

# 1-MHz, Micro-Power, Low-Noise, RRIO, 1.8-V CMOS OPERATIONAL AMPLIFIER Precision Value Line Series

Check for Samples: [OPA313](#), [OPA2313](#), [OPA4313](#)

## FEATURES

- **Low  $I_Q$ :** 50  $\mu\text{A}/\text{ch}$
- **Wide Supply Range:** 1.8 V to 5.5 V
- **Low Noise:** 25  $\text{nV}/\sqrt{\text{Hz}}$  at 1 kHz
- **Gain Bandwidth:** 1 MHz
- **Low Input Bias Current:** 0.2 pA
- **Low Offset Voltage:** 0.5 mV
- **Unity-Gain Stable**
- **Internal RF/EMI Filter**
- **Extended Temperature Range:**  
–40°C to +125°C

## APPLICATIONS

- **Battery-Powered Instruments:**
  - Consumer, Industrial, Medical
  - Notebooks, Portable Media Players
- **Sensor Signal Conditioning:**
  - Loop-Powered
  - Notebooks, Portable Media Players
- **Wireless Sensors:**
  - Home Security
  - Remote Sensing
  - Wireless Metering

## DESCRIPTION

The OPA313 family of single-, dual-, and quad-channel op amps represents a new generation of low-cost, general purpose, micro-power operational amplifiers. Featuring rail-to-rail input and output swings, and low quiescent current (50  $\mu\text{A}$ , typ) combined with a wide bandwidth of 1 MHz and very low noise (25  $\text{nV}/\sqrt{\text{Hz}}$  at 1 kHz) makes this family very attractive for a variety of battery-powered applications that require a good balance between cost and performance. The low input bias current supports those op amps to be used in applications with megaohm source impedances.

The robust design of the OPA313 devices provides ease-of-use to the circuit designer: unity-gain stability with capacitive loads of up to 150 pF, integrated RF/EMI rejection filter, no phase reversal in overdrive conditions, and high electrostatic discharge (ESD) protection (4-kV HBM).

These devices are optimized for operation at voltages as low as +1.8 V ( $\pm 0.9$  V) and up to +5.5 V ( $\pm 2.75$  V), and are specified over the extended temperature range of –40°C to +125°C.

The OPA313 (single) is available in both SC70-5 and SOT23-5 packages. The OPA2313 (dual) is offered in SO-8, MSOP-8, and DFN-8 packages. The quad-channel OPA4313 is offered in a TSSOP-14 package.



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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### PACKAGE INFORMATION<sup>(1)</sup>

PRODUCT	PACKAGE-LEAD	PACKAGE DESIGNATOR	SPECIFIED TEMPERATURE RANGE	PACKAGE MARKING
OPA313	SC70-5	DCK	–40°C to +125°C	SIE
	SOT23-5	DBV	–40°C to +125°C	SIF
OPA2313	SO-8	D	–40°C to +125°C	OP2313
	MSOP-8	DGK	–40°C to +125°C	OUSS
	DFN-8	DRG	–40°C to +125°C	SDY
OPA4313	TSSOP-14	PW	–40°C to +125°C	OPA4313

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or visit the device product folder at [www.ti.com](http://www.ti.com).

### ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>

Over operating free-air temperature range, unless otherwise noted.

		OPA313, OPA2313, OPA4313	UNIT
Supply voltage		7	V
Signal input terminals	Voltage <sup>(2)</sup>	(V–) – 0.5 to (V+) + 0.5	V
	Current <sup>(2)</sup>	±10	mA
Output short-circuit <sup>(3)</sup>		Continuous	mA
Operating temperature, T <sub>A</sub>		–40 to +150	°C
Storage temperature, T <sub>stg</sub>		–65 to +150	°C
Junction temperature, T <sub>J</sub>		+150	°C
ESD rating	Human body model (HBM)	4000	V
	Charged device model (CDM)	1000	V
	Machine model (MM)	200	V

- (1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not supported.
- (2) Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5 V beyond the supply rails should be current limited to 10 mA or less.
- (3) Short-circuit to ground, one amplifier per package.

**ELECTRICAL CHARACTERISTICS: +5.5 V<sup>(1)</sup>**

 At  $T_A = +25^\circ\text{C}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S / 2$ , and  $V_{CM} = V_{OUT} = V_S / 2$ , unless otherwise noted.

PARAMETER		TEST CONDITIONS	OPA313, OPA2313, OPA4313			UNIT
			MIN	TYP	MAX	
<b>OFFSET VOLTAGE</b>						
$V_{OS}$	Input offset voltage			0.5	2.5	mV
$dV_{OS}/dT$	vs Temperature	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$		2		$\mu\text{V}/^\circ\text{C}$
PSRR	vs power supply	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	74	90		dB
	Channel separation, dc	At dc		10		$\mu\text{V}/\text{V}$
<b>INPUT VOLTAGE RANGE</b>						
$V_{CM}$	Common-mode voltage range	No phase reversal, rail-to-rail input	$(V_-) - 0.2$		$(V_+) + 0.2$	V
CMRR	Common-mode rejection ratio	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$ , $(V_{S-}) - 0.2\text{ V} < V_{CM} < (V_{S+}) - 1.3\text{ V}$	70	85		dB
		$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$ , $V_{CM} = -0.2\text{ V}$ to $5.7\text{ V}$	64	80		dB
<b>INPUT BIAS CURRENT</b>						
$I_B$	Input bias current			$\pm 0.2$	$\pm 10$	pA
		$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}^{(2)}$			$\pm 50$	pA
		$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}^{(2)}$			$\pm 600$	pA
$I_{OS}$	Input offset current			$\pm 0.2$	$\pm 10$	pA
		$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}^{(2)}$			$\pm 50$	pA
		$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}^{(2)}$			$\pm 600$	pA
<b>NOISE</b>						
	Input voltage noise (peak-to-peak)	$f = 0.1\text{ Hz}$ to $10\text{ Hz}$		6		$\mu\text{V}_{PP}$
$e_n$	Input voltage noise density	$f = 10\text{ kHz}$		22		$\text{nV}/\sqrt{\text{Hz}}$
		$f = 1\text{ kHz}$		25		$\text{nV}/\sqrt{\text{Hz}}$
$i_n$	Input current noise density	$f = 1\text{ kHz}$		5		$\text{fA}/\sqrt{\text{Hz}}$
<b>INPUT CAPACITANCE</b>						
$C_{IN}$	Differential			1		pF
	Common-mode			5		pF
<b>OPEN-LOOP GAIN</b>						
$A_{OL}$	Open-loop voltage gain	$0.05\text{ V} < V_O < (V_+) - 0.05\text{ V}$ , $R_L = 100\text{ k}\Omega$	90	104		dB
		$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$ , $0.1\text{ V} < V_O < (V_+) - 0.1\text{ V}$	104	116		dB
		$0.3\text{ V} < V_O < (V_+) - 0.3\text{ V}$ , $R_L = 2\text{ k}\Omega$	100	110		dB
	Phase margin	$V_S = 5.0\text{ V}$ , $G = +1$		65		degrees

- (1) Parameters with minimum or maximum specification limits are 100% production tested at  $+25^\circ\text{C}$ , unless otherwise noted. Over temperature limits are based on characterization and statistical analysis.
- (2) Specified by design and characterization; not production tested.

**ELECTRICAL CHARACTERISTICS: +5.5 V<sup>(1)</sup> (continued)**

At  $T_A = +25\text{ }^\circ\text{C}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S / 2$ , and  $V_{CM} = V_{OUT} = V_S / 2$ , unless otherwise noted.

PARAMETER		TEST CONDITIONS	OPA313, OPA2313, OPA4313			UNIT
			MIN	TYP	MAX	
<b>FREQUENCY RESPONSE</b>						
GBW	Gain-bandwidth product	$V_S = 5.0\text{ V}$ , $C_L = 10\text{ pF}$		1		MHz
SR	Slew rate	$V_S = 5.0\text{ V}$ , $G = +1$		0.5		V/ $\mu\text{s}$
$t_s$	Settling time	To 0.1%, $V_S = 5.0\text{ V}$ , 2-V step, $G = +1$		5		$\mu\text{s}$
		To 0.01%, $V_S = 5.0\text{ V}$ , 2-V step, $G = +1$		6		$\mu\text{s}$
	Overload recovery time	$V_S = 5.0\text{ V}$ , $V_{IN} \times \text{Gain} > V_S$		3		$\mu\text{s}$
THD+N	Total harmonic distortion + noise <sup>(3)</sup>	$V_S = 5.0\text{ V}$ , $V_O = 1\text{ V}_{RMS}$ , $G = +1$ , $f = 1\text{ kHz}$		0.0045%		
<b>OUTPUT</b>						
$V_O$	Voltage output swing from supply rails	$R_L = 100\text{ k}\Omega$ <sup>(4)</sup>		5	20	mV
		$T_A = -40\text{ }^\circ\text{C}$ to $+125\text{ }^\circ\text{C}$ , $R_L = 100\text{ k}\Omega$ <sup>(4)</sup>			30	mV
		$R_L = 2\text{ k}\Omega$ <sup>(4)</sup>		75	100	mV
		$T_A = -40\text{ }^\circ\text{C}$ to $+125\text{ }^\circ\text{C}$ , $R_L = 2\text{ k}\Omega$			125	mV
$I_{SC}$	Short-circuit current			$\pm 15$		mA
		$T_A = -40\text{ }^\circ\text{C}$ to $+125\text{ }^\circ\text{C}$		$\pm 12$		mA
$R_O$	Open-loop output impedance			2300		$\Omega$
<b>POWER SUPPLY</b>						
$V_S$	Specified voltage range		1.8 ( $\pm 0.9$ )	5.5 ( $\pm 2.75$ )		V
$I_Q$	Quiescent current per amplifier	$I_O = 0\text{ mA}$ , $V_S = 5.0\text{ V}$		50	60	$\mu\text{A}$
		$T_A = -40\text{ }^\circ\text{C}$ to $+125\text{ }^\circ\text{C}$ , $V_S = 5.0\text{ V}$ , $I_O = 0\text{ mA}$			85	$\mu\text{A}$
	Power-on time	$V_S = 0\text{ V}$ to $5\text{ V}$ , to 90% $I_Q$ level		10		$\mu\text{s}$
<b>TEMPERATURE</b>						
	Specified range		-40	+125		$^\circ\text{C}$
	Operating range		-40	+150		$^\circ\text{C}$
	Storage range		-65	+150		$^\circ\text{C}$

(3) Third-order filter; bandwidth = 80 kHz at -3 dB.  
 (4) Specified by design and characterization; not production tested.

**ELECTRICAL CHARACTERISTICS: +1.8 V<sup>(1)</sup>**

 At  $T_A = +25^\circ\text{C}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S / 2$ ,  $V_{CM} = V_{S+} - 1.3\text{ V}$ , and  $V_{OUT} = V_S / 2$ , unless otherwise noted.

PARAMETER		TEST CONDITIONS	OPA313, OPA2313, OPA4313			UNIT
			MIN	TYP	MAX	
<b>OFFSET VOLTAGE</b>						
$V_{OS}$	Input offset voltage			0.5	2.5	mV
$dV_{OS}/dT$	vs Temperature	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$		2		$\mu\text{V}/^\circ\text{C}$
PSRR	vs power supply	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	74	90		dB
	Channel separation, dc	At dc		10		$\mu\text{V}/\text{V}$
<b>INPUT VOLTAGE RANGE</b>						
$V_{CM}$	Common-mode voltage range	No phase reversal, rail-to-rail input	$(V_-) - 0.2$		$(V_+) + 0.2$	V
CMRR	Common-mode rejection ratio	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$ , $(V_{S-}) - 0.2\text{ V} < V_{CM} < (V_{S+}) - 1.3\text{ V}$	70	85		dB
		$V_S = 1.8\text{ V}$ , $V_{CM} = -0.2\text{ V}$ to $+1.8\text{ V}$	58	73		
		$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$ , $V_{CM} = -0.2\text{ V}$ to $1.6\text{ V}$	58	70		dB
<b>INPUT BIAS CURRENT</b>						
$I_B$	Input bias current			$\pm 0.2$	$\pm 10$	$\mu\text{A}$
		$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}^{(2)}$			$\pm 50$	$\mu\text{A}$
		$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}^{(2)}$			$\pm 600$	$\mu\text{A}$
$I_{OS}$	Input offset current			$\pm 0.2$	$\pm 10$	$\mu\text{A}$
		$T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}^{(2)}$			$\pm 50$	$\mu\text{A}$
		$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}^{(2)}$			$\pm 600$	$\mu\text{A}$
<b>NOISE</b>						
	Input voltage noise (peak-to-peak)	$f = 0.1\text{ Hz}$ to $10\text{ Hz}$		6		$\mu\text{V}_{PP}$
$e_n$	Input voltage noise density	$f = 10\text{ kHz}$		22		$\text{nV}/\sqrt{\text{Hz}}$
		$f = 1\text{ kHz}$		25		$\text{nV}/\sqrt{\text{Hz}}$
$i_n$	Input current noise density	$f = 1\text{ kHz}$		5		$\text{fA}/\sqrt{\text{Hz}}$
<b>INPUT CAPACITANCE</b>						
$C_{IN}$	Differential			1		pF
	Common-mode			5		pF
<b>OPEN-LOOP GAIN</b>						
$A_{OL}$	Open-loop voltage gain	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$ , $0.1\text{ V} < V_O < (V_+) - 0.1\text{ V}$	90	110		dB
		$0.05\text{ V} < V_O < (V_+) - 0.05\text{ V}$ , $R_L = 100\text{ k}\Omega$	100	110		dB

- (1) Parameters with minimum or maximum specification limits are 100% production tested at  $+25^\circ\text{C}$ , unless otherwise noted. Over temperature limits are based on characterization and statistical analysis.
- (2) Specified by design and characterization; not production tested.

**ELECTRICAL CHARACTERISTICS: +1.8 V<sup>(1)</sup> (continued)**

At  $T_A = +25\text{ }^\circ\text{C}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S / 2$ ,  $V_{CM} = V_{S+} - 1.3\text{ V}$ , and  $V_{OUT} = V_S / 2$ , unless otherwise noted.

PARAMETER		TEST CONDITIONS	OPA313, OPA2313, OPA4313			UNIT
			MIN	TYP	MAX	
<b>FREQUENCY RESPONSE</b>						
GBW	Gain-bandwidth product	$C_L = 10\text{ pF}$	0.9			MHz
SR	Slew rate	$G = +1$	0.45			V/ $\mu\text{s}$
$t_S$	Settling time	To 0.1%, $V_S = 5.0\text{ V}$ , 2-V step, $G = +1$	5			$\mu\text{s}$
		To 0.01%, $V_S = 5.0\text{ V}$ , 2-V step, $G = +1$	6			$\mu\text{s}$
	Overload recovery time	$V_S = 5.0\text{ V}$ , $V_{IN} \times \text{Gain} > V_S$	3			$\mu\text{s}$
THD+N	Total harmonic distortion + noise <sup>(3)</sup>	$V_S = 5.0\text{ V}$ , $V_O = 1\text{ V}_{RMS}$ , $G = +1$ , $f = 1\text{ kHz}$	0.0045%			
<b>OUTPUT</b>						
$V_O$	Voltage output swing from supply rails	$R_L = 100\text{ k}\Omega$ <sup>(4)</sup>	5		15	mV
		$T_A = -40\text{ }^\circ\text{C}$ to $+125\text{ }^\circ\text{C}$ , $R_L = 100\text{ k}\Omega$ <sup>(4)</sup>			30	mV
		$R_L = 2\text{ k}\Omega$ <sup>(4)</sup>	25		50	mV
		$T_A = -40\text{ }^\circ\text{C}$ to $+125\text{ }^\circ\text{C}$ , $R_L = 2\text{ k}\Omega$			125	mV
$I_{SC}$	Short-circuit current		$\pm 6$			mA
$R_O$	Open-loop output impedance		2300			$\Omega$
<b>POWER SUPPLY</b>						
$V_S$	Specified voltage range		1.8 ( $\pm 0.9$ )	5.5 ( $\pm 2.75$ )		V
$I_Q$	Quiescent current per amplifier	$I_O = 0\text{ mA}$	50		60	$\mu\text{A}$
	Power-on time	$V_S = 0\text{ V}$ to $5\text{ V}$ , to 90% $I_Q$ level	10			$\mu\text{s}$
<b>TEMPERATURE</b>						
	Specified range		-40	+125		$^\circ\text{C}$
	Operating range		-40	+150		$^\circ\text{C}$
	Storage range		-65	+150		$^\circ\text{C}$

(3) Third-order filter; bandwidth = 80 kHz at -3 dB.

(4) Specified by design and characterization; not production tested.

**THERMAL INFORMATION: OPA313**

THERMAL METRIC <sup>(1)</sup>		OPA313		UNITS
		DBV (SOT23)	DCK (SC70)	
		5 PINS	5 PINS	
$\theta_{JA}$	Junction-to-ambient thermal resistance	228.5	281.4	°C/W
$\theta_{JC(top)}$	Junction-to-case(top) thermal resistance	99.1	91.6	
$\theta_{JB}$	Junction-to-board thermal resistance	54.6	59.6	
$\Psi_{JT}$	Junction-to-top characterization parameter	7.7	1.5	
$\Psi_{JB}$	Junction-to-board characterization parameter	53.8	58.8	
$\theta_{JC(bottom)}$	Junction-to-case(bottom) thermal resistance	N/A	N/A	

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

**THERMAL INFORMATION: OPA2313**

THERMAL METRIC <sup>(1)</sup>		OPA2313			UNITS
		D (SO)	DGK (MSOP)	DRG (DFN)	
		8 PINS	8 PINS	8 PINS	
$\theta_{JA}$	Junction-to-ambient thermal resistance	138.4	191.2	53.8	°C/W
$\theta_{JC(top)}$	Junction-to-case(top) thermal resistance	89.5	61.9	69.2	
$\theta_{JB}$	Junction-to-board thermal resistance	78.6	111.9	20.1	
$\Psi_{JT}$	Junction-to-top characterization parameter	29.9	5.1	3.8	
$\Psi_{JB}$	Junction-to-board characterization parameter	78.1	110.2	20.0	
$\theta_{JC(bottom)}$	Junction-to-case(bottom) thermal resistance	N/A	N/A	11.6	

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

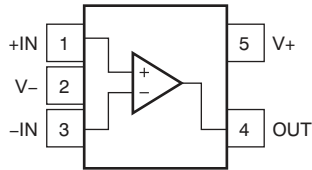
**THERMAL INFORMATION: OPA4313**

THERMAL METRIC <sup>(1)</sup>		OPA4313	UNITS
		PW (TSSOP)	
		14 PINS	
$\theta_{JA}$	Junction-to-ambient thermal resistance	121.0	°C/W
$\theta_{JC(top)}$	Junction-to-case(top) thermal resistance	49.4	
$\theta_{JB}$	Junction-to-board thermal resistance	62.8	
$\Psi_{JT}$	Junction-to-top characterization parameter	5.9	
$\Psi_{JB}$	Junction-to-board characterization parameter	62.2	
$\theta_{JC(bottom)}$	Junction-to-case(bottom) thermal resistance	N/A	

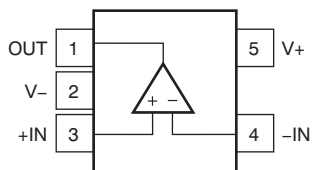
(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

## PIN CONFIGURATIONS

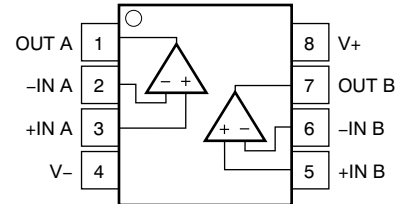
**DCK PACKAGE  
SC70-5  
(TOP VIEW)**



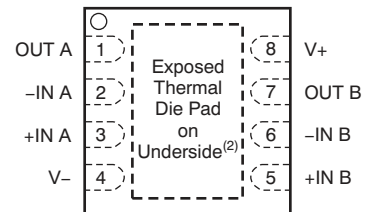
**DBV PACKAGE  
SOT23-5  
(TOP VIEW)**



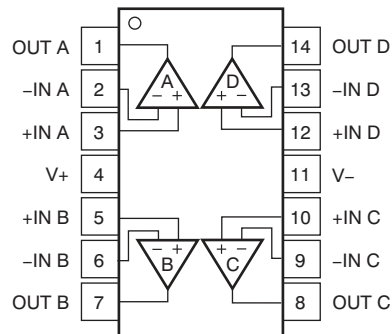
**D, DGK PACKAGES  
SO-8, MSOP-8  
(TOP VIEW)**



**DRG PACKAGE<sup>(1)</sup>  
DFN-8  
(TOP VIEW)**



**PW PACKAGE  
TSSOP-14  
(TOP VIEW)**



(1) Pitch: 0,65 mm.

(2) Connect thermal pad to V-. Pad size: 1,8 mm x 1,5 mm.



## TYPICAL CHARACTERISTICS

**Table 1. Characteristic Performance Measurements**

TITLE	FIGURE
Open-Loop Gain and Phase vs Frequency	<a href="#">Figure 1</a>
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THD+N vs Amplitude (G = -1, 2 k $\Omega$ , 10 k $\Omega$ )	<a href="#">Figure 31</a>
THD+N vs Frequency (0.5 V <sub>RMS</sub> , G = +1, 2 k $\Omega$ , 10 k $\Omega$ )	<a href="#">Figure 32</a>
EMIRR IN+ vs Frequency	<a href="#">Figure 33</a>

### TYPICAL CHARACTERISTICS

At  $T_A = +25\text{ }^\circ\text{C}$ ,  $V_S = 5\text{ V}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S / 2$ , and  $V_{CM} = V_{OUT} = V_S / 2$ , unless otherwise noted.

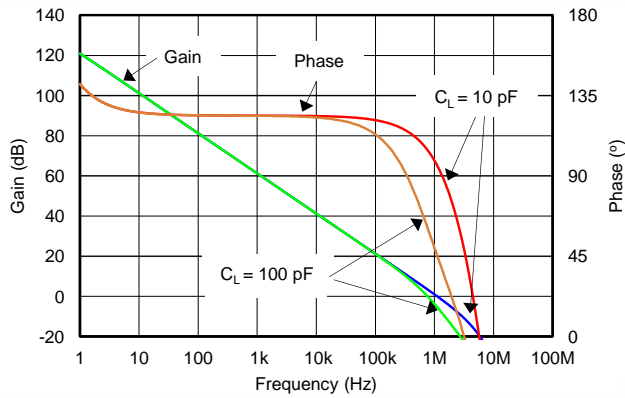


Figure 1. OPEN-LOOP GAIN AND PHASE vs FREQUENCY

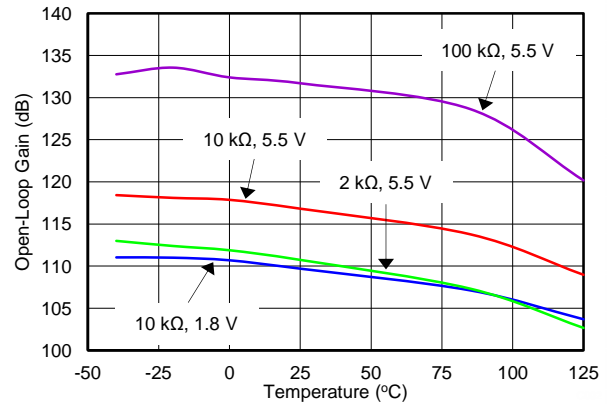


Figure 2. OPEN-LOOP GAIN vs TEMPERATURE

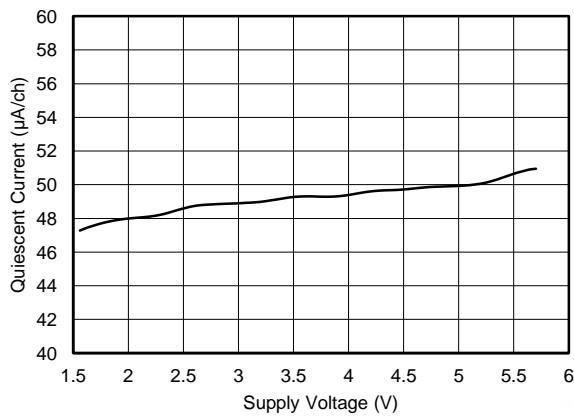


Figure 3. QUIESCENT CURRENT vs SUPPLY

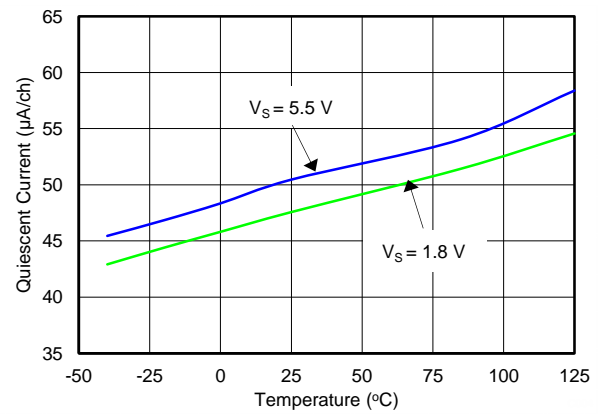


Figure 4. QUIESCENT CURRENT vs TEMPERATURE

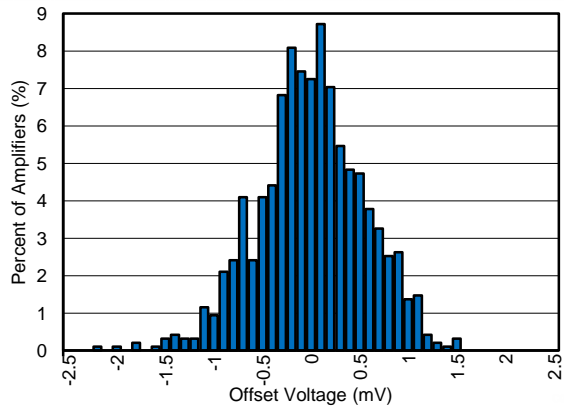


Figure 5. OFFSET VOLTAGE PRODUCTION DISTRIBUTION

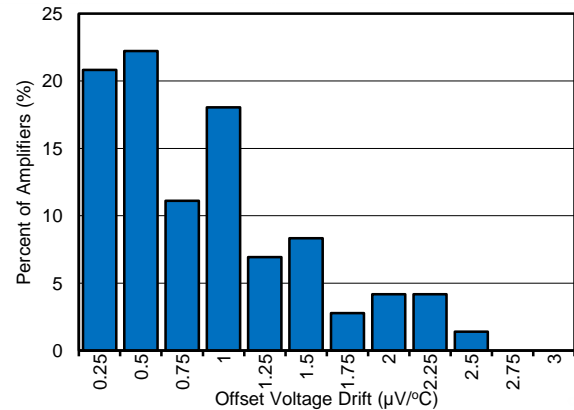
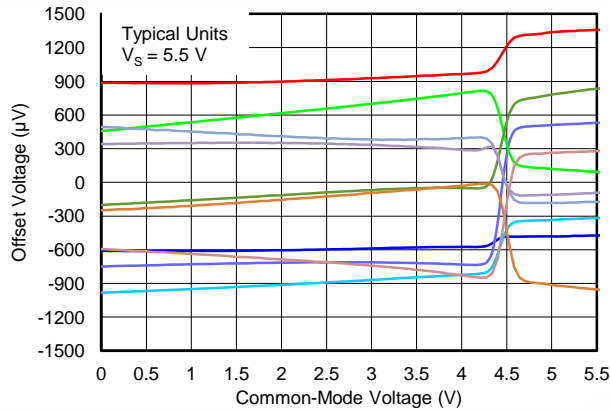


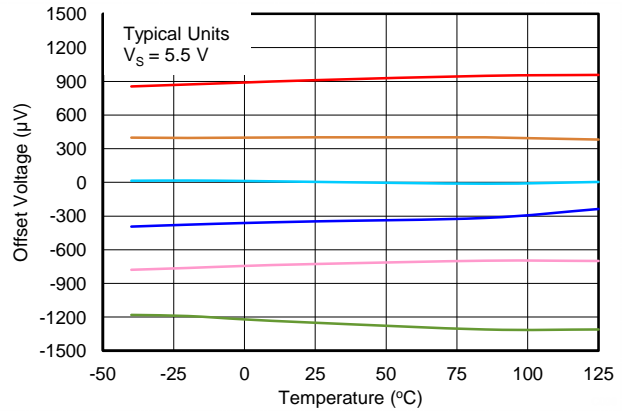
Figure 6. OFFSET VOLTAGE DRIFT DISTRIBUTION

**TYPICAL CHARACTERISTICS (continued)**

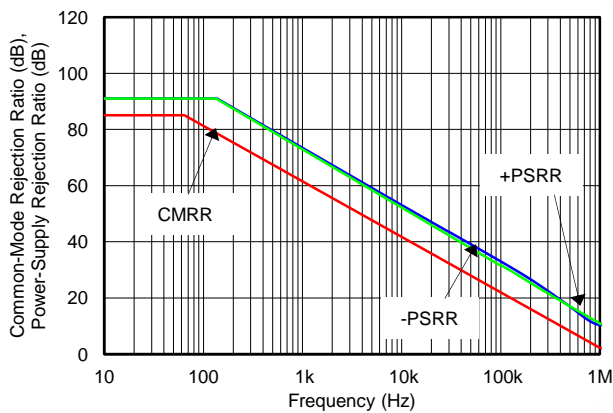
At  $T_A = +25\text{ }^\circ\text{C}$ ,  $V_S = 5\text{ V}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S / 2$ , and  $V_{CM} = V_{OUT} = V_S / 2$ , unless otherwise noted.



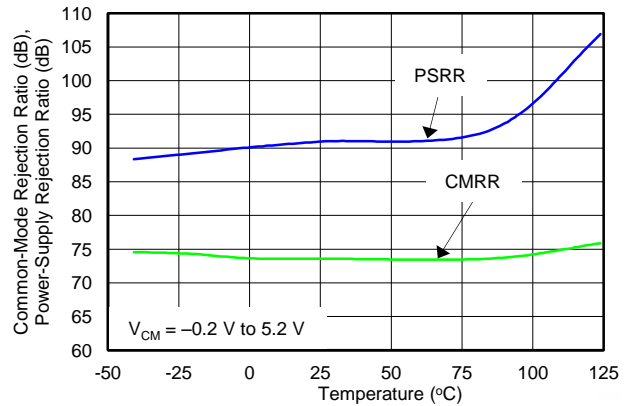
**Figure 7. OFFSET VOLTAGE vs COMMON-MODE VOLTAGE**



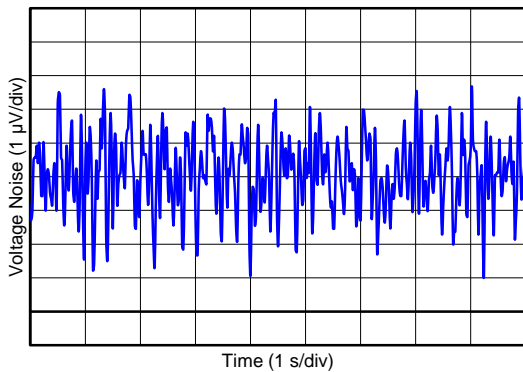
**Figure 8. OFFSET VOLTAGE vs TEMPERATURE**



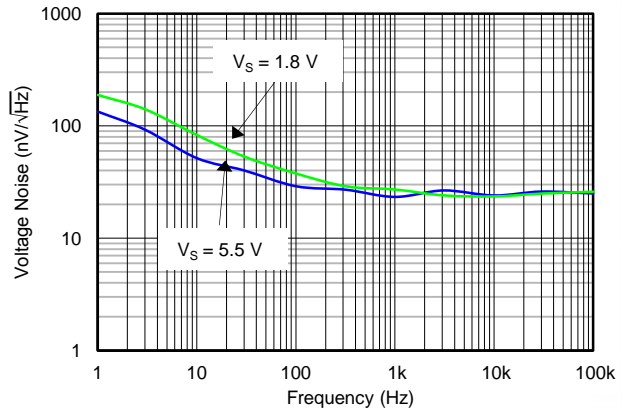
**Figure 9. CMRR AND PSRR vs FREQUENCY (Referred-to-Input)**



**Figure 10. CMRR AND PSRR vs TEMPERATURE**



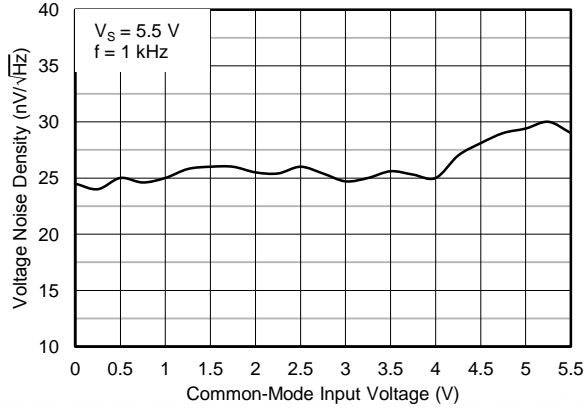
**Figure 11. 0.1-Hz TO 10-Hz INPUT VOLTAGE NOISE**



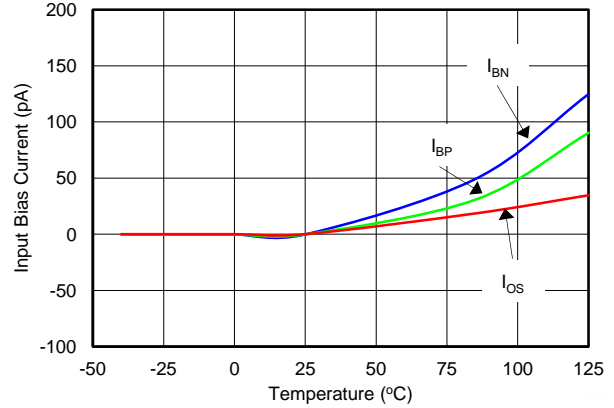
**Figure 12. INPUT VOLTAGE NOISE SPECTRAL DENSITY vs FREQUENCY**

**TYPICAL CHARACTERISTICS (continued)**

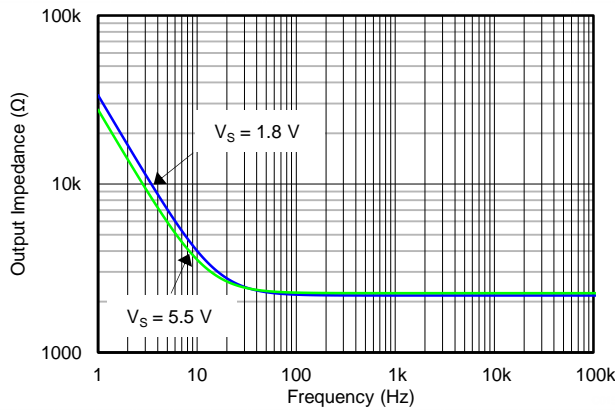
At  $T_A = +25\text{ }^\circ\text{C}$ ,  $V_S = 5\text{ V}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S / 2$ , and  $V_{CM} = V_{OUT} = V_S / 2$ , unless otherwise noted.



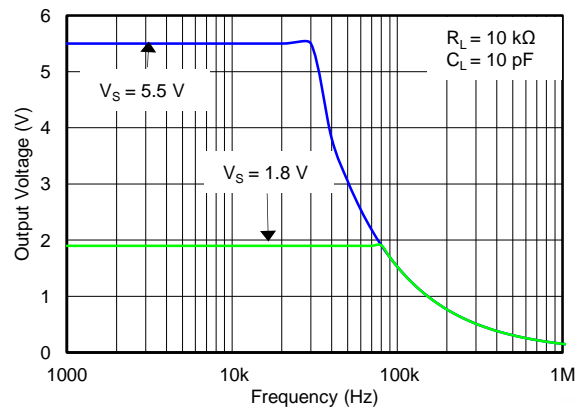
**Figure 13. VOLTAGE NOISE vs COMMON-MODE VOLTAGE**



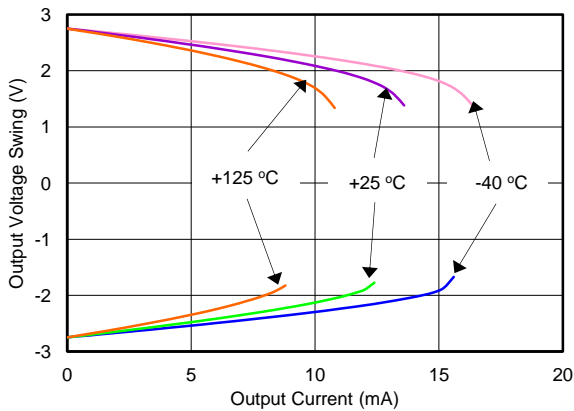
**Figure 14. INPUT BIAS AND OFFSET CURRENT vs TEMPERATURE**



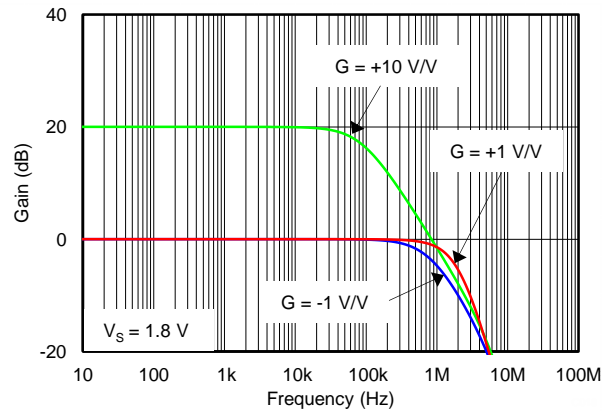
**Figure 15. OPEN-LOOP OUTPUT IMPEDANCE vs FREQUENCY**



**Figure 16. MAXIMUM OUTPUT VOLTAGE vs FREQUENCY AND SUPPLY VOLTAGE**



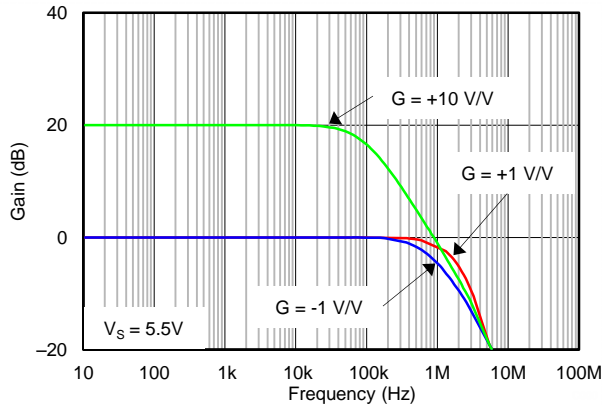
**Figure 17. OUTPUT VOLTAGE SWING vs OUTPUT CURRENT (Over Temperature)**



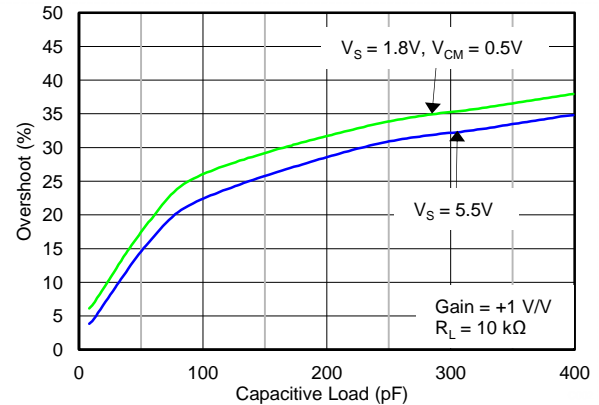
**Figure 18. CLOSED-LOOP GAIN vs FREQUENCY (Minimum Supply)**

**TYPICAL CHARACTERISTICS (continued)**

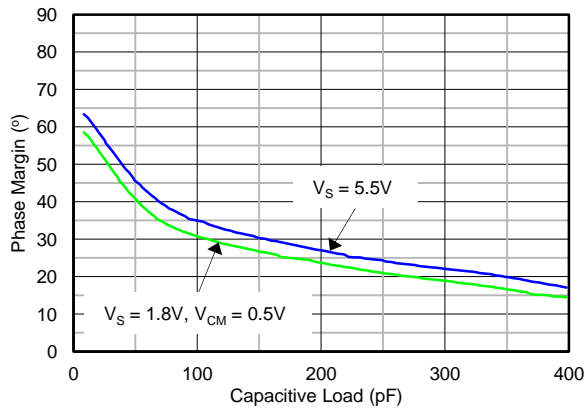
At  $T_A = +25\text{ }^\circ\text{C}$ ,  $V_S = 5\text{ V}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S / 2$ , and  $V_{CM} = V_{OUT} = V_S / 2$ , unless otherwise noted.



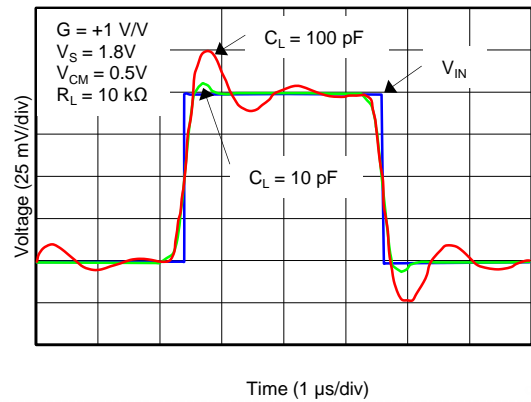
**Figure 19. CLOSED-LOOP GAIN vs FREQUENCY (Maximum Supply)**



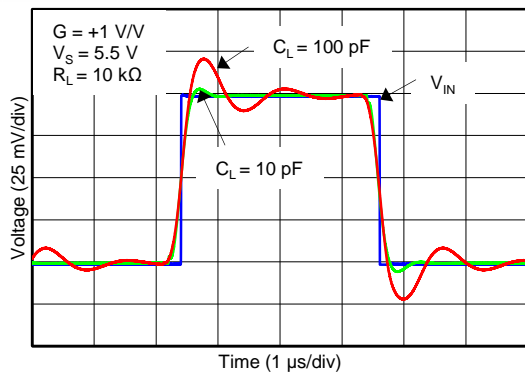
**Figure 20. SMALL-SIGNAL OVERSHOOT vs LOAD CAPACITANCE**



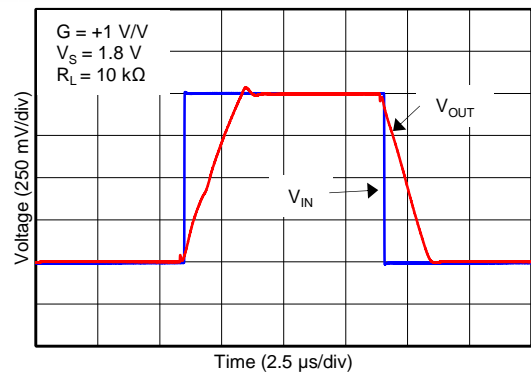
**Figure 21. PHASE MARGIN vs CAPACITIVE LOAD**



**Figure 22. SMALL-SIGNAL PULSE RESPONSE (Minimum Supply)**



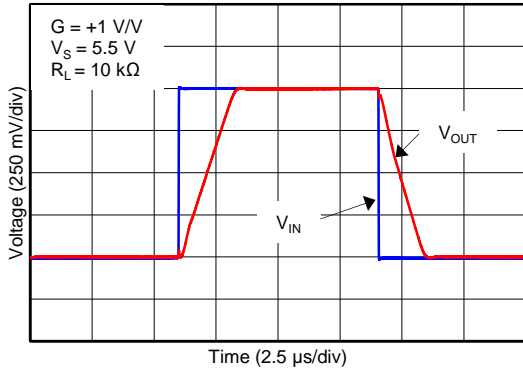
**Figure 23. SMALL-SIGNAL PULSE RESPONSE (Maximum Supply)**



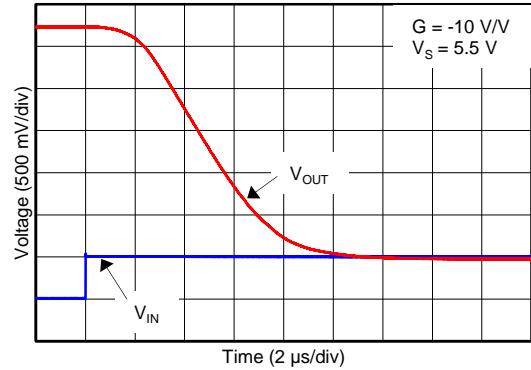
**Figure 24. LARGE-SIGNAL PULSE RESPONSE (Minimum Supply)**

**TYPICAL CHARACTERISTICS (continued)**

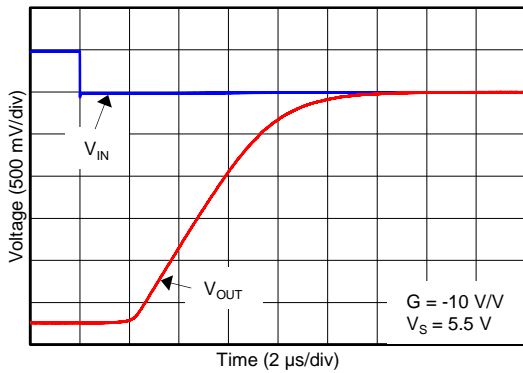
At  $T_A = +25\text{ }^\circ\text{C}$ ,  $V_S = 5\text{ V}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S / 2$ , and  $V_{CM} = V_{OUT} = V_S / 2$ , unless otherwise noted.



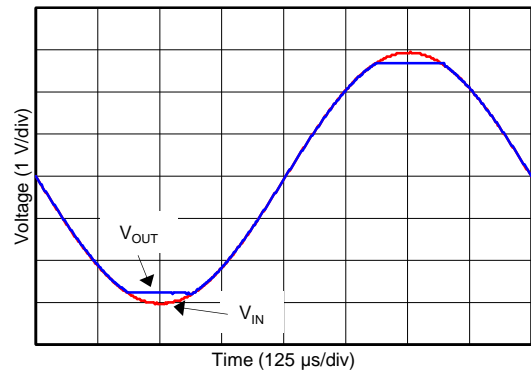
**Figure 25. LARGE-SIGNAL PULSE RESPONSE (Maximum Supply)**



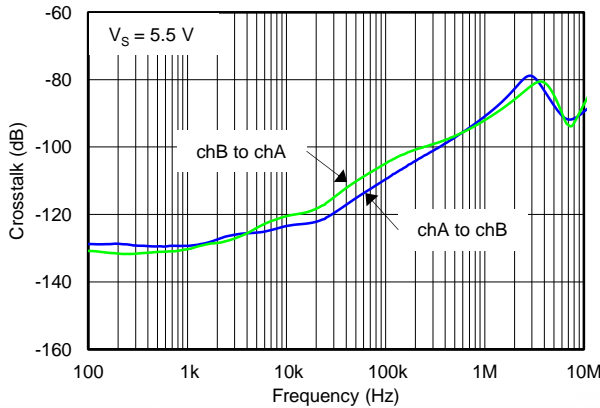
**Figure 26. POSITIVE OVERLOAD RECOVERY**



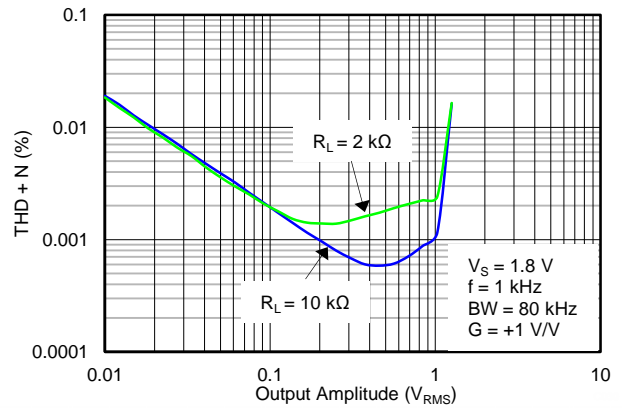
**Figure 27. NEGATIVE OVERLOAD RECOVERY**



**Figure 28. NO PHASE REVERSAL**



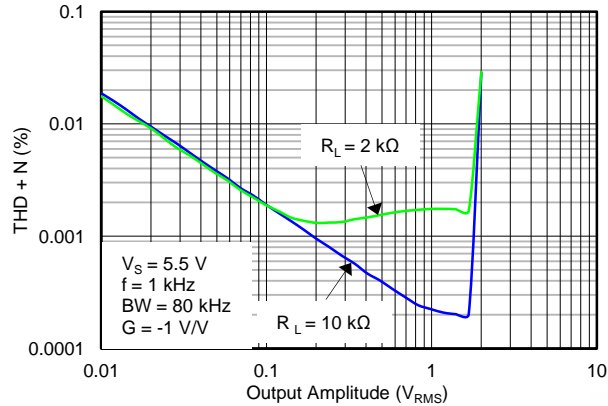
**Figure 29. CHANNEL SEPARATION vs FREQUENCY**



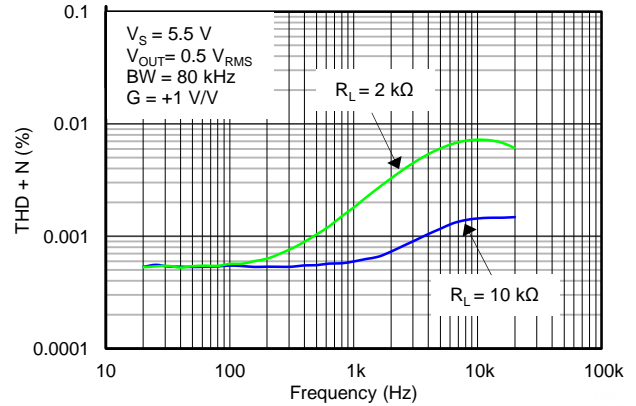
**Figure 30. THD+N vs OUTPUT AMPLITUDE (Minimum Supply)**

**TYPICAL CHARACTERISTICS (continued)**

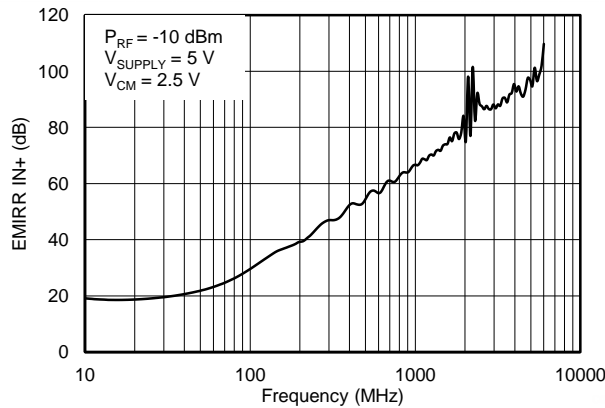
At  $T_A = +25\text{ }^\circ\text{C}$ ,  $V_S = 5\text{ V}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S / 2$ , and  $V_{CM} = V_{OUT} = V_S / 2$ , unless otherwise noted.



**Figure 31. THD+N vs OUTPUT AMPLITUDE (Maximum Supply)**



**Figure 32. THD+N vs FREQUENCY**



**Figure 33. EMIRR IN+ vs FREQUENCY**

## APPLICATION INFORMATION

The OPA313 is a family of low-power, rail-to-rail input and output operational amplifiers specifically designed for portable applications. These devices operate from 1.8 V to 5.5 V, are unity-gain stable, and suitable for a wide range of general-purpose applications. The class AB output stage is capable of driving  $\leq 10\text{-k}\Omega$  loads connected to any point between  $V_+$  and ground. The input common-mode voltage range includes both rails, and allows the OPA313 family to be used in virtually any single-supply application. Rail-to-rail input and output swing significantly increases dynamic range, especially in low-supply applications, and makes them ideal for driving sampling analog-to-digital converters (ADCs).

The OPA313 features 1-MHz bandwidth and  $0.5\text{-V}/\mu\text{s}$  slew rate with only  $50\text{-}\mu\text{A}$  supply current per channel, providing good ac performance at very low power consumption. DC applications are also well served with a low input noise voltage of  $25\text{ nV}/\sqrt{\text{Hz}}$  at 1 kHz, low input bias current ( $0.2\text{ pA}$ ), and an input offset voltage of  $0.5\text{ mV}$  (typical). The typical offset voltage drift is  $2\text{ }\mu\text{V}/^\circ\text{C}$ ; over the full temperature range the input offset voltage changes only  $200\text{ }\mu\text{V}$  ( $0.5\text{ mV}$  to  $0.7\text{ mV}$ ).

## OPERATING VOLTAGE

The OPA313 series op amps are fully specified and ensured for operation from  $+1.8\text{ V}$  to  $+5.5\text{ V}$ . In addition, many specifications apply from  $-40^\circ\text{C}$  to  $+125^\circ\text{C}$ . Parameters that vary significantly with operating voltages or temperature are shown in the [Typical Characteristics](#) graphs. Power-supply pins should be bypassed with  $0.01\text{-}\mu\text{F}$  ceramic capacitors.

## RAIL-TO-RAIL INPUT

The input common-mode voltage range of the OPA313 series extends  $200\text{ mV}$  beyond the supply rails. This performance is achieved with a complementary input stage: an N-channel input differential pair in parallel with a P-channel differential pair, as shown in [Figure 34](#). The N-channel pair is active for input voltages close to the positive rail, typically  $(V_+) - 1.3\text{ V}$  to  $200\text{ mV}$  above the positive supply, while the P-channel pair is on for inputs from  $200\text{ mV}$  below the negative supply to approximately  $(V_+) - 1.3\text{ V}$ . There is a small transition region, typically  $(V_+) - 1.4\text{ V}$  to  $(V_+) - 1.2\text{ V}$ , in which both pairs are on. This  $200\text{-mV}$  transition region can vary up to  $300\text{ mV}$  with process variation. Thus, the transition region (both stages on) can range from  $(V_+) - 1.7\text{ V}$  to  $(V_+) - 1.5\text{ V}$  on the low end, up to  $(V_+) - 1.1\text{ V}$  to  $(V_+) - 0.9\text{ V}$  on the high end. Within this transition region, PSRR, CMRR, offset voltage, offset drift, and THD may be degraded compared to device operation outside this region.

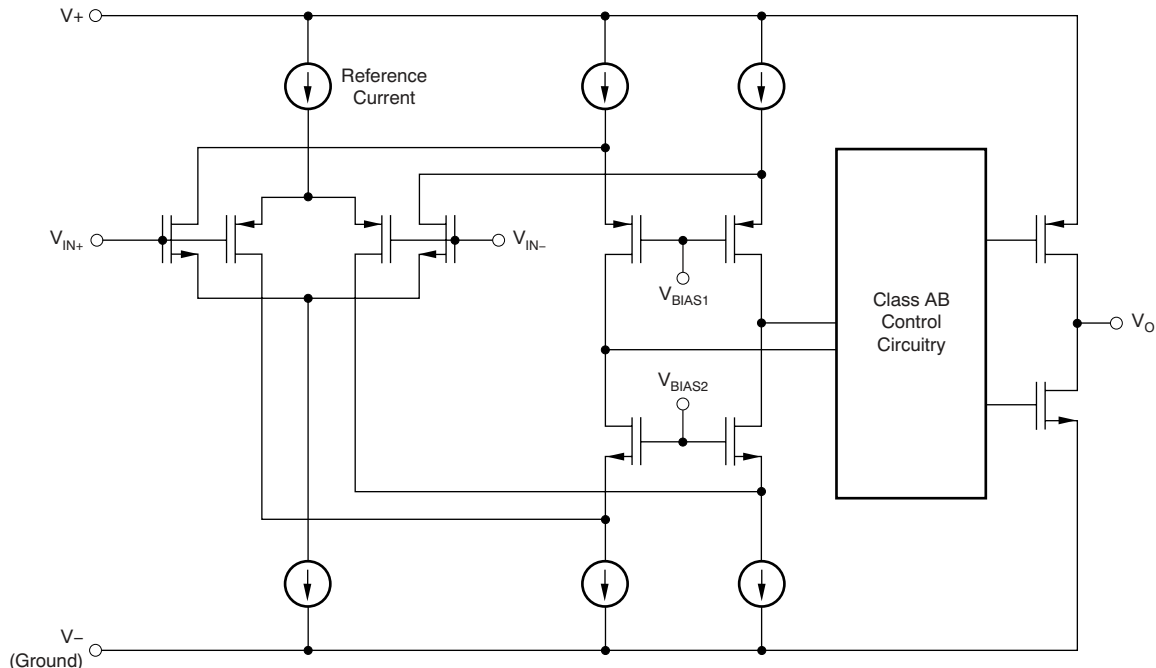
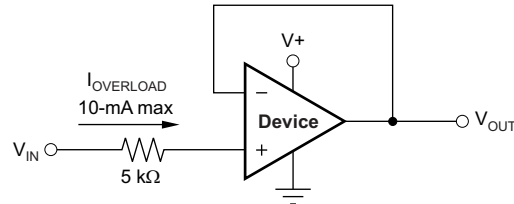


Figure 34. Simplified Schematic



## INPUT AND ESD PROTECTION

The OPA313 family incorporates internal electrostatic discharge (ESD) protection circuits on all pins. In the case of input and output pins, this protection primarily consists of current-steering diodes connected between the input and power-supply pins. These ESD protection diodes also provide in-circuit, input overdrive protection, as long as the current is limited to 10 mA as stated in the [Absolute Maximum Ratings](#). [Figure 35](#) shows how a series input resistor may be added to the driven input to limit the input current. The added resistor contributes thermal noise at the amplifier input and its value should be kept to a minimum in noise-sensitive applications.



**Figure 35. Input Current Protection**

## COMMON-MODE REJECTION RATIO (CMRR)

CMRR for the OPA313 is specified in several ways so the best match for a given application may be used; see the [Electrical Characteristics](#). First, the CMRR of the device in the common-mode range below the transition region [ $V_{CM} < (V+) - 1.3 \text{ V}$ ] is given. This specification is the best indicator of the capability of the device when the application requires use of one of the differential input pairs. Second, the CMRR over the entire common-mode range is specified at ( $V_{CM} = -0.2 \text{ V to } 5.7 \text{ V}$ ). This last value includes the variations seen through the transition region (see [Figure 7](#)).

## EMI SUSCEPTIBILITY AND INPUT FILTERING

Operational amplifiers vary with regard to the susceptibility of the device to electromagnetic interference (EMI). If conducted EMI enters the op amp, the dc offset observed at the amplifier output may shift from its nominal value while EMI is present. This shift is a result of signal rectification associated with the internal semiconductor junctions. While all op amp pin functions can be affected by EMI, the signal input pins are likely to be the most susceptible. The OPA313 operational amplifier family incorporate an internal input low-pass filter that reduces the amplifiers response to EMI. Both common-mode and differential mode filtering are provided by this filter. The filter is designed for a cutoff frequency of approximately 35 MHz (–3 dB), with a roll-off of 20 dB per decade.

Texas Instruments has developed the ability to accurately measure and quantify the immunity of an operational amplifier over a broad frequency spectrum extending from 10 MHz to 6 GHz. The EMI rejection ratio (EMIRR) metric allows op amps to be directly compared by the EMI immunity. [Figure 33](#) illustrates the results of this testing on the OPA313 family. Detailed information can also be found in the application report, *EMI Rejection Ratio of Operational Amplifiers* ([SBOA128](#)), available for download from [www.ti.com](http://www.ti.com).

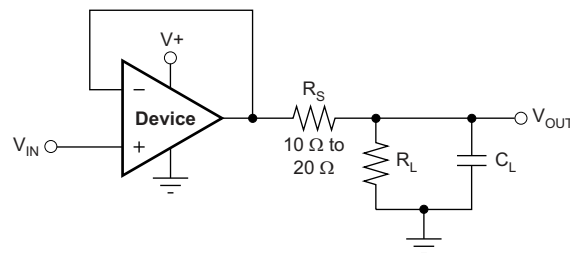
## RAIL-TO-RAIL OUTPUT

Designed as a micro-power, low-noise operational amplifier, the OPA313 delivers a robust output drive capability. A class AB output stage with common-source transistors is used to achieve full rail-to-rail output swing capability. For resistive loads up to 10 kΩ, the output swings typically to within 5 mV of either supply rail regardless of the power-supply voltage applied. Different load conditions change the ability of the amplifier to swing close to the rails; refer to the typical characteristic graph, [Output Voltage Swing vs Output Current](#).

## CAPACITIVE LOAD AND STABILITY

The OPA313 is designed to be used in applications where driving a capacitive load is required. As with all op amps, there may be specific instances where the OPA313 can become unstable. The particular op amp circuit configuration, layout, gain, and output loading are some of the factors to consider when establishing whether or not an amplifier is stable in operation. An op amp in the unity-gain (+1-V/V) buffer configuration that drives a capacitive load exhibits a greater tendency to be unstable than an amplifier operated at a higher noise gain. The capacitive load, in conjunction with the op amp output resistance, creates a pole within the feedback loop that degrades the phase margin. The degradation of the phase margin increases as the capacitive loading increases. When operating in the unity-gain configuration, the OPA313 remains stable with a pure capacitive load up to approximately 1 nF. The equivalent series resistance (ESR) of some very large capacitors ( $C_L$  greater than 1  $\mu$ F) is sufficient to alter the phase characteristics in the feedback loop such that the amplifier remains stable. Increasing the amplifier closed-loop gain allows the amplifier to drive increasingly larger capacitance. This increased capability is evident when observing the overshoot response of the amplifier at higher voltage gains. See the typical characteristic graph, [Small-Signal Overshoot vs. Capacitive Load](#).

One technique for increasing the capacitive load drive capability of the amplifier operating in a unity-gain configuration is to insert a small resistor, typically 10  $\Omega$  to 20  $\Omega$ , in series with the output, as shown in [Figure 36](#). This resistor significantly reduces the overshoot and ringing associated with large capacitive loads. One possible problem with this technique, however, is that a voltage divider is created with the added series resistor and any resistor connected in parallel with the capacitive load. The voltage divider introduces a gain error at the output that reduces the output swing.



**Figure 36. Improving Capacitive Load Drive**

## DFN PACKAGE

The OPA2313 (dual version) uses the DFN style package (also known as SON); this package is a QFN with contacts on only two sides of the package bottom. This leadless package maximizes printed circuit board (PCB) space and offers enhanced thermal and electrical characteristics through an exposed pad. One of the primary advantages of the DFN package is its low, 0.9-mm height. DFN packages are physically small, have a smaller routing area, improved thermal performance, reduced electrical parasitics, and use a pinout scheme that is consistent with other commonly-used packages, such as SO and MSOP. Additionally, the absence of external leads eliminates bent-lead issues.

The DFN package can easily be mounted using standard PCB assembly techniques. See Application Note, [QFN/SON PCB Attachment \(SLUA271\)](#) and Application Report, [Quad Flatpack No-Lead Logic Packages \(SCBA017\)](#), both available for download from [www.ti.com](http://www.ti.com).

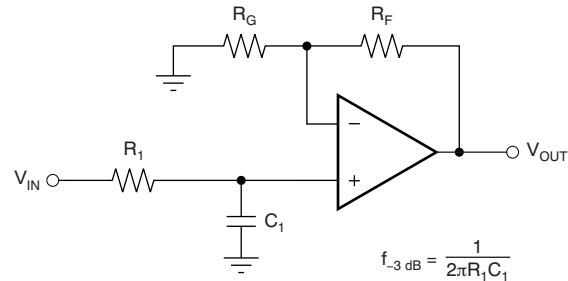
### NOTE

The exposed leadframe die pad on the bottom of the DFN package should be connected to the most negative potential (V-).

## APPLICATION EXAMPLES

### GENERAL CONFIGURATIONS

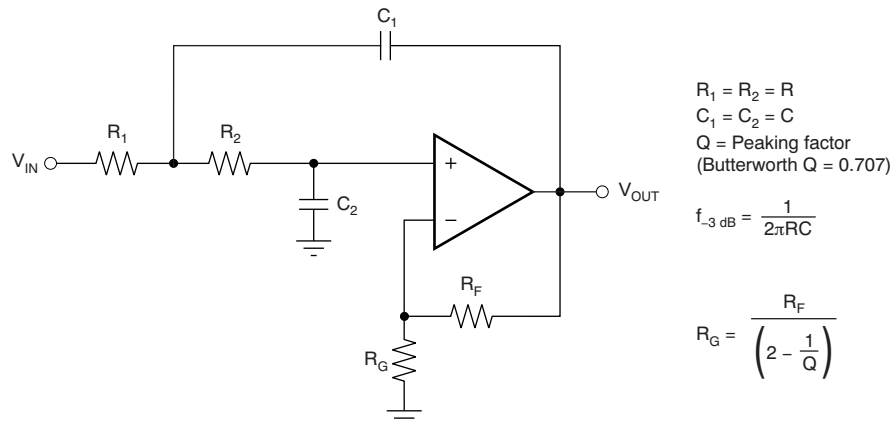
When receiving low-level signals, limiting the bandwidth of the incoming signals into the system is often required. The simplest way to establish this limited bandwidth is to place an RC filter at the noninverting terminal of the amplifier, as [Figure 37](#) shows.



$$\frac{V_{OUT}}{V_{IN}} = \left(1 + \frac{R_F}{R_G}\right) \left(\frac{1}{1 + sR_1C_1}\right)$$

**Figure 37. Single-Pole Low-Pass Filter**

If even more attenuation is needed, a multiple pole filter is required. The Sallen-Key filter can be used for this task, as [Figure 38](#) shows. For best results, the amplifier should have a bandwidth that is eight to 10 times the filter frequency bandwidth. Failure to follow this guideline can result in phase shift of the amplifier.



**Figure 38. Two-Pole Low-Pass Sallen-Key Filter**

## REVISION HISTORY

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision B (December 2012) to Revision C	Page
• Changed first Open-Loop Gain, $A_{OL}$ typical specification in Electrical Characteristics: +5.5 V table .....	3
• Updated <a href="#">Figure 10</a> .....	11
• Updated <a href="#">Figure 19</a> through <a href="#">Figure 22</a> .....	12

Changes from Revision A (September 2012) to Revision B	Page
• Changed title of document .....	1
• Changed third paragraph of Description section .....	1
• Changed title of Electrical Characteristics: +5.5 V table .....	3
• Deleted middle two rows from Input Voltage Range, $CMRR$ parameter in Electrical Characteristics: +5.5 V table .....	3
• Changed test conditions of Input Voltage Range, $CMRR$ parameter in Electrical Characteristics: +5.5 V table .....	3
• Added footnote to Input Bias Current, $I_B$ and $I_{OS}$ parameters in Electrical Characteristics: +5.5 V table .....	3
• Changed Open-Loop Gain, $A_{OL}$ parameter in Electrical Characteristics: +5.5 V table .....	3
• Deleted first row from Frequency Response, $GBW$ parameter in Electrical Characteristics: +5.5 V table .....	4
• Deleted first row from Frequency Response, $SR$ parameter in Electrical Characteristics: +5.5 V table .....	4
• Changed Output, $V_O$ parameter in Electrical Characteristics: +5.5 V table .....	4
• Changed Output, $I_{SC}$ parameter in Electrical Characteristics: +5.5 V table .....	4
• Changed test conditions for the first row in the Power Supply, $I_Q$ parameter in Electrical Characteristics: +5.5 V table .....	4
• Changed Electrical Characteristics: +1.8 V table .....	5
• Changed conditions of Electrical Characteristics: +1.8 V table .....	5
• Changed last row of Input Voltage Range, $CMRR$ parameter in Electrical Characteristics: +1.8 V table .....	5
• Changed footnote to Input Bias Current, $I_B$ and $I_{OS}$ parameters in Electrical Characteristics: +1.8 V table .....	5
• Changed Open-Loop Gain, $A_{OL}$ parameter in Electrical Characteristics: +1.8 V table .....	5
• Changed Frequency Response, $GBW$ parameter test conditions in Electrical Characteristics: +1.8 V table .....	6
• Changed Frequency Response, $SR$ parameter test conditions in Electrical Characteristics: +1.8 V table .....	6
• Changed Output, $V_O$ parameter test conditions in Electrical Characteristics: +1.8 V table .....	6
• Changed Output, $I_{SC}$ parameter in Electrical Characteristics: +1.8 V table .....	6
• Deleted last row from Power Supply, $I_Q$ parameter in Electrical Characteristics: +1.8 V table .....	6
• Updated <a href="#">Figure 2</a> .....	10

Changes from Original (September 2012) to Revision A	Page
• Changed from product preview to production data .....	1

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
OPA2313ID	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OP2313	<a href="#">Samples</a>
OPA2313IDGK	ACTIVE	VSSOP	DGK	8	80	RoHS & Green	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	OUSS	<a href="#">Samples</a>
OPA2313IDGKR	ACTIVE	VSSOP	DGK	8	2500	RoHS & Green	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	OUSS	<a href="#">Samples</a>
OPA2313IDR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OP2313	<a href="#">Samples</a>
OPA2313IDRGR	ACTIVE	SON	DRG	8	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	SDY	<a href="#">Samples</a>
OPA2313IDRGT	ACTIVE	SON	DRG	8	250	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	SDY	<a href="#">Samples</a>
OPA313IDBVR	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	SIE	<a href="#">Samples</a>
OPA313IDBVT	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	SIE	<a href="#">Samples</a>
OPA313IDCKR	ACTIVE	SC70	DCK	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	SIF	<a href="#">Samples</a>
OPA313IDCKT	ACTIVE	SC70	DCK	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	SIF	<a href="#">Samples</a>
OPA4313IPW	ACTIVE	TSSOP	PW	14	90	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA4313	<a href="#">Samples</a>
OPA4313IPWR	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA4313	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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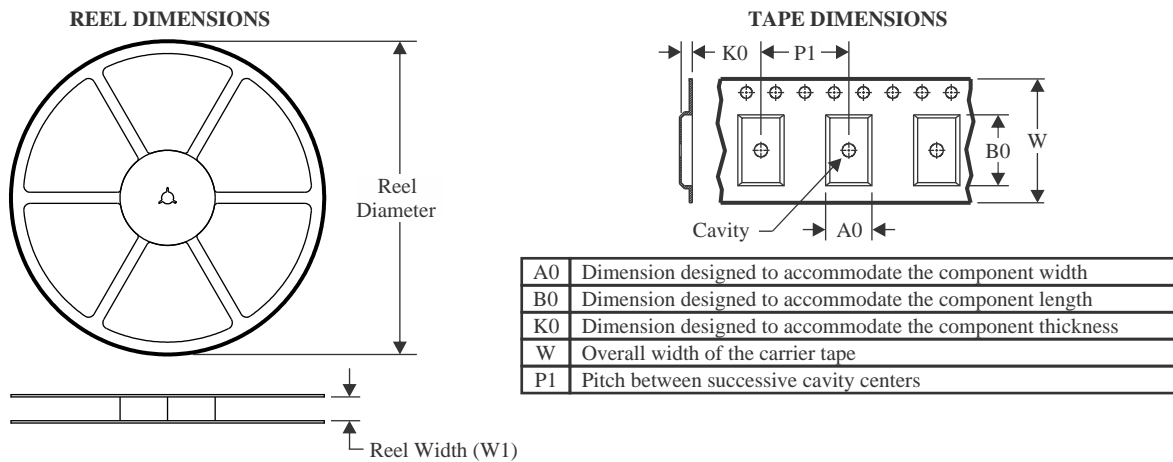
**OTHER QUALIFIED VERSIONS OF OPA2313 :**

- Automotive: [OPA2313-Q1](#)

NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

## TAPE AND REEL INFORMATION



### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
OPA2313IDGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA2313IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA2313IDRGR	SON	DRG	8	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
OPA2313IDRGT	SON	DRG	8	250	180.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
OPA313IDBVR	SOT-23	DBV	5	3000	178.0	9.0	3.3	3.2	1.4	4.0	8.0	Q3
OPA313IDBVT	SOT-23	DBV	5	250	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
OPA313IDCKR	SC70	DCK	5	3000	178.0	9.0	2.4	2.5	1.2	4.0	8.0	Q3
OPA4313IPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
OPA2313IDGKR	VSSOP	DGK	8	2500	366.0	364.0	50.0
OPA2313IDR	SOIC	D	8	2500	340.5	336.1	25.0
OPA2313IDRGR	SON	DRG	8	3000	346.0	346.0	33.0
OPA2313IDRGT	SON	DRG	8	250	210.0	185.0	35.0
OPA313IDBVR	SOT-23	DBV	5	3000	180.0	180.0	18.0
OPA313IDBVT	SOT-23	DBV	5	250	180.0	180.0	18.0
OPA313IDCKR	SC70	DCK	5	3000	180.0	180.0	18.0
OPA4313IPWR	TSSOP	PW	14	2000	356.0	356.0	35.0



**TUBE**


\*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
OPA2313ID	D	SOIC	8	75	507	8	3940	4.32
OPA2313IDGK	DGK	VSSOP	8	80	330	6.55	500	2.88
OPA4313IPW	PW	TSSOP	14	90	530	10.2	3600	3.5



# EXAMPLE BOARD LAYOUT

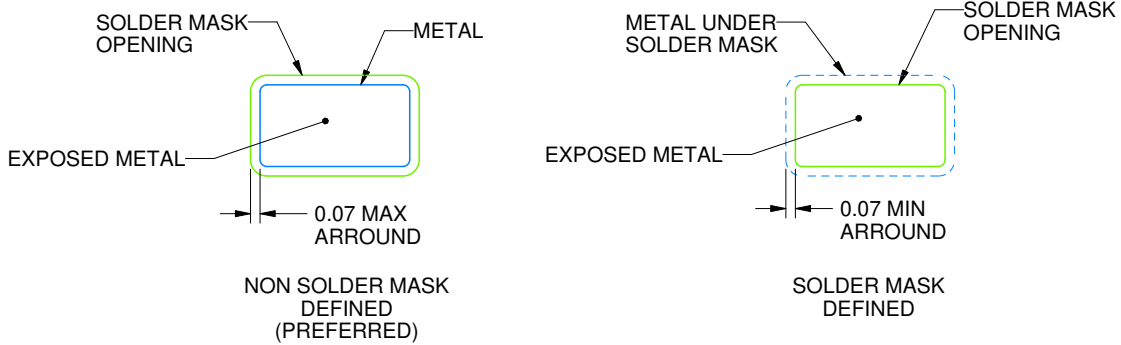
DBV0005A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:15X



SOLDER MASK DETAILS

4214839/H 09/2023

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DBV0005A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE:15X

4214839/H 09/2023

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.



PW (R-PDSO-G14)

PLASTIC SMALL OUTLINE



- NOTES:
- All linear dimensions are in millimeters.
  - This drawing is subject to change without notice.
  - Publication IPC-7351 is recommended for alternate designs.
  - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
  - Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

# D0008A



## PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



### NOTES:

1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
4. This dimension does not include interlead flash.
5. Reference JEDEC registration MS-012, variation AA.

# EXAMPLE BOARD LAYOUT

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



# EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE  
BASED ON .005 INCH [0.125 MM] THICK STENCIL  
SCALE:8X

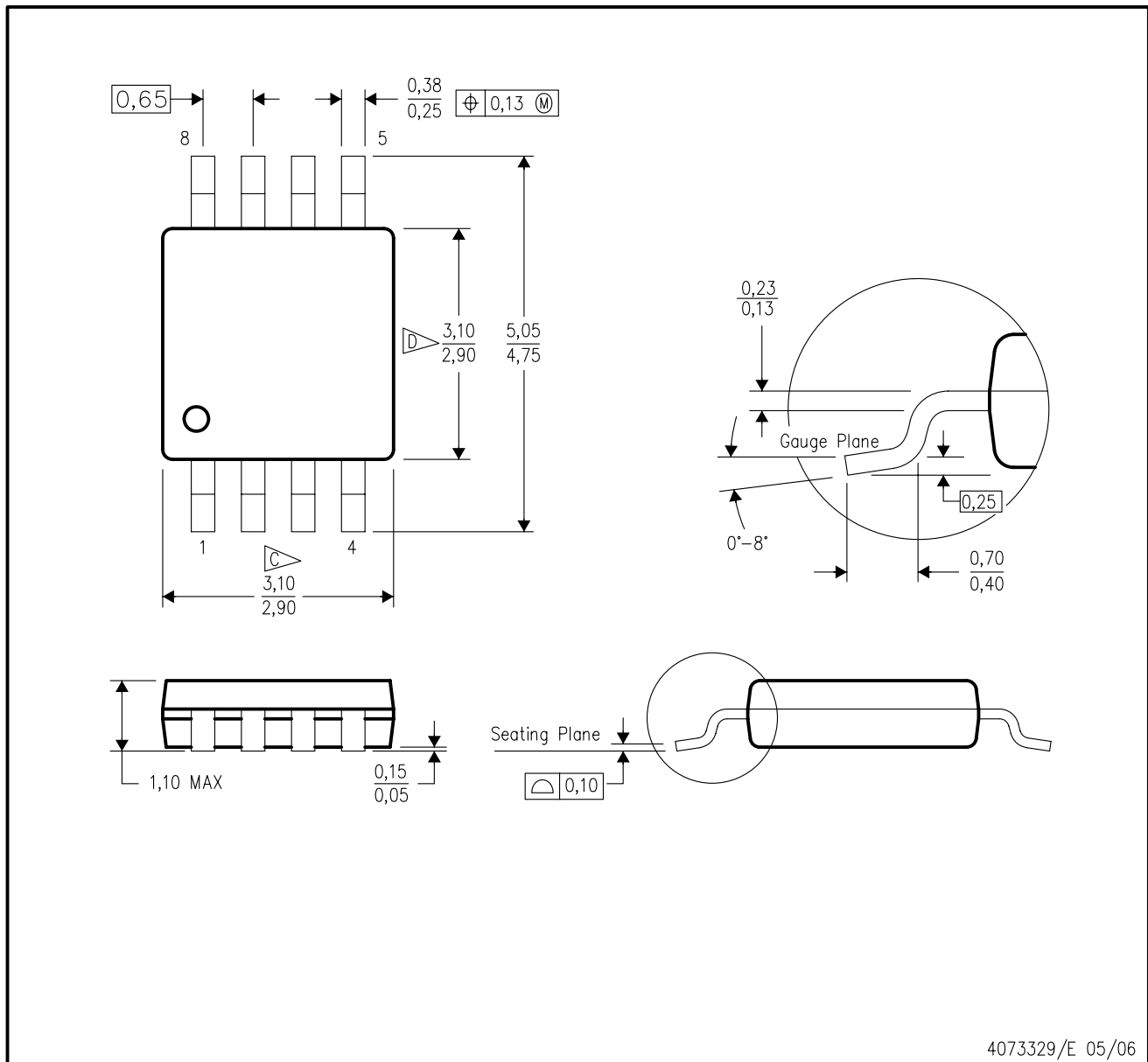
4214825/C 02/2019

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

DGK (S-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



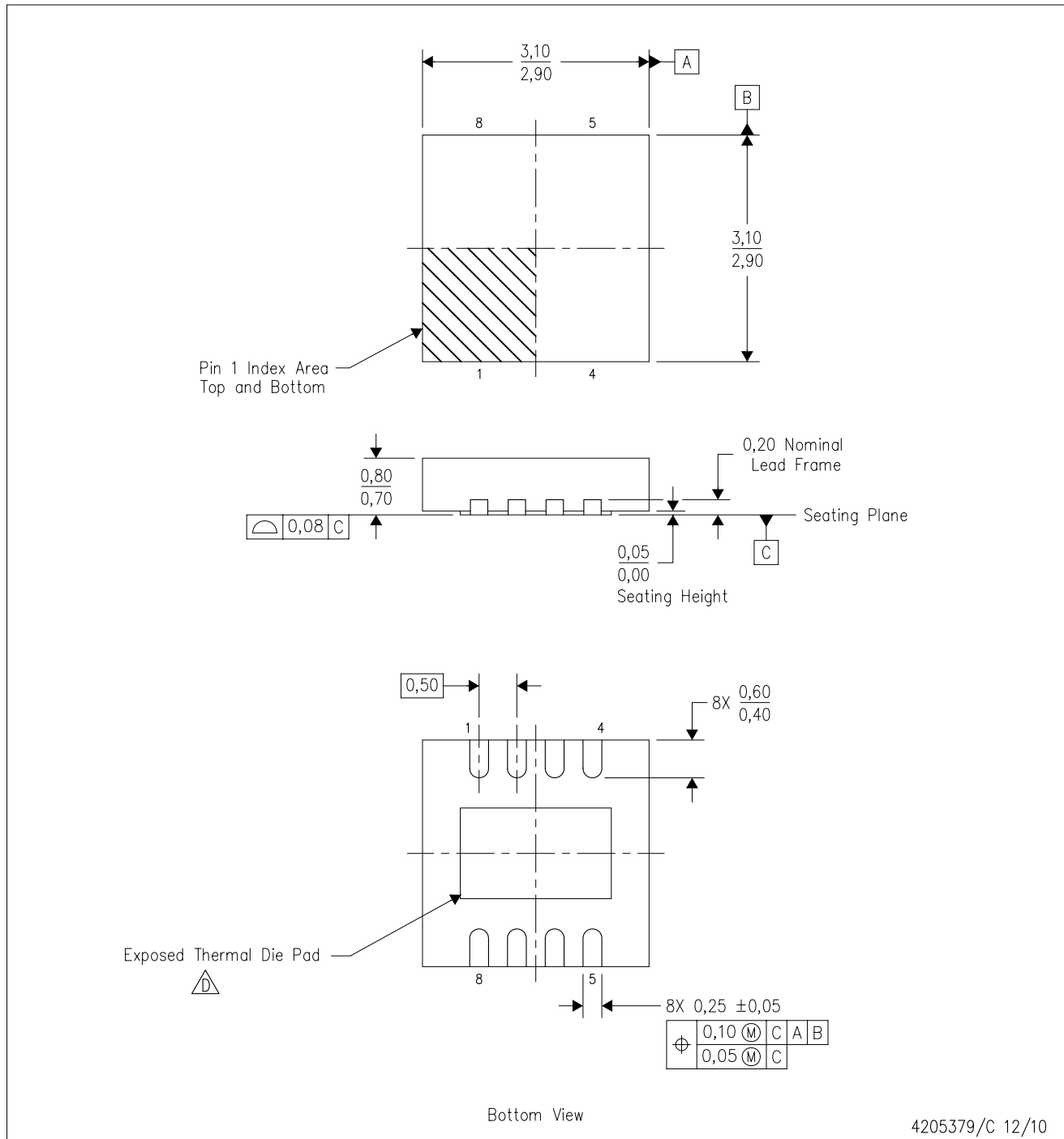
- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - $\nabla$  C Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 per end.
  - $\nabla$  D Body width does not include interlead flash. Interlead flash shall not exceed 0.50 per side.
  - E. Falls within JEDEC MO-187 variation AA, except interlead flash.



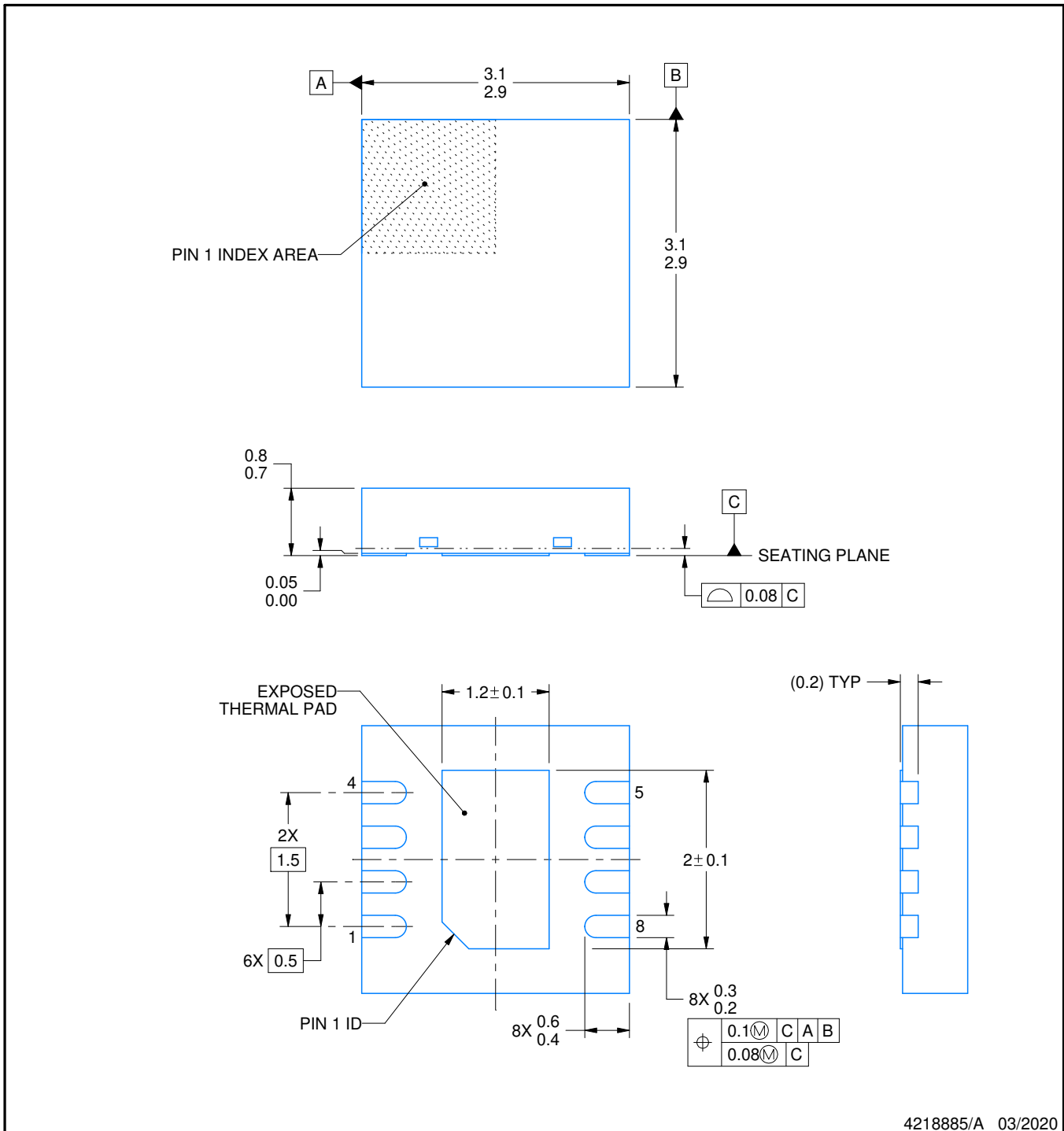
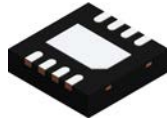
- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate designs.
  - D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
  - E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

DRG (S-PWSON-N8)

PLASTIC SMALL OUTLINE NO-LEAD



- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
  - B. This drawing is subject to change without notice.
  - C. SON (Small Outline No-Lead) package configuration.
  - D. The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.
  - E. JEDEC MO-229 package registration pending.



4218885/A 03/2020

**NOTES:**

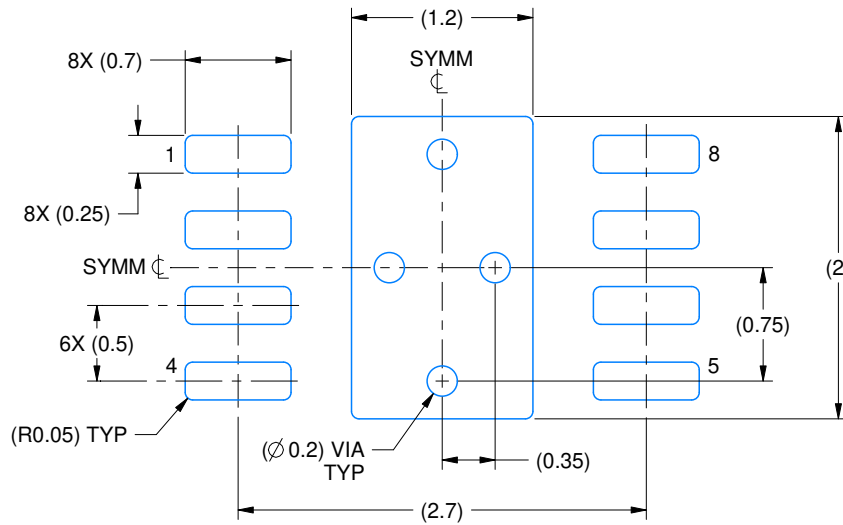
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

# EXAMPLE BOARD LAYOUT

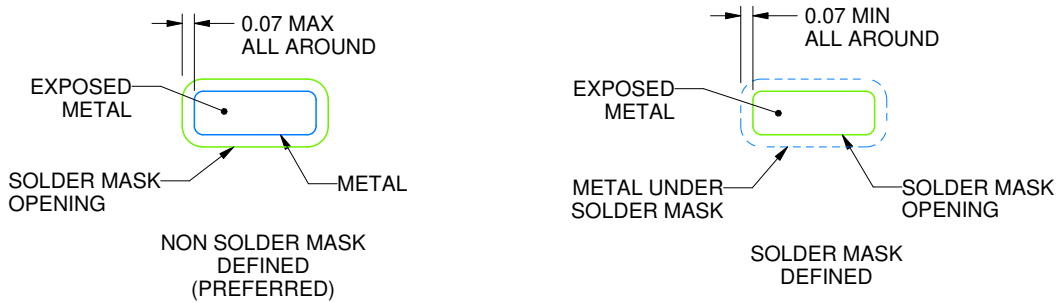
**DRG0008A**

**WSON - 0.8 mm max height**

PLASTIC SMALL OUTLINE - NO LEAD



**LAND PATTERN EXAMPLE**  
EXPOSED METAL SHOWN  
SCALE:20X



**SOLDER MASK DETAILS**

4218885/A 03/2020

NOTES: (continued)

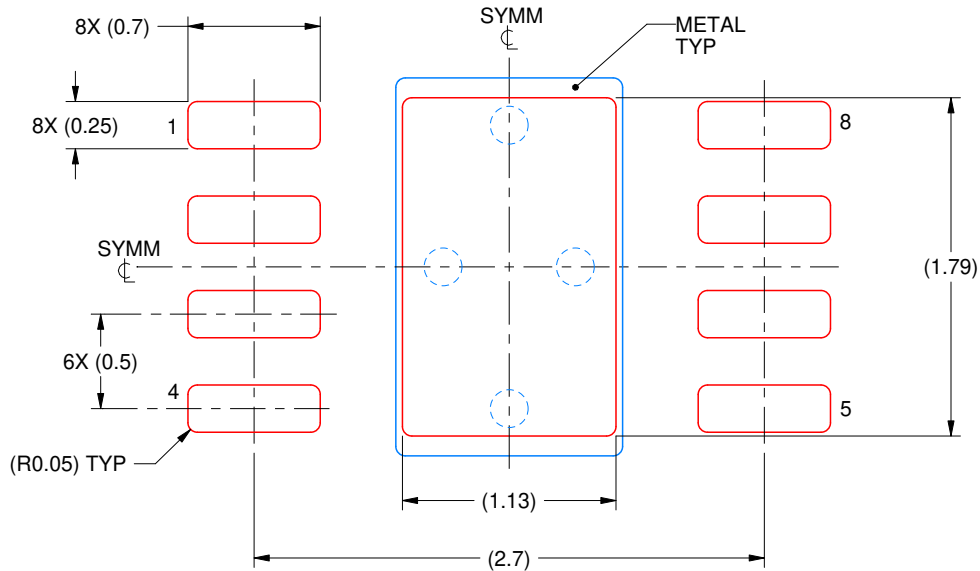
- This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 ([www.ti.com/lit/slUA271](http://www.ti.com/lit/slUA271)).
- Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

# EXAMPLE STENCIL DESIGN

DRG0008A

WSON - 0.8 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD  
84% PRINTED SOLDER COVERAGE BY AREA  
SCALE:25X

4218885/A 03/2020

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

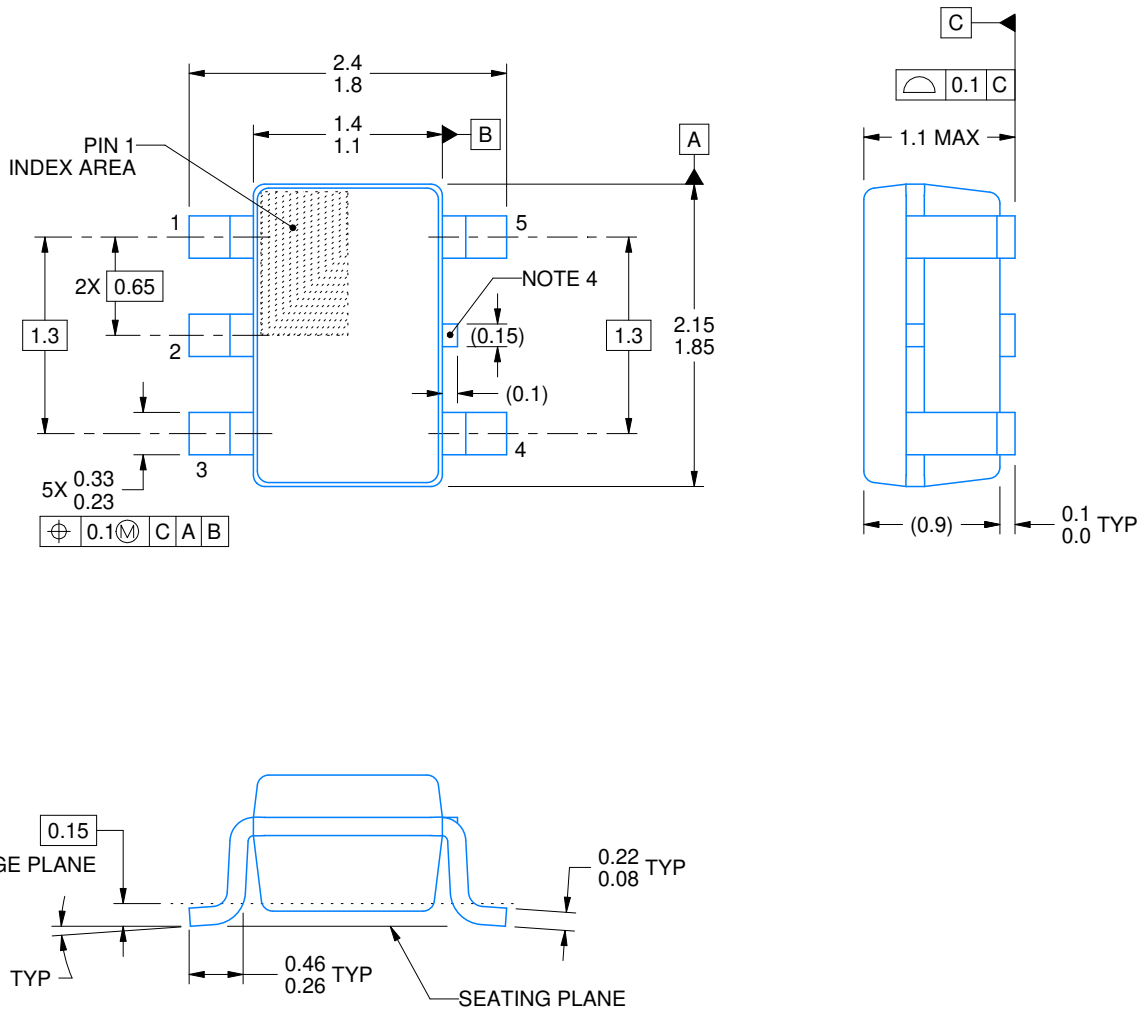
DCK0005A



# PACKAGE OUTLINE

SOT - 1.1 max height

SMALL OUTLINE TRANSISTOR



4214834/C 03/2023

## NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Reference JEDEC MO-203.
4. Support pin may differ or may not be present.

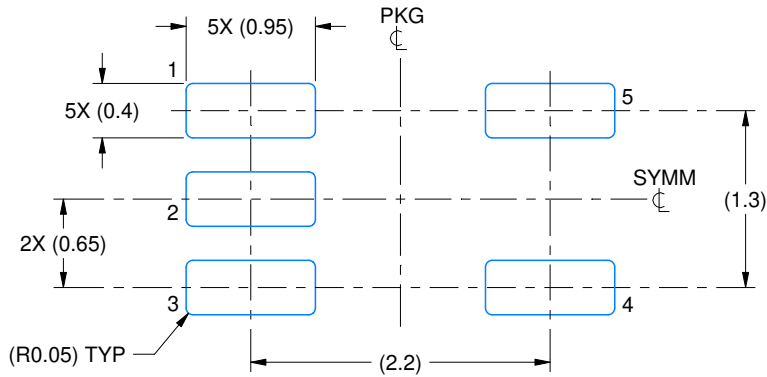


# EXAMPLE BOARD LAYOUT

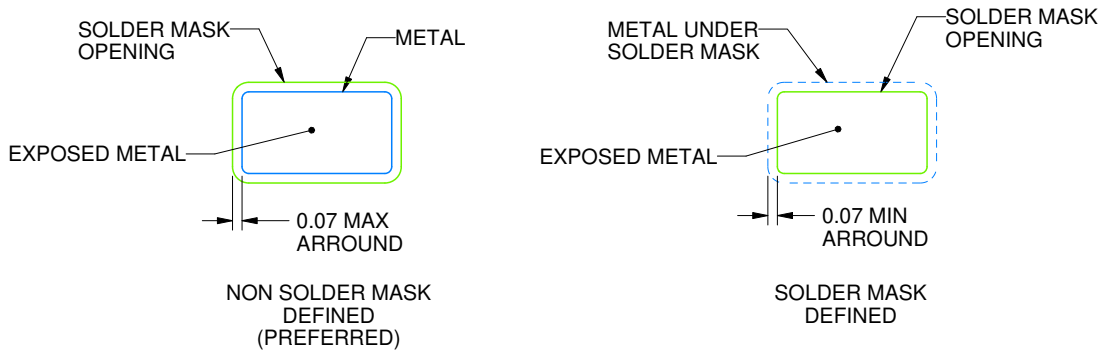
DCK0005A

SOT - 1.1 max height

SMALL OUTLINE TRANSISTOR



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:18X



SOLDER MASK DETAILS

4214834/C 03/2023

NOTES: (continued)

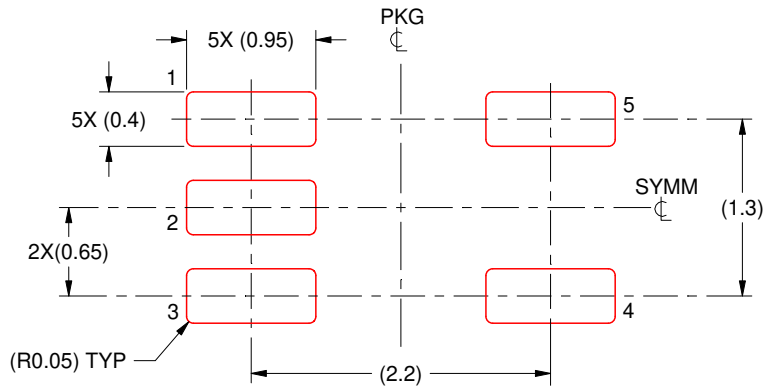
- 4. Publication IPC-7351 may have alternate designs.
- 5. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DCK0005A

SOT - 1.1 max height

SMALL OUTLINE TRANSISTOR



SOLDER PASTE EXAMPLE  
BASED ON 0.125 THICK STENCIL  
SCALE:18X

4214834/C 03/2023

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
7. Board assembly site may have different recommendations for stencil design.

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