

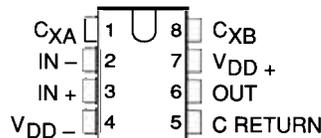
# ICL7652

## Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIER

D3343, OCTOBER 1989

- Extremely Low Offset Voltage . . . 5  $\mu\text{V}$  Max
- Extremely Low Change in Offset Voltage with Temperature . . . 0.003  $\mu\text{V}/^\circ\text{C}$  Typ
- Low Input Offset Current . . . 30 pA Max
- $A_{\text{VD}}$  . . . 120 dB Min
- CMRR and  $k_{\text{SVR}}$  . . . 110 dB Min
- Single-Supply Operation
- Common-Mode Input Voltage Range Includes the Negative Rail
- No Noise Degradation with External Capacitors Connected to  $V_{\text{DD-}}$

D, JG, OR P PACKAGE  
(TOP VIEW)



AVAILABLE OPTIONS

$T_A$	$V_{\text{IO max}}$ AT 25°C	PACKAGE		
		SMALL- OUTLINE (D)	CERAMIC DIP (JG)	PLASTIC DIP (P)
0°C to 70°C	5 $\mu\text{V}$	ICL7652D	ICL7652JG	ICL7652P

D package is available taped and reeled. Add "R" suffix to device type (i.e., ICL7652DR).

### description

The ICL7652 is a precision chopper-stabilized operational amplifier manufactured using Texas Instruments Advanced LinCMOS™ process. This process, in conjunction with unique chopper-stabilization circuitry, produces an operational amplifier whose performance matches or exceeds that of similar devices available today.

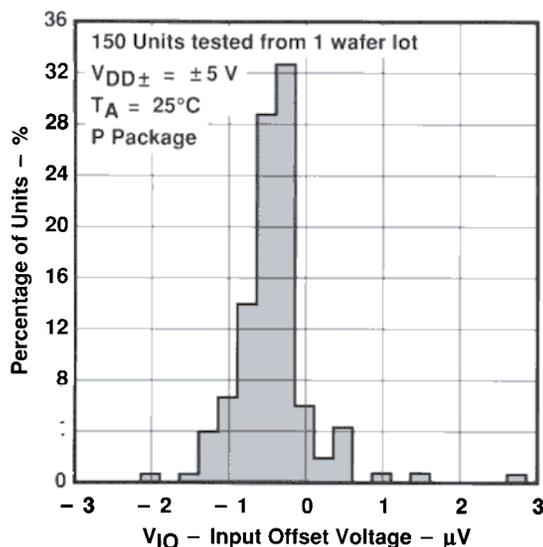
Chopper stabilization techniques make possible extremely high dc precision by continuously nulling input offset voltage even with variations in temperature, time, common-mode voltage, and supply voltage. Additionally, low-frequency noise voltage is significantly reduced. This precision, coupled with the extremely high input impedance of the CMOS input stage, makes the ICL7652 an ideal choice for low-level signal-processing applications such as strain gauges, thermocouples, and other transducer amplifiers.

The ICL7652 input common-mode range includes the negative rail, thereby providing superior performance in either single-supply or split-supply applications, even at supply voltages as low as  $\pm 1.9$  V. The ICL7652 also features fast overload recovery time. Two external capacitors are required to operate the device; however, the on-chip chopper control circuitry is transparent to the user.

The device inputs and output are designed to withstand  $-100$ -mA surge currents without sustaining latchup. Additionally, the ICL7652 incorporates internal ESD protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015; however, care should be exercised in handling these devices, as exposure to ESD may result in degradation of the device parametric performance.

The ICL7652 is characterized for operation from 0°C to 70°C.

DISTRIBUTION OF ICL7652  
INPUT OFFSET VOLTAGE



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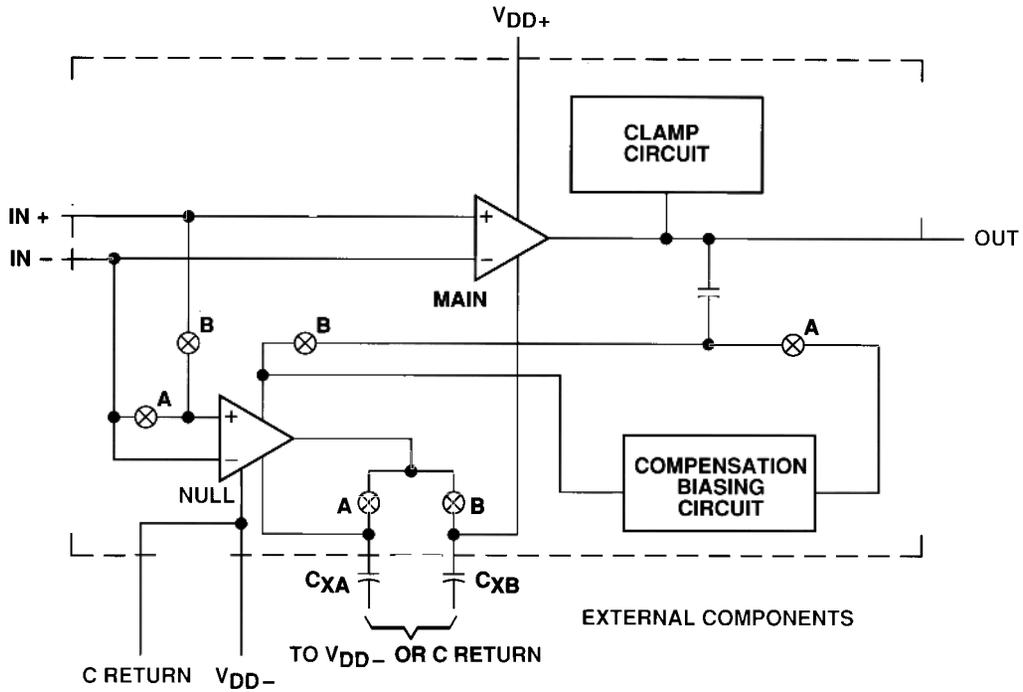


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**OPERATIONAL AMPLIFIER**

functional block diagram



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**absolute maximum ratings over operating free-air temperature range (unless otherwise noted)**

Supply voltage, $V_{DD+}$ (see Note 1) .....	8 V
Supply voltage, $V_{DD-}$ .....	-8 V
Differential input voltage (see Note 2) .....	$\pm 16$ V
Input voltage range, $V_I$ (any input) .....	$\pm 8$ V
Input current, $I_I$ (each input) .....	$\pm 5$ mA
Output current, $I_O$ .....	$\pm 50$ mA
Duration of short-circuit current at (or below) 25°C (see Note 3) .....	unlimited
Continuous total dissipation .....	See Dissipation Rating Table
Operating free-air temperature range, $T_A$ .....	0°C to 70°C
Storage temperature range .....	-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package .....	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package .....	300°C

- NOTES: 1. All voltage values, except differential voltages, are with respect to the midpoint between  $V_{DD+}$  and  $V_{DD-}$ .  
 2. Differential voltages are at the noninverting input with respect to the inverting input.  
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW
JG	1050 mW	8.4 mW/°C	672 mW
P	1000 mW	8.0 mW/°C	640 mW

**recommended operating conditions**

	MIN	MAX	UNIT
Supply voltage, $V_{DD}$	$\pm 1.9$	$\pm 8$	V
Common-mode input voltage, $V_{IC}$	$V_{DD-}$	$V_{DD+} - 1.9$	V
Operating free-air temperature, $T_A$	0	70	°C

# ICL7652

## Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIER

electrical characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS	$T_A^\dagger$	MIN	TYP	MAX	UNIT
$V_{IO}$	Input offset voltage	$V_{IC} = 0, R_S = 50\ \Omega$	25°C	0.6		5	$\mu\text{V}$
			Full range			7.25	
$\alpha_{VIO}$	Temperature coefficient of input offset voltage		25°C	0.003		0.05	$\mu\text{V}/^\circ\text{C}$
			Full range			0.06	
Input offset voltage long-term drift (see Note 4)			25°C	0.003		0.06	$\mu\text{V}/\text{mo}$
$I_{IO}$	Input offset current		25°C	2		30	$\text{pA}$
		Full range			100		
$I_{IB}$	Input bias current	25°C	4		30	$\text{pA}$	
		Full range			100		
$V_{ICR}$	Common-mode input voltage range	$R_S = 50\ \Omega$	Full range	-5 to 3.1			V
$V_{OM+}$	Maximum positive peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	4.7		4.8	V
			Full range	4.7			
$V_{OM-}$	Maximum negative peak output voltage swing	$R_L = 10\ \text{k}\Omega$	25°C	-4.7		-4.9	V
			Full range	-4.7			
$A_{VD}$	Large-signal differential voltage amplification	$V_O = \pm 4\ \text{V}, R_L = 10\ \text{k}\Omega$	25°C	120		150	dB
			Full range	114			
$f_{ch}$	Internal chopping frequency		25°C		450		Hz
CMRR	Common-mode rejection ratio	$V_O = 0, V_{IC} = V_{ICR\ \text{min}}, R_S = 50\ \Omega$	25°C	110		140	dB
			Full range	104			
$k_{SVR}$	Supply-voltage rejection ratio ( $\Delta V_{DD\pm} / \Delta V_{IO}$ )	$V_{DD\pm} = \pm 1.9\ \text{V to } \pm 8\ \text{V}, V_O = 0, R_S = 50\ \Omega$	25°C	110		135	dB
			Full range	104			
$I_{DD}$	Supply current	$V_O = 0, \text{ No load}$	25°C		1.5	2.4	mA
			Full range			3.5	

NOTE 4: Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at  $T_A = 150^\circ\text{C}$  extrapolated to  $T_A = 25^\circ\text{C}$  using the Arrhenius equation and assuming an activation energy of 0.96 eV.

operating characteristics at specified free-air temperature,  $V_{DD\pm} = \pm 5\text{ V}$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS	$T_A^\dagger$	MIN	TYP	MAX	UNIT
SR +	Positive slew rate at unity gain	$V_O = \pm 2.3\ \text{V}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C	2		2.8	$\text{V}/\mu\text{s}$
			Full range	1.5			
SR -	Negative slew rate at unity gain		25°C	2.3		3.1	$\text{V}/\mu\text{s}$
			Full range	1.8			
$V_n$	Equivalent input noise voltage (see Note 5)	$f = 10\ \text{Hz}$	25°C		94		$\text{nV}/\sqrt{\text{Hz}}$
		$f = 1\ \text{kHz}$	25°C		23		
$V_{N(PP)}$	Peak-to-peak equivalent input noise voltage	$f = 0\ \text{to } 1\ \text{Hz}$	25°C		0.8		$\mu\text{V}$
		$f = 0\ \text{to } 10\ \text{Hz}$	25°C		2.8		
$I_n$	Equivalent input noise current	$f = 1\ \text{kHz}$	25°C		0.004		$\text{pA}/\sqrt{\text{Hz}}$
Gain-bandwidth product		$f = 10\ \text{kHz}, R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		1.9		MHz
$\phi_m$	Phase margin at unity gain	$R_L = 10\ \text{k}\Omega, C_L = 100\ \text{pF}$	25°C		48°		

$^\dagger$ Full range is  $0^\circ\text{C}$  to  $70^\circ\text{C}$ .

NOTE 5: This parameter is tested on a sample basis for the ICL7652. For other test requirements, please contact the factory. This statement has no bearing on testing or nontesting of other parameters.

**ICL7652**  
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**OPERATIONAL AMPLIFIER**

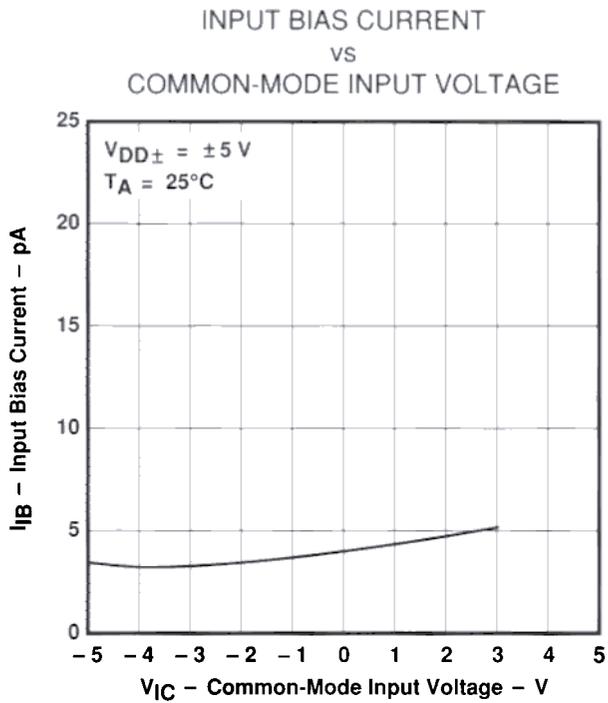
**TYPICAL CHARACTERISTICS**

table of graphs

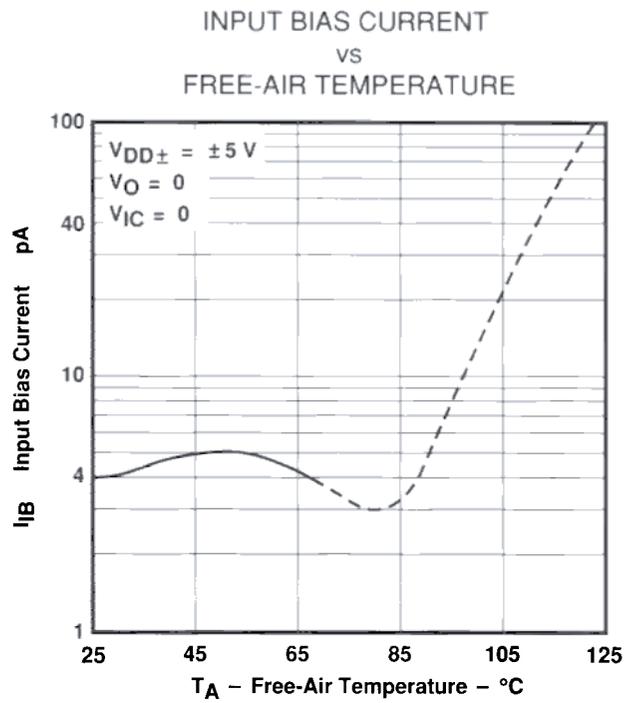
		<b>FIGURE</b>	
$I_{IB}$	Input bias current	vs Common-mode input voltage	1
		vs Temperature	2
$I_{IO}$	Input offset current	vs Temperature	3
$V_{O(PP)}$	Maximum peak-to-peak output voltage swing	vs Frequency	4
$V_{OM}$	Maximum peak output voltage swing	vs Output current	5, 6
		vs Temperature	7, 8
$A_{VD}$	Differential voltage amplification	vs Frequency	9
		vs Temperature	10
$I_{DD}$	Supply current	vs Supply voltage	11
		vs Temperature	12
$I_{OS}$	Short-circuit output current	vs Supply voltage	13
		vs Temperature	14
SR	Slew rate	vs Supply voltage	15
		vs Temperature	16
	Pulse response	Small-signal	17
		Large-signal	18
	Gain-bandwidth product	vs Supply voltage	19
		vs Temperature	20
$V_n$	Equivalent input noise voltage	vs Frequency	21
$\phi_m$	Phase margin	vs Supply voltage	22
		vs Temperature	23
		vs Load capacitance	24
	Phase shift	vs Frequency	9

**ICL7652**  
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**OPERATIONAL AMPLIFIER**

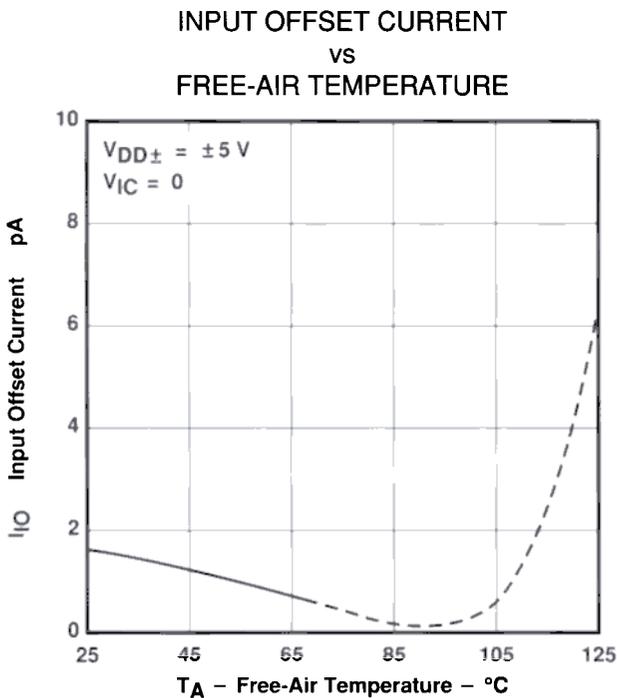
**TYPICAL CHARACTERISTICS†**



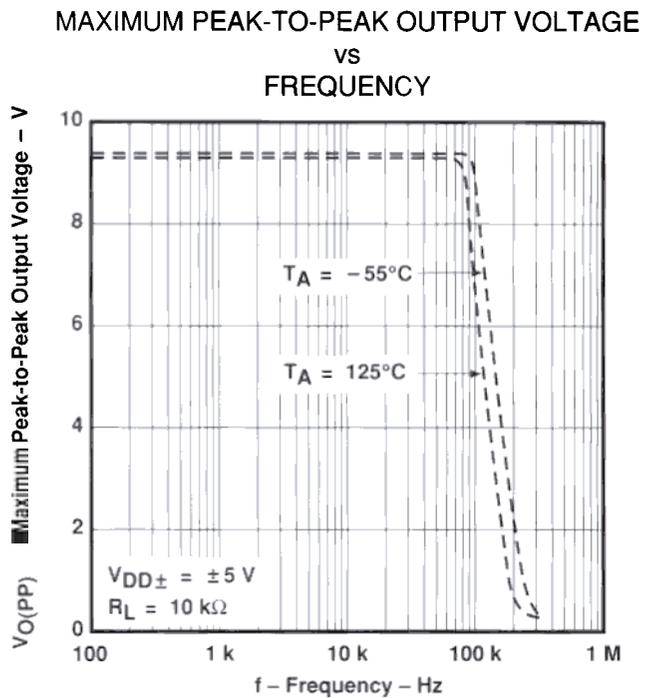
**FIGURE 1**



**FIGURE 2**



**FIGURE 3**



**FIGURE 4**

†Data at high and low temperatures are applicable to ICL7652 only within its noted operating free-air temperature range of 0°C to 70°C. The extended data have been retained to show typical performance of devices selected from the same process but characterized for operation from -55°C to 125°C.

**TYPICAL CHARACTERISTICS†**

MAXIMUM PEAK OUTPUT VOLTAGE  
 VS  
 OUTPUT CURRENT

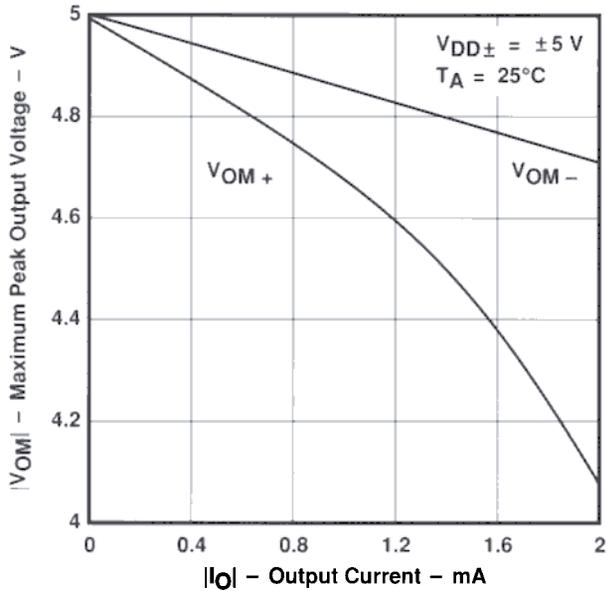


FIGURE 5

MAXIMUM PEAK OUTPUT VOLTAGE  
 VS  
 OUTPUT CURRENT

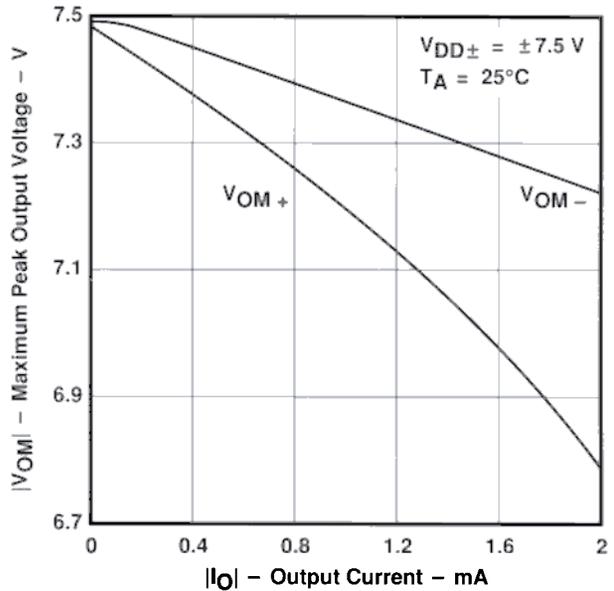


FIGURE 6

MAXIMUM PEAK OUTPUT VOLTAGE  
 VS  
 FREE-AIR TEMPERATURE

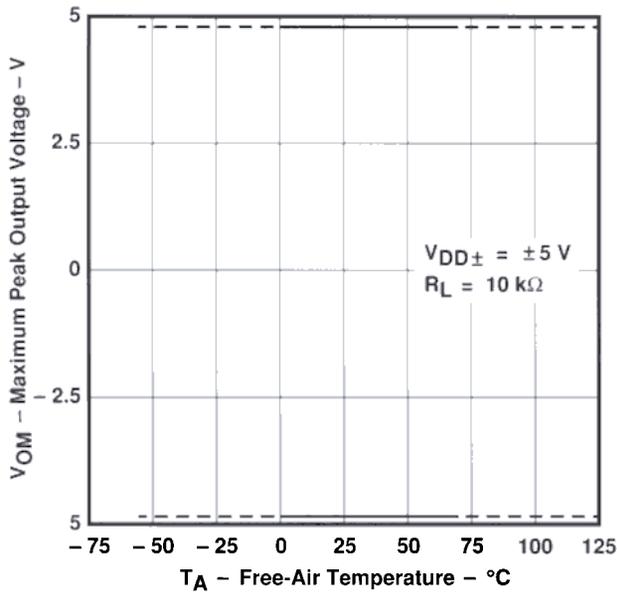


FIGURE 7

MAXIMUM PEAK OUTPUT VOLTAGE  
 VS  
 FREE-AIR TEMPERATURE

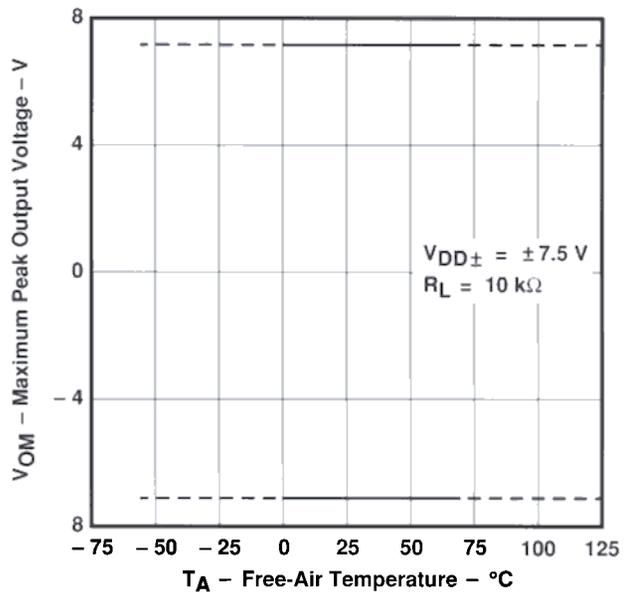


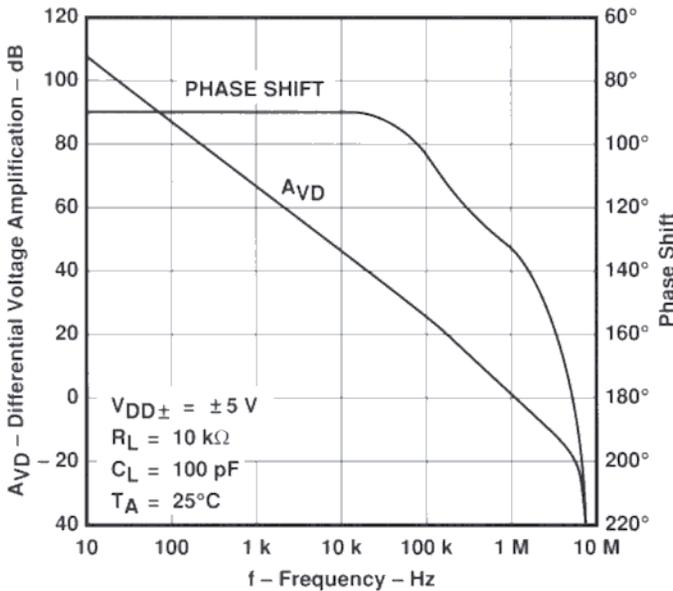
FIGURE 8

†Data at high and low temperatures are applicable to ICL7652 only within its noted operating free-air temperature range of 0°C to 70°C. The extended data have been retained to show typical performance of devices selected from the same process but characterized for operation from -55°C to 125°C.

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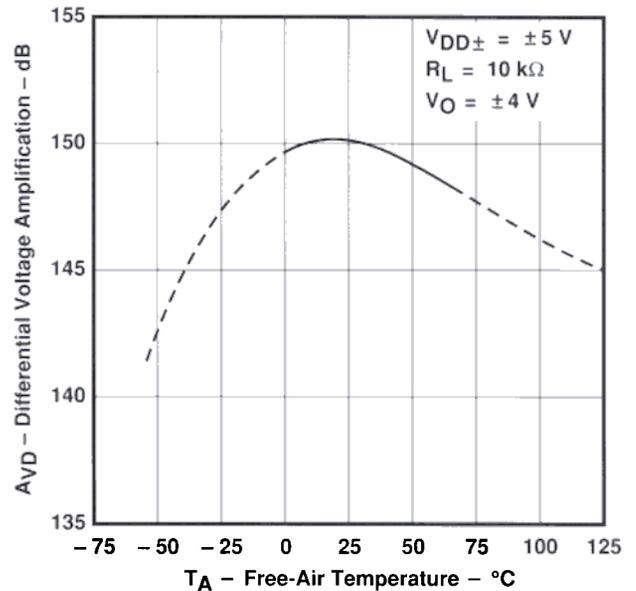
**TYPICAL CHARACTERISTICS†**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION and PHASE SHIFT vs FREQUENCY**



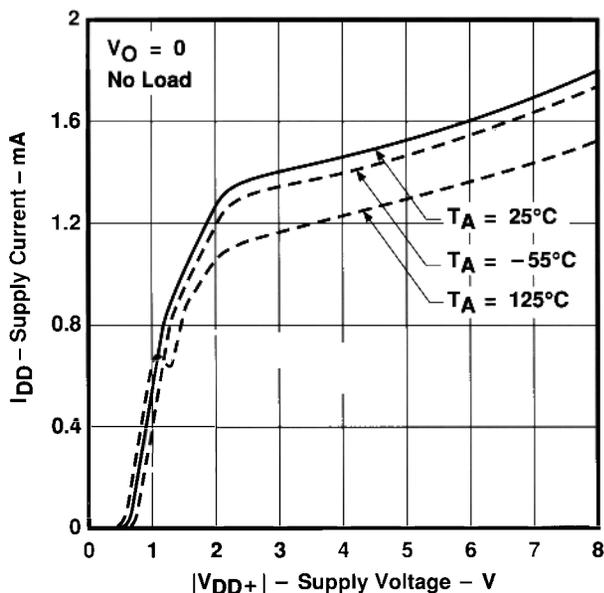
**FIGURE 9**

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION vs FREE-AIR TEMPERATURE**



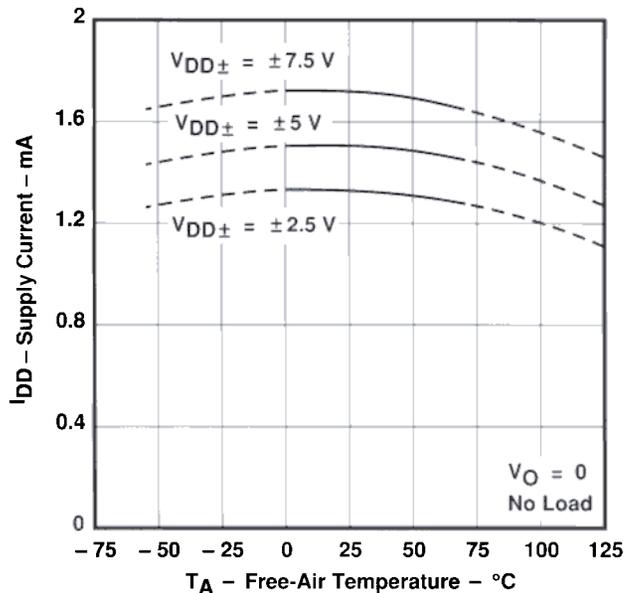
**FIGURE 10**

**SUPPLY CURRENT vs SUPPLY VOLTAGE**



**FIGURE 11**

**SUPPLY CURRENT vs FREE-AIR TEMPERATURE**

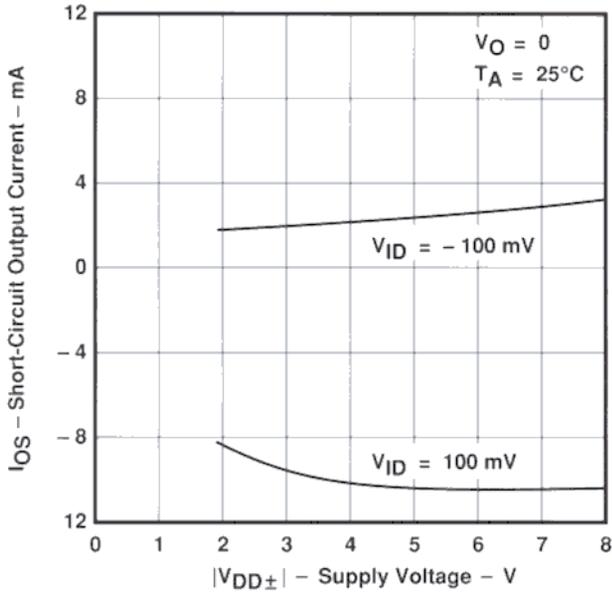


**FIGURE 12**

†Data at high and low temperatures are applicable to ICL7652 only within its noted operating free-air temperature range of 0°C to 70°C. The extended data have been retained to show typical performance of devices selected from the same process but characterized for operation from -55°C to 125°C.

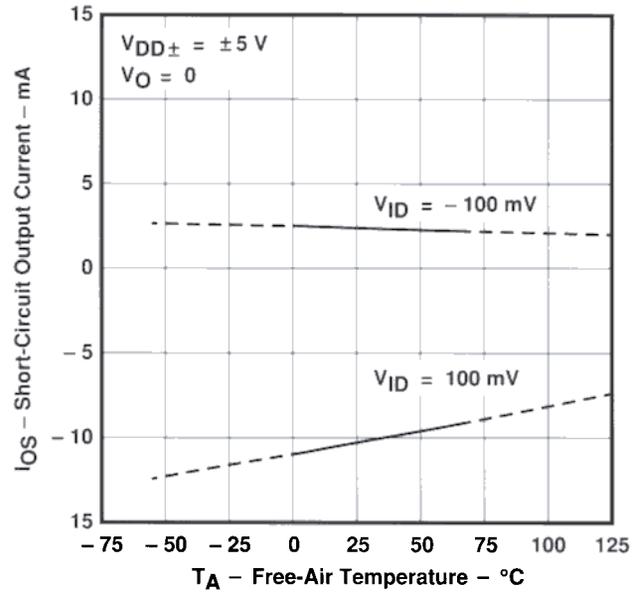
**TYPICAL CHARACTERISTICS†**

SHORT-CIRCUIT OUTPUT CURRENT  
 VS  
 SUPPLY VOLTAGE



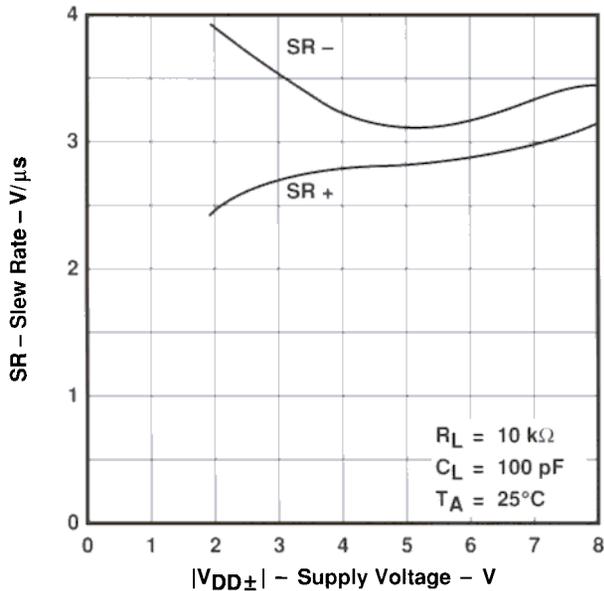
**FIGURE 13**

SHORT-CIRCUIT OUTPUT CURRENT  
 VS  
 FREE-AIR TEMPERATURE



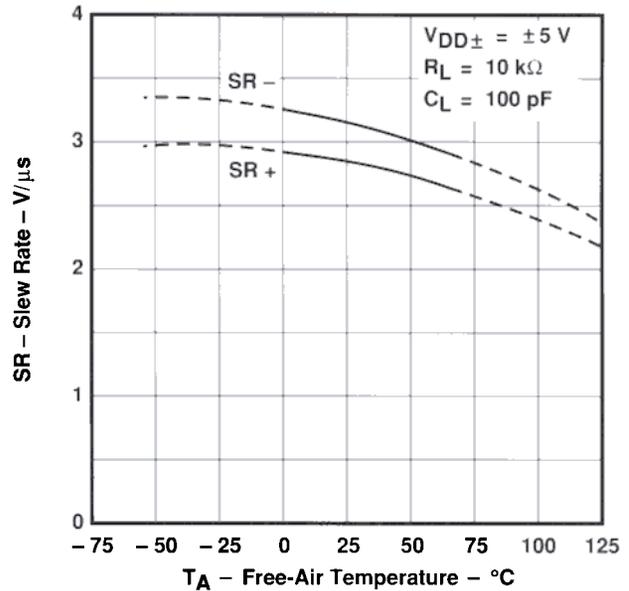
**FIGURE 14**

SLEW RATE  
 VS  
 SUPPLY VOLTAGE



**FIGURE 15**

SLEW RATE  
 VS  
 FREE-AIR TEMPERATURE



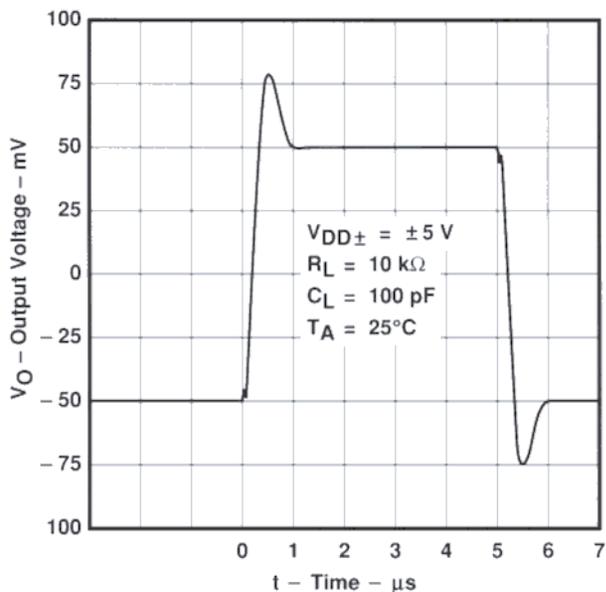
**FIGURE 16**

†Data at high and low temperatures are applicable to ICL7652 only within its noted operating free-air temperature range of 0°C to 70°C. The extended data have been retained to show typical performance of devices selected from the same process but characterized for operation from -55°C to 125°C.

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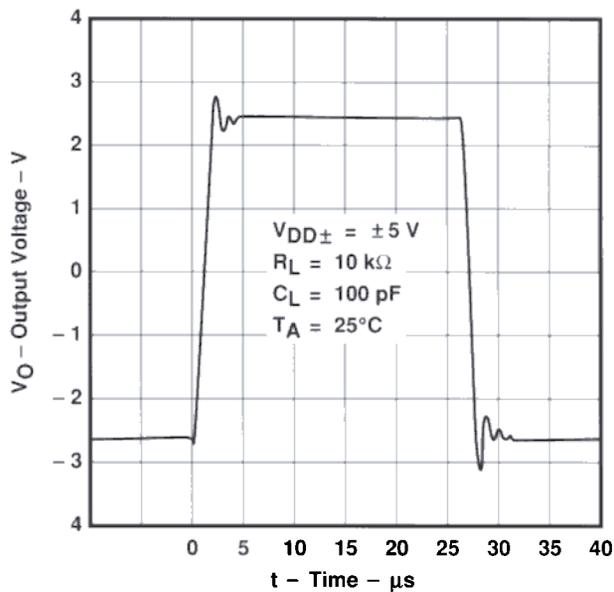
**TYPICAL CHARACTERISTICS†**

VOLTAGE-FOLLOWER  
 SMALL-SIGNAL  
 PULSE RESPONSE



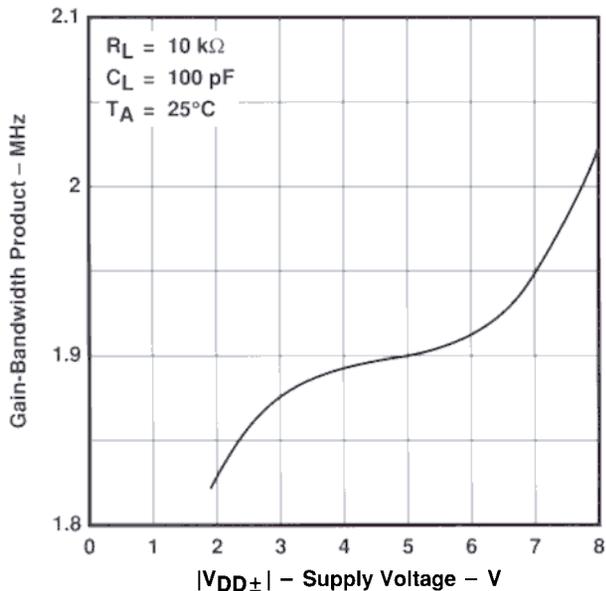
**FIGURE 17**

VOLTAGE-FOLLOWER  
 LARGE-SIGNAL  
 PULSE RESPONSE



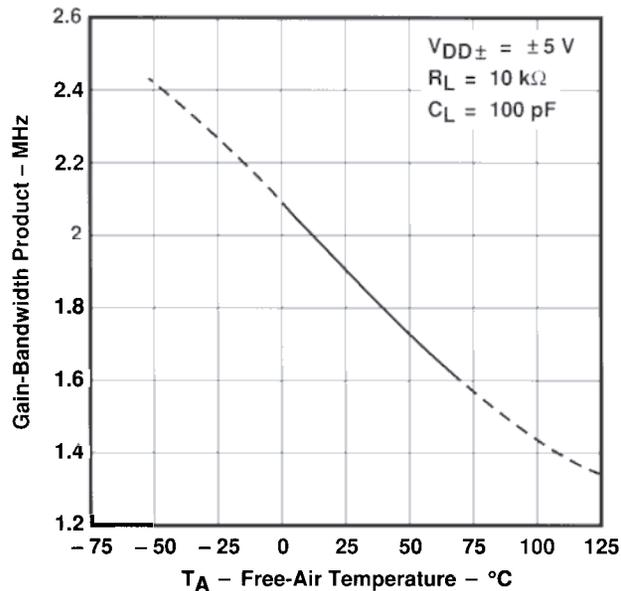
**FIGURE 18**

GAIN-BANDWIDTH PRODUCT  
 VS  
 SUPPLY VOLTAGE



**FIGURE 19**

GAIN-BANDWIDTH PRODUCT  
 VS  
 FREE-AIR TEMPERATURE



**FIGURE 20**

†Data at high and low temperatures are applicable to ICL7652 only within its noted operating free-air temperature range of 0°C to 70°C. The extended data have been retained to show typical performance of devices selected from the same process but characterized for operation from -55°C to 125°C.

**TYPICAL CHARACTERISTICS†**

EQUIVALENT INPUT NOISE VOLTAGE  
 VS  
 FREQUENCY

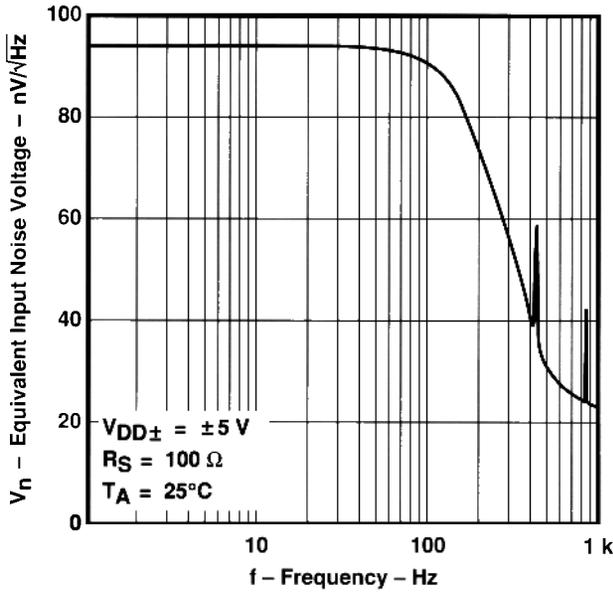


FIGURE 21

PHASE MARGIN  
 VS  
 SUPPLY VOLTAGE

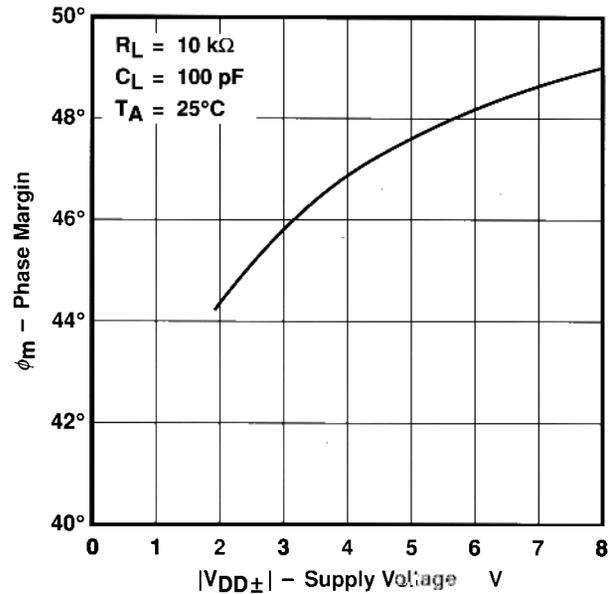


FIGURE 22

PHASE MARGIN  
 VS  
 FREE-AIR TEMPERATURE

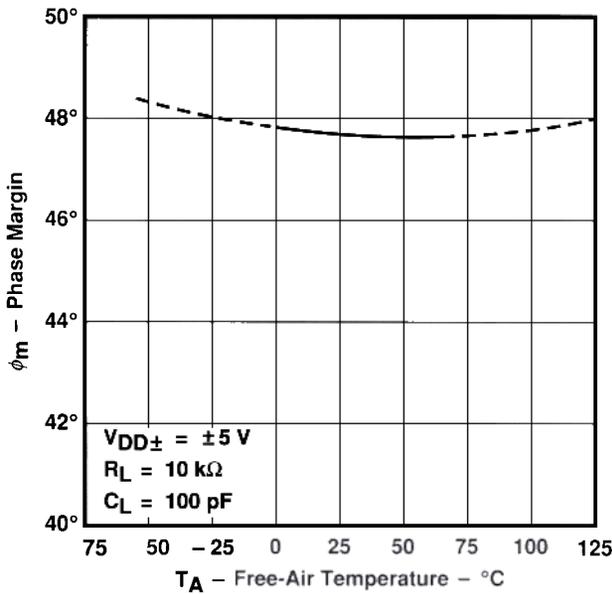


FIGURE 23

PHASE MARGIN  
 VS  
 LOAD CAPACITANCE

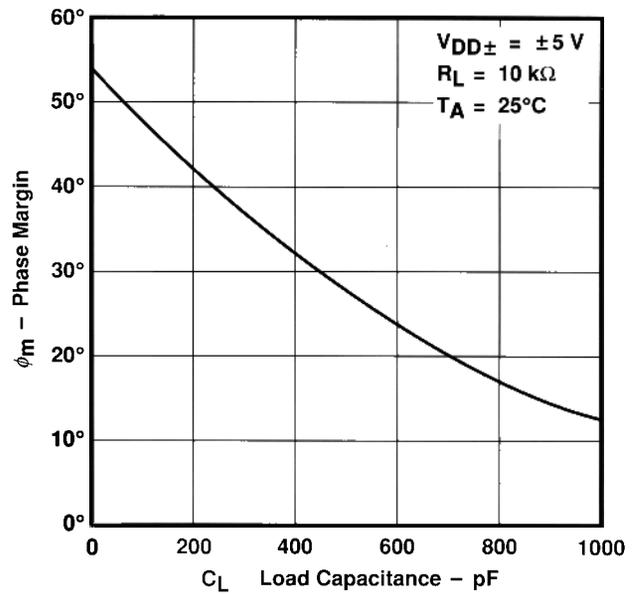


FIGURE 24

†Data at high and low temperatures are applicable to ICL7652 only within its noted operating free-air temperature range of 0°C to 70°C. The extended data have been retained to show typical performance of devices selected from the same process but characterized for operation from -55°C to 125°C.

# ICL7652

## Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED OPERATIONAL AMPLIFIER

### TYPICAL APPLICATION DATA

#### capacitor selection and placement

The two important factors to consider when selecting external capacitors  $C_{XA}$  and  $C_{XB}$  are leakage and dielectric absorption. Both factors can cause system degradation that negate the performance advantages realized by using the ICL7652.

Degradation from capacitor leakage becomes more apparent with increasing temperatures. Low-leakage capacitors and standoffs are recommended for operation at  $T_A = 125^\circ\text{C}$ . In addition, guardbands around the capacitor connections on both sides of the printed circuit board are recommended to alleviate problems caused by surface leakage on circuit boards.

Capacitors with high dielectric absorption tend to take several seconds to settle upon application of power, which directly affects input offset voltage. In applications needing fast settling of input offset voltage, it is recommended that high-quality film capacitors, such as mylar, polystyrene, or polypropylene, be used. In other applications, however, a ceramic or other low-grade capacitor may suffice.

Unlike many choppers available today, the ICL7652 is designed to function with values of  $C_{XA}$  and  $C_{XB}$  in the range of  $0.1\ \mu\text{F}$  to  $1\ \mu\text{F}$  without degradation to input offset voltage or input noise voltage. These capacitors should be located as close as possible to the  $C_{XA}$  and  $C_{XB}$  pins and returned to either the  $V_{DD-}$  pin or the C RETURN pin. Note that on many choppers, connecting these capacitors to the  $V_{DD-}$  pin causes degradation in noise performance, a problem that is eliminated on the ICL7652.

#### overload recovery/output clamp

When large differential input voltage conditions are applied to the ICL7652, the nulling loop attempts to prevent the output from saturating by driving  $C_{XA}$  and  $C_{XB}$  to internally-clamped voltage levels. Once the overdrive condition is removed, a period of time is required to allow the built-up charge to dissipate. This time period is defined as overload recovery time (see Figure 25). Typical overload recovery time for the ICL7652 is significantly faster than that of competitive products.

#### thermoelectric effects

To take advantage of the extremely low offset voltage temperature coefficient of the ICL7652, care must be taken to compensate for the thermoelectric effects present when two dissimilar metals are brought into contact with each other (such as device leads being soldered to a printed circuit board). It is not uncommon for dissimilar metal junctions to produce thermoelectric voltages in the range of several microvolts per degree Celsius (orders of magnitude greater than the  $0.01\text{-}\mu\text{V}/^\circ\text{C}$  typical of the ICL7652).

To help minimize thermoelectric effects, careful attention should be paid to component selection and circuit board layout. Avoid the use of nonsoldered connections (such as sockets, relays, switches, etc.) in the input signal path. Cancel thermoelectric effects by duplicating the number of components and junctions in each device input. The use of low-thermoelectric-coefficient components, such as wire-wound resistors, is also beneficial.

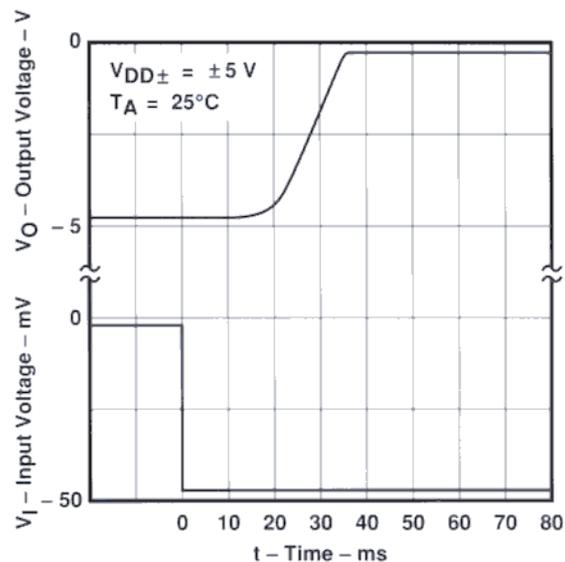


FIGURE 25. OVERLOAD RECOVERY

**TYPICAL APPLICATION DATA**

**avoiding latchup**

The ICL7652 inputs and output are designed to withstand  $-100\text{-mA}$  surge currents without sustaining latchup. However, because CMOS devices are susceptible to latchup due to their inherent parasitic thyristors, techniques to reduce the chance of latchup should be used whenever possible. Internal protection diodes should not be forward biased in normal operation. Applied input and output voltages should not exceed the supply voltage by more than  $300\text{ mV}$ . Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by using decoupling capacitors ( $0.1\text{ }\mu\text{F}$  typical) located across the supply rails as close to the device as possible.

The current path established if latchup occurs is usually between the supply rails and is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor. The chance of latchup occurring increases with increasing temperature and supply voltage.

**electrostatic discharge protection**

The ICL7652 incorporates internal ESD protection circuits that prevent functional failures at voltages at or below  $2000\text{ V}$ . Care should be exercised in handling these devices, as exposure to ESD may result in degradation of the device parametric performance.

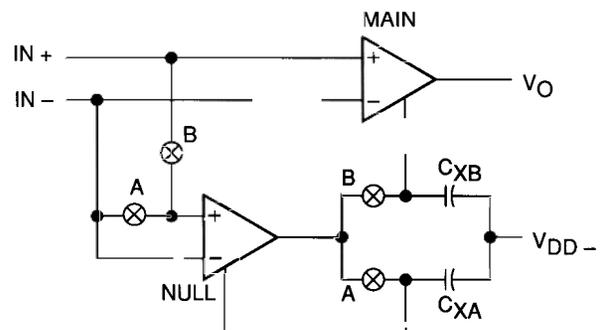
**theory of operation**

Chopper-stabilized operational amplifiers offer the best dc performance of any monolithic operational amplifier. This superior performance is the result of using two operational amplifiers – a main amplifier and a nulling amplifier – plus oscillator-controlled logic and two external capacitors to create a system that behaves as a single amplifier. With this approach, the ICL7652 achieves submicrovolt input offset voltage, submicrovolt noise voltage, and offset voltage variations with temperature in the  $\text{nV}/^\circ\text{C}$  range.

The ICL7652 on-chip control logic produces two dominant clock phases – a nulling phase and an amplifying phase. The term “chopper-stabilized” derives from the process of switching between these two clock phases. Figure 26 shows a simplified block diagram of the ICL7652. Switches A and B are make-before-break types. During the nulling phase, switch A is closed, shorting the nulling amplifier inputs together and allowing the nulling amplifier to reduce its own input offset voltage by feeding its output signal back to an inverting input node. Simultaneously, external capacitor  $C_{XA}$  stores the nulling potential to allow the offset voltage of the amplifier to remain nulled during the amplifying phase.

During the amplifying phase, switch B is closed, connecting the output of the nulling amplifier to a noninverting input of the main amplifier. In this configuration, the input offset voltage of the main amplifier is nulled. Also, external capacitor  $C_{XB}$  stores the nulling potential to allow the offset voltage of the main amplifier to remain nulled during the next nulling phase.

This continuous chopping process allows offset-voltage nulling during variations in time and temperature and over the common-mode input voltage range and power supply range. In addition, because the low-frequency signal path is through both the null and main amplifiers, extremely high gain is achieved.



**FIGURE 26. SIMPLIFIED BLOCK DIAGRAM**

**ICL7652**  
**Advanced LinCMOS™ PRECISION CHOPPER-STABILIZED**  
**OPERATIONAL AMPLIFIER**

**TYPICAL APPLICATION DATA**

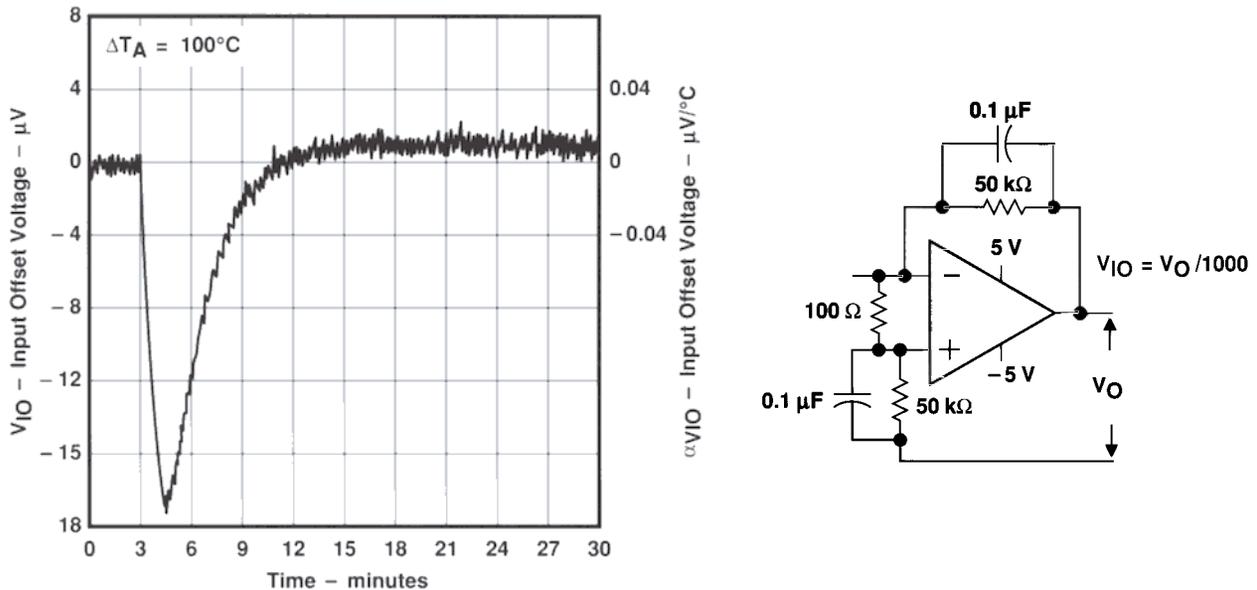
The low-frequency noise of a chopper amplifier depends on the magnitude of the component noise prior to chopping and the capability of the circuit to reduce this noise while chopping. The use of the Advanced LinCMOS™ process, with its low-noise analog MOS transistors and patent-pending input stage design, significantly reduces the input noise voltage.

The primary source of nonideal operation in chopper-stabilized amplifiers is error charge from the switches. As charge imbalance accumulates on critical nodes, input offset voltage can increase, especially with increasing chopping frequency. This problem has been significantly reduced in the ICL7652 by use of a patent-pending compensation circuit and the Advanced LinCMOS™ process.

The ICL7652 incorporates a feed-forward design that ensures continuous frequency response. Essentially, the gain magnitude of the nulling amplifier and compensation network crosses unity at the break frequency of the main amplifier. As a result, the high-frequency response of the system is the same as the frequency response of the main amplifier. This approach also ensures that the slewing characteristics remain the same during both the nulling and amplifying phases.

**temperature coefficient of input offset voltage**

Figure 27 shows the effects of package-induced thermal EMF. The ICL7652 can null only the offset voltage within its nulling loop. There are metal-to-metal junctions outside the nulling loop (bonding wires, solder joints, etc.) that produce EMF. In Figure 27, an ICL7652 was placed in an oven at 25°C at t = 0, biased up, and allowed to stabilize. At t = 3 min, the oven was turned on and allowed to rise in temperature to 125°C. As evidenced by the curve, the overall change in input offset voltage with temperature is much less than the specified maximum limit of 0.05 μV/°C.



**FIGURE 27. EFFECTS OF PACKAGE-INDUCED THERMAL EMF**

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
ICL7652P	LIFEBUY	PDIP	P	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	ICL7652P	

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

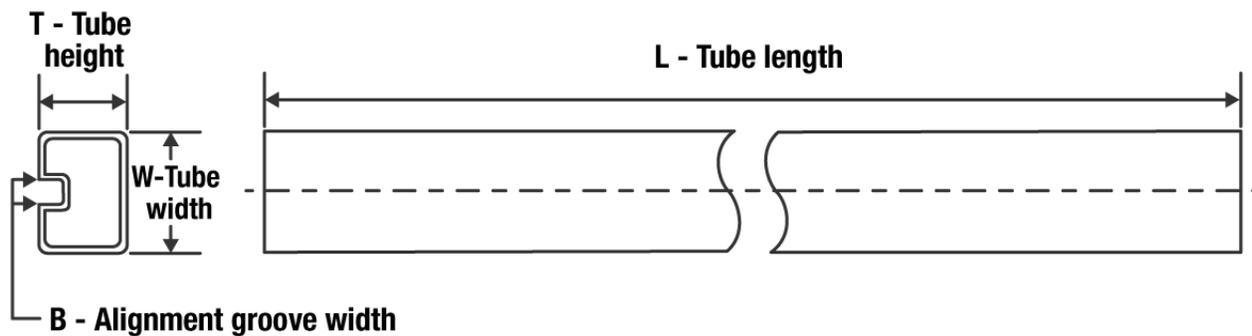
(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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**TUBE**


\*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
ICL7652P	P	PDIP	8	50	506	13.97	11230	4.32

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