

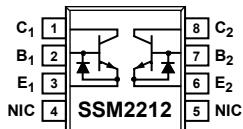
Data Sheet

SSM2212

FEATURES

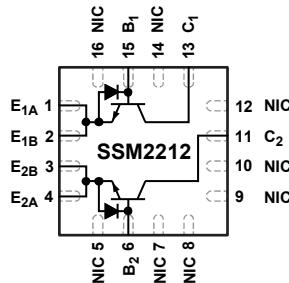
- Very low voltage noise: 1 nV/ $\sqrt{\text{Hz}}$ maximum at 100 Hz**
- Excellent current gain match: 0.5%**
- Low offset voltage (V_{OS}): 200 μV maximum (SOIC)**
- Outstanding offset voltage drift: 0.03 $\mu\text{V}/^\circ\text{C}$**
- High gain bandwidth product: 200 MHz**

PIN CONNECTIONS



NOTES
1. NIC = NO INTERNAL CONNECTION.
09043-001

Figure 1. 8-Lead SOIC_N



NOTES
1. NIC = NO INTERNAL CONNECTION.
09043-020

Figure 2. 16-Lead LFCSP_WQ

GENERAL DESCRIPTION

The **SSM2212** is a dual, NPN-matched transistor pair that is specifically designed to meet the requirements of ultralow noise audio systems.

With its extremely low input base spreading resistance ($r_{bb'}$ is typically 28Ω) and high current gain (h_{FE} typically exceeds 600 at $I_C = 1 \text{ mA}$), the **SSM2212** can achieve outstanding signal-to-noise ratios. The high current gain results in superior performance compared to systems incorporating commercially available monolithic amplifiers.

Excellent matching of the current gain (Δh_{FE}) to approximately 0.5% and low V_{OS} of less than $10 \mu\text{V}$ typical make the **SSM2212** ideal for symmetrically balanced designs, which reduce high-order amplifier harmonic distortion.

Stability of the matching parameters is guaranteed by protection diodes across the base-emitter junction. These diodes prevent degradation of beta and matching characteristics due to reverse biasing of the base-emitter junction.

The **SSM2212** is also an ideal choice for accurate and reliable current biasing and mirroring circuits. Furthermore, because the accuracy of a current mirror degrades exponentially with mismatches of V_{BE} between transistor pairs, the low V_{OS} of the **SSM2212** does not need offset trimming in most circuit applications.

The **SSM2212** SOIC performance and characteristics are guaranteed over the extended temperature range of -40°C to $+85^\circ\text{C}$.

The **SSM2212** is available in 8-lead SOIC and 16-lead LFCSP packages.

Rev. C

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Document Feedback

TABLE OF CONTENTS

Features	1	Thermal Resistance	5
Pin Connections	1	ESD Caution.....	5
General Description	1	Typical Performance Characteristics	6
Revision History	2	Pin Configurations and Function Descriptions	9
Specifications.....	3	Applications Information	10
Electrical Characteristics—SOIC Package	3	Fast Logarithmic Amplifier.....	10
Electrical Characteristics—LFCSP Package.....	4	Outline Dimensions	11
Absolute Maximum Rating	5	Ordering Guide	11

REVISION HISTORY

6/15—Rev. B to Rev. C

Added LFCSP Package.....Universal Changes to Features Section and General Description Section.....	1
Changed Pin Configuration Section to Pin Connections Section ..	1
Added Figure 2; Renumbered Sequentially.....	1
Added Electrical Characteristics—LFCSP Package Section and Table 2; Renumbered Sequentially	4
Changes to Table 4.....	5
Added Pin Configurations and Function Descriptions Section, Figure 17, Table 5, Figure 18, and Table 6	9
Added Figure 21.....	11
Updated Outline Dimensions	11
Changes to Ordering Guide.....	11

7/10—Rev. A to Rev. B

Changes to Figure 1	1
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6/10—Rev. 0 to Rev. A

Changes to Fast Logarithmic Amplifier Section	8
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6/10—Revision 0: Initial Version

SPECIFICATIONS

ELECTRICAL CHARACTERISTICS—SOIC PACKAGE

$V_{CB} = 15 \text{ V}$, $I_O = 10 \mu\text{A}$, $T_A = 25^\circ\text{C}$, unless otherwise specified.

Table 1.

Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
DC AND AC CHARACTERISTICS						
Current Gain ¹	h_{FE}	$I_C = 1 \text{ mA}$ $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$	300	605		
		$I_C = 10 \mu\text{A}$ $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$	200	550		
Current Gain Match ²	Δh_{FE}	$10 \mu\text{A} \leq I_C \leq 1 \text{ mA}$		0.5	5	%
Noise Voltage Density ³	e_N	$I_C = 1 \text{ mA}$, $V_{CB} = 0 \text{ V}$ $f_0 = 10 \text{ Hz}$ $f_0 = 100 \text{ Hz}$ $f_0 = 1 \text{ kHz}$ $f_0 = 10 \text{ kHz}$		1.6	2	$\text{nV}/\sqrt{\text{Hz}}$
		$I_C = 1 \text{ mA}$ $f_0 = 10 \text{ Hz}$ $f_0 = 100 \text{ Hz}$ $f_0 = 1 \text{ kHz}$ $f_0 = 10 \text{ kHz}$		0.9	1	$\text{nV}/\sqrt{\text{Hz}}$
		$I_C = 1 \text{ mA}$ $f_0 = 10 \text{ Hz}$ $f_0 = 100 \text{ Hz}$ $f_0 = 1 \text{ kHz}$ $f_0 = 10 \text{ kHz}$		0.85	1	$\text{nV}/\sqrt{\text{Hz}}$
Low Frequency Noise (0.1 Hz to 10 Hz)	$e_N \text{ p-p}$	$I_C = 1 \text{ mA}$		0.4		$\mu\text{V p-p}$
Offset Voltage	V_{OS}	$V_{CB} = 0 \text{ V}$, $I_C = 1 \text{ mA}$ $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$		10	200	μV
Offset Voltage Change vs. V_{CB}	$\Delta V_{OS}/\Delta V_{CB}$	$0 \text{ V} \leq V_{CB} \leq V_{MAX}^4$, $1 \mu\text{A} \leq I_C \leq 1 \text{ mA}^5$		10	50	μV
Offset Voltage Change vs. I_C	$\Delta V_{OS}/\Delta I_C$	$1 \mu\text{A} \leq I_C \leq 1 \text{ mA}^5$, $V_{CB} = 0 \text{ V}$		5	70	μV
Offset Voltage Drift	$\Delta V_{OS}/\Delta T$	$-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$, V_{OS} trimmed to 0 V		0.08	1	$\mu\text{V}/^\circ\text{C}$
				0.03	0.3	$\mu\text{V}/^\circ\text{C}$
Breakdown Voltage	BV_{CEO}		40			V
Gain Bandwidth Product	f_T	$I_C = 10 \text{ mA}$, $V_{CE} = 10 \text{ V}$		200		MHz
Collector-to-Base Leakage Current	I_{CBO}	$V_{CB} = V_{MAX}$ $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$		25	500	pA
Collector-to-Collector Leakage Current	I_{CC}	$V_{CC} = V_{MAX}^{6,7}$ $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$		3		nA
Collector-to-Emitter Leakage Current	I_{CES}	$V_{CE} = V_{MAX}$, $V_{BE} = 0 \text{ V}^{6,7}$ $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$		35	500	pA
Input Bias Current	I_B	$I_C = 10 \mu\text{A}$ $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$		4		nA
Input Offset Current	I_{OS}	$I_C = 10 \mu\text{A}$ $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$		50		nA
Input Offset Current Drift	$\Delta I_{OS}/\Delta T$	$I_C = 10 \mu\text{A}^6$, $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$		6.2		nA
Collector Saturation Voltage	$V_{CE(\text{SAT})}$	$I_C = 1 \text{ mA}$, $I_B = 100 \mu\text{A}$		13		nA
Output Capacitance	C_{OB}	$V_{CB} = 15 \text{ V}$, $I_E = 0 \mu\text{A}$		23		pF
Bulk Resistance	R_{BE}	$10 \mu\text{A} \leq I_C \leq 10 \text{ mA}^6$		0.3	1.6	Ω
Collector-to-Collector Capacitance	C_{CC}	$V_{CC} = 0 \text{ V}$		35		pF

¹ Current gain is guaranteed with collector-to-base voltage (V_{CB}) swept from 0 V to V_{MAX} at the indicated collector currents.

² Current gain match (Δh_{FE}) is defined as follows: $\Delta h_{FE} = (100(\Delta I_B)(h_{FE,\text{min}})/I_C)$.

³ Noise voltage density is guaranteed, but not 100% tested.

⁴ This is the maximum change in V_{OS} as V_{CB} is swept from 0 V to 40 V.

⁵ Measured at $I_C = 10 \mu\text{A}$ and guaranteed by design over the specified range of I_C .

⁶ Guaranteed by design.

⁷ I_{CC} and I_{CES} are verified by measurement of I_{CBO} .

ELECTRICAL CHARACTERISTICS—LFCSP PACKAGE

$V_{CB} = 15 \text{ V}$, $I_O = 100 \mu\text{A}$, $T_A = 25^\circ\text{C}$, unless otherwise specified.

Table 2.

Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
DC AND AC CHARACTERISTICS						
Current Gain ¹	h_{FE}	$I_C = 1 \text{ mA}, V_{CB} = 15 \text{ V}$ $I_C = 1 \text{ mA}, V_{CB} = 0 \text{ V}$ $I_C = 100 \mu\text{A}, V_{CB} = 15 \text{ V}$ $I_C = 100 \mu\text{A}, V_{CB} = 0 \text{ V}$	300	1800	2400	
Current Gain Match ²	Δh_{FE}	$100 \mu\text{A} \leq I_C \leq 1 \text{ mA}$	200	1300	2200	
Noise Voltage Density ³	e_N	$I_C = 1 \text{ mA}, V_{CB} = 0 \text{ V}$ $f_O = 10 \text{ Hz}$ $f_O = 100 \text{ Hz}$ $f_O = 1 \text{ kHz}$ $f_O = 10 \text{ kHz}$	350	2100	2500	
Low Frequency Noise (0.1 Hz to 10 Hz)	$e_N \text{ p-p}$	$I_C = 1 \text{ mA}$	250	1500	2300	
Offset Voltage	V_{OS}	$V_{CB} = 0 \text{ V}, I_C = 1 \text{ mA}$ $V_{CB} = 0 \text{ V}, I_C = 100 \mu\text{A}$	0.5	5	%	
Gain Bandwidth Product	f_T	$I_C = 10 \text{ mA}, V_{CE} = 10 \text{ V}$	1.6	2	nV/ $\sqrt{\text{Hz}}$	
Input Bias Current	I_B	$I_C = 100 \mu\text{A}$	0.9	1	nV/ $\sqrt{\text{Hz}}$	
Input Offset Current	I_{OS}	$I_C = 100 \mu\text{A}$	0.85	1	nV/ $\sqrt{\text{Hz}}$	
Output Capacitance	C_{OB}	$V_{CB} = 15 \text{ V}, I_E = 0 \mu\text{A}$	0.85	1	nV/ $\sqrt{\text{Hz}}$	
Collector-to-Collector Capacitance	C_{CC}	$V_{CC} = 0 \text{ V}$	200	250	250	MHz
			10	200	250	μV
			23	35	35	μV
			10	200	250	nA
			35	35	35	pF

¹ Current gain is guaranteed with collector-to-base voltage (V_{CB}) swept from 0 V to V_{MAX} at the indicated collector currents.

² Current gain match (Δh_{FE}) is defined as follows: $\Delta h_{FE} = (100(\Delta I_B)(h_{FE \text{ min}})/I_C)$.

³ Noise voltage density is guaranteed, but not 100% tested.

ABSOLUTE MAXIMUM RATING

Table 3.

Parameter	Rating
Breakdown Voltage of Collector-to-Base Voltage (BV_{CBO})	40 V
Breakdown Voltage of Collector-to-Emitter Voltage (BV_{CEO})	40 V
Breakdown Voltage of Collector-to-Collector Voltage (BV_{CC})	40 V
Breakdown Voltage of Emitter-to-Emitter Voltage (BV_{EE})	40 V
Collector Current (I_C)	20 mA
Emitter Current (I_E)	20 mA
Storage Temperature Range	-65°C to +150°C
Operating Temperature Range	-40°C to +85°C
Junction Temperature Range	-65°C to +150°C
Lead Temperature (Soldering, 60 sec)	300°C

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

θ_{JA} is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages.

Table 4. Thermal Resistance

Package Type	θ_{JA}	θ_{JC}	Unit
8-Lead SOIC (R-8)	120	45	°C/W
16-Lead LFCSP (CP-16-22)	75	4.4	°C/W

ESD CAUTION



ESD (electrostatic discharge) sensitive device.

Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

TYPICAL PERFORMANCE CHARACTERISTICS

$T_A = 25^\circ\text{C}$, $V_{CE} = 5\text{ V}$, unless otherwise specified.

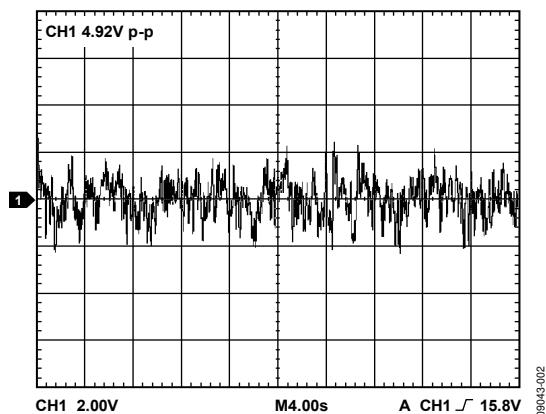


Figure 3. Low Frequency Noise (0.1 Hz to 10 Hz), $I_C = 1\text{ mA}$, Gain = $10,000,000$

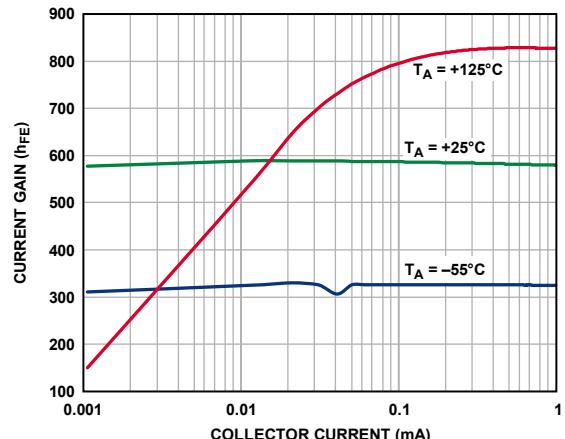


Figure 6. Current Gain vs. Collector Current ($V_{CE} = 0\text{ V}$)

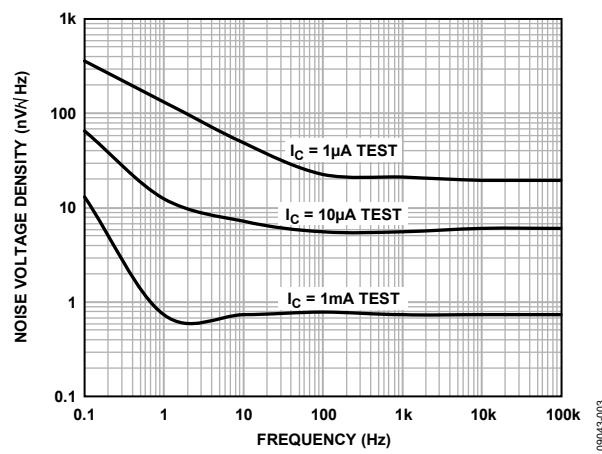


Figure 4. Noise Voltage Density vs. Frequency

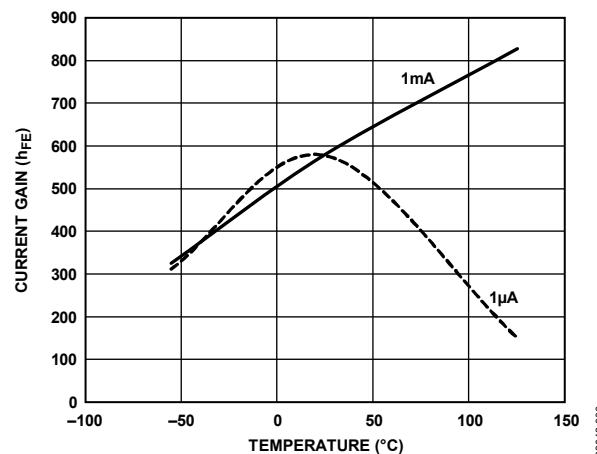


Figure 7. Current Gain vs. Temperature (Excludes I_{CB0})

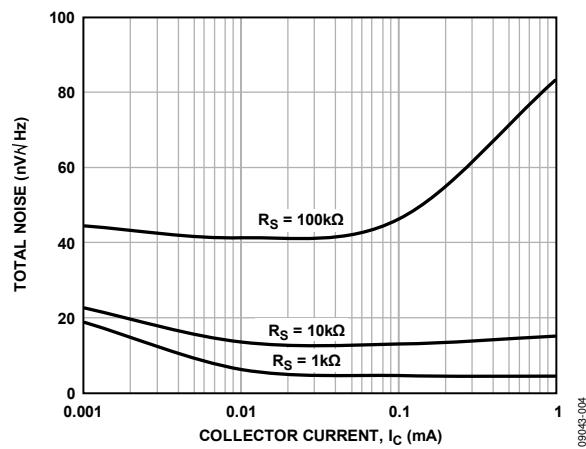


Figure 5. Total Noise vs. Collector Current, $f = 1\text{ kHz}$

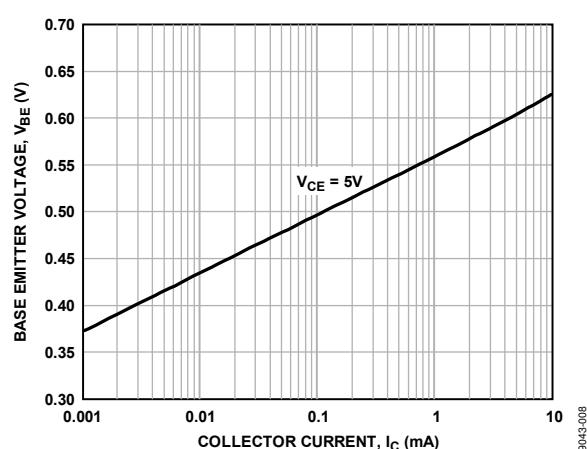


Figure 8. Base Emitter Voltage vs. Collector Current

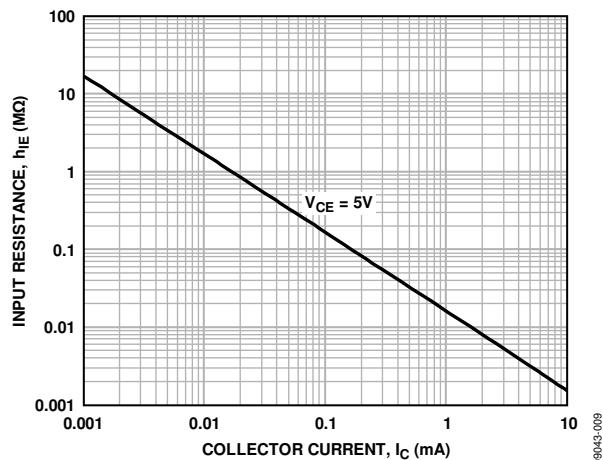


Figure 9. Small Signal Input Resistance vs. Collector Current

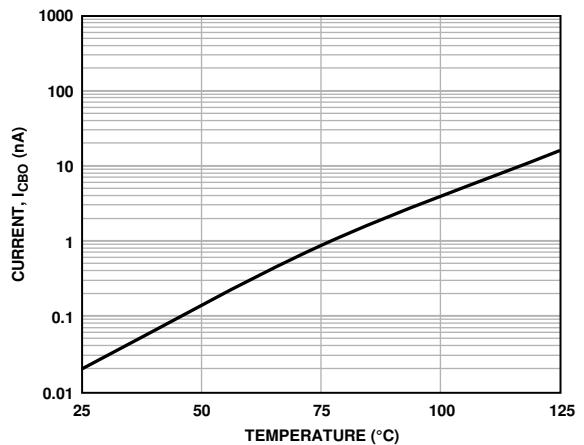


Figure 12. Collector-to-Base Leakage Current vs. Temperature

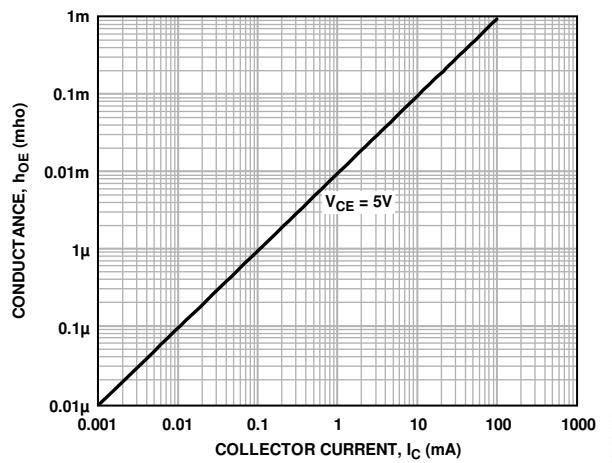


Figure 10. Small Signal Output Conductance vs. Collector Current

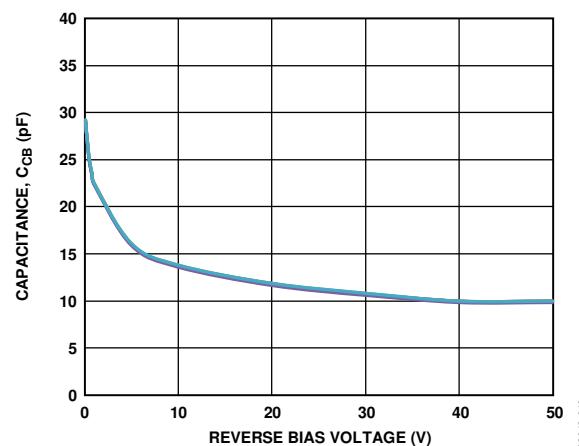


Figure 13. Collector-to-Base Capacitance vs. Reverse Bias Voltage

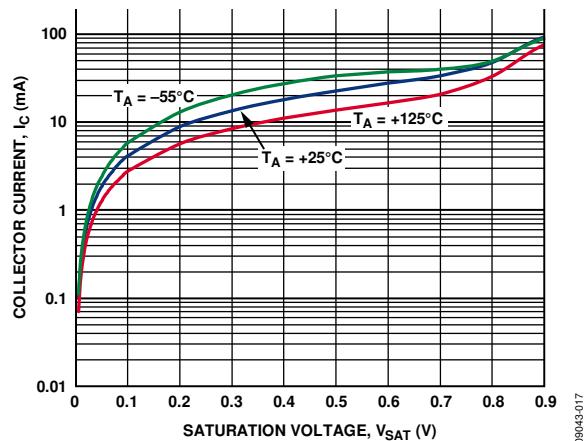


Figure 11. Collector Current vs. Saturation Voltage

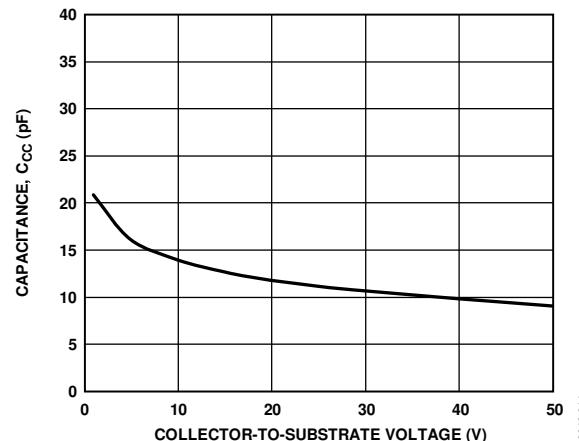


Figure 14. Collector-to-Collector Capacitance vs. Collector-to-Substrate Voltage

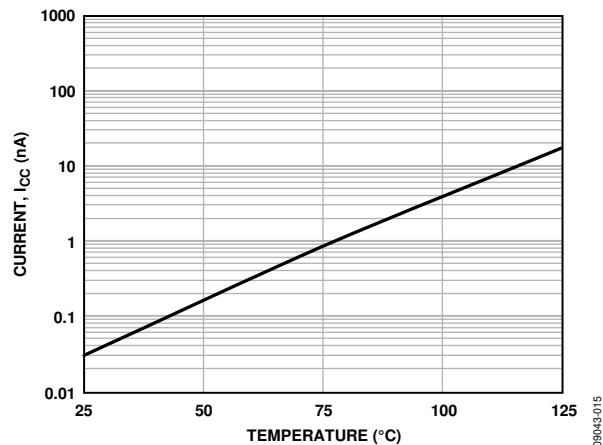


Figure 15. Collector-to-Collector Leakage Current vs. Temperature

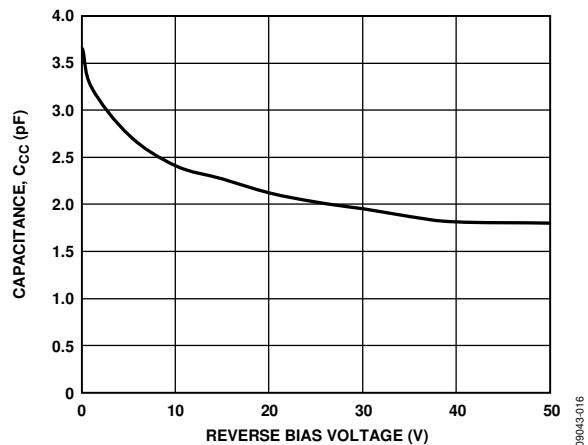


Figure 16. Collector-to-Collector Capacitance vs. Reverse Bias Voltage

PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS

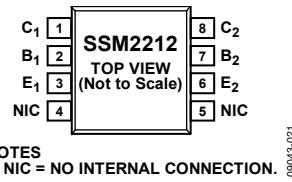


Figure 17. SOIC Pin Configuration

Table 5. SOIC Pin Function Descriptions

Pin No.	Mnemonic	Description
1	C ₁	Collector of Channel 1.
2	B ₁	Base of Channel 1.
3	E ₁	Emitter of Channel 1.
4, 5	NIC	No Internal Connection.
6	E ₂	Emitter of Channel 2.
7	B ₂	Base of Channel 2.
8	C ₂	Collector of Channel 2.

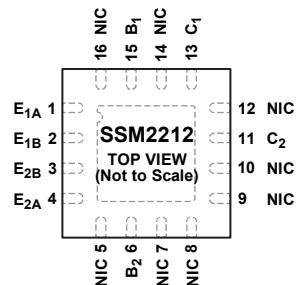


Figure 18. LFCSP Pin Configuration

Table 6. LFCSP Pin Function Descriptions

Pin No.	Mnemonic	Description
1	E _{1A}	Emitter of Channel 1. Must be connect to E _{1B} .
2	E _{1B}	Emitter of Channel 1. Must be connect to E _{1A} .
3	E _{2B}	Emitter of Channel 2. Must be connect to E _{2A} .
4	E _{2A}	Emitter of Channel 2. Must be connect to E _{2B} .
5, 7, 8, 9, 10, 12, 14, 16	NIC	No Internal Connection.
6	B ₂	Base of Channel 2.
11	C ₂	Collector of Channel 2.
13	C ₁	Collector of Channel 1.
15	B ₁	Base of Channel 1.
	EPAD	Exposed Pad. The exposed pad must be connected to the lowest potential.

APPLICATIONS INFORMATION

FAST LOGARITHMIC AMPLIFIER

The circuit in Figure 19 is a modification of a standard logarithmic amplifier configuration. Running the SSM2212 at 2.5 mA per side (full scale) allows a fast response with a wide dynamic range. The circuit has a 7 decade current range and a 5 decade voltage range, and it is capable of 2.5 µs settling time to 1% with a 1 V to 10 V step. The output follows the equation:

$$V_O = \frac{R3 + R2}{R2} \frac{kT}{q} \ln \frac{V_{REF}}{V_{IN}}$$

To compensate for the temperature dependence of the kT/q term, a resistor with a positive 0.35%/°C temperature coefficient is chosen for R_2 . The output is inverted with respect to the input and is nominally -1 V/decade using the component values indicated.

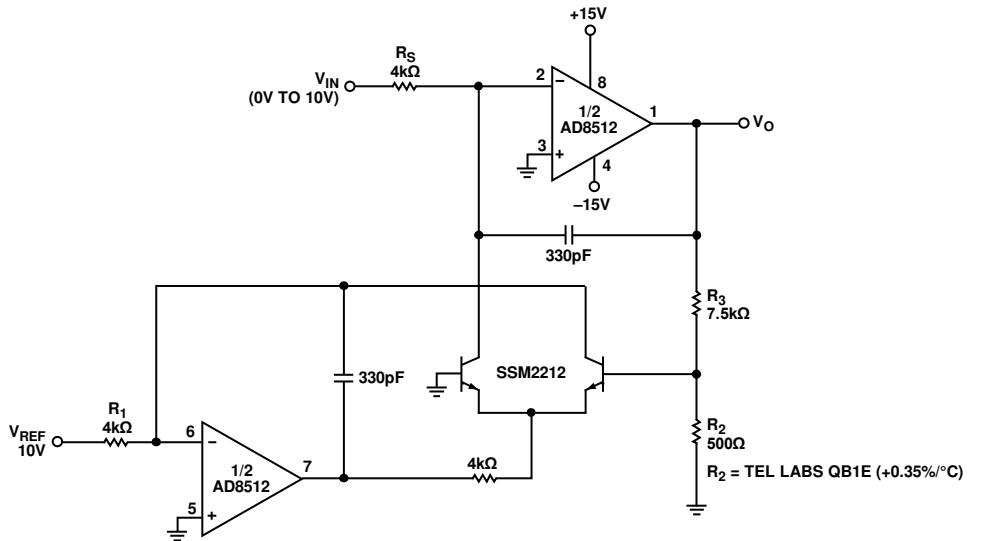
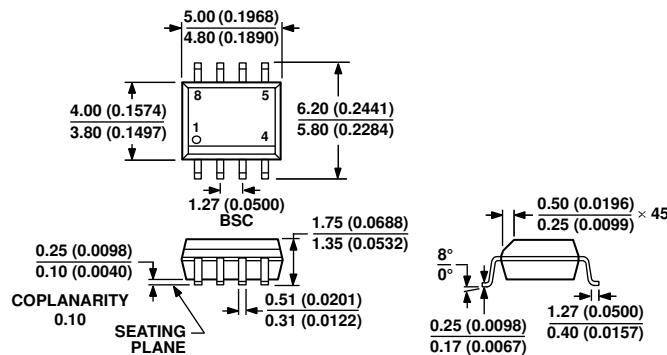


Figure 19. Fast Logarithmic Amplifier

09043-01B

OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MS-012-AA
 CONTROLLING DIMENSIONS ARE IN MILLIMETERS; INCH DIMENSIONS
 (IN PARENTHESES) ARE ROUNDED-OFF MILLIMETER EQUIVALENTS FOR
 REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.

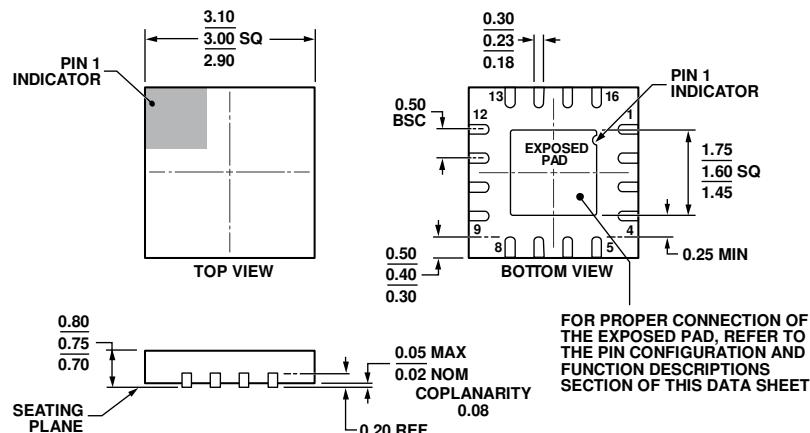
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Figure 20. 8-Lead Standard Small Outline Package [SOIC_N]

Narrow Body

(R-8)

Dimensions shown in millimeters (and inches)



COMPLIANT TO JEDEC STANDARDS MO-220-WEED-6.

Figure 21. 16-Lead Lead Frame Chip Scale Package [LFCSP_WQ]

3 mm × 3 mm Body, Very Very Thin Quad

(CP-16-22)

Dimensions shown in millimeters

08-16-2010-E

ORDERING GUIDE

Model ¹	Temperature Range	Package Description	Package Option	Branding
SSM2212RZ	-40°C to +85°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
SSM2212RZ-R7	-40°C to +85°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
SSM2212RZ-RL	-40°C to +85°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
SSM2212CPZ-R7	-40°C to +85°C	16-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-16-22	A3F
SSM2212CPZ-RL	-40°C to +85°C	16-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-16-22	A3F

¹ Z = RoHS Compliant Part.

NOTES