A True System-on-Chip Solution for 2.4-GHz IEEE 802.15.4 and ZigBee Applications
Check for Samples: CC2530F32, CC2530F64, CC2530F128, CC2530F256

FEATURES

- **RF/Layout**
  - **2.4-GHz IEEE 802.15.4 Compliant RF Transceiver**
  - **Excellent Receiver Sensitivity and Robustness to Interference**
  - **Programmable Output Power Up to 4.5 dBm**
  - **Very Few External Components**
  - **Only a Single Crystal Needed for Asynchronous Networks**
  - **6-mm × 6-mm QFN40 Package**
  - **Suitable for Systems Targeting Compliance With Worldwide Radio-Frequency Regulations: ETSI EN 300 328 and EN 300 440 (Europe), FCC CFR47 Part 15 (US) and ARIB STD-T-66 (Japan)**
- **Low Power**
  - **Active-Mode RX (CPU Idle): 24 mA**
  - **Active Mode TX at 1 dBm (CPU Idle): 29 mA**
  - **Power Mode 1 (4 μs Wake-Up): 0.2 mA**
  - **Power Mode 2 (Sleep Timer Running): 1 μA**
  - **Power Mode 3 (External Interrupts): 0.4 μA**
  - **Wide Supply-Voltage Range (2 V–3.6 V)**
- **Microcontroller**
  - **High-Performance and Low-Power 8051 Microcontroller Core With Code Prefetch**
  - **32-, 64-, 128-, or 256-KB In-System-Programmable Flash**
  - **8-KB RAM With Retention in All Power Modes**
  - **Hardware Debug Support**
- **Peripherals**
  - **Powerful Five-Channel DMA**
  - **Integrated High-Performance Op-Amp and Ultralow-Power Comparator**

- **IEEE 802.15.4 MAC Timer, General-Purpose Timers (One 16-Bit, Two 8-Bit)**
  - **IR Generation Circuitry**
  - **32-kHz Sleep Timer With Capture**
  - **CSMA/CA Hardware Support**
  - **Accurate Digital RSSI/LQI Support**
  - **Battery Monitor and Temperature Sensor**
  - **12-Bit ADC With Eight Channels and Configurable Resolution**
  - **AES Security Coprocessor**
  - **Two Powerful USARTs With Support for Several Serial Protocols**
  - **21 General-Purpose I/O Pins (19 × 4 mA, 2 × 20 mA)**
  - **Watchdog Timer**
- **Development Tools**
  - **CC2530 Development Kit**
  - **CC2530 ZigBee® Development Kit**
  - **CC2530 RemoTI™ Development Kit for RF4CE**
  - **SmartRF™ Software**
  - **Packet Sniffer**
  - **IAR Embedded Workbench™ Available**

APPLICATIONS
- **2.4-GHz IEEE 802.15.4 Systems**
- **RF4CE Remote Control Systems (64-KB Flash and Higher)**
- **ZigBee Systems (256-KB Flash)**
- **Home/Building Automation**
- **Lighting Systems**
- **Industrial Control and Monitoring**
- **Low-Power Wireless Sensor Networks**
- **Consumer Electronics**
- **Health Care**

DESCRIPTION

The CC2530 is a true system-on-chip (SoC) solution for IEEE 802.15.4, Zigbee and RF4CE applications. It enables robust network nodes to be built with very low total bill-of-material costs. The CC2530 combines the excellent performance of a leading RF transceiver with an industry-standard enhanced 8051 MCU, in-system programmable flash memory, 8-KB RAM, and many other powerful features. The CC2530 comes in four different flash versions: CC2530F32/64/128/256, with 32/64/128/256 KB of flash memory, respectively. The CC2530 has various operating modes, making it highly suited for systems where ultralow power consumption is required. Short transition times between operating modes further ensure low energy consumption.

Combined with the industry-leading and golden-unit-status ZigBee protocol stack (Z-Stack™) from Texas Instruments, the CC2530F256 provides a robust and complete ZigBee solution.

Combined with the golden-unit-status RemoTI stack from Texas Instruments, the CC2530F64 and higher provide a robust and complete ZigBee RF4CE remote-control solution.
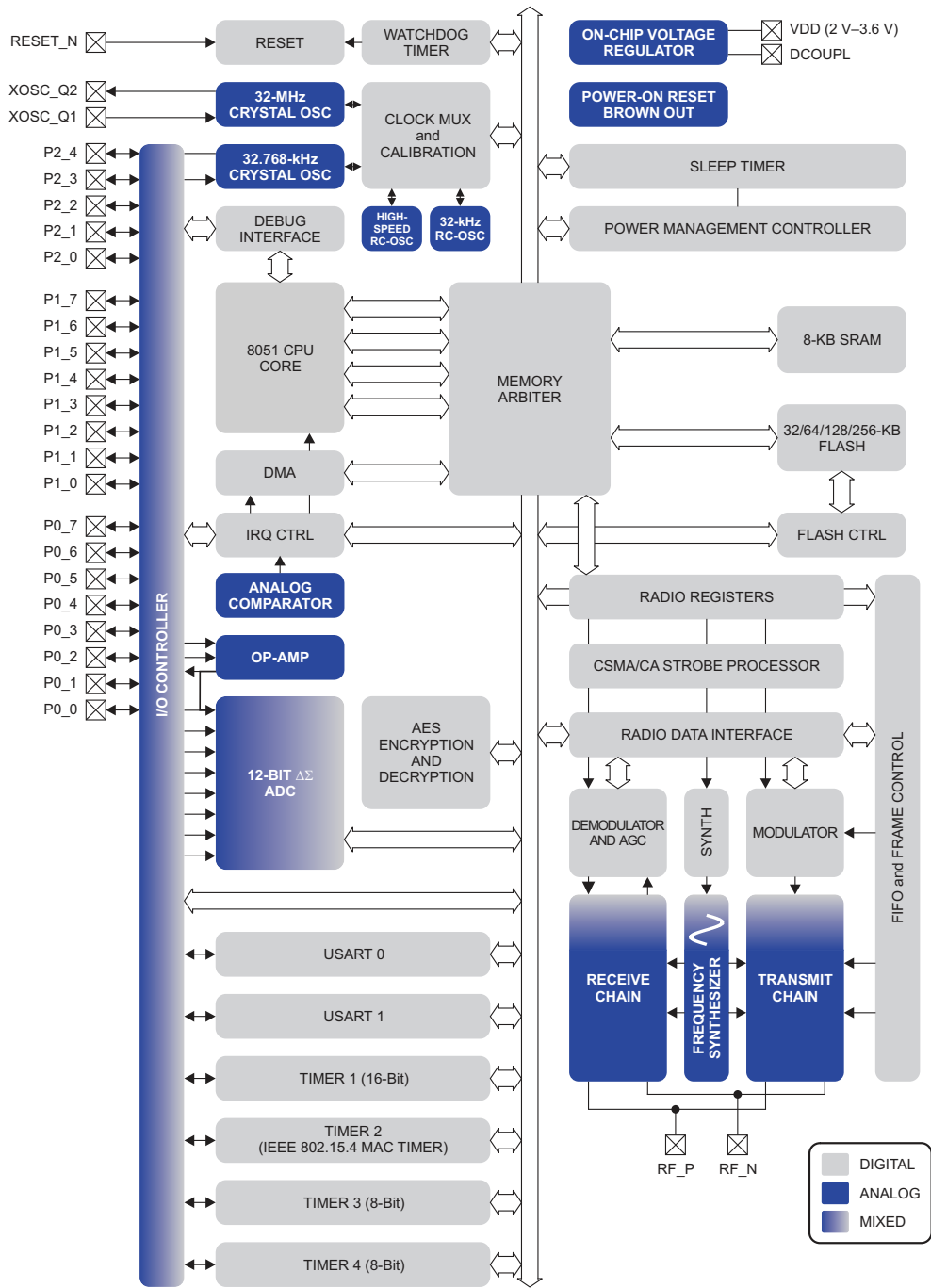
This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

Product Folder Link(s): *CC2530F32 CC2530F64 CC2530F128 CC2530F256*

ABSOLUTE MAXIMUM RATINGS[1]

		MIN	MAX	UNIT
Supply voltage	All supply pins must have the same voltage	–0.3	3.9	V
Voltage on any digital pin		–0.3	VDD + 0.3, ≤ 3.9	V
Input RF level			10	dBm
Storage temperature range		–40	125	°C
ESD[2]	All pads, according to human-body model, JEDEC STD 22, method A114		2	kV
	According to charged-device model, JEDEC STD 22, method C101		500	V

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) CAUTION: ESD sensitive device. Precaution should be used when handling the device in order to prevent permanent damage.

RECOMMENDED OPERATING CONDITIONS

	MIN	MAX	UNIT
Operating ambient temperature range, T_A	–40	125	°C
Operating supply voltage	2	3.6	V

ELECTRICAL CHARACTERISTICS

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C and VDD = 3 V, unless otherwise noted. **Boldface** limits apply over the entire operating range, T_A = –40°C to 125°C, VDD = 2 V to 3.6 V, and f_c = 2394 MHz to 2507 MHz.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_{core}	Core current consumption	Digital regulator on. 16-MHz RCOSC running. No radio, crystals, or peripherals active. Medium CPU activity: normal flash access[1], no RAM access		3.4		mA
		32-MHz XOSC running. No radio or peripherals active. Medium CPU activity: normal flash access[1], no RAM access		6.5	**8.9**	mA
		32-MHz XOSC running, radio in RX mode, –50-dBm input power, no peripherals active, CPU idle		20.5		mA
		32-MHz XOSC running, radio in RX mode at -100-dBm input power (waiting for signal), no peripherals active, CPU idle		24.3	**29.6**	mA
		32-MHz XOSC running, radio in TX mode, 1-dBm output power, no peripherals active, CPU idle		28.7		mA
		32-MHz XOSC running, radio in TX mode, 4.5-dBm output power, no peripherals active, CPU idle		33.5	**39.6**	mA
		Power mode 1. Digital regulator on; 16-MHz RCOSC and 32-MHz crystal oscillator off; 32.768-kHz XOSC, POR, BOD and sleep timer active; RAM and register retention		0.2	0.3	mA
		Power mode 2. Digital regulator off; 16-MHz RCOSC and 32-MHz crystal oscillator off; 32.768-kHz XOSC, POR, and sleep timer active; RAM and register retention		1	2	µA
		Power mode 3. Digital regulator off; no clocks; POR active; RAM and register retention		0.4	1	µA
I_{peri}	**Peripheral Current Consumption** (Adds to core current I_{core} for each peripheral unit activated)					
	Timer 1	Timer running, 32-MHz XOSC used		90		µA
	Timer 2	Timer running, 32-MHz XOSC used		90		µA
	Timer 3	Timer running, 32-MHz XOSC used		60		µA
	Timer 4	Timer running, 32-MHz XOSC used		70		µA
	Sleep timer	Including 32.753-kHz RCOSC		0.6		µA
	ADC	When converting		1.2		mA
	Flash	Erase		1		mA
		Burst write peak current		6		mA

(1) Normal flash access means that the code used exceeds the cache storage, so cache misses happen frequently.

GENERAL CHARACTERISTICS

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C and VDD = 3 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
WAKE-UP AND TIMING					

GENERAL CHARACTERISTICS (continued)

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C and VDD = 3 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Power mode 1 → active	Digital regulator on, 16-MHz RCOSC and 32-MHz crystal oscillator off. Start-up of 16-MHz RCOSC		4		µs
Power mode 2 or 3 → active	Digital regulator off, 16-MHz RCOSC and 32-MHz crystal oscillator off. Start-up of regulator and 16-MHz RCOSC		0.1		ms
Active → TX or RX	Initially running on 16-MHz RCOSC, with 32-MHz XOSC OFF		0.5		ms
	With 32-MHz XOSC initially on			192	µs
RX/TX and TX/RX turnaround				192	µs
RADIO PART					
RF frequency range	Programmable in 1-MHz steps, 5 MHz between channels for compliance with [1]	2394		2507	MHz
Radio baud rate	As defined by [1]		250		kbps
Radio chip rate	As defined by [1]		2		MChip/s
Flash erase cycles				20	k cycles
Flash page size			2		KB

RF RECEIVE SECTION

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C, VDD = 3 V, and f_c = 2440 MHz, unless otherwise noted.

Boldface limits apply over the entire operating range, T_A = –40°C to 125°C, VDD = 2 V to 3.6 V, and f_c = 2394 MHz to 2507 MHz.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Receiver sensitivity	PER = 1%, as specified by [1] [1] requires –85 dBm		–97	–92 **–88**	dBm
Saturation (maximum input level)	PER = 1%, as specified by [1] [1] requires –20 dBm		10		dBm
Adjacent-channel rejection, 5-MHz channel spacing	Wanted signal –82 dBm, adjacent modulated channel at 5 MHz, PER = 1 %, as specified by [1]. [1] requires 0 dB		49		dB
Adjacent-channel rejection, –5-MHz channel spacing	Wanted signal –82 dBm, adjacent modulated channel at –5 MHz, PER = 1 %, as specified by [1]. [1] requires 0 dB		49		dB
Alternate-channel rejection, 10-MHz channel spacing	Wanted signal –82 dBm, adjacent modulated channel at 10 MHz, PER = 1%, as specified by [1] [1] requires 30 dB		57		dB
Alternate-channel rejection, –10-MHz channel spacing	Wanted signal –82 dBm, adjacent modulated channel at –10 MHz, PER = 1 %, as specified by [1] [1] requires 30 dB		57		dB
Channel rejection ≥ 20 MHz ≤ –20 MHz	Wanted signal at –82 dBm. Undesired signal is an IEEE 802.15.4 modulated channel, stepped through all channels from 2405 to 2480 MHz. Signal level for PER = 1%.		57 57		dB
Co-channel rejection	Wanted signal at –82 dBm. Undesired signal is 802.15.4 modulated at the same frequency as the desired signal. Signal level for PER = 1%.		–3		dB
Blocking/desensitization 5 MHz from band edge 10 MHz from band edge 20 MHz from band edge 50 MHz from band edge –5 MHz from band edge –10 MHz from band edge –20 MHz from band edge –50 MHz from band edge	Wanted signal 3 dB above the sensitivity level, CW jammer, PER = 1%. Measured according to EN 300 440 class 2.		–33 –33 –32 –31 –35 –35 –34 –34		dBm
Spurious emission. Only largest spurious emission stated within each band. 30 MHz–1000 MHz 1 GHz–12.75 GHz	Conducted measurement with a 50-Ω single-ended load. Suitable for systems targeting compliance with EN 300 328, EN 300 440, FCC CFR47 Part 15 and ARIB STD-T-66.		< –80 –57		dBm
Frequency error tolerance[1]	[1] requires minimum 80 ppm		±150		ppm
Symbol rate error tolerance[2]	[1] requires minimum 80 ppm		±1000		ppm

(1) Difference between center frequency of the received RF signal and local oscillator frequency.
(2) Difference between incoming symbol rate and the internally generated symbol rate

Product Folder Link(s): *CC2530F32   CC2530F64 CC2530F128   CC2530F256*

RF TRANSMIT SECTION

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C, VDD = 3 V and f_c = 2440 MHz, unless otherwise noted.

Boldface limits apply over the entire operating range, T_A = –40°C to 125°C, VDD = 2 V to 3.6 V and f_c = 2394 MHz to 2507 MHz.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Nominal output power	Delivered to a single-ended 50-Ω load through a balun using maximum-recommended output-power setting [1] requires minimum –3 dBm	0 **–8**	4.5	8 **10**	dBm
Programmable output power range			32		dB
Spurious emissions Measured conducted according to stated regulations. Only largest spurious emission stated within each band.	Max recommended output power setting[1] 25 MHz–1000 MHz (outside restricted bands) 25 MHz–2400 MHz (within FCC restricted bands) 25 MHz–1000 MHz (within ETSI restricted bands) 1800–1900 MHz (ETSI restricted band) 5150–5300 MHz (ETSI restricted band) At 2 × f_c and 3 × f_c (FCC restricted band) At 2 × f_c and 3 × f_c (ETSI EN 300-440 and EN 300-328)[2] 1 GHz–12.75 GHz (outside restricted bands) At 2483.5 MHz and above (FCC restricted band) f_c= 2480 MHz[3]		 –60 –60 –60 –57 –55 –42 –31 –53 –42		dBm
Error vector magnitude (EVM)	Measured as defined by [1] using maximum-recommended output-power setting [1] requires maximum 35%.		2%		
Optimum load impedance	Differential impedance as seen from the RF port (RF_P and RF_N) towards the antenna		69 + j29		Ω

(1)   Texas Instruments CC2530 EM reference design is suitable for systems targeting compliance with EN 300 328, EN 300 440, FCC CFR47 Part 15 and ARIB STD-T-66.
(2)   Margins for passing conducted requirements at the third harmonic can be improved by using a simple band-pass filter connected between matching network and RF connector (1.8 pF in parallel with 1.6 nH); this filter must be connected to a good RF ground.
(3)   Margins for passing FCC requirements at 2483.5 MHz and above when transmitting at 2480 MHz can be improved by using a lower output-power setting or having less than 100% duty cycle.

32-MHz CRYSTAL OSCILLATOR

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C and VDD = 3 V, unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Crystal frequency			32		MHz
	Crystal frequency accuracy requirement[1]		–40		40	ppm
ESR	Equivalent series resistance			6	60	Ω
C_0	Crystal shunt capacitance			1	7	pF
C_L	Crystal load capacitance			10	16	pF
	Start-up time			0.3		ms
	Power-down guard time	The crystal oscillator must be in power down for a guard time before it is used again. This requirement is valid for all modes of operation. The need for power-down guard time can vary with crystal type and load.		3		ms

(1)   Including aging and temperature dependency, as specified by [1]

32.768-kHz CRYSTAL OSCILLATOR

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C and VDD = 3 V, unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Crystal frequency			32.768		kHz
	Crystal frequency accuracy requirement[1]		–40		40	ppm
ESR	Equivalent series resistance			40	130	kΩ
C_0	Crystal shunt capacitance			0.9	2	pF
C_L	Crystal load capacitance			12	16	pF
	Start-up time			0.4		s

(1)   Including aging and temperature dependency, as specified by [1]

32-kHz RC OSCILLATOR

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C and VDD = 3 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Calibrated frequency[1]			32.753		kHz
Frequency accuracy after calibration			±0.2%		
Temperature coefficient[2]			0.4		%/°C
Supply-voltage coefficient[3]			3		%/V
Calibration time[4]			2		ms

(1)   The calibrated 32-kHz RC oscillator frequency is the 32-MHz XTAL frequency divided by 977.
(2)   Frequency drift when temperature changes after calibration
(3)   Frequency drift when supply voltage changes after calibration
(4)   When the 32-kHz RC oscillator is enabled, it is calibrated when a switch from the 16-MHz RC oscillator to the 32-MHz crystal oscillator is performed while SLEEPCMD.OSC32K_CALDIS is 0.

16-MHz RC OSCILLATOR

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C and VDD = 3 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Frequency[1]			16		MHz
Uncalibrated frequency accuracy			±18%		
Calibrated frequency accuracy			±0.6%	±1%	
Start-up time				10	µs
Initial calibration time[2]			50		µs

(1)   The calibrated 16-MHz RC oscillator frequency is the 32-MHz XTAL frequency divided by 2.
(2)   When the 16-MHz RC oscillator is enabled, it is calibrated when a switch from the 16-MHz RC oscillator to the 32-MHz crystal oscillator
        is performed while SLEEPCMD.OSC_PD is set to 0.

RSSI/CCA CHARACTERISTICS

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C and VDD = 3 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
RSSI range			100		dB
Absolute uncalibrated RSSI/CCA accuracy			±4		dB
RSSI/CCA offset[1]			73		dB
Step size (LSB value)			1		dB

(1)   Real RSSI = Register value – offset

FREQEST CHARACTERISTICS

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C and VDD = 3 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
FREQEST range			±250		kHz
FREQEST accuracy			±40		kHz
FREQEST offset[1]			20		kHz
Step size (LSB value)			7.8		kHz

(1)   Real FREQEST = Register value – offset

FREQUENCY SYNTHESIZER CHARACTERISTICS

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C, VDD = 3 V and f_c = 2440 MHz, unless
otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Phase noise, unmodulated carrier	At ±1-MHz offset from carrier		−110		dBc/Hz
	At ±2-MHz offset from carrier		−117		
	At ±5-MHz offset from carrier		−122		

ANALOG TEMPERATURE SENSOR

Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C and VDD = 3 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output at 25°C			1480		12-bit ADC
Temperature coefficient			4.5		/1°C
Voltage coefficient	Measured using integrated ADC using internal bandgap voltage reference and maximum resolution		1		/0.1 V
Initial accuracy without calibration			±10		°C
Accuracy using 1-point calibration (entire temperature range)			±5		°C
Current consumption when enabled (ADC current not included)			0.5		mA

OP-AMP CHARACTERISTICS

T_A = 25°C, VDD = 3 V . All measurement results are obtained using the CC2530 reference designs post-calibration.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Chopping Configuration, Register APCFG = 0x07, OPAMPMC = 0x03, OPAMPC = 0x01						
	Output maximum voltage			VDD – 0.07		V
	Output minimum voltage			0.07		V
	Open-loop gain			108		dB
	Gain-bandwidth product			2		MHz
	Slew rate			107		V/µs
	Input maximum voltage			VDD + 0.13		V
	Intput minimum voltage			–55		mV
	Input offset voltage			40		µV
CMRR	Common-mode rejection ratio			90		dB
	Supply current			0.4		mA
	Input noise voltage	f = 0.01 Hz to 1 Hz		1.1		nV/√(Hz)
		f = 0.1 Hz to 10 Hz		1.7		
Non-Chopping Configuration, Register APCFG = 0x07, OPAMPMC = 0x00, OPAMPC = 0x01						
	Output maximum voltage			VDD – 0.07		V
	Output minimum voltage			0.07		V
	Open-loop gain			108		dB
	Gain-bandwidth product			2		MHz
	Slew rate			107		V/µs
	Input maximum voltage			VDD + 0.13		V
	Intput minimum voltage			–55		mV
	Input offset voltage			0.8		mV
CMRR	Common-mode rejection ratio			90		dB
	Supply current			0.4		mA
	Input noise voltage	f = 0.01 Hz to 1 Hz		60		nV/√(Hz)
		f = 0.1 Hz to 10 Hz		65		

COMPARATOR CHARACTERISTICS

T_A = 25°C, VDD = 3 V. All measurement results are obtained using the CC2530 reference designs, post-calibration.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Common-mode maximum voltage			VDD		V
Common-mode minimum voltage			–0.3		
Input offset voltage			1		mV
Offset vs temperature			16		µV/°C
Offset vs operating voltage			4		mV/V
Supply current			230		nA
Hysteresis			0.15		mV

Product Folder Link(s): *CC2530F32 CC2530F64 CC2530F128 CC2530F256*

ADC CHARACTERISTICS

T_A = 25°C and VDD = 3 V, unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Input voltage	VDD is voltage on AVDD5 pin	0		VDD	V
	External reference voltage	VDD is voltage on AVDD5 pin	0		VDD	V
	External reference voltage differential	VDD is voltage on AVDD5 pin	0		VDD	V
	Input resistance, signal	Using 4-MHz clock speed		197		kΩ
	Full-scale signal[1]	Peak-to-peak, defines 0 dBFS		2.97		V
ENOB[1]	Effective number of bits	Single-ended input, 7-bit setting		5.7		bits
		Single-ended input, 9-bit setting		7.5		
		Single-ended input, 10-bit setting		9.3		
		Single-ended input, 12-bit setting		10.8		
		Differential input, 7-bit setting		6.5		
		Differential input, 9-bit setting		8.3		
		Differential input, 10-bit setting		10.0		
		Differential input, 12-bit setting		11.5		
	Useful power bandwidth	7-bit setting, both single and differential		0–20		kHz
THD[1]	Total harmonic distortion	Single-ended input, 12-bit setting, –6 dBFS		–75.2		dB
		Differential input, 12-bit setting, –6 dBFS		–86.6		
	Signal to nonharmonic ratio[1]	Single-ended input, 12-bit setting		70.2		dB
		Differential input, 12-bit setting		79.3		
		Single-ended input, 12-bit setting, –6 dBFS		78.8		
		Differential input, 12-bit setting, –6 dBFS		88.9		
CMRR	Common-mode rejection ratio	Differential input, 12-bit setting, 1-kHz sine (0 dBFS), limited by ADC resolution		>84		dB
	Crosstalk	Single-ended input, 12-bit setting, 1-kHz sine (0 dBFS), limited by ADC resolution		>84		dB
	Offset	Midscale		–3		mV
	Gain error			0.68		%
DNL[1]	Differential nonlinearity	12-bit setting, mean		0.05		LSB
		12-bit setting, maximum		0.9		
INL[1]	Integral nonlinearity	12-bit setting, mean		4.6		LSB
		12-bit setting, maximum		13.3		
SINAD[1] (–THD+N)	Signal-to-noise-and-distortion	Single-ended input, 7-bit setting		35.4		dB
		Single-ended input, 9-bit setting		46.8		
		Single-ended input, 10-bit setting		57.5		
		Single-ended input, 12-bit setting		66.6		
		Differential input, 7-bit setting		40.7		
		Differential input, 9-bit setting		51.6		
		Differential input, 10-bit setting		61.8		
		Differential input, 12-bit setting		70.8		
	Conversion time	7-bit setting		20		µs
		9-bit setting		36		
		10-bit setting		68		
		12-bit setting		132		
	Power consumption			1.2		mA
	Internal reference voltage			1.15		V
	Internal reference VDD coefficient			4		mV/V
	Internal reference temperature coefficient			0.4		mV/10°C

(1)   Measured with 300-Hz sine-wave input and VDD as reference.

CONTROL INPUT AC CHARACTERISTICS

T_A = –40°C to 125°C, VDD = 2 V to 3.6 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
System clock, f_{SYSCLK} t_{SYSCLK} = 1/f_{SYSCLK}	The undivided system clock is 32 MHz when crystal oscillator is used. The undivided system clock is 16 MHz when calibrated 16-MHz RC oscillator is used.	16		32	MHz
RESET_N low duration	See item 1, Figure 1. This is the shortest pulse that is recognized as a complete reset pin request. Note that shorter pulses may be recognized but might not lead to complete reset of all modules within the chip.	1			μs
Interrupt pulse duration	See item 2, Figure 1.This is the shortest pulse that is recognized as an interrupt request.	20			ns

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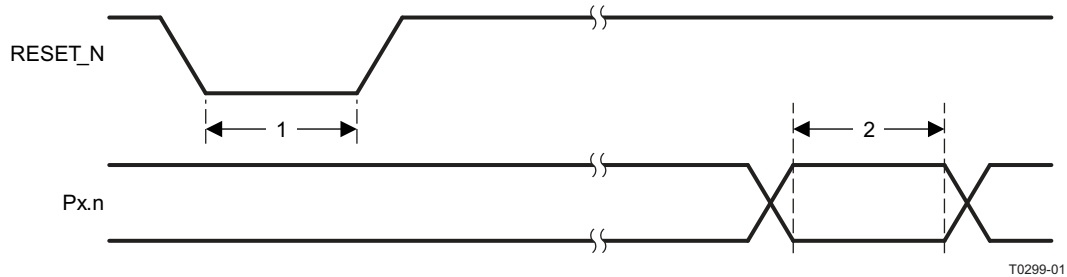
Figure 1. Control Input AC Characteristics

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Product Folder Link(s): *CC2530F32   CC2530F64 CC2530F128   CC2530F256*

SPI AC CHARACTERISTICS

T_A = –40°C to 125°C, VDD = 2 V to 3.6 V

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_1	SCK period	Master, RX and TX	250			ns
		Slave, RX and TX	250			
	SCK duty cycle	Master		50%		
t_2	SSN low to SCK	Master	63			ns
		Slave	63			
t_3	SCK to SSN high	Master	63			ns
		Slave	63			
t_4	MOSI early out	Master, load = 10 pF			7	ns
t_5	MOSI late out	Master, load = 10 pF			10	ns
t_6	MISO setup	Master	90			ns
t_7	MISO hold	Master	10			ns
	SCK duty cycle	Slave		50%		ns
t_{10}	MOSI setup	Slave	35			ns
t_{11}	MOSI hold	Slave	10			ns
t_9	MISO late out	Slave, load = 10 pF			95	ns
	Operating frequency	Master, TX only			8	MHz
		Master, RX and TX			4	
		Slave, RX only			8	
		Slave, RX and TX			4	
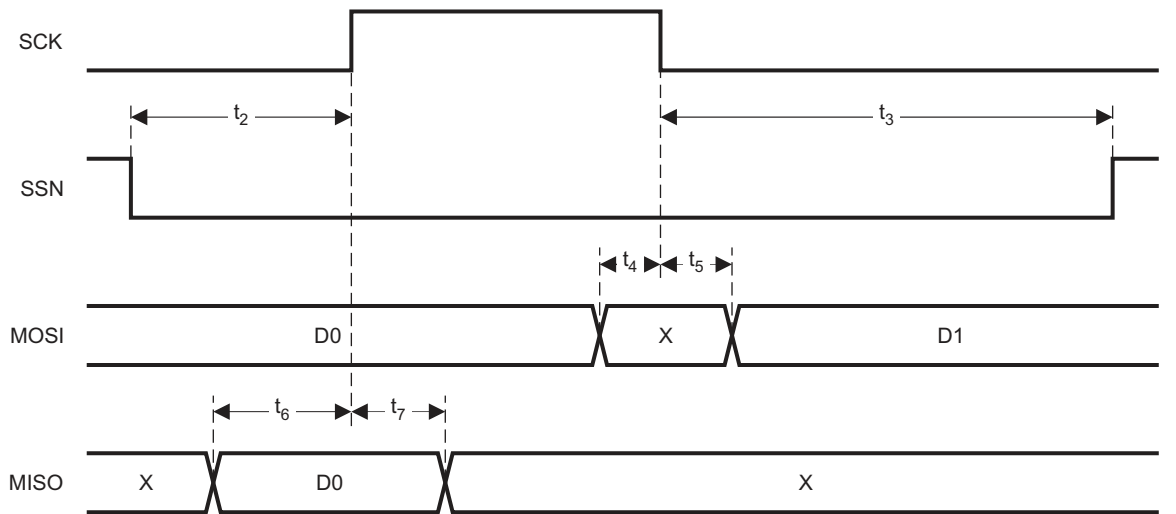
Figure 2. SPI Master AC Characteristics
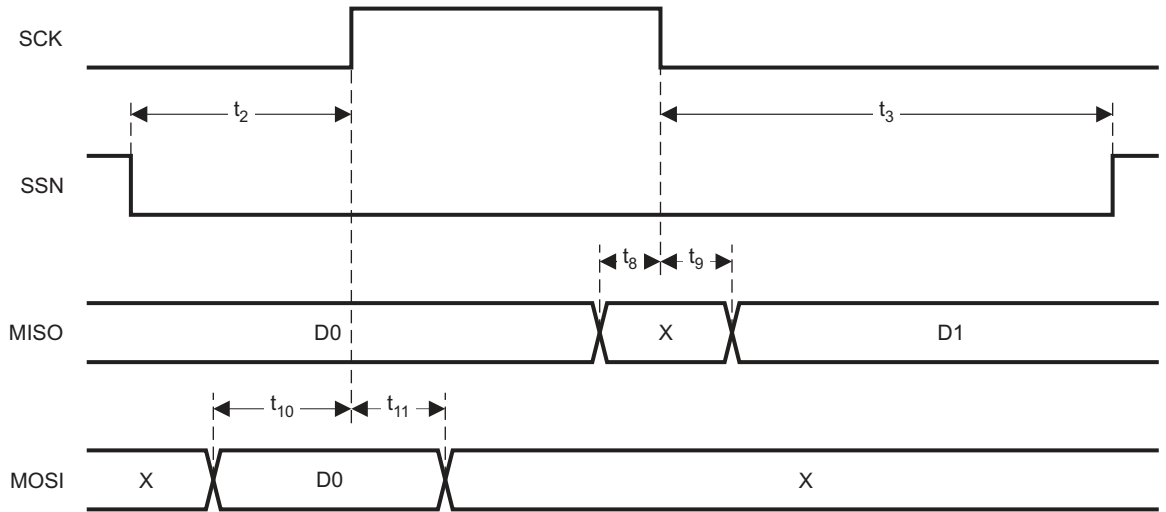
Figure 3. SPI Slave AC Characteristics

Product Folder Link(s): *CC2530F32   CC2530F64 CC2530F128   CC2530F256*

DEBUG INTERFACE AC CHARACTERISTICS

T_A = –40°C to 125°C, VDD = 2 V to 3.6 V, unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
f_{clk_dbg}	Debug clock frequency (see Figure 4)				12	MHz
t_1	Allowed high pulse on clock (see Figure 4)		35			ns
t_2	Allowed low pulse on clock (see Figure 4)		35			ns
t_3	EXT_RESET_N low to first falling edge on debug clock (see Figure 5)		167			ns
t_4	Falling edge on clock to EXT_RESET_N high (see Figure 5)		83			ns
t_5	EXT_RESET_N high to first debug command (see Figure 5)		83			ns
t_6	Debug data setup (see Figure 6)		2			ns
t_7	Debug data hold (see Figure 6)		4			ns
t_8	Clock-to-data delay (see Figure 6)	Load = 10 pF			30	ns
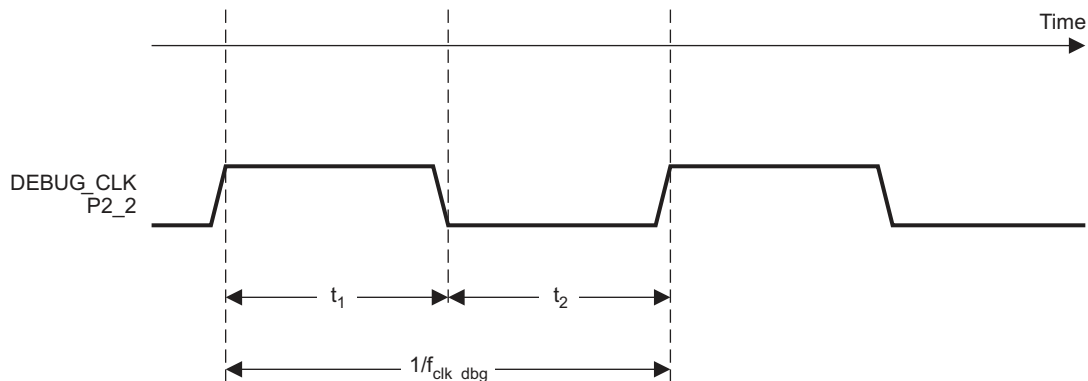
Figure 4. Debug Clock – Basic Timing
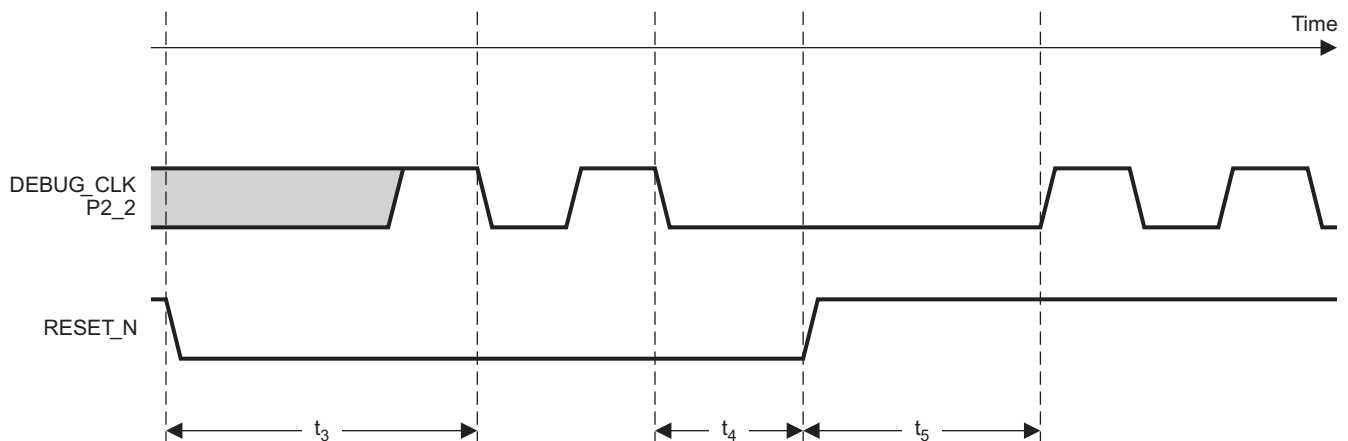
Figure 5. Data Setup and Hold Timing

Product Folder Link(s): *CC2530F32 CC2530F64 CC2530F128 CC2530F256*
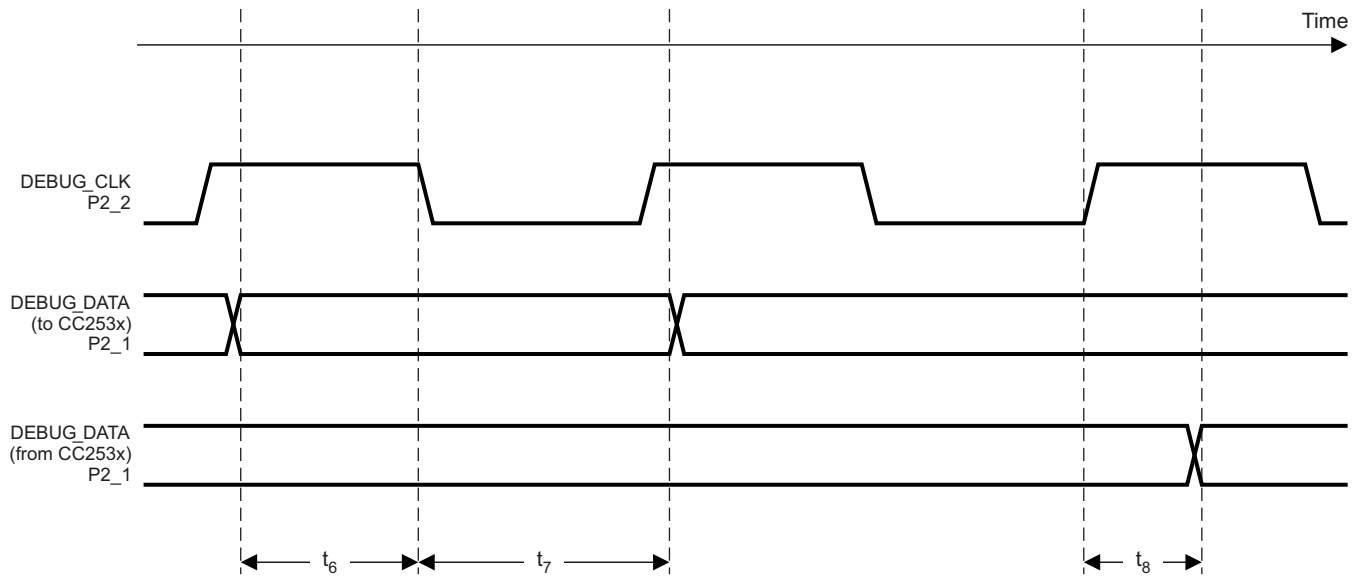
Figure 6. Debug Enable Timing

TIMER INPUTS AC CHARACTERISTICS

$T_A = -40°C$ to $125°C$, VDD = 2 V to 3.6 V, unless otherwise noted.

| PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input capture pulse duration	Synchronizers determine the shortest input pulse that can be recognized. The synchronizers operate at the current system clock rate (16 or 32 MHz).	1.5			t_{SYSCLK}

Product Folder Link(s): *CC2530F32   CC2530F64 CC2530F128   CC2530F256*

DC CHARACTERISTICS

T_A = 25°C, VDD = 3 V, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Logic-0 input voltage				0.5	V
Logic-1 input voltage		2.5			V
Logic-0 input current	Input equals 0 V	−50		50	nA
Logic-1 input current	Input equals VDD	−50		50	nA
I/O-pin pullup and pulldown resistors			20		kΩ
Logic-0 output voltage, 4-mA pins	Output load 4 mA			0.5	V
Logic-1 output voltage, 4-mA pins	Output load 4 mA	2.4			V
Logic-0 output voltage, 20-mA pins	Output load 20 mA			0.5	V
Logic-1 output voltage, 20-mA pins	Output load 20 mA	2.4			V

DEVICE INFORMATION

PIN DESCRIPTIONS

The CC2530 pinout is shown in Figure 7 and a short description of the pins follows.
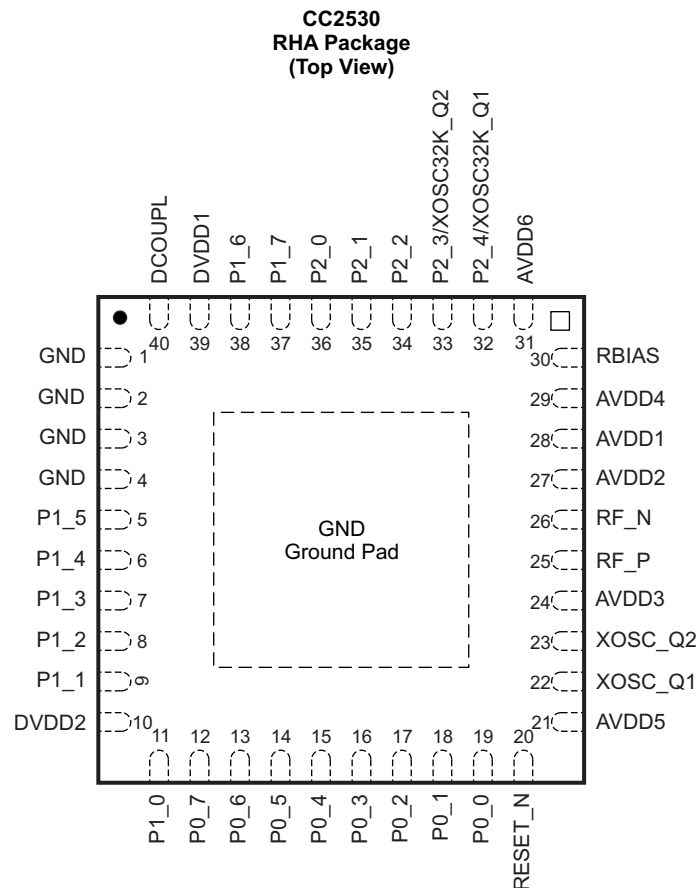
NOTE: The exposed ground pad must be connected to a solid ground plane, as this is the ground connection for the chip.

Figure 7. Pinout Top View

Table 1. Pin Descriptions

PIN NAME	PIN	PIN TYPE	DESCRIPTION
AVDD1	28	Power (analog)	2-V–3.6-V analog power-supply connection
AVDD2	27	Power (analog)	2-V–3.6-V analog power-supply connection
AVDD3	24	Power (analog)	2-V–3.6-V analog power-supply connection
AVDD4	29	Power (analog)	2-V–3.6-V analog power-supply connection
AVDD5	21	Power (analog)	2-V–3.6-V analog power-supply connection
AVDD6	31	Power (analog)	2-V–3.6-V analog power-supply connection
DCOUPL	40	Power (digital)	1.8-V digital power-supply decoupling. Do not use for supplying external circuits.
DVDD1	39	Power (digital)	2-V–3.6-V digital power-supply connection
DVDD2	10	Power (digital)	2-V–3.6-V digital power-supply connection
GND	—	Ground	The ground pad must be connected to a solid ground plane.
GND	1, 2, 3, 4	Unused pins	Connect to GND
P0_0	19	Digital I/O	Port 0.0
P0_1	18	Digital I/O	Port 0.1
P0_2	17	Digital I/O	Port 0.2
P0_3	16	Digital I/O	Port 0.3
P0_4	15	Digital I/O	Port 0.4
P0_5	14	Digital I/O	Port 0.5
P0_6	13	Digital I/O	Port 0.6
P0_7	12	Digital I/O	Port 0.7
P1_0	11	Digital I/O	Port 1.0 – 20-mA drive capability
P1_1	9	Digital I/O	Port 1.1 – 20-mA drive capability
P1_2	8	Digital I/O	Port 1.2
P1_3	7	Digital I/O	Port 1.3
P1_4	6	Digital I/O	Port 1.4
P1_5	5	Digital I/O	Port 1.5
P1_6	38	Digital I/O	Port 1.6
P1_7	37	Digital I/O	Port 1.7
P2_0	36	Digital I/O	Port 2.0
P2_1	35	Digital I/O	Port 2.1
P2_2	34	Digital I/O	Port 2.2
P2_3/ XOSC32K_Q2	33	Digital I/O, Analog I/O	Port 2.3/32.768 kHz XOSC
P2_4/ XOSC32K_Q1	32	Digital I/O, Analog I/O	Port 2.4/32.768 kHz XOSC
RBIAS	30	Analog I/O	External precision bias resistor for reference current
RESET_N	20	Digital input	Reset, active-low
RF_N	26	RF I/O	Negative RF input signal to LNA during RX Negative RF output signal from PA during TX
RF_P	25	RF I/O	Positive RF input signal to LNA during RX Positive RF output signal from PA during TX
XOSC_Q1	22	Analog I/O	32-MHz crystal oscillator pin 1 or external-clock input
XOSC_Q2	23	Analog I/O	32-MHz crystal oscillator pin 2

Product Folder Link(s): *CC2530F32 CC2530F64 CC2530F128 CC2530F256*

CIRCUIT DESCRIPTION
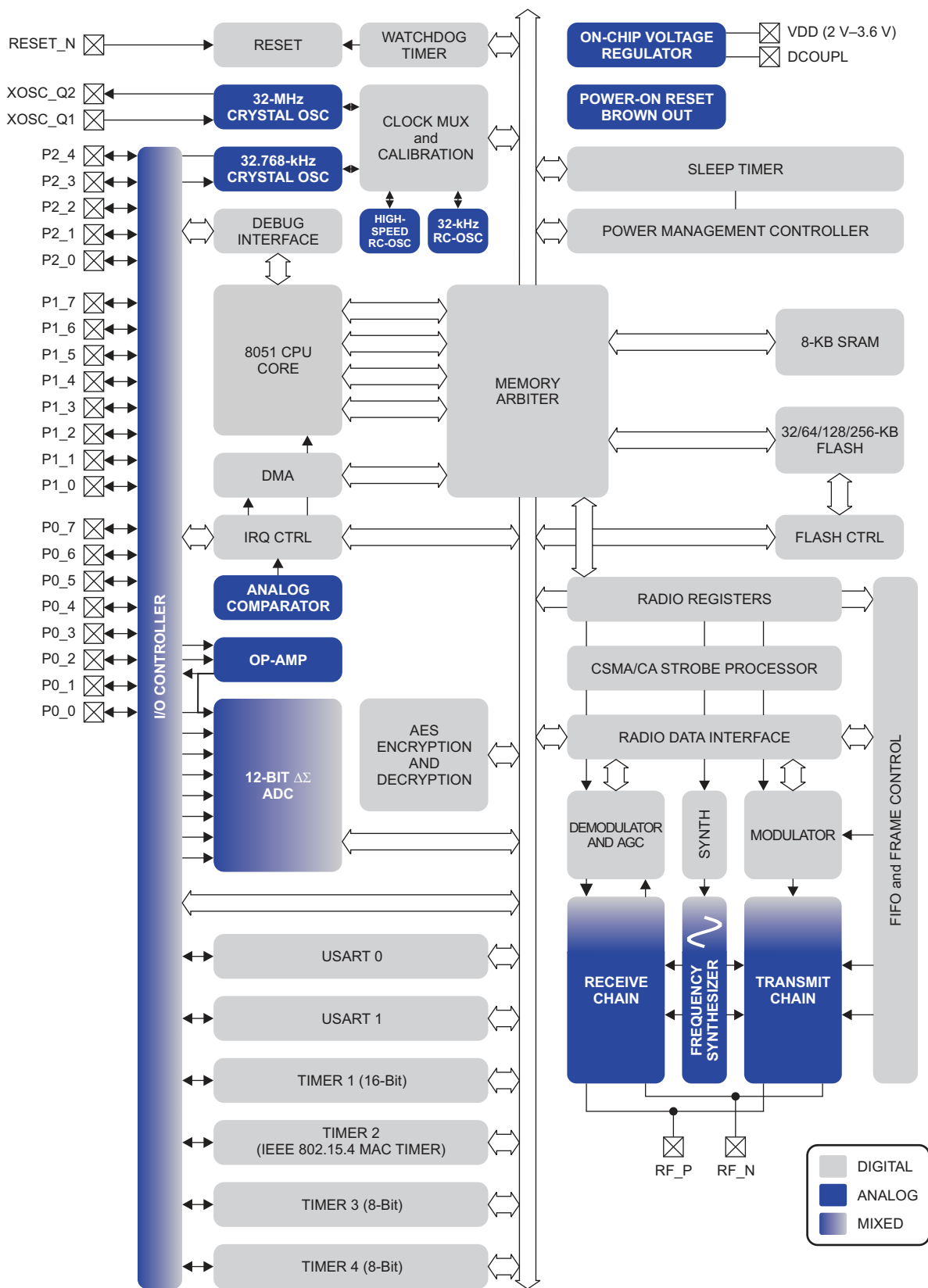
B0301-02

Figure 8. CC2530 Block Diagram

A block diagram of the CC2530 is shown in Figure 8. The modules can be roughly divided into one of three categories: CPU- and memory-related modules; modules related to peripherals, clocks, and power management; and radio-related modules. In the following subsections, a short description of each module that appears in Figure 8 is given.

For more details about the modules and their usage, see the corresponding chapters in the CC253x User's Guide (SWRU191).

CPU and Memory

The **8051 CPU** core used in the CC253x device family is a single-cycle 8051-compatible core. It has three different memory-access buses (SFR, DATA and CODE/XDATA) with single-cycle access to SFR, DATA, and the main SRAM. It also includes a debug interface and an 18-input extended interrupt unit.

The **interrupt controller** services a total of 18 interrupt sources, divided into six interrupt groups, each of which is associated with one of four interrupt priorities. Any interrupt service request is serviced also when the device is in idle mode by going back to active mode. Some interrupts can also wake up the device from sleep mode (power modes 1–3).

The **memory arbiter** is at the heart of the system, as it connects the CPU and DMA controller with the physical memories and all peripherals through the SFR bus. The memory arbiter has four memory access points, access of which can map to one of three physical memories: an 8-KB SRAM, flash memory, and XREG/SFR registers. It is responsible for performing arbitration and sequencing between simultaneous memory accesses to the same physical memory.

The **8-KB SRAM** maps to the DATA memory space and to parts of the XDATA memory spaces. The 8-KB SRAM is an ultralow-power SRAM that retains its contents even when the digital part is powered off (power modes 2 and 3). This is an important feature for low-power applications.

The **32/64/128/256 KB flash block** provides in-circuit programmable non-volatile program memory for the device, and maps into the CODE and XDATA memory spaces. In addition to holding program code and constants, the non-volatile memory allows the application to save data that must be preserved such that it is available after restarting the device. Using this feature one can, e.g., use saved network-specific data to avoid the need for a full start-up and network find-and-join process .

Clocks and Power Management

The digital core and peripherals are powered by a 1.8-V low-dropout **voltage regulator**. It provides **power management** functionality that enables low power operation for long battery life using different power modes. Five different **reset** sources exist to reset the device.

Peripherals

The CC2530 includes many different peripherals that allow the application designer to develop advanced applications.

The **debug interface** implements a proprietary two-wire serial interface that is used for in-circuit debugging. Through this debug interface, it is possible to perform an erasure of the entire flash memory, control which oscillators are enabled, stop and start execution of the user program, execute supplied instructions on the 8051 core, set code breakpoints, and single-step through instructions in the code. Using these techniques, it is possible to perform in-circuit debugging and external flash programming elegantly.

The device contains flash memory for storage of program code. The flash memory is programmable from the user software and through the debug interface. The **flash controller** handles writing and erasing the embedded flash memory. The flash controller allows page-wise erasure and 4-bytewise programming.

The **I/O controller** is responsible for all general-purpose I/O pins. The CPU can configure whether peripheral modules control certain pins or whether they are under software control, and if so, whether each pin is configured as an input or output and if a pullup or pulldown resistor in the pad is connected. CPU interrupts can be enabled on each pin individually. Each peripheral that connects to the I/O pins can choose between two different I/O pin locations to ensure flexibility in various applications.

A versatile five-channel **DMA controller** is available in the system, accesses memory using the XDATA memory space, and thus has access to all physical memories. Each channel (trigger, priority, transfer mode, addressing mode, source and destination pointers, and transfer count) is configured with DMA descriptors anywhere in memory. Many of the hardware peripherals (AES core, flash controller, USARTs, timers, ADC interface) achieve highly efficient operation by using the DMA controller for data transfers between SFR or XREG addresses and flash/SRAM.

Timer 1 is a 16-bit timer with timer/counter/PWM functionality. It has a programmable prescaler, a 16-bit period value, and five individually programmable counter/capture channels, each with a 16-bit compare value. Each of the counter/capture channels can be used as a PWM output or to capture the timing of edges on input signals. It can also be configured in **IR Generation Mode** where it counts Timer 3 periods and the output is ANDed with the output of Timer 3 to generate modulated consumer IR signals with minimal CPU interaction.

Timer 2 (the MAC Timer) is specially designed for supporting an IEEE 802.15.4 MAC or other time-slotted protocol in software. The timer has a configurable timer period and a 24-bit overflow counter that can be used to keep track of the number of periods that have transpired. A 40-bit capture register is also used to record the exact time at which a start-of-frame delimiter is received/transmitted or the exact time at which transmission ends, as well as two 16-bit output compare registers and two 24-bit overflow compare registers that can send various command strobes (start RX, start TX, etc.) at specific times to the radio modules.

Timer 3 and Timer 4 are 8-bit timers with timer/counter/PWM functionality. They have a programmable prescaler, an 8-bit period value, and one programmable counter channel with an 8-bit compare value. Each of the counter channels can be used as a PWM output.

The **sleep timer** is an ultralow-power timer that counts 32-kHz crystal oscillator or 32-kHz RC oscillator periods. The sleep timer runs continuously in all operating modes except power mode 3 (PM3). Typical applications of this timer are as a real-time counter or as a wake-up timer to come out of power mode 1 (PM1) or 2 (PM2).

The **ADC** supports 7 to 12 bits of resolution in a 30 kHz to 4 kHz bandwidth, respectively. DC and audio conversions with up to eight input channels (Port 0) are possible. The inputs can be selected as single-ended or differential. The reference voltage can be internal, AVDD, or a single-ended or differential external signal. The ADC also has a temperature-sensor input channel. The ADC can automate the process of periodic sampling or conversion over a sequence of channels.

The **operational amplifier** is intended to provide front-end buffering and gain for the ADC. Both inputs as well as the output are available on pins, so the feedback network is fully customizable. A chopper-stabilized mode is available for applications that need good accuracy with high gain.

The ultralow-power **analog comparator** enables applications to wake up from PM2 or PM3 based on an analog signal. Both inputs are brought out to pins; the reference voltage must be provided externally. The comparator output is connected to the I/O controller interrupt detector and can be treated by the MCU as a regular I/O pin interrupt.

The **random-number generator** uses a 16-bit LFSR to generate pseudorandom numbers, which can be read by the CPU or used directly by the command strobe processor. It can be seeded with random data from noise in the radio ADC.

The **AES encryption/decryption core** allows the user to encrypt and decrypt data using the AES algorithm with 128-bit keys. The core is able to support the AES operations required by IEEE 802.15.4 MAC security, the ZigBee network layer, and the application layer.

A built-in **watchdog timer** allows the CC2530 to reset itself in case the firmware hangs. When enabled by software, the watchdog timer must be cleared periodically; otherwise, it resets the device when it times out. It can alternatively be configured for use as a general 32-kHz timer.

USART 0 and USART 1 are each configurable as either a SPI master/slave or a UART. They provide double buffering on both RX and TX and hardware flow control and are thus well suited to high-throughput full-duplex applications. Each has its own high-precision baud-rate generator, thus leaving the ordinary timers free for other uses.

Radio

The CC2530 features an IEEE 802.15.4-compliant radio transceiver. The RF core controls the analog radio modules. In addition, it provides an interface between the MCU and the radio which makes it possible to issue commands, read status, and automate and sequence radio events. The radio also includes a packet-filtering and address-recognition module.
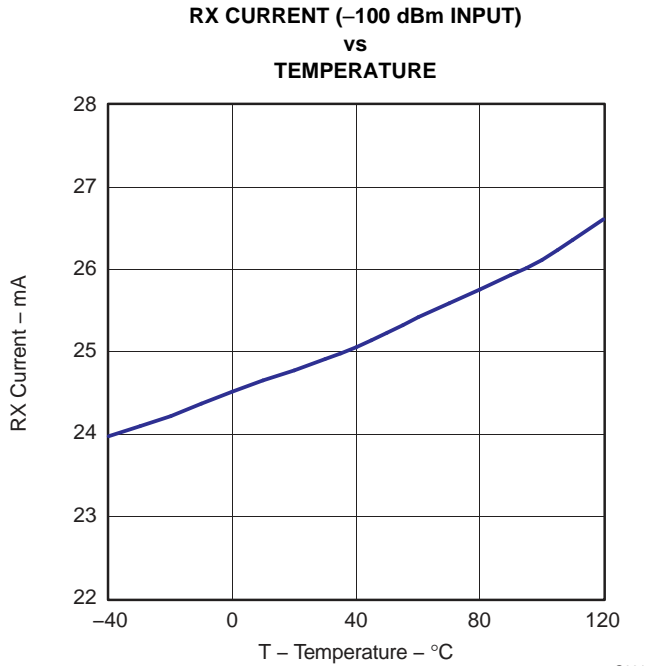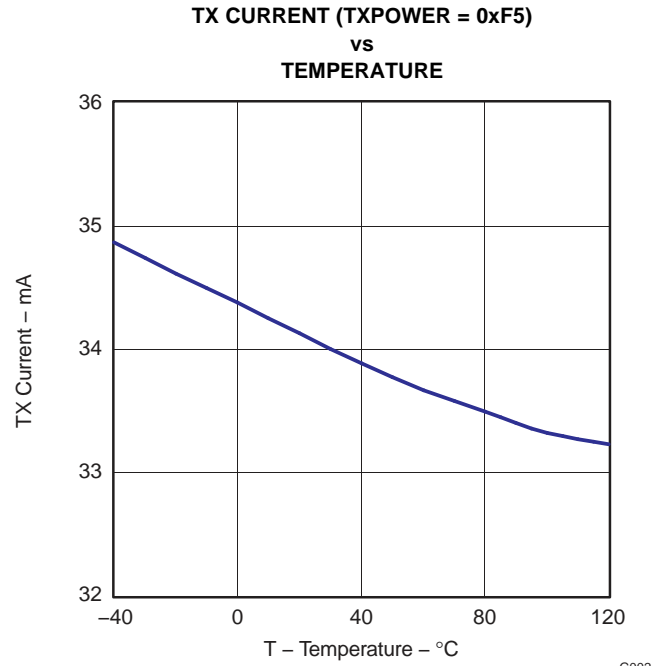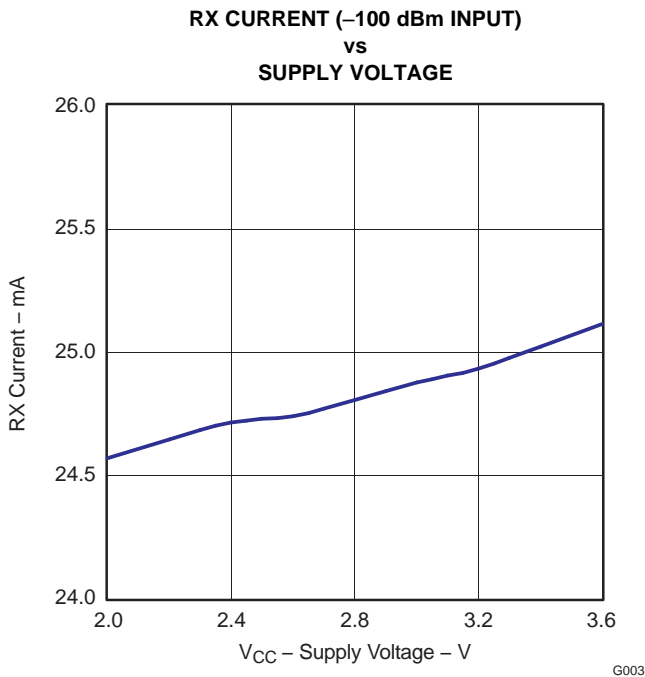
TYPICAL CHARACTERISTICS

**RX CURRENT (−100 dBm INPUT)
vs
TEMPERATURE**

Figure 9.

**TX CURRENT (TXPOWER = 0xF5)
vs
TEMPERATURE**

Figure 10.

**RX CURRENT (−100 dBm INPUT)
vs
SUPPLY VOLTAGE**

Figure 11.

**TX CURRENT (TXPOWER = 0xF5)
vs
SUPPLY VOLTAGE**

Figure 12.

Product Folder Link(s): *CC2530F32 CC2530F64 CC2530F128 CC2530F256*
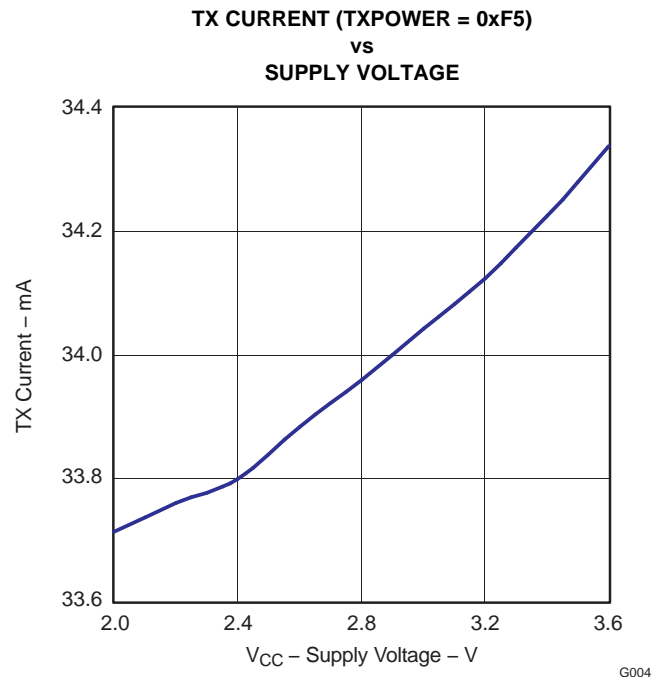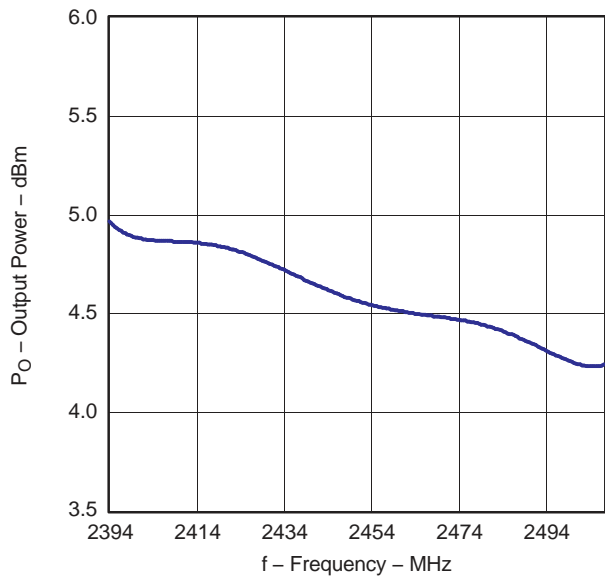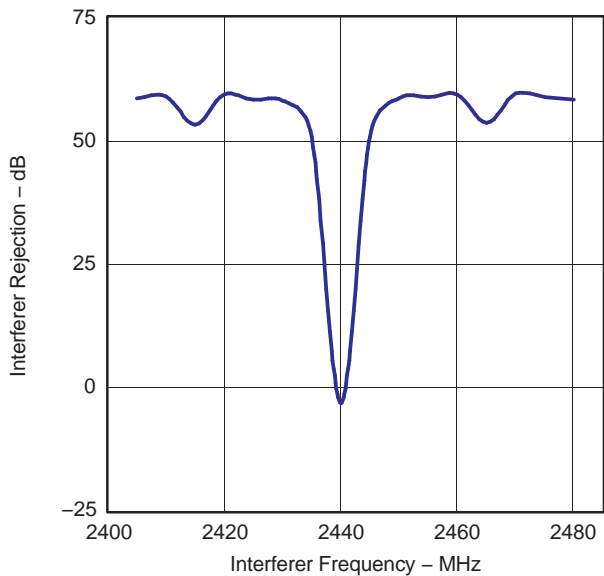
TYPICAL CHARACTERISTICS (continued)

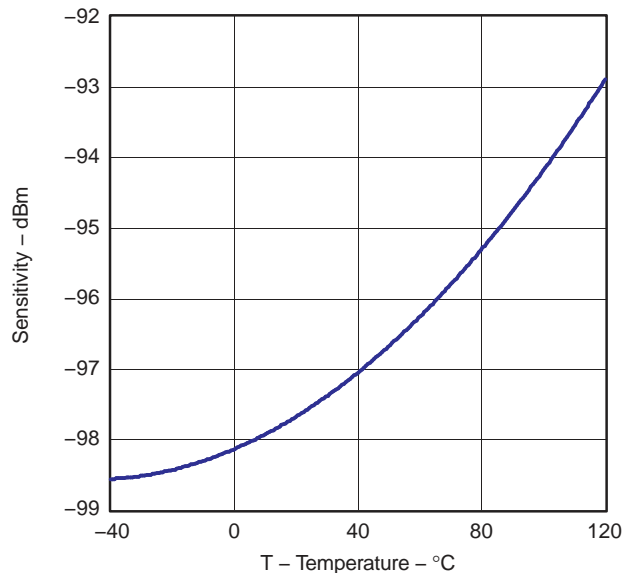OUTPUT POWER (TXPOWER = 0xF5)
vs
FREQUENCY

Figure 13.

INTERFERER REJECTION (802.15.4 INTERFERER)
vs
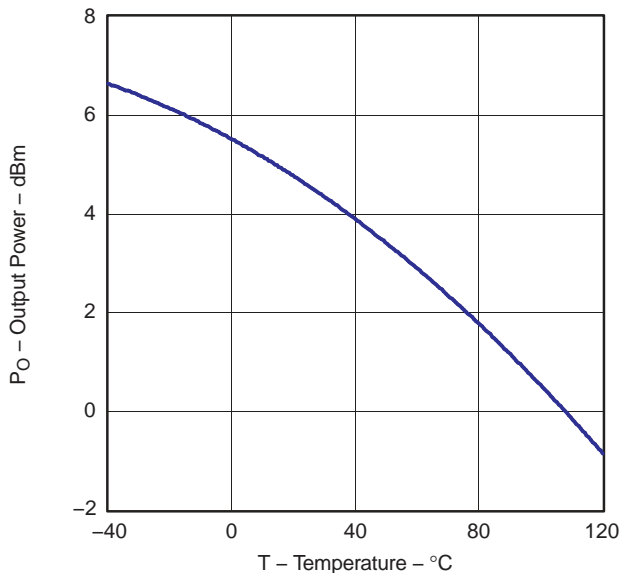INTERFERER FREQUENCY (CARRIER AT −82 dBm, 2440 MHz)

Figure 14.

SENSITIVITY
vs
TEMPERATURE

Figure 15.

OUTPUT POWER (TXPOWER = 0xF5)
vs
TEMPERATURE

Figure 16.

Product Folder Link(s): *CC2530F32 CC2530F64 CC2530F128 CC2530F256*

TYPICAL CHARACTERISTICS (continued)

OUTPUT POWER (TXPOWER = 0xF5)
vs
SUPPLY VOLTAGE

Figure 17.

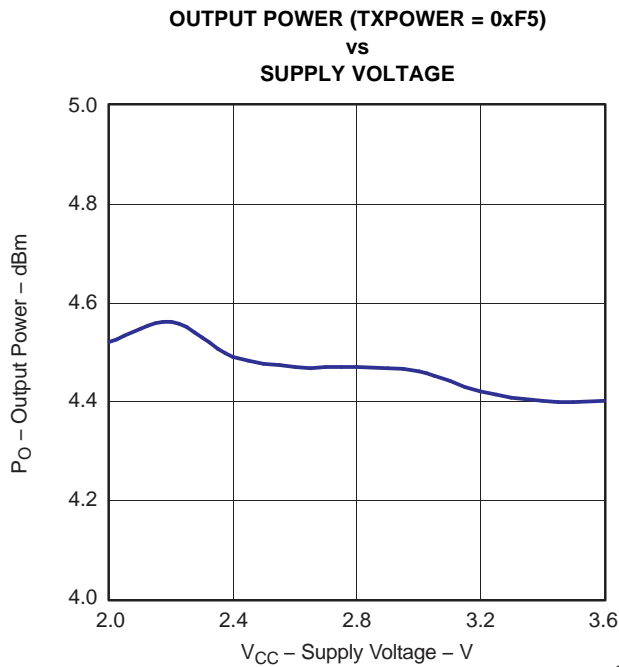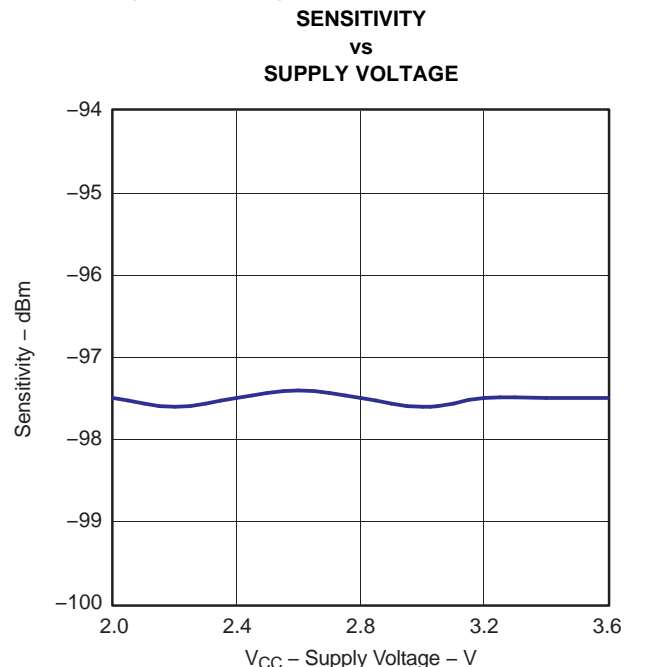SENSITIVITY
vs
SUPPLY VOLTAGE

Figure 18.

Table 2. Recommended Output Power Settings[1]

TXPOWER Register Setting	Typical Output Power (dBm)	Typical Current Consumption (mA)
0xF5	4.5	34
0xE5	2.5	31
0xD5	1	29
0xC5	–0.5	28
0xB5	–1.5	27
0xA5	–3	27
0x95	–4	26
0x85	–6	26
0x75	–8	25
0x65	–10	25
0x55	–12	25
0x45	–14	25
0x35	–16	25
0x25	–18	24
0x15	–20	24
0x05	–22	23
0x05 and TXCTRL = 0x09	–28	23

(1) Measured on Texas Instruments CC2530 EM reference design with T_A = 25°C, VDD = 3 V and f_c = 2440 MHz, unless otherwise noted. See References, Item 1, for recommended register settings.

APPLICATION INFORMATION

Few external components are required for the operation of the CC2530. A typical application circuit is shown in Figure 19. Typical values and description of external components are shown in Table 3.
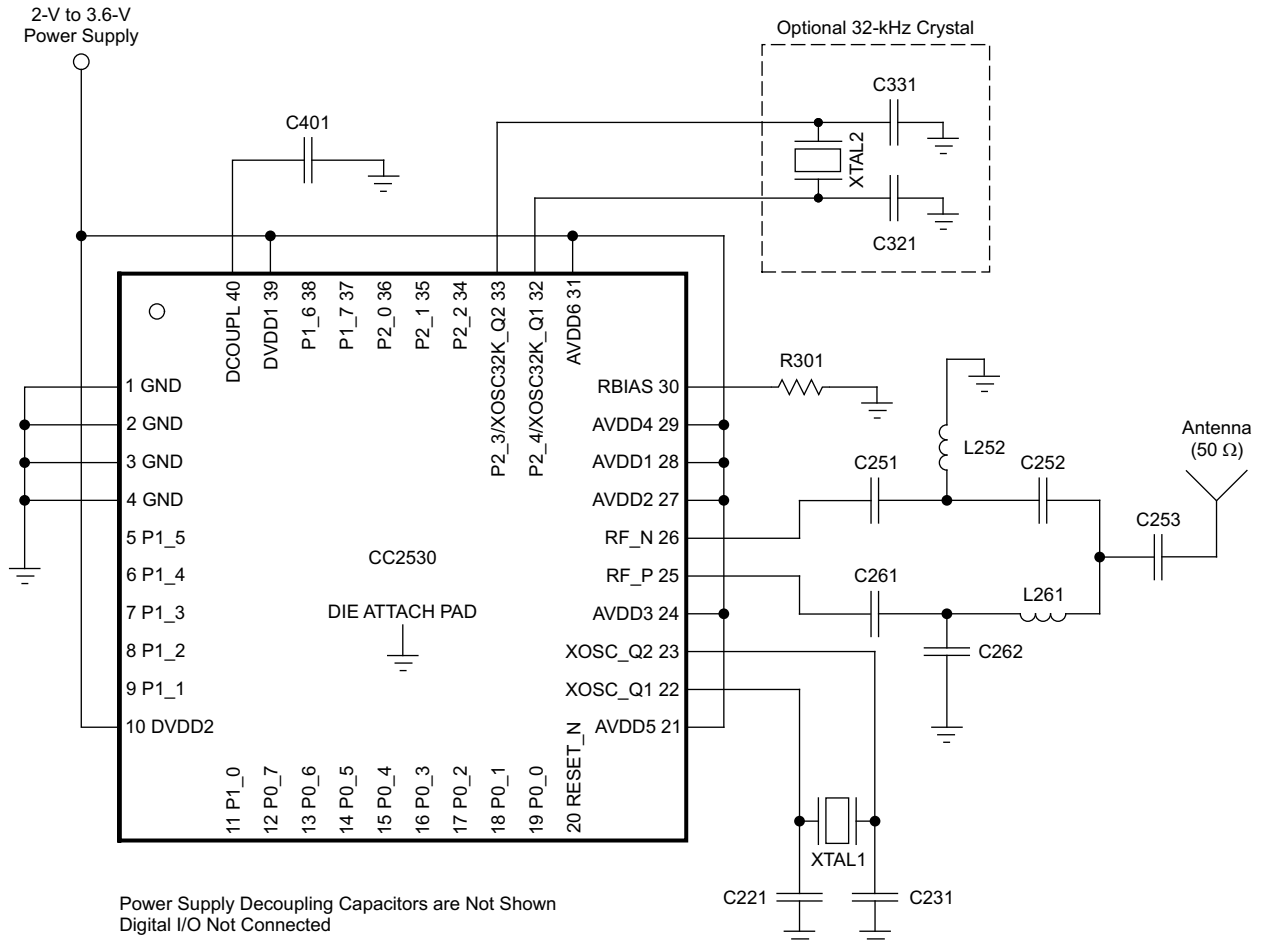
Figure 19. CC2530 Application Circuit

Table 3. Overview of External Components (Excluding Supply Decoupling Capacitors)

Component	Description	Value
C251	Part of the RF matching network	18 pF
C261	Part of the RF matching network	18 pF
L252	Part of the RF matching network	2 nH
L261	Part of the RF matching network	2 nH
C262	Part of the RF matching network	1 pF
C252	Part of the RF matching network	1 pF
C253	Part of the RF matching network	2.2 pF
C331	32kHz xtal loading capacitor	15 pF
C321	32kHz xtal loading capacitor	15 pF
C231	32MHz xtal loading capacitor	27 pF
C221	32MHz xtal loading capacitor	27 pF
C401	Decoupling capacitor for the internal digital regulator	1 μF

Product Folder Link(s): *CC2530F32 CC2530F64 CC2530F128 CC2530F256*

Table 3. Overview of External Components (Excluding Supply Decoupling Capacitors) (continued)

Component	Description	Value
R301	Resistor used for internal biasing	56 kΩ

Input/Output Matching

When using an unbalanced antenna such as a monopole, a balun should be used to optimize performance. The balun can be implemented using low-cost discrete inductors and capacitors. The recommended balun shown consists of C262, L261, C252, and L252.

If a balanced antenna such as a folded dipole is used, the balun can be omitted.

Crystal

An external 32-MHz crystal, XTAL1, with two loading capacitors (C221 and C231) is used for the 32-MHz crystal oscillator. See the *32-MHz Crystal Oscillator* section for details. The load capacitance seen by the 32-MHz crystal is given by:

$$C_L = \frac{1}{\frac{1}{C_{221}} + \frac{1}{C_{231}}} + C_{parasitic}$$

(1)

XTAL2 is an optional 32.768-kHz crystal, with two loading capacitors (C321 and C331) used for the 32.768-kHz crystal oscillator. The 32.768-kHz crystal oscillator is used in applications where both very low sleep-current consumption and accurate wake-up times are needed. The load capacitance seen by the 32.768-kHz crystal is given by:

$$C_L = \frac{1}{\frac{1}{C_{321}} + \frac{1}{C_{331}}} + C_{parasitic}$$

(2)

A series resistor may be used to comply with the ESR requirement.

On-Chip 1.8-V Voltage-Regulator Decoupling

The 1.8-V on-chip voltage regulator supplies the 1.8-V digital logic. This regulator requires a decoupling capacitor (C401) for stable operation.

Power-Supply Decoupling and Filtering

Proper power-supply decoupling must be used for optimum performance. The placement and size of the decoupling capacitors and the power supply filtering are very important to achieve the best performance in an application. TI provides a compact reference design that should be followed very closely.

References

1. IEEE Std. 802.15.4-2006: Wireless Medium Access Control (MAC) and Physical Layer (PHY) Specifications for Low-Rate Wireless Personal Area Networks (LR-WPANs)
   http://standards.ieee.org/getieee802/download/802.15.4-2006.pdf
2. CC253x User's Guide – CC253x System-on-Chip Solution for 2.4 GHz IEEE 802.15.4 and ZigBee Applications (SWRU191)

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Texas Instruments offers a wide selection of cost-effective, low-power RF solutions for proprietary and standard-based wireless applications for use in industrial and consumer applications. Our selection includes RF transceivers, RF transmitters, RF front ends, and System-on-Chips as well as various software solutions for the sub-1- and 2.4-GHz frequency bands.

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- Low-power RF and ZigBee module solutions and development tools
- RF certification services and RF circuit manufacturing

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TEXAS INSTRUMENTS

REVISION HISTORY

Product Folder Link(s): *CC2530F32 CC2530F64 CC2530F128 CC2530F256*

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
CC2530F128RHAR	ACTIVE	VQFN	RHA	40	2500	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 125	CC2530 F128	Samples
CC2530F128RHAT	ACTIVE	VQFN	RHA	40	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 125	CC2530 F128	Samples
CC2530F256RHAR	ACTIVE	VQFN	RHA	40	2500	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 125	CC2530 F256	Samples
CC2530F256RHAT	ACTIVE	VQFN	RHA	40	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 125	CC2530 F256	Samples
CC2530F32RHAR	ACTIVE	VQFN	RHA	40	2500	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 125	CC2530 F32	Samples
CC2530F32RHAT	ACTIVE	VQFN	RHA	40	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 125	CC2530 F32	Samples
CC2530F64RHAR	ACTIVE	VQFN	RHA	40	2500	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 125	CC2530 F64	Samples
CC2530F64RHAT	ACTIVE	VQFN	RHA	40	250	Green (RoHS & no Sb/Br)	CU NIPDAU | CU NIPDAUAG	Level-3-260C-168 HR	-40 to 125	CC2530 F64	Samples
HPA00700F256RHAR	ACTIVE	VQFN	RHA	40	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 125	CC2530 F256	Samples
HPA00702F128RHAR	ACTIVE	VQFN	RHA	40	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 125	CC2530 F128	Samples
HPA01064RHAR	ACTIVE	VQFN	RHA	40	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 125	CC2530 F256	Samples
HPA02146RHAR	ACTIVE	VQFN	RHA	40	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 125	CC2530 F256	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

[3] MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

[4] There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

[5] Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

[6] Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION

REEL DIMENSIONS

TAPE DIMENSIONS

A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

User Direction of Feed

Pocket Quadrants

*All dimensions are nominal

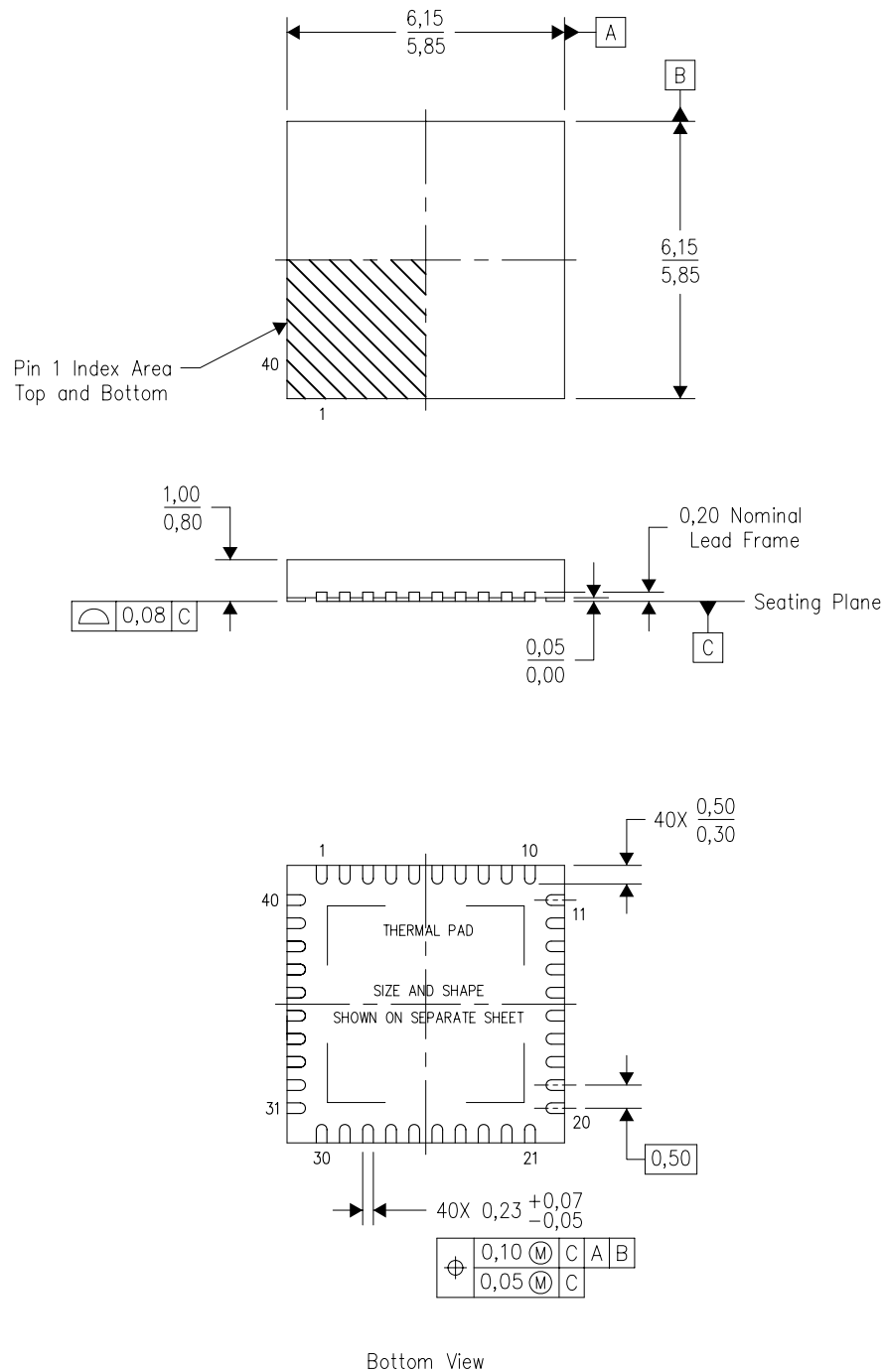| Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CC2530F128RHAR	VQFN	RHA	40	2500	330.0	16.4	6.3	6.3	1.5	12.0	16.0	Q2
CC2530F128RHAT	VQFN	RHA	40	250	180.0	16.4	6.3	6.3	1.5	12.0	16.0	Q2
CC2530F256RHAR	VQFN	RHA	40	2500	330.0	16.4	6.3	6.3	1.5	12.0	16.0	Q2
CC2530F256RHAT	VQFN	RHA	40	250	180.0	16.4	6.3	6.3	1.5	12.0	16.0	Q2
CC2530F32RHAR	VQFN	RHA	40	2500	330.0	16.4	6.3	6.3	1.5	12.0	16.0	Q2
CC2530F32RHAT	VQFN	RHA	40	250	180.0	16.4	6.3	6.3	1.5	12.0	16.0	Q2
CC2530F64RHAT	VQFN	RHA	40	250	180.0	16.4	6.3	6.3	1.5	12.0	16.0	Q2

TAPE AND REEL BOX DIMENSIONS

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
CC2530F128RHAR	VQFN	RHA	40	2500	336.6	336.6	28.6
CC2530F128RHAT	VQFN	RHA	40	250	213.0	191.0	55.0
CC2530F256RHAR	VQFN	RHA	40	2500	336.6	336.6	28.6
CC2530F256RHAT	VQFN	RHA	40	250	213.0	191.0	55.0
CC2530F32RHAR	VQFN	RHA	40	2500	336.6	336.6	28.6
CC2530F32RHAT	VQFN	RHA	40	250	213.0	191.0	55.0
CC2530F64RHAT	VQFN	RHA	40	250	213.0	191.0	55.0

RHA (S−PVQFN−N40)   PLASTIC QUAD FLATPACK NO−LEAD

Pin 1 Index Area
Top and Bottom

$\frac{6,15}{5,85}$

$\frac{6,15}{5,85}$

40

1

$\frac{1,00}{0,80}$

0,20 Nominal
Lead Frame

Seating Plane

$\bigtriangleup$ 0,08 C

$\frac{0,05}{0,00}$

1   10

40   11

THERMAL PAD

SIZE AND SHAPE
SHOWN ON SEPARATE SHEET

31   20

30   21

40X $\frac{0,50}{0,30}$

0,50

40X 0,23 $^{+0,07}_{-0,05}$

$\oplus$	0,10 Ⓜ	C	A	B
	0,05 Ⓜ	C		

Bottom View

4204276/E   06/11

NOTES: A.   All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M−1994.
         B.   This drawing is subject to change without notice.
         C.   QFN (Quad Flatpack No−Lead) Package configuration.
         D.   The package thermal pad must be soldered to the board for thermal and mechanical performance.
         E.   See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
         F.   Package complies to JEDEC MO−220 variation VJJD−2.

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RHA (S−PVQFN−N40)   PLASTIC QUAD FLATPACK NO−LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No−Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

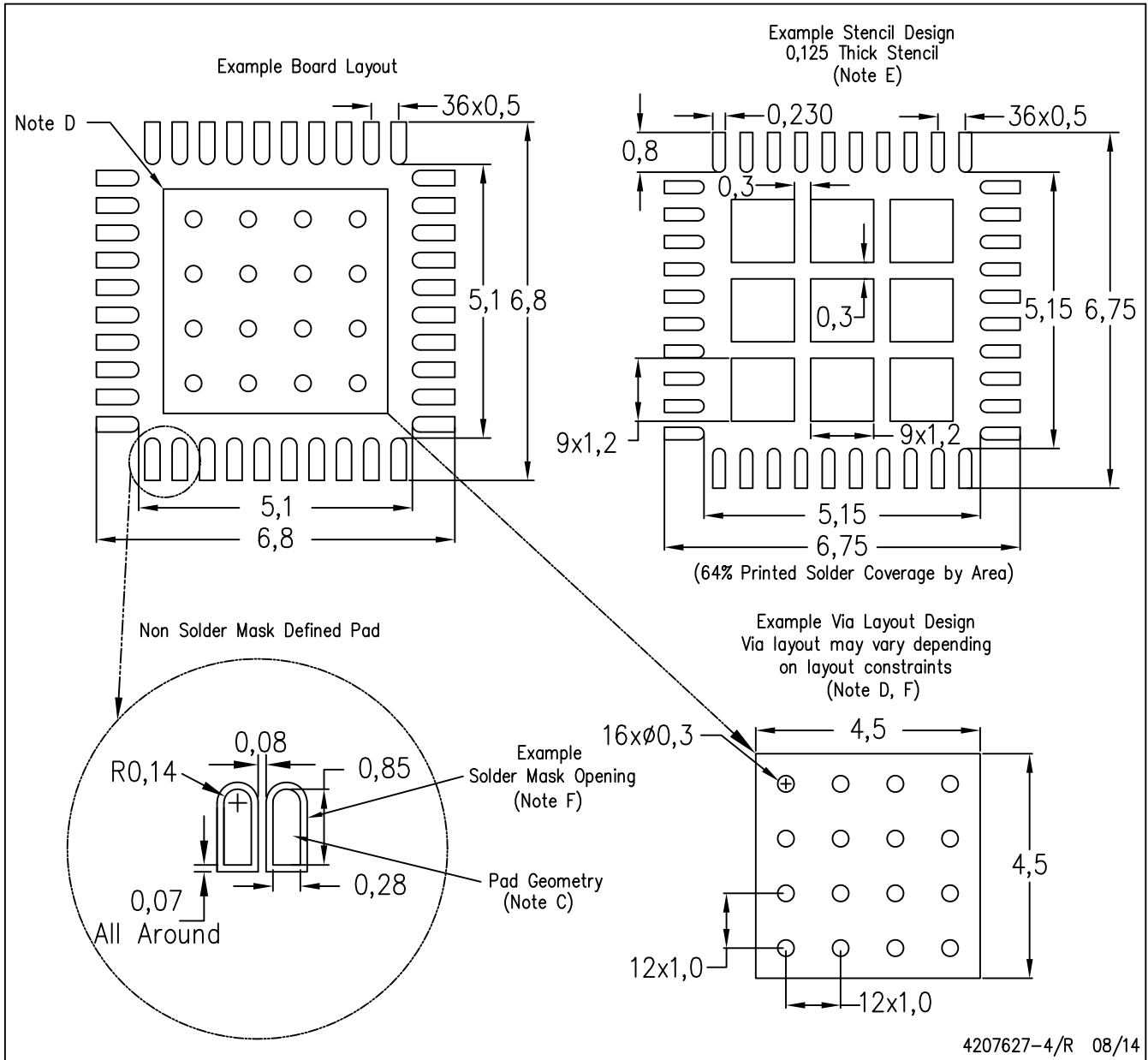The exposed thermal pad dimensions for this package are shown in the following illustration.

PIN 1 INDICATOR
C0,30

1   10

40

11
                                    Exposed Thermal Pad

4,50±0,10

31

20

30   21

4,50±0,10
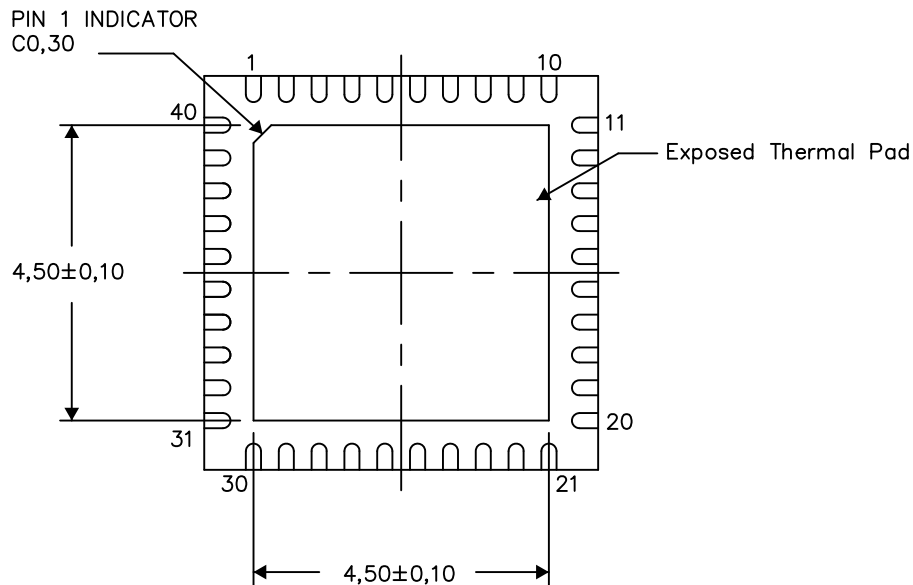
Bottom View

Exposed Thermal Pad Dimensions

4206355−4/X 08/14

NOTES: A. All linear dimensions are in millimeters

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RHA (S−PVQFN−N40)   PLASTIC QUAD FLATPACK NO−LEAD

Example Board Layout

Note D

36x0,5

5,1 6,8

5,1

6,8

Non Solder Mask Defined Pad

R0,14

0,08

0,85

0,07

0,28

All Around

Example
Solder Mask Opening
(Note F)

Pad Geometry
(Note C)

Example Stencil Design
0,125 Thick Stencil
(Note E)

0,8

0,230

36x0,5

0,3

0,3

5,15 6,75

9x1,2

9x1,2

5,15

6,75

(64% Printed Solder Coverage by Area)

Example Via Layout Design
Via layout may vary depending
on layout constraints
(Note D, F)

16xØ0,3

4,5

4,5

12x1,0

12x1,0

4207627−4/R   08/14

NOTES: A.   All linear dimensions are in millimeters.
          B.   This drawing is subject to change without notice.
          C.   Publication IPC−7351 is recommended for alternate designs.
          D.   This package is designed to be soldered to a thermal pad on the board.   Refer to Application Note, Quad Flat−Pack
                Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets
                for specific thermal information, via requirements, and recommended board layout.   These documents are available at
                www.ti.com <http://www.ti.com>.
          E.   Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release.   Customers should
                contact their board assembly site for stencil design recommendations.   Refer to IPC 7525 for stencil design considerations.
          F.   Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting
                recommendations for vias placed in the thermal pad.

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