

## FEATURES

- ±4000 V HBM ESD**
- Ideal for current shunt applications**
- High common-mode voltage range**
  - 4 V to +65 V operating
  - 40 V to +80 V survival
- 3 MHz bandwidth**
- <100 ns output propagation delay**
- Gain: 3 V/V**
- Wide operating temperature range**
  - Die: –40°C to +150°C
  - 8-lead SOIC: –40°C to +125°C
- Adjustable output offset**
- Available in 8-lead SOIC**
- Excellent ac and dc performance**
  - 10  $\mu\text{V}/^\circ\text{C}$  offset drift
  - 10 ppm/ $^\circ\text{C}$  gain drift
- Qualified for automotive applications**

## APPLICATIONS

- High-side current sensing in**
  - DC-to-dc converters
  - Motor controls
  - Transmission controls
  - Diesel-injection controls
  - Engine management
  - Suspension controls
  - Vehicle dynamic controls

## GENERAL DESCRIPTION

The **AD8216** is a single-supply, difference amplifier ideal for amplifying small differential voltages in the presence of large common-mode voltages. The operating input common-mode voltage range extends from –4 V to +65 V. The typical supply voltage is 5 V. The **AD8216** features a 3 MHz bandwidth, allowing for the input-to-output propagation delay that is always less than 150 ns. This feature is ideal for applications monitoring rapidly increasing and decreasing load currents.

The **AD8216** is offered in a SOIC package. The operating temperature range is –40°C to +125°C.

## FUNCTIONAL BLOCK DIAGRAM

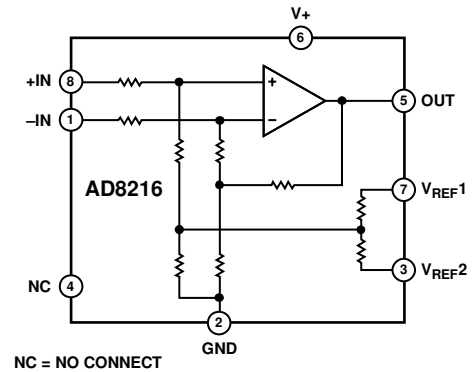


Figure 1.

Excellent ac and dc performance over temperature keep errors in the measurement loop to a minimum. Offset and gain drift are guaranteed to a maximum of 20  $\mu\text{V}/^\circ\text{C}$  and 15 ppm/ $^\circ\text{C}$ , respectively.

The output offset can be adjusted from 0.06 V to 4.9 V with a 5 V supply by using the  $V_{\text{REF1}}$  pin and the  $V_{\text{REF2}}$  pin. With the  $V_{\text{REF1}}$  pin attached to the  $V_+$  pin and the  $V_{\text{REF2}}$  pin attached to the GND pin, the output is set at half scale. Attaching both  $V_{\text{REF1}}$  and  $V_{\text{REF2}}$  to GND causes the output to be unipolar, starting near ground. Attaching both  $V_{\text{REF1}}$  and  $V_{\text{REF2}}$  to  $V_+$  causes the output to be unipolar, starting near  $V_+$ . Other offsets can be obtained by applying an external voltage to  $V_{\text{REF1}}$  and  $V_{\text{REF2}}$ .

Rev. C

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## REVISION HISTORY

### 9/15—Rev. B to Rev. C

|                           |    |
|---------------------------|----|
| Changes to Figure 23..... | 10 |
|---------------------------|----|

### 10/11—Rev. A to Rev. B

|   |    |
|---|----|
| Changes to Ordering Guide .....         | 14 |
| Added Automotive Products Section ..... | 14 |

### 9/08—Rev. 0 to Rev. A

|                                       |    |
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| Changes to Features Section.....      | 1  |
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### 11/07—Revision 0: Initial Version

## SPECIFICATIONS

$T_A$  = operating temperature range,  $V_S = 5$  V, unless otherwise noted.

Table 1.

| Parameter                          | Test Conditions/Comments                                     | Min   | Typ  | Max   | Unit   |
|------------------------------------|--|-------|------|-------|--------|
| <b>GAIN</b>                        |  |       |      |       |        |
| Initial                            |  |       | 3    |       | V/V    |
| Accuracy                           | $V_{OUT} \geq 0.1$ V dc, 25°C                                |       |      | ±0.25 | %      |
| Accuracy Over Temperature          | Specified temperature range                                  |       |      | ±0.4  | %      |
| Gain vs. Temperature               |  |       | 10   | 15    | ppm/°C |
| <b>VOLTAGE OFFSET</b>              |  |       |      |       |        |
| Offset Voltage, RTI                | 25°C   |       | ±0.5 |       | mV     |
| Over Temperature, RTI              | Specified temperature range                                  |       |      | ±3    | mV     |
| Offset Drift                       |  |       | ±10  | ±20   | µV/°C  |
| <b>INPUT</b>                       |  |       |      |       |        |
| Input Impedance                    |  |       |      |       |        |
| Differential                       |  |       | 400  |       | kΩ     |
| Common Mode                        |  |       | 200  |       | kΩ     |
| Input Voltage Range                | Common mode, continuous                                      | −4    |      | +65   | V      |
|                                    | Differential, $V_{REF1}$ and $V_{REF2}$ tied to GND          |       | 1.6  |       | V      |
|                                    | Differential, $V_{REF1}$ @ GND and $V_{REF2}$ @ 5 V          | −800  |      | +800  | mV     |
| Common-Mode Rejection              | 25°C, $f =$ dc to 20 kHz <sup>1</sup>                        | 80    | 90   |       | dB     |
|                                    | Operating temperature range, $f =$ dc to 20 kHz <sup>1</sup> | 80    | 90   |       | dB     |
| <b>OUTPUT</b>                      |  |       |      |       |        |
| Output Voltage Range               | $R_L = 25$ kΩ  | 0.06  |      | 4.9   | V      |
| Output Resistance                  |  |       | 200  |       | Ω      |
| <b>DYNAMIC RESPONSE</b>            |  |       |      |       |        |
| Small Signal −3 dB Bandwidth       |  |       | 3    |       | MHz    |
| Slew Rate                          |  |       | 15   |       | V/µs   |
| Propagation Delay                  | Input-to-output response                                     |       | 100  | 150   | ns     |
| <b>NOISE</b>                       |  |       |      |       |        |
| 0.1 Hz to 10 Hz, RTI               |  |       | 20   |       | µV p-p |
| Spectral Density, 1 kHz, RTI       |  |       | 0.5  |       | µV/√Hz |
| <b>OFFSET ADJUSTMENT</b>           |  |       |      |       |        |
| Ratiometric Accuracy <sup>2</sup>  | Divider to supplies  | 0.499 |      | 0.501 | V/V    |
| Accuracy, RTO                      | Voltage applied to $V_{REF1}$ and $V_{REF2}$ in parallel     |       |      | ±0.5  | mV/V   |
| Output Offset Adjustment Range     | $V_S = 5$ V  | 0.06  |      | 4.9   | V      |
| VREF Input Voltage Range           |  | 0.0   |      | $V_S$ | V      |
| VREF Divider Resistor Values       |  | 24    | 32   | 40    | kΩ     |
| <b>POWER SUPPLY</b>                |  |       |      |       |        |
| Operating Range                    |  | 4.5   |      | 5.5   | V      |
| Quiescent Current Over Temperature | $V_{OUT} = 0.1$ V dc   |       | 1    | 2     | mA     |
| Power Supply Rejection Ratio       |  | 70    |      |       | dB     |
| <b>TEMPERATURE RANGE</b>           |  |       |      |       |        |
| For Specified Performance          | Operating temperature range                                  | −40   |      | +125  | °C     |

<sup>1</sup> Source imbalance < 2 Ω.

<sup>2</sup> The offset adjustment is ratiometric to the power supply when  $V_{REF1}$  and  $V_{REF2}$  are used as a divider between the supplies.

## ABSOLUTE MAXIMUM RATINGS

Table 2.

| Parameter                             | Rating          |
|---------------------------------------|-----------------|
| Supply Voltage                        | 12.5 V          |
| Continuous Input Common-Mode Voltage  | -40 V to +80 V  |
| Continuous Input Differential Voltage | 6 V             |
| Reverse Supply Voltage                | 0.3 V           |
| ESD Rating                            |                 |
| HBM (Human Body Model)                | ±4000 V         |
| CDM (Charged Device Model)            | ±1000 V         |
| Operating Temperature Range           | -40°C to +125°C |
| Storage Temperature Range             | -65°C to +150°C |
| Output Short-Circuit Duration         | Indefinite      |

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

### ESD CAUTION



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

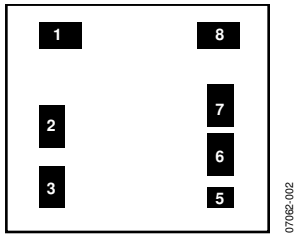


Figure 2. Metallization Diagram

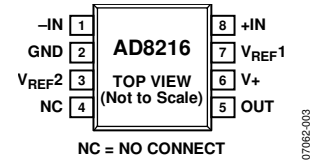


Figure 3. Pin Configuration

Table 3. Pin Function Descriptions

| Pin No. | Mnemonic          | X    | Y    |
|---------|-------------------|------|------|
| 1       | -IN               | -320 | +390 |
| 2       | GND               | -357 | +14  |
| 3       | V <sub>REF2</sub> | -349 | -201 |
| 4       | NC                | NC   | NC   |
| 5       | OUT               | +348 | -325 |
| 6       | V+                | +349 | -194 |
| 7       | V <sub>REF1</sub> | +349 | -26  |
| 8       | +IN               | +318 | +390 |

Die size is 1100  $\mu\text{m}$  by 1035  $\mu\text{m}$ .

Die thickness is 13 mil.

Minimum passivation opening (minimum bond pad size) is 92  $\mu\text{m}$   $\times$  92  $\mu\text{m}$ .

Passivation type is 8 kA USG (Oxide) + 10 kA Oxynitride.

Bond pad metal composition is 98.5% Al, 1% Si, and 0.5% Cu.

Backside potential is V+.

TYPICAL PERFORMANCE CHARACTERISTICS

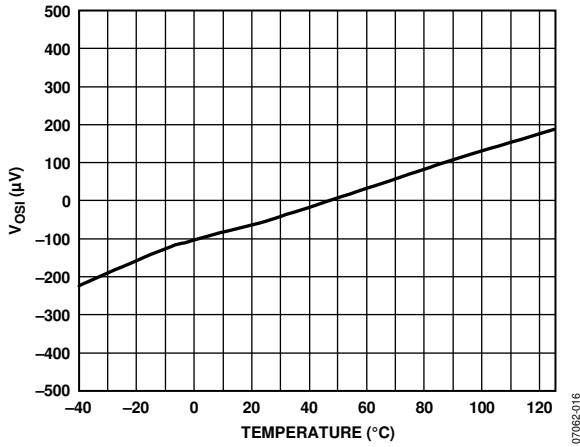


Figure 4. Typical Offset Drift

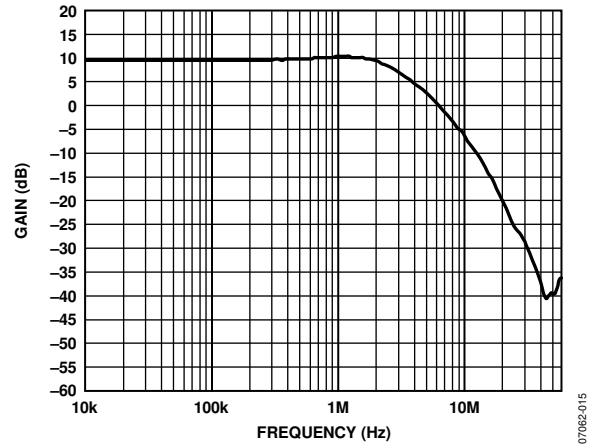


Figure 7. Typical Small Signal Bandwidth ( $V_{OUT} = 200\text{ mV p-p}$ )

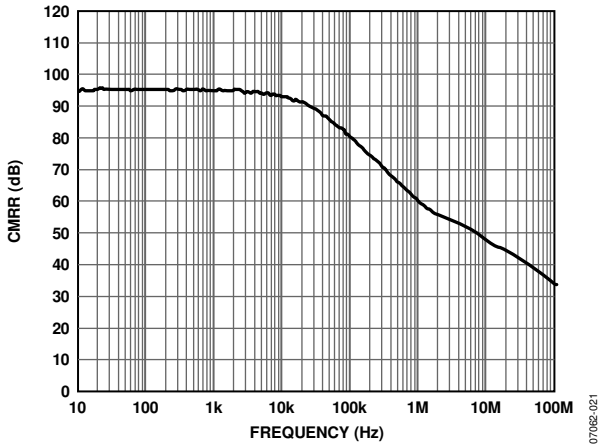


Figure 5. CMRR vs. Frequency

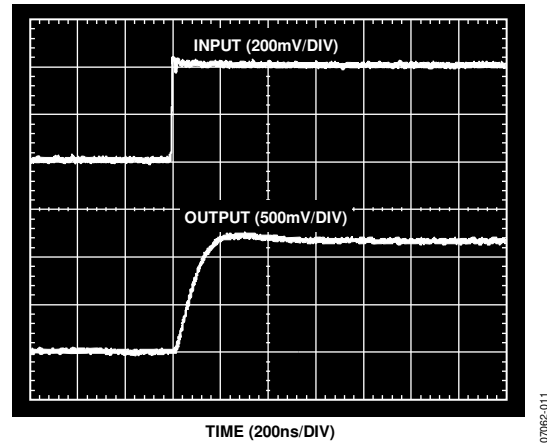


Figure 8. Rise Time

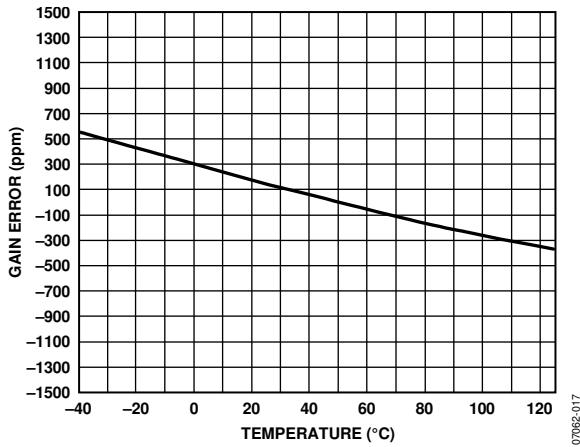


Figure 6. Typical Gain Drift

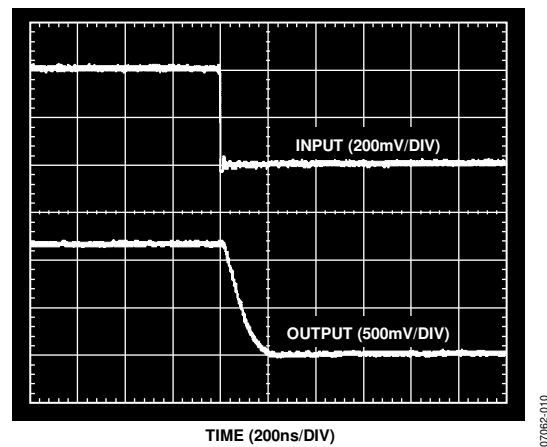


Figure 9. Fall Time

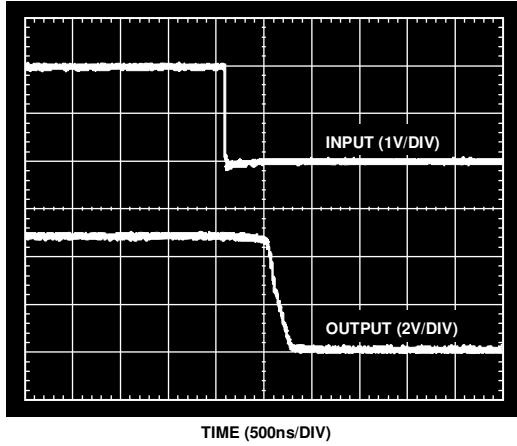


Figure 10. Differential Overload Recovery (Falling)

07062-008

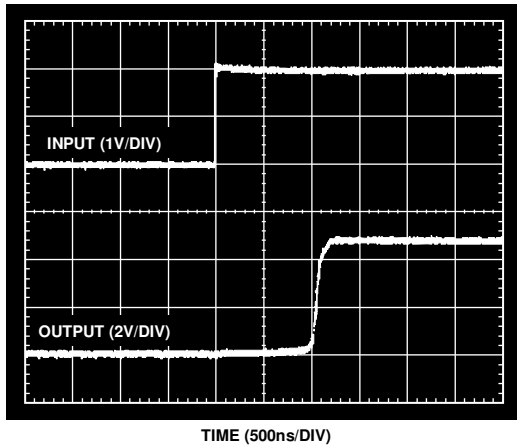


Figure 11. Differential Overload Recovery (Rising)

07062-009

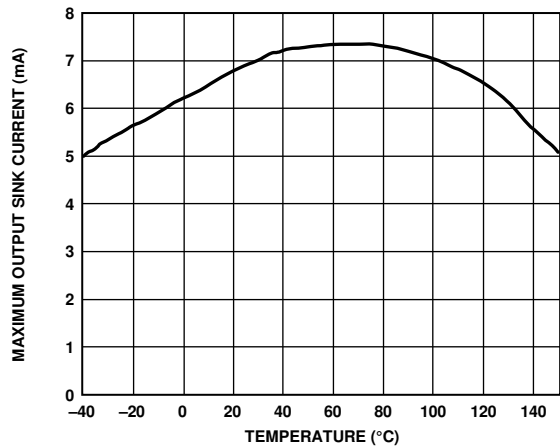


Figure 12. Maximum Output Sink Current vs. Temperature

07062-020

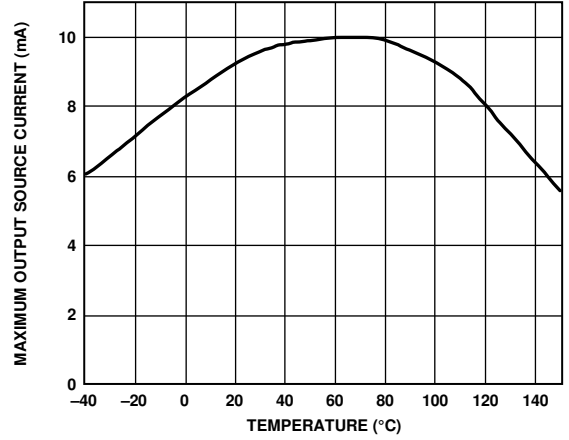


Figure 13. Maximum Output Source Current vs. Temperature

07062-019

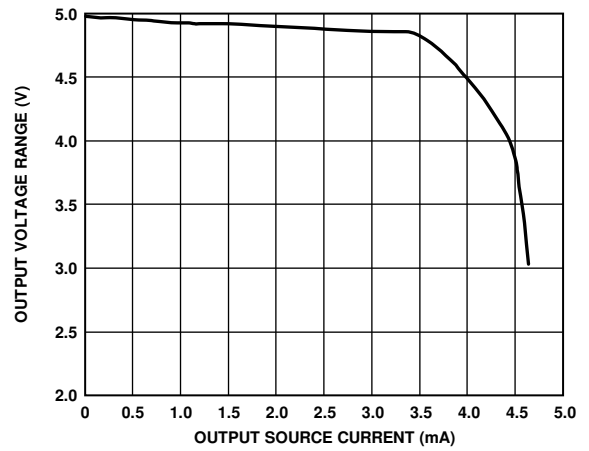


Figure 14. Output Voltage Range vs. Output Source Current

07062-007

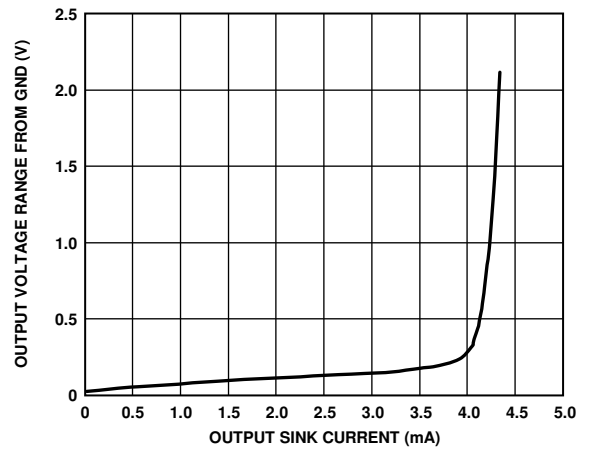


Figure 15. Output Voltage Range from GND vs. Output Sink Current

07062-006

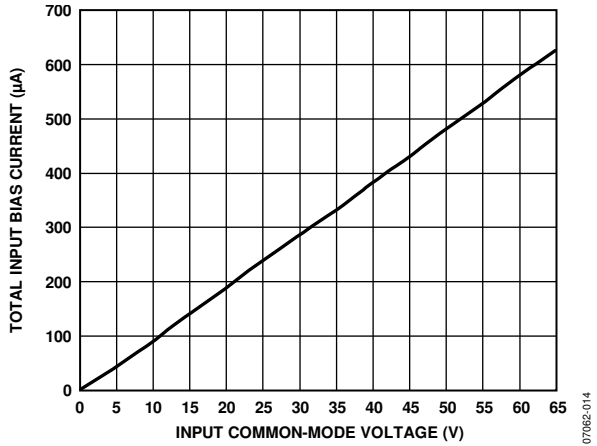


Figure 16. Total Input Bias Current vs. Input Common-Mode Voltage

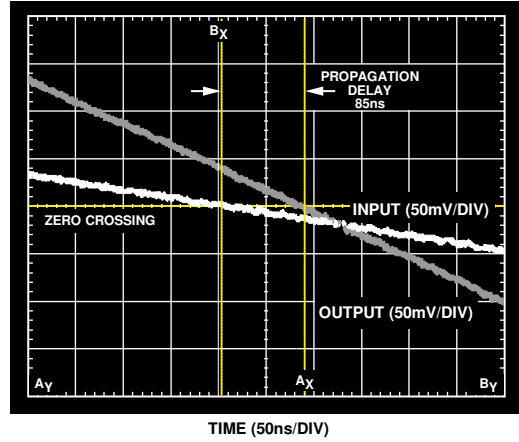


Figure 19. Propagation Delay (Falling)

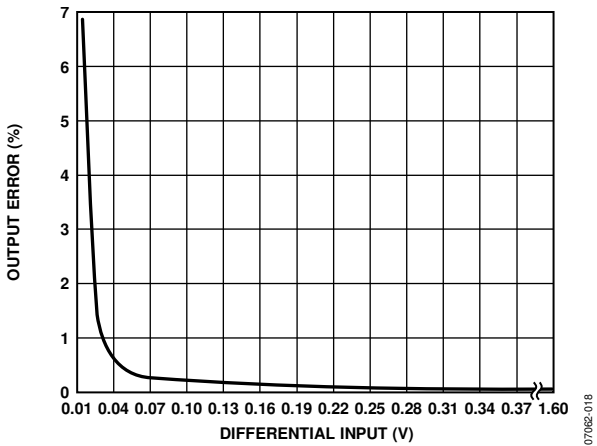


Figure 17. Output Error (%) vs. Differential Input Voltage (Unidirectional Operation ( $V_{REF1}$  and  $V_{REF2}$  Connected to GND))

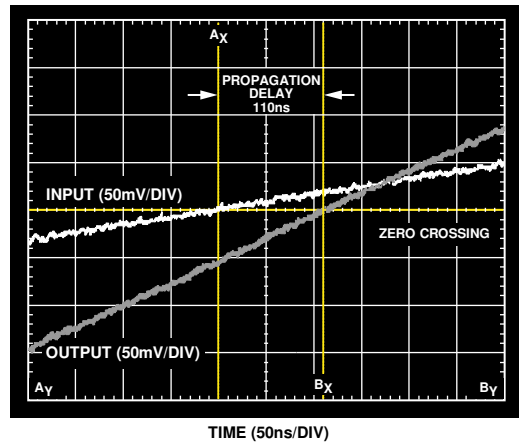


Figure 20. Propagation Delay (Rising)

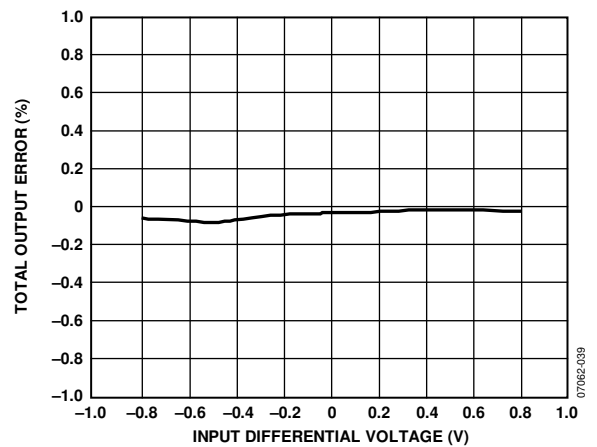


Figure 18. Output Error (%) vs. Differential Input Voltage (Bidirectional Operation ( $V_{REF1}$  and  $V_{REF2}$  Connected to 2.5 V))

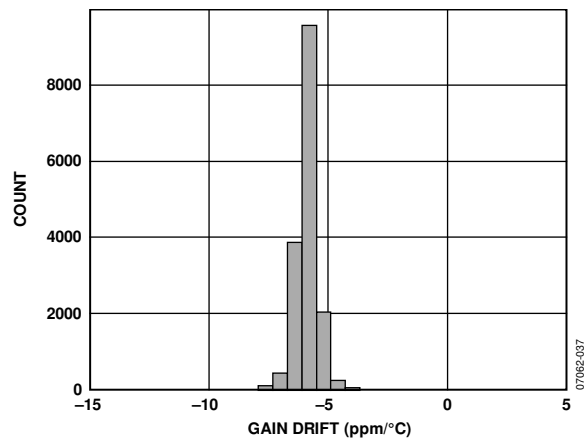


Figure 21. Gain Drift Distribution



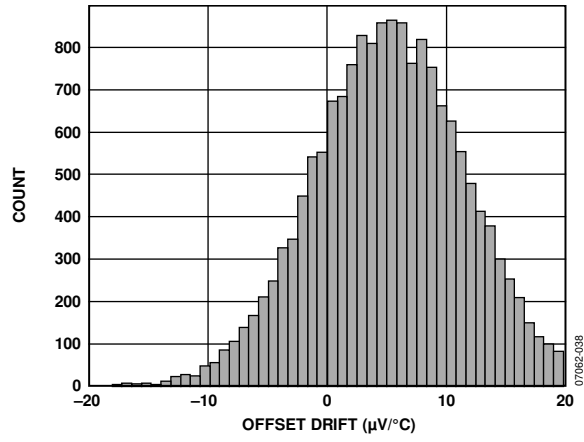


Figure 22. Offset Drift Distribution

## THEORY OF OPERATION

The AD8216 is a single-supply difference amplifier typically used to accurately amplify a small differential current shunt voltage in the presence of a rapidly changing common-mode voltage.

The AD8216 consists of an amplifier (A1), a resistor network, a small voltage reference, and a bias circuit (not shown), see Figure 23.

The set of input attenuators preceding A1 consist of  $R_A$ ,  $R_B$ , and  $R_C$ , which feature a combined series resistance of approximately  $200\text{ k}\Omega \pm 20\%$ . The purpose of these resistors is to attenuate the input voltage to match the input voltage range of A1. This balanced resistor network attenuates the common-mode signal by a ratio of  $1/18$ . A1 amplifier inputs are held within the power supply range, even if Pin 1 and Pin 8 exceed the supply or fall below common (ground). A reference voltage of  $250\text{ mV}$  biases the attenuator above ground, which allows the amplifier to operate in the presence of negative common-mode voltages.

The input resistor network also attenuates normal (differential) mode voltages. Therefore, Amplifier A1 features a gain of  $54\text{ V/V}$  to provide a total system gain of  $3\text{ V/V}$ .

$$\text{Total Gain (V/V)} = 1/18 (\text{V/V}) \times 54 (\text{V/V}) = 3 \text{ V/V}$$

The reference inputs,  $V_{\text{REF1}}$  and  $V_{\text{REF2}}$ , are tied through resistors to the positive input of A1, which allows the output offset to be adjusted anywhere in the output operating range. The gain is  $1\text{ V/V}$  from the reference pins to the output when the reference pins are used in parallel. When they are used to divide the supply, the gain is  $0.5\text{ V/V}$ .

The ratios of  $R_A$ ,  $R_B$ ,  $R_C$ , and  $R_F$  are trimmed to a high level of precision to allow the CMRR to exceed  $80\text{ dB}$ . This performance is accomplished by laser trimming the resistor ratio matching to better than  $0.01\%$ .

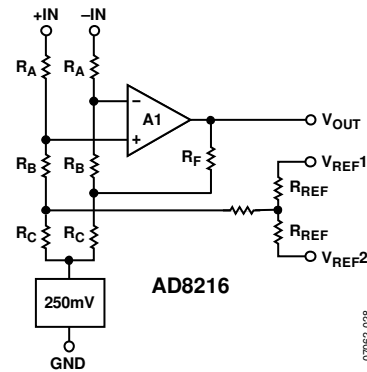


Figure 23. Simplified Schematic

## OUTPUT OFFSET ADJUSTMENT

The output of the AD8216 can be adjusted for unidirectional or bidirectional operation.

### UNIDIRECTIONAL OPERATION

Unidirectional operation allows the AD8216 to measure currents through a resistive shunt in one direction. The basic modes for unidirectional operation are ground referenced output mode and V+ referenced output mode.

For unidirectional operation, the output can be set at the negative rail (near ground) or at the positive rail (near V+) when the differential input is 0 V. The output moves to the opposite rail when a correct polarity differential input voltage is applied. In this case, full scale is approximately 1.6 V. The required polarity of the differential input depends on the output voltage setting. If the output is set at the positive rail, the input polarity needs to be negative to move the output down. If the output is set at ground, the polarity is positive to move the output up.

### GROUND REFERENCED OUTPUT

When using the AD8216 in this mode, both referenced inputs are tied to ground, which causes the output to sit at the negative rail when there are zero differential volts at the input (see Figure 24).

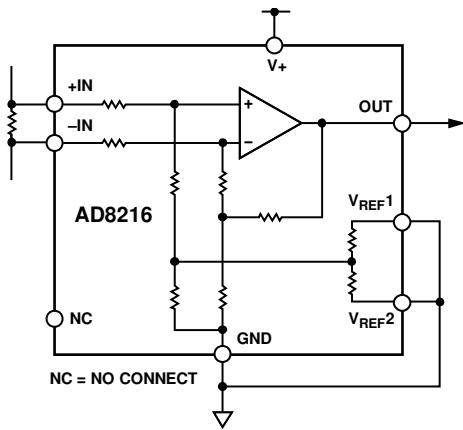


Figure 24. Ground Referenced Output

Table 4. V+ = 5 V

| V <sub>IN</sub> (Referred to -IN) | V <sub>OUT</sub> |
|-----------------------------------|------------------|
| 0 V                               | 0.1 V            |
| 1.6 V                             | 4.8 V            |

### V+ REFERENCED OUTPUT

The V+ referenced output mode is set when both reference pins are tied to the positive supply. It is typically used when the diagnostic scheme requires detection of the amplifier and the wiring before power is applied to the load (see Figure 25).

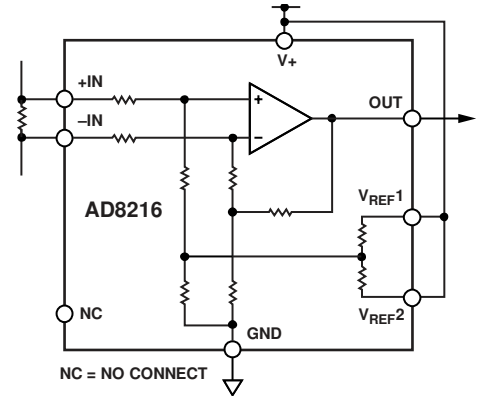


Figure 25. V+ Referenced Output

Table 5. V+ = 5 V

| V <sub>IN</sub> (Referred to -IN) | V <sub>OUT</sub> |
|-----------------------------------|------------------|
| 0 V                               | 4.9 V            |
| -1.6 V                            | 0.1 V            |

### BIDIRECTIONAL OPERATION

Bidirectional operation allows the AD8216 to measure currents through a resistive shunt in two directions.

In this case, the output is set anywhere within the output range. Typically, it is set at half scale for equal range in both directions. In some cases, however, it is set at a voltage other than half scale when the bidirectional current is nonsymmetrical.

Table 6. V+ = 5 V, V<sub>OUT</sub> = 2.5 V with V<sub>IN</sub> = 0 V

| V <sub>IN</sub> (Referred to -IN) | V <sub>OUT</sub> |
|-----------------------------------|------------------|
| +800 mV                           | 4.9 V            |
| -800 mV                           | 0.1 V            |

Adjusting the output is accomplished by applying voltage(s) to the referenced inputs.

V<sub>REF1</sub> and V<sub>REF2</sub> are tied to internal resistors that connect to an internal offset node. There is no operational difference between the pins.

**EXTERNAL REFERENCED OUTPUT**

Tying both pins together and to a reference produces an output equal to the reference voltage when there is no differential input (see Figure 26). The output moves down from the reference voltage when the input is negative, relative to the  $-IN$  pin, and up when the input is positive, relative to the  $-IN$  pin.

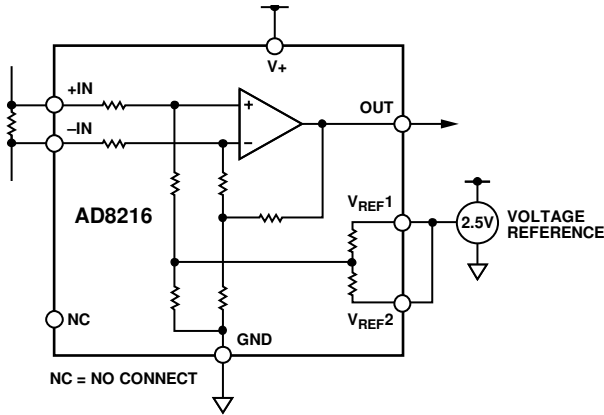


Figure 26. External Referenced Output

**SPLITTING THE SUPPLY**

By tying one reference pin to  $V+$  and the other to the ground pin, the output is set at half of the supply when there is no differential input (see Figure 27). The benefit is that no external reference is required to offset the output for bidirectional current measurement. This configuration creates a midscale offset that is ratiometric to the supply, which means that if the supply increases or decreases, the output remains at half the supply. For example, if the supply is 5.0 V, the output is at half scale or 2.5 V. If the supply increases by 10% (to 5.5 V), the output goes to 2.75 V.

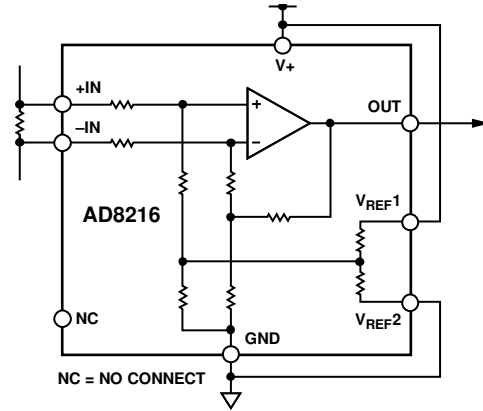


Figure 27. Split Supply

**SPLITTING AN EXTERNAL REFERENCE**

In this case, an external reference is divided by 2 with an accuracy of approximately 0.5% by connecting one  $V_{REF}$  pin to ground and the other  $V_{REF}$  pin to the reference (see Figure 28).

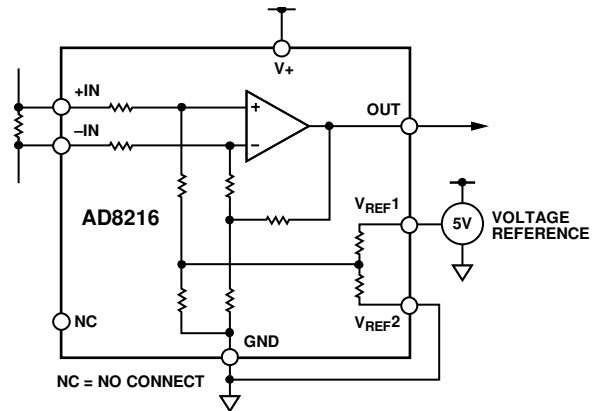


Figure 28. Split External Reference

## APPLICATIONS INFORMATION

A typical application for the AD8216 is high-side measurement of a current through a solenoid for PWM control of the solenoid opening. Typical applications include dc-to-dc converters, motor controls, and solenoid controls.

Two typical circuit configurations are used for this type of application.

### HIGH-SIDE CURRENT SENSE WITH A LOW-SIDE SWITCH

In this case, the PWM control switch is ground referenced. An inductive load (solenoid) is tied to a power supply. A resistive shunt is placed between the switch and the load (see Figure 29). An advantage of placing the shunt on the high side is that the entire current, including the recirculation current, can be measured because the shunt remains in the loop when the switch is off. In addition, diagnostics can be enhanced because shorts to ground can be detected with the shunt on the high side.

In this circuit configuration, when the switch is closed, the common-mode voltage moves down to near the negative rail. When the switch is opened, the voltage reversal across the inductive load causes the common-mode voltage to be held one diode drop above the battery by the clamp diode.

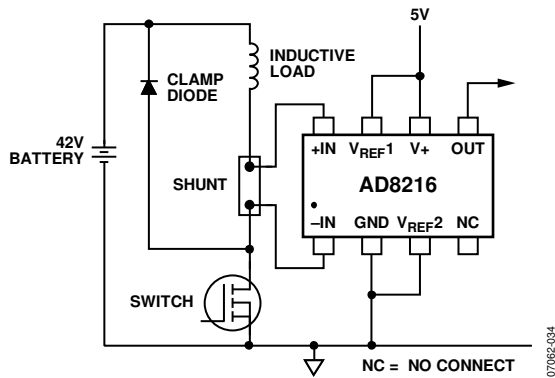


Figure 29. Low-Side Switch

### HIGH-SIDE CURRENT SENSE WITH A HIGH-SIDE SWITCH

This configuration minimizes the possibility of unexpected solenoid activation and excessive corrosion (see Figure 30). In this case, both the switch and the shunt are on the high side. When the switch is off, it removes the battery from the load, which prevents damage from potential shorts to ground, while still allowing the recirculating current to be measured and providing for diagnostics. Removing the power supply from the load for the majority of the time minimizes the corrosive effects that could be caused by the differential voltage between the load and ground.

When using a high-side switch, the battery voltage is connected to the load when the switch is closed, causing the common-mode voltage to increase to the battery voltage. In this case, when the switch is opened, the voltage reversal across the inductive load causes the common-mode voltage to be held one diode drop below ground by the clamp diode.

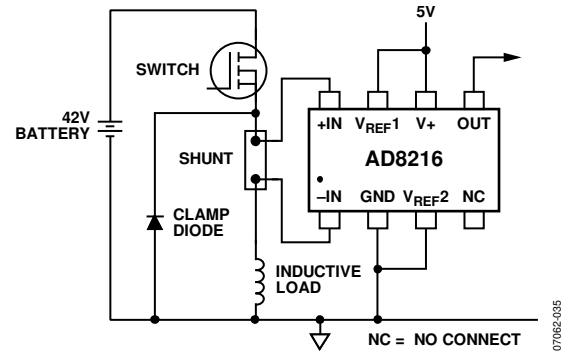


Figure 30. High-Side Switch

Another typical application for the AD8216 is as part of the control loop in H-bridge motor controls. In this case, the AD8216 is placed in the middle of the H-bridge so that it can accurately measure current in both directions by using the shunt available at the motor (see Figure 31). This is a better solution than a ground referenced op amp because ground is not typically a stable reference voltage in this type of application. This instability in the ground reference causes the measurements that could be made with a simple ground referenced op amp to be inaccurate.

The AD8216 measures current in both directions as the H-bridge switches and the motor changes direction. The output of the AD8216 is configured in an external reference bidirectional mode (see the Output Offset Adjustment section).

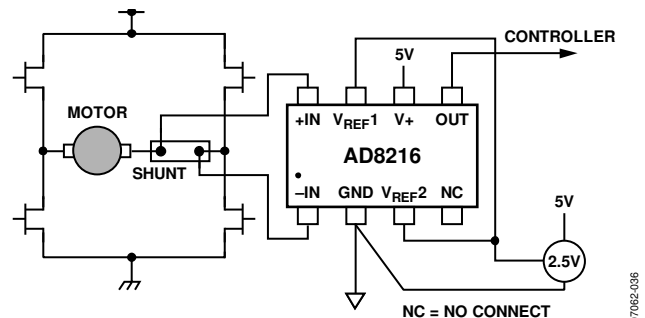
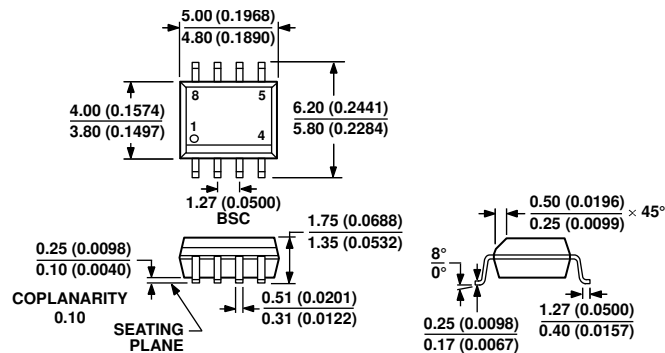


Figure 31. Motor Control Application

## OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MS-012-AA  
 CONTROLLING DIMENSIONS ARE IN MILLIMETERS; INCH DIMENSIONS  
 (IN PARENTHESES) ARE ROUNDED-OFF MILLIMETER EQUIVALENTS FOR  
 REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.

012807-A

Figure 32. 8-Lead Standard Small Outline Package [SOIC\_N]  
 Narrow Body (R-8)  
 Dimensions shown in millimeters and (inches)

## ORDERING GUIDE

| Model <sup>1,2</sup> | Temperature Range | Package Description              | Package Option |
|----------------------|-------------------|----------------------------------|----------------|
| AD8216YRZ            | -40°C to +125°C   | 8-Lead SOIC_N                    | R-8            |
| AD8216YRZ-RL         | -40°C to +125°C   | 8-Lead SOIC_N, 13" Tape and Reel | R-8            |
| AD8216YRZ-R7         | -40°C to +125°C   | 8-Lead SOIC_N, 7" Tape and Reel  | R-8            |
| AD8216WYRZ           | -40°C to +125°C   | 8-Lead SOIC_N                    | R-8            |
| AD8216WYRZ-RL        | -40°C to +125°C   | 8-Lead SOIC_N, 13" Tape and Reel | R-8            |
| AD8216WYRZ-R7        | -40°C to +125°C   | 8-Lead SOIC_N, 7" Tape and Reel  | R-8            |

<sup>1</sup> Z = RoHS Compliant Part.

<sup>2</sup> W = Qualified for Automotive Applications.

## AUTOMOTIVE PRODUCTS

The [AD8216W](#) models are available with controlled manufacturing to support the quality and reliability requirements of automotive applications. Note that these automotive models may have specifications that differ from the commercial models; therefore, designers should review the Specifications section of this data sheet carefully. Only the automotive grade products shown are available for use in automotive applications. Contact your local Analog Devices account representative for specific product ordering information and to obtain the specific Automotive Reliability reports for these models.

**NOTES**

**NOTES**