Laser Driver with Extinction Ratio Control

General Description

The MAX3646 is a +3.3V laser driver designed for multirate transceiver modules with data rates from 155Mbps to 622Mbps. Lasers can be DC-coupled to the MAX3646 for reduced component count and ease of multirate operation.

Laser extinction ratio control (ERC) combines the features of automatic power control (APC), modulation compensation, and built-in thermal compensation. The APC loop maintains constant average optical power. Modulation compensation increases the modulation current in proportion to the bias current. These control loops, combined with thermal compensation, maintain a constant optical extinction ratio over temperature and lifetime.

The MAX3646 accepts differential data input signals. The wide 5mA to 60mA (up to 85mA AC-coupled) modulation current range and up to 100mA bias current range, make the MAX3646 ideal for driving FP/DFB lasers in fiber optic modules. External resistors set the required laser current levels. The MAX3646 provides transmit disable control (TX_DISABLE), single-point fault tolerance, bias-current monitoring, and photocurrent monitoring. The device also offers a latched failure output (TX_FAULT) to indicate faults, such as when the APC loop is no longer able to maintain the average optical power at the required level. The MAX3646 is compliant with the SFF-8472 transmitter diagnostic and SFP MSA timing requirements.

The MAX3646 is offered in a 4mm x 4mm, 24-pin thin QFN package and operates over the extended -40°C to $+85^{\circ}$ C temperature range.

Applications

Multirate OC-3 to OC-12 FEC Transceivers 125Mbps Ethernet SFP, GBIC, and 1 x 9 Transceivers

Typical Application Circuit appears at end of data sheet.

_Features

- Single +3.3V Power Supply
- 47mA Power-Supply Current
- 85mA Modulation Current
- 100mA Bias Current
- Automatic Power Control (APC)
- Modulation Compensation
- On-Chip Temperature Compensation
- Self-Biased Inputs for AC-Coupling
- ♦ Ground-Referenced Current Monitors
- Laser Shutdown and Alarm Outputs
- Enable Control and Laser Safety Feature

Ordering Information

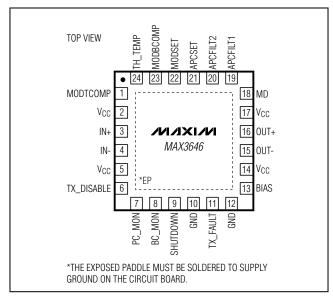
| PART | TEMP RANGE | PIN-PACKAGE |
|--------------|----------------|-----------------|
| MAX3646ETG | -40°C to +85°C | 24 Thin QFN-EP* |
| MAX3646ETG-T | -40°C to +85°C | 24 Thin QFN-EP* |
| MAX3646ETG+ | -40°C to +85°C | 24 Thin QFN-EP* |
| MAX3646ETG+T | -40°C to +85°C | 24 Thin QFN-EP* |

+Denotes a lead-free/RoHS-compliant package.

*EP = Exposed pad.

T = Tape and reel.

Pin Configuration



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For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

ABSOLUTE MAXIMUM RATINGS

Supply Voltage V_C.....-0.5V to +6.0V IN+, IN-, TX_DISABLE, TX_FAULT, SHUTDOWN, BC_MON, PC_MON, APCFILT1, APCFILT2, MD, TH_TEMP, MODTCOMP, MODBCOMP, MODSET, and APCSET Voltage......-0.5V to (V_{CC} + 0.5V) OUT+, OUT-, BIAS Current....-20mA to +150mA Continuous Power Dissipation (T_A = +70°C) 24-Pin TQFN (derate 27.8mW/°C above +70°C)2222mW

| Operating Junction Temperature Range55°C | to +150°C |
|--|-----------|
| Lead Temperature (soldering, 10s) | +300°C |
| Soldering Temperature (reflow) | |
| Lead (Pb)-free packages | +260°C |
| Packages containing lead (Pb) | +240°C |
| Storage Temperature Range55°C | |

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{CC} = +2.97V to +3.63V, T_A = -40°C to +85°C. Typical values are at V_{CC} = +3.3V, I_{BIAS} = 60mA, I_{MOD} = 60mA, T_A = +25°C, unless otherwise noted.) (Notes 1, 2)

| PARAMETER | SYMBOL | C | ONDITIONS | MIN | ТҮР | MAX | UNITS |
|---|-----------------|--|----------------------------|------|------|--|--------|
| POWER SUPPLY | | · | | | | | |
| Supply Current | ICC | (Note 3) | | | 47 | 60 | mA |
| Power-Supply Noise Rejection | PSNR | f ≤ 1MHz, 100mA _F | P-P (Note 4) | | 33 | | dB |
| I/O SPECIFICATIONS | • | | | | | | • |
| Differential Input Swing | VID | DC-coupled, Figu | re 1 | 0.2 | | 2.4 | Vp-p |
| Common-Mode Input | V _{CM} | | | 1.7 | | V _{CC} - V _{ID} / 4 | V |
| LASER BIAS | | • | | | | | |
| Bias-Current-Setting Range | | | | 1 | | 100 | mA |
| Bias Off Current | | TX_DISABLE = hig | gh | | | 0.1 | mA |
| Bias-Current Monitor Ratio | | IBIAS / IBC_MON | | 62 | 76 | 90 | mA/mA |
| LASER MODULATION | | • | | | | | |
| Modulation Current-Setting Range | IMOD | (Note 5) | (Note 5) | | | 85 | mA |
| Output Edge Speed | | 20% to 80% (Notes 6, 7) | $5mA \le I_{MOD} \le 85mA$ | | 100 | 200 | ps |
| Output Overshoot/Undershoot | | (Note 7) | | | ±6 | | % |
| Random Jitter | | (Notes 6, 7) | | | 1.1 | 2.5 | psrms |
| | | 622Mbps, 5mA \leq I _{MOD} \leq 85mA | | | 24 | 46 | |
| Deterministic Jitter (Notes 6, 8) | | 155Mbps, 5mA ≤ I _{MOD} ≤ 85mA | | | 45 | 100 | psp-p |
| Modulation-Current Temperature Stability | | (N_{Label}, C) 5mA $\leq I_{\text{MOD}} \leq 10$ mA | | ±175 | ±600 | 0.0 | |
| | | (Note 6) $10mA < I_{MOD} \le 85mA$ | | | ±125 | ±480 | ppm/°C |
| Madulation Current Sotting Free | | | | | | ±20 | % |
| Modulation-Current-Setting Error | | | | | | ±15 | 70 |
| Modulation Off Current | | TX_DISABLE = hig | gh | | | 0.1 | mA |

ELECTRICAL CHARACTERISTICS (continued)

(V_{CC} = +2.97V to +3.63V, T_A = -40°C to +85°C. Typical values are at V_{CC} = +3.3V, I_{BIAS} = 60mA, I_{MOD} = 60mA, T_A = +25°C, unless otherwise noted.) (Notes 1, 2)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|-----------------|--|-----------------------|------|------|--------|
| AUTOMATIC POWER AND EXTIN | ICTION RA | TIO CONTROLS | | | | • |
| Monitor-Diode Input Current Range | I _{MD} | Average current into the MD pin | 18 | | 1500 | μA |
| MD Pin Voltage | | | | | 1.4 | V |
| MD Current Monitor Ratio | | IMD / IPC_MON | 0.85 | 0.93 | 1.15 | mA/mA |
| APC Loop Time Constant | | $C_{APC_FILT} = 0.01 \mu F$, $\Delta I_{MD} / \Delta I_{BIAS} = 1/70$ | | 3.3 | | μs |
| APC Setting Stability | | (Note 6) | | ±100 | ±480 | ppm/°C |
| APC Setting Accuracy | | $T_A = +25^{\circ}C$ | | | ±15 | % |
| I _{MOD} Compensation-Setting Range by Bias | К | $K = \Delta I_{MOD} / \Delta I_{BIAS}$ | 0 | | 1.5 | mA/mA |
| I _{MOD} Compensation-Setting Range by Temperature | TC | $TC = \Delta I_{MOD} / \Delta T \text{ (Note 6)}$ | 0 | | 1.0 | mA/°C |
| Threshold-Setting Range for Temperature Compensation | T _{TH} | (Note 6) | +10 | | +60 | °C |
| LASER SAFETY AND CONTROL | | | | | | • |
| Bias and Modulation Turn-Off Delay | | $C_{APC_FILT} = 0.01 \mu F$, $\Delta I_{MD} / \Delta I_{BIAS} = 1/80$ (Note 6) | | | 5 | μs |
| Bias and Modulation Turn-On Delay | | $C_{APC}_{FILT} = 0.01 \mu F$, $\Delta I_{MD} / \Delta I_{BIAS} = 1/80$ (Note 6) | | | 600 | μs |
| Threshold Voltage at Monitor Pins | VREF | Figure 5 | 1.14 | 1.3 | 1.39 | V |
| INTERFACE SIGNALS | • | • | • | | | • |
| TX_DISABLE Input High | V _{HI} | | 2.0 | | | V |
| TX_DISABLE Input Low | VLO | $R_{PULL} = 45k\Omega$ (typical) | | | 0.8 | V |
| | | V _{HI} = V _{CC} | | | 15 | |
| TX_DISABLE Input Current | | $V_{LO} = GND$ | | -70 | -140 | μA |
| TX_FAULT Output Low | | Sinking 1mA, open collector | | | 0.4 | V |
| Shutdown Output High | | Sourcing 100µA | V _{CC} - 0.4 | | | V |
| Shutdown Output Low | | Sinking 100µA | | | 0.4 | V |

Note 1: AC characterization is performed using the circuit in Figure 2 using a PRBS 2²³ - 1 or equivalent pattern.

Note 2: Specifications at -40°C are guaranteed by design and characterization.

Note 3: Excluding IBIAS and IMOD. Input data is AC-coupled. TX_FAULT open, SHUTDOWN open.

Note 4: Power-supply noise rejection (PSNR) = $20\log_{10}(V_{noise (on VCC)} / \Delta V_{OUT})$. V_{OUT} is the voltage across the 15 Ω load when IN+ is high.

Note 5: The minimum required voltage at the OUT+ and OUT- pins is +0.75V.

Note 6: Guaranteed by design and characterization.

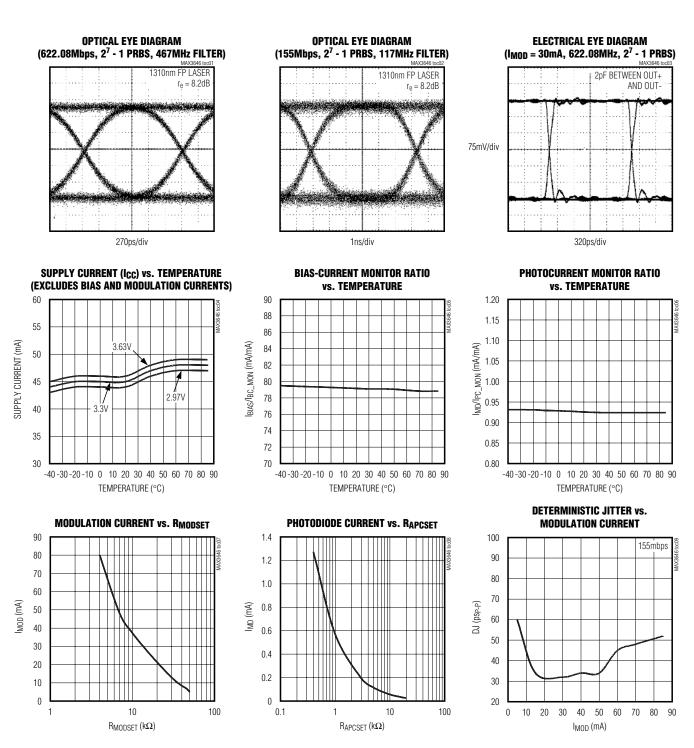
Note 7: Tested with 00001111 pattern at 622Mbps.

Note 8: DJ includes pulse-width distortion (PWD).

Typical Operating Characteristics

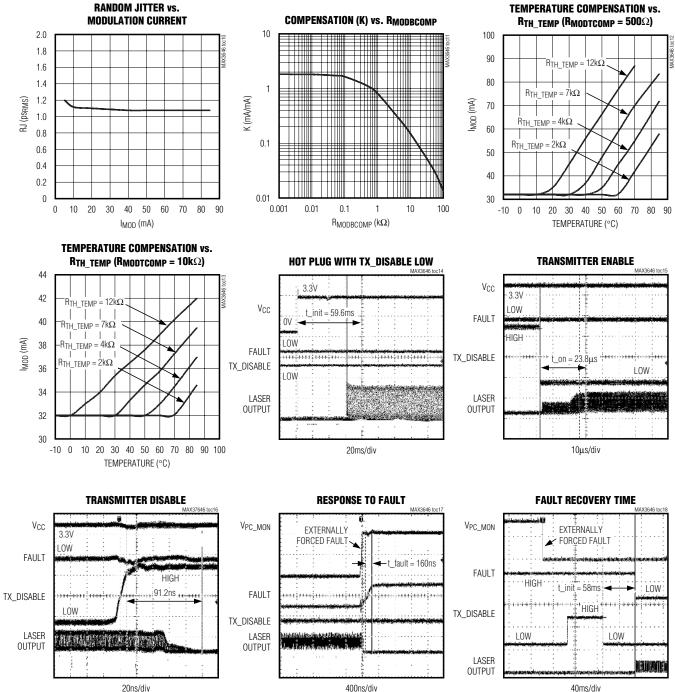
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 $(V_{CC} = +3.3V, C_{APC} = 0.01\mu$ F, $I_{BIAS} = 20$ mA, $I_{MOD} = 30$ mA, $T_A = +25$ °C, unless otherwise noted.)



Typical Operating Characteristics (continued)

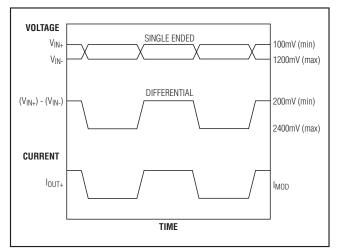
(V_{CC} = +3.3V, C_{APC} = 0.01μ F, I_{BIAS} = 20mA, I_{MOD} = 30mA, T_A = +25°C, unless otherwise noted.)



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_____Pin Description

| PIN | NAME | FUNCTION |
|--------------|------------|--|
| 1 | MODTCOMP | Modulation-Current Compensation from Temperature. A resistor at this pin sets the temperature coefficient of the modulation current when above the threshold temperature. Leave open for zero temperature compensation. |
| 2, 5, 14, 17 | Vcc | +3.3V Supply Voltage |
| 3 | IN+ | Noninverted Data Input |
| 4 | IN- | Inverted Data Input |
| 6 | TX_DISABLE | Transmitter Disable, TTL. Laser output is disabled when TX_DISABLE is asserted high or left unconnected. The laser output is enabled when this pin is asserted low. |
| 7 | PC_MON | Photodiode-Current Monitor Output. Current out of this pin develops a ground-referenced voltage across an external resistor that is proportional to the monitor diode current. |
| 8 | BC_MON | Bias-Current Monitor Output. Current out of this pin develops a ground-referenced voltage across an external resistor that is proportional to the bias current. |
| 9 | SHUTDOWN | Shutdown Driver Output. Voltage output to control an external transistor for optional shutdown circuitry. |
| 10, 12 | GND | Ground |
| 11 | TX_FAULT | Open-Collector Transmit Fault Indicator (Table 1) |
| 13 | BIAS | Laser Bias-Current Output |
| 15 | OUT- | Inverted Modulation-Current Output. IMOD flows into this pin when input data is low. |
| 16 | OUT+ | Noninverted Modulation-Current Output. IMOD flows into this pin when input data is high. |
| 18 | MD | Monitor Photodiode Input. Connect this pin to the anode of a monitor photodiode. A capacitor to ground is required to filter the high-speed AC monitor photocurrent. |
| 19 | APCFILT1 | Connect a capacitor (CAPC) between pin 19 (APCFILT1) and pin 20 (APCFILT2) to set the dominant pole of the APC feedback loop. |
| 20 | APCFILT2 | (See Pin 19) |
| 21 | APCSET | A resistor connected from this pin to ground sets the desired average optical power. The total capacitive load at the APCSET pin should be no more than 10pF. Minimize metal resistance for ground connections. |
| 22 | MODSET | A resistor connected from this pin to ground sets the desired constant portion of the modulation current. The total capacitive load at the MODSET pin should be no more than 10pF. Minimize metal resistance for ground connections. |
| 23 | MODBCOMP | Modulation-Current Compensation from Bias. Couples the bias current to the modulation current. Mirrors IBIAS through an external resistor. Leave open for zero-coupling. |
| 24 | TH_TEMP | Threshold for Temperature Compensation. A resistor at this pin programs the temperature above which compensation is added to the modulation current. |
| _ | EP | Exposed Pad. Solder the exposed pad to the circuit board ground for specified thermal and electrical performance. |



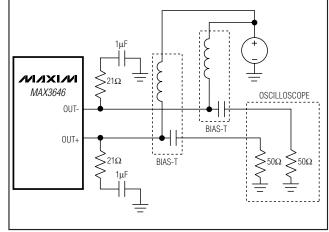


Figure 1. Required Input Signal and Output Polarity

Figure 2. Test Circuit for Characterization

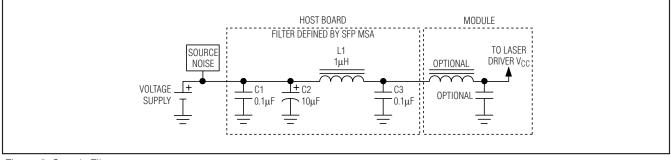


Figure 3. Supply Filter

Detailed Description

The MAX3646 laser driver consists of three main parts: a high-speed modulation driver, biasing block with ERC, and safety circuitry. The circuit design is optimized for high-speed, low-voltage (+3.3V) operation (Figure 4).

High-Speed Modulation Driver

The output stage is composed of a high-speed differential pair and a programmable modulation current source. The MAX3646 is optimized for driving a 15Ω load. The minimum instantaneous voltage required at OUT- is 0.7V for modulation currents up to 60mA and 0.75V for currents from 60mA to 85mA. Operation above 60mA can be accomplished by AC-coupling or with sufficient voltage at the laser to meet the driver output voltage requirement.

To interface with the laser diode, a damping resistor (R_D) is required. The combined resistance damping

resistor and the equivalent series resistance (ESR) of the laser diode should equal 15Ω . To further damp aberrations caused by laser diode parasitic inductance, an RC shunt network may be necessary. Refer to Application Note 274: *HFAN-02.0: Interfacing Maxim Laser Drivers with Laser Diodes* for more information.

Any capacitive load at the cathode of a laser diode degrades optical output performance. Because the BIAS output is directly connected to the laser cathode, minimize the parasitic capacitance associated with the pin by using an inductor to isolate the BIAS pin parasitics form the laser cathode.

Extinction Ratio Control

The extinction ratio (r_e) is the laser on-state power divided by the off-state power. Extinction ratio remains constant if peak-to-peak and average power are held constant:

$$r_e = (2P_{AVG} + P_{P-P}) / (2P_{AVG} - P_{P-P})$$



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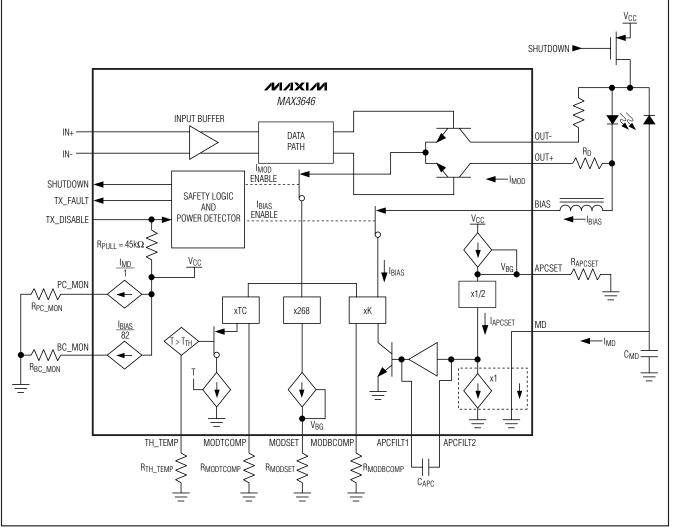


Figure 4. Functional Diagram

Average power is regulated using APC, which keeps constant current from a photodiode coupled to the laser. Peak-to-peak power is maintained by compensating the modulation current for reduced slope efficiency (η) of laser over time and temperature:

$$P_{AVG} = I_{MD}/\rho_{MON}$$

 $P_{P-P} = \eta \times I_{MOD}$

Modulation compensation from bias increases the modulation current by a user-selected proportion (K) needed to maintain peak-to-peak laser power as bias current increases with temperature. Refer to Application Note 1119: *HFAN-02.2.1: Maintaining the Extinction Ratio of Optical Transmitters Using K-Factor Control* for details:

$\mathsf{K} = \Delta \mathsf{I}_{\mathsf{MOD}} \, / \, \Delta \mathsf{I}_{\mathsf{BIAS}}$

This provides a first-order approximation of the current increase needed to maintain peak-to-peak power. Slope efficiency decreases more rapidly as temperature increases. The MAX3646 provides additional temperature compensation as temperature increases past a user-defined threshold (T_{TH}).



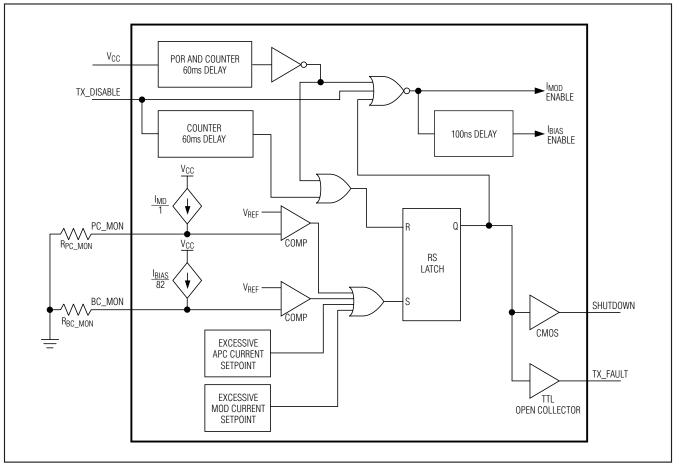


Figure 5. Simplified Safety Circuit

Table 1. Typical Fault Conditions

| 1 | If any of the I/O pins are shorted to GND or V _{CC} (single-point failure; see Table 2), and the bias current or the photocurrent exceeds the programmed threshold. |
|---|--|
| 2 | End-of-life (EOL) condition of the laser diode. The bias current and/or the photocurrent exceed the programmed threshold. |
| 3 | Laser cathode is grounded and photocurrent exceeds the programming threshold. |
| 4 | No feedback for the APC loop (broken interconnection, defective monitor photodiode), and the bias current exceeds the programmed threshold. |

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Table 2. Circuit Responses to Various Single-Point Faults

| PIN | CIRCUIT RESPONSE TO OVERVOLTATGE OR SHORT TO V _{CC} | CIRCUIT RESPONSE TO UNDERVOLTAGE OR SHORT TO GROUND |
|--|---|---|
| TX_FAULT | Does not affect laser power. | Does not affect laser power. |
| TX_DISABLE | Modulation and bias currents are disabled. | Normal condition for circuit operation. |
| IN+ | The optical average power increases and a fault occurs if VPC_MON exceeds the threshold. The APC loop responds by decreasing the bias current. | The optical average power decreases and the APC loop responds by increasing the bias current. A fault state occurs if V_{BC_MON} exceeds the threshold voltage. |
| IN- | The optical average power decreases and the APC loop responds by increasing the bias current. A fault state occurs if V_{BC_MON} exceeds the threshold voltage. | The optical average power increases and a fault occurs if V _{PC_MON} exceeds the threshold. The APC loop responds by decreasing the bias current. |
| MD | This disables bias current. A fault state occurs. | The APC circuit responds by increasing the bias current until a fault is detected, then a fault* state occurs. |
| SHUTDOWN | Does not affect laser power. If the shutdown circuitry is used, the laser current is disabled. | Does not affect laser power. |
| BIAS | In this condition, the laser forward voltage is 0V and no light is emitted. | Fault state* occurs. If the shutdown circuitry is used, the laser current is disabled. |
| OUT+ | The APC circuit responds by increasing the bias current until a fault is detected, then a fault state* occurs. | Fault state* occurs. If the shutdown circuitry is used, the laser current is disabled. |
| OUT- | Does not affect laser power. | Does not affect laser power. |
| PC_MON | Fault state* occurs. | Does not affect laser power. |
| BC_MON | Fault state* occurs. | Does not affect laser power. |
| APCFILT1 IBIAS increases until VBC_MON exceeds the threshold voltage. IBIAS increases until VBC_MON exceeds the threshold voltage. | | I_{BIAS} increases until V_{BC_MON} exceeds the threshold voltage. |
| APCFILT2 | IBIAS increases until VBC_MON exceeds the threshold voltage. | I_{BIAS} increases until V_{BC_MON} exceeds the threshold voltage. |
| MODSET | Does not affect laser power. | Fault state* occurs. |
| APCSET | Does not affect laser power. | Fault state* occurs. |

*A fault state asserts the TX_FAULT pin, disables the modulation and bias currents, and asserts the SHUTDOWN pin.

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| PARAMETER | SYMBOL | RELATION |
|------------------------------|----------------|---|
| Average power | Pavg | $P_{AVG} = (P_0 + P_1) / 2$ |
| Extinction ratio | r _e | $r_{e} = P_{1} / P_{0}$ |
| Optical power of a one | P ₁ | $P_1 = 2P_{AVG} \times r_e / (r_e + 1)$ |
| Optical power of a zero | P ₀ | $P_0 = 2P_{AVG} / (r_e + 1)$ |
| Optical amplitude | Pp-p | $P_{P-P} = P_1 - P_0$ |
| Laser slope efficiency | η | $\eta = P_{P-P} / I_{MOD}$ |
| Modulation current | IMOD | $I_{MOD} = P_{P-P} / \eta$ |
| Threshold current | ITH | P ₀ at I ≥ I _{TH} |
| Bias current (AC-coupled) | IBIAS | $I_{BIAS} \ge I_{TH} + I_{MOD} / 2$ |
| Laser to monitor transfer | ρμον | I _{MD} / Pavg |

Table 3. Optical Power Relations

Note: Assuming a 50% average input duty cycle and mark density.

Safety Circuitry

The safety circuitry contains a disable input (TX_DISABLE), a latched fault output (TX_FAULT), and fault detectors (Figure 5). This circuitry monitors the operation of the laser driver and forces a shutdown if a fault is detected (Table 1). The TX_FAULT pin should be pulled high with a 4.7k Ω to 10k Ω resistor to V_{CC} as required by the SFP MSA. A single-point fault can be a short to V_{CC} or GND. See Table 2 to view the circuit response to various single-point failure. The transmit fault condition is latched until reset by a toggle or TX_DISABLE or V_{CC}. The laser driver offers redundant laser diode shutdown through the optional shutdown circuitry as shown in the *Typical Application Circuit*. This shutdown transistor prevents a single-point fault at the laser from creating an unsafe condition.

Safety Circuitry Current Monitors

The MAX3646 features monitors (BC_MON, PC_MON) for bias current (I_{BIAS}) and photocurrent (I_{MD}). The monitors are realized by mirroring a fraction of the currents and developing voltages across external resistors connected to ground. Voltages greater than V_{REF} at PC_MON or BC_MON result in a fault state. For example, connecting a 100 Ω resistor to ground at each monitor output gives the following relationships:

 $V_{BC}MON = (I_{BIAS} / 82) \times 100\Omega$

 $V_{PC}MON = I_{MD} \times 100\Omega$

External sense resistors can be used for high-accuracy measurement of bias and photodiode currents. On-chip

isolation resistors are included to reduce the number of components needed to implement this function.

_Design Procedure

When designing a laser transmitter, the optical output is usually expressed in terms of average power and extinction ratio. Table 3 shows relationships that are helpful in converting between the optical average power and the modulation current. These relationships are valid if the mark density and duty cycle of the optical waveform are 50%.

For a desired laser average optical power (PAVG) and optical extinction ratio (r_e), the required bias and modulation currents can be calculated using the equations in Table 3. Proper setting of these currents requires knowledge of the laser to monitor transfer (p_{MON}) and slope efficiency (η).

Programming the Monitor-Diode Current Set Point

The MAX3646 operates in APC mode at all times. The bias current is automatically set so average laser power is determined by the APCSET resistor:

$$PAVG = IMD / PMON$$

The APCSET pin controls the set point for the monitor diode current. An internal current regulator establishes the APCSET current in the same manner as the MODSET pin. See the I_{MD} vs. RAPCSET graph in the *Typical Operating Characteristics* and select the value of RAPCSET that corresponds to the required current at +25°C:

The laser driver automatically adjusts the bias to maintain the constant average power. For DC-coupled laser diodes:

$$I_{AVG} = I_{BIAS} + I_{MOD} / 2$$

Determine the modulation current form the laser slope efficiency:

$$MOD = 2 \times PAVG / \eta \times (r_e - 1)/(r_{e+} + 1)$$

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The modulation current of the MAX3646 consists of a static modulation current (I_{MODS}), a current proportional to I_{BIAS}, and a current proportional to temperature. The portion of I_{MOD} set by MODSET is established by an internal current regulator, which maintains the reference voltage of V_{REF} across the external programming resistor. See the I_{MOD} vs. R_{MODSET} graph in the *Typical Operating Characteristics* and select the value

of RMODSET that corresponds to the required current at +25°C:

> IMOD = IMODS + K X IBIAS + IMODT IMODS = 268 x VREF / RMODSET $I_{MODT} = TC \times (T - T_{TH})$ IT > Ттн

 $I_{MODT} = 0$ IT ≤ TTH

An external resistor at the MODBCOMP pin sets current proportional to IBIAS. Open circuiting the MODBCOMP pin can turn off the interaction between IBIAS and IMOD:

$$K = 1700 / (1000 + R_{MODBCOMP}) \pm 10\%$$

If I_{MOD} must be increased from I_{MOD1} to I_{MOD2} to maintain the extinction ratio at elevated temperatures, the required compensation factor is:

 $K = (I_{MOD2} - I_{MOD1}) / (I_{BIAS2} - I_{BIAS1})$

A threshold for additional temperature compensation can be set with a programming resistor at the TH_TEMP pin:

 $T_{TH} = -70^{\circ}C + 1.45M\Omega / (9.2k\Omega + R_{TH} T_{EMP})^{\circ}C \pm 10\%$

The temperature coefficient of thermal compensation above TTH is set by RMODTCOMP. Leaving the MODTCOMP pin open disables additional thermal compensation:

 $TC = 1 / (0.5 + R_{MODTCOMP}(k\Omega)) \text{ mA/°C} \pm 10\%$

Current Compliance ($I_{MOD} \leq 60mA$), **DC-Coupled**

The minimum voltage at the OUT+ and OUT- pins is 0.7V.

For:

 V_{DIODE} = Diode bias point voltage (1.2V typ)

 R_{L} = Diode bias point resistance (5 Ω typ)

 R_D = Series matching resistor (20 Ω typ)

For compliance:

 $VOUT + = VCC - VDIODE - IMOD \times (RD + RL) -$ IBIAS X RL $\geq 0.7V$

Current Compliance (IMOD > 60mA), AC-Coupled

For applications requiring modulation current greater than 60mA, headroom is insufficient from proper operation of the laser driver if the laser is DC-coupled. To avoid this problem, the MAX3646's modulation output can be AC-coupled to the cathode of a laser diode. An external pullup inductor is necessary to DC-bias the modulation output at V_{CC}. Such a configuration isolates laser forward voltage from the output circuitry and allows the output at OUT+ to swing above and below the supply voltage (VCC). When AC-coupled, the MAX3646 modulation current can be programmed up to 85mA. Refer to Application Note 274: HFAN-02.0: Interfacing

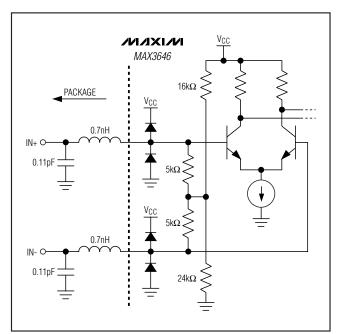


Figure 6. Simplified Input Structure

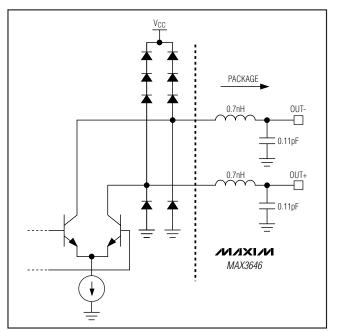


Figure 7. Simplified Output Structure



Maxim Laser Drivers with Laser Diodes for more information on AC-coupling laser drivers to laser diodes.

For compliance:

 $V_{OUT+} = V_{CC} - I_{MOD} / 2 \times (R_D + R_L) \ge 0.75 V$

Determine CAPC

The APC loop filter capacitor (CAPC) must be selected to balance the requirements for fast turn-on and minimal interaction with low frequencies in the data pattern. The low-frequency cutoff is:

$C_{APC}(\mu F) \cong 68 / (f_{3dB}(kHz) \times (\eta \times \rho_{MON})^{1.1}$

High-frequency noise can be filtered with an additional cap, $C_{\text{MD}},$ from the MD pin to ground:

 $C_{MD} \cong C_{APC} / 4$

The MAX3646 is designed so turn-on time is faster than 1ms for most laser gain values ($\eta \times \rho_{MON}$). Choosing a smaller value of CAPC reduces turn-on time. Careful balance between turn-on time and low-frequency cutoff may be needed at low data rates for some values of laser gain.

Interface Models

Figures 6 and 7 show simplified input and output circuits for the MAX3646 laser driver. If dice are used, replace package parasitic elements with bondwire parasitic elements.

Layout Considerations

To minimize loss and crosstalk, keep the connections between the MAX3646 output and the laser diode as short as possible. Use good high-frequency layout techniques and multilayer boards with uninterrupted ground plane to minimize EMI and crosstalk. Circuit boards should be made using low-loss dielectrics. Use controlled-impedance lines for data inputs, as well as the module output.

Laser Safety and IEC 825

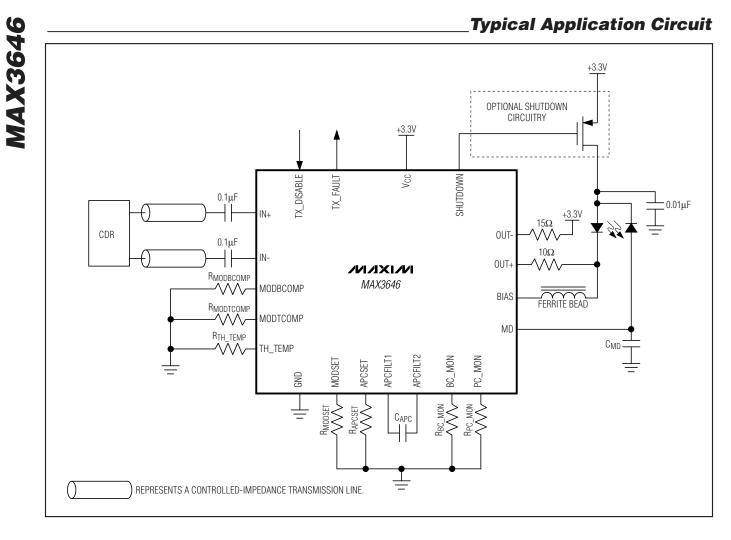
Using the MAX3646 laser driver alone does not ensure that a transmitter design is IEC 825 compliant. The entire transmitter circuit and component selections must be considered. Each customer must determine the level of fault tolerance required by their application, recognizing that Maxim products are not designed or authorized for use as components in systems intended for surgical implant into the body, for applications intended to support or sustain life, or for any other application where the failure of a Maxim product could create a situation where personal injury or death may occur.

Exposed-Pad (EP) Package

The exposed pad on the 24-pin QFN provides a very low thermal resistance path for heat removal from the IC. The pad is also electrical ground on the MAX3646 and should be soldered to the circuit board ground for proper thermal and electrical performance. Refer to Application Note 862: *HFAN-08.1: Thermal Consideration of QFN and Other Exposed-Paddle Packages* at <u>www.maxim-ic.com</u> for additional information.

Chip Information

PROCESSS: SiGe/BIPOLAR



Package Information

For the latest package outline information and land patterns (footprints), go to **www.maxim-ic.com/packages**. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

| PACKAGE TYPE | PACKAGE CODE | OUTLINE NO. | LAND PATTERN NO. |
|--------------|--------------|----------------|------------------|
| 24 TQFN-EP | T2444-3 | <u>21-0139</u> | <u>90-0021</u> |

Revision History

MAX3646

| REVISION NUMBER | REVISION DATE | DESCRIPTION | |
|--------------------|------------------|--|------------|
| 0 | 1/04 | Initial release | |
| 1 | 7/04 | Added the lead(Pb)-free package option to the Ordering Information | 1 |
| 2 | 6/11 | Updated Ordering Information; changed Absolute Maximum Ratings to reflect lead and soldering specs; changed continuous power dissipation specs; updated bias- current monitor ratio in <i>Electrical Characteristics</i> table; changed the APCSET and MODSET function description in <i>Pin Description</i> table; replaced Figure 2 | 1, 2, 6, 7 |

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

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