

MAX5113

9-Channel, 14-Bit Current DAC with SPI Interface

General Description

The MAX5113 is a 14-bit, 9-channel, current-output digital-to-analog converter (DAC). The device operates from a low +3.0V power supply and provides 14-bit performance without any adjustment.

The device's output ranges are optimized to bias a high-power tunable laser source. Each of the 9 channels provides a current source. Channels 1 and 2 provide 10mA current. An internal multiplexer switches the outputs of each channel to one of four external nodes. Channel 3 provides a selectable current of 2mA or 20mA. Channel 4 provides 90mA. Channel 5 provides 180mA. Channel 6 provides a selectable current of -60mA or +300mA. Channel 7 provides 90mA. Channels 8 and 9 provide a selectable current of 15mA or 35mA. Connect DAC outputs in parallel to obtain additional current or to achieve higher resolution. The device contains an internal reference.

An SPI interface drives the device with clock rates of up to 25MHz. An active-high asynchronous CLR input resets DAC codes to zero independent of the serial interface. The device provides a separate power-supply input for driving the interface logic.

The MAX5113 is specified over the -40°C to +105°C temperature range, and is available in 3mm x 3mm, 36-bump WLP and 5mm x 5mm, 32-pin TQFN packages.

Features

- ◆ Low 3.0V Supply
- ◆ Integrated Multiplexers for Outputs 1 and 2
- ◆ Increased Current or Resolution with Outputs Connected in Parallel
- ◆ SPI-Compatible Serial Interface
- ◆ Internal Reference
- ◆ Overtemperature Protection
- ◆ Operates Over the -40°C to +105°C Temperature Range
- ◆ Available in 36-Bump WLP or 32-Pin TQFN Packages

Applications

Tunable Laser Diode Biasing

Ordering Information

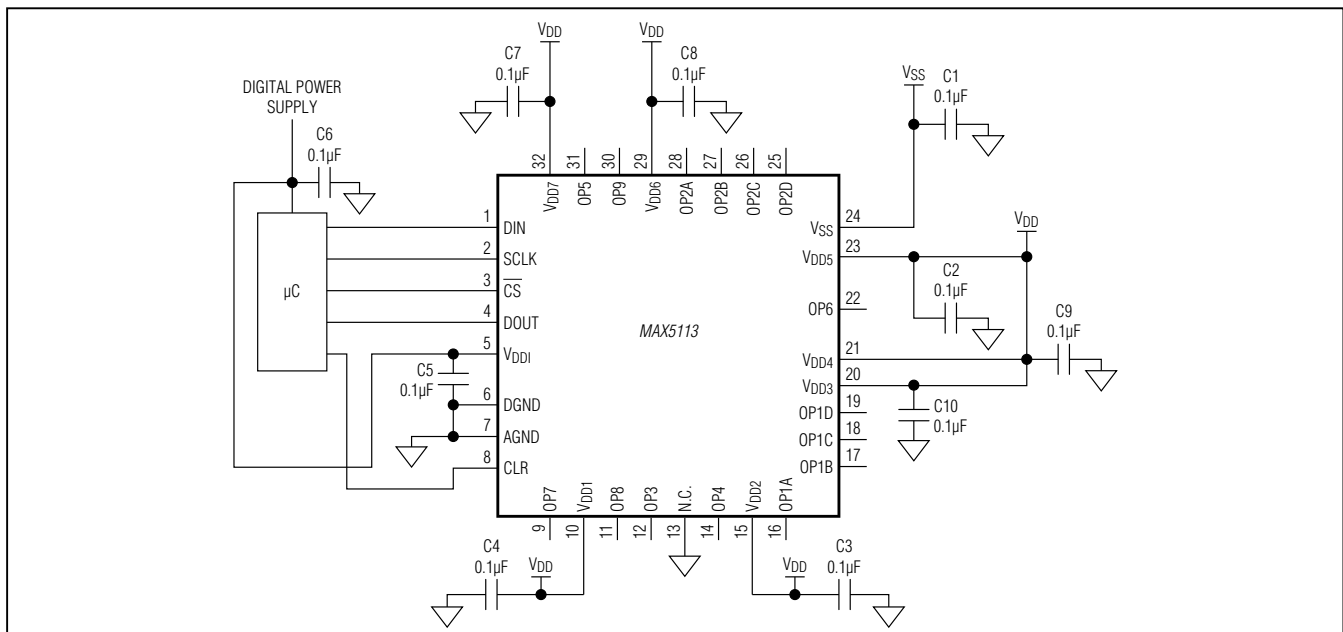
PART	INTERFACE	PIN-PACKAGE
MAX5113GWX+T	SPI	36 WLP
MAX5113GTJ+	SPI	32 TQFN-EP*

Note: All devices are specified over the -40°C to +105°C operating temperature range.

+Denotes a lead(Pb)-free/RoHS-compliant package.

*EP = Exposed pad.

Typical Operating Circuit



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ABSOLUTE MAXIMUM RATINGS

VDD to AGND	-0.3V to +4.0V
VSS to AGND	-6.0V to +0.3V
VDDI to AGND	-0.3V to +6.0V
OP6 to AGN	the higher of (VDD - 9V), (VSS - 0.3V) and -6.0V to the lower of (VDD + 0.3V) and +4.0V
OP1 to OP5 and OP7, OP8,	
OP9 to AGND	-0.3V to the lower of (VDD + 0.3V) and +4.0V
DOUT to DGND	-0.3V to the lower of (VDDIO + 0.3V) and +6.0V
N.C. to AGND	-0.3V to the lower of (VDD + 0.3V) and +4.0V
Digital I/Os to DGND	-0.3V to +6.0V

AGND to DGND	-0.3V to +0.3V
All Other Pins to AGND	-0.3V to +4.0V
Continuous Power Dissipation (TA = +70°C)	
WLP (derate at 26.3mW/°C above +70°C)	2104mW
TQFN (derate at 34.5mW/°C above +70°C)	2758mW
Maximum Current Into Any Pin	380mA
Operating Temperature Range	-40°C to +105°C
Storage Temperature Range	-65°C to +150°C
Junction Temperature	+150°C
Lead Temperature (TQFN only, soldering, 10s)	+300°C
Soldering Temperature (reflow)	+260°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Package Thermal Characteristics (Note 1)

TQFN	Junction-to-Ambient Thermal Resistance (θ_{JA})	29°C/W	WLP	Junction-to-Ambient Thermal Resistance (θ_{JA})	38°C/W
	Junction-to-Case Thermal Resistance (θ_{JC})	1.7°C/W			

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

ELECTRICAL CHARACTERISTICS

(VDD = +2.6V to +3.3V, VSS = -4.75V to -5.46V, VDDI = +1.8V to +5.25V, AGND = DGND, TA = -40°C to +105°C, VOP1-VOP5 = VOP6 sourcing = VOP7, VOP8, and VOP9 = VDD - 1V, VOP6 sinking = VSS + 1V, unless otherwise noted. Typical specifications are at VDD = +3.0V, VSS = -5.2V, TA = +25°C. Specifications apply to all DACs and outputs, unless otherwise noted.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
STATIC PERFORMANCE						
Resolution	N		14			Bits
Differential Nonlinearity	DNL	Guaranteed monotonic		±0.5	±1.0	LSB
Integral Nonlinearity	INL	OP1-OP6 source and OP9		±2	±8	LSB
		OP6 sink		±8		
Ideal Gain	IGAIN			$I_{MAX}/2^{14}$		mA/LSB
Gain Error (Note 3)	GE	All but OP3, 2mA and OP6 sink			±1.3	%FS
		OP3, 2mA			±1.5	
		OP6 sink			±5	
Gain Error Tempco (Note 4)	GETC	All but OP6 sink			±50	ppm/°C
		OP6 sink			±15	
Full-Scale Output	IMAX	OP1 and OP2			10	mA
		OP3	2mA FS range		2	
			20mA FS range		20	
		OP4			90	
		OP5			180	
		OP6 current source			300	
		OP6 current sink			-60	
		OP7			90	
		OP8 and OP9	15mA FS range		15	
35mA FS range			35			

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ELECTRICAL CHARACTERISTICS (continued)

(V_{DD} = +2.6V to +3.3V, V_{SS} = -4.75V to -5.46V, V_{DDI} = +1.8V to +5.25V, AGND = DGND, T_A = -40°C to +105°C, V_{OP1}–V_{OP5} = V_{OP6} sourcing = V_{OP7}, V_{OP8}, and V_{OP9} = V_{DD} - 1V, V_{OP6} sinking = V_{SS} + 1V, unless otherwise noted. Typical specifications are at V_{DD} = +3.0V, V_{SS} = -5.2V, T_A = +25°C. Specifications apply to all DACs and outputs, unless otherwise noted.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
Offset Error (Note 3)	OE	OP1 and OP2	-120	-60	0	μA	
		OP3	2mA FS range	-24	-12		0
			20mA FS range	-240	-120		0
		OP4	-1080	-540	0		
		OP5	-2160	-1080	0		
		OP6 current source	-3600	-1800	0		
		OP6 current sink	0	+360	+720		
		OP7	-1080	-540	0		
OP8 and OP9	15mA FS range	-180	-90	0			
	35mA FS range	-420	-210	0			
Offset Error Tempco (Note 4)	OETC	OP1 and OP2			±250	nA/°C	
		OP3	2mA FS range				±50
			20mA FS range				±500
		OP4			±2250		
		OP5			±4500		
		OP6 current source			±7500		
		OP6 current sink			±1500		
		OP7			±2250		
OP8 and OP9	15mA FS range			±375			
	35mA FS range			±875			
Output Compliance Range	V _{OR}	All but OP6 sink	V _{GN} D		V _{DD} - 1	V	
		OP6 sink	V _{SS} + 1		V _{DD}		
DYNAMIC PERFORMANCE							
Output Resistance	R _{OUT}	OP1 and OP2		2		MΩ	
		OP3	2mA FS range		10		
			20mA FS range		1		
		OP4		0.2			
		OP5		0.1			
		OP6 current source		0.06			
		OP6 current sink		0.04			
		OP7		0.2			
OP8 and OP9	15mA FS range		1.3				
	35mA FS range		0.56				
Current-Output Slew Rate	SR	OP1 and OP2		5		mA/μs	
		OP3	2mA FS range		1		
			20mA FS range		10		
		OP4		45			
		OP5		90			
		OP6 current source		150			
		OP6 current sink		30			
		OP7		45			
OP8 and OP9	15mA FS range		7.5				
	35mA FS range		17.5				

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DD} = +2.6V$ to $+3.3V$, $V_{SS} = -4.75V$ to $-5.46V$, $V_{DDI} = +1.8V$ to $+5.25V$, $AGND = DGND$, $T_A = -40^{\circ}C$ to $+105^{\circ}C$, $V_{OP1}-V_{OP5} = V_{OP6}$ sourcing = V_{OP7} , V_{OP8} , and $V_{OP9} = V_{DD} - 1V$, V_{OP6} sinking = $V_{SS} + 1V$, unless otherwise noted. Typical specifications are at $V_{DD} = +3.0V$, $V_{SS} = -5.2V$, $T_A = +25^{\circ}C$. Specifications apply to all DACs and outputs, unless otherwise noted.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Output Settling Time	t_{OS}	$t_O \pm 0.1\%$		15		μs
Noise at Full Scale (10kHz)	INO	OP1 and OP2		1.6		nA/ \sqrt{Hz}
		OP3	2mA FS range	0.4		
			20mA FS range	3.4		
		OP4		16		
		OP5		31		
		OP6 current source		56		
		OP6 current sink		11		
		OP7		16		
		OP8 and OP9	15mA FS range	2.8		
35mA FS range	6.5					
DAC Glitch Impulse Major-Carry Transition	IOGE	OP1 and OP2		60		μC
		OP3, 20mA		120		
		OP4		540		
		OP5		1080		
		OP6 current source		1800		
		OP6 current sink		360		
		OP7		540		
		OP8 and OP9	15mA FS range	90		
			35mA FS range	210		
DAC Output GND Switch Resistance	R_{GSW}	At 0.7V			50	Ω
DAC Output GND Switch Current	I_{GSW}	At 0.7V	14			mA
OVERTEMPERATURE DETECTORS						
Overtemperature Disable Threshold	T_{OVTD}			+160		$^{\circ}C$
Overtemperature Warning Threshold	T_{OVTW}			+150		$^{\circ}C$
POWER REQUIREMENTS						
Power-Supply Range	V_{DD}		2.6		3.3	V
Interface Power-Supply Range	V_{DDI}		1.8		5.25	V
Negative Supply Range	V_{SS}		-5.46	-5.2	-4.75	V
Supply Current	I_{DD}	No load, no I/O		500	600	μA
Negative Supply Current	I_{SS}		-20	-11		μA
POWER-ON RESET (POR)						
POR Threshold	V_{POR}			1.6		V
POR Threshold Hysteresis	V_{PORH}			0.025		V

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ELECTRICAL CHARACTERISTICS (continued)

(V_{DD} = +2.6V to +3.3V, V_{SS} = -4.75V to -5.46V, V_{DDI} = +1.8V to +5.25V, AGND = DGND, T_A = -40°C to +105°C, V_{OP1}–V_{OP5} = V_{OP6} sourcing = V_{OP7}, V_{OP8}, and V_{OP9} = V_{DD} - 1V, V_{OP6} sinking = V_{SS} + 1V, unless otherwise noted. Typical specifications are at V_{DD} = +3.0V, V_{SS} = -5.2V, T_A = +25°C. Specifications apply to all DACs and outputs, unless otherwise noted.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
DIGITAL INPUT CHARACTERISTICS (DIN, SCLK, $\overline{\text{CS}}$, CLR)						
Input Low Voltage	V _{IL}	V _{DDI} = 2.2V to 5.25V			0.3 x V _{DDI}	V
		V _{DDI} = 1.8V to 2.2V			0.2 x V _{DDI}	V
Input High Voltage	V _{IH}	V _{DDI} = 2.2V to 5.25V	0.7 x V _{DDI}			V
		V _{DDI} = 1.8V to 2.2V	0.8 x V _{DDI}			V
Input Hysteresis	V _{HYS}			250		mV
Input Capacitance	C _{IN}			10		pF
Input Leakage Current	I _{IN}	Input = 0V or V _{DDI}			±10	µA
DIGITAL OUTPUT CHARACTERISTICS (DOUT)						
Output Low Voltage	V _{OL}	V _{DDI} = 2.2V to 5.5V, I _{SINK} = 5mA			0.40	V
		V _{DDI} = 1.8V to 2.2V, I _{SINK} = 2mA			0.25	
Output High Voltage	V _{OH}	V _{DDI} = 2.2V to 5.5V, I _{SOURCE} = 5mA	V _{DDI} - 0.5			V
		V _{DDI} = 1.8V to 2.2V, I _{SOURCE} = 2mA	V _{DDI} - 0.3			V
Output Three-State Leakage	I _{OZ}	BHEN = 0		±0.01	±1	µA
Output Bus Hold Sinking Current	I _{BHK}	DOUT low, BHEN = 1		-12		µA
Output Bus Hold Sourcing Current	I _{BHC}	DOUT high, BHEN = 1		+5		µA
Output Three-State Capacitance	C _{OZ}			11		pF
Output Short-Circuit Current	I _{OSS}	V _{DDI} = 5.25V		±100		mA
TIMING CHARACTERISTICS (Note 5)						
Serial Clock Frequency	f _{SCLK}		0		25	MHz
SCLK Period	t _{CP}	V _{DDI} = 1.8V to 5.25V	40			ns
SCLK Pulse Width Low	t _{CL}	V _{DDI} = 1.8V to 5.25V, 60% duty cycle	16			ns
SCLK Pulse Width High	t _{CH}	V _{DDI} = 1.8V to 5.25V, 40% duty cycle	16			ns
$\overline{\text{CS}}$ Fall to SCLK Fall Setup Time	t _{CSS0}	To first SCLK falling edge	16			ns
$\overline{\text{CS}}$ Fall to SCLK Fall Hold Time	t _{CSh0}	Applies to inactive FE preceding first FE	0			ns
$\overline{\text{CS}}$ Rise to SCLK Fall Hold Time	t _{CSh1}	Applies to 24th FE	0			ns
DIN-to-SCLK Fall Setup Time	t _{DS}		8			ns
DIN-to-SCLK Fall Hold Time	t _{DH}		8			ns
SCLK Fall to DOUT Settling Time	t _{DOT}	V _{DDI} = 2.2V to 5.5V, C _{LOAD} = 20pF			30	ns
		V _{DDI} = 1.8V to 2.2V, C _{LOAD} = 20pF			60	
SCLK Fall to DOUT Hold Time	t _{DOH}	C _{LOAD} = 0pF	2			ns
SCLK Fall to DOUT Disable	t _{DOZ}	24th active FE deassertion	2		30	ns

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DD} = +2.6V$ to $+3.3V$, $V_{SS} = -4.75V$ to $-5.46V$, $V_{DDI} = +1.8V$ to $+5.25V$, $AGND = DGND$, $T_A = -40^{\circ}C$ to $+105^{\circ}C$, $V_{OP1}\text{--}V_{OP5} = V_{OP6}$ sourcing = V_{OP7} , V_{OP8} , and $V_{OP9} = V_{DD} - 1V$, V_{OP6} sinking = $V_{SS} + 1V$, unless otherwise noted. Typical specifications are at $V_{DD} = +3.0V$, $V_{SS} = -5.2V$, $T_A = +25^{\circ}C$. Specifications apply to all DACs and outputs, unless otherwise noted.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
\overline{CS} Fall to DOUT Enable	tDOE	$V_{DDI} = 2.2V$ to $5.5V$, asynchronous	1		30	ns
		$V_{DDI} = 1.8V$ to $2.2V$, asynchronous	1		40	
\overline{CS} Rise to DOUT Disable	tCSDOZ	Stand alone, aborted sequence			35	ns
\overline{CS} Rise to SCLK Fall	tCSA	Applies to 24th FE, aborted sequence			20	ns
\overline{CS} Pulse-Width High	tCSPW	Stand alone	20			ns
SCLK Fall to \overline{CS} Fall	tCSF	Applies to 24th active FE	100			ns
CLR Fall to \overline{CS} Fall	tCLRCS	Applies to DACs in reset mode only	20			ns
CLR Pulse-Width High	tCLRPW	No DAC is in shutter or gate mode	40			ns
		Any DAC is in shutter or gate mode (Note 6)	500			

Note 2: Specifications are 100% production tested at $T_A \geq +25^{\circ}C$. Specifications for $T_A < +25^{\circ}C$ are guaranteed by design and characterization.

Note 3: Configuration register write operation required following power-up for output offset adjustment. See the *DAC Outputs* section in the *Detailed Description*. All gain and offset errors include the effect of the internal reference and are guaranteed over temperature. Gain error = (measured gain - I_{GAIN})/I_{GAIN}. Measured gain = (code 16383 DAC output - code 500 DAC output)/15883. Offset error = code 500 DAC output - (500 x measured gain). The device is trimmed such that offset error is negative, ensuring linear operation down to zero current output.

Note 4: Guaranteed by design and characterization. Not production tested. All gain and offset temperature coefficients include the effect of the internal reference. Temperature coefficients are calculated by the “box” method. For additional information, refer to Maxim Application Note 4300: *Calculating the error budget in precision digital-to-analog converter (DAC) applications*.

Note 5: Timing characteristics are tested and guaranteed with digital input conditions $V_{IH} = V_{DDI}$ and $V_{IL} = 0V$.

Note 6: Minimum pulse width required to realize functionally useful DAC transitions. Not production tested. See Shutter Mode Settling Time Down and Shutter Mode Settling Time Up graphs in the *Typical Operating Characteristics* section.

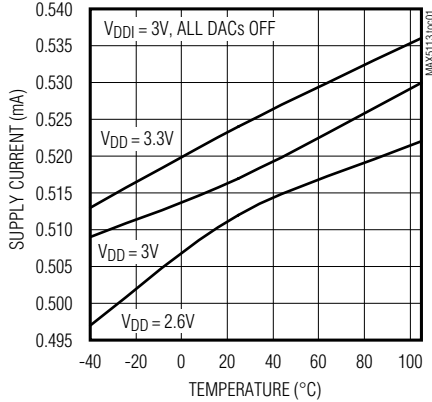
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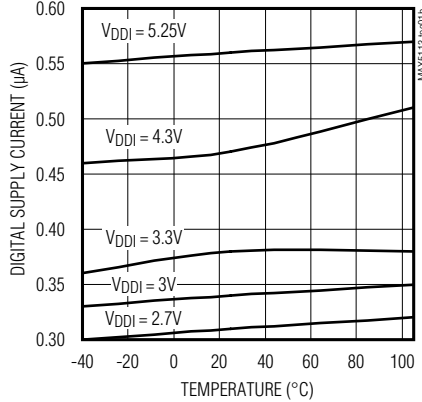
Typical Operating Characteristics

($V_{DD} = 3.0V$, $T_A = +25^\circ C$, unless otherwise noted.)

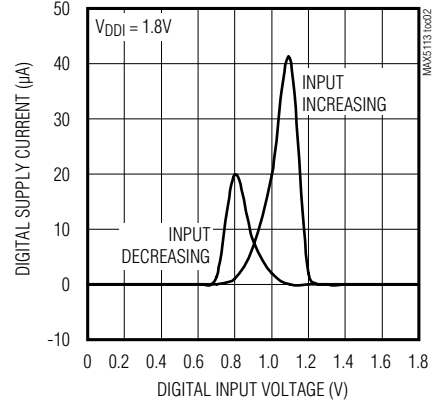
ANALOG SUPPLY CURRENT vs. TEMPERATURE



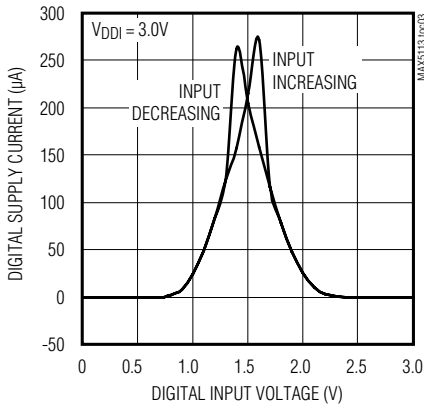
DIGITAL SUPPLY CURRENT vs. TEMPERATURE



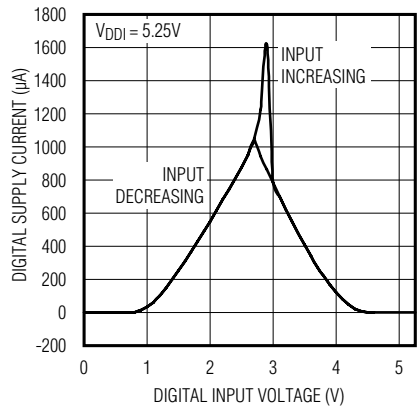
DIGITAL SUPPLY CURRENT vs. DIGITAL INPUT VOLTAGE



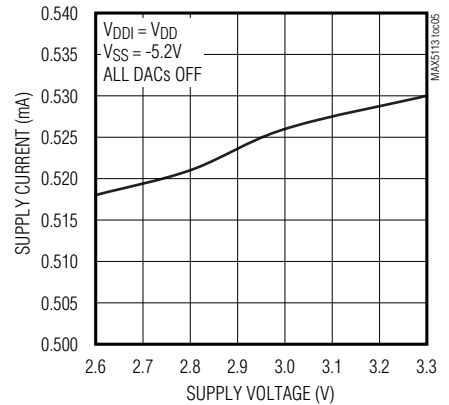
DIGITAL SUPPLY CURRENT vs. DIGITAL INPUT VOLTAGE



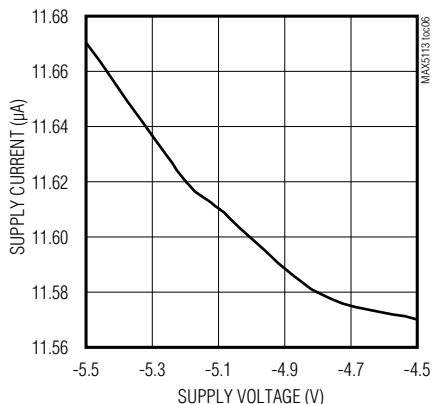
DIGITAL SUPPLY CURRENT vs. DIGITAL SUPPLY VOLTAGE



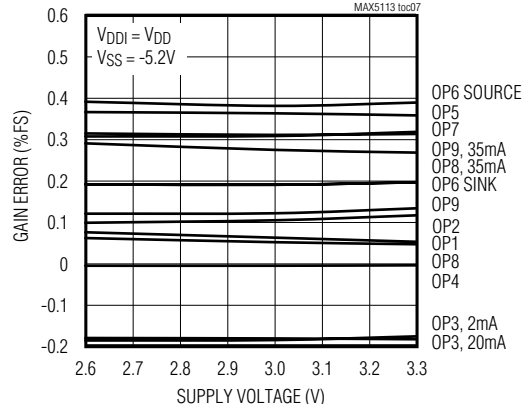
POSITIVE ANALOG SUPPLY CURRENT vs. ANALOG SUPPLY VOLTAGE



NEGATIVE ANALOG SUPPLY CURRENT vs. NEGATIVE SUPPLY VOLTAGE



GAIN ERROR vs. ANALOG SUPPLY VOLTAGE

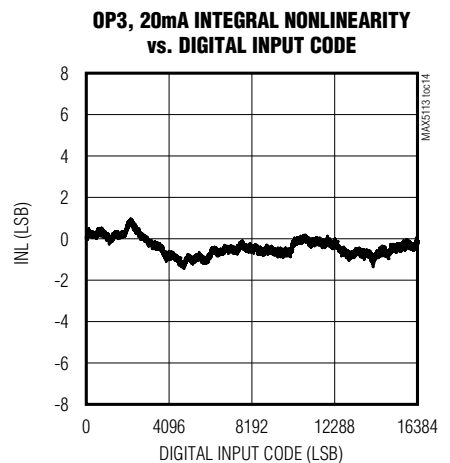
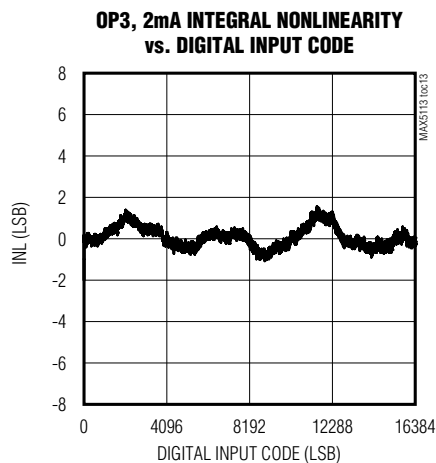
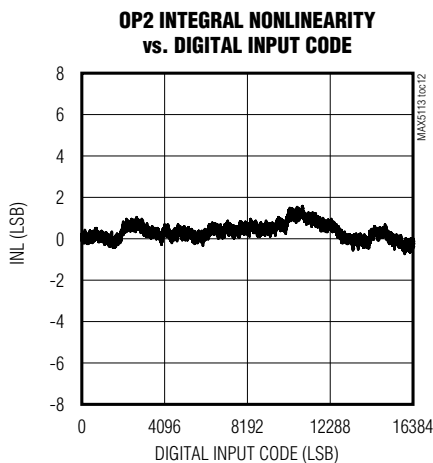
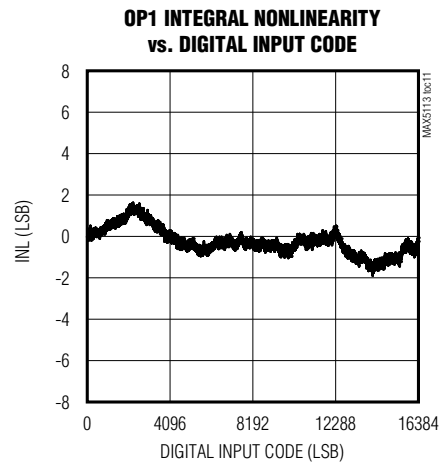
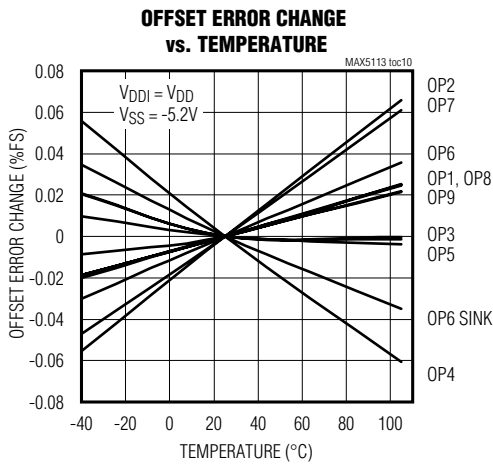
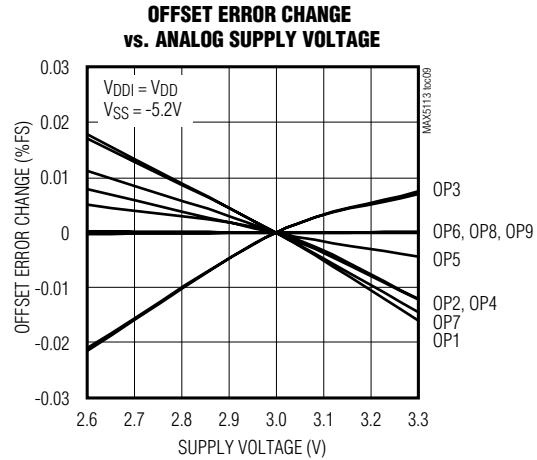
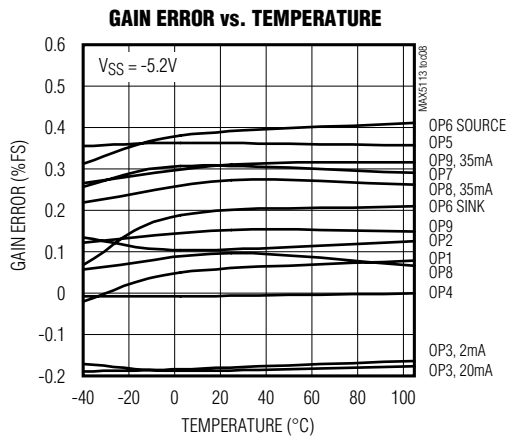


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Typical Operating Characteristics (continued)

($V_{DD} = 3.0V$, $T_A = +25^\circ C$, unless otherwise noted.)



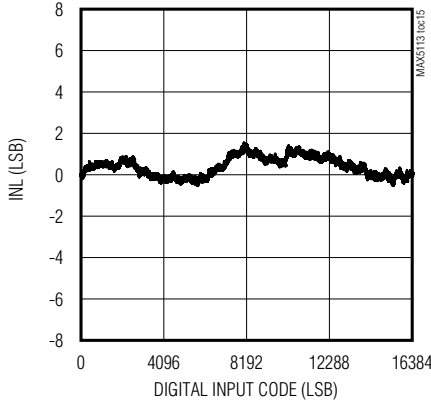
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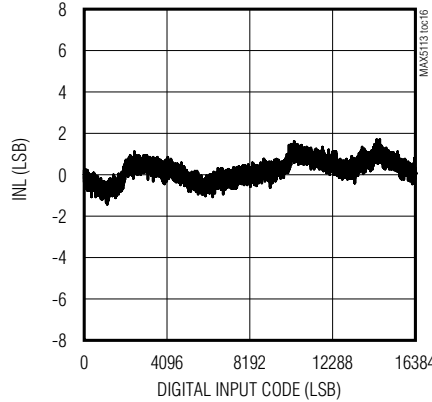
Typical Operating Characteristics (continued)

($V_{DD} = 3.0V$, $T_A = +25^\circ C$, unless otherwise noted.)

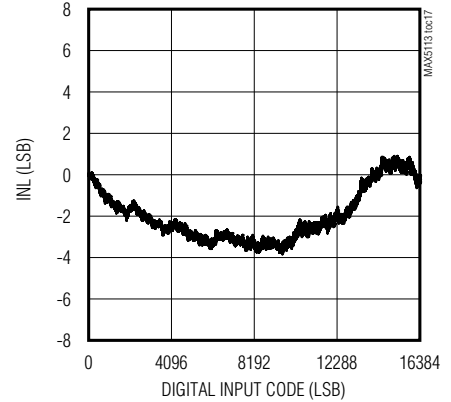
OP4 INTEGRAL NONLINEARITY vs. DIGITAL INPUT CODE



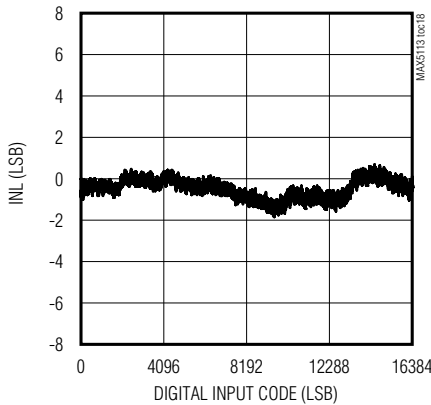
OP5 INTEGRAL NONLINEARITY vs. DIGITAL INPUT CODE



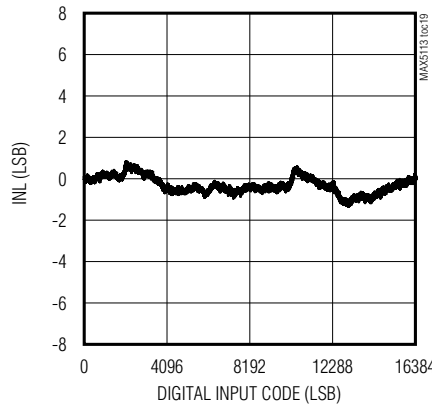
OP6 SINK INTEGRAL NONLINEARITY vs. DIGITAL INPUT CODE



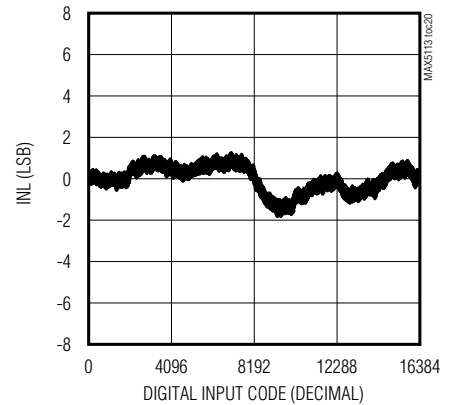
OP6 SOURCE INTEGRAL NONLINEARITY vs. DIGITAL INPUT CODE



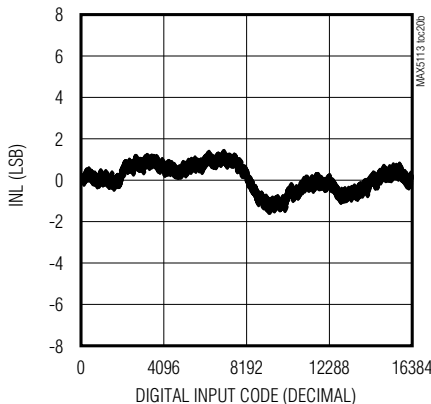
OP7 INTEGRAL NONLINEARITY vs. DIGITAL INPUT CODE



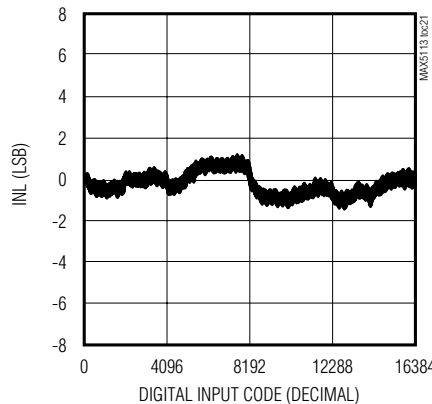
OP8, 15mA INTEGRAL NONLINEARITY vs. DIGITAL INPUT CODE



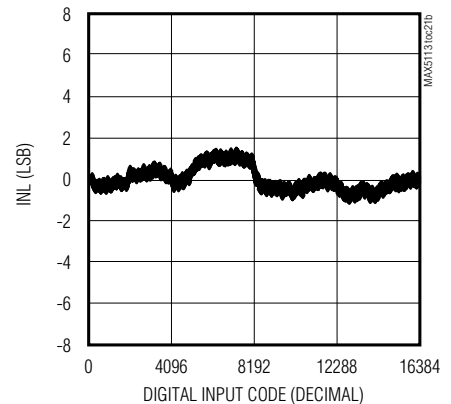
OP8, 35mA INTEGRAL NONLINEARITY vs. DIGITAL INPUT CODE



OP9, 15mA INTEGRAL NONLINEARITY vs. DIGITAL INPUT CODE



OP9, 35mA INTEGRAL NONLINEARITY vs. DIGITAL INPUT CODE

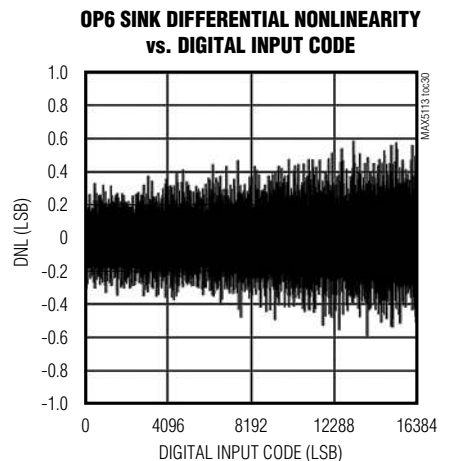
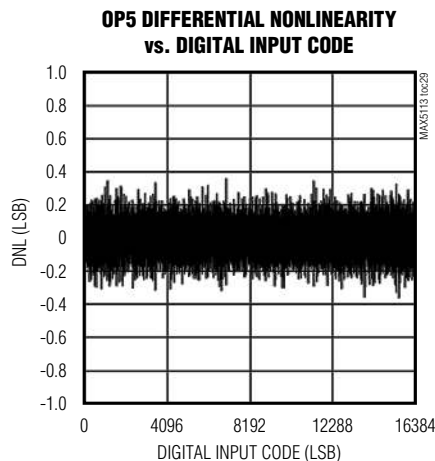
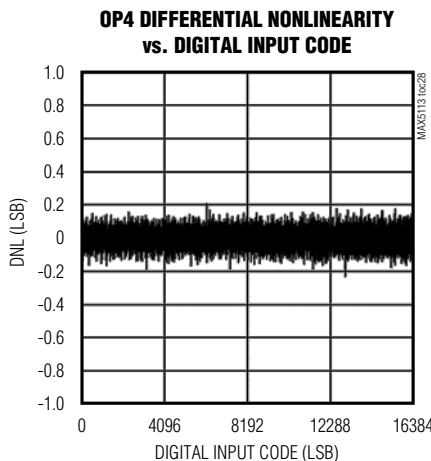
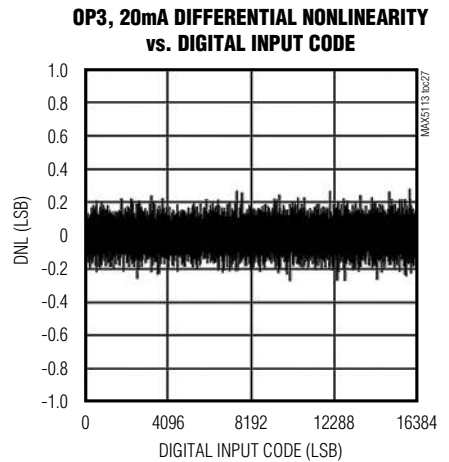
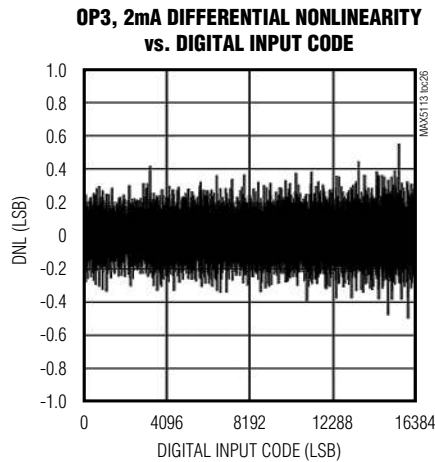
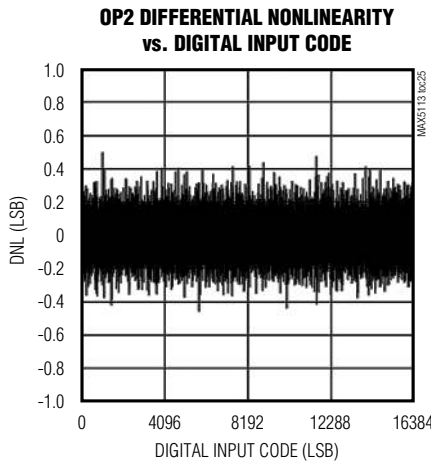
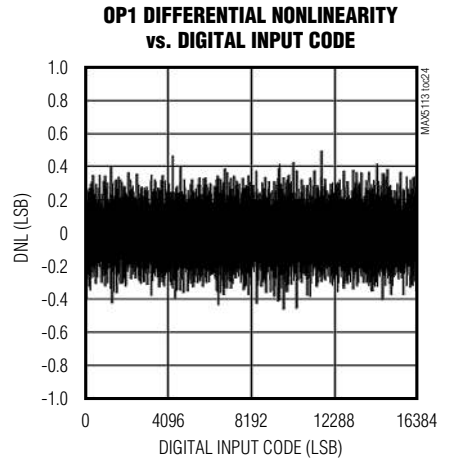
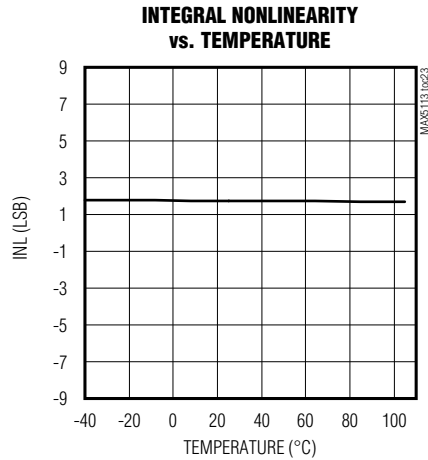
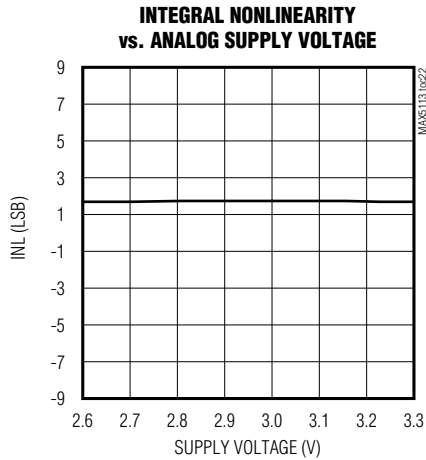


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Typical Operating Characteristics (continued)

($V_{DD} = 3.0V$, $T_A = +25^\circ C$, unless otherwise noted.)



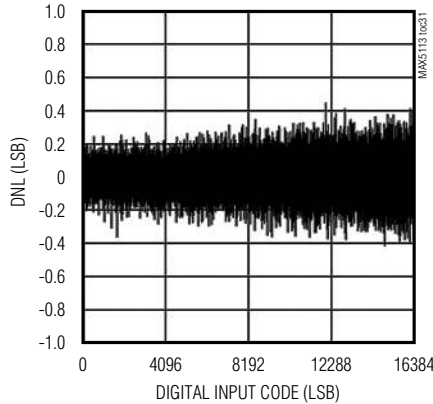
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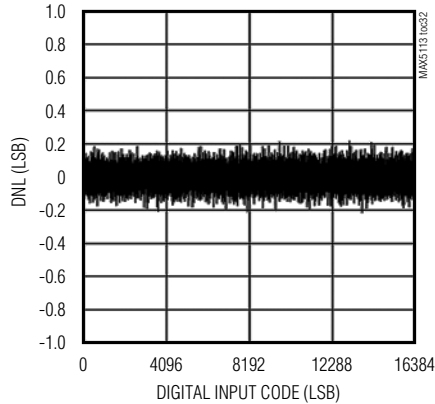
Typical Operating Characteristics (continued)

($V_{DD} = 3.0V$, $T_A = +25^\circ C$, unless otherwise noted.)

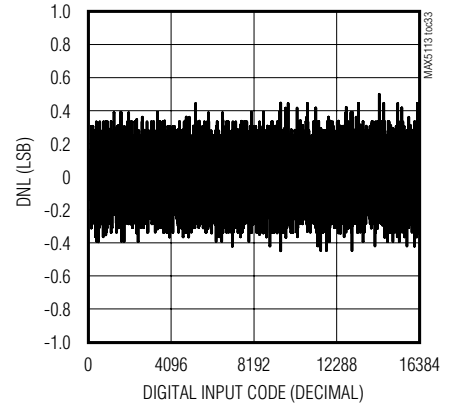
OP6 SOURCE DIFFERENTIAL NONLINEARITY vs. DIGITAL INPUT CODE



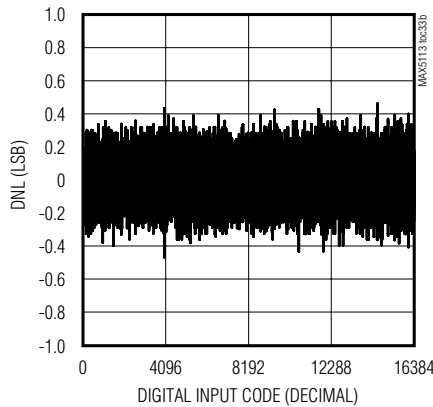
OP7 DIFFERENTIAL NONLINEARITY vs. DIGITAL INPUT CODE



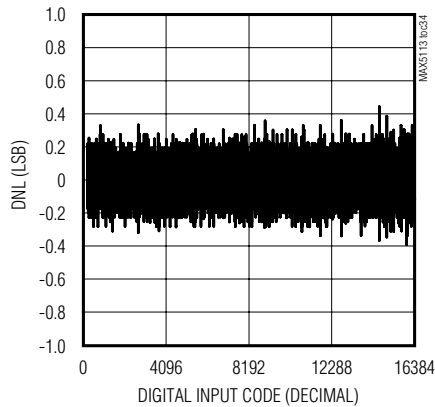
OP8, 15mA DIFFERENTIAL NONLINEARITY vs. DIGITAL INPUT CODE



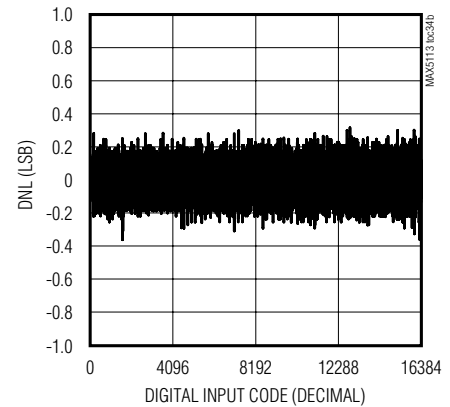
OP8, 35mA DIFFERENTIAL NONLINEARITY vs. DIGITAL INPUT CODE



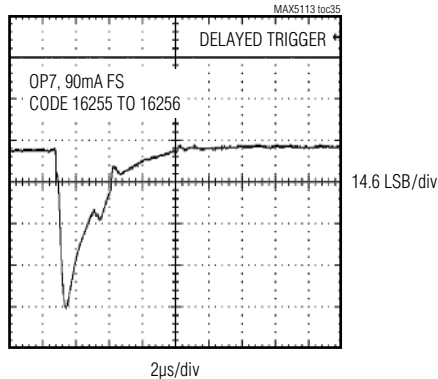
OP8, 15mA DIFFERENTIAL NONLINEARITY vs. DIGITAL INPUT CODE



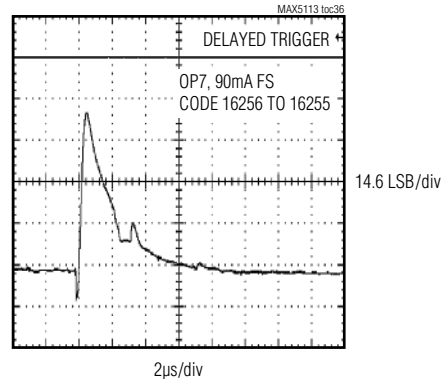
OP9, 35mA DIFFERENTIAL NONLINEARITY vs. DIGITAL INPUT CODE



LOW-TO-HIGH MAJOR CODE TRANSITION OUTPUT TRANSIENT



HIGH-TO-LOW MAJOR CODE TRANSITION OUTPUT TRANSIENT



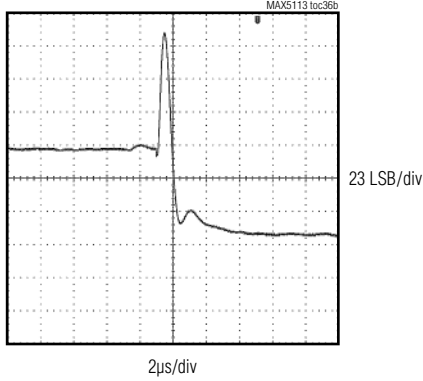
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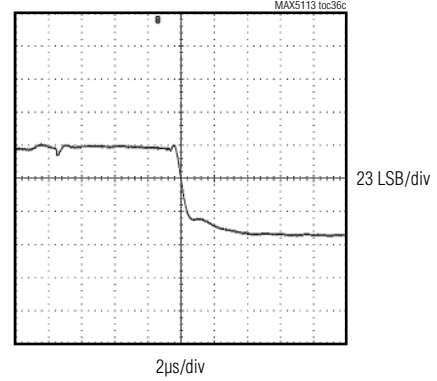
Typical Operating Characteristics (continued)

(VDD = 3.0V, TA = +25°C, unless otherwise noted.)

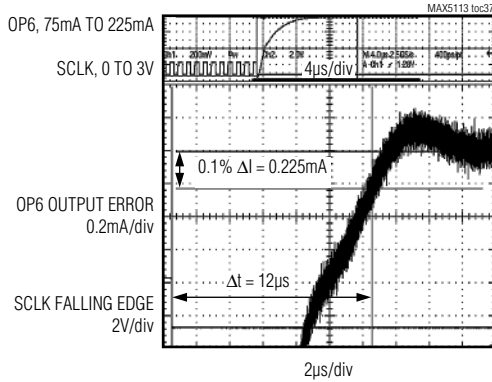
**MULTICODE TRANSITION TRANSIENT
OP8 CODE 9690 TO 9630 (T/H OFF)**



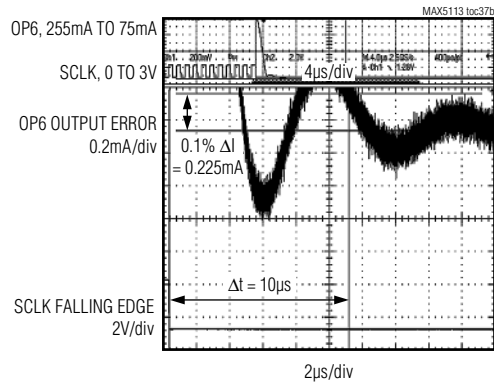
**MULTICODE TRANSITION TRANSIENT
OP8 CODE 9690 TO 9630 (T/H ON)**



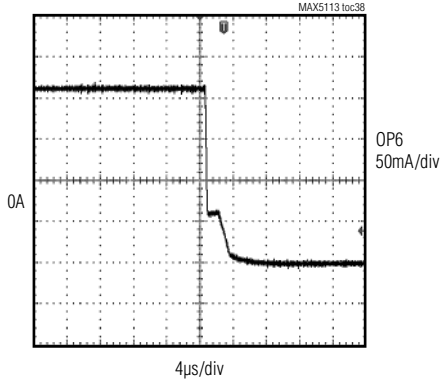
**OUTPUT 6 LOW TO HIGH CODE
0.1% SETTling TIME**



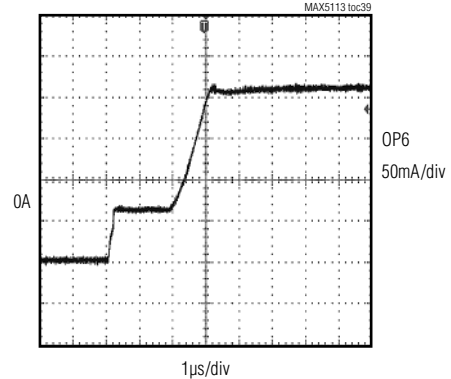
**OUTPUT6 HIGH TO LOW CODE
0.1% SETTling TIME**



SHUTTER MODE SETTling TIME DOWN



SHUTTER MODE SETTling TIME UP

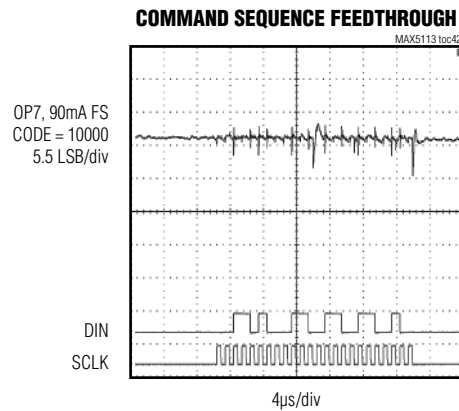
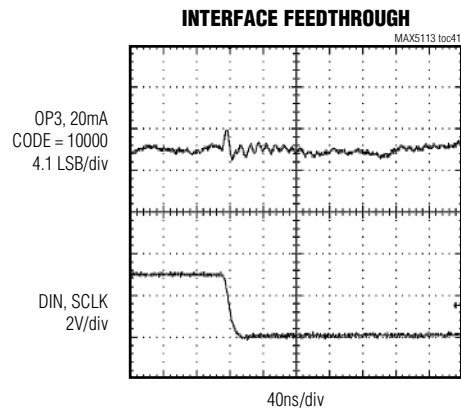
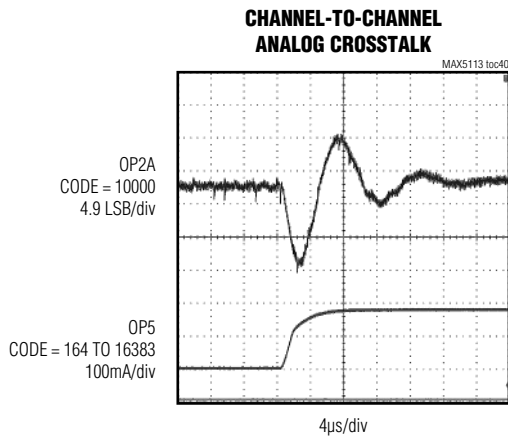


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Typical Operating Characteristics (continued)

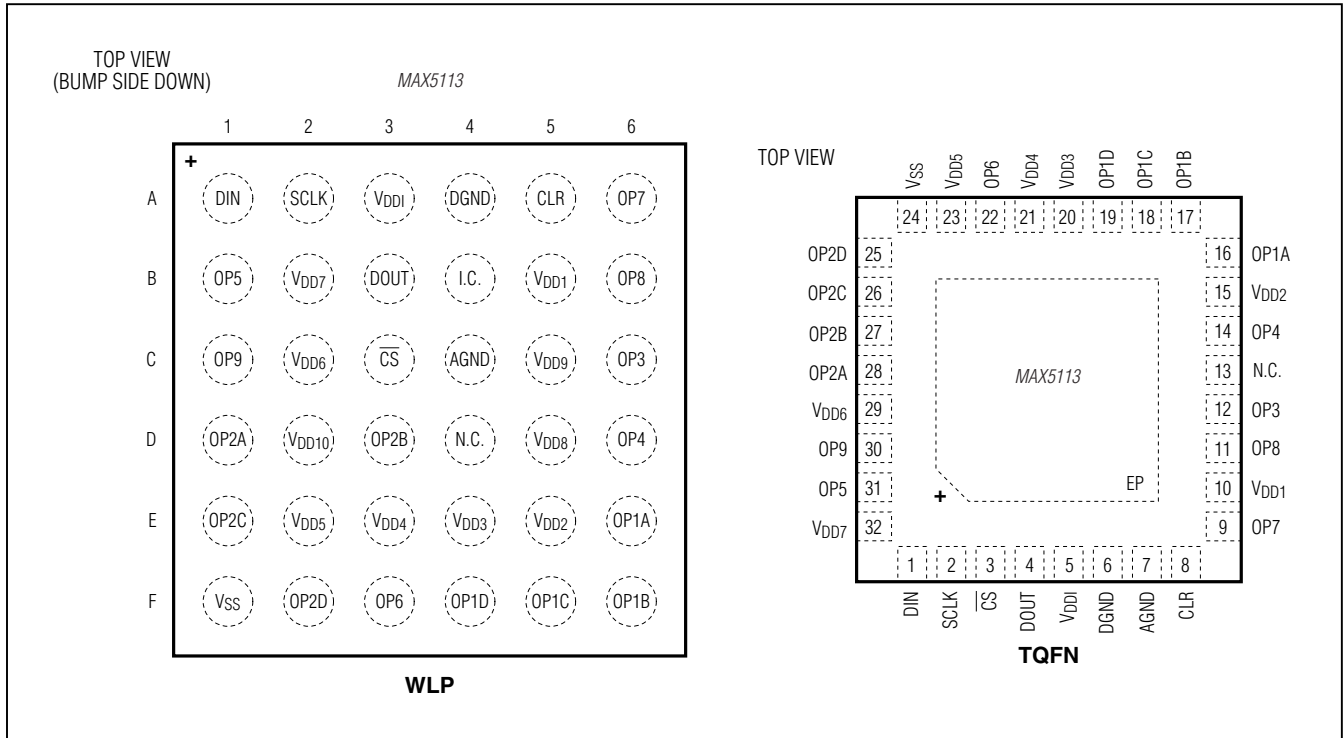
(VDD = 3.0V, TA = +25°C, unless otherwise noted.)



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Pin Configurations



Pin Description

PIN		NAME	FUNCTION
WLP	TQFN-EP		
A1	1	DIN	SPI Data In
B1	31	OP5	DAC 5 Output, 180mA Full Scale
C1	30	OP9	DAC 9 Output, 15mA or 35mA Full Scale
D1	28	OP2A	DAC 2 Multiplexer Output A, 10mA Full Scale
E1	26	OP2C	DAC 2 Multiplexer Output C, 10mA Full Scale
F1	24	VSS	Negative Power Supply
A2	2	SCLK	SPI Clock Input
B2	32	VDD7	DAC 5 Output Positive Power Supply. Internally connected to VDD6 and VDD10.
C2	29	VDD6	DAC 5 Output Positive Power Supply (WLP). Internally connected to VDD7 and VDD10.
			DAC 5 Output and DAC 2 Output Positive Power Supply (TQFN). Internally connected to VDD7 and VDD10.
D2	—	VDD10	DAC 2 Output and DAC 9 Output Positive Power Supply. Internally connected to VDD6 and VDD7.
E2	23	VDD5	DAC 6 Output Positive Power Supply. Internally connected to VDD3 and VDD4.
F2	25	OP2D	DAC 2 Multiplexer Output D, 10mA Full Scale

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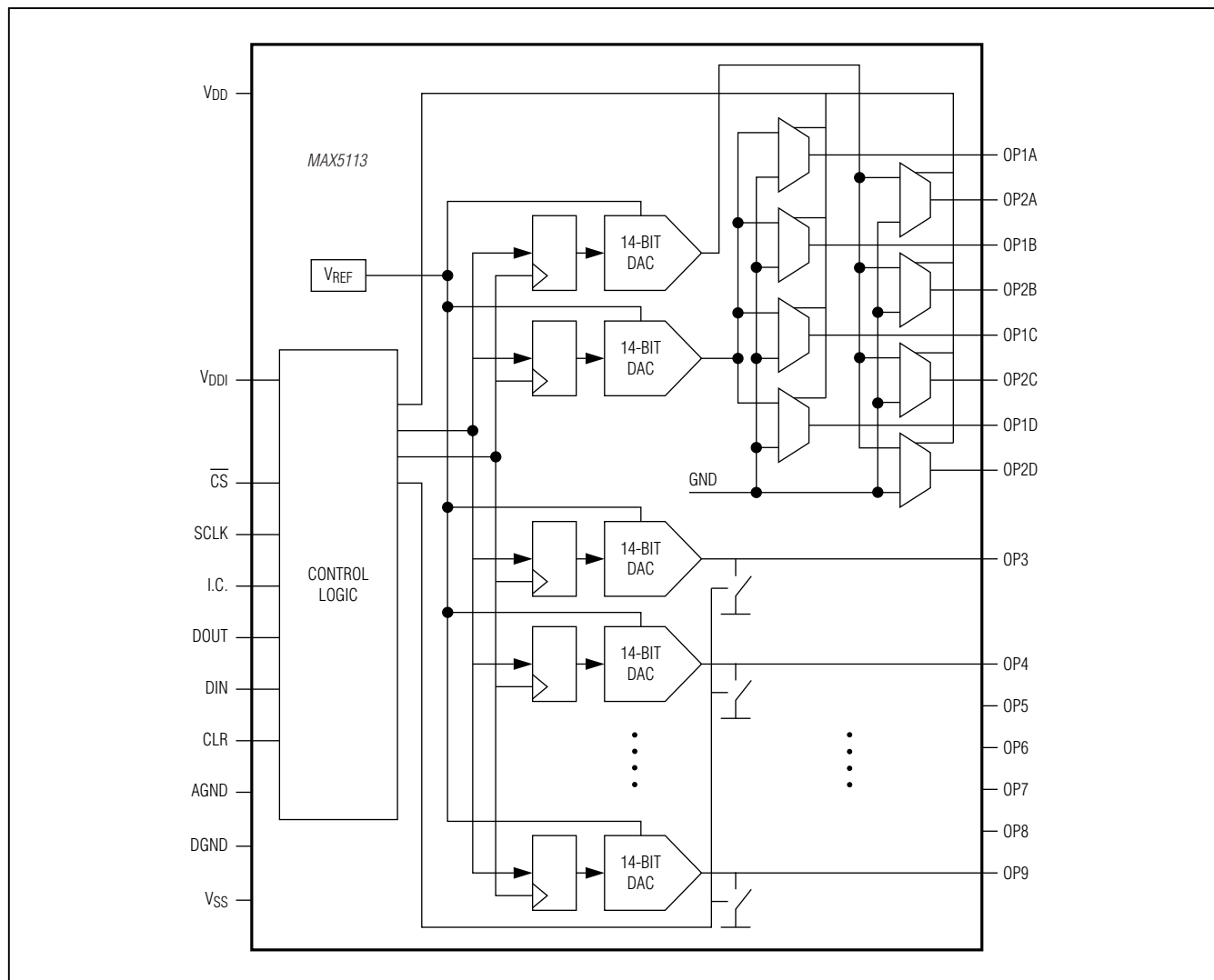
Pin Description (continued)

PIN		NAME	FUNCTION
WLP	TQFN-EP		
A3	5	VDDI	Interface Power Supply. Connect to V _{DD} or to a separate supply to allow for a different interface voltage.
B3	4	DOUT	SPI Data Out
C3	3	\overline{CS}	Active-Low SPI Chip Select
D3	27	OP2B	DAC 2 Multiplexer Output B, 10mA Full Scale
E3	21	VDD4	DAC 6 Output Positive Power Supply. Internally connected to V _{DD5} and V _{DD3} .
F3	22	OP6	DAC 6 Output, -60mA or 300mA Full Scale
A4	6	DGND	Digital Ground
B4	—	I.C.	Internally Connected. Connect to DGND.
C4	7	AGND	Analog Ground
D4	13	N.C.	No Internal Connection. Must obey <i>Absolute Maximum Ratings</i> .
E4	20	VDD3	DAC6 Output Positive Power Supply. Internally connected to V _{DD5} and V _{DD4} .
F4	19	OP1D	DAC 1 Multiplexer Output D, 10mA Full Scale
A5	8	CLR	Active-High Clear
B5	10	VDD1	DAC 7 Output Positive Power Supply (WLP)
			DAC 3 Output and DAC 7 Output and DAC8 Output Positive Power Supply (TQFN)
C5	—	VDD9	DAC 3 Output and DAC 8 Output Positive Power Supply
D5	—	VDD8	DAC 4 Output Positive Power Supply
E5	15	VDD2	DAC 1 Output Positive Power Supply (WLP)
			DAC 1 Output and DAC 4 Output Positive Power Supply (TQFN)
F5	18	OP1C	DAC 1 Multiplexer Output C, 10mA Full Scale
A6	9	OP7	DAC 7 Output, 90mA Full Scale
B6	11	OP8	DAC 8 Output, 15mA or 35mA Full Scale
C6	12	OP3	DAC 3 Output, 2mA or 20mA Full Scale
D6	14	OP4	DAC 4 Output, 90mA Full Scale
E6	16	OP1A	DAC 1 Multiplexer Output A, 10mA Full Scale
F6	17	OP1B	DAC 1 Multiplexer Output B, 10mA Full Scale
—	—	EP	Exposed Pad (TQFN only). Internally connected to AGND. Connect to a ground plane to enhance thermal dissipation.

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Functional Diagram



9-Channel, 14-Bit Current DAC with SPI Interface

Detailed Description

The MAX5113 output ranges are optimized to bias a high-power tunable laser source. See Table 1 for the output current range available on each DAC output.

The DACs and highly stable internal reference are factory trimmed to ensure the outputs are within the specifications. Connect DACs in parallel to increase current drive or resolution.

DAC Outputs

The DAC configuration registers (01h–09h) control the configuration of each DAC. The Individual Configuration Register for each channel must be written to after a power-up event, even if the default values are written. This ensures the device will meet guaranteed offset performance specifications. DACs 1 and 2 drive four 2:1 multiplexers. The multiplexers route each DAC output to one of four outputs. Configure unused outputs as high impedance or connect to AGND. DAC 3 full-scale output is selectable between 2mA and 20mA.

DAC 6 provides 300mA full-scale output when selected as a current source. When selected as a current sink, the full 14 bits are available between 0 and -60mA. A typical application for DAC 6 is to drive an optical amplifier where a current source is varied to set the gain or where a current sink is varied to set the attenuation.

All other DACs are positive current source DACs. DACs 8 and 9 full-scale outputs are selectable between 15mA and 35mA. The output range of DACs 3, 8 and 9 is selected using the RNG bit in the Individual Configuration

Table 1. Typical Full-Scale Output Currents

OUTPUT	OUTPUT-CURRENT RANGE CAPABILITY (mA)	
	LOW RANGE (DEFAULT)	HIGH RANGE
OP1	0 to 10	Reserved
OP2	0 to 10	Reserved
OP3	0 to 2	0 to 20
OP4	0 to 90	Reserved
OP5	0 to 180	Reserved
OP6	-60 to 0 or 0 to 300	Reserved
OP7	0 to 90	Reserved
OP8	0 to 15	0 to 35
OP9	0 to 15	0 to 35

registers. The DAC 6 polarity and full-scale output is set by the SW_POL bit in the DAC 6 register.

Output Track and Hold

All channels feature a track-and-hold circuit to improve glitch performance. In common with all DACs of this type, the MAX5113 DACs will glitch when in transition from one code to another. The size of the glitch is defined by the size of the transition and where in the overall range the transition occurs. In general, a small transition results in a small glitch. However, this is not absolute. The track-and-hold circuit may be enabled to reduce the glitch size to close to zero. The track and hold can be enabled independently for each channel by setting bit 12 in the Individual DAC Configuration registers (01h–09h).

When enabled, the track and hold will engage after the 24th SCLK in the SPI frame, setting a new DAC code. This will hold the output level until the DAC section has settled. There is a small negative offset present in the output level while the track and hold is engaged. Approximately 10 LSB. The track and hold is engaged for 6μs, typical. It then disengages and the channel will transition to its new level with no glitch.

DAC Ground Switch

All DACs include a programmable switch to connect the output to ground when the DAC code is set to zero. The switch is open when the configuration bit is set to 0 and code zero is programmed. In this case, the output drivers are disabled, and the outputs set to high impedance. The DAC switch configuration is set for each individual DAC; see the *01h–09h: Individual DAC (1 to 9) Configuration Registers* section. The global DAC switch-override bits (GSWG[1:0]) in the General Configuration register (00h) override all switch selections when applied.

Clear Function (CLR)

The clear function allows the access of modes of operation through a single active-high input, CLR. The behavior of each DAC with CLR asserted is independently configurable. See the *CLR Interaction* section.

The clear function can also be asserted in software by setting the SW_CLR bit in the Software Reset Command register; see the *0Fh: Software Reset Command Register* section.

The clear function for each DAC is programmed through the CLR_CFG[1:0] bits in the Individual DAC Configuration registers (01h–09h) as shown in the following:

- **00 (Ignore):** The assertion of CLR does not affect the DAC.

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- **01 (Shutter):** Shutter mode applies to OP6 only. For all other DACs, shutter mode produces the same effect as ignore. For OP6, the output polarity stays negative for as long as CLR is asserted (level sensitive). The current sink level is defined by the DAC 6 Shutter Mode Code register (1Bh). Once CLR releases, the DAC output returns to the previously programmed value as set in the DAC 6 Source Mode Code register (16h).
- **10 (Gate):** The DAC is held at code zero (with ground switches engaged if enabled) as long as CLR is asserted (level sensitive). Once CLR releases, the DAC output returns to the previously programmed value as set in the DAC 1–9 Code register (10h–1Ah).
- **11 (Reset):** The DAC is set to code zero (with ground switches engaged if enabled) when CLR is asserted and remains at code zero after CLR is released (edge sensitive).

While the clear operation is in effect, DAC channels configured in ignore, shutter, or gate mode continue to accept new code settings. DAC channels configured in reset mode do not accept code changes until the clear operation is terminated.

Software Clear Interactions

The device provides a software-accessible version of the clear function (SW_CLR), which allows access to the clear functionality directly through the SPI interface (see the *0Fh: Software Reset Command Register* section). When the command 0Fh is used to launch a clear operation, the affected DAC outputs are held in the clear position, determined by the clear configuration settings. This happens from the time when the 0Fh command requesting a clear operation is completed until a second 0Fh command requesting removal of the clear operation is completed. The software- and pin-based clear operations are independently controlled and can be used individually or together without conflict. The devices provide an internal logic-OR circuitry.

Power-On Reset (POR), Power Brownout

The device contains a POR circuit with a threshold of 1.6V (typ) and a hysteresis of 0.025V (typ). POR ensures that the device resets all registers to default conditions as V_{DD} rises through the upper POR threshold. The default condition of all DAC registers is code zero with ground switches engaged, ensuring that no large output current transients damage the load during initial power-up.

In a V_{DD} brownout situation, V_{DD} must fall below the lower POR threshold before a POR is issued when V_{DD} rises again. As V_{DD} falls, the device eventually loses

regulation. However, the device is designed to avoid any large output current transients that could damage the load.

Software Reset and Standby Functions

The device contains a software reset function. The software reset function resets all code and configuration registers to default conditions. Write a 1 to the RST bit in the Software Reset Command register (0Fh) to initiate reset. The RST bit is not persistent, so writing a 0 to reset the bit is not required.

The device includes a software standby function that causes all DAC code registers (10h–1Bh) to be set to code zero. Write a 1 to the STDBY bit in the Software Reset Command register (0Fh) to initiate the standby function. STDBY bit is not persistent, so writing a 0 to reset the bit is not required.

The software standby function is a subset of the software reset function. The software reset function takes effect when both functions are issues.

Overtemperature Error Handling

The device features an on-chip temperature protection circuit to prevent the device from overheating when all DACs output the maximum programmed current. When the die temperature rises above the threshold temperature, +160°C, the PRO_TEMP bit in the Status/Revision Readback Command register (0Eh) is set and the device enters an overtemperature shutdown mode. All DACs are set to code zero, but the control interface remains active, thereby allowing the host processor to read back the device status. The PRO_TEMP bit is latched and, therefore, the device can only be reset by a software reset command, a software standby command, or by cycling the power.

The device features an overtemperature status bit, OVR_TEMP. The OVR_TEMP bit is not latched, and is set if the device temperature is above the protection threshold. The OVR_TEMP bit allows the host processor to determine if the device is too hot to reset. If a software reset is attempted while the device is above the protection threshold, the command is ignored. Similarly, above the threshold die temperature, the device immediately enters shutdown mode when power is cycled.

The device features a warning bit, HI_TEMP. The warning bit is not latched and serves as a high-temperature status indicator bit. The HI_TEMP bit is set when the die temperature is typically 10°C below the overtemperature protection threshold.

See the *Applications Information* section for more detail on calculating die temperature and heat-sinking requirements.

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User Configuration Registers

Table 2 shows a summary of the register map.

Table 2. User Register/Command Summary

REGISTER ADDRESS (hex)	ACCESS	PAIRABLE	REGISTER NAME
00h	W	Y	General Configuration
01h	W	Y	DAC 1 Configuration
02h	W	Y	DAC 2 Configuration
03h	W	Y	DAC 3 Configuration
04h	W	Y	DAC 4 Configuration
05h	W	Y	DAC 5 Configuration
06h	W	Y	DAC 6 Configuration
07h	W	Y	DAC 7 Configuration
08h	W	Y	DAC 8 Configuration
09h	W	Y	DAC 9 Configuration
0Ah	—	—	Reserved
0Bh	—	—	Reserved
0Ch	—	—	Reserved
0Dh	—	—	Reserved
0Eh	R	N	Status Feedback and Part ID
0Fh	W	Y	Software Reset/Standby/Clear
10h	W	Y	DAC 1–9 Code
11h	W	Y	DAC 1 Code
12h	W	Y	DAC 2 Code
13h	W	Y	DAC 3 Code
14h	W	Y	DAC 4 Code
15h	W	Y	DAC 5 Code
16h	W	Y	DAC 6 Source Mode Code
17h	W	Y	DAC 7 Code
18h	W	Y	DAC 8 Code
19h	W	Y	DAC 9 Code
1Ah	W	Y	DAC 6 Sink Mode Code
1Bh	W	Y	DAC 6 Shutter Mode Code
1Ch	—	—	Reserved
1Dh	—	—	Reserved
1Eh	—	—	Reserved
1Fh	W	Y	DAC 6 Polarity Control

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Register Details

00h: General Configuration Register

BIT	15	14	13	12	11	10	9	8
NAME	GSWG[1:0]		BHEN	X	X	X	X	X
DEFAULT	0	0	1	0	0	0	0	0

BIT	7	6	5	4	3	2	1	0
NAME	X	X	X	X	X	X	X	X
DEFAULT	0	0	0	0	0	0	0	0

BIT	NAME	DESCRIPTION
15:14	GSWG[1:0]	Global GSW Configuration Override 00: Individual DAC GSW settings are unaltered 01: Individual DAC GSW settings are set to 0 (ground switches disabled) 10: Individual DAC GSW settings are set to 1 (ground switches enabled) 11: Individual DAC GSW settings are unaltered
13	BHEN	DOUT Bus Hold Enable 0: Bus hold circuit is disabled 1: Bus hold circuit is enabled
12:0	X	Reserved

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01h–09h: Individual DAC (1 to 9) Configuration Registers

BIT	15	14	13	12	11	10	9	8
NAME	GSW	CLR_CFG[1:0]		T/H_EN	RNG	MUX[3:1]		
DEFAULT	1	0	0	0	1	0	1	0

BIT	7	6	5	4	3	2	1	0
NAME	MUX0	X	X	X	X	X	X	X
DEFAULT	0	0	0	0	0	0	0	0

BIT	NAME	DESCRIPTION
15	GSW	Ground Switch Control 0: Output is left open when DAC code = 0000h 1: Output is connected to ground when DAC code = 0000h. For DACs 1 and 2, this setting applies to the active mux output.
14:13	CLR_CFG[1:0]	Clear Configuration Settings (determine how CLR pin affects each DAC) 00 (Ignore): The DAC is not affected by the CLR pin (default) 01 (Shutter): DAC output polarity is held negative (current level determined by 1Bh) as long as the CLR pin is asserted (level sensitive, applies to DAC 6 only; otherwise, implements the ignore function) 10 (Gate): DAC output is held at zero scale (with ground switches engaged if enabled) as long as the CLR pin is asserted (level sensitive) 11 (Reset): DAC output is set to zero scale (with ground switches engaged if enabled) when CLR is asserted and remains valid after CLR is removed (edge sensitive)
12	T/H_EN	Track and Hold Enable 0: Track and Hold disabled 1: Track and Hold enabled
11	RNG	Range (DAC 3, 8, and 9) 0: DAC full-scale output level is set to high range. 1: DAC full-scale output level is set to low range. Note: For all DACs not supporting RNG settings, this bit is reserved and should be set to 1 (default).
10:7	MUX[3:0]	Output Mux Settings for DAC (mux settings are only supported for DAC 1 and DAC 2) 0000: Output A active, all others high impedance 0001: Output B active, all others high impedance 0010: Output C active, all others high impedance 0011: Output D active, all others high impedance 01XX: All outputs high impedance (open) 1000: Output A active, all others shunted to GND 1001: Output B active, all others shunted to GND 1010: Output C active, all others shunted to GND 1011: Output D active, all others shunted to GND 11XX: All outputs shunted to GND (default)
6:0	X	Reserved

Note: Any change to individual DAC configuration settings resets the affected DAC code to 0000h.

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0Eh: Status/Revision Readback Command Register

BIT	15	14	13	12	11	10	9	8
NAME	PRO_TEMP	OVR_TEMP	HI_TEMP	X	PART_ID[3:0]			
DEFAULT	0	0	0	0	0	0	0	1

BIT	7	6	5	4	3	2	1	0
NAME	REV_ID[3:0]				X	X	X	X
DEFAULT	0	1	0	0	0	0	0	0

BIT	NAME	DESCRIPTION
15	PRO_TEMP	Overtemperature Protection Indicator 0: Normal operation 1: Device overtemperature protection engaged
14	OVR_TEMP	Overtemperature Warning Indicator 0: Normal operation 1: Device temperature is too high (exceeding protection limit)
13	HI_TEMP	High-Temperature Warning Indicator 0: Normal operation 1: Device temperature is high (nearing protection limit)
12	X	Reserved
11:8	PART_ID[3:0]	Part ID Code (0001)
7:4	REV_ID[3:0]	Revision Code (0100)
3:0	X	Reserved

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0Fh: Software Reset Command Register

BIT	15	14	13	12	11	10	9	8
NAME	RST	STDBY	SW_CLR	X	X	X	X	X
DEFAULT	0	0	0	0	0	0	0	0

BIT	7	6	5	4	3	2	1	0
NAME	X	X	X	X	X	X	X	X
DEFAULT	0	0	0	0	0	0	0	0

BIT	NAME	DESCRIPTION
15	RST	Global Reset (identical to a POR) 0: No operation 1: Reset: All DAC modes, configurations, and codes are returned to their default settings <i>Not Persistent: The reset operation is contained within the command. It is not necessary to issue a second 0Fh command to remove the reset condition.</i>
14	STDBY	Global Standby (identical to a global power-down) 0: No operation 1: Standby: All DAC codes are set to zero, but retain all configuration information <i>Not Persistent: The standby operation is contained within the command. It is not necessary to issue a second 0Fh command to remove the standby condition.</i> <i>Exclusive: If RST and STDBY are requested, STDBY is not issued.</i>
13	SW_CLR	Software Clear 0: No operation/remove SW_CLR 1: Assert SW_CLR <i>Persistent: The status of SW_CLR remains in effect until changed by a later 0Fh command.</i> <i>Exclusive: If SW_CLR and RST and/or STDBY are requested, SW_CLR is not issued.</i>
12:0	X	Reserved

Note: A software reset or standby command is required to exit overtemperature-protection mode once engaged (software clear does not qualify for an exit).

10h: Group DAC (1 to 9) Code Command Register

BIT	15	14	13	12	11	10	9	8
NAME	B13	B12	B11	B10	B9	B8	B7	B6
DEFAULT	0	0	0	0	0	0	0	0

BIT	7	6	5	4	3	2	1	0
NAME	B5	B4	B3	B2	B1	B0	X	X
DEFAULT	0	0	0	0	0	0	0	0

BIT	NAME	DESCRIPTION
15:2	B[13:0]	Group DAC Code Setting in Straight Binary Format. All DACs outputs update to code B[13:0] upon command completion. <i>This command is primarily useful for speeding up testing and qualification.</i> <i>Deglitching circuitry is not activated by group DAC code operations.</i>
1:0	X	Reserved

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11h–1Bh: Individual DAC (1 to 9) Code Setting Registers

BIT	15	14	13	12	11	10	9	8
NAME	B13	B12	B11	B10	B9	B8	B7	B6
DEFAULT	0	0	0	0	0	0	0	0

BIT	7	6	5	4	3	2	1	0
NAME	B5	B4	B3	B2	B1	B0	X	X
DEFAULT	0	0	0	0	0	0	0	0

BIT	NAME	DESCRIPTION
15:2	B[13:0]	DAC Code Settings in Straight Binary Format 3FFFh = Full-scale output 0000h = Zero-scale output (GSW configuration settings apply)
1:0	X	Reserved

Note: 11h–19h are DAC code settings for DACs 1–9, respectively. 1Ah is the sink mode setting for DAC 6. 1Bh is the shutter mode setting for DAC 6. See Table 2.

1Fh: DAC 6 Polarity Command Register

BIT	15	14	13	12	11	10	9	8
NAME	SW_POL	X	X	X	X	X	X	X
DEFAULT	0	0	0	0	0	0	0	0

BIT	7	6	5	4	3	2	1	0
NAME	X	X	X	X	X	X	X	X
DEFAULT	0	0	0	0	0	0	0	0

BIT	NAME	DESCRIPTION
15	SW_POL	Software Polarity Control (to DAC 6 only) 0: Source-mode operation (0 to 300mA determined by 16h code, with GSW operation) 1: Sink-mode operation (0 to -60mA determined by 1Ah code, GSW operation disabled)
14:0	X	Reserved

DAC 6 Polarity Operations

Software command 1Fh (SW_POL) or the CLR operation in shutter (01) mode controls the polarity of DAC 6. DAC 6 operates in a sink-current mode (0 to -60mA, determined by register 1Ah in sink mode) when SW_POL is set high. When the software command is used, the requested polarity is held in effect from the time when the 1Fh command requesting a polarity change is completed until a second 1Fh command requesting a polarity change operation is completed. When the shutter mode is used, DAC 6 remains in shutter mode as long as CLR is held high. The software- and CLR-driven polarity operations

are independently controlled and can be used individually or together without conflict. The device provides an internal logic-OR operation. Shutter allows the fast access to a programmable negative code, based on register 1Bh, from either a source or sink mode with a controlled return to the original operating state upon release.

Gate mode is activated by asserting the CLR or through the SW_CLR bit. If gate mode is activated while the DAC is set to sink mode, the DAC remains in sink mode, but the current is reduced to 0mA for the duration of the gating event.

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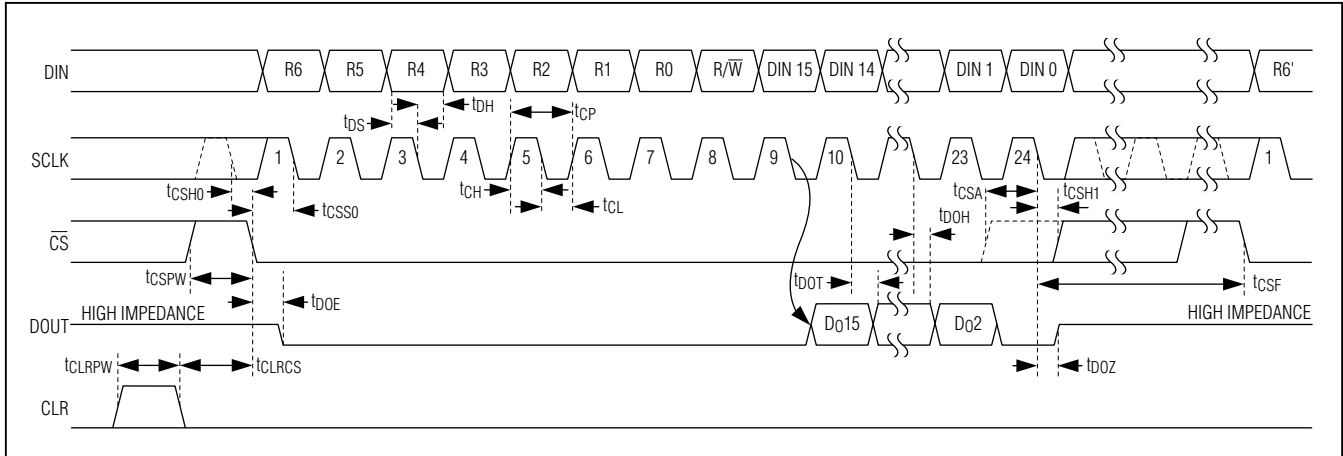


Figure 1. Minimum SPI Programming Operation

Table 3. SPI User Commands and Data Mapping with Clock Falling Edges (24-Bit Frame)

EDGE	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24
DIN	R6	R5	R4	R3	R2	R1	R0	R/W	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
DOUT	0	0	0	0	0	0	0	0	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	0	High-Z

Similarly, when reset mode is set, the DAC remains in sink mode, but the current is reduced to 0mA and remains there during and after the reset event. All source-, sink-, and shutter-mode current settings are reset to zero by this operation. Shutter mode is inaccessible while DAC 6 is configured for reset.

Regardless of polarity/shutter setting, DAC 6 continues to accept updated code settings for either source (16h), sink (1Ah), or shutter mode (1Bh) code registers, provided the DAC is not being held in any reset mode (through CLR or SW_CLR).

SPI Interface

The device features an SPI interface capable of clock speeds up to 25MHz. Figure 1 shows programming operation with 24 SCLK periods, which is the minimum for a valid frame.

For free-running SCLK applications, tCSH0 is shown with respect to the SCLK falling edge preceding the first SCLK falling edge (optional cycles shown in gray). To abort a command sequence, the rise of CS must precede the 24th falling edge of SCLK by tCSA.

Table 3 shows how the contents of the user-command data are mapped into the command registry. The device requires a 1-byte command, also referred to as a register address, R[6:0] paired with a R/W bit, followed by a 2-byte data word, D[15:0]. Set R/W to 0 for a write. Set R/W to 1 for a read. A full 24-bit SPI command sequence is required for all SPI command operations, regardless of the number of data bits actually used for the command. Any commands terminating with less than a full 24-bit sequence are aborted without affecting the operation of the device, subject to tCSA timing requirements. When a command sequence with more than 24 bits is provided, the command is executed on the 24th SCLK falling edge and the remainder of the command is ignored. In addition, the SPI interface elements are disabled to reduce transient current consumption.

All SPI commands cause the device to assume control of the DOUT line from the first SCLK edge through the 24th SCLK edge. Write-mode commands read out all zeros. After relinquishing the DOUT line, the device returns to a high-impedance mode. An optional bus hold circuit can be engaged to prevent leaving the DOUT line unbiased while not interfering with other devices on the bus. This is bit 13 in the General Configuration register (00h).

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CLR Interaction

The device's clear function allows the access of operation modes through a single input, CLR. Each DAC mode can be configured independently. The CLR input interacts with DAC code settings only. The CLR input does not interfere with configurations or readback operations. On-going SPI transfers continue uninterrupted when CLR is driven high. The effect of CLR being driven depends on the clear configurations of the individual DAC channel.

Code changes to any DAC channels configured in ignore (00), shutter (01), or gate (10) mode are recognized, regardless of the status of CLR. In shutter or gate mode, the DACs remain in the shutter or off positions for the duration of the CLR assertion. Once CLR is released, the DACs return to the most recently programmed output values.

Any DAC channels configured in reset (11) mode ignore code changes contained in the SPI commands during when CLR is or has been asserted. These channels do not recognize code-change commands until a subsequent \overline{CS} high condition is recognized, if the removal of the clear condition is observed at least t_{CLRCS} prior to the falling edge of \overline{CS} . In reset mode, the DAC code memories are reset to a zero code state and remain in that state until programmed by a subsequent command.

Applications Information

Thermal Design

To reduce thermal resistance, include V_{DD} and ground planes in the application PCB. Connect the TQFN exposed pad to the ground plane through a large via. Connect the multiple V_{DD} inputs to the V_{DD} plane through multiple vias and bypass each V_{DD} pin with a separate 0.1 μ F capacitor as close as possible to the supply pin. Connect AGND and DGND to the ground plane.

Noise Immunity

Each V_{DD} pin should be bypassed with a separate 0.1 μ F capacitor as close as possible to the supply pin. Pay particular attention to the ESR value of the capacitors and add a 100pF capacitor in parallel to each 100nF capacitor. Noise is particularly important in fiber applications; thus, it may be necessary to add 100pF capacitors to decouple the optional electrodes to ground. This ensures that any crosstalk between the interface and the DAC outputs caused by PCB parasitic is minimized.

Chip Information

PROCESS: BiCMOS

Package Information

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
36 WLP	W363A3+1	21-0024	Refer to Application Note 1891
32 TQFN-EP	T3255+4	21-0140	90-0012

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Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	7/12	Initial release	—
1	9/12	Released the WLP package and revised the <i>Electrical Characteristics</i> , Table 1, and the <i>01h–09h: Individual DAC (1 to 9) Configuration Registers</i> table.	1, 2, 17, 21
2	5/13	Updated Note 3 in the <i>Electrical Characteristics</i> and revised the <i>DAC Outputs</i> section.	6, 17



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