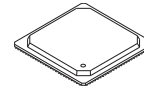


## MPC5121e



TEPBGA  
27 mm x 27 mm

## MPC5121e (Early) Data Sheet

The MPC5121e integrates a high performance MPC603e series e300 core with a rich set of peripheral functions focused on communications and systems integration.

Major features of the MPC5121e are:

- e300 Power Architecture processor core (enhanced version of the MPC603e core), operates up to 400 MHz
- Power modes include doze, nap, sleep, deep sleep, and hibernate
- AXE – fully programmable, 200 MHz 32-bit RISC core for real-time acceleration tasks, such as audio.
- MBX Lite – 2D/3D graphics engine with PowerVR vector processing
- DIU – Display interface unit
- DDR1, DDR2, and low-power mobile DDR (LPDDR) SDRAM memory controller
- USB 2.0 OTG controller with integrated physical layer (PHY)
- DMA subsystem
- EMB – Flexible multi-function external memory bus interface
- NFC – NAND flash controller
- 10/100Base Ethernet
- PCI interface, version 2.3
- PATA – Parallel ATA integrated development environment (IDE) controller
- SATA – Serial ATA controller with integrated physical layer (PHY)
- SDHC – MMC/SD/SDIO card host controller
- PSC – Programmable serial controller
- S/PDIF – Serial audio interface
- CAN – Controller area network

Figure 1 shows a simplified MPC5121e block diagram.

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**Preliminary—Subject to Change Without Notice**

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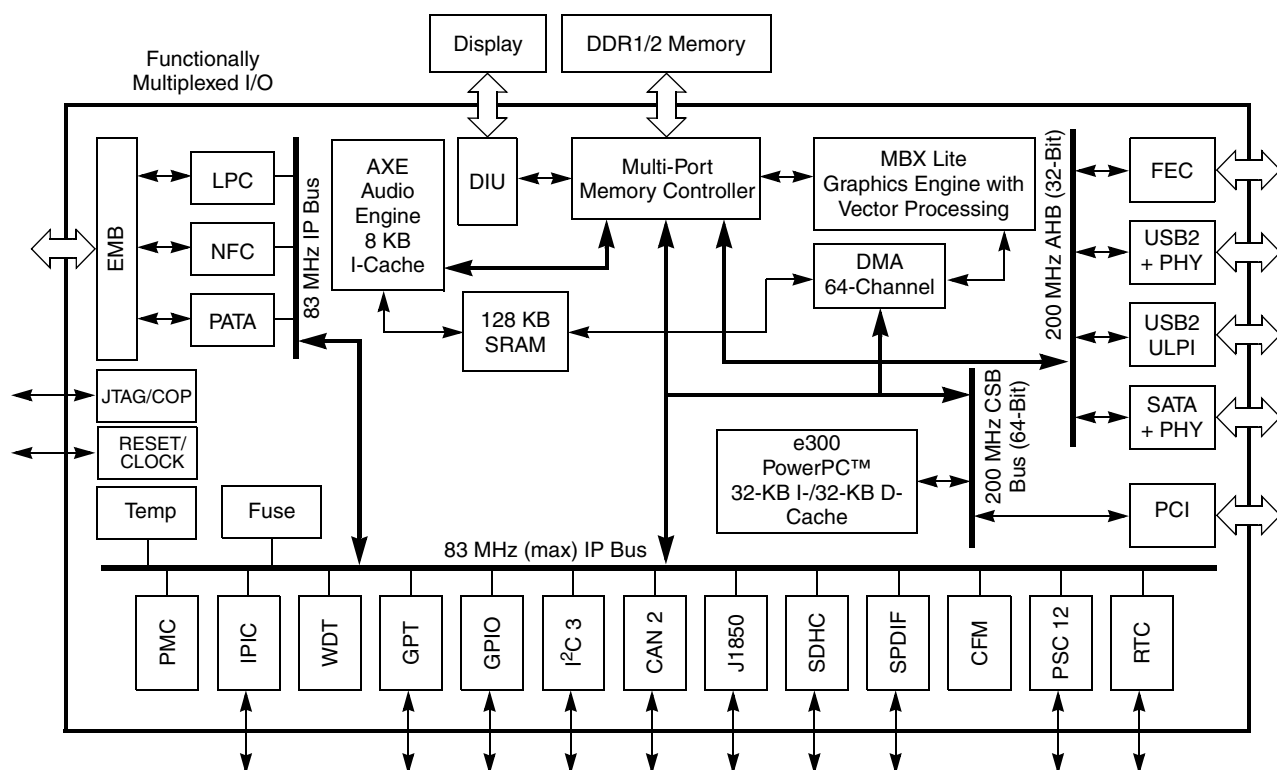


Figure 1. Simplified MPC5121e Block Diagram

# 1 Pin Assignments

This section details pin assignments.

## 1.1 Pinout Listings

Table 1 provides the pin-out listing for the MPC5121e.

Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 1 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
<b>DDR Memory Interface (67 Total)</b>				
MDQ0	AF5	DDR	VDD_MEM_IO	—
MDQ1	AB6	DDR	VDD_MEM_IO	—
MDQ2	AE4	DDR	VDD_MEM_IO	—
MDQ3	AF6	DDR	VDD_MEM_IO	—
MDQ4	AF7	DDR	VDD_MEM_IO	—
MDQ5	AB8	DDR	VDD_MEM_IO	—
MDQ6	AD6	DDR	VDD_MEM_IO	—
MDQ7	AE6	DDR	VDD_MEM_IO	—
MDQ8	AC7	DDR	VDD_MEM_IO	—

Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 2 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
MDQ9	AF8	DDR	VDD_MEM_IO	—
MDQ10	AB9	DDR	VDD_MEM_IO	—
MDQ11	AD7	DDR	VDD_MEM_IO	—
MDQ12	AE9	DDR	VDD_MEM_IO	—
MDQ13	AF10	DDR	VDD_MEM_IO	—
MDQ14	AC9	DDR	VDD_MEM_IO	—
MDQ15	AF11	DDR	VDD_MEM_IO	—
MDQ16	AD10	DDR	VDD_MEM_IO	—
MDQ17	AF12	DDR	VDD_MEM_IO	—
MDQ18	AD11	DDR	VDD_MEM_IO	—
MDQ19	AB12	DDR	VDD_MEM_IO	—
MDQ20	AD12	DDR	VDD_MEM_IO	—
MDQ21	AB13	DDR	VDD_MEM_IO	—
MDQ22	AF14	DDR	VDD_MEM_IO	—
MDQ23	AD13	DDR	VDD_MEM_IO	—
MDQ24	AE13	DDR	VDD_MEM_IO	—
MDQ25	AC13	DDR	VDD_MEM_IO	—
MDQ26	AF15	DDR	VDD_MEM_IO	—
MDQ27	AB14	DDR	VDD_MEM_IO	—
MDQ28	AE16	DDR	VDD_MEM_IO	—
MDQ29	AD15	DDR	VDD_MEM_IO	—
MDQ30	AC15	DDR	VDD_MEM_IO	—
MDQ31	AB15	DDR	VDD_MEM_IO	—
MDM0	AC6	DDR	VDD_MEM_IO	—
MDM1	AE8	DDR	VDD_MEM_IO	—
MDM2	AF13	DDR	VDD_MEM_IO	—
MDM3	AF16	DDR	VDD_MEM_IO	—
MDQS0	AD5	DDR	VDD_MEM_IO	—
MDQS1	AD8	DDR	VDD_MEM_IO	—
MDQS2	AC11	DDR	VDD_MEM_IO	—
MDQS3	AD14	DDR	VDD_MEM_IO	—
MBA0	AD16	DDR	VDD_MEM_IO	—
MBA1	AC16	DDR	VDD_MEM_IO	—
MBA2	AF19	DDR	VDD_MEM_IO	—
MA0	AD17	DDR	VDD_MEM_IO	—

Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 3 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
MA1	AB16	DDR	VDD_MEM_IO	—
MA2	AE18	DDR	VDD_MEM_IO	—
MA3	AF20	DDR	VDD_MEM_IO	—
MA4	AD18	DDR	VDD_MEM_IO	—
MA5	AB17	DDR	VDD_MEM_IO	—
MA6	AE19	DDR	VDD_MEM_IO	—
MA7	AC18	DDR	VDD_MEM_IO	—
MA8	AF21	DDR	VDD_MEM_IO	—
MA9	AD19	DDR	VDD_MEM_IO	—
MA10	AF22	DDR	VDD_MEM_IO	—
MA11	AC19	DDR	VDD_MEM_IO	—
MA12	AE21	DDR	VDD_MEM_IO	—
MA13	AD20	DDR	VDD_MEM_IO	—
MA14	AB19	DDR	VDD_MEM_IO	—
MA15	AE22	DDR	VDD_MEM_IO	—
$\overline{MWE}$	AD21	DDR	VDD_MEM_IO	—
$\overline{MRAS}$	AF23	DDR	VDD_MEM_IO	—
$\overline{MCAS}$	AF24	DDR	VDD_MEM_IO	—
$\overline{MCS}$	AD22	DDR	VDD_MEM_IO	—
MCKE	AD20	DDR	VDD_MEM_IO	—
MCK	AF17	DDR	VDD_MEM_IO	—
$\overline{MCK}$	AF18	DDR	VDD_MEM_IO	—
MODT	AC21	DDR	VDD_MEM_IO	—
<b>LPC Interface (8 Total)</b>				
LPC_CLK	AA4	General IO	VDD_IO	—
$\overline{LPC\_OE}$	Y5	General IO	VDD_IO	—
$\overline{LPC\_RW}$	AA1	General IO	VDD_IO	—
$\overline{LPC\_CS0}$	W5	General IO	VDD_IO	—
$\overline{LPC\_CS1}$	Y3	General IO	VDD_IO	—
$\overline{LPC\_CS2}$	Y1	General IO	VDD_IO	—
$\overline{LPC\_ACK}$	AA2	General IO	VDD_IO	—
LPC_AX03	W4	General IO	VDD_IO	—
<b>EMB Interface (35 Total)</b>				
EMB_AX02	W3	General IO	VDD_IO	—

Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 4 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
EMB_AX01	V5	General IO	VDD_IO	—
EMB_AX00	W2	General IO	VDD_IO	—
EMB_AD31	W1	General IO	VDD_IO	—
EMB_AD30	V4	General IO	VDD_IO	—
EMB_AD29	U5	General IO	VDD_IO	—
EMB_AD28	V3	General IO	VDD_IO	—
EMB_AD27	V2	General IO	VDD_IO	—
EMB_AD26	V1	General IO	VDD_IO	—
EMB_AD25	U1	General IO	VDD_IO	—
EMB_AD24	U3	General IO	VDD_IO	—
EMB_AD23	T5	General IO	VDD_IO	—
EMB_AD22	T1	General IO	VDD_IO	—
EMB_AD21	T4	General IO	VDD_IO	—
EMB_AD20	T3	General IO	VDD_IO	—
EMB_AD19	R5	General IO	VDD_IO	—
EMB_AD18	T2	General IO	VDD_IO	—
EMB_AD17	R1	General IO	VDD_IO	—
EMB_AD16	R3	General IO	VDD_IO	—
EMB_AD15	P1	General IO	VDD_IO	—
EMB_AD14	P2	General IO	VDD_IO	—
EMB_AD13	P4	General IO	VDD_IO	—
EMB_AD12	P5	General IO	VDD_IO	—
EMB_AD11	P3	General IO	VDD_IO	—
EMB_AD10	N1	General IO	VDD_IO	—
EMB_AD09	N2	General IO	VDD_IO	—
EMB_AD08	N3	General IO	VDD_IO	—
EMB_AD07	N4	General IO	VDD_IO	—
EMB_AD06	M1	General IO	VDD_IO	—
EMB_AD05	M3	General IO	VDD_IO	—
EMB_AD04	M5	General IO	VDD_IO	—
EMB_AD03	L1	General IO	VDD_IO	—
EMB_AD02	L2	General IO	VDD_IO	—

Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 5 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
EMB_AD01	L3	General IO	VDD_IO	—
EMB_AD00	L4	General IO	VDD_IO	—
<b>PATA Interface (9 Total)</b>				
$\overline{\text{PATA\_CE1}}$	K1	General IO	VDD_IO	ATA name: CS0
$\overline{\text{PATA\_CE2}}$	L5	General IO	VDD_IO	ATA name: CS1
PATA_ISOLATE	K3	General IO	VDD_IO	—
$\overline{\text{PATA\_IOR}}$	J1	General IO	VDD_IO	ATA name: DIOR
$\overline{\text{PATA\_IOW}}$	K5	General IO	VDD_IO	ATA name: DIOW
PATA_IOCHRDY	J2	General IO	VDD_IO	ATA name: IORDY
PATA_INTRQ	J3	General IO	VDD_IO	—
PATA_DRQ	J4	General IO	VDD_IO	ATA name: DMARQ
$\overline{\text{PATA\_DACK}}$	H2	General IO	VDD_IO	ATA name: DMACK
<b>NFC Interface (7 Total)</b>				
$\overline{\text{NFC\_WP}}$	G4	General IO	VDD_IO	—
NFC_RB	H1	General IO	VDD_IO	—
$\overline{\text{NFC\_WE}}$	G3	General IO	VDD_IO	—
$\overline{\text{NFC\_RE}}$	G2	General IO	VDD_IO	—
NFC_ALE	H4	General IO	VDD_IO	—
NFC_CLE	H5	General IO	VDD_IO	—
$\overline{\text{NFC\_CE0}}$	H3	General IO	VDD_IO	—
<b>I2C Interface (6 Total)</b>				
I2C0_SCL	AC23	General IO	VDD_IO	—
I2C0_SDA	AD26	General IO	VDD_IO	—
I2C1_SCL	AB22	General IO	VDD_IO	—
I2C1_SDA	AB23	General IO	VDD_IO	—
I2C2_SCL	AC25	General IO	VDD_IO	—
I2C2_SDA	AA22	General IO	VDD_IO	—
<b>IRQ Interface (2 Total)</b>				
IRQ0	AC26	General IO	VDD_IO	—
IRQ1	AB25	General IO	VDD_IO	—
<b>CAN Interface (4 Total)</b>				
CAN1_RX	C19	Analog Input	VBAT_RTC	—
CAN1_TX	A18	General IO	VDD_IO	—
CAN2_RX	B19	Analog Input	VBAT_RTC	—

Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 6 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
CAN2_TX	E16	General IO	VDD_IO	—
<b>J1850 Interface (2 Total)</b>				
J1850_TX	Y22	General IO	VDD_IO	—
J1850_RX	AA24	General IO	VDD_IO	—
<b>SPDIF Interface (3 Total)</b>				
SPDIF_TXCLK	AB21	General IO	VDD_IO	—
SPDIF_TX	AD24	General IO	VDD_IO	—
SPDIF_RX	AC24	General IO	VDD_IO	—
<b>PCI (54 Total)</b>				
PCI_INTA	U23	PCI	VDD_IO	—
PCI_RST_OUT	F22	PCI	VDD_IO	—
PCI_AD0	U24	PCI	VDD_IO	—
PCI_AD1	V26	PCI	VDD_IO	—
PCI_AD2	U25	PCI	VDD_IO	—
PCI_AD3	R22	PCI	VDD_IO	—
PCI_AD4	U26	PCI	VDD_IO	—
PCI_AD5	T24	PCI	VDD_IO	—
PCI_AD6	R23	PCI	VDD_IO	—
PCI_AD7	T26	PCI	VDD_IO	—
PCI_AD8	R26	PCI	VDD_IO	—
PCI_AD9	P23	PCI	VDD_IO	—
PCI_AD10	R24	PCI	VDD_IO	—
PCI_AD11	R25	PCI	VDD_IO	—
PCI_AD12	P26	PCI	VDD_IO	—
PCI_AD13	P24	PCI	VDD_IO	—
PCI_AD14	P25	PCI	VDD_IO	—
PCI_AD15	N26	PCI	VDD_IO	—
PCI_AD16	L22	PCI	VDD_IO	—
PCI_AD17	K25	PCI	VDD_IO	—
PCI_AD18	J26	PCI	VDD_IO	—
PCI_AD19	K24	PCI	VDD_IO	—
PCI_AD20	J25	PCI	VDD_IO	—
PCI_AD21	H26	PCI	VDD_IO	—
PCI_AD22	K23	PCI	VDD_IO	—



Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 7 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
PCI_AD23	J24	PCI	VDD_IO	—
PCI_AD24	H24	PCI	VDD_IO	—
PCI_AD25	J23	PCI	VDD_IO	—
PCI_AD26	G25	PCI	VDD_IO	—
PCI_AD27	J22	PCI	VDD_IO	—
PCI_AD28	F26	PCI	VDD_IO	—
PCI_AD29	G24	PCI	VDD_IO	—
PCI_AD30	F24	PCI	VDD_IO	—
PCI_AD31	H22	PCI	VDD_IO	—
PCI_C/BE0	P22	PCI	VDD_IO	—
PCI_C/BE1	N24	PCI	VDD_IO	—
PCI_C/BE2	L24	PCI	VDD_IO	—
PCI_C/BE3	G26	PCI	VDD_IO	—
PCI_PAR	N22	PCI	VDD_IO	—
PCI_FRAME <sup>1</sup>	M23	PCI	VDD_IO	4
PCI_TRDY	M22	PCI	VDD_IO	4
PCI_IRDY	K26	PCI	VDD_IO	4
PCI_STOP	M24	PCI	VDD_IO	4
PCI_DEVSEL	L26	PCI	VDD_IO	4
PCI_IDSEL	K22	PCI	VDD_IO	—
PCI_SERR	M26	PCI	VDD_IO	4
PCI_PERR	M25	PCI	VDD_IO	4
PCI_REQ0	G23	PCI	VDD_IO	4
PCI_REQ1	E26	PCI	VDD_IO	4
PCI_REQ2	D26	PCI	VDD_IO	4
PCI_GNT0	E25	PCI	VDD_IO	—
PCI_GNT1	G22	PCI	VDD_IO	—
PCI_GNT2	E24	PCI	VDD_IO	—
PCI_CLK	C26	PCI	VDD_IO	—
<b>PSC Interface (61 Total)</b>				
PSC_MCLK_IN	C17	General IO	VDD_IO	—
PSC0_0	D16	General IO	VDD_IO	—
PSC0_1	A17	General IO	VDD_IO	—
PSC0_2	E15	General IO	VDD_IO	—

Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 8 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
PSC0_3	C16	General IO	VDD_IO	—
PSC0_4	B16	General IO	VDD_IO	—
PSC1_0	C15	General IO	VDD_IO	—
PSC1_1	A16	General IO	VDD_IO	—
PSC1_2	E14	General IO	VDD_IO	—
PSC1_3	A15	General IO	VDD_IO	—
PSC1_4	D14	General IO	VDD_IO	—
PSC2_0	C14	General IO	VDD_IO	—
PSC2_1	B14	General IO	VDD_IO	—
PSC2_2	E13	General IO	VDD_IO	—
PSC2_3	A14	General IO	VDD_IO	—
PSC2_4	D13	General IO	VDD_IO	—
PSC3_0	AF3	General IO	VDD_IO	—
PSC3_1	AB5	General IO	VDD_IO	—
PSC3_2	AC4	General IO	VDD_IO	—
PSC3_3	AD4	General IO	VDD_IO	—
PSC3_4	AF4	General IO	VDD_IO	—
PSC4_0	AB1	General IO	VDD_IO	—
PSC4_1	AA3	General IO	VDD_IO	—
PSC4_2	AB3	General IO	VDD_IO	—
PSC4_3	AA5	General IO	VDD_IO	—
PSC4_4	AC2	General IO	VDD_IO	—
PSC5_0	AC1	General IO	VDD_IO	—
PSC5_1	AC3	General IO	VDD_IO	—
PSC5_2	AD1	General IO	VDD_IO	—
PSC5_3	AD2	General IO	VDD_IO	—
PSC5_4	AE3	General IO	VDD_IO	—
PSC6_0	A11	General IO	VDD_IO	—
PSC6_1	C10	General IO	VDD_IO	—
PSC6_2	A10	General IO	VDD_IO	—
PSC6_3	B9	General IO	VDD_IO	—
PSC6_4	A9	General IO	VDD_IO	—
PSC7_0	B8	General IO	VDD_IO	—
PSC7_1	E10	General IO	VDD_IO	—
PSC7_2	C8	General IO	VDD_IO	—

Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 9 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
PSC7_3	A8	General IO	VDD_IO	—
PSC7_4	A7	General IO	VDD_IO	—
PSC8_0	E9	General IO	VDD_IO	—
PSC8_1	D8	General IO	VDD_IO	—
PSC8_2	C7	General IO	VDD_IO	—
PSC8_3	B6	General IO	VDD_IO	—
PSC8_4	E8	General IO	VDD_IO	—
PSC9_0	C6	General IO	VDD_IO	—
PSC9_1	D7	General IO	VDD_IO	—
PSC9_2	E7	General IO	VDD_IO	—
PSC9_3	D6	General IO	VDD_IO	—
PSC9_4	E6	General IO	VDD_IO	—
PSC10_0	C13	General IO	VDD_IO	—
PSC10_1	B13	General IO	VDD_IO	—
PSC10_2	A13	General IO	VDD_IO	—
PSC10_3	C12	General IO	VDD_IO	—
PSC10_4	E12	General IO	VDD_IO	—
PSC11_0	A12	General IO	VDD_IO	—
PSC11_1	B11	General IO	VDD_IO	—
PSC11_2	C11	General IO	VDD_IO	—
PSC11_3	E11	General IO	VDD_IO	—
PSC11_4	D11	General IO	VDD_IO	—
<b>JTAG (5 Total)</b>				
TCK	AB26	General IO	VDD_IO	6
TDI	Y23	General IO	VDD_IO	3
TDO	W22	General IO	VDD_IO	—
TMS	Y25	General IO	VDD_IO	3
$\overline{\text{TRST}}$	AA26	General IO	VDD_IO	3
<b>Test / Debug (2 Total)</b>				
TEST	W25	General IO	VDD_IO	2, 5
$\overline{\text{CKSTP\_OUT}}$	Y26	General IO	VDD_IO	—
<b>System Control (3 Total)</b>				
$\overline{\text{HRESET}}$	W24	General IO	VDD_IO	1, 6
$\overline{\text{PORESET}}$	W23	General IO	VDD_IO	2, 6

Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 10 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
$\overline{\text{SRESET}}$	V22	General IO	VDD_IO	1, 6
<b>System Clock (2 Total)</b>				
SYS_XTALI	V24	Analog Input	SYS_PLL_AVDD	Oscillator Input
SYS_XTALO	W26	Analog Output	SYS_PLL_AVDD	Oscillator Output
<b>RTC (3 Total)</b>				
XTALI_RTC	C20	Analog Input	VBAT_RTC	Oscillator Input
XTALO_RTC	A20	Analog Output	VBAT_RTC	Oscillator Output
HIB_MODE	D18	Analog Output	VBAT_RTC	—
<b>GP Input Only (4 Total)</b>				
GPIO28	A19	Analog Input	VBAT_RTC	—
GPIO29	E17	Analog Input	VBAT_RTC	—
GPIO30	C18	Analog Input	VBAT_RTC	—
GPIO31	B18	Analog Input	VBAT_RTC	—
<b>DDR Reference Voltage</b>				
MVREF	AB11	Analog Input	Voltage Reference for SSTL input pads	
<b>USB – PHY without Power and Ground Supplies (7 Total)</b>				
USB_XTALI	C24	Analog Input	USB_PLL_PWR3	Oscillator Input
USB_XTALO	B24	Analog Output	USB_PLL_PWR3	Oscillator Output
USB_DP	A23	Analog IO	USB_VDDA	—
USB_DM	A22	Analog IO	USB_VDDA	—
USB_TPA	A24	Analog Output	—	—
USB_VBUS	D21	Analog IO	—	—
USB_UID	E19	Analog Input	—	—
<b>USB digital IOs (2 Total)</b>				
USB2_VBUS_PWR_FAULT	B21	General IO	VDD_IO	—
USB2_DRVVBUS	A21	General IO	VDD_IO	—
<b>SATA PHY without Power and Ground Supplies (7 Total)</b>				
SATA_XTALI	C3	Analog Input	SATA_VDDA_3P3	Oscillator Input
SATA_XTALO	C2	Analog Output	SATA_VDDA_3P3	Oscillator Output
SATA_ANAVIZ	E5	Analog Output	—	SATA PHY debug output
SATA_TXN	E1	Analog Output	SATA_VDDA_1P2	—
SATA_TXP	F1	Analog Output	SATA_VDDA_1P2	—

Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 11 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
SATA_RXP	A5	Analog Input	SATA_VDDA_1P2	—
SATA_RXN	A4	Analog Input	SATA_VDDA_1P2	—
<b>Power and Ground Supplies (without SATA PHY and USB PHY)</b>				
VDD_CORE	K10, K11, K12, K13, K14, K15, K16, K17, L10, L17, M10, M17, N10, N17, P10, P17, R10, R17, T10, T17, U10, U11, U12, U13, U14, U15, U16, U17	Power	—	—
VDD_IO	B10, B15, B25, D10, D15, F11, F13, F14, F19, F23, F25, H21, J5, K2, K4, L23, L25, N6, N21, P6, P21, R2, R4, T23, T25, W6, W21, Y2, Y4, AA23, AA25, AE1, AE2, AE24, AE25, AF2, AF25	Power	—	—
VDD_MEM_IO	AA8, AA13, AA14, AB18, AC5, AC10, AC14, AC20, AD9, AE5, AE10, AE15, AE20	Power	—	—
VSS	A2, A3, A25, B1, B2, B3, B5, B7, B12, B20, B22, B26, C1, C4, C23, C25, D2, D12, D17, D24, D25, E18, F2, F3, F4, F5, F6, F8, F10, F16, F17, F21, G5, H6, H23, H25, K6, K21, L6, L11, L12, L13, L14, L15, L16, L21, M2, M4, M11, M12, M13, M14, M15, M16, N5, N11, N12, N13, N14, N15, N16,	Ground	—	—

Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 12 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
VSS continued	N23, N25, P11, P12, P13, P14, P15, P16, R11, R12, R13, R14, R15, R16, T6, T11, T12, T13, T14, T15, T16, T21, U2, U4, U6, U21, V23, V25, Y24, AA6, AA10, AA11, AA16, AA17, AA21, AB2, AB4, AB10, AB24, AC8, AC12, AC17, AC22, AD3, AD25, AE7, AE12, AE17, AE23, AE26	Ground	—	—
SYS_PLL_AVDD	T22	Analog Power	—	—
SYS_PLL_AVSS	U22	Analog Ground	—	—
CORE_PLL_AVDD	AA19	Analog Power	—	—
CORE_PLL_AVSS	AD23	Analog Ground	—	—
VBAT_RTC	D19	Power	—	—
AVDD_FUSEWR	C9	Power	—	—
AVDD_FUSERD	D9	Power	—	—
MVTT0	AB7	Analog Input	SSTL(DDR2) Termination (ODT) Voltage	
MVTT1	AF9	Analog Input	SSTL(DDR2) Termination (ODT) Voltage	
MVTT2	AE11	Analog Input	SSTL(DDR2) Termination (ODT) Voltage	
MVTT3	AE14	Analog Input	SSTL(DDR2) Termination (ODT) Voltage	
<b>Power and Ground Supplies (USB PHY)</b>				
USB_PLL_GND	E23	Analog Ground	—	—
USB_PLL_PWR3	D23	Analog Power	—	—
USB_RREF	E22	Analog Power	—	—
USB_VSSA_BIAS	B23	Analog Ground	—	—
USB_VDDA_BIAS	D22	Analog Power	—	—
USB_VSSA	C22, E20, E21	Analog Ground	—	—
USB_VDDA	D20	Analog Power	—	—
<b>Power and Ground Supplies (SATA PHY)</b>				
SATA_RESREF	E4	Analog Power	—	—
SATA_VDDA_3P3	D4	Analog Power	—	—
SATA_VDDA_1P2	C5, D1, E2	Analog Power	—	—
SATA_VDDA_VREG	D5	Analog Power	—	—
SATA_PLL_VDDA1P2	E3	Analog Power	—	—

Table 1. MPC5121e TE-PBGA Pinout Listing (Sheet 13 of 13)

Signal	Package Pin Number	Pad Type	Power Supply	Notes
SATA_PLL_VSSA	D3	Analog Ground	—	—
SATA_RX_VSSA	B4, A6	Analog Ground	—	—
SATA_TX_VSSA	G1	Analog Ground	—	—

<sup>1</sup> This pin is an input or open-drain output. This pin can not be configured. An external pull-up resistor is required.

2) This pin is an input only. This pin can not be configured.

3) These JTAG pins have internal pull-up P-FETs. This pin can not be configured.

4) This pins should have an external pull-up resistor. Follow PCI specification and see System Design Information.

5) This test pin must be tied to VSS.

6) This pin contains an enabled internal schmitt-trigger.

### NOTE

This table indicates only the pins with permanently enabled internal pull-up, pull-down, or schmitt-trigger. Most of the digital I/O pins can be configured to enable internal pull-up, pull-down, or schmitt-trigger. See MPC5121e Reference Manual, IO Control chapter.

## 2 Electrical and Thermal Characteristics

### 2.1 DC Electrical Characteristics

#### 2.1.1 Absolute Maximum Ratings

The tables in this section describe the MPC5121e DC Electrical characteristics. [Table 2](#) gives the absolute maximum ratings.

Table 2. Absolute Maximum Ratings<sup>1</sup>

Characteristic	Sym	Min	Max	Unit	SpecID
Supply voltage – e300 core and peripheral logic	VDD_CORE	-0.3	1.47	V	D1.1
Supply voltage – I/O buffers	VDD_IO, VDD_MEM_IO	-0.3	3.6	V	D1.2
Supply voltage – System APLL, System Oscillator	SYS_PLL_AVDD	-0.3	3.6	V	D1.3
Supply voltage – e300 APLL	CORE_PLL_AVDD	-0.3	3.6	V	D1.4
Supply voltage – RTC (Hibernation)	VBAT_RTC	-0.3	3.6	V	D1.5
Supply voltage – FUSE Programming	AVDD_FUSEWR	-0.3	3.6	V	D1.6
Supply voltage – FUSE Reading and Tempsensor	AVDD_FUSERD	-0.3	3.6	V	D1.7
Supply voltage – SATA PHY analog	SATA_VDDA_3P3	-0.3	3.6	V	D1.8
Supply voltage – SATA PHY voltage regulator	SATA_VDDA_VREG	-0.3	2.6	V	D1.9
Supply voltage – SATA PHY Tx/Rx	SATA_VDDA_1P2	-0.3	1.47	V	D1.10
Supply voltage – SATA PHY PLL	SATA_PLL_VDDA1P2	-0.3	1.47	V	D1.11

Table 2. Absolute Maximum Ratings<sup>1</sup> (continued)

Characteristic	Sym	Min	Max	Unit	SpecID
Supply voltage – USB PHY PLL and OSC	USB_PLL_PWR3	-0.3	3.6	V	D1.12
Supply voltage – USB PHY transceiver	USB_VDDA	-0.3	3.6	V	D1.13
Supply voltage – USB PHY bandgap bias	USB_VDDA_BIAS	-0.3	3.6	V	D1.14
Input voltage – USB PHY cable	USB_VBUS	-0.3	3.6	V	D1.15
Input voltage (VDD_IO)	V <sub>in</sub>	-0.3	VDD_IO + 0.3	V	D1.16
Input voltage (VDD_MEM_IO)	V <sub>in</sub>	-0.3	VDD_MEM_IO + 0.3	V	D1.17
Input voltage (VBAT_RTC)	V <sub>in</sub>	-0.3	VBAT_RTC + 0.3	V	D1.18
Input voltage overshoot	V <sub>inos</sub>	—	1	V	D1.19
Input voltage undershoot	V <sub>inus</sub>	—	1	V	D1.20
Storage temperature range	T <sub>stg</sub>	-55	150	°C	D1.21

<sup>1</sup> Absolute maximum ratings are stress ratings only, and functional operation at the maximums is not guaranteed. Stresses beyond those listed may affect device reliability or cause permanent damage.

## 2.1.2 Recommended Operating Conditions

Table 3 gives the recommended operating conditions.

Table 3. Recommended Operating Conditions

Characteristic	Sym	Min <sup>1</sup>	Typ	Max <sup>1</sup>	Unit	SpecID
Supply voltage – e300 core and peripheral logic	VDD_CORE	1.33	1.4	1.47	V	D2.1
State Retention voltage – e300 core and peripheral logic <sup>2</sup>		1.08	—	—	V	D2.2
Supply voltage – standard I/O buffers	VDD_IO	3.0	3.3	3.6	V	D2.3
Supply voltage – memory I/O buffers (DDR)	VDD_MEM_IO <sub>DDR</sub>	2.3	2.5	2.7	V	D2.4
Supply voltage – memory I/O buffers (DDR2, LPDDR)	VDD_MEM_IO <sub>DDR2</sub> VDD_MEM_IO <sub>LPDDR</sub>	1.7	1.8	1.9	V	D2.5
Input Reference Voltage (DDR/DDR2)	MVREF	0.49*VDD_M EM_IO	0.50*VDD_M EM_IO	0.51*VDD_M MEM_IO	V	D2.6
Termination Voltage (DDR2)	MVTT	MVREF-0.04	MVREF	MVREF+ 0.04	V	D2.7
Supply voltage – System APLL, System Oscillator	SYS_PLL_AVDD	3.0	3.3	3.6	V	D2.8
Supply voltage – e300 APLL	CORE_PLL_AVDD	3.0	3.3	3.6	V	D2.9
Supply voltage – RTC (Hibernation)	VBAT_RTC	3.0	3.3	3.6	V	D2.10
Supply voltage – FUSE Programming	AVDD_FUSEWR	3.0		3.6	V	D2.11



Table 3. Recommended Operating Conditions (continued)

Characteristic	Sym	Min <sup>1</sup>	Typ	Max <sup>1</sup>	Unit	SpecID
Supply voltage – FUSE Reading	AVDD_FUSERD	3.0	3.3	3.6	V	D2.12
Supply voltage – SATA PHY analog and OSC	SATA_VDDA_3P3	3.0	3.3	3.6	V	D2.13
Supply voltage – SATA PHY voltage regulator	SATA_VDDA_VREG	1.7		2.6	V	D2.14
Supply voltage – SATA PHY Tx/Rx	SATA_VDDA_1P2	1.14	1.2	1.47	V	D2.15
Supply voltage – SATA PHY PLL	SATA_PLL_VDDA1P2	1.14	1.2	1.47	V	D2.16
Supply voltage – USB PHY PLL and OSC	USB_PLL_PWR3	3.0	3.3	3.6	V	D2.17
Supply voltage – USB PHY transceiver	USB_VDDA	3.0	3.3	3.6	V	D2.18
Supply voltage – USB PHY bandgap bias	USB_VDDA_BIAS	3.0	3.3	3.6	V	D2.19
Input voltage – USB PHY cable	USB_VBUS	1.4	—	3.6	V	D2.20
Input voltage – standard I/O buffers	V <sub>in</sub>	0	—	VDD_IO	V	D2.21
Input voltage – memory I/O buffers (DDR)	V <sub>in</sub> DDR	0	—	VDD_MEM_IO <sub>DDR</sub>	V	D2.22
Input voltage – memory I/O buffers (DDR2)	V <sub>in</sub> DDR2	0	—	VDD_MEM_IO <sub>DDR2</sub>	V	D2.23
Ambient operating temperature range	TA	0	—	+70	°C	D2.24

<sup>1</sup> These are recommended and tested operating conditions. Proper device operation outside these conditions is not guaranteed.

<sup>2</sup> The State Retention voltage can be applied to VDD\_CORE after the device is placed in Deep-Sleep mode.

### 2.1.3 DC Electrical Specifications

Table 4 gives the DC Electrical characteristics for the MPC5121e at recommended operating conditions.

Table 4. DC Electrical Specifications

Characteristic	Condition	Sym	Min	Max	Unit	SpecID
Input high voltage	Input type = TTL VDD_IO	V <sub>IH</sub>	0.51*VDD_IO	—	V	D3.1
Input high voltage	Input type = TTL VDD_MEM_IO <sub>DDR</sub>	V <sub>IH</sub>	MVREF+0.15	—	V	D3.2
Input high voltage	Input type = TTL VDD_MEM_IO <sub>DDR2</sub>	V <sub>IH</sub>	MVREF+0.125	—	V	D3.3
Input high voltage	Input type = TTL VDD_MEM_IO <sub>LPDDR</sub>	V <sub>IH</sub>	0.7*VDD_IO <sub>MEM<sub>L</sub></sub> PDDR	—	V	D3.4
Input high voltage	Input type = PCI VDD_IO	V <sub>IH</sub>	0.5*VDD_IO	—	V	D3.5
Input high voltage	Input type = SCHMITT VDD_IO	V <sub>IH</sub>	0.65*VDD_IO	—	V	D3.6
Input high voltage	SYS_XTALI crystal mode <sup>1</sup> bypass mode <sup>2</sup>	CV <sub>IH</sub>	V <sub>x</sub> tal+0.4V (VDD_IO/2)+0.4V	—	V	D3.7
Input high voltage	SATA_XTALI crystal mode bypass mode	SV <sub>IH</sub>	V <sub>x</sub> tal+0.4V (VDD_IO/2)+0.4V	—	V	D3.8

Table 4. DC Electrical Specifications (continued)

Characteristic	Condition	Sym	Min	Max	Unit	SpecID
Input high voltage	USB_XTALI crystal mode bypass mode	UV <sub>IH</sub>	V <sub>x</sub> tal+0.4V (VDD_IO/2)+0.4V	—	V	D3.9
Input high voltage	RTC_XTALI crystal mode <sup>3</sup> bypass mode <sup>4</sup>	RV <sub>IH</sub>	(VBAT_RTC/5)+0.5V (VBAT_RTC/2)+0.4V	—	V	D3.10
Input low voltage	Input type = TTL VDD_IO	V <sub>IL</sub>	—	0.42*VDD_IO	V	D3.11
Input low voltage	Input type = TTL VDD_MEM_IO <sub>DDR</sub>	V <sub>IL</sub>	—	MVREF-0.15	V	D3.12
Input low voltage	Input type = TTL VDD_MEM_IO <sub>DDR2</sub>	V <sub>IL</sub>	—	MVREF-0.125	V	D3.13
Input low voltage	Input type = TTL VDD_MEM_IO <sub>LPDDR</sub>	V <sub>IL</sub>	—	0.3*VDD_IO <sub>MEM<sub>L</sub></sub> PDDR	V	D3.14
Input low voltage	Input type = PCI VDD_IO	V <sub>IL</sub>	—	0.3*VDD_IO	V	D3.15
Input low voltage	Input type = SCHMITT VDD_IO	V <sub>IL</sub>	—	0.35*VDD_IO	V	D3.16
Input low voltage	SYS_XTALI crystal mode bypass mode	CV <sub>IL</sub>	—	V <sub>x</sub> tal-0.4V (VDD_IO/2)-0.4V	V	D3.17
Input low voltage	SATA_XTALI crystal mode bypass mode	SV <sub>IL</sub>	—	V <sub>x</sub> tal-0.4V (VDD_IO/2)-0.4V	V	D3.18
Input low voltage	USB_XTALI crystal mode bypass mode	UV <sub>IL</sub>	—	V <sub>x</sub> tal-0.4V (VDD_IO/2)-0.4V	V	D3.19
Input low voltage	RTC_XTALI crystal mode bypass mode	RV <sub>IL</sub>	—	(VBAT_RTC/5)-0.5V (VBAT_RTC/2)-0.4V	V	D3.20
Input leakage current	V <sub>in</sub> = 0 or VDD_IO/VDD_MEM_IO <sub>DDR/2</sub> (depending on input type) <sup>5</sup>	I <sub>IN</sub>	-2.5	2.5	μA	D3.21
Input leakage current	SYS_XTAL_IN V <sub>in</sub> = 0 or VDD_IO	I <sub>IN</sub>	—	20	μA	D3.22
Input leakage current	RTC_XTAL_IN V <sub>in</sub> = 0 or VDD_IO	I <sub>IN</sub>	—	1.01	μA	D3.23
Input current, pullup resistor <sup>6</sup>	PULLUP VDD_IO V <sub>in</sub> = V <sub>IL</sub>	I <sub>INpu</sub>	25	270	μA	D3.24
Input current, pulldown resistor <sup>8</sup>	PULLDOWN VDD_IO V <sub>in</sub> = V <sub>IH</sub>	I <sub>INpd</sub>	25	150	μA	D3.25
Output high voltage	IOH is driver dependent <sup>7</sup> VDD_IO	V <sub>OH</sub>	0.8*VDD_IO	—	V	D3.26
Output high voltage	IOH is driver dependent <sup>7</sup> VDD_MEM_IO <sub>DDR</sub>	V <sub>OHDDR</sub>	1.94	—	V	D3.27
Output high voltage	IOH is driver dependent <sup>7</sup> VDD_MEM_IO <sub>DDR2</sub>	V <sub>OHDDR2</sub>	VDD_MEM_IO-0.28	—	V	D3.28
Output high voltage	IOH is driver dependent <sup>7</sup> VDD_MEM_IO <sub>LPDDR</sub>	V <sub>OHLPDDR</sub>	0.9*VDD_MEM_IO	—	V	D3.28
Output low voltage	IOL is driver dependent <sup>7</sup> VDD_IO	V <sub>OL</sub>	—	0.2*VDD_IO	V	D3.30
Output low voltage	IOL is driver dependent <sup>7</sup> VDD_MEM_IO <sub>DDR</sub>	V <sub>OLDDR</sub>	—	0.36	V	D3.31

Table 4. DC Electrical Specifications (continued)

Characteristic	Condition	Sym	Min	Max	Unit	SpecID
Output low voltage	IOL is driver dependent <sup>7</sup> VDD_MEM_IO <sub>DDR2</sub>	V <sub>OLDDR2</sub>	—	0.28	V	D3.32
Output low voltage	IOL is driver dependent <sup>7</sup> VDD_MEM_IO <sub>LPDDR</sub>	V <sub>OLLPDDR</sub>	—	0.1*VDD_MEM_IO	V	D3.33
Differential cross point voltage (DDR MCK/MCK)	—	V <sub>OXMCK</sub>	0.5*VDD_MEM_IO – 0.125	0.5*VDD_MEM_IO + 0.125	V	D3.34
DC Injection Current Per Pin <sup>8</sup>	—	I <sub>CS</sub>	–1.0	1.0	mA	D3.35
Input Capacitance (digital pins)	—	C <sub>in</sub>	—	7	pF	D3.36
Input Capacitance (analog pins)	—	C <sub>in</sub>	—	10	pF	D3.37
On Die Termination (DDR2)	—	R <sub>ODT</sub>	120	180	Ω	D3.38

- <sup>1</sup> This parameter is meant for those who do not use quartz crystals or resonators, but CAN osc, in crystal mode. In that case, Vextal - Vxtal - 400mV criteria has to be met for oscillator's comparator to produce output clock.
- <sup>2</sup> This parameter is meant for those who do not use quartz crystals or resonators, but signal generator clock to drive, in bypass mode. In that case, drive only the EXTAL pin not connecting anything to other pin for the oscillator's comparator to produce output clock.
- <sup>3</sup> This parameter is meant for those who do not use quartz crystals or resonators, but CAN osc, in crystal mode. In that case, drive one of the XTAL\_IN or XTAL\_OUT pins not connecting anything to other pin for the oscillator's comparator to produce output clock.
- <sup>4</sup> This parameter is meant for those who do not use quartz crystals or resonators, but signal generator clock to drive, in bypass mode. In that case, drive only the xtal\_in pin not connecting anything to other pin for the oscillator's comparator to produce output clock.
- <sup>5</sup> Leakage current is measured with output drivers disabled and pull-up/pull-downs inactive.
- <sup>6</sup> Pullup current is measured at VIL and pulldown current is measured at VIH.
- <sup>7</sup> See Table 5 for the typical drive capability of a specific signal pin based on the type of output driver associated with that pin as listed in Table 1.
- <sup>8</sup> All injection current is transferred to VDD\_IO/VDD\_MEM\_IO. An external load is required to dissipate this current to maintain the power supply within the specified voltage range.  
Total injection current for all digital input-only and all digital input/output pins must not exceed 10 mA. Exceeding this limit can cause disruption of normal operation.

Table 5. I/O Pads - Drive Current, Slew Rate

Pad Type	Supply Voltage	Drive Select/Slew Rate Control	Rise time max (ns)	Fall time max (ns)	Current loh (mA)	Current lol (mA)	SpecID
General IO	VDD_IO = 3.3V	11	1.4	1.6	35	35	D3.41
		10	9.8	12			D3.42
		01	19	24			D3.43
		00	140	183			D3.44

Table 5. I/O Pads (continued)- Drive Current, Slew Rate

Pad Type	Supply Voltage	Drive Select/Slew Rate Control	Rise time max (ns)	Fall time max (ns)	Current loh (mA)	Current lol (mA)	SpecID
DDR	VDD_MEM_IO = 2.5V (DDR)	011	2	2	15.4	16.2	D3.45
		000 001	1	1	8.1	8.1	D3.46
	4.6				4.6	D3.47	
	VDD_MEM_IO = 1.8V (LPDDR)	010	1	1	12.1	13.4	D3.48
		110			5.3	5.3	D3.49
PCI	VDD_IO = 3.3V	1	1.4	1.4	11	17	D3.50
		0	2	2			D3.51

Notes:

1. General IO – Rise and Fall Times at Drive load 50pF.
2. PCI – Rise and Fall Times at Drive load 10pF.
3. DDR – Rise and Fall Times at 70 ohm transmission line with 0.167 ns td terminated at the destination with 70 ohm to MVTT (0.5\*VDD\_IO\_MEM) with 4pF, representing the DDR input capacitance.

## 2.1.4 Electrostatic Discharge

### CAUTION

This device contains circuitry that protects against damage due to high-static voltage or electrical fields. However, it is advised that normal precautions be taken to avoid application of any voltages higher than maximum-rated voltages. Operational reliability is enhanced if unused inputs are tied to an appropriate logic voltage level (GND or VDD). Table 8 gives package thermal characteristics for this device.

Table 6. ESD and Latch-Up Protection Characteristics

Sym	Rating	Min	Max	Unit	SpecID
V <sub>HBM</sub>	Human Body Model (HBM) – JEDEC JESD22-A114-B	2000	—	V	D4.1
V <sub>MM</sub>	Machine Model (MM) – JEDEC JESD22-A115	200	—	V	D4.2
V <sub>CDM</sub>	Charge Device Model (CDM) – JEDEC JESD22-C101	250	—	V	D4.3
I <sub>LAT</sub>	Latch-up Current at TA=25 °C	±90	—	mA	D4.4

## 2.1.5 Power Dissipation

Power dissipation of the MPC5121e is caused by 4 different components: the dissipation of the internal or core digital logic (supplied by VDD\_CORE), the dissipation of the analog circuitry (supplied by SYS\_PLL\_AVDD and CORE\_PLL\_AVDD), the dissipation of the IO logic (supplied by VDD\_MEM\_IO and VDD\_IO) and the dissipation of the PHYs (supplied by own supplies). Table 7 details typical measured core and analog power dissipation figures for a range of operating modes. However, the dissipation due to the switching of the IO pins can not be given in general, but must be calculated for each application case using the following formula:

$$P_{IO} = P_{IOint} + \sum_M N \times C \times VDD_{IO}^2 \times f \quad \text{Eqn. 1}$$

where N is the number of output pins switching in a group M, C is the capacitance per pin, VDD\_IO is the IO voltage swing, f is the switching frequency and P<sub>IOint</sub> is the power consumed by the unloaded IO stage. The total power consumption of the MPC5121e processor must not exceed the value, which would cause the maximum junction temperature to be exceeded.

$$P_{total} = P_{core} + P_{analog} + P_{IO} + P_{PHYs} \quad \text{Eqn. 2}$$

**Table 7. Power Dissipation**

Core Power Supply (VDD_CORE)			SpecID
Mode	High-Performance e300 = 300 MHz, CSB = 200 MHz	Unit	
Operational <sup>1</sup> ,	800	mW	D5.1
Deep-Sleep <sup>1</sup> ,	1	mW	D5.2
Hibernation	20	uW	D5.3
PLL/OSC Power Supplies (SYS_PLL_AVDD, CORE_PLL_AVDD)			
Typical	25	mW	D5.4
Unloaded I/O Power Supplies (VDD_IO, VDD_MEM_IO)			
Typical	300	mW	D5.5
PHY Power Supplies (USB_VDDA, SATA_VDDA)			
Typical	200	mW	D5.6

<sup>1</sup> Typical core power is measured at VDD\_CORE = 1.4 V, T<sub>j</sub> = 25 C.

## 2.1.6 Thermal Characteristics

**Table 8. Thermal Resistance Data**

Rating	Board Layers	Sym	Value	Unit	SpecID
Junction to Ambient Natural Convection <sup>1,2</sup>	Single layer board (1s)	R <sub>θJA</sub>	30	°C/W	D6.1
Junction to Ambient Natural Convection <sup>1,3</sup>	Four layer board (2s2p)	R <sub>θJMA</sub>	22	°C/W	D6.2

**Table 8. Thermal Resistance Data**

Junction to Ambient (@200 ft/min) <sup>1,3</sup>	Single layer board (1s)	R <sub>θJMA</sub>	24	°C/W	D6.3
Junction to Ambient (@200 ft/min) <sup>1,3</sup>	Four layer board (2s2p)	R <sub>θJMA</sub>	19	°C/W	D6.4
Junction to Board <sup>4</sup>	—	R <sub>θJB</sub>	14	°C/W	D6.5
Junction to Case <sup>5</sup>	—	R <sub>θJC</sub>	8	°C/W	D6.6
Junction to Package Top <sup>6</sup>	Natural Convection	Ψ <sub>JT</sub>	2	°C/W	D6.7

<sup>1</sup> Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.

<sup>2</sup> Per SEMI G38-87 and JEDEC JESD51-2 with the single layer board horizontal.

<sup>3</sup> Per JEDEC JESD51-6 with the board horizontal.

<sup>4</sup> Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.

<sup>5</sup> Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1).

<sup>6</sup> Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2. When Greek letters are not available, the thermal characterization parameter is written as Psi-JT.

### 2.1.6.1 Heat Dissipation

An estimation of the chip-junction temperature, T<sub>J</sub>, can be obtained from the following equation:

$$T_J = T_A + (R_{\theta JA} \times P_D) \quad \text{Eqn. 3}$$

where:

T<sub>A</sub> = ambient temperature for the package (°C)

R<sub>θJA</sub> = junction to ambient thermal resistance (°C/W)

P<sub>D</sub> = power dissipation in package (W)

The junction to ambient thermal resistance is an industry standard value, which provides a quick and easy estimation of thermal performance. Unfortunately, there are two values in common usage: the value determined on a single layer board, and the value obtained on a board with two planes. For packages such as the PBGA, these values can be different by a factor of two. Which value is correct depends on the power dissipated by other components on the board. The value obtained on a single layer board is appropriate for the tightly packed printed circuit board. The value obtained on the board with the internal planes is usually appropriate if the board has low power dissipation and the components are well separated.

Historically, the thermal resistance has frequently been expressed as the sum of a junction to case thermal resistance and a case to ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA} \quad \text{Eqn. 4}$$

where:

R<sub>θJA</sub> = junction to ambient thermal resistance (°C/W)

R<sub>θJC</sub> = junction to case thermal resistance (°C/W)

R<sub>θCA</sub> = case to ambient thermal resistance (°C/W)

R<sub>θJC</sub> is device related and cannot be influenced by the user. You control the thermal environment to change the case to ambient thermal resistance, R<sub>θCA</sub>. For instance, you can change the air flow around the device, add a heat sink, change the mounting

arrangement on printed circuit board, or change the thermal dissipation on the printed circuit board surrounding the device. This description is most useful for ceramic packages with heat sinks where some 90% of the heat flow is through the case to the heat sink to ambient. For most packages, a better model is required.

A more accurate thermal model can be constructed from the junction to board thermal resistance and the junction to case thermal resistance. The junction to case covers the situation where a heat sink is used or where a substantial amount of heat is dissipated from the top of the package. The junction to board thermal resistance describes the thermal performance when most of the heat is conducted to the printed circuit board. This model can be used for hand estimations or for a computational fluid dynamics (CFD) thermal model.

To determine the junction temperature of the device in the application after prototypes are available, the Thermal Characterization Parameter ( $\Psi_{JT}$ ) can be used to determine the junction temperature with a measurement of the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D) \quad \text{Eqn. 5}$$

where:

$T_T$  = thermocouple temperature on top of package ( $^{\circ}\text{C}$ )

$\Psi_{JT}$  = thermal characterization parameter ( $^{\circ}\text{C}/\text{W}$ )

$P_D$  = power dissipation in package (W)

The thermal characterization parameter is measured per JESD51-2 specification using a 40-gauge type T thermocouple epoxied to the top center of the package case. The thermocouple should be positioned, so that the thermocouple junction rests on the package. A small amount of epoxy is placed over the thermocouple junction and over approximately one mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by cooling effects of the thermocouple wire.

## 2.2 Oscillator and PLL Electrical Characteristics

The MPC5121e System requires a system-level clock input SYS\_XTALI. This clock input may be driven directly from an external oscillator or with a crystal using the internal oscillator.

There is a separate oscillator for the independent Real-Time Clock (RTC) system.

The MPC5121e clock generation uses two phase locked loop (PLL) blocks.

- The system PLL (SYS\_PLL) takes an external reference frequency and generates the internal system clock. The system clock frequency is determined by the external reference frequency and the settings of the SYS\_PLL configuration.
- The e300 core PLL (CORE\_PLL) generates a master clock for all of the CPU circuitry. The e300 core clock frequency is determined by the system clock frequency and the settings of the CORE\_PLL configuration.

The USB PHY contains its own oscillator with the input USB\_XTALI and an embedded PLL.

The SATA PHY contains its own oscillator with the input SATA\_XTALI and an embedded PLL.

### 2.2.1 System Oscillator Electrical Characteristics

**Table 9. System Oscillator Electrical Characteristics**

Characteristic	Sym	Min	Typical	Max	Unit	SpecID
SYS_XTAL frequency	$f_{\text{sys\_xtal}}$	15.6	33.3	35.0	MHz	O1.1

The System Oscillator can work in Oscillator mode or in bypass mode to support an external input Clock as clock reference.

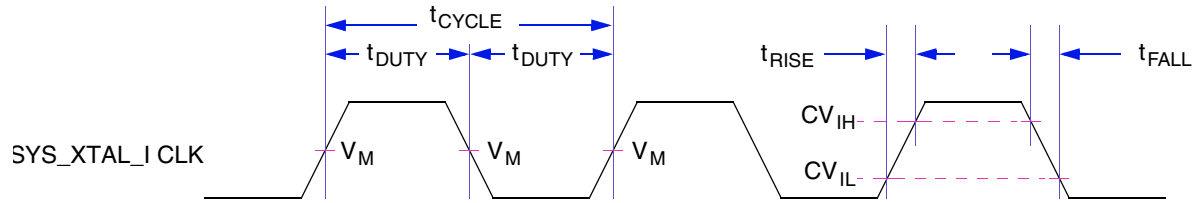


Figure 2. Timing Diagram—SYS\_XTAL\_IN

Table 10. SYS\_XTAL\_IN Timing

Sym	Description	Min	Max	Units	SpecID
$t_{CYCLE}$	SYS_XTALI cycle time. <sup>1,2</sup>	64.1	28.57	ns	O.1.2
$t_{RISE}$	SYS_XTALI rise time. <sup>3</sup>	1	4	ns	O.1.3
$t_{FALL}$	SYS_XTALI fall time. <sup>4</sup>	1	4	ns	O.1.4
$t_{DUTY}$	SYS_XTALI duty cycle (measured at $V_M$ ) <sup>5</sup>	40	60	%	O.1.5

- <sup>1</sup> The SYS\_XTALI frequency and system PLL settings must be chosen such that the resulting system frequencies do not exceed their respective maximum or minimum operating frequencies. See the MPC5121e Reference Manual.
- <sup>2</sup> The MIN/Max cycle times are calculated using  $1/f_{sys\_xtal}$  (MIN/MAX) where the  $f_{sys\_xtal}$  (MIN/MAX) (15.6/35 MHz) are taken from Table 9 (System oscillator electrical characteristics).
- <sup>3</sup> Rise time is measured from 20% of vdd to 80% of vdd.
- <sup>4</sup> Fall time is measured from 20% of vdd to 80% of vdd.
- <sup>5</sup> SYS\_XTALI duty cycle is measured at  $V_M$ .

## 2.2.2 RTC Oscillator Electrical Characteristics

Table 11. RTC Oscillator Electrical Characteristics

Characteristic	Sym	Min	Typical	Max	Unit	SpecID
RTC_XTAL frequency	$f_{rtc\_xtal}$	—	32.768	—	kHz	O2.1

## 2.2.3 System PLL Electrical Characteristics

Table 12. System PLL Specifications

Characteristic	Sym	Min	Typical	Max	Unit	SpecID
Sys PLL input clock frequency <sup>1</sup>	$f_{sys\_xtal}$	16	33.3	67	MHz	O3.1
Sys PLL input clock jitter <sup>2</sup>	$t_{jitter}$	—	—	10	ps	O3.2
Sys PLL VCO frequency <sup>1</sup>	$f_{VCOsys}$	400	—	800	MHz	O3.3
Sys PLL VCO output jitter (Dj), peak to peak / cycle	$f_{VCOjitterDj}$	—	—	40	ps	O3.4
Sys PLL VCO output jitter (Rj), rms 1 sigma	$f_{VCOjitterRj}$	—	—	12	ps	O3.5



Table 12. System PLL Specifications

Characteristic	Sym	Min	Typical	Max	Unit	SpecID
Sys PLL relock time - after power up <sup>3</sup>	t <sub>lock1</sub>	—	—	200	μs	O3.6
Sys PLL relock time - when power was on <sup>4</sup>	t <sub>lock2</sub>	—	—	170	μs	O3.7

<sup>1</sup> The SYS\_XTAL frequency and PLL Configuration bits must be chosen such that the resulting system frequency, CPU (core) frequency, and PLL (VCO) frequency do not exceed their respective maximum or minimum operating frequencies.

<sup>2</sup> This represents total input jitter - short term and long term combined. Two different types of jitter can exist on the input to CORE\_SYSCLK, systemic and true random jitter. True random jitter is rejected. Systemic jitter is passed into and through the PLL to the internal clock circuitry.

<sup>3</sup> PLL-relock time is the maximum amount of time required for the PLL lock after a stable VDD and CORE\_SYSCLK are reached during the power-on reset sequence.

<sup>4</sup> PLL-relock time is the maximum amount of time required for the PLL lock after the PLL has been disabled and subsequently re-enabled during sleep modes.

## 2.2.4 e300 Core PLL Electrical Characteristics

The internal clocking of the e300 core is generated from and synchronized to the system clock by means of a voltage-controlled core PLL.

Table 13. e300 PLL Specifications

Characteristic	Sym	Min	Typical	Max	Unit	SpecID
e300 frequency <sup>1</sup>	f <sub>core</sub>	50	—	400	MHz	O4.1
e300 PLL VCO frequency <sup>1</sup>	f <sub>VCOcore</sub>	400	—	800	MHz	O4.3
e300 PLL input clock frequency	f <sub>CSB_CLK</sub>	50	—	200	MHz	O4.4
e300 PLL input clock cycle time	t <sub>CSB_CLK</sub>	5	—	20	ns	O4.5
e300 PLL relock time <sup>2</sup>	t <sub>lock</sub>	—	—	200	μs	O4.6

<sup>1</sup> The frequency and e300 PLL Configuration bits must be chosen such that the resulting system frequencies, CPU (core) frequency, and e300 PLL (VCO) frequency do not exceed their respective maximum or minimum operating frequencies in [Table 14](#).

<sup>2</sup> PLL-relock time is the maximum amount of time required for the PLL lock after a stable VDD and CORE\_SYSCLK are reached during the power-on reset sequence. This specification also applies when the PLL has been disabled and subsequently re-enabled during sleep modes.

## 2.3 AC Electrical Characteristics

### 2.3.1 Overview

Hyperlinks to the indicated timing specification sections are provided below.

- AC Operating Frequency Data
- Resets
- External Interrupts
- SDRAM (DDR)
- PCI
- LPC
- NFC
- PATA
- SATA PHY
- FEC
- USB ULPI
- On-Chip USB PHY
- SDHC
- DIU
- SPDIF
- CAN
- I<sup>2</sup>C
- J1850
- PSC
- GPIOs and Timers
- Fusebox
- IEEE 1149.1 (JTAG)

AC Test Timing Conditions:

Unless otherwise noted, all test conditions are as follows:

- TA = 0 to 70°C
- VDD\_CORE = 1.33 to 1.47 V  
VDD\_IO = 3.0 to 3.6 V
- Input conditions:  
All Inputs: tr, tf ≤ 1 ns
- Output Loading:  
All Outputs: 50 pF

### 2.3.2 AC Operating Frequency Data

Table 14 provides the operating frequency information for the MPC5121e.

**Table 14. Clock Frequencies**

	Min	Max	Units	SpecID
e300 Processor Core	214.5	400	MHz	A1.1
SDRAM Clock	28.6	200	MHz	A1.2
CSB Bus Clock	28.6	200	MHz	A1.3
IP Bus Clock	4.8	83	MHz	A1.4
PCI Clock	4.43	66	MHz	A1.5
LPC Clock	1.2	83	MHz	A1.6
NFC Clock	1.2	83	MHz	A1.7

Table 14. Clock Frequencies (continued)

	Min	Max	Units	SpecID
DIU Clock	0.45	66.6	MHz	A1.8
MBX Clock	3.57	100	MHz	A1.9

## NOTES:

1. The SYS\_XTAL\_IN frequency, Sys PLL, and CORE PLL settings must be chosen so that the resulting e300 clk, csb\_clk, MCK, frequencies do not exceed their respective maximum or minimum operating frequencies.
2. The values are valid for the user-operation mode. There can be deviations for test modes.
3. The selection of the peripheral clock frequencies needs to take care about requirements for baud rates and minimum frequency limitation.
4. The DDR data rate is 2x the DDR memory bus frequency.

SYS\_XTAL\_IN is the input clock multiplied up by the system phase-locked loop (SYS PLL) and the clock unit to create the coherent system bus clock (csb\_clk), the internal clock for the DDR controller (ddr\_clk), and the clocks for the peripherals. The csb\_clk serves as the clock input to the e300 core. A second PLL inside the e300 core multiplies the csb\_clk frequency to create the internal clock for the e300 core (core\_clk). The system and core PLL multipliers are selected by the SPMF and COREPLL fields in the reset configuration word, which is loaded at power-on reset

See the MPC5121e Reference Manual for more information on the clock subsystem.

### 2.3.3 Resets

The MPC5121e has three reset pins:

- $\overline{\text{PORESET}}$  - Power on Reset
- $\overline{\text{HRESET}}$  - Hard Reset
- $\overline{\text{SRESET}}$  - Software Reset

These signals are asynchronous I/O signals and can be asserted at any time. The input side uses a Schmitt trigger and requires the same input characteristics as other MPC5121e inputs, as specified in the DC Electrical Specifications section.

As long as VDD is not stable the  $\overline{\text{HRESET}}$  output is not stable.

Table 15. Reset Rise / Fall Timing

Description	Min	Max	Unit	SpecID
$\overline{\text{PORESET}}$ <sup>1</sup> fall time	—	1	ms	A3.4
$\overline{\text{PORESET}}$ rise time	—	1	ms	A3.5
$\overline{\text{HRESET}}$ <sup>2,3</sup> fall time	—	1	ms	A3.6
$\overline{\text{HRESET}}$ rise time	—	1	ms	A3.7
$\overline{\text{SRESET}}$ fall time	—	1	ms	A3.8
$\overline{\text{SRESET}}$ rise time	—	1	ms	A3.9

<sup>1</sup> Make sure that the  $\overline{\text{PORESET}}$  does not carry any glitches. The MPC5121e has no filter to prevent them from getting into the chip.

<sup>2</sup>  $\overline{\text{HRESET}}$  and  $\overline{\text{SRESET}}$  must have a monotonous rise time.

<sup>3</sup> The assertion of  $\overline{\text{HRESET}}$  becomes active at Power on Reset without any SYS\_XTAL clock.

The timing relationship can be seen below.

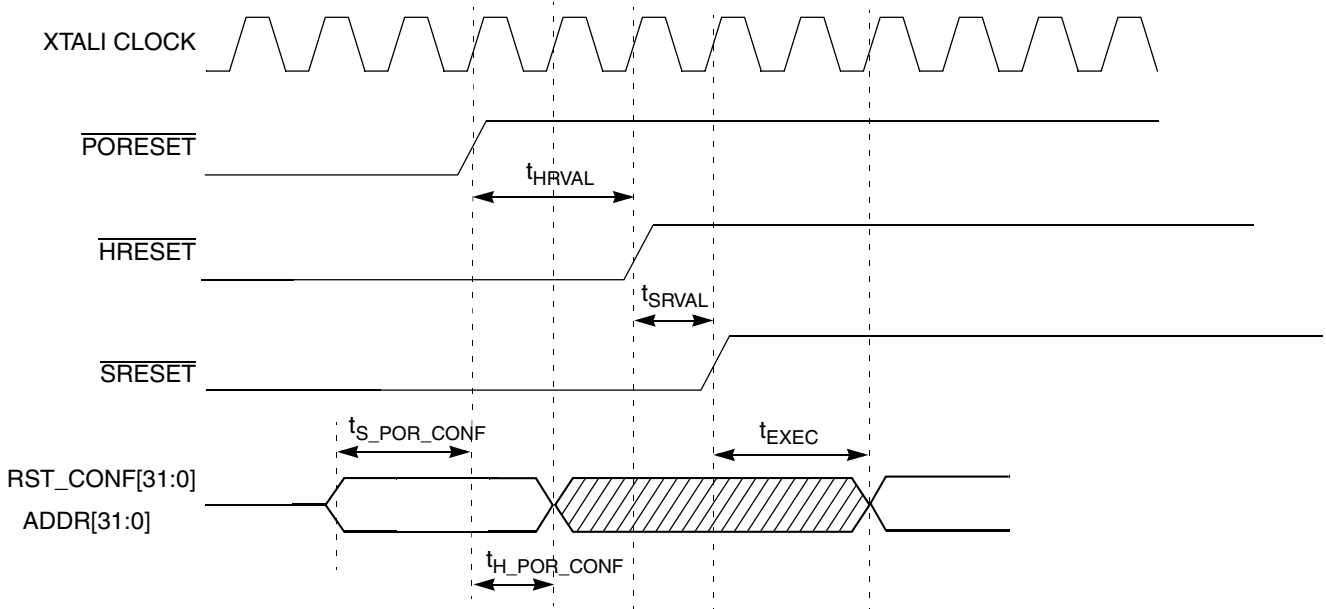


Figure 3. Power-Up Behavior

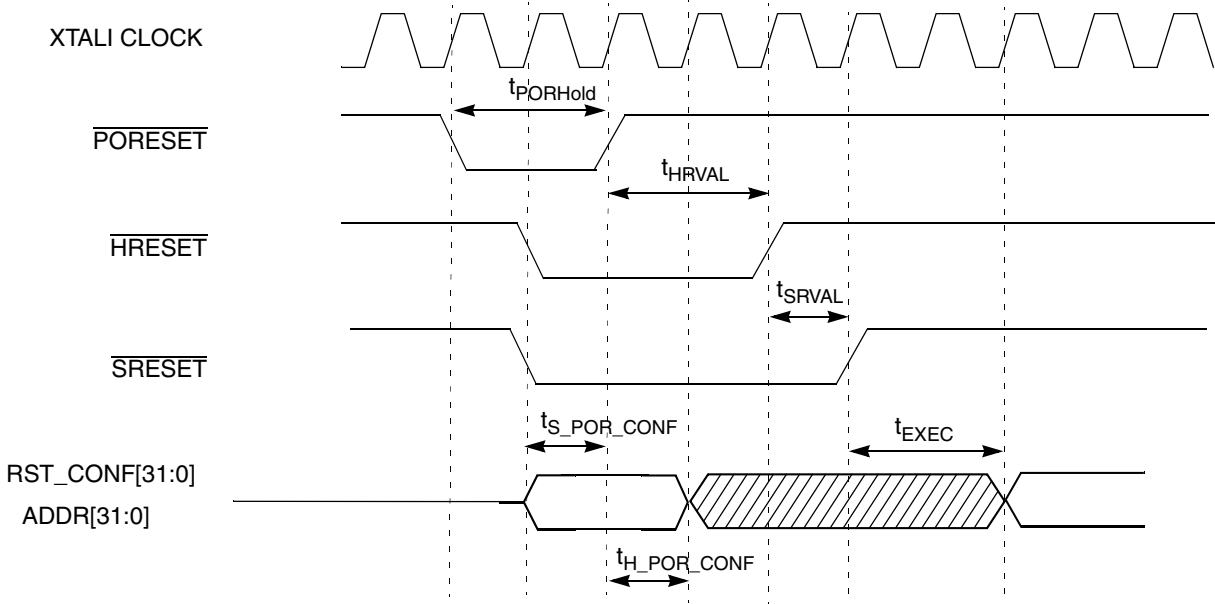


Figure 4. Power-On Reset Behavior

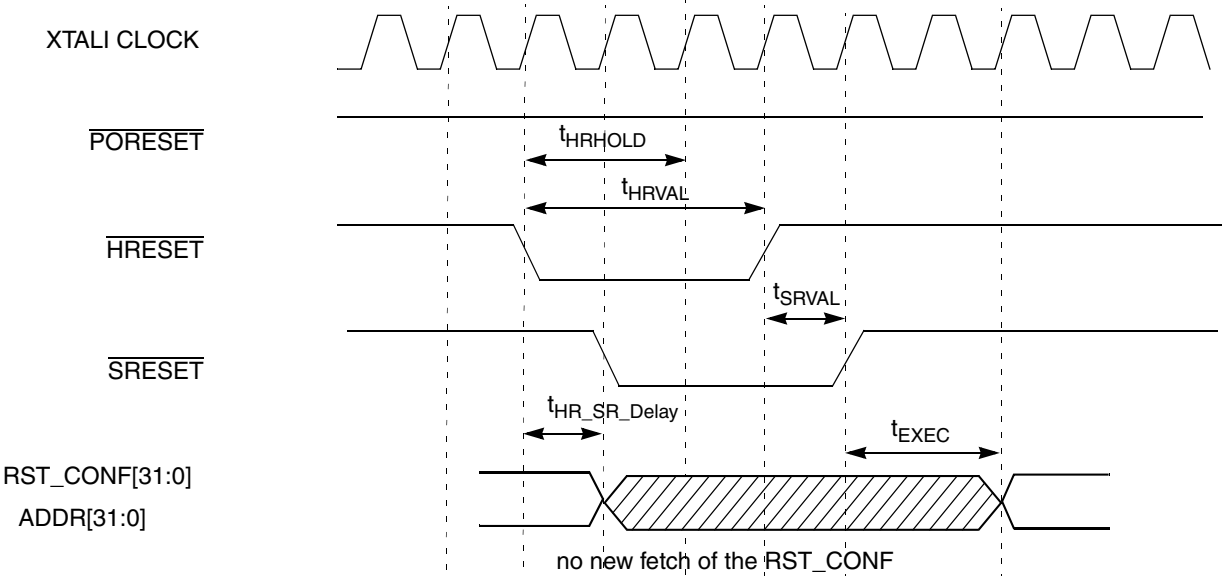


Figure 5. HRESET Behavior

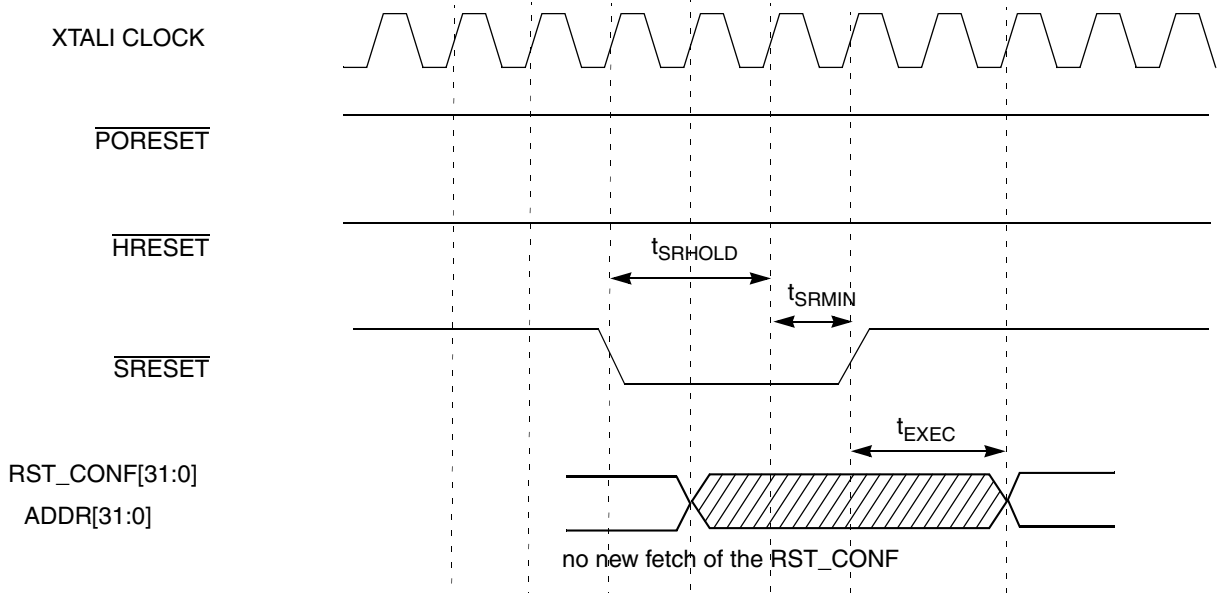


Figure 6. SRESET Behavior

Table 16. Reset Timing

Symbol	Description	Value (XTALI CLOCK)	SpecID
t <sub>PORHOLD</sub>	Time $\overline{\text{PORESET}}$ must be held low before a qualified reset occurs	4 cycles	A3.10
t <sub>HRVAL</sub>	Time $\overline{\text{HRESET}}$ is asserted after a qualified reset occurs	26810 cycles	A3.11
t <sub>SRVAL</sub>	Time $\overline{\text{SRESET}}$ is asserted after assertion of $\overline{\text{HRESET}}$	21 cycles	A3.12
t <sub>EXEC</sub>	Time between $\overline{\text{SRESET}}$ assertion and first core instruction fetch	4 cycles	A3.13

Table 16. Reset Timing (continued)

Symbol	Description	Value (XTALI CLOCK)	SpecID
$t_{S\_POR\_CONF}$	Reset configuration setup time before assertion of $\overline{PORESET}$	1 cycle	A3.14
$t_{H\_POR\_CONF}$	Reset configuration hold time after assertion of $\overline{PORESET}$	1 cycle	A3.15
$t_{HR\_SR\_DELAY}$	Time $\overline{HRESET}$ must be held low before a qualified $\overline{SRESET}$ occurs	2 cycles	A3.16
$t_{HRHOLD}$	Time $\overline{HRESET}$ must be held low before a qualified reset occurs	4 cycles	A3.17
$t_{SRHOLD}$	Time $\overline{SRESET}$ must be held low before a qualified reset occurs	4 cycles	A3.18
$t_{SRMIN}$	Time $\overline{SRESET}$ is asserted after it has been qualified	1 cycles	A3.19

### 2.3.4 External Interrupts

The MPC5121e provides three different kinds of external interrupts:

- IRQ interrupts
- GPIO interrupts with simple interrupt capability (not available in power-down mode)
- WakeUp interrupts

Table 17. IPIC Input AC Timing Specifications

Description	Symbol	Min	Unit	SpecID
IPIC inputs - minimum pulse width	$t_{PICWID}$	20	ns	A4.1

IPIC inputs must be valid for at least  $t_{PICWID}$  to ensure proper operation in edge triggered mode.

### 2.3.5 SDRAM (DDR)

The MPC5121e memory controller supports three types of DDR devices:

- DDR-1 (SSTL\_2 class II interface)
- DDR-2 (SSTL\_18 interface)
- LPDDR (1.8V I/O supply voltage)

JEDEC standards define the minimum set of requirements for compliant memory devices:

- JEDEC STANDARD, DDR2 SDRAM SPECIFICATION, JESD79-2C, MAY 2006
- JEDEC STANDARD, Double Data Rate (DDR) SDRAM Specification, JESD79E, May 2005
- JEDEC STANDARD, Low Power Double Data Rate (LPDDR) SDRAM Specification, JESD79-4, May 2006

The MPC5121e supports the configuration of two output drive strength for DDR2 and LPDDR:

- full drive strength
- half drive strength (intended for lighter loads or point-to-point environments)

The MPC5121e memory controller supports dynamic on-die termination in the host device and in the DDR2 memory device.

This section includes AC specifications for all DDR SDRAM pins. The DC parameters are specified in the DC Electrical Characteristics.

### 2.3.5.1 DDR and DDR2 SDRAM AC Timing Specifications

Table 18. DDR and DDR2 (DDR2-400) SDRAM Timing Specifications

At recommended operating conditions with VDD\_MEM\_IO of  $\pm 5\%$

Parameter	Symbol	Min	Max	Unit	SpecID
Clock cycle time, CL=x	$t_{CK}$	5000	8000	ps	A5.1
CK HIGH pulse width	$t_{CH}$	0.45	0.55	$t_{CK}$	A5.2
CK LOW pulse width	$t_{CL}$	0.45	0.55	$t_{CK}$	A5.3
DQS latching rising transitions to associated clock edges	$t_{DQSS}$	-0.25	0.25	$t_{CK}$	A5.4
DQS falling edge to CK setup time	$t_{DSS}$	-0.2	—	$t_{CK}$	A5.5
DQS falling edge hold time from CK	$t_{DSH}$	-0.2	—	$t_{CK}$	A5.6
Address and control output setup time relative to MCK rising edge	$t_{IS(base)}$	350	—	ps	A5.9
Address and control output hold time relative to MCK rising edge	$t_{IH(base)}$	475	—	ps	A5.10
DQ and DM output setup time relative to DQS	$t_{DS1(base)}$	25	—	ps	A5.12
DQ and DM output hold time relative to DQS	$t_{DH1(base)}$	25	—	ps	A5.13
DQS-DQ skew for DQS and associated DQ inputs	$t_{DQSQ}$	—	350	ps	A5.15
CK half pulse width	$t_{HP}$	$\min(t_{CL}, t_{CH})$	—	ps	A5.16
DQ/DQS output hold time relative to DQS	$t_{QH}$	$t_{HP} - t_{QHS} - 200$	—	ps	A5.17
ODT turn-on delay	$t_{AOND}$	2	—	$t_{CK}$	A5.18
ODT turn-on	$t_{AON}$	$t_{AC}(\min)$	—	ns	A5.19
ODT turn-on (Power-Down mode)	$t_{AONPD}$	$t_{AC}(\min) + 2$	—	ns	A5.20
ODT turn-off delay	$t_{AOFD}$	2.5	—	$t_{CK}$	A5.21
ODT turn-off	$t_{AOF}$	$t_{AC}(\min)$	—	ns	A5.22
ODT turn-off (Power-Down mode)	$t_{AOFPD}$	$t_{AC}(\min) + 2$	—	ns	A5.23
ODT to power down entry latency	$t_{ANPD}$	3	—	$t_{CK}$	A5.24
ODT power down exit latency	$t_{AXPD}$	8	—	$t_{CK}$	A5.25
OCD drive mode output delay	$t_{OIT}$	0	—	$t_{CK}$	A5.26
Minimum time clocks remains ON after CKE asynchronously drops LOW	$t_{Delay}$	$t_{IS} + t_{CK} + t_{IH}$	—	ns	A5.27

The JEDEC parameter  $t_{QHS}$  is defined by the DDR memory.

Figure 7 shows the DDR SDRAM write timing.

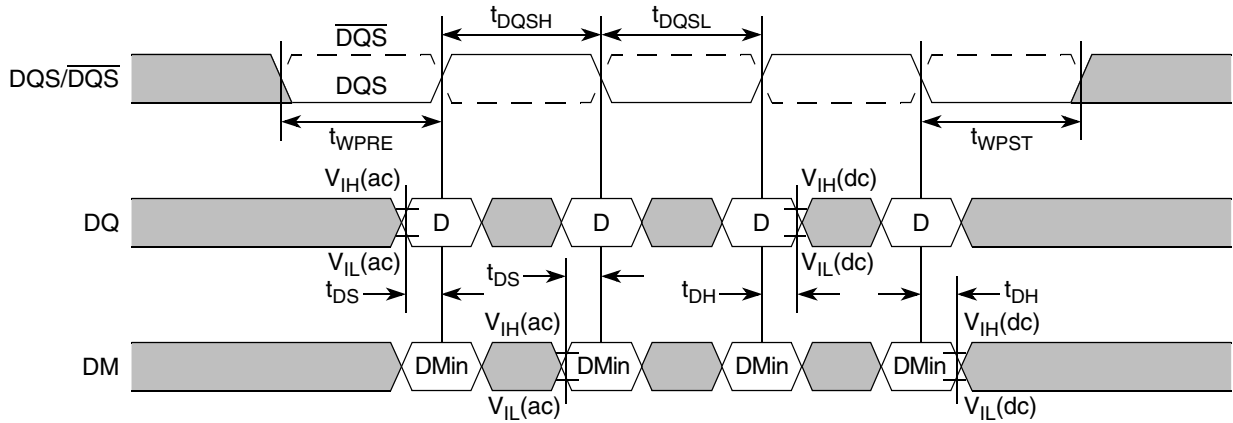


Figure 7. DDR Write Timing

Figure 8 shows the DDR SDRAM read timing.

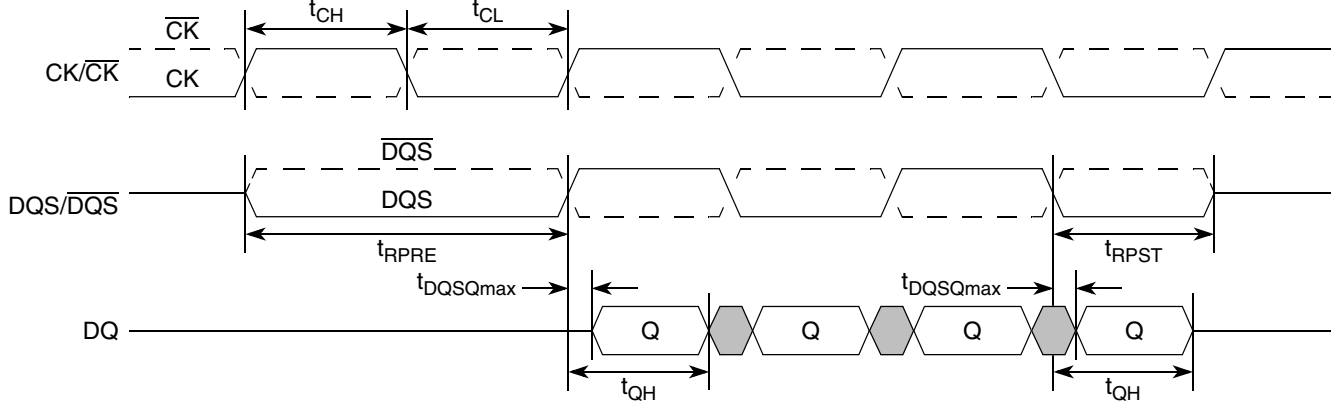


Figure 8. DDR Read Timing

Figure 9 provides the AC test load for the DDR bus.

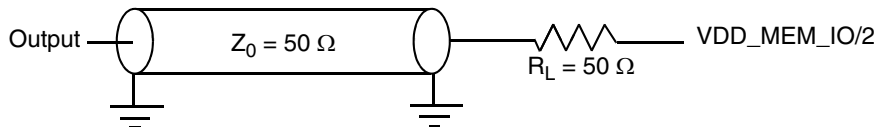


Figure 9. DDR AC Test Load

### 2.3.6 PCI

The PCI interface on the MPC5121e is designed to PCI Version 2.3 and supports 33 and 66 MHz PCI operations. See the PCI Local Bus Specification; the component section specifies the electrical and timing parameters for PCI components with the intent that components connect directly together whether on the planar or an expansion board, without any external buffers or other glue logic. Parameters apply at the package pins, not at expansion board edge connectors.

The PCI\_CLK is used as output clock, the MPC5121e is a PCI host device only.

Figure 10 shows the clock waveform and required measurement points for 3.3 V signaling environments. Table 20 summarizes the clock specifications.



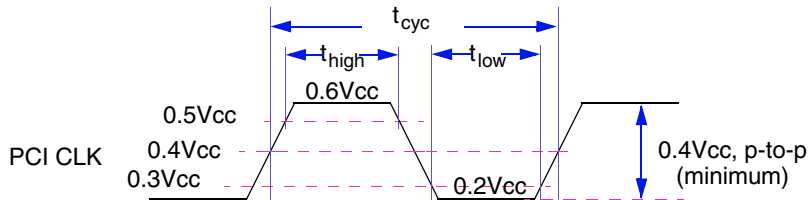


Figure 10. PCI CLK Waveform

Table 20. PCI CLK Specifications

Sym	Description	66 MHz <sup>1</sup>		33 MHz		Units	SpecID
		Min <sup>2</sup>	Max	Min	Max		
$t_{cyc}$	PCI CLK Cycle Time <sup>1,3</sup>	15	30	30	—	ns	A6.1
$t_{high}$	PCI CLK High Time	6	—	11	—	ns	A6.2
$t_{low}$	PCI CLK Low Time	6	—	11	—	ns	A6.3
—	PCI CLK Slew Rate <sup>2</sup>	1.5	4	1	4	V/ns	A6.4

<sup>1</sup> In general, all 66 MHz PCI components must work with any clock frequency up to 66 MHz. CLK requirements vary depending upon whether the clock frequency is above 33 MHz.

<sup>2</sup> Rise and fall times are specified in terms of the edge rate measured in V/ns. This slew rate must be met across the minimum peak-to-peak portion of the clock waveform as shown in Figure 10.

<sup>3</sup> The minimum clock period must not be violated for any single clock cycle, i.e., accounting for all system jitter.

Table 21. PCI Timing Parameters<sup>1</sup>

Sym	Description	66 MHz		33 MHz		Units	SpecID
		Min <sup>2</sup>	Max	Min	Max		
$t_{val}$	CLK to Signal Valid Delay – bused signals <sup>1,2,3</sup>	2	6	2	11	ns	A6.5
$t_{val}(ptp)$	CLK to Signal Valid Delay – point to point <sup>1,2,3</sup>	2	6	2	12	ns	A6.6
$t_{on}$	Float to Active Delay <sup>1</sup>	2	—	2	—	ns	A6.7
$t_{off}$	Active to Float Delay <sup>1</sup>	—	14	—	28	ns	A6.8
$t_{su}$	Input Setup Time to CLK – bused signals <sup>3,4</sup>	3	—	7	—	ns	A6.9
$t_{su}(ptp)$	Input Setup Time to CLK – point to point <sup>3,4</sup>	5	—	10,12	—	ns	A6.10
$t_h$	Input Hold Time from CLK <sup>4</sup>	0	—	0	—	ns	A6.11

<sup>1</sup> See the timing measurement conditions in the PCI Local Bus Specification. It is important that all driven signal transitions drive to their Voh or Vol level within one Tcyc.

<sup>2</sup> Minimum times are measured at the package pin with the load circuit, and maximum times are measured with the load circuit as shown in the PCI Local Bus Specification.

<sup>3</sup> REQ# and GNT# are point-to-point signals and have different input setup times than do bused signals. GNT# and REQ# have a setup of 5 ns at 66 MHz. All other signals are bused.

<sup>4</sup> See the timing measurement conditions in the PCI Local Bus Specification.

## Electrical and Thermal Characteristics

For Measurement and Test Conditions, see the PCI Local Bus Specification.

### 2.3.7 LPC

The Local Plus Bus is the external bus interface of the MPC5121e. A maximum of eight configurable chip selects (CS) are provided. There are two main modes of operation: non-MUXed and MUXED. The reference clock is the LPC CLK. The maximum bus frequency is 83 MHz.

Definition of Acronyms and Terms:

WS = Wait State

DC = Dead Cycle

HC = Hold Cycle

DS = Data Size in Bytes

LB = Long Burst

AL = Address latch enable Length

OR = Odd Ratio between  $t_{IPScck}$  and  $t_{LPCcck}$ ; e.g. OR is one when  $t_{IPScck}:t_{LPCcck}$  is 3:1, 5:1 or 7:1

$t_{LPCcck}$  = LPC clock period

$t_{IPScck}$  = IPS clock period

**Table 22. LPC Timing**

Sym	Description	Min	Max	Units	SpecID
$t_{OD}$	$\overline{CS}[x]$ , ADDR, R/ $\overline{W}$ , TSIZ, DATA (wr), $\overline{TS}$ , $\overline{OE}$ valid after LPC CLK (Output Delay related to LPC CLK)	0	5	ns	A7.1
$t_1$	non-Burst $\overline{CS}[x]$ pulse width	$(2+WS)*t_{LPCcck}$	$(2+WS)*t_{LPCcck}$	ns	A7.2
$t_2$	ADDR, R/ $\overline{W}$ , TSIZ, DATA (wr) valid before $\overline{CS}[x]$ assertion	$t_{LPCcck}-t_{OD}$	$t_{LPCcck}+t_{OD}$	ns	A7.3
$t_3$	$\overline{OE}$ assertion after $\overline{CS}[x]$ assertion	$t_{LPCcck}-t_{OD}$	$t_{LPCcck}+t_{OD}$	ns	A7.4
$t_4$	ADDR, R/ $\overline{W}$ , TSIZ, Data (wr) hold after $\overline{CS}[x]$ negation	$t_{LPCcck}-t_{OD}$	$(HC + 1) * t_{LPCcck} + t_{OD}$	ns	A7.5
$t_5$	$\overline{TS}$ pulse width	$t_{LPCcck}$	$t_{LPCcck}$	ns	A7.6
$t_6$	DATA (rd) setup before LPC CLK	4	-	ns	A7.7
$t_7$	DATA (rd) input hold	0	$(DC+1)*t_{LPCcck}$	ns	A7.8
$t_8$	read Burst $\overline{CS}[x]$ pulse width	$(2+WS+4^{LB*2*(32/DS)}) * t_{LPCcck}$	$(2+WS+4^{LB*2*(32/DS)}) * t_{LPCcck}$	ns	A7.9
$t_9$	Burst $\overline{ACK}$ pulse width	$(4^{LB*2*(32/DS)}) * t_{LPCcck}$	$(4^{LB*2*(32/DS)}) * t_{LPCcck}$	ns	A7.10
$t_{10}$	Burst DATA (rd) input hold	0	-	ns	A7.11
$t_{11}$	read Burst $\overline{ACK}$ assertion after $\overline{CS}[x]$ assertion	$(2+WS)*t_{LPCcck}$	$(2+WS)*t_{LPCcck}$	ns	A7.12
$t_{12}$	write Burst $\overline{CS}[x]$ pulse width	$(t_{OD}+WS+4^{LB*2*(32/DS)}) * t_{LPCcck} + OR * t_{IPScck}$	$(t_{OD}+WS+4^{LB*2*(32/DS)}) * t_{LPCcck} + OR * t_{IPScck}$	ns	A7.13
$t_{13}$	write Burst ADDR, R/ $\overline{W}$ , TSIZ, DATA (wr) hold after $\overline{CS}[x]$ negation	$0.5 * t_{LPCcck} - OR * t_{IPScck} - t_{OD}$	$(HC+0.5) * t_{LPCcck} - OR * t_{IPScck} + t_{OD}$	ns	A7.14
$t_{14}$	write Burst $\overline{ACK}$ assertion after $\overline{CS}[x]$ assertion	$(t_{OD}+WS) * t_{LPCcck} + OR * t_{IPScck} - t_{OD}$	$(t_{OD}+WS) * t_{LPCcck} + OR * t_{IPScck} + t_{OD}$	ns	A7.15

Table 22. LPC Timing (continued)

Sym	Description	Min	Max	Units	SpecID
t <sub>15</sub>	write Burst DATA valid	t <sub>LPCck</sub> -t <sub>OD</sub>	-	ns	A7.16
t <sub>16</sub>	non-Muxed Mode: asynchronous write Burst ADDR valid before write DATA valid	0.5*t <sub>LPCck</sub> -OR*t <sub>IPScck</sub> -t <sub>OD</sub>	0.5*t <sub>LPCck</sub> -OR*t <sub>IPScck</sub> +t <sub>OD</sub>	ns	A7.17
t <sub>17</sub>	MUXed Mode: ADDR cycle	AL*2*t <sub>LPCck</sub> -t <sub>OD</sub>	AL*2*t <sub>LPCck</sub>	ns	A7.18
t <sub>18</sub>	MUXed Mode: $\overline{\text{ALE}}$ cycle	AL*t <sub>LPCck</sub>	AL*t <sub>LPCck</sub>	ns	A7.19
t <sub>19</sub>	non-MUXed Mode Page Burst: ADDR cycle	t <sub>LPCck</sub> -t <sub>OD</sub>	t <sub>LPCck</sub>	ns	A7.20
t <sub>20</sub>	non-MUXed Mode Page Burst: Burst DATA (rd) input setup before next ADDR cycle	t <sub>OD</sub> + t <sub>6</sub>	—	ns	A7.21
t <sub>21</sub>	non-MUXed Mode Page Burst: Burst DATA (rd) input hold after next ADDR cycle	0	—	ns	A7.22

### 2.3.7.1 Non-MUXed Mode

#### 2.3.7.1.1 Non-Muxed non-Burst Mode

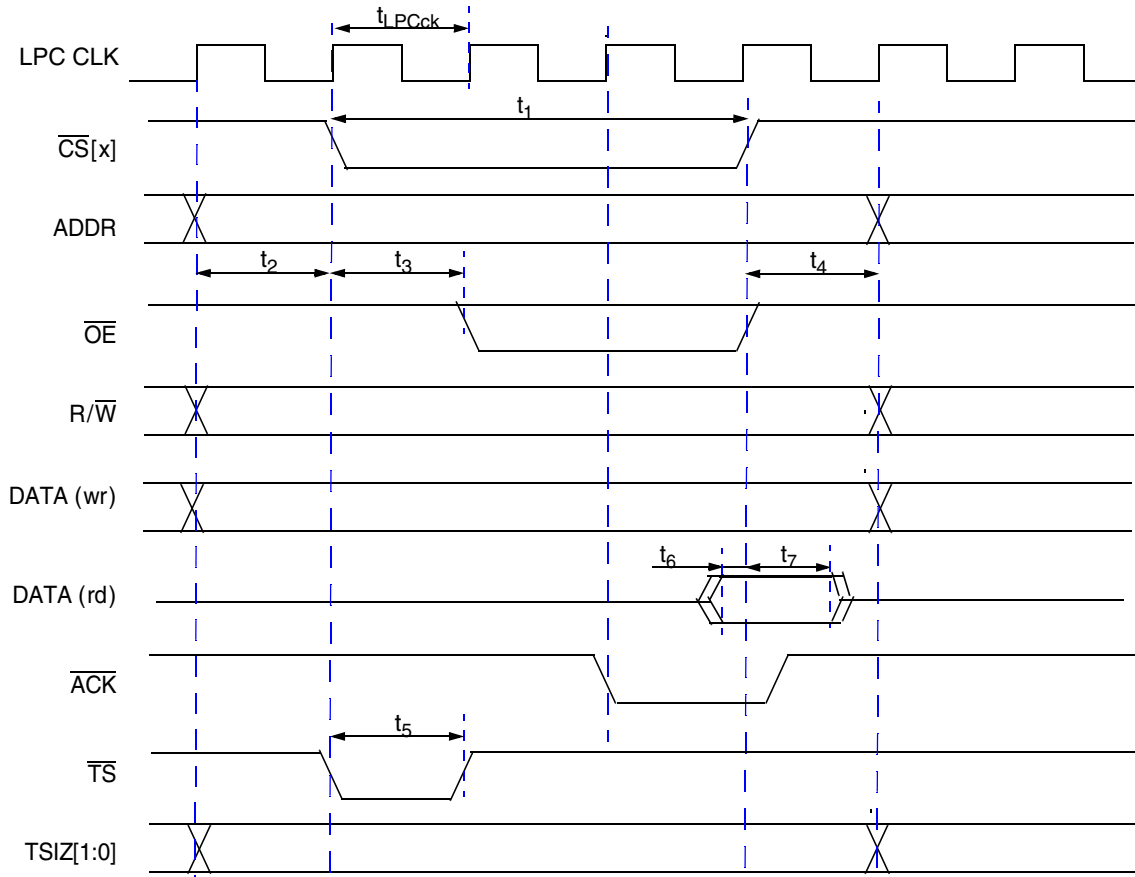


Figure 11. Timing Diagram – Non-MUXed non-Burst Mode

**NOTE**

$\overline{ACK}$  is asynchronous input signal and has no timing requirements.  $\overline{ACK}$  needs to be deasserted after  $\overline{CS}[x]$  is deasserted.

2.3.7.1.2 Non-Muxed Synchronous Read Burst Mode

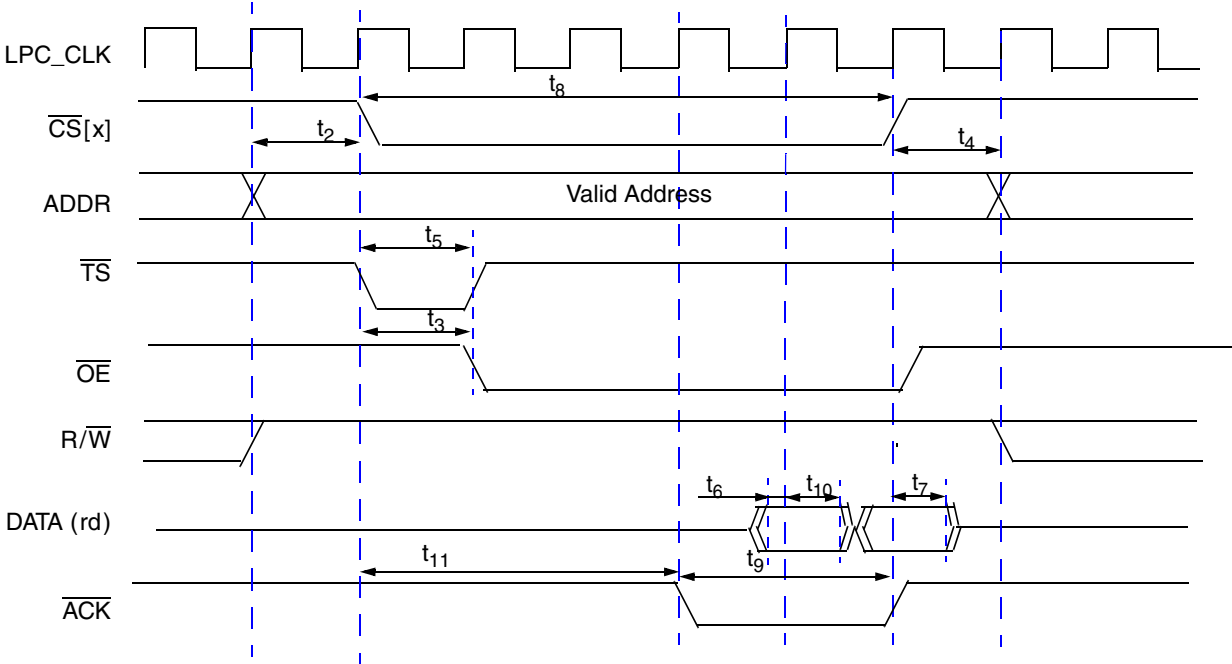


Figure 12. Timing Diagram – Non-MUXed Synchronous Read Burst Mode

2.3.7.1.3 Non-Muxed Synchronous Write Burst Mode

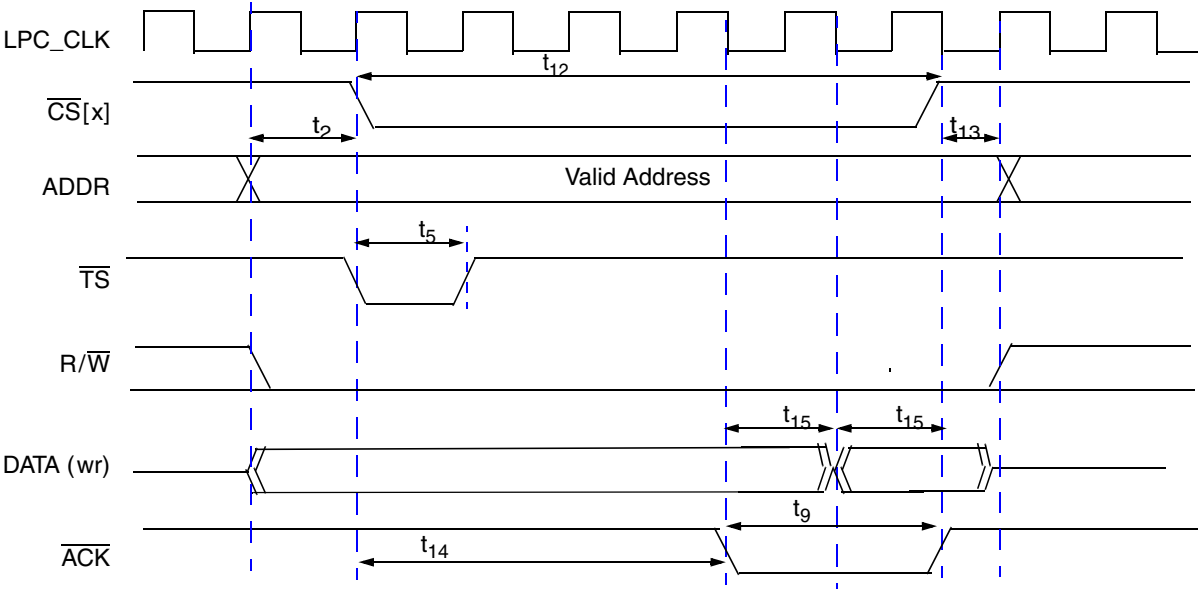


Figure 13. Timing Diagram – Non-MUXed Synchronous Write Burst

### 2.3.7.1.4 Non-MUXed Asynchronous Read Burst Mode (Page Mode)

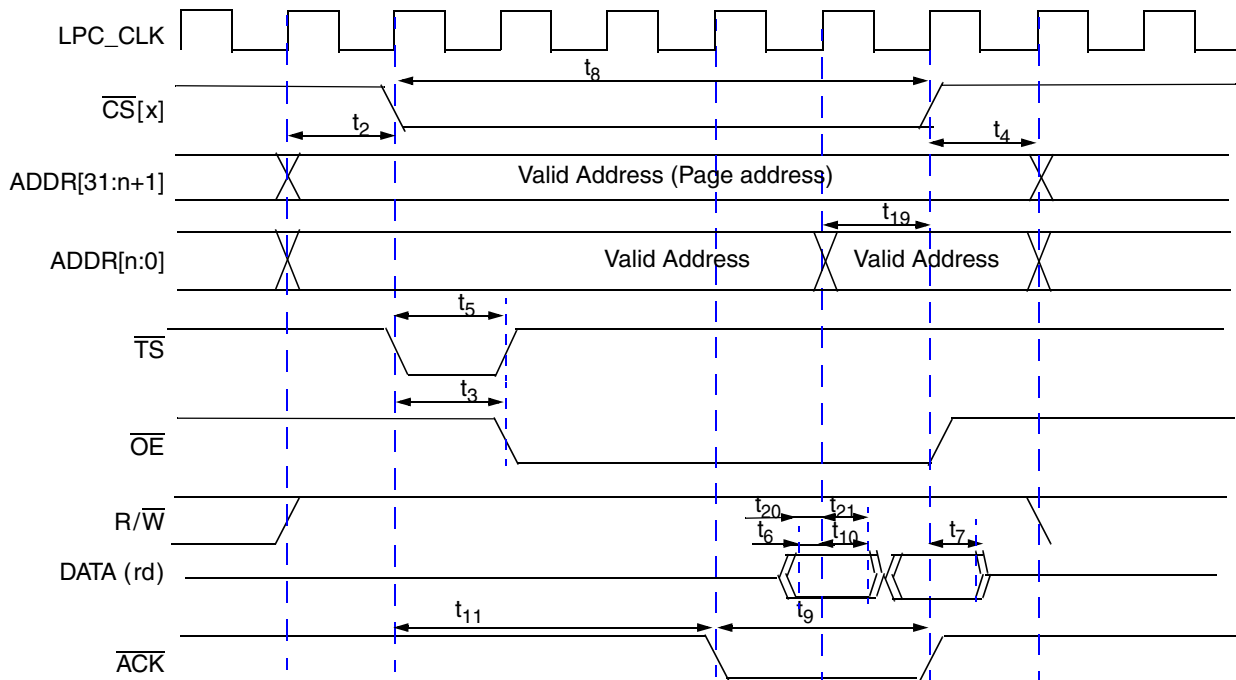


Figure 14. Timing Diagram – Non-MUXed Asynchronous Read Burst

### 2.3.7.1.5 Non-MUXed Asynchronous Write Burst Mode

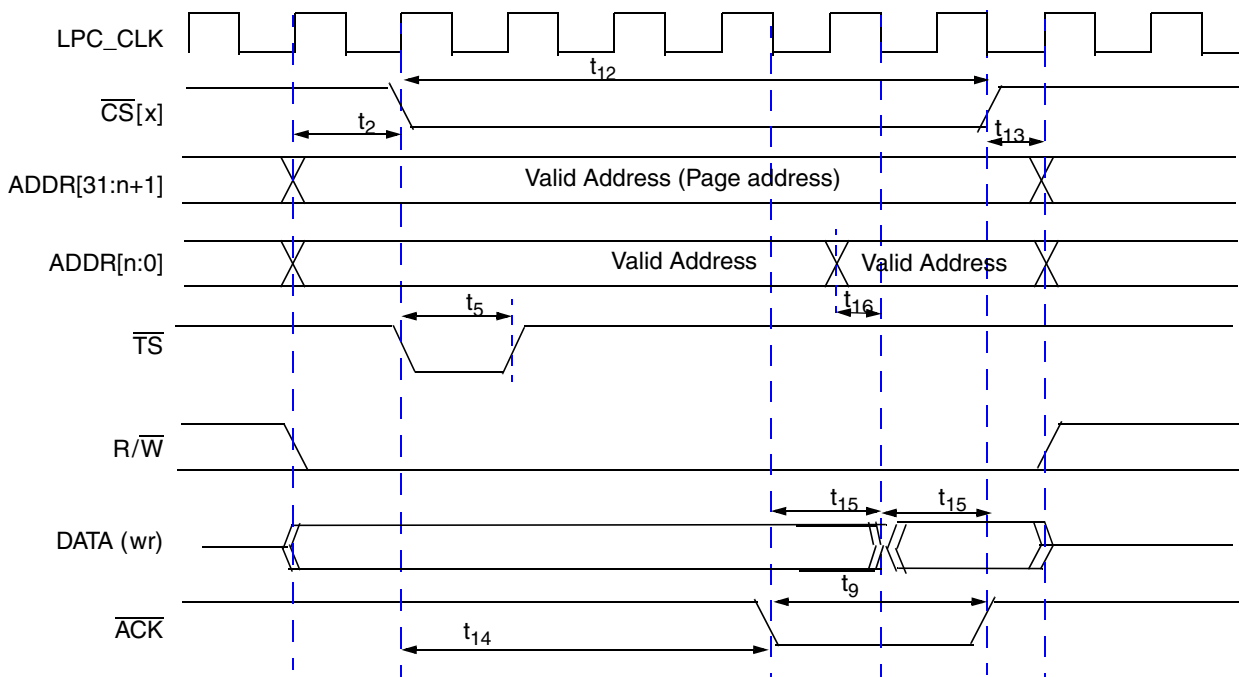


Figure 15. Timing Diagram – Non-MUXed Asynchronous Write Burst

### 2.3.7.2 MUXed Mode

#### 2.3.7.2.1 MUXed non-Burst Mode

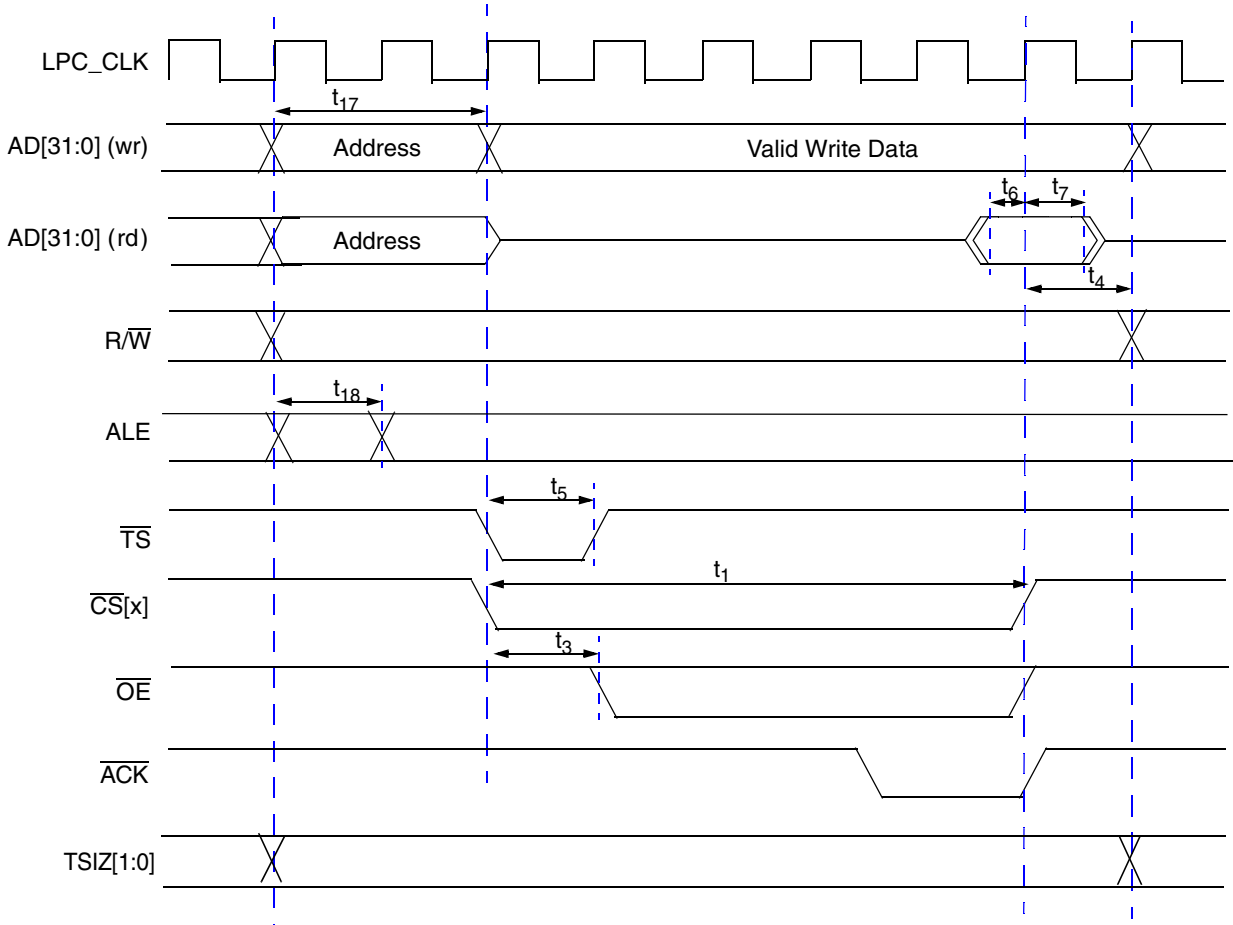


Figure 16. Timing Diagram – MUXed non-Burst Mode

**NOTE**

$\overline{ACK}$  is asynchronous input signal and has no timing requirements.  $\overline{ACK}$  needs to be deasserted after  $CS[x]$  is deasserted.

### 2.3.7.2.2 MUXed Synchronous Read Burst Mode

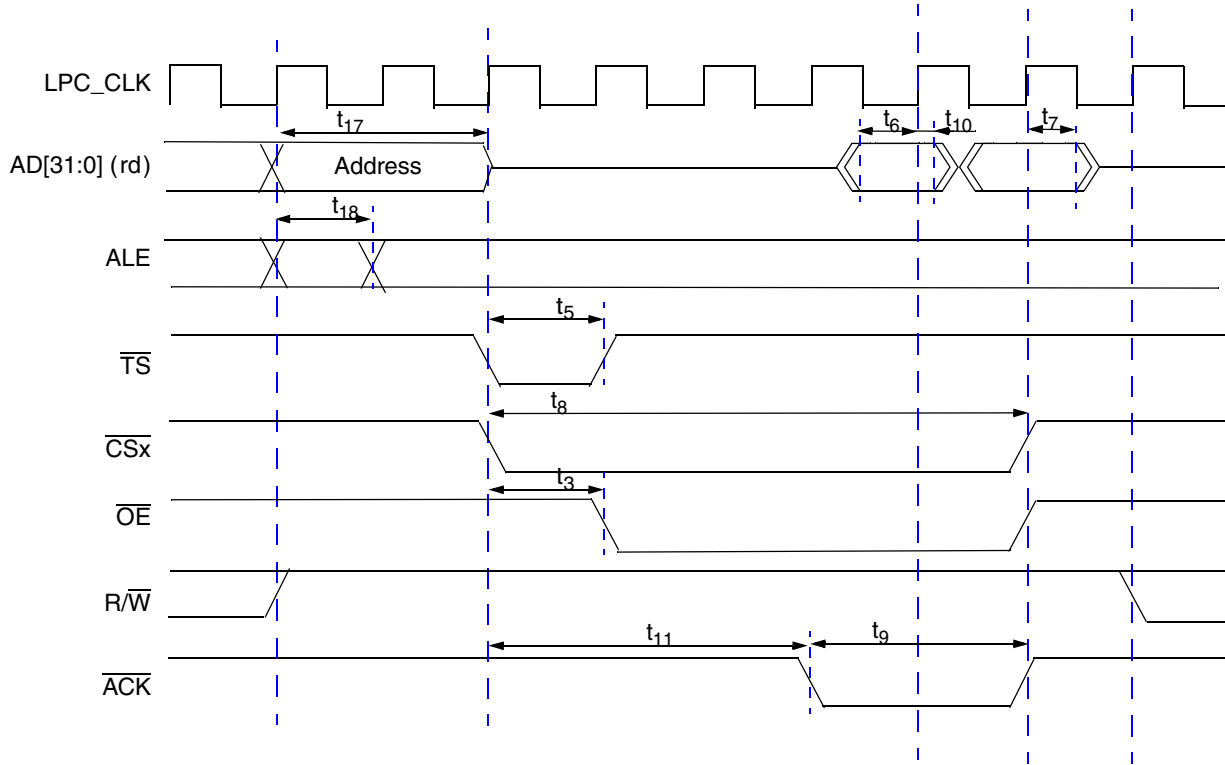


Figure 17. Timing Diagram – MUXed Synchronous Read Burst

### 2.3.7.2.3 MUXed Synchronous Write Burst Mode

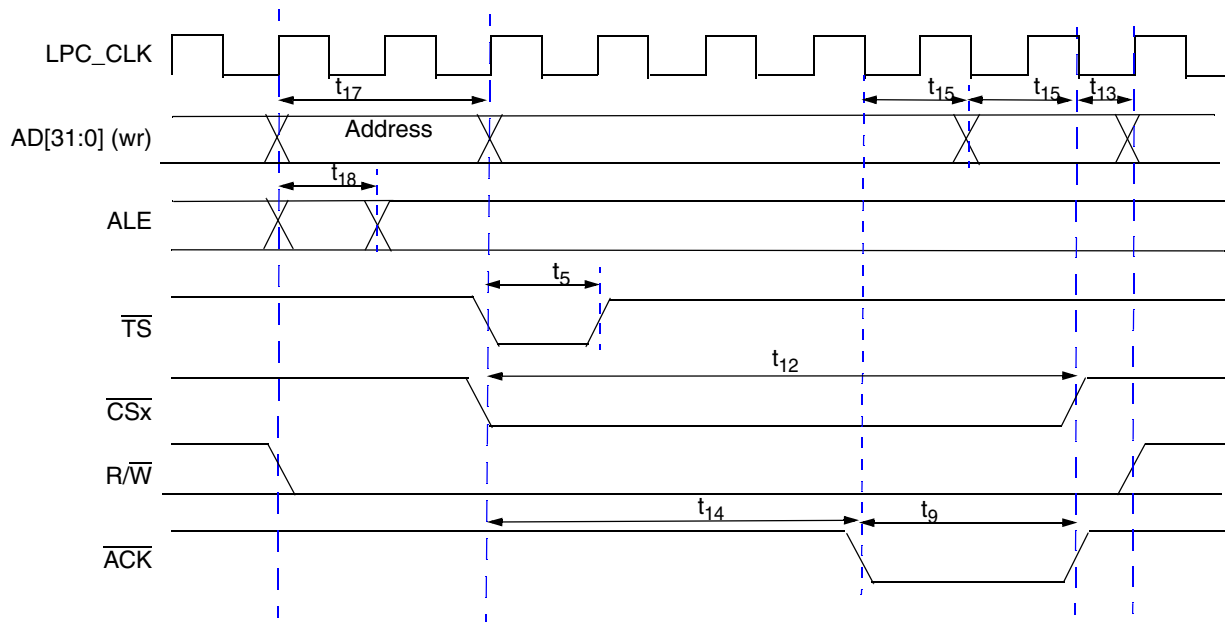
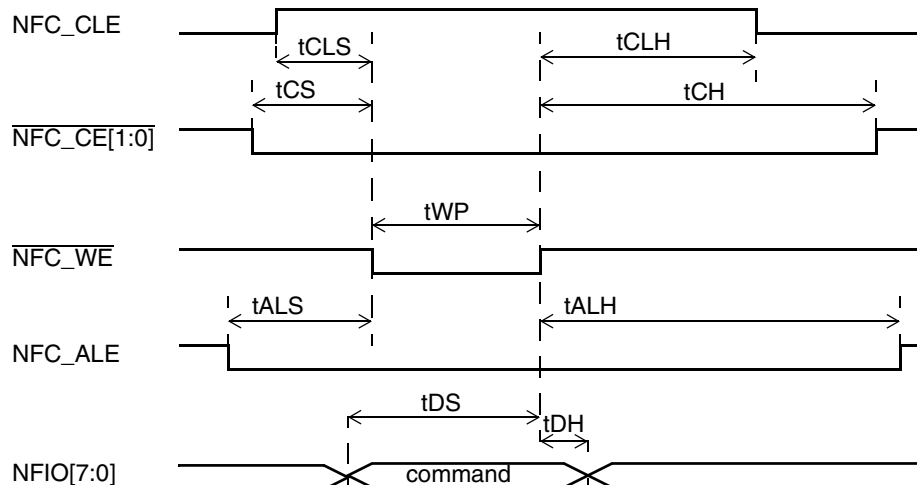


Figure 18. Timing Diagram – MUXed Synchronous Write Burst

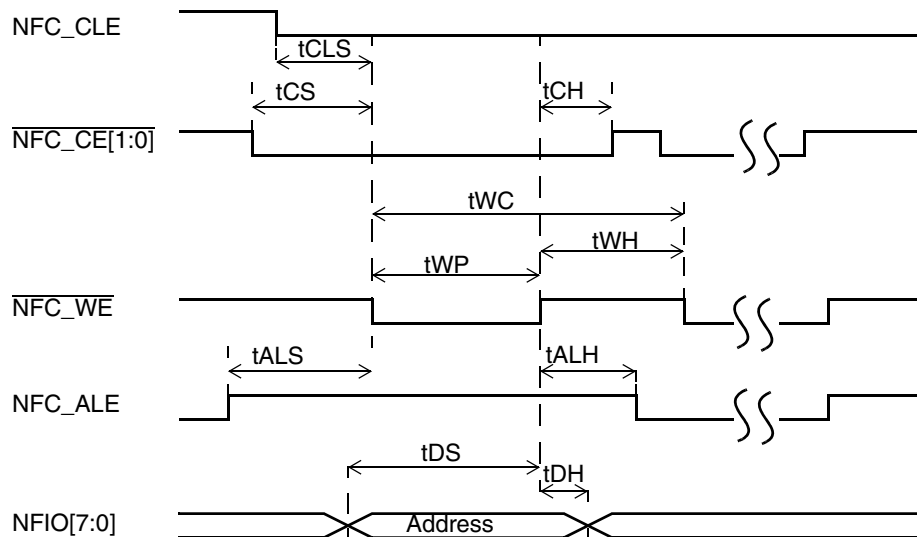


### 2.3.8 NFC

The NAND flash controller (NFC) implements the interface to standard NAND Flash memory devices. This section describes the timing parameters of the NFC.

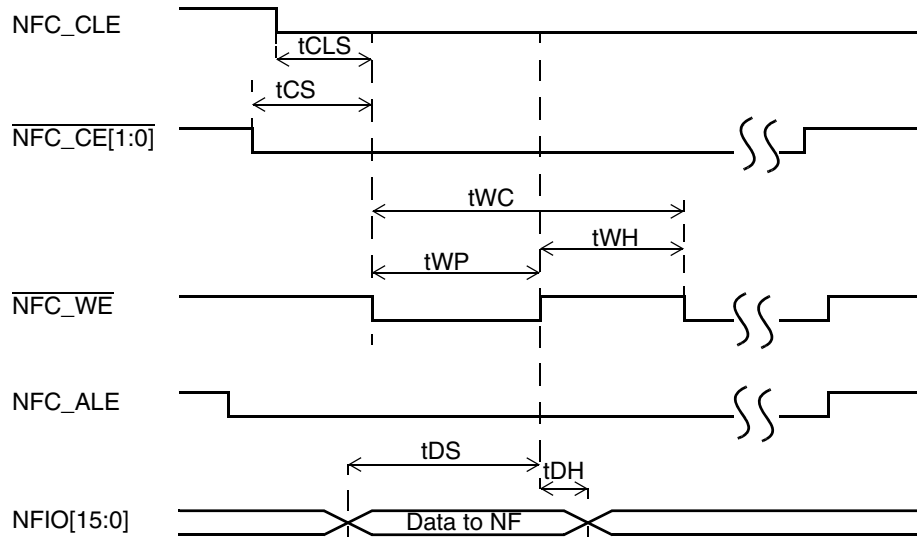


**Figure 19. Command Latch Cycle Timing**

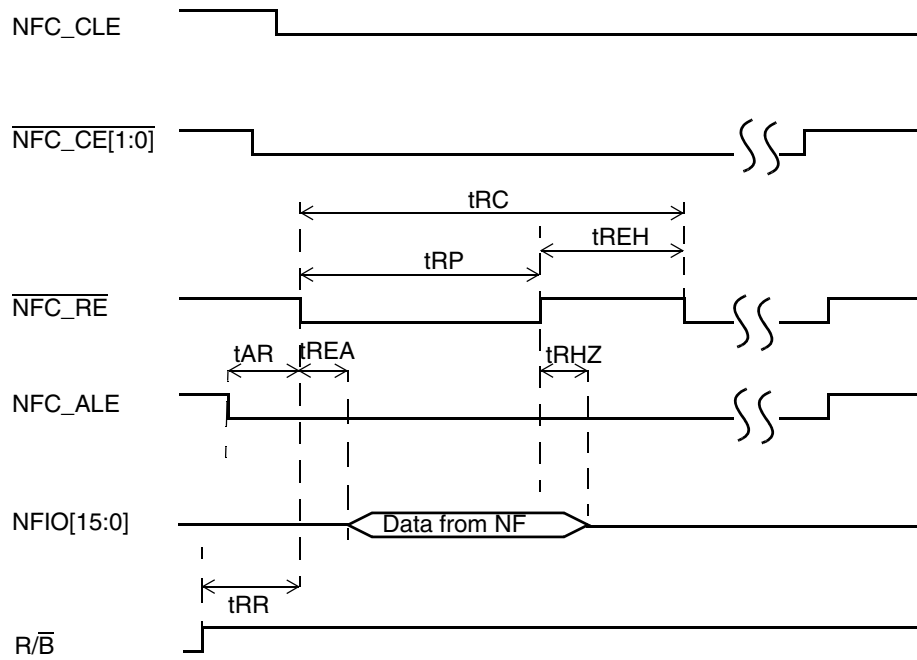


**Figure 20. Address Latch Cycle Timing**

## Electrical and Thermal Characteristics



**Figure 21. Write Data Latch Timing**



**Figure 22. Read Data Latch Timing**

**Table 23. NFC Target Timing Characteristics**

Timing parameter	Description	Min. value	Max. value	Unit	SpecID
tCLS	NFC_CLE setup Time	T+1	—	ns	A8.1
tCLH	NFC_CLE Hold Time	T-1	—	ns	A8.2
tCS	$\overline{\text{NFC\_CE}}[1:0]$ Setup Time	2T-1	—	ns	A8.3
tCH	$\overline{\text{NFC\_CE}}[1:0]$ Hold Time	T	—	ns	A8.4

Table 23. NFC Target Timing Characteristics (continued)

Timing parameter	Description	Min. value	Max. value	Unit	SpecID
tWP	NFC_WP Pulse Width	T-1	—	ns	A8.5
tALS	NFC_ALE Setup Time	T-1	—	ns	A8.6
tALH	NFC_ALE Hold Time	T-1	—	ns	A8.7
tDS	Data Setup Time	T-2	—	ns	A8.8
tDH	Data Hold Time	T-1	—	ns	A8.9
tWC	Write Cycle Time	2T	—	ns	A8.10
tWH	NFC_WE Hold Time	T-1	—	ns	A8.11
tRR	Ready to $\overline{\text{NFC\_RE}}$ Low	5T+2	—	ns	A8.12
tRP	$\overline{\text{NFC\_RE}}$ Pulse Width	1.5T-1	—	ns	A8.13
tRC	READ Cycle Time	2T	—	ns	A8.14
tREH	$\overline{\text{NFC\_RE}}$ High Hold Time	0.5T	—	ns	A8.15

T is the flash clock cycle.

T= 45 ns, frequency = 22 MHz (boot configuration, IP bus = 66 MHz)

T= 36 ns, frequency = 27 MHz (maximum configurable frequency, IP bus = 83 MHz)

### 2.3.9 PATA

The MPC5121e ATA Controller (PATA) is completely software programmable. It can be programmed to operate with ATA protocols using their respective timing, as described in the ANSI ATA-4 specification. The ATA interface is completely asynchronous in nature. Signal relationships are based on specific fixed timing in terms of timing units (nanoseconds).

ATA data setup and hold times, with respect to Read/Write strobes, are software programmable inside the ATA Controller. Data setup and hold times are implemented using counters. The counters count the number of ATA clock cycles needed to meet the ANSI ATA-4 timing specifications. For details, see the ANSI ATA-4 specification and how to program an ATA Controller and ATA drive for different ATA protocols and their respective timing. See the MPC5121e Reference Manual.

The MPC5121e ATA Host Controller design makes data available coincidentally with the active edge of the WRITE strobe in PIO and Multiword DMA modes.

- Write data is latched by the drive at the inactive edge of the WRITE strobe. This gives ample setup-time beyond that required by the ATA-4 specification.
- Data is held unchanged until the next active edge of the WRITE strobe. This gives ample hold-time beyond that required by the ATA-4 specification.

All ATA transfers are programmed in terms of system clock cycles (IP bus clocks) in the ATA Host Controller timing registers. This puts constraints on the ATA protocols and their respective timing modes in which the ATA Controller can communicate with the drive.

## Electrical and Thermal Characteristics

Faster ATA modes (i.e., UDMA 0, 1, 2) are supported when the system is running at a sufficient frequency to provide adequate data transfer rates. Adequate data transfer rates are a function of the following:

- The MPC5121e operating frequency (IP bus clock frequency)
- Internal MPC5121e bus latencies
- Other system load dependent variables

The ATA clock is the same frequency as the IP bus clock in MPC5121e. See the MPC5121e Reference Manual.

### NOTE

All output timing numbers are specified for nominal 50 pF loads.

## 2.3.9.1 PATA Timing Parameters

In the timing equations, some timing parameters are used. These parameters depend on the implementation of the ATA interface in silicon, the bus transceiver used, the cable delay and cable skew. The parameters shown in [Table 2-24](#) specify the ATA timing.

**Table 2-24. PATA Timing Parameters**

Name	Meaning	Controlled by	Value	SpecID
T	PATA Bus clock period	MPC5121e	15 ns	A9.1
ti_ds	Set-up time ATA_DATA to ATA_IORDY edge (UDMA-in only)	MPC5121e	2 ns	A9.2
ti_dh	Hold time ATA_IORDY edge to ATA_DATA (UDMA-in only)	MPC5121e	5 ns	A9.3
tco	Propagation delay bus clock L-to-H to: ATA_CS0, ATA_CS1, ATA_DA2, ATA_DA1, ATA_DA0, ATA_DIOR, ATA_DIOW, ATA_DMACK, ATA_DATA, ATA_BUFFER_EN	MPC5121e	2 ns	A9.4
tsu	Set-up time ATA_DATA to bus clock L-to-H	MPC5121e	2 ns	A9.5
tsui	Set-up time ATA_IORDY to bus clock H-to-L	MPC5121e	2 ns	A9.6
thi	Hold time ATA_IORDY to bus clock H to L	MPC5121e	2 ns	A9.7
tskew1	Max difference in propagation delay bus clock L-to-H to any of following signals: ATA_CS0, ATA_CS1, ATA_DA2, ATA_DA1, ATA_DA0, ATA_DIOR, ATA_DIOW, ATA_DMACK, ATA_DATA (WRITE), ATA_BUFFER_EN	MPC5121e	1.7 ns	A9.8
tskew2	Max difference in buffer propagation delay for any of following signals: ATA_CS0, ATA_CS1, ATA_DA2, ATA_DA1, ATA_DA0, ATA_DIOR, ATA_DIOW, ATA_DMACK, ATA_DATA (WRITE), ATA_BUFFER_EN	Transceiver		A9.9
tskew3	Max difference in buffer propagation delay for any of following signals: ATA_IORDY, ATA_DATA (read)	Transceiver		A9.10
tbuf	Max buffer propagation delay	Transceiver		A9.11
tcable1	Cable propagation delay for ata_data	Cable		A9.12
tcable2	Cable propagation delay for control signals: ATA_DIOR, ATA_DIOW, ATA_IORDY, ATA_DMACK	Cable		A9.13
tskew4	Max difference in cable propagation delay between: ATA_IORDY and ATA_DATA (read)	Cable		A9.14

Table 2-24. PATA Timing Parameters (continued)

Name	Meaning	Controlled by	Value	SpecID
tskew5	Max difference in cable propagation delay between: ATA_DIOR, ATA_DIOW, ATA_DMACK and ATA_CS0, ATA_CS1, ATA_DA2, ATA_DA1, ATA_DA0, ATA_DATA (write)	Cable		A9.15
tskew6	Max difference in cable propagation delay without accounting for ground bounce	Cable		A9.16

### 2.3.9.2 PIO Mode Timing

A timing diagram for the PIO read mode is given in Figure 23.

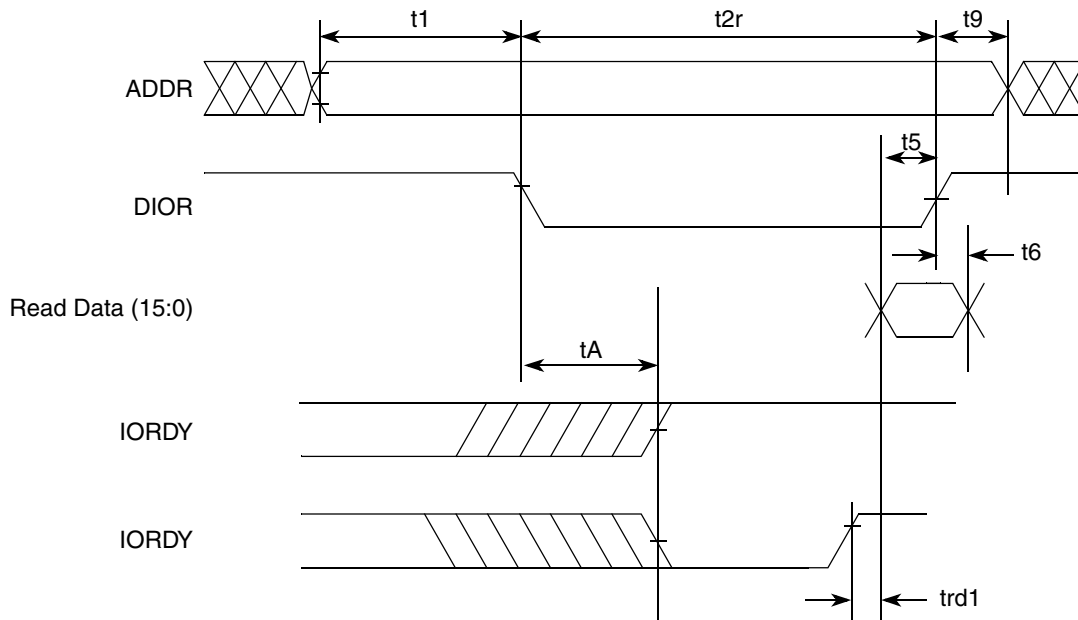


Figure 23. PIO Read Mode Timing

To fulfill read mode timing, the different timing parameters given in Table 2-25 must be observed.

Table 2-25. Timing Parameters PIO Read

ATA Parameter	PIO Read Mode Timing Parameter	Value	How to meet	SpecID
t1	t1	$t1(\min) = \text{time\_1} * T - (\text{tskew1} + \text{tskew2} + \text{tskew5})$	calculate and programming time_1, see Reference Manual	A9.20
t2	t2r	$t2(\min) = \text{time\_2r} * T - (\text{tskew1} + \text{tskew2} + \text{tskew5})$	calculate and programming time_2r, see Reference Manual	A9.21
t9	t9	$t9(\min) = \text{time\_9} * T - (\text{tskew1} + \text{tskew2} + \text{tskew6})$	calculate and programming time_9, see Reference Manual	A9.22
t5	t5	$t5(\min) = t_{co} + t_{su} + t_{buf} + t_{buf} + t_{cable1} + t_{cable2}$	If not met, increase time_2r	A9.23
t6	t6	0	—	A9.24

Table 2-25. Timing Parameters PIO Read (continued)

ATA Parameter	PIO Read Mode Timing Parameter	Value	How to meet	SpecID
tA	tA	$tA(\min) = (1.5 + \text{time\_ax}) * T - (\text{tco} + \text{tsui} + \text{tcable2} + \text{tcable2} + 2 * \text{tbuf})$	calculate and programming time_ax, see Reference Manual	A9.25
trd	trd1	$\text{trd1}(\max) = (-\text{trd}) + (\text{tskew3} + \text{tskew4})$ $\text{trd1}(\min) = (\text{time\_pio\_rdx} - 0.5) * T - (\text{tsu} + \text{thi})$ $(\text{time\_pio\_rdx} - 0.5) * T > \text{tsu} + \text{thi} + \text{tskew3} + \text{tskew4}$	calculate and programming time_pio_rdx, see Reference Manual	A9.26
t0	—	$t0(\min) = (\text{time\_1} + \text{time\_2} + \text{time\_9}) * T$	time_1, time_2r, time_9	A9.27

In PIO write mode, timing waveforms are somewhat different as shown in Figure 24.

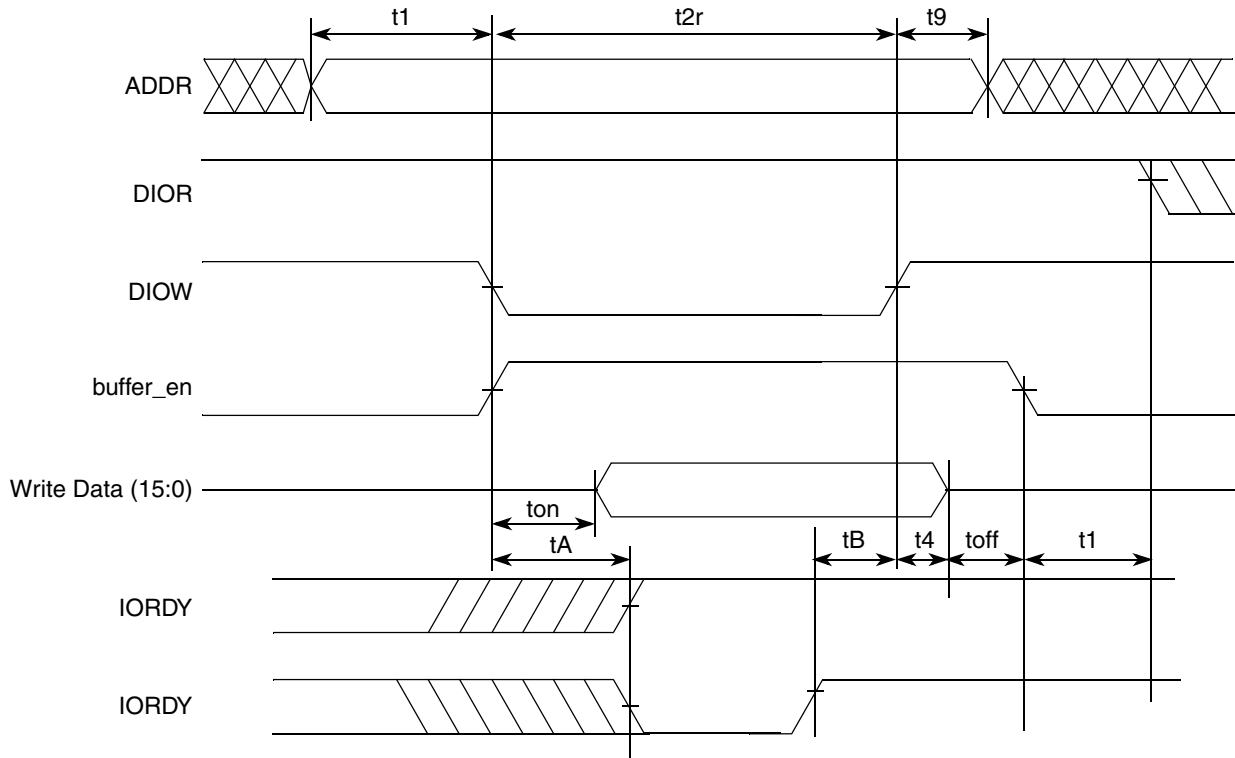


Figure 24. PIO Write Mode Timing

To fulfill this timing, several parameters need to be observed as shown in Table 2-26.

Table 2-26. Timing Parameters PIO Write

ATA Parameter	PIO Write Mode Timing Parameter	Value	How to meet	SpecID
t1	t1	$t1(\text{min}) = \text{time\_1} * T - (\text{tskew1} + \text{tskew2} + \text{tskew5})$	time_1, see Reference Manual	A9.30
t2	t2r	$t2(\text{min}) = \text{time\_2w} * T - (\text{tskew1} + \text{tskew2} + \text{tskew5})$	calculate and programming time_2w, see Reference Manual	A9.31
t9	t9	$t9(\text{min}) = \text{time\_9} * T - (\text{tskew1} + \text{tskew2} + \text{tskew6})$	time_9, see Reference Manual	A9.32
t3	—	$t3(\text{min}) = (\text{time\_2w} - \text{time\_on}) * T - (\text{tskew1} + \text{tskew2} + \text{tskew5})$	If not met, increase time_2w	A9.33
t4	t4	$t4(\text{min}) = \text{time\_4} * T - \text{tskew1}$	calculate and programming time_4, see Reference Manual	A9.34
tA	tA	$tA = (1.5 + \text{time\_ax}) * T - (\text{tco} + \text{tsui} + \text{tcable2} + \text{tcable2} + 2 * \text{tbuf})$	calculate and programming time_ax, see Reference Manual	A9.35
t0	—	$t0(\text{min}) = (\text{time\_1} + \text{time\_2} + \text{time\_9}) * T$	time_1, time_2r, time_9	A9.36
—	—	Avoid bus contention when switching buffer on by making ton long enough	—	A9.37
—	—	Avoid bus contention when switching buffer off by making toff long enough	—	A9.38

### 2.3.9.3 Timing in Multiword DMA Mode

Timing in multiword DMA mode is given in Figure 25 and Figure 26.

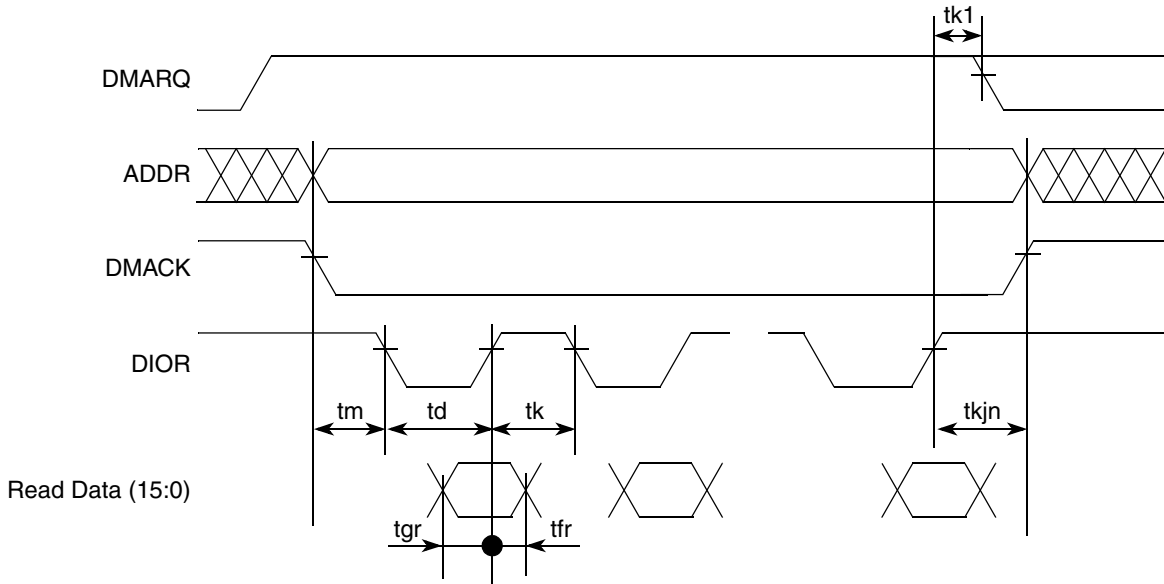


Figure 25. MDMA Read Timing

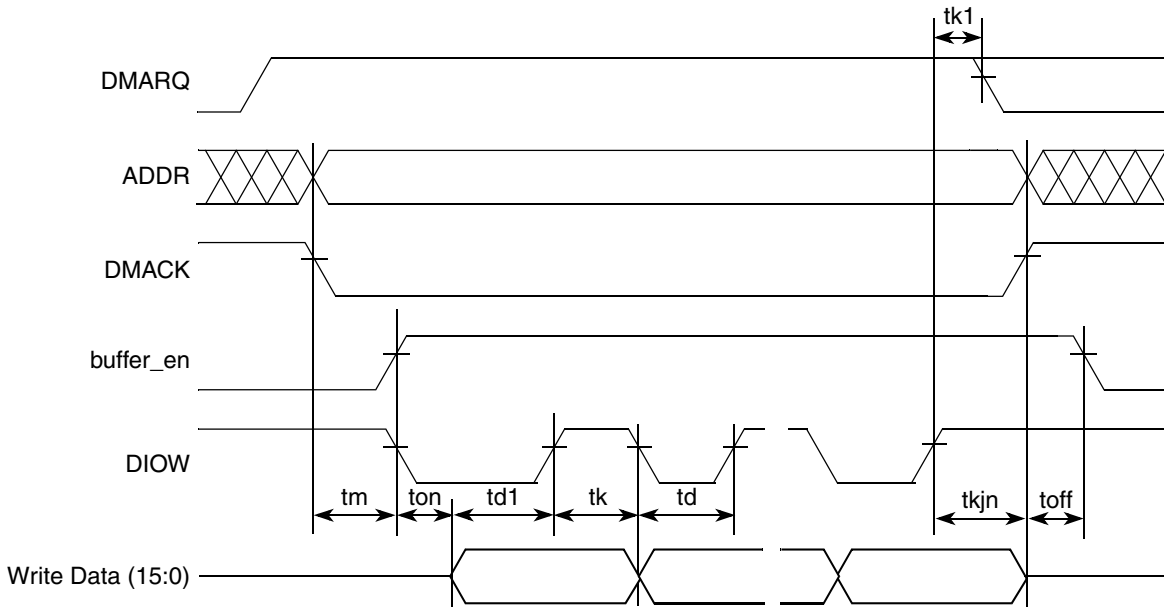


Figure 26. MDMA Write Timing

To meet this timing, a number of timing parameters must be controlled as shown in Table 2-27.



Table 2-27. Timing Parameters MDMA Read and Write

ATA Parameter	MDMA Read/Write Timing Parameter	Value	How to meet	SpecID
tm, ti	tm	$tm(\min) = ti(\min) = \text{time\_m} * T - (\text{tskew1} + \text{tskew2} + \text{tskew5})$	calculate and programming time_m, see Reference Manual	A9.40
td	td, td1	$td1(\min) = td(\min) = \text{time\_d} * T - (\text{tskew1} + \text{tskew2} + \text{tskew6})$	calculate and programming time_d, see Reference Manual	A9.41
tk	tk	$tk(\min) = \text{time\_k} * T - (\text{tskew1} + \text{tskew2} + \text{tskew6})$	calculate and programming time_k, see Reference Manual	A9.42
t0	—	$t0(\min) = (\text{time\_d} + \text{time\_k}) * T$	time_d, time_k	A9.43
tg(read)	tgr	$tgr(\min\text{-read}) = \text{tco} + \text{tsu} + \text{tbuf} + \text{tbuf} + \text{tcable1} + \text{tcable2}$ $tgr(\min\text{-drive}) = \text{td} - \text{te}(\text{drive})$	time_d, see Reference Manual	A9.44
tf(read)	tfr	$tfr(\min\text{-drive}) = 0$	—	A9.45
tg(write)	—	$tg(\min\text{-write}) = \text{time\_d} * T - (\text{tskew1} + \text{tskew2} + \text{tskew5})$	time_d	A9.46
tf(write)	—	$tf(\min\text{-write}) = \text{time\_k} * T - (\text{tskew1} + \text{tskew2} + \text{tskew6})$	time_k	A9.47
tL	—	$tL(\max) = (\text{time\_d} + \text{time\_k} * 2) * T - (\text{tsu} + \text{tco} + 2 * \text{tbuf} + 2 * \text{tcable2})$	time_d, time_k	A9.48
tn, tj	tkjn	$tn = tj = tkjn = (\max(\text{time\_k}, \text{time\_jn}) * T - (\text{tskew1} + \text{tskew2} + \text{tskew6}))$	calculate and programming time_jn, see Reference Manual	A9.49
—	ton toff	$\text{ton} = \text{time\_on} * T - \text{tskew1}$ $\text{toff} = \text{time\_off} * T - \text{tskew1}$	—	A9.50

### 2.3.9.4 UDMA In Timing Diagrams

UDMA mode timing is more complicated than PIO mode or MDMA mode. In this section, timing diagrams for UDMA in are given:

- [Figure 27](#) gives timing for UDMA in transfer start
- [Figure 28](#) gives timing for host terminating UDMA in transfer
- [Figure 29](#) gives timing for device terminating UDMA in transfer.

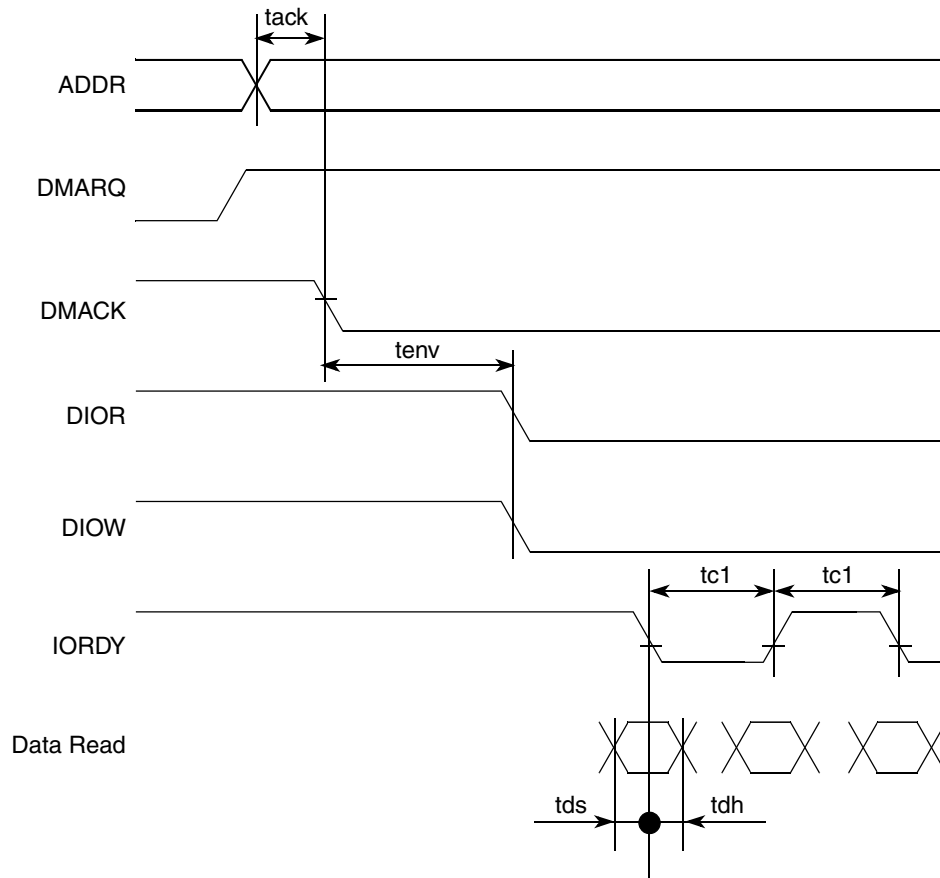


Figure 27. UDMA In Transfer Start Timing Diagram

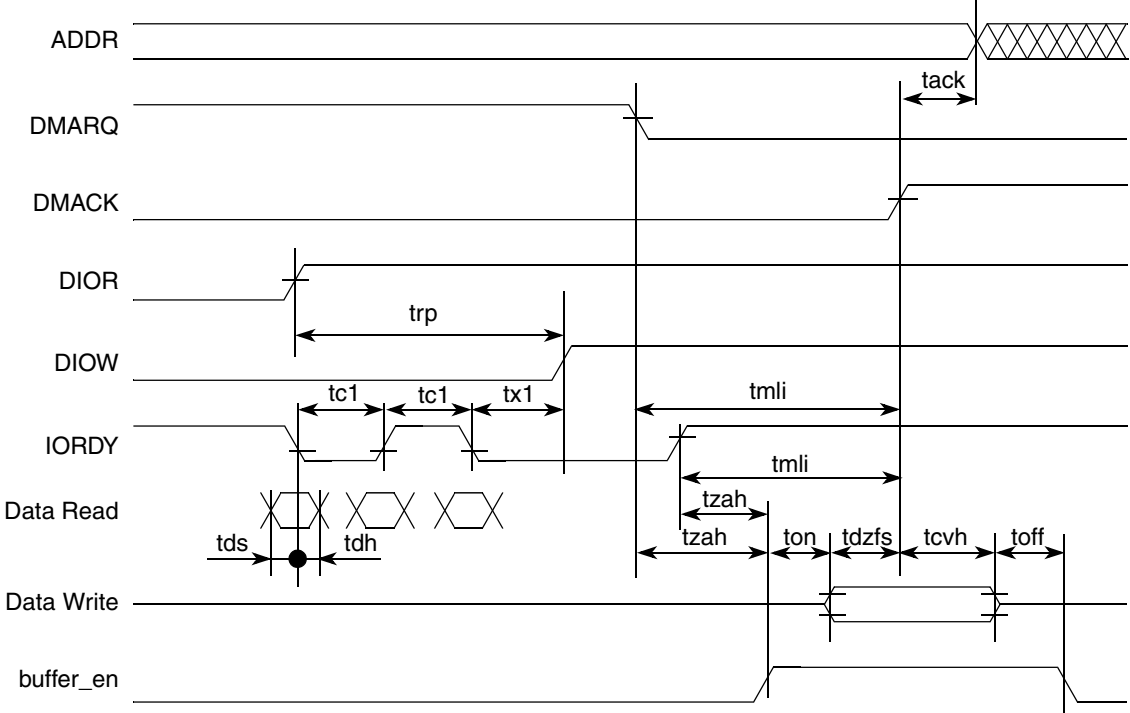


Figure 28. UDMA In Host Terminates Transfer

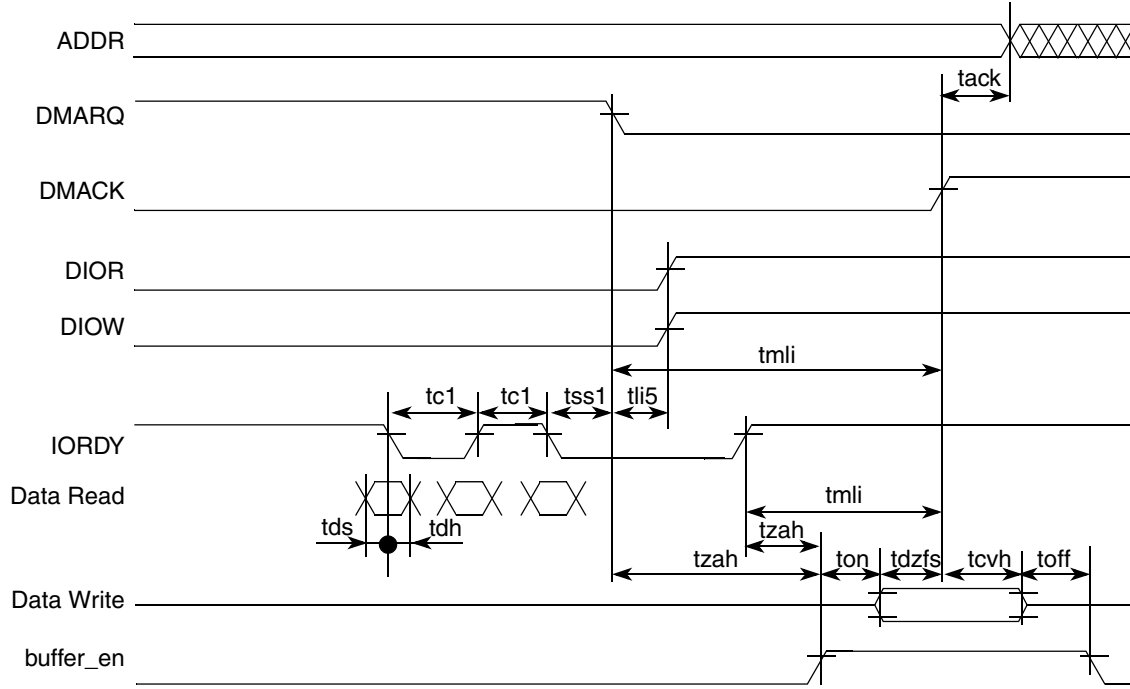


Figure 29. UDMA In Device Terminates Transfer

Timing parameters are explained in [Table 28](#).

**Table 28. Timing Parameters UDMA in Burst**

ATA Parameter	UDMA In Transfer Start Timing Diagram Parameter	Value	How to Meet	SpecID
	UDMA In Host/Device Terminates Transfer			
tack	tack	$tack(min) = (time\_ack * T) - (tskew1 + tskew2)$	calculate and programming time_ack, see Reference Manual	A9.51
tenv	tenv	$tenv(min) = (time\_env * T) - (tskew1 + tskew2)$ $tenv(max) = (time\_env * T) + (tskew1 + tskew2)$	calculate and programming time_env, see Reference Manual	A9.52
tds	tds1	$tds - (tskew3) - ti\_ds > 0$	tskew3, ti_ds, ti_dh should be low enough	A9.53
tdh	tdh1	$tdh - (tskew3) - ti\_dh > 0$		A9.54
tcyc	tc1	$(tcyc - tskew) > T$	Bus clock period T big enough	A9.55
trp	trp	$trp(min) = time\_rp * T - (tskew1 + tskew2 + tskew6)$	calculate and programming time_rp, see Reference Manual	A9.56
—	tx1 <sup>1</sup>	$(time\_rp * T) - (tco + tsu + 3T + 2 * tbuf + 2 * tcable2) > trfs$ (drive)	calculate and programming time_rp, see Reference Manual	A9.57
tml1	tml1	$tml1(min) = (time\_ml1x + 0.4) * T$	calculate and programming time_ml1x, see Reference Manual	A9.58
tzah	tzah	$tzah(min) = (time\_zah + 0.4) * T$	calculate and programming time_zah, see Reference Manual	A9.59
tdzfs	tdzfs	$tdzfs = (time\_dzfs * T) - (tskew1 + tskew2)$	calculate and programming time_dzfs, see Reference Manual	A9.60
tcvh	tcvh	$tcvh = (time\_cvh * T) - (tskew1 + tskew2)$	calculate and programming time_cvh, see Reference Manual	A9.61
—	ton toff <sup>2</sup>	$ton = time\_on * T - tskew1$ $toff = time\_off * T - tskew1$	—	A9.62

<sup>1</sup> There is a special timing requirement in the ATA host that requires the internal DIOW to go only high three clocks after the last active edge on the DSTROBE signal. The equation given on this line tries to capture this constraint.

<sup>2</sup> Make TON and TOFF big enough to avoid bus contention.

### 2.3.9.5 UDMA Out Timing Diagrams

UDMA mode timing is more complicated than PIO mode or MDMA mode. In this section, timing diagrams for UDMA out are given:

- [Figure 30](#) gives timing for UDMA out transfer start
- [Figure 31](#) gives timing for host terminating UDMA out transfer
- [Figure 32](#) gives timing for device terminating UDMA out transfer.

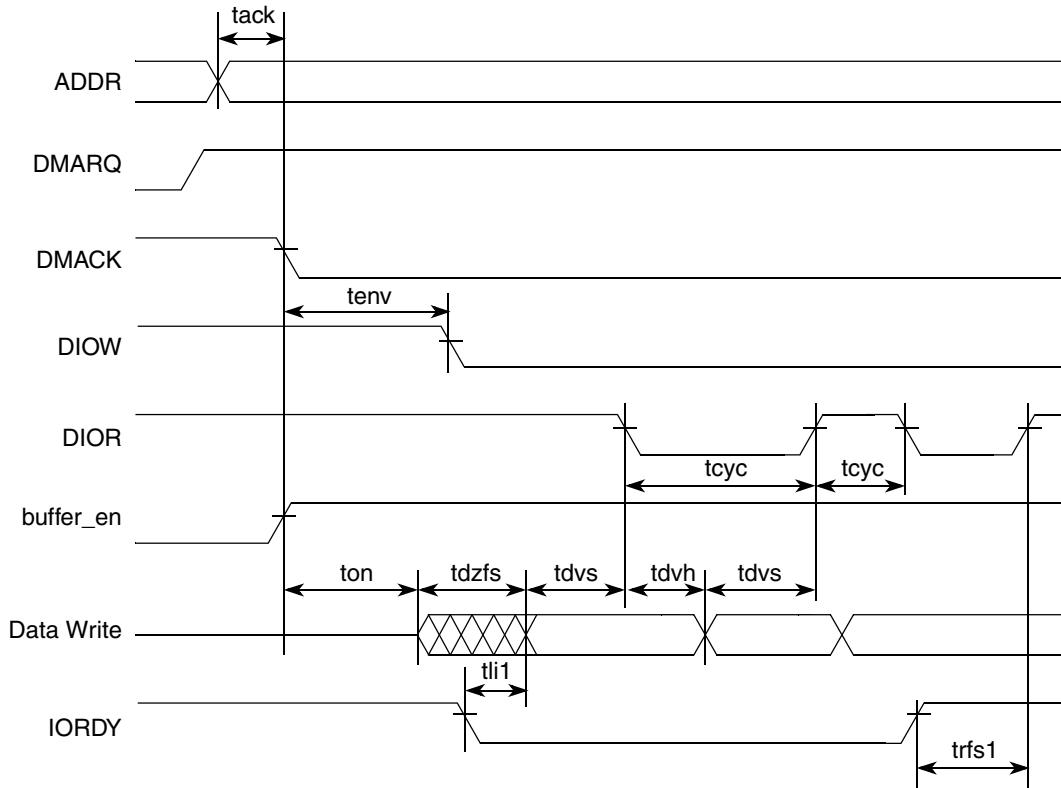


Figure 30. UDMA Out Transfer Start Timing Diagram

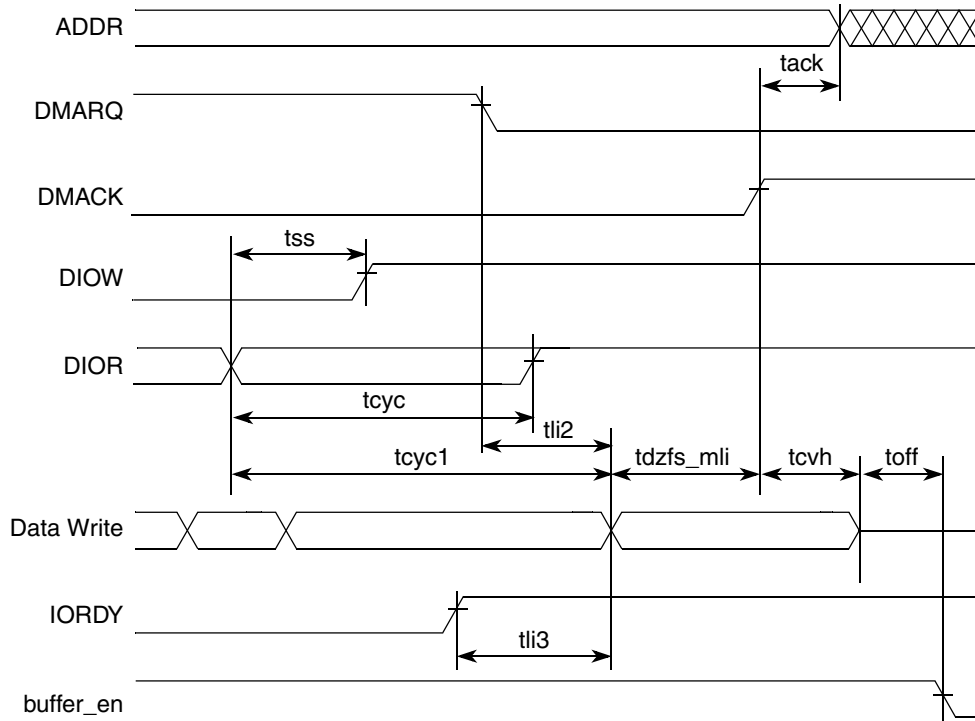


Figure 31. UDMA Out Host Terminates Transfer

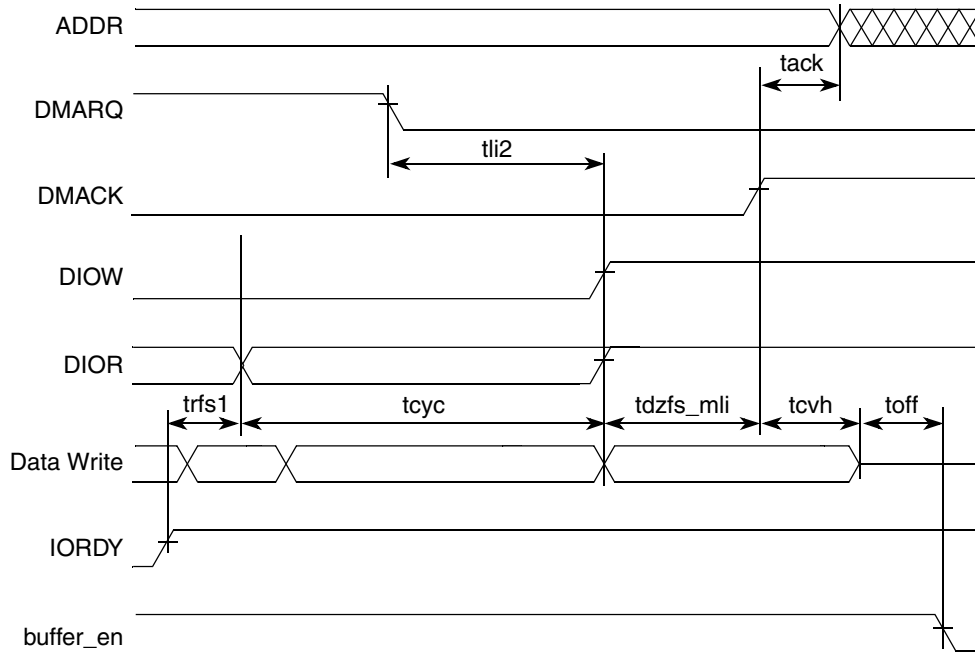


Figure 32. UDMA Out Device Terminates Transfer

Timing parameters are explained in [Table 29](#).

Table 29. Timing Parameters UDMA Out Burst

ATA Parameter	UDMA Out Transfer Start Timing Diagram Parameter	Value	How to meet	SpecID
	Out Host/Device Terminates Transfer			
tack	tack	$tack(min) = (time\_ack * T) - (tskew1 + tskew2)$	calculate and programming time_ack, see Reference Manual	A9.63
tenv	tenv	$tenv(min) = (time\_env * T) - (tskew1 + tskew2)$ $tenv(max) = (time\_env * T) + (tskew1 + tskew2)$	calculate and programming time_env, see Reference Manual	A9.64
tdvs	tdvs	$tdvs = (time\_dvs * T) - (tskew1 + tskew2)$	calculate and programming time_dvs, see Reference Manual	A9.65
tdvh	tdvh	$tdvs = (time\_dvh * T) - (tskew1 + tskew2)$	calculate and programming time_dvh, see Reference Manual	A9.66
tcyc	tcyc	$tcyc = time\_cyc * T - (tskew1 + tskew2)$	calculate and programming time_cyc, see Reference Manual	A9.67
t2cyc	—	$t2cyc = time\_cyc * 2 * T$	calculate and programming time_cyc, see Reference Manual	A9.68
trfs1	trfs	$trfs = 1.6 * T + tsui + tco + tbuf + tbuf$	—	A9.69
—	tdzfs	$tdzfs = time\_dzfs * T - (tskew1)$	calculate and programming time_dzfs, see Reference Manual	A9.70
tss	tss	$tss = time\_ss * T - (tskew1 + tskew2)$	calculate and programming time_ss, see Reference Manual	A9.71
tmli	tdzfs_mli	$tdzfs\_mli = \max(time\_dzfs, time\_mli) * T - (tskew1 + tskew2)$	—	A9.72
tli	tli1	$tli1 > 0$	—	A9.73
tli	tli2	$tli2 > 0$	—	A9.74
tli	tli3	$tli3 > 0$	—	A9.75

Table 29. Timing Parameters UDMA Out Burst (continued)

ATA Parameter	UDMA Out Transfer Start Timing Diagram Parameter	Value	How to meet	SpecID
	Out Host/Device Terminates Transfer			
tcvh	tcvh	$t_{cvh} = (\text{time\_cvh} * T) - (\text{tskew1} + \text{tskew2})$	calculate and programming time_cvh, see Reference Manual	A9.76
—	ton toff	ton = time_on * T – tskew1 toff = time_off * T – tskew1	—	A9.77

### 2.3.10 SATA PHY

1.5 Gbps SATA PHY Layer

See “Serial ATA: High Speed Serialized AT Attachment” Revision 1.0a, 7-January-2003.

### 2.3.11 FEC

AC Test Timing Conditions:

- Output Loading  
All Outputs: 25 pF

Table 30. MII Rx Signal Timing

Sym	Description	Min	Max	Unit	SpecID
t <sub>1</sub>	RXD[3:0], RX_DV, RX_ER to RX_CLK setup	5	—	ns	A11.1
t <sub>2</sub>	RX_CLK to RXD[3:0], RX_DV, RX_ER hold	5	—	ns	A11.2
t <sub>3</sub>	RX_CLK pulse width high	35%	65%	RX_CLK Period <sup>1</sup>	A11.3
t <sub>4</sub>	RX_CLK pulse width low	35%	65%	RX_CLK Period <sup>1</sup>	A11.4

<sup>1</sup> RX\_CLK shall have a frequency of 25% of data rate of the received signal. See the IEEE 802.3 Specification.

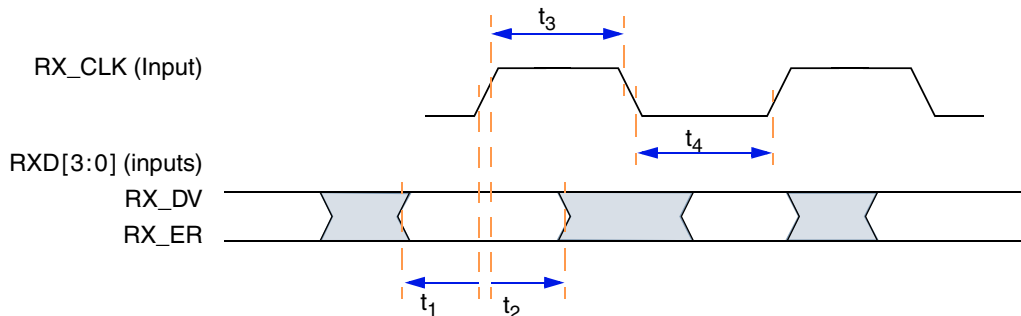


Figure 33. Ethernet Timing Diagram – MII Rx Signal



Table 31. MII Tx Signal Timing

Sym	Description	Min	Max	Unit	SpecID
t <sub>5</sub>	TX_CLK rising edge to TXD[3:0], TX_EN, TX_ER invalid	5	—	ns	A11.5
t <sub>6</sub>	TX_CLK rising edge to TXD[3:0], TX_EN, TX_ER valid	—	25	ns	A11.6
t <sub>7</sub>	TX_CLK pulse width high	35%	65%	TX_CLK Period <sup>1</sup>	A11.7
t <sub>8</sub>	TX_CLK pulse width low	35%	65%	TX_CLK Period <sup>1</sup>	A11.8

<sup>1</sup> The TX\_CLK frequency shall be 25% of the nominal transmit frequency, e.g., a PHY operating at 100 Mb/s must provide a TX\_CLK frequency of 25 MHz and a PHY operating at 10 Mb/s must provide a TX\_CLK frequency of 2.5 MHz. See the IEEE 802.3 Specification.

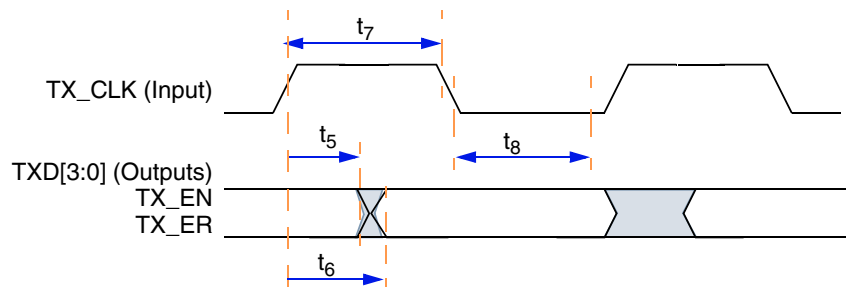


Figure 34. Ethernet Timing Diagram – MII Tx Signal

Table 32. MII Async Signal Timing

Sym	Description	Min	Max	Unit	SpecID
t <sub>9</sub>	CRS, COL minimum pulse width	1.5	—	TX_CLK Period	A11.9

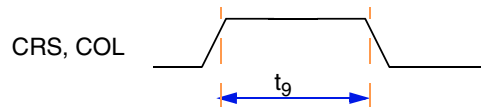


Figure 35. Ethernet Timing Diagram – MII Async

Table 33. MII Serial Management Channel Signal Timing

Sym	Description	Min	Max	Unit	SpecID
t <sub>10</sub>	MDC falling edge to MDIO output delay	0	25	ns	A11.10
t <sub>11</sub>	MDIO (input) to MDC rising edge setup	10	—	ns	A11.11
t <sub>12</sub>	MDIO (input) to MDC rising edge hold	0	—	ns	A11.12
t <sub>13</sub>	MDC pulse width high <sup>1</sup>	160	—	ns	A11.13
t <sub>14</sub>	MDC pulse width low <sup>1</sup>	160	—	ns	A11.14
t <sub>15</sub>	MDC period <sup>2</sup>	400	—	ns	A11.15

<sup>1</sup> MDC is generated by MPC5121e with a duty cycle of 50% except when MII\_SPEED in the FEC MII\_SPEED control register is changed during operation. See the MPC5121e Reference Manual.

<sup>2</sup> The MDC period must be set to a value of less than or equal to 2.5 MHz (to be compliant with the IEEE MII characteristic) by programming the FEC MII\_SPEED control register. See the MPC5121e Reference Manual.

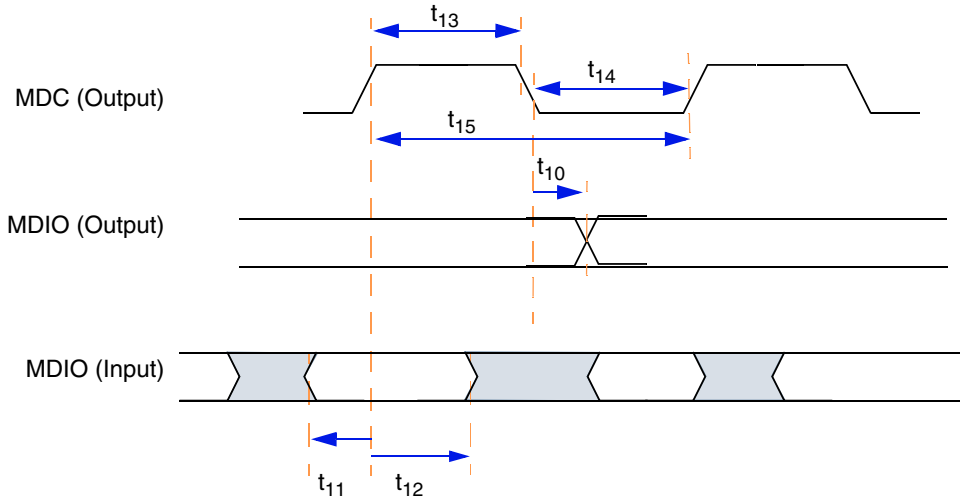


Figure 36. Ethernet Timing Diagram – MII Serial Management

### 2.3.12 USB ULPI

This section specifies the USB ULPI timing.

For more information refer to UTMI+ Low Pin Interface (ULPI) Specification, Revision 1.1, October 20, 2004.

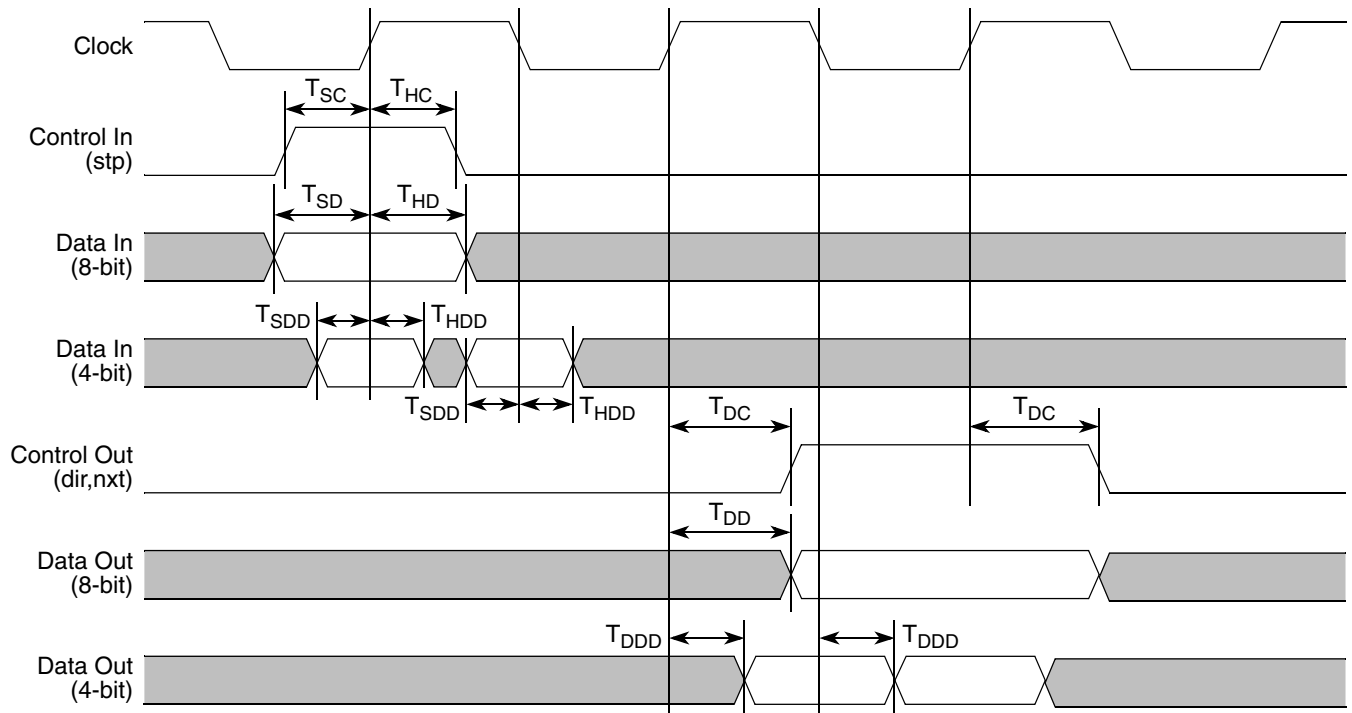


Figure 37. ULPI Timing Diagram

**Table 34. Timing Specifications – USB Output Line**

Sym	Description	Min	Max	Units	SpecID
T <sub>SC</sub> , T <sub>SD</sub>	Setup time (control in, 8-bit data in)	—	6.0	ns	A12.1
T <sub>HC</sub> , T <sub>HD</sub>	Hold time (control in, 8-bit data in)	0.0	-	ns	A12.2
T <sub>DC</sub> , T <sub>DD</sub>	Output delay (control out, 8-bit data out)	—	9.0	ns	A12.3

**NOTE**

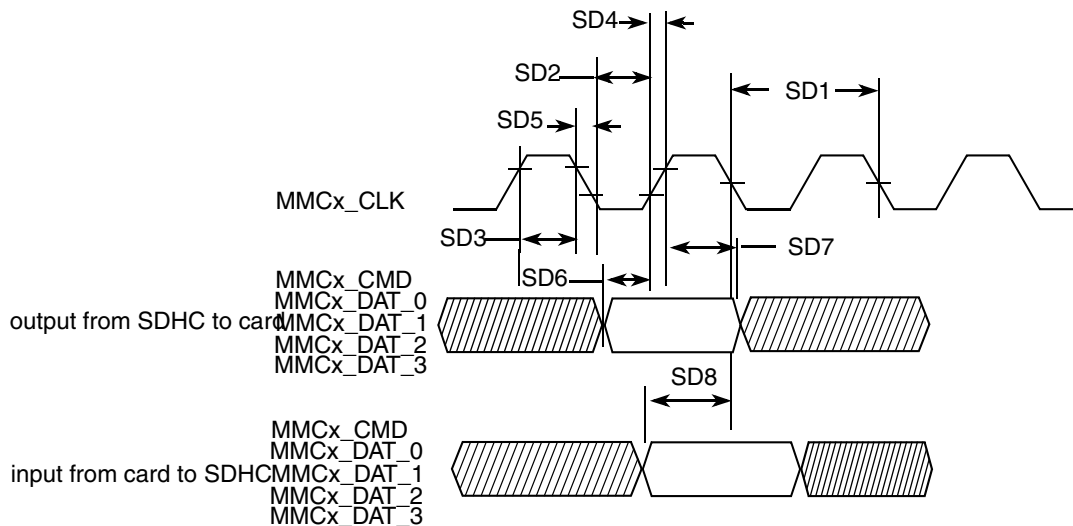
Output timing is specified at a nominal 50 pF load.

### 2.3.13 On-Chip USB PHY

The USB PHY is an USB2.0 compatible PHY integrated on-chip. See Chapter 7 in the USB Specification Rev. 2.0 at [www.usb.org](http://www.usb.org).

### 2.3.14 SDHC

Figure 38 depicts the timings of the SDHC.



**Figure 38. SDHC Timing Diagram**

Table 35 lists the timing parameters.

Table 35. MMC/SD Interface Timing Parameters

ID	Parameter	Symbols	Min	Max	Unit	SpecID
Card Input Clock						
SD1	Clock Frequency (Low Speed)	$f_{PP}^1$	0	400	kHz	A14.1
	Clock Frequency (SD/SDIO Full Speed/High Speed)	$f_{PP}^2$	0	25/50	MHz	A14.2
	Clock Frequency (MMC Full Speed/High Speed)	$f_{PP}^3$	0	20/52	MHz	A14.3
	Clock Frequency (Identification Mode)	$f_{OD}^4$	100	400	kHz	A14.4
SD2	Clock Low Time (Full Speed/High Speed)	$t_{WL}$	10/7		ns	A14.5
SD3	Clock High Time (Full Speed/High Speed)	$t_{WH}$	10/7		ns	A14.6
SD4	Clock Rise Time (Full Speed/High Speed)	$t_{TLH}$		10/3	ns	A14.7
SD5	Clock Fall Time (Full Speed/High Speed)	$t_{THL}$		10/3	ns	A14.8
SDHC Output / Card Inputs CMD, DAT (Reference to CLK)						
SD6	SDHC Output for Card Input Setup	$t_{OSU}$	15		ns	A14.9
SD7	SDHC Output for Card Input Hold	$t_{OH}$	15		ns	A14.10
SDHC Input / Card Outputs CMD, DAT (Reference to CLK)						
SD8	SDHC Input Setup Time	$t_{ISU}$	8		ns	A14.11

<sup>1</sup> In low speed mode, card clock must be lower than 400 kHz, voltage ranges from 2.7 to 3.6 V.

<sup>2</sup> In normal data transfer mode for SD/SDIO card, clock frequency can be any value between 0 ~ 25 MHz.

<sup>3</sup> In normal data transfer mode for MMC card, clock frequency can be any value between 0 ~ 20 MHz.

<sup>4</sup> In card identification mode, card clock must be 100 kHz ~ 400 kHz, voltage ranges from 2.7 to 3.6 V.

## 2.3.15 DIU

The DIU is a display controller designed to manage the TFT LCD display.

### 2.3.15.1 Interface to TFT LCD Panels, Functional Description

Figure 39 depicts the LCD interface timing for a generic active matrix color TFT panel. In this figure signals are shown with positive polarity. The sequence of events for active matrix interface timing is:

- DIU\_CLK latches data into the panel on its positive edge (when positive polarity is selected). In active mode, DIU\_CLK runs continuously. This signal frequency could be from 5 to 66 MHz depending on the panel type.
- DIU\_HSYNC causes the panel to start a new line. It always encompasses at least one PCLK pulse.
- DIU\_VSYNC causes the panel to start a new frame. It always encompasses at least one HSYNC pulse.
- DIU\_DE acts like an output enable signal to the LCD panel. This output enables the data to be shifted onto the display. When disabled, the data is invalid and the trace is off.

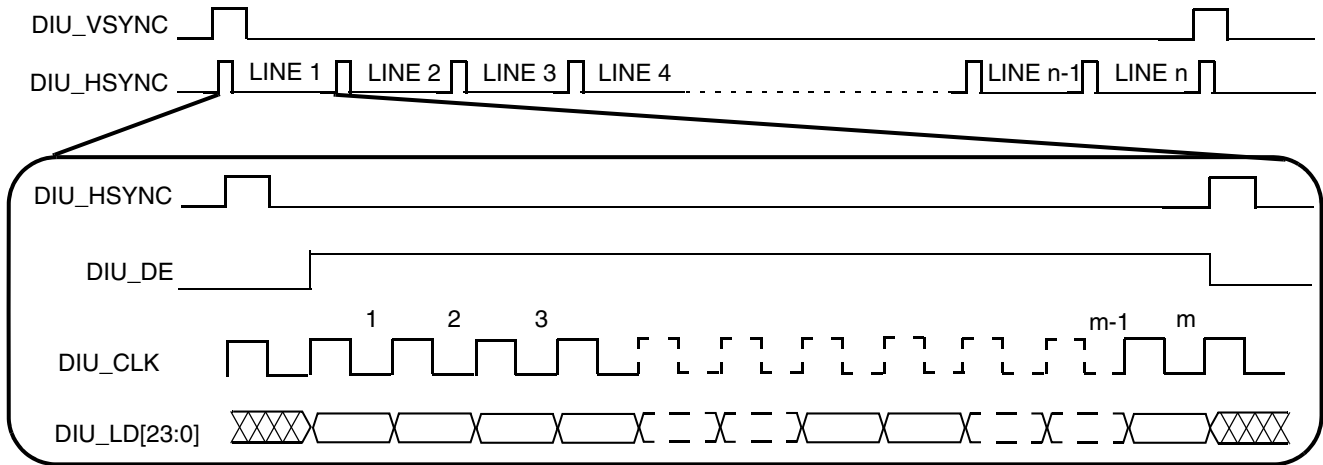


Figure 39. Interface Timing Diagram for TFT LCD Panels

### 2.3.15.2 Interface to TFT LCD Panels, Electrical Characteristics

Figure 40 depicts the horizontal timing (timing of one line), including the horizontal sync pulse and the data. All parameters shown in the diagram are programmable. This timing diagram corresponds to positive polarity of the DIU\_CLK signal (meaning the data and sync. signals change at the rising edge of it) and active-high polarity of the DIU\_HSYNC, DIU\_VSYNC and DIU\_DE signal. You can select the polarity of the DIU\_HSYNC and DIU\_VSYNC signal via the SYN\_POL register, whether active-high or active-low, the default is active-high. The DIU\_DE signal is always active-high. And, pixel clock inversion and a flexible programmable pixel clock delay is also supported, programed via the DIU Clock Config Register (DCCR) in the system clock module.

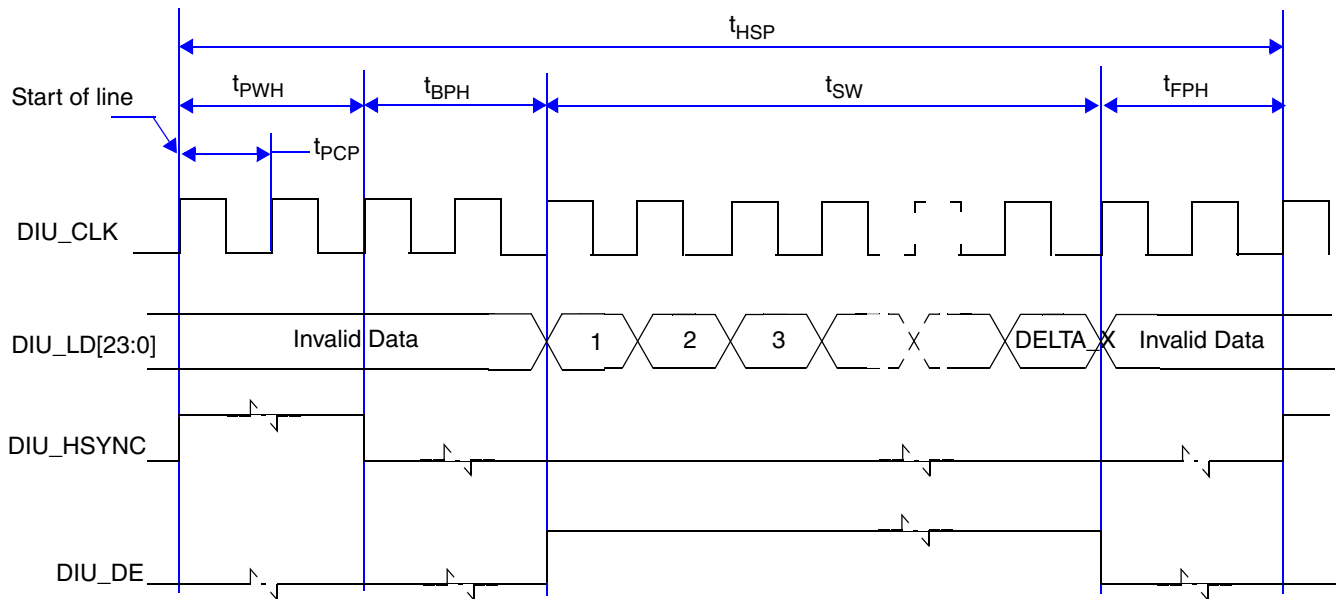


Figure 40. TFT LCD Interface Timing Diagram – Horizontal Sync Pulse

Figure 41 depicts the vertical timing (timing of one frame), including the vertical sync pulse and the data. All parameters shown in the diagram are programmable.

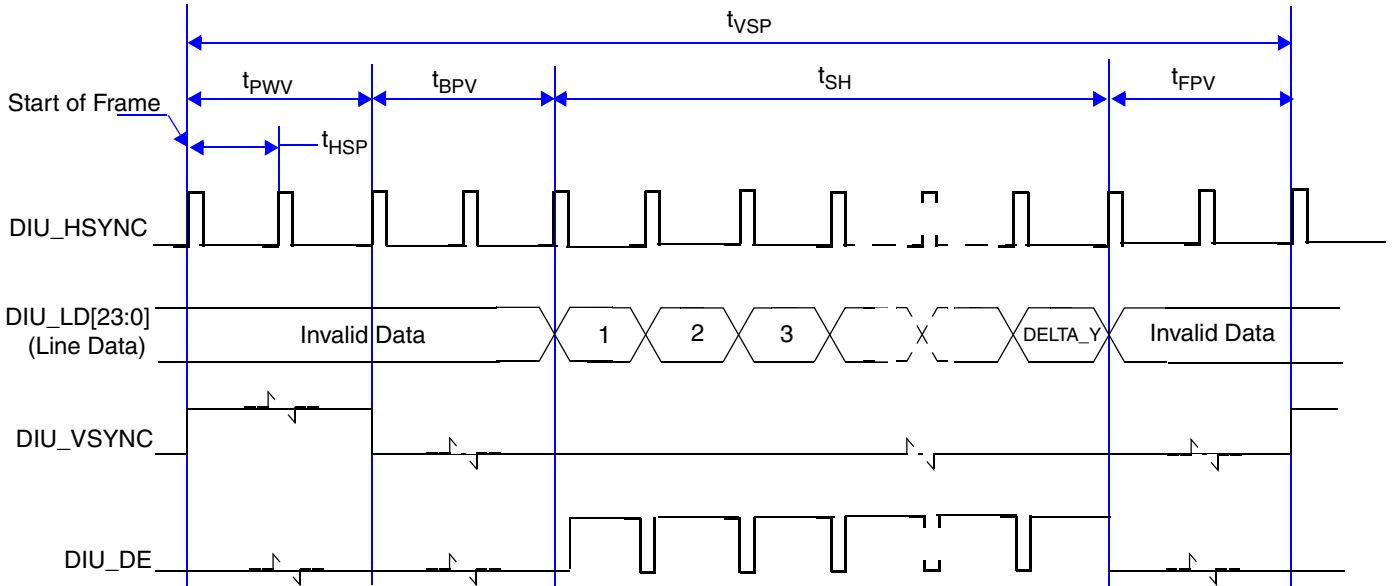


Figure 41. TFT LCD Interface Timing Diagram – Vertical Sync Pulse

Table 39 shows timing parameters of signals.

Table 36. LCD Interface Timing Parameters – Pixel Level

Name	Description	Value	Unit	SpecID
t <sub>PCP</sub>	Display Pixel Clock Period	15 <sup>1</sup>	ns	A15.1
t <sub>PWH</sub>	HSYNC Pulse Width	PW_H * t <sub>PCP</sub>	ns	A15.2
t <sub>BPH</sub>	HSYNC Back Porch Width	BP_H * t <sub>PCP</sub>	ns	A15.3
t <sub>FPH</sub>	HSYNC Front Porch Width	FP_H * t <sub>PCP</sub>	ns	A15.4
t <sub>SW</sub>	Screen Width	DELTA_X * t <sub>PCP</sub>	ns	A15.5
t <sub>HSP</sub>	HSYNC (Line) Period	(PW_H + BP_H + DELTA_X + FP_H) * t <sub>PCP</sub>	ns	A15.6
t <sub>PWV</sub>	VSYNC Pulse Width	PW_V * t <sub>HSP</sub>	ns	A15.7
t <sub>BPV</sub>	VSYNC Back Porch Width	BP_V * t <sub>HSP</sub>	ns	A15.8
t <sub>FPV</sub>	VSYNC Front Porch Width	FP_V * t <sub>HSP</sub>	ns	A15.9
t <sub>SH</sub>	Screen Height	DELTA_Y * t <sub>HSP</sub>	ns	A15.10
t <sub>VSP</sub>	VSYNC (Frame) Period	(PW_V + BP_V + DELTA_Y + FP_V) * t <sub>HSP</sub>	ns	A15.11

<sup>1</sup> Display interface pixel clock period immediate value (in nanosecond).

The DELTA\_X and DELTA\_Y parameters are programmed via the DISP\_SIZE register; The PW\_H, BP\_H, and FP\_H parameters are programmed via the HSYN\_PARA register; And the PW\_V, BP\_V and FP\_V parameters are programmed via the VSYN\_PARA register. See appropriate section in the reference manual for detailed descriptions on these parameters.

Figure 38 depicts the synchronous display interface timing for access level, and Table 39 lists the timing parameters.

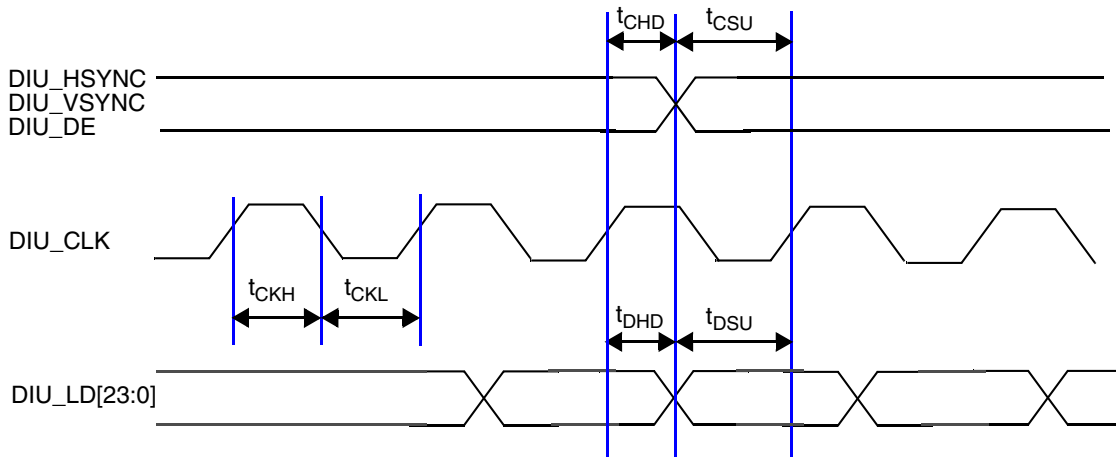


Figure 42. LCD Interface Timing Diagram – Access Level

Table 37. LCD Interface Timing Parameters – Access Level

Parameter	Description	Min	Typ	Max	Unit	SpecID
$t_{CKH}$	LCD Interface Pixel Clock High Time	$t_{PCP} * 0.4$	$t_{PCP} * 0.5$	$t_{PCP} * 0.6$	ns	A15.12
$t_{CKL}$	LCD Interface Pixel Clock Low Time	$t_{PCP} * 0.4$	$t_{PCP} * 0.5$	$t_{PCP} * 0.6$	ns	A15.13
$t_{DSU}$	LCD Interface Data Setup Time	5.0	-	-	ns	A15.14
$t_{DHD}$	LCD Interface Data Hold Time	6.0	-	-	ns	A15.15
$t_{CSU}$	LCD Interface Control Signal Setup Time	5.0	-	-	ns	A15.16
$t_{CHD}$	LCD Interface Control Signal Hold Time	6.0	-	-	ns	A15.17

### 2.3.16 SPDIF

The Sony/Philips Digital Interface (SPDIF) timing is totally asynchronous, therefore there is no need for relationship with the clock.

### 2.3.17 CAN

The CAN functions are available as TX pins at normal IO pads and as RX pins at the VBAT\_RTC domain. There is no filter for the WakeUp dominant pulse. Any High-to-Low edge can cause WakeUp, if configured.

### 2.3.18 I<sup>2</sup>C

This section specifies the timing parameters of the Inter-Integrated Circuit (I2C) interface. Refer to the I2C-Bus Specification.

Table 38. I<sup>2</sup>C Input Timing Specifications – SCL and SDA

Sym	Description	Min	Max	Units	SpecID
1	Start condition hold time	2	—	IP-Bus Cycle <sup>1</sup>	A18.1
2	Clock low time	8	—	IP-Bus Cycle <sup>1</sup>	A18.2
4	Data hold time	0.0	—	ns	A18.3
6	Clock high time	4	—	IP-Bus Cycle <sup>1</sup>	A18.4

**Table 38. I<sup>2</sup>C Input Timing Specifications – SCL and SDA (continued)**

Sym	Description	Min	Max	Units	SpecID
7	Data setup time	0.0	—	ns	A18.5
8	Start condition setup time (for repeated start condition only)	2	—	IP-Bus Cycle <sup>1</sup>	A18.6
9	Stop condition setup time	2	—	IP-Bus Cycle <sup>1</sup>	A18.7

<sup>1</sup> Inter Peripheral Clock is defined in the MPC5121e Reference Manual.

**Table 39. I<sup>2</sup>C Output Timing Specifications – SCL and SDA**

Sym	Description	Min	Max	Units	SpecID
1 <sup>1</sup>	Start condition hold time	6	—	IP-Bus Cycle <sup>2</sup>	A18.8
2 <sup>1</sup>	Clock low time	10	—	IP-Bus Cycle <sup>2</sup>	A18.9
3 <sup>3</sup>	SCL/SDA rise time	—	7.9	ns	A18.10
4 <sup>1</sup>	Data hold time	7	—	IP-Bus Cycle <sup>2</sup>	A18.11
5 <sup>1</sup>	SCL/SDA fall time	—	7.9	ns	A18.12
6 <sup>1</sup>	Clock high time	10	—	IP-Bus Cycle <sup>2</sup>	A18.13
7 <sup>1</sup>	Data setup time	2	—	IP-Bus Cycle <sup>2</sup>	A18.14
8 <sup>1</sup>	Start condition setup time (for repeated start condition only)	20	—	IP-Bus Cycle <sup>2</sup>	A18.15
9 <sup>1</sup>	Stop condition setup time	10	—	IP-Bus Cycle <sup>2</sup>	A18.16

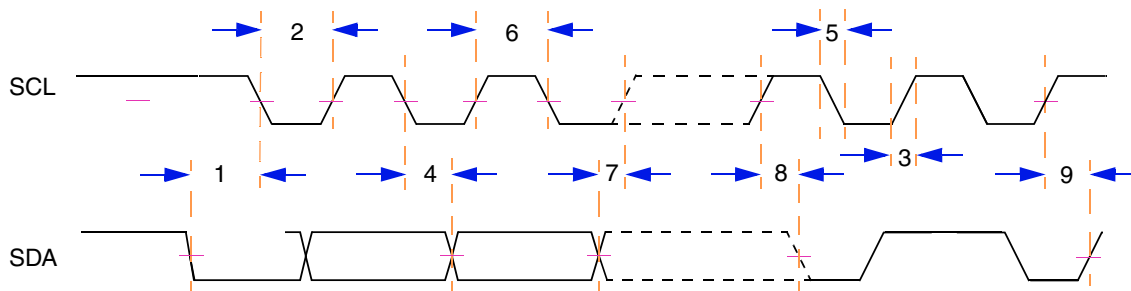
<sup>1</sup> Programming IFDR with the maximum frequency results in the minimum output timings listed. The I<sup>2</sup>C interface is designed to scale the data transition time, moving it to the middle of the SCL low period. The actual position is affected by the prescale and division values programmed in IFDR.

<sup>2</sup> Because SCL and SDA are open-drain-type outputs, which the processor can only actively drive low, the time SCL or SDA takes to reach a high level depends on external signal capacitance and pull-up resistor values

<sup>3</sup> Inter Peripheral Clock is defined in the MPC5121e Reference Manual

**NOTE**

Output timing is specified at a nominal 50 pF load.



**Figure 43. Timing Diagram – I<sup>2</sup>C Input/Output**

**2.3.19 J1850**

See the MPC5121e Reference Manual.



## 2.3.20 PSC

The Programmable Serial Controllers (PSC) support different modes of operation (Codec, AC97, IrDA, SPI).

All the timing numbers specified for different PSC modes are design targets.

### 2.3.20.1 Codec Mode (8,16,24 and 32-bit)/I<sup>2</sup>S Mode

Table 40. Timing Specifications – 8,16, 24, and 32-bit CODEC/I<sup>2</sup>S Master Mode

Sym	Description	Min	Typ	Max	Units	SpecID
1	Bit Clock cycle time, programmed in CCS register	40.0	—	—	ns	A20.1
2	Clock duty cycle	45	50	55	% <sup>1</sup>	A20.2
3	Bit Clock fall time	—	—	7.9	ns	A20.3
4	Bit Clock rise time	—	—	7.9	ns	A20.4
5	FrameSync valid after clock edge	—	—	8.4	ns	A20.5
6	FrameSync invalid after clock edge	—	—	8.4	ns	A20.6
7	Output Data valid after clock edge	—	—	9.3	ns	A20.7
8	Input Data setup time	6.0	—	—	ns	A20.8

<sup>1</sup> Bit Clock cycle time

#### NOTE

Output timing is specified at a nominal 50 pF load.

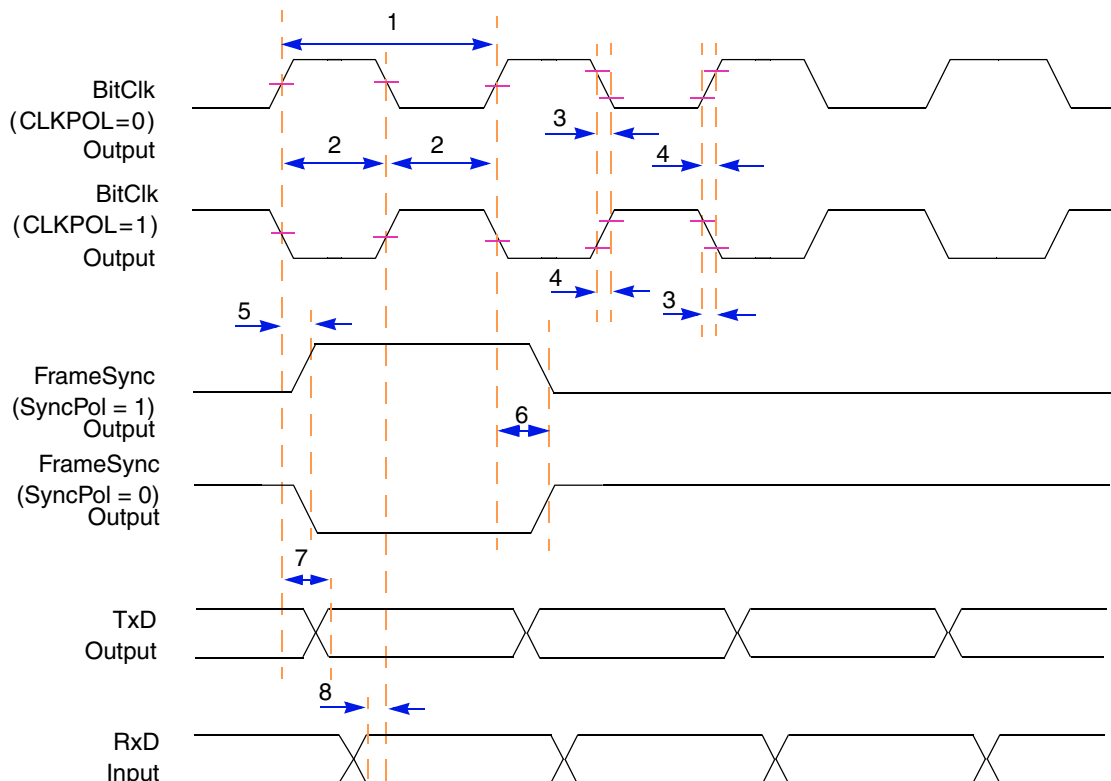


Figure 44. Timing Diagram – 8,16, 24, and 32-bit CODEC/I<sup>2</sup>S Master Mode

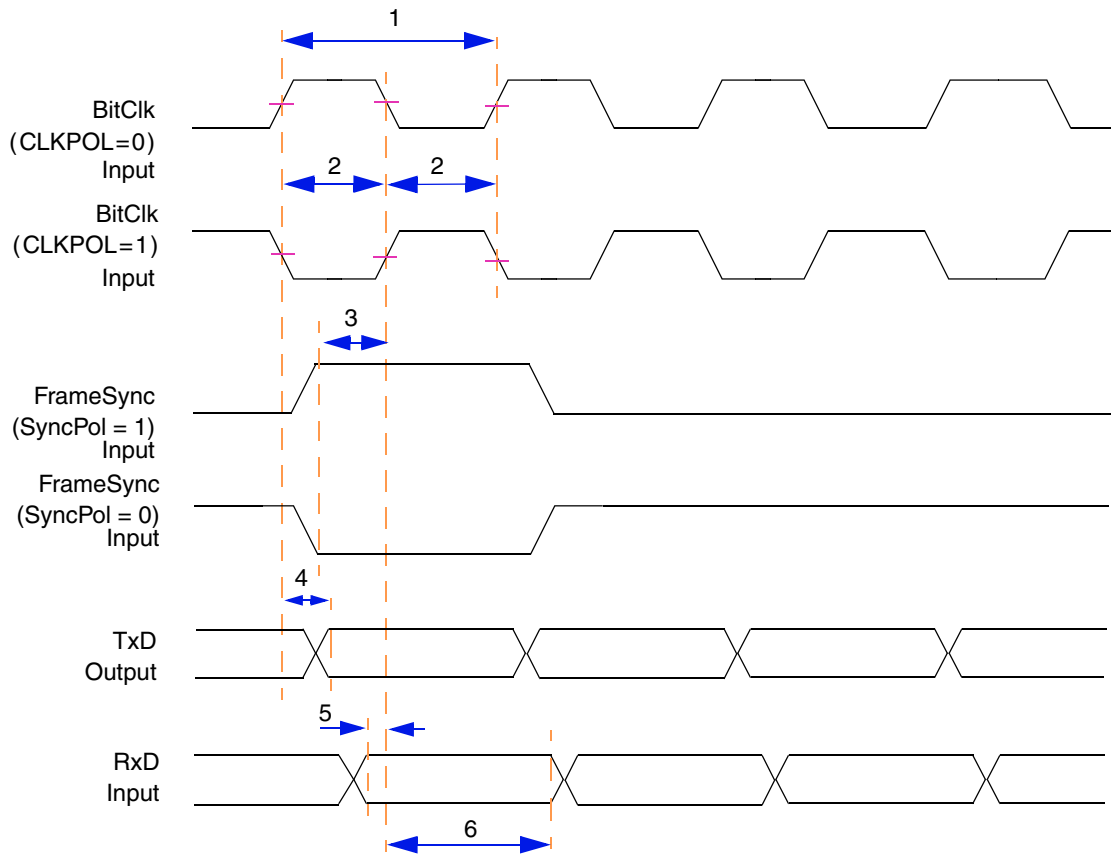
**Table 41. Timing Specifications – 8,16, 24, and 32-bit CODEC/I<sup>2</sup>S Slave Mode**

Sym	Description	Min	Typ	Max	Units	SpecID
1	Bit Clock cycle time	40.0	—	—	ns	A20.9
2	Clock duty cycle	—	50	—	% <sup>1</sup>	A20.10
3	FrameSync setup time	1.0	—	—	ns	A20.11
4	Output Data valid after clock edge	—	—	14.0	ns	A20.12
5	Input Data setup time	1.0	—	—	ns	A20.13
6	Input Data hold time	1.0	—	—	ns	A20.14

<sup>1</sup> Bit Clock cycle time

**NOTE**

Output timing is specified at a nominal 50 pF load.



**Figure 45. Timing Diagram – 8,16, 24, and 32-bit CODEC/I<sup>2</sup>S Slave Mode**

### 2.3.20.2 AC97 Mode

Table 42. Timing Specifications – AC97 Mode

Sym	Description	Min	Typ	Max	Units	SpecID
1	Bit Clock cycle time	—	81.4	—	ns	A20.15
2	Clock pulse high time	—	40.7	—	ns	A20.16
3	Clock pulse low time	—	40.7	—	ns	A20.17
4	FrameSync valid after rising clock edge	—	—	13.0	ns	A20.18
5	Output Data valid after rising clock edge	—	—	14.0	ns	A20.19
6	Input Data setup time	1.0	—	—	ns	A20.20
7	Input Data hold time	1.0	—	—	ns	A20.21

#### NOTE

Output timing is specified at a nominal 50 pF load.

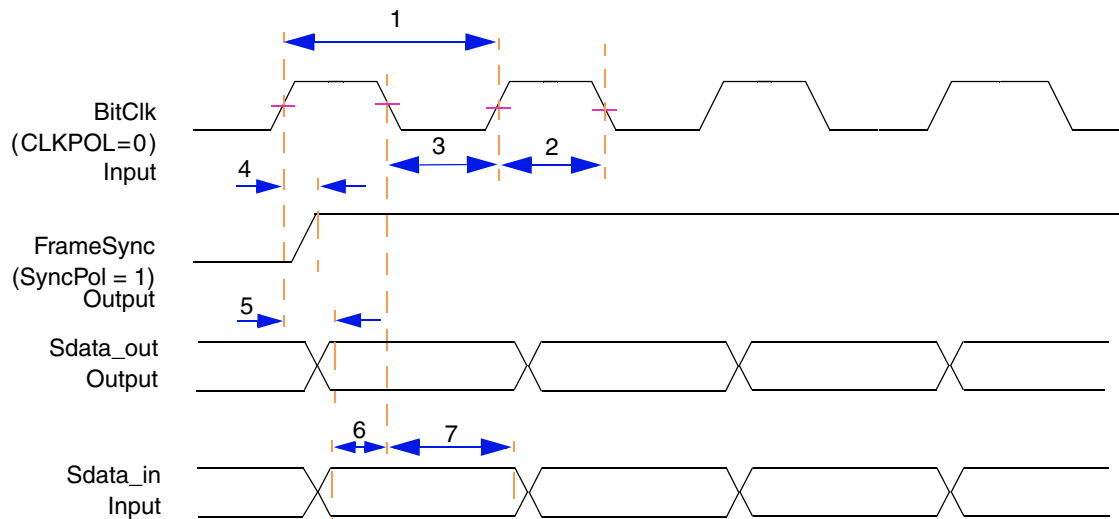


Figure 46. Timing Diagram – AC97 Mode

### 2.3.20.3 IrDA Mode

Table 43. Timing Specifications – IrDA Transmit Line

Sym	Description	Min	Max	Units	SpecID
1	Pulse high time, defined in the IrDA protocol definition	0.125	10000	μs	A20.22
2	Pulse low time, defined in the IrDA protocol definition	0.125	10000	μs	A20.23
3	Transmitter rising time	—	7.9	ns	A20.24
4	Transmitter falling time	—	7.9	ns	A20.25

#### NOTE

Output timing is specified at a nominal 50 pF load.

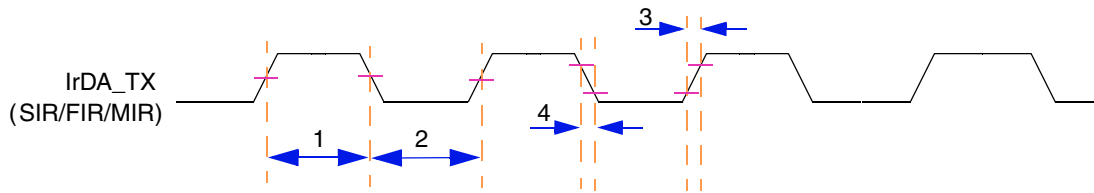


Figure 47. Timing Diagram – IrDA Transmit Line

### 2.3.20.4 SPI Mode

Table 44. Timing Specifications – SPI Master Mode, Format 0 (CPHA = 0)

Sym	Description	Min	Max	Units	SpecID
1	SCK cycle time, programmable in the PSC CCS register	30.0	—	ns	A20.26
2	SCK pulse width, 50% SCK duty cycle	15.0	—	ns	A20.27
3	Slave select clock delay, programmable in the PSC CCS register	30.0	—	ns	A20.28
4	Output Data valid after Slave Select ( $\overline{SS}$ )	—	8.9	ns	A20.29
5	Output Data valid after SCK	—	8.9	ns	A20.30
6	Input Data setup time	6.0	—	ns	A20.31
7	Input Data hold time	1.0	—	ns	A20.32
8	Slave disable lag time	—	TSCK	ns	A20.33
9	Sequential Transfer delay, programmable in the PSC CTUR / CTLR register	15.0	—	ns	A20.34
10	Clock falling time	—	7.9	ns	A20.35
11	Clock rising time	—	7.9	ns	A20.36

**NOTE**

Output timing is specified at a nominal 50 pF load.

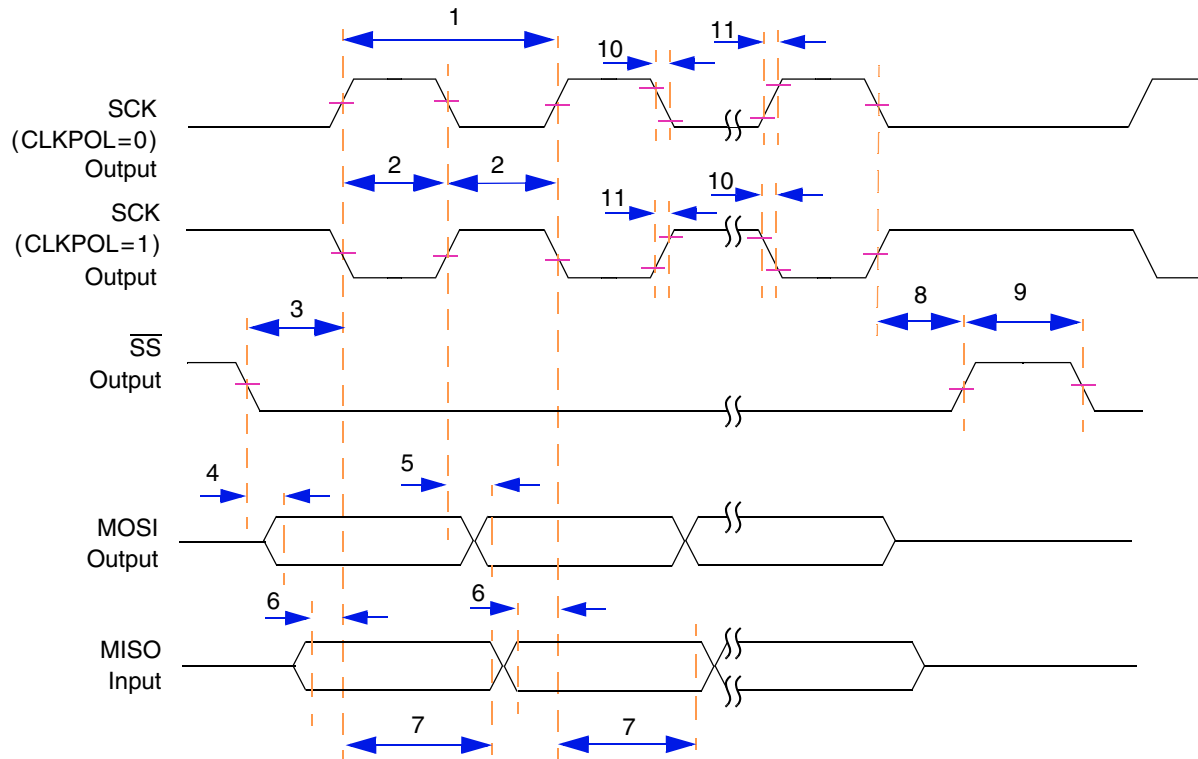


Figure 48. Timing Diagram – SPI Master Mode, Format 0 (CPHA = 0)

Table 45. Timing Specifications – SPI Slave Mode, Format 0 (CPHA = 0)

Sym	Description	Min	Max	Units	SpecID
1	SCK cycle time, programmable in the PSC CCS register	30.0	—	ns	A20.37
2	SCK pulse width, 50% SCK duty cycle	15.0	—	ns	A20.38
3	Slave select clock delay	1.0	—	ns	A20.39
4	Input Data setup time	1.0	—	ns	A20.40
5	Input Data hold time	1.0	—	ns	A20.41
6	Output data valid after $\overline{SS}$	—	14.0	ns	A20.42
7	Output data valid after SCK	—	14.0	ns	A20.43
8	Slave disable lag time	0.0	—	ns	A20.44
9	Minimum Sequential Transfer delay = 2 * IP Bus clock cycle time	30.0	—	—	A20.45

**NOTE**

Output timing is specified at a nominal 50 pF load.

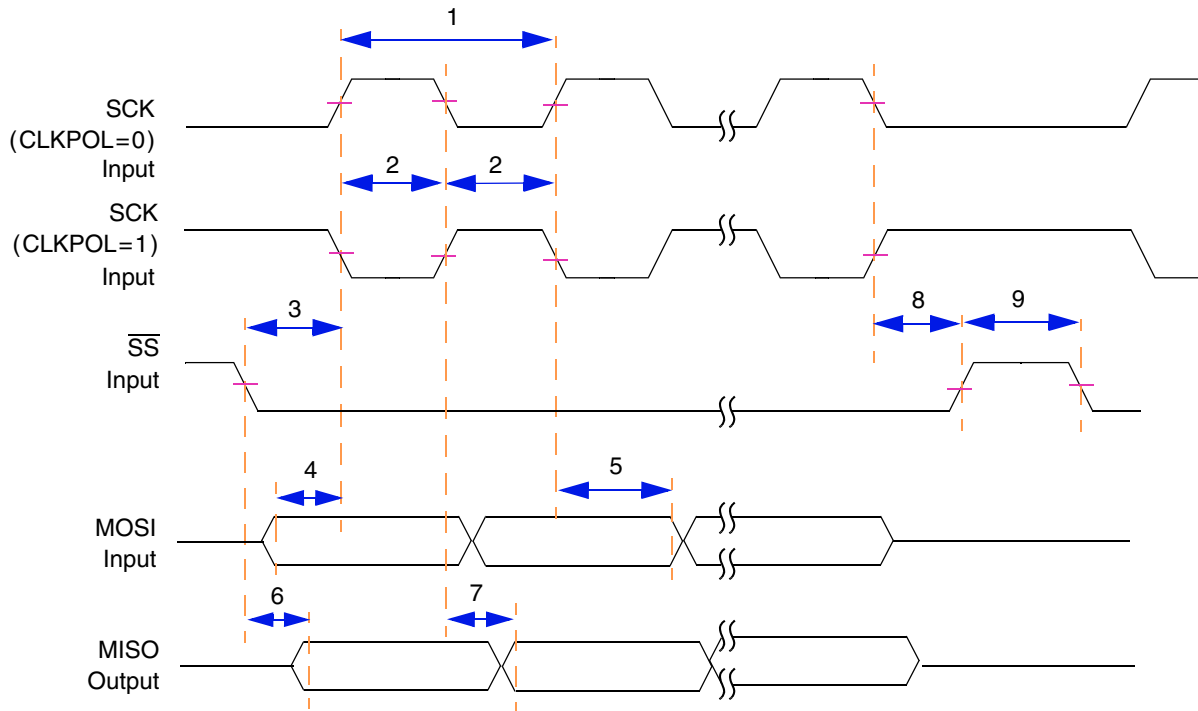


Figure 49. Timing Diagram – SPI Slave Mode, Format 0 (CPHA = 0)

Table 46. Timing Specifications – SPI Master Mode, Format 1 (CPHA = 1)

Sym	Description	Min	Max	Units	SpecID
1	SCK cycle time, programable in the PSC CCS register	30.0	—	ns	A20.46
2	SCK pulse width, 50% SCK duty cycle	15.0	—	ns	A20.47
3	Slave select clock delay, programable in the PSC CCS register	30.0	—	ns	A20.48
4	Output data valid	—	8.9	ns	A20.49
5	Input Data setup time	6.0	—	ns	A20.50
6	Input Data hold time	1.0	—	ns	A20.51
7	Slave disable lag time	—	T <sub>SCK</sub>	ns	A20.52
8	Sequential Transfer delay, programable in the PSC CTUR / CTRL register	15.0	—	ns	A20.53
9	Clock falling time	—	7.9	ns	A20.54
10	Clock rising time	—	7.9	ns	A20.55

**NOTE**

Output timing is specified at a nominal 50 pF load.

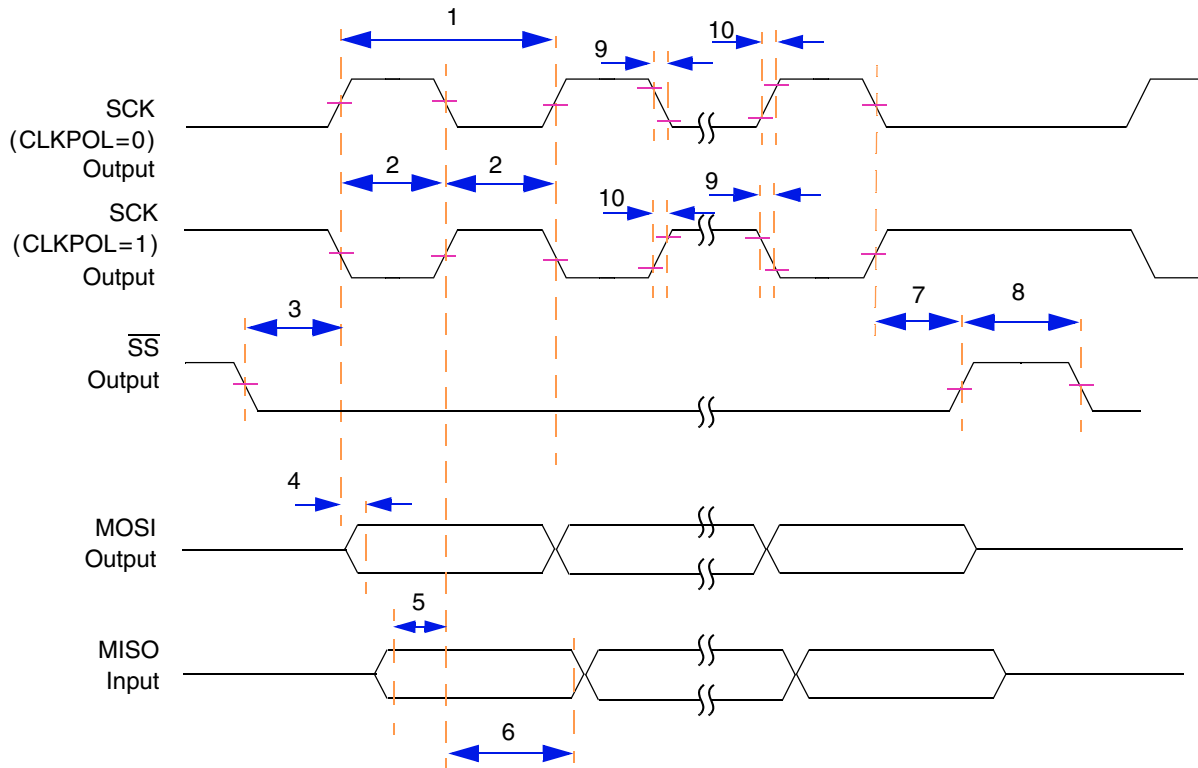


Figure 50. Timing Diagram – SPI Master Mode, Format 1 (CPHA = 1)

Table 47. Timing Specifications – SPI Slave Mode, Format 1 (CPHA = 1)

Sym	Description	Min	Max	Units	SpecID
1	SCK cycle time, programmable in the PSC CCS register	30.0	—	ns	A20.56
2	SCK pulse width, 50% SCK duty cycle	15.0	—	ns	A20.57
3	Slave select clock delay	0.0	—	ns	A20.58
4	Output data valid	—	14.0	ns	A20.59
5	Input Data setup time	2.0	—	ns	A20.60
6	Input Data hold time	1.0	—	ns	A20.61
7	Slave disable lag time	0.0	—	ns	A20.62
8	Minimum Sequential Transfer delay = 2 * IP-Bus clock cycle time	30.0	—	ns	A20.63

**NOTE**

Output timing is specified at a nominal 50 pF load.

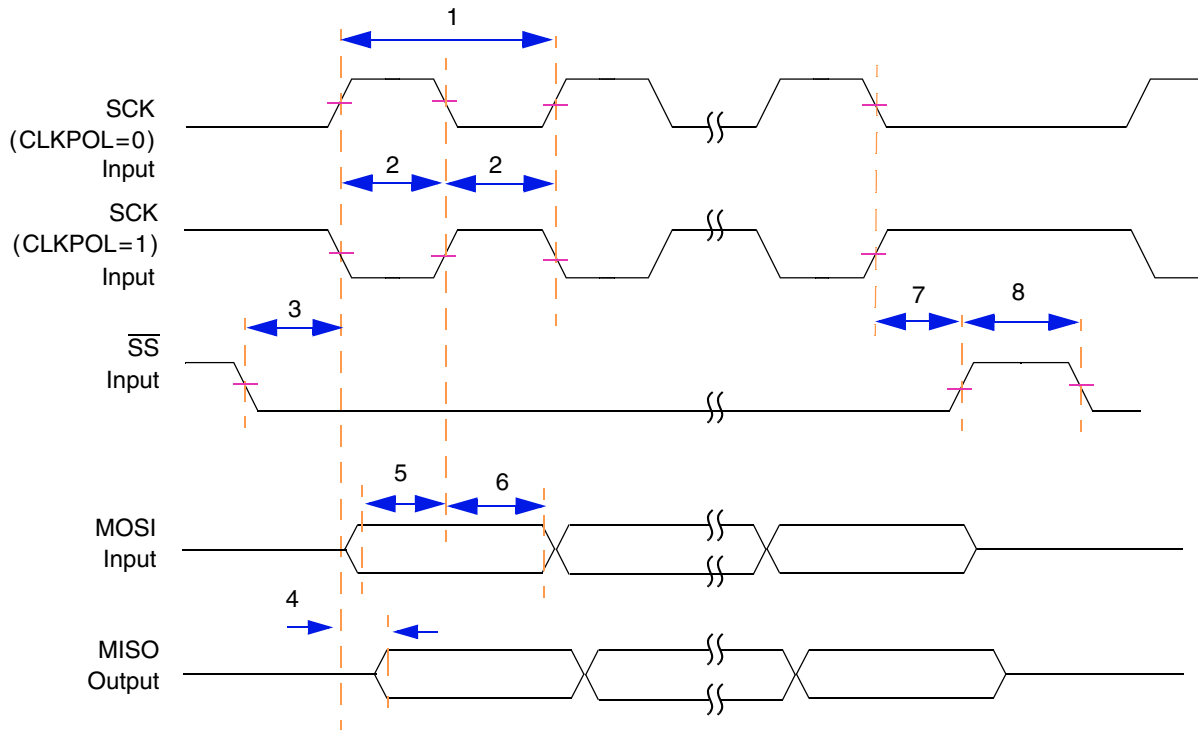


Figure 51. Timing Diagram – SPI Slave Mode, Format 1 (CPHA = 1)

### 2.3.21 GPIOs and Timers

The MPC5121e contains several sets of I/Os that do not require special setup, hold, or valid requirements. The external events (GPIO or timer inputs) are asynchronous to the system clock. The inputs must be valid for at least  $t_{IOWID}$  to ensure proper capture by the internal IP clock.

Table 48. GPIO/Timers Input AC Timing Specifications

Symbol	Description	Min	Unit	SpecID
$t_{IOWID}$	GPIO/Timers inputs - minimum pulse width	$2T^1$	ns	A21.1

<sup>1</sup> T is the IP bus clock cycle. T = 12 ns is the minimum value (for the maximum IP bus frequency of 83 MHz).

### 2.3.22 Fusebox

Table 49 gives the Fusebox timing specification.

Table 49. Fusebox Timing Characteristics

Sym	Description	Min	Max	Units	SpecID
$t_{FUSEWR}$	Program time <sup>1</sup> for Fuse	62.5	—	us	A22.1

<sup>1</sup> The program length is defined by the value defined in the EPM\_PGM\_LENGTH bits of the IIM module.



## 2.3.23 IEEE 1149.1 (JTAG)

Table 50. JTAG Timing Specification

Sym	Characteristic	Min	Max	Unit	SpecID
—	TCK frequency of operation	0	25	MHz	A23.1
1	TCK cycle time	40	—	ns	A23.2
2	TCK clock pulse width measured at 1.5V	1.08	—	ns	A23.3
3	TCK rise and fall times	0	3	ns	A23.4
4	$\overline{\text{TRST}}$ setup time to tck falling edge <sup>1</sup>	10	—	ns	A23.5
5	$\overline{\text{TRST}}$ assert time	5	—	ns	A23.6
6	Input data setup time <sup>2</sup>	5	—	ns	A23.7
7	Input data hold time <sup>2</sup>	15	—	ns	A23.8
8	TCK to output data valid <sup>3</sup>	0	30	ns	A23.9
9	TCK to output high impedance <sup>3</sup>	0	30	ns	A23.10
10	TMS, TDI data setup time.	5	—	ns	A23.11
11	TMS, TDI data hold time.	1	—	ns	A23.12
12	TCK to TDO data valid.	0	15	ns	A23.13
13	TCK to TDO high impedance.	0	15	ns	A23.14

<sup>1</sup>  $\overline{\text{TRST}}$  is an asynchronous signal. The setup time is for test purposes only.

<sup>2</sup> Non-test, other than TDI and TMS, signal input timing with respect to TCK.

<sup>3</sup> Non-test, other than TDO, signal output timing with respect to TCK.

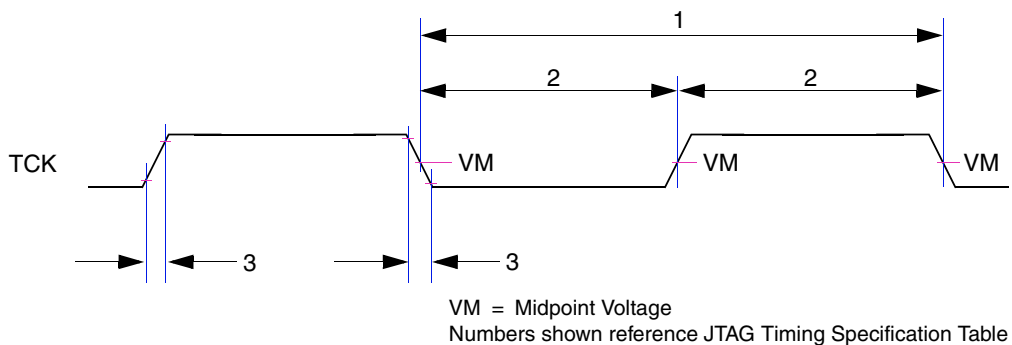


Figure 52. Timing Diagram – JTAG Clock Input

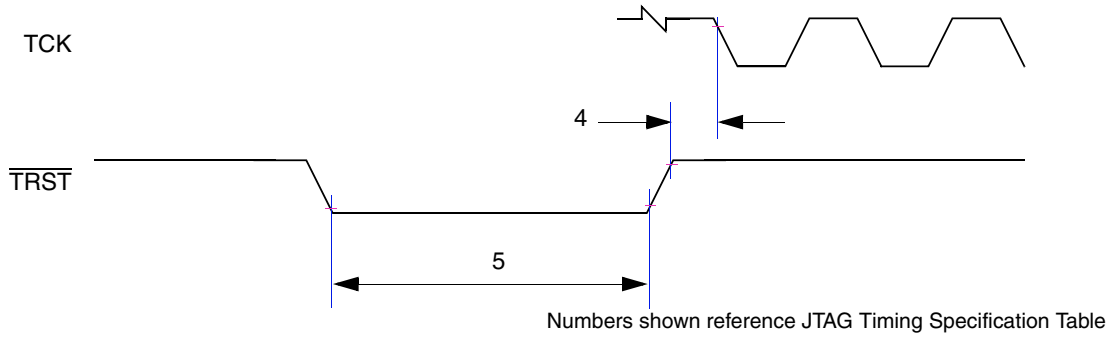


Figure 53. Timing Diagram – JTAG TRST

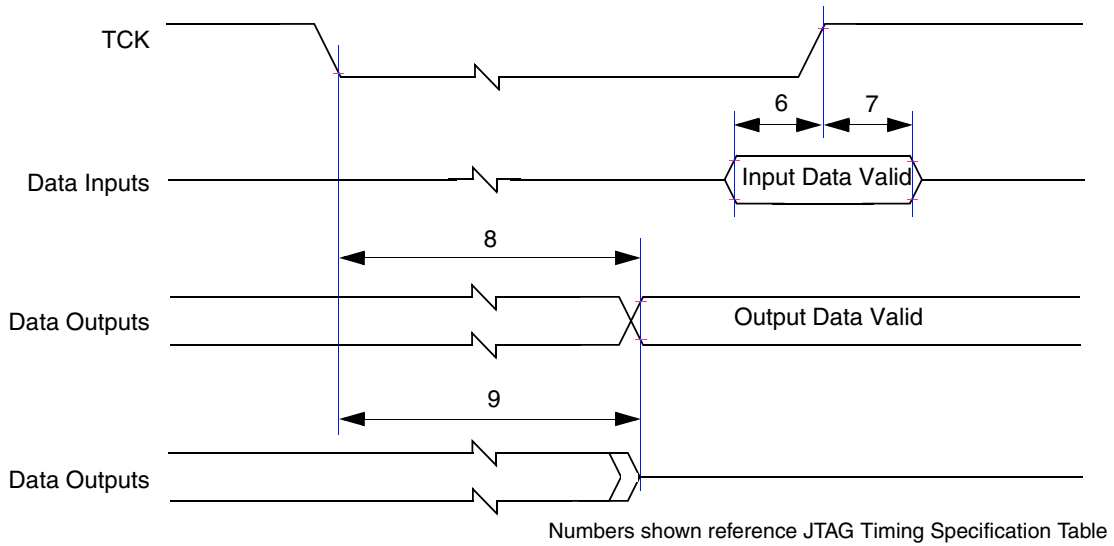


Figure 54. Timing Diagram – JTAG Boundary Scan

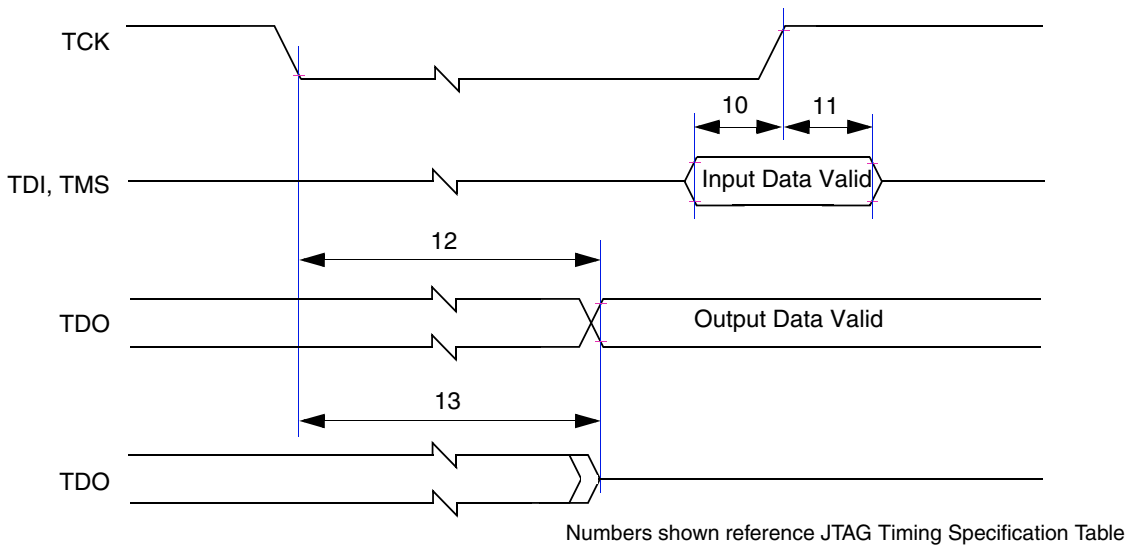


Figure 55. Timing Diagram – Test Access Port

## 3 System Design Information

### 3.1 Power Up/Down Sequencing

Power sequencing between the 1.4 V power supply VDD\_CORE and the remaining supplies is required to prevent excessive current during power up phase.

The recommended power sequence is as follows:

- Use 12V/millisecond or slower time for all supplies.
- Power up VDD\_IO, PLL\_AVDD, VBAT\_RTC (if not applied permanently), VDD\_MEM\_IO, AVDD\_FUSERD, USB PHY & SATA PHY supplies first in any order and then power up VDD\_CORE. If required AVDD\_FUSEWR should be powered up afterwards.
- All the supplies must reach the specified operating conditions before the  $\overline{\text{PORESET}}$  can be released.
- For power down, drop AVDD\_FUSEWR to 0V first, drop VDD\_CORE to 0V, and then drop all other supplies.
- VDD\_CORE should not exceed VDD\_IO, VDD\_MEM\_IO, VBAT\_RTC or PLL\_AVDDs by more than 0.4 V at any time, including power-up.

### 3.2 System and CPU Core AVDD Power Supply Filtering

Each of the independent PLL power supplies require filtering external to the device. The following drawing [Figure 56](#) is a recommendation for the required filter circuit.

Each circuit should be placed as close as possible to the specific AV<sub>DD</sub> pin being supplied to minimize noise coupled from nearby circuits.

All traces should be as low impedance as possible, especially ground pins to the ground plane.

The filter for System/Core PLLVDD to VSS should be connected to the power and ground planes, respectively, not fingers of the planes.

In addition to keeping the filter components for System/Core PLLVDD as close as practical to the body of the MPC5121e as previously mentioned, special care should be taken to avoid coupling switching power supply noise or digital switching noise onto the portion of that supply between the filter and the MPC5121e.

The capacitors for C2 in the figure below should be rated X5R or better due to temperature performance.

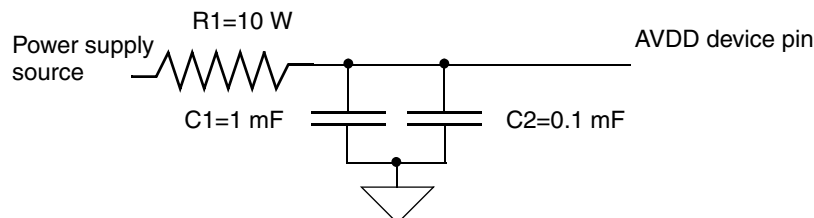


Figure 56. Power Supply Filtering

### 3.3 Connection Recommendations

To ensure reliable operation, connect unused inputs to an appropriate signal level. Unused active low inputs should be tied to VDD\_IO. Unused active high inputs should be connected to VSS. All NC (no-connect) signals must remain unconnected.

Power and ground connections must be made to all external VDD and VSS pins of the MPC5121e.

The unused AVDD\_FUSEWR power should be connected to VSS directly or via a resistor.

For DDR or LPDDR modes the unused pins VTT[3:0] for DDR2 Termination voltage can be unconnected.

## System Design Information

The SATA PHY needs to be powered even if it is not used in an application. In this case, you should not enable the SATA oscillator and the SATA PHY by software.

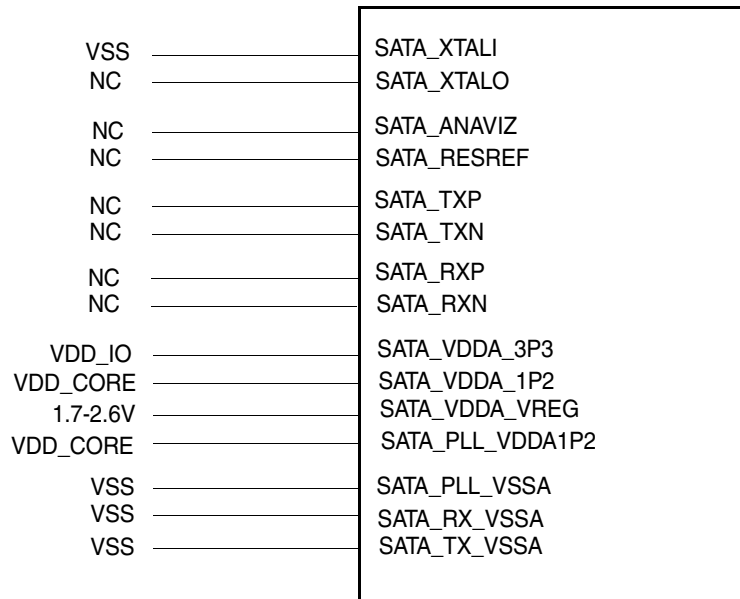


Figure 57. Recommended Connection for Pins of Unused SATA PHY

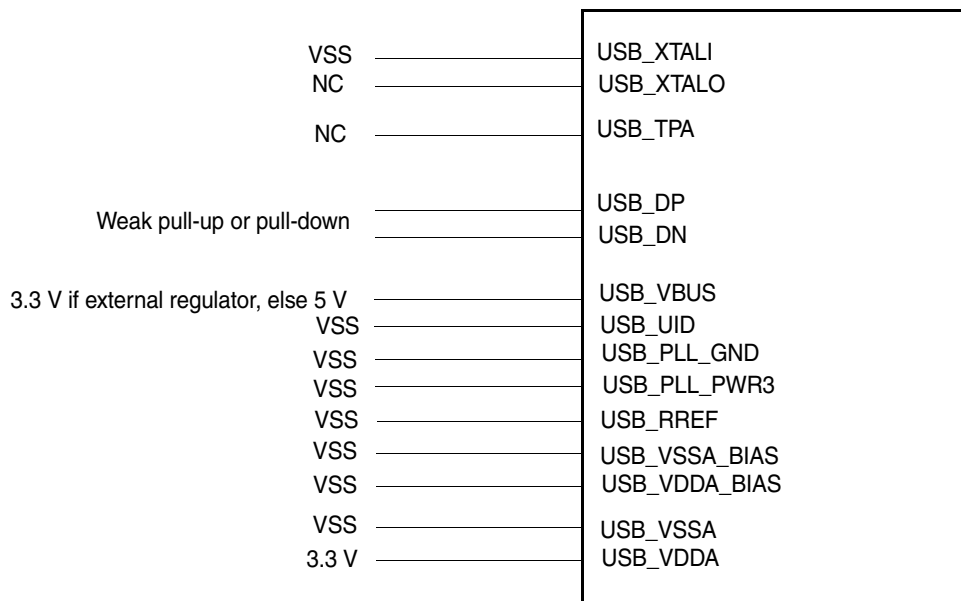


Figure 58. Recommended connection for pins of unused USB PHY

## 3.4 Pull-Up/Pull-Down Resistor Requirements

The MPC5121e requires external pull-up or pull-down resistors on certain pins.

### 3.4.1 Pull-Down Resistor Requirements for TEST pin

The MPC5121e requires a pull-down resistor on the test pin TEST.

## 3.4.2 Pull-Up Requirements for the PCI Control Lines

PCI control signals always require pull-up resistors on the motherboard (not the expansion board) to ensure that they contain stable values when no agent is actively driving the bus. This includes  $\overline{\text{PCI\_FRAME}}$ ,  $\overline{\text{PCI\_TRDY}}$ ,  $\overline{\text{PCI\_IRDY}}$ ,  $\overline{\text{PCI\_DEVSEL}}$ ,  $\overline{\text{PCI\_STOP}}$ ,  $\overline{\text{PCI\_SERR}}$ ,  $\overline{\text{PCI\_PERR}}$ , and  $\overline{\text{PCI\_REQ}}$ .

Refer to the PCI Local Bus specification.

## 3.5 JTAG

The MPC5121e provides you with an IEEE 1149.1 JTAG interface to facilitate board/system testing. It also provides a Common On-Chip Processor (COP) Interface, which shares the IEEE 1149.1 JTAG port.

The COP Interface provides access to the MPC5121e's embedded e300 processor and to other on-chip resources. This interface provides a means for executing test routines and for performing software development and debug functions.

### 3.5.1 JTAG\_TRST

Boundary scan testing is enabled through the JTAG interface signals. The JTAG\_TRST signal is optional in the IEEE 1149.1 specification but is provided on all processors that implement the PowerPC architecture. To obtain a reliable power-on reset performance, the JTAG\_TRST signal must be asserted during power-on reset.

#### 3.5.1.1 TRST and PORESET

The JTAG interface can control the direction of the MPC5121e I/O pads via the boundary scan chain. The JTAG module must be reset before the MPC5121e comes out of power-on reset; do this by asserting TRST before PORESET is released.

For more details refer to the Reset and JTAG Timing Specification.

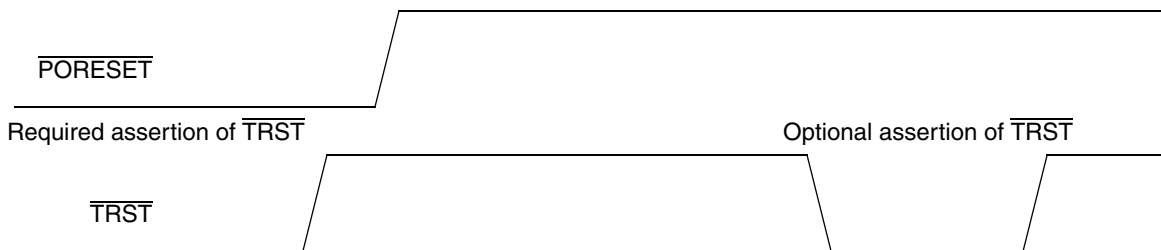


Figure 59.  $\overline{\text{PORESET}}$  vs.  $\overline{\text{TRST}}$

### 3.5.2 e300 COP/BDM Interface

There are two possibilities to connect the JTAG interface: using it with a COP connector and without a COP connector.

#### 3.5.2.1 Boards Interfacing the JTAG Port via a COP Connector

The MPC5121e functional pin interface and internal logic provides access to the embedded e300 processor core through the Freescale standard COP/BDM interface. Table 51 gives the COP/BDM interface signals. The pin order shown reflects only the COP/BDM connector order.

Table 51. COP/BDM Interface Signals

BDM Pin #	MPC5121e I/O Pin	BDM Connector	Internal Pull Up/Down	External Pull Up/Down	I/O <sup>1</sup>
16	—	GND	—	—	—
15	$\overline{\text{CKSTP\_OUT}}$	ckstp_out	—	10k Pull-Up	I
14	—	KEY	—	—	—
13	$\overline{\text{HRESET}}$	hreset	—	10k Pull-Up	O
12	—	GND	—	—	—
11	$\overline{\text{SRESET}}$	sreset	—	10k Pull-Up	O
10	—	N/C	—	—	—
9	TMS	tms	Pull-Up	10k Pull-Up	O
8	$\overline{\text{CKSTP\_IN}}$	ckstp_in	—	10k Pull-Up	O
7	TCK	tck	Pull-Up	10k Pull-Up	O
6	—	VDD <sup>2</sup>	—	—	—
5	See Note <sup>3</sup>	halted <sup>3</sup>	—	—	I
4	$\overline{\text{TRST}}$	trst	Pull-Up	10k Pull-Up	O
3	TDI	tdi	Pull-Up	10k Pull-Up	O
2	See Note <sup>4</sup>	qack <sup>4</sup>	—	—	O
1	TDO	tdo	—	—	I

<sup>1</sup> With respect to the emulator tool's perspective:  
Input is really an output from the embedded e300 core.  
Output is really an input to the core.

<sup>2</sup> From the board under test, power sense for chip power.

<sup>3</sup> HALTED is not available from e300 core.

<sup>4</sup> Input to the e300 core to enable/disable soft-stop condition during breakpoints. MPC5121e internally ties CORE\_QACK to GND in its normal/functional mode (always asserted).

For a board with a COP (common on-chip processor) connector that accesses the JTAG interface and needs to reset the JTAG module, only wiring  $\overline{\text{JTAG\_TRST}}$  and  $\overline{\text{PORESET}}$  is not recommended.

To reset the MPC5121e via the COP connector, the  $\overline{\text{HRESET}}$  pin of the COP should be connected to the  $\overline{\text{HRESET}}$  pin of the MPC5121e. The circuitry shown in Figure 60 allows the COP to assert  $\overline{\text{HRESET}}$  or  $\overline{\text{JTAG\_TRST}}$  separately, while any other board sources can drive  $\overline{\text{PORESET}}$ .

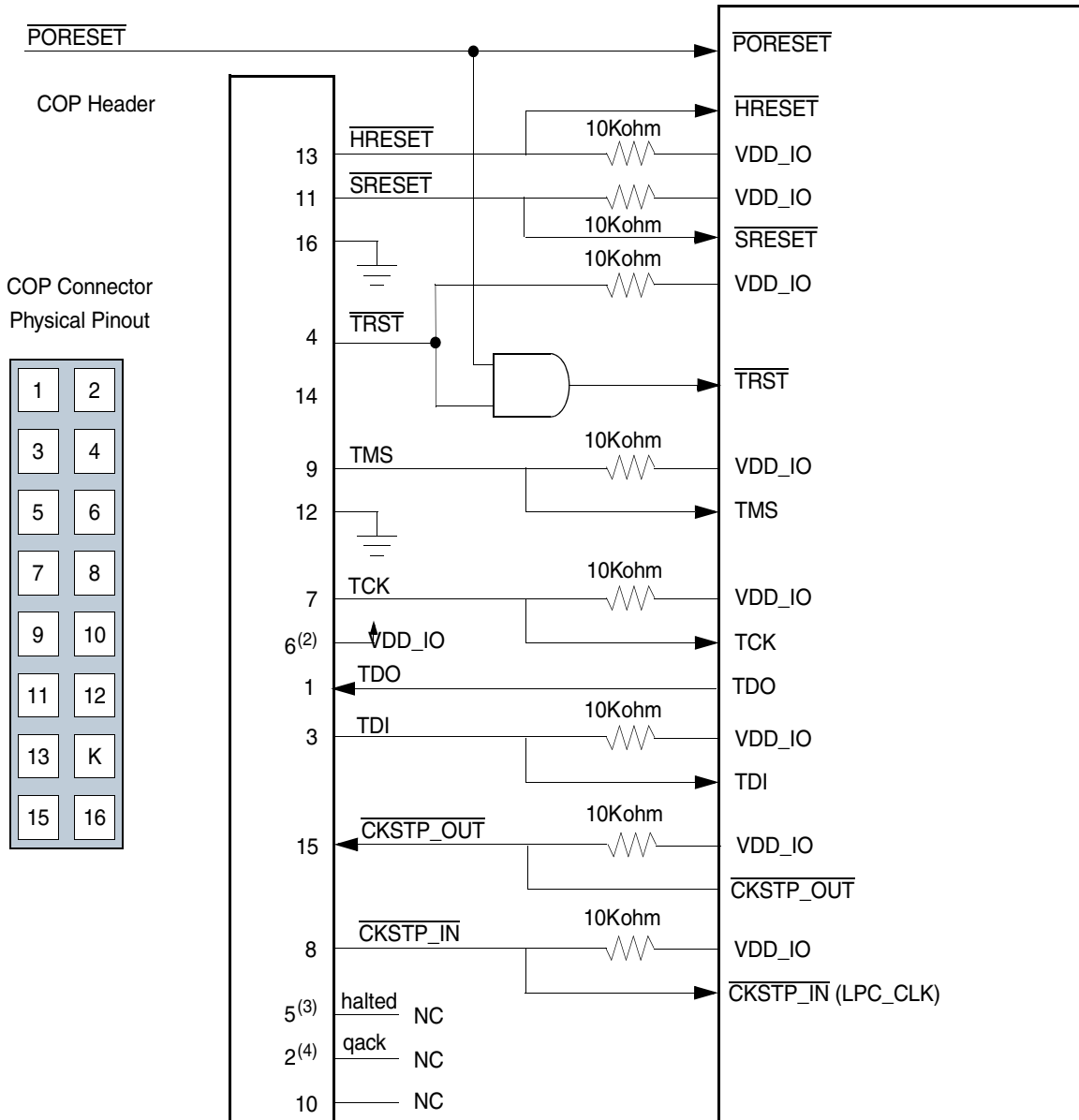


Figure 60. COP Connector Diagram

### 3.5.2.2 Boards Without COP Connector

If the JTAG interface is not used,  $\overline{\text{JTAG\_TRST}}$  should be tied to  $\overline{\text{PORESET}}$ , so that it is asserted when the system reset signal ( $\overline{\text{PORESET}}$ ) is asserted. This ensures that the JTAG scan chain is initialized during power on. Figure 61 shows the connection of the JTAG interface without COP connector.





## 4.1 Package Parameters

Table 52. TEPBGA Parameters

Package outline	27 mm × 27 mm
Interconnects	516
Pitch	1.00 mm
Module height (typical)	2.25 mm
Solder Balls	96.5 Sn/3.5Ag (VY package)
Ball diameter (typical)	0.6 mm

## 4.2 Mechanical Dimensions

# Package Information

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	26			
A		VSS	VSS	SATA_RXN	SATA_RXP	SATA_RX_VSSA	PSC7_4	PSC7_3	PSC6_4	PSC6_2	PSC6_0	PSC11_0	PSC10_2	PSC2_3	PSC1_3	PSC1_1	PSC0_1	CAN1_TX	GPIO2_8	XTALO_RTC	USB2_DRVVBUS	USB2_DM	USB2_DP	USB2_TPA	VSS				
B	VSS	VSS	VSS	SATA_RX_VSSA	VSS	PSC8_3	VSS	PSC7_0	PSC6_3	VDD_I_O	PSC11_1	VSS	PSC10_1	PSC2_1	VDD_I_O	PSC0_4	VSS	GPIO3_1	CAN2_RX	VSS	USB2_VBUS_PWR_FAULT	VSS	USB2_VSSA_BIAS	USB2_XTALO	VDD_I_O	VSS			
C	VSS	SATA_XTALO	SATA_XTALI	VSS	SATA_VDDA_1P2	PSC9_0	PSC8_2	PSC7_2	AVDD_FUSEWR	PSC6_1	PSC11_2	PSC10_3	PSC10_0	PSC2_0	PSC1_0	PSC0_3	PSC_MCLK_IN	GPIO3_0	CAN1_RX	XTALI_RTC	USB2_VDDA	USB2_VSSA	VSS	USB2_XTALI	VSS	PCI_C_LK			
D	SATA_VDDA_1P2	VSS	SATA_PLL_VSSA	SATA_VDDA_3P3	SATA_VDDA_VREG	PSC9_3	PSC9_1	PSC8_1	AVDD_FUSERD	VDD_I_O	PSC11_4	VSS	PSC2_4	PSC1_4	VDD_I_O	PSC0_0	VSS	HIB_MODE	VBAT_RTC	USB2_VDDA	USB2_VBUS	USB2_VSSA_BIAS	USB2_PLL_PWR3	VSS	VSS	PCI_REO2			
E	SATA_TXN	SATA_VDDA_1P2	SATA_PLL_VDDA1P2	SATA_RESREF	SATA_ANAVIZ	PSC9_4	PSC9_2	PSC8_4	PSC8_0	PSC7_1	PSC11_3	PSC10_4	PSC2_2	PSC1_2	PSC0_2	CAN2_TX	GPIO2_9	VSS	USB2_UID	USB2_VSSA	USB2_VSSA	USB2_RREF	USB2_PLL_GND	PCI_G_NT2	PCI_G_NT0	PCI_REO1			
F	SATA_TXP	VSS	VSS	VSS	VSS	VSS	VSS	VSS	VSS	VDD_I_O	VDD_I_O	VDD_I_O	VDD_I_O	VSS	VSS	VSS	VSS	VDD_I_O	VDD_I_O	VSS	PCI_RST	VDD_I_O	PCI_A_D30	VDD_I_O	PCI_A_D28	PCI_A_D26			
G	SATA_TX_VSSA	NFC_RE	NFC_WE	NFC_WP	VSS																	PCI_G_NT1	PCI_REO0	PCI_A_D29	PCI_A_D26	PCI_C_BE3			
H	NFC_RB	PATA_DACK	NFC_CE0	NFC_ALE	NFC_CLE	VSS															VDD_I_O	PCI_A_D31	VSS	PCI_A_D24	VSS	PCI_A_D21			
J	PATA_OR	PATA_OCHR DY	PATA_NTRQ	PATA_DRQ	VDD_I_O																	PCI_A_D27	PCI_A_D25	PCI_A_D23	PCI_A_D20	PCI_A_D18			
K	PATA_CET	VDD_I_O	PATA_SOLATE	VDD_I_O	PATA_OW	VSS				VDD	VDD	VDD	VDD	VDD	VDD	VDD	VDD						PCI_I_DSEL	PCI_A_D22	PCI_A_D19	PCI_A_D17	PCI_RDY		
L	EMB_AD03	EMB_AD02	EMB_AD01	EMB_AD00	PATA_CE2	VSS				VDD	VSS	VSS	VSS	VSS	VSS	VSS	VDD						VSS	PCI_A_D16	VDD_I_O	PCI_C_BE2	VDD_I_O	PCI_D_EVSE L	
M	EMB_AD06	VSS	EMB_AD05	VSS	EMB_AD04					VDD	VSS	VSS	VSS	VSS	VSS	VSS	VDD							PCI_T_RDY	PCI_F_RAME	PCI_S_TOP	PCI_P_ERR	PCI_S_ERR	
N	EMB_AD10	EMB_AD09	EMB_AD08	EMB_AD07	VSS	VDD_I_O				VDD	VSS	VSS	VSS	VSS	VSS	VSS	VDD						VDD_I_O	PCI_P_AR	VSS	PCI_C_BE1	VSS	PCI_A_D15	
P	EMB_AD15	EMB_AD14	EMB_AD11	EMB_AD13	EMB_AD12	VDD_I_O				VDD	VSS	VSS	VSS	VSS	VSS	VSS	VDD						VDD_I_O	PCI_C_BE0	PCI_A_D09	PCI_A_D13	PCI_A_D14	PCI_A_D12	
R	EMB_AD17	VDD_I_O	EMB_AD16	VDD_I_O	EMB_AD19					VDD	VSS	VSS	VSS	VSS	VSS	VSS	VDD							PCI_A_D03	PCI_A_D06	PCI_A_D10	PCI_A_D11	PCI_A_D08	
T	EMB_AD22	EMB_AD18	EMB_AD20	EMB_AD21	EMB_AD23	VSS				VDD	VSS	VSS	VSS	VSS	VSS	VSS	VDD						VSS	SYS_PLL_A_VDD	VDD_I_O	PCI_A_D05	VDD_I_O	PCI_A_D07	
U	EMB_AD25	VSS	EMB_AD24	VSS	EMB_AD29	VSS				VDD	VDD	VDD	VDD	VDD	VDD	VDD	VDD							VSS	SYS_PLL_A_VSS	PCI_NTA	PCI_A_D00	PCI_A_D02	PCI_A_D04
V	EMB_AD26	EMB_AD27	EMB_AD28	EMB_AD30	EMB_AX01																			SRES_ET	VSS	SYS_XTALI	VSS	PCI_A_D01	
W	EMB_AD31	EMB_AX00	EMB_AX02	LPC_AX03	LPC_CS0	VDD_I_O																	VDD_I_O	TDO	PORESET	HRES_ET	TEST	SYS_XTALO	
Y	LPC_CS2	VDD_I_O	LPC_CS1	VDD_I_O	LPC_OE																			J1850_TX	TDI	VSS	TMS	CKST_P_OUT	
AA	LPC_RWB	LPC_A_CK	PSC4_1	LPC_CLK	PSC4_3	VSS		VDD_MEM_I_O		VSS	VSS		VDD_MEM_I_O	VDD_MEM_I_O		VSS	VSS		CORE_PLL_AVDD		VSS	I2C2_SDA	VDD_I_O	J1850_RX	VDD_I_O	TRST			
AB	PSC4_0	VSS	PSC4_2	VSS	PSC3_1	MDQ1	MVTT_0	MDQ5	MDQ1_0	VSS	MVRE_F	MDQ1_9	MDQ2_1	MDQ2_7	MDQ3_1	MA1	MA5	VDD_MEM_I_O	MA14	MCKE	SPDIF_TXCLK	I2C1_SCL	I2C1_SDA	VSS	IRQ1	TCK			
AC	PSC5_0	PSC4_4	PSC5_1	PSC3_2	VDD_MEM_I_O	MDM0	MDQ8	VSS	MDQ1_4	VDD_MEM_I_O	MDQ5_2	VSS	MDQ2_5	VDD_MEM_I_O	MDQ3_0	MBA1	VSS	MA7	MA11	VDD_MEM_I_O	MODT	VSS	I2C0_SCL	SPDIF_RX	I2C2_SCL	IRQ0			
AD	PSC5_2	PSC5_3	VSS	PSC3_3	MDQ5_0	MDQ6	MDQ1_1	MDQ5_1	VDD_MEM_I_O	MDQ1_6	MDQ1_8	MDQ2_0	MDQ2_3	MDQ5_3	MDQ2_9	MBA0	MA0	MA4	MA9	MA13	MWE	MCS	CORE_PLL_AVSS	SPDIF_TX	VSS	I2C0_SDA			
AE	VDD_I_O	VDD_I_O	PSC5_4	MDQ2	VDD_MEM_I_O	MDQ7	VSS	MDM1	MDQ1_2	VDD_MEM_I_O	MVTT_2	VSS	MDQ2_4	MVTT_3	VDD_MEM_I_O	MDQ2_8	VSS	MA2	MA6	VDD_MEM_I_O	MA12	MA15	VSS	VDD_I_O	VDD_I_O	VSS			
AF		VDD_I_O	PSC3_0	PSC3_4	MDQ0	MDQ3	MDQ4	MDQ9	MVTT_1	MDQ1_3	MDQ1_5	MDQ1_7	MDM2	MDQ2_2	MDQ2_6	MDM3	MCK	MCK	MBA2	MA3	MA8	MA10	MRAS	MCAS	VDD_I_O				

Figure 62. Ball Map for the MPC5121e 516-PBGA Package

Figure 63 shows the mechanical dimensions and bottom surface nomenclature of the MPC5121e 516-PBGA package.

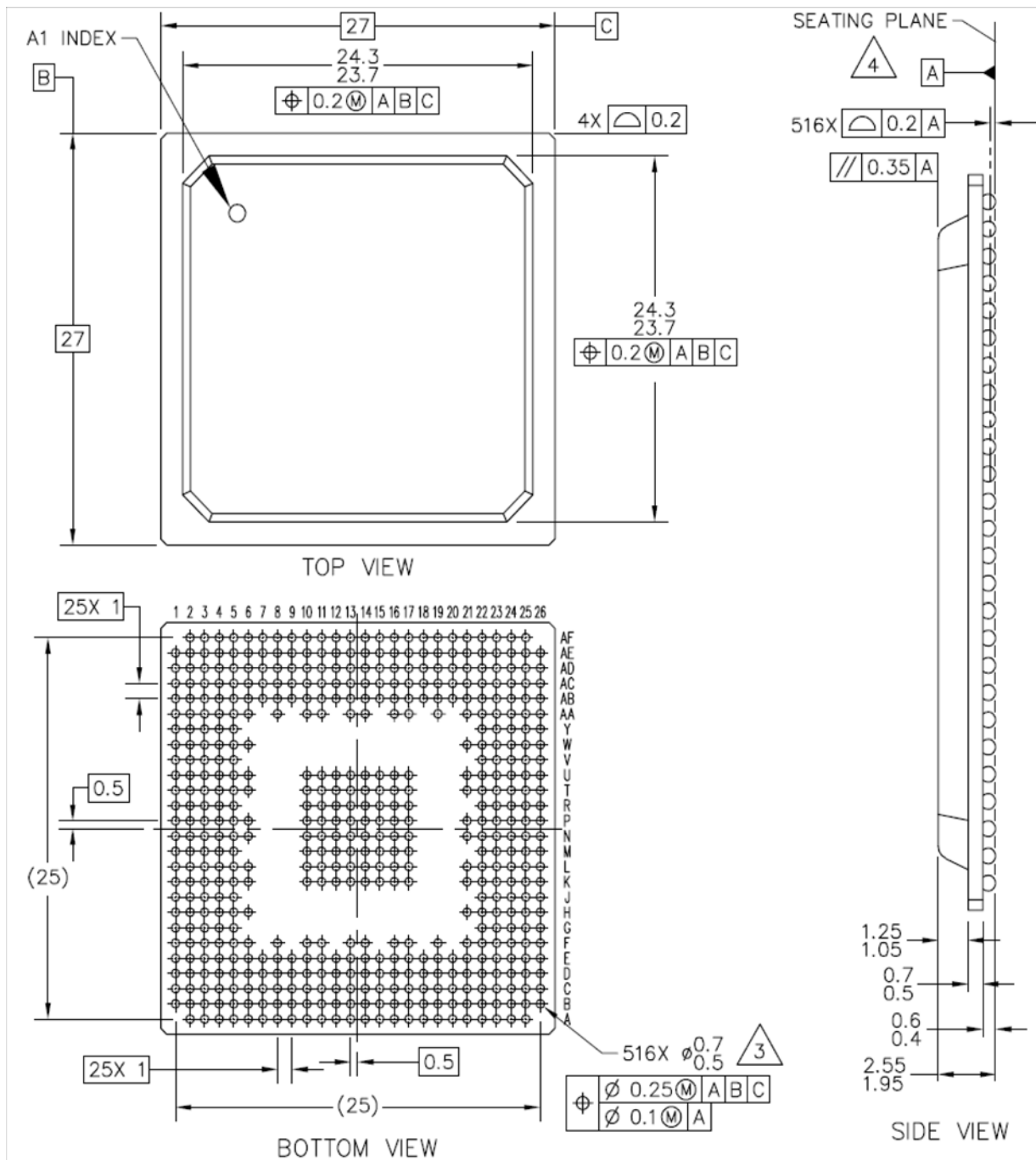


Figure 63. Mechanical Dimension and Bottom Surface Nomenclature of the MPC5121e TEPBGA

- 1 All dimensions are in millimeters.
- 2 Dimensions and tolerances per ASME Y14.5M-1994.
- 3 Maximum solder ball diameter measured parallel to datum A.
- 4 Datum A, the seating plane, is determined by the spherical crowns of the solder balls.

## 5 Product Documentation

This Data Sheet is labeled as a particular type: Product Preview, Advance Information, or Technical Data. Definitions of these types are available at: <http://www.freescale.com>.

Table 53 provides a revision history for this document.

**Table 53. Document Revision History**

Revision	Substantive Change(s)
Rev. 0.01	First preliminary release of AC Characteristics for design implementation (4/2006)
Rev. 0.02	Updated Package Information, Pinout listing, Power Sequencing, DC, AC Characteristics (7/2006)
Rev. 0, Draft A	Reformatted version (7/2006)
Rev. 0, Draft B	Updates to IO list, DC, AC Characteristics, System Information (10/2006)
Rev. 0, Draft D	Updates to IO list, DC, AC Characteristics, (01/2007)
Rev. 0, Draft E	Updates to DC, AC Characteristics, Package, Pinout (02/2007)
Rev. 0, Draft F	Updates to DC, AC Characteristics, System Information (02/2007)
Rev. 0, Draft G	Updates to DC, AC Characteristics, Ball Map, System Information (04/2007)
Rev. 0, Draft H	Updates to DC, AC Characteristics, Ball Map (05/2007)
Rev. 0, Draft I	Updates to DC, AC Characteristics, SpecIDs (07/2007)
Rev. 1, Draft A	Updates to DC VDD_CORE, SpecIDs (08/2007)
Rev. 1, Draft B	Updates to DC: VOH/OL, AC Characteristics, System Info (11/2007)
Rev. 1, Draft C	Updates to DC, AC Characteristics, System Info, Consumer Part (02/2008)



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