fixed internally at 16V and 10V. During startup, the hold-up capacitor must be charged to 16V through the startup resistor to enable the IC. The hold-up capacitor continues to supply  $V_{DD}$  before the energy can be delivered from auxiliary winding of the main transformer.  $V_{DD}$  must not drop below 10V during this startup process. This UVLO hysteresis window ensures that hold-up capacitor is adequate to supply  $V_{DD}$  during startup.

#### Gate Output / Soft Driving

The SG5841/J BiCMOS output stage is a fast totempole gate driver. Cross conduction has been avoided to minimize heat dissipation, increase efficiency, and enhance reliability. The output driver is clamped by an internal 18V Zener diode to protect power MOSFET transistors against undesirable gate over-voltage. A soft driving waveform is implemented to minimize EMI.

#### **Built-in Slope Compensation**

The sensed voltage across the current-sense resistor is used for peak-current-mode control and pulse-by-pulse current limiting. Built-in slope compensation improves stability or prevents sub-harmonic oscillation. SG5841/J inserts a synchronized, positive-going ramp at every switching cycle.

#### **Constant Output Power Limit**

When the SENSE voltage across the sense resistor,  $R_{S}$ , reaches the threshold voltage, around 0.85V, the output GATE drive is turned off after delay,  $t_{PD}$ . This delay introduces additional current, proportional to  $t_{PD} \bullet V_{IN} / L_P$ . The delay is nearly constant, regardless of the input voltage  $V_{IN}$ . Higher input voltage results in larger additional current and the output power limit is higher than under low-input line voltage. To compensate this variation for a wide AC input range, a sawtooth power-limiter (saw limiter) is designed to solve the unequal power-limit problem. The saw limiter is designed as a positive ramp signal (V\_limit\_ramp) and fed to the inverting input of the OCP comparator. This results in a lower current limit at high-line inputs than at low-line inputs.

#### **V**<sub>DD</sub> Over-Voltage Clamping

 $V_{\text{DD}}$  over-voltage clamping prevents damage due to abnormal conditions. If  $V_{\text{DD}}$  voltage is over the  $V_{\text{DD}}$  over-voltage clamping voltage ( $V_{\text{DD-CLAMP}}$ ) and lasts for  $t_{\text{D-VDDCLAMP}}$ , the PWM pulses are disabled until the  $V_{\text{DD}}$  drops below the  $V_{\text{DD}}$  over-voltage clamping voltage.

#### **Thermal Protection**

An NTC thermistor  $R_{NTC}$  in series with a resistor  $R_A$  can be connected from the RT pin to ground. A constant current  $I_{RT}$  is output from pin RT. The voltage on the RT pin can be expressed as  $V_{RT} = I_{RT} \times (R_{NTC} + R_A)$ , in which  $I_{RT} = 2 \times (1.3V / R_I)$ . At high ambient temperature,  $R_{NTC}$  is smaller, such that  $V_{RT}$  decreases. When  $V_{RT}$  is less than 0.62V, the PWM is completely turned off.

#### **Limited Power Control**

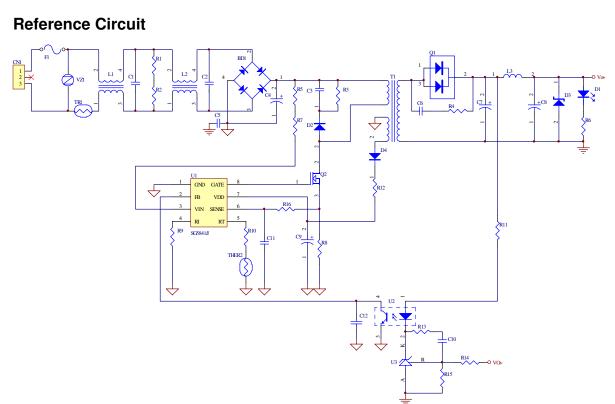
The FB voltage increases every time the output of the power supply is shorted or overloaded. If the FB voltage remains higher than a built-in threshold for longer than  $t_{D-OLP}$ , PWM output is turned off. As PWM output is turned off, the supply voltage  $V_{DD}$  begins decreasing.

$$t_{\text{D-OLP (ms)}} = 1.115 \times R_{\text{I}}(K\Omega)$$
(2)

When  $V_{DD}$  goes below the turn-off threshold (e.g. 10V) the controller totally shuts down.  $V_{DD}$  is charged up to the turn-on threshold voltage of 16V through the startup resistor until PWM output is restarted. This protection remains activated as long as the overloading condition persists. This prevents the power supply from overheating due to overloading conditions.

#### **Noise Immunity**

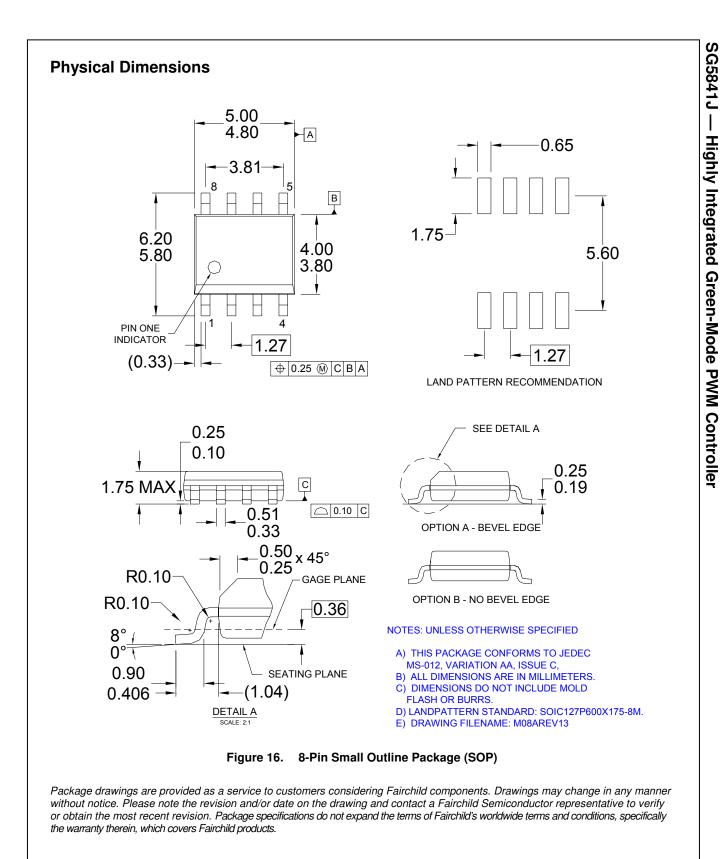
Noise on the current-sense or control signal may cause significant pulse-width jitter, particularly in the continuous-conduction mode. Slope compensation helps alleviate this problem. Good placement and layout practices should be followed. Avoiding long PCB traces and component leads, locating compensation and filter components near SG5841/J, and increasing power MOS gate resistance improve performance.

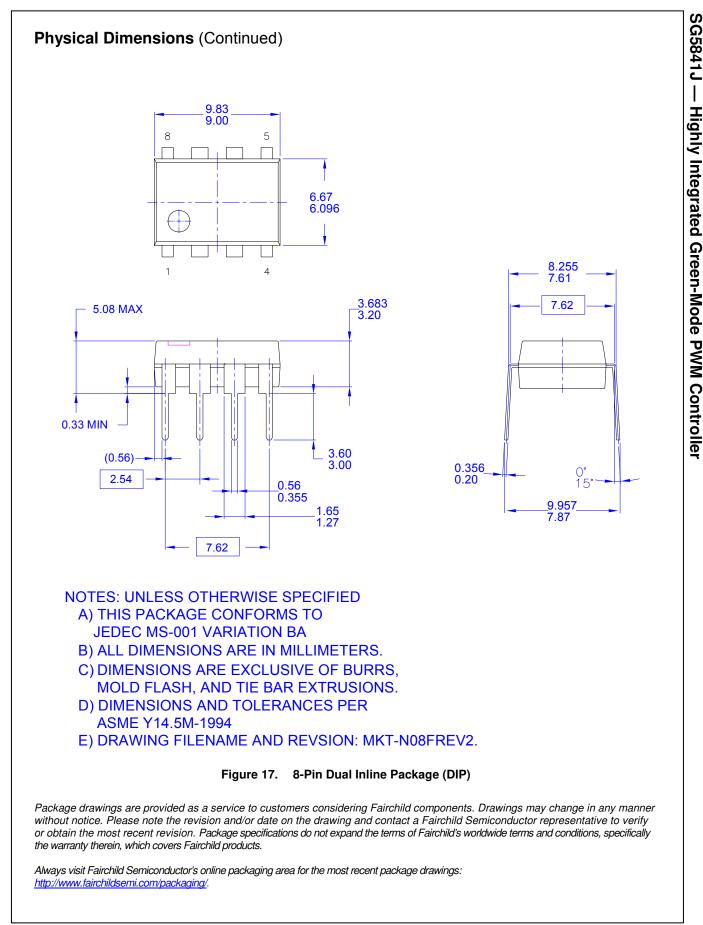




#### BOM

Reference Reference Component Component BD1 BD 4A/600V Q2 MOS 7A/600V C1 XC 0.68µF/300V R1, R2 R 1MΩ 1/4W C2 R3 XC 0.1µF/300V R 100KW 1/2W C3 CC 0.01µF/500V R4 R 47Ω 1/4W C4 EC 120µ/400V R5, R7 R 750KΩ 1/4W C5 YC 222p/250V R6 R 2KΩ 1/8W C6 R8 CC 1000pF/100V R 0.3Ω 2W C7 EC 1000µF/25V R9 R 33KΩ 1/8W C8 EC 470µF/25V R10 R 4.7KΩ 1/8W 1% C9 R11 EC 10µF/50V R 470Ω 1/8W R12 C10 CC 222pF/50V R 0Ω 1/8W C11 CC 470pF/50V R13 R 4.7KΩ 1/8W C12 CC 102pF/50V (Option) R14 R 154KΩ 1/8W D1 LED R15 R 39KΩ 1/8W D2 Diode BYV95C R16 R 100Ω 1/8W D3 TVS P6KE16A THER2 Thermistor TTC104 D4 Diode FR103 Τ1 Transformer (600µH-PQ2620) F1 FUSE 4A/250V U1 IC SG5841/J U2 L1 Choke (900µH) **IC PC817** U3 IC TL431 L2 Choke (10mH) L3 Inductor (2µH) VZ1 VZ 9G Q1 Diode 20A/100V





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