











SN65LVDS93A

SLLS992B - AUGUST 2009 - REVISED MARCH 2015

SN65LVDS93A FlatLink™ Transmitter

Features

- Industrial Temperature Range -40°C to 85°C
- LVDS Display Serdes Interfaces Directly to LCD Display Panels With Integrated LVDS
- Package Options: 4.5-mm × 7-mm BGA, and 8.1mm × 14-mm TSSOP
- 1.8 V up to 3.3-V Tolerant Data Inputs to Connect Directly to Low-Power, Low-Voltage Application and Graphic Processors
- Transfer Rate up to 135 Mpps (Mega Pixels Per Second); Pixel Clock Frequency Range 10 MHz to 135 MHz
- Suited for Display Resolutions Ranging From HVGA up to HD With Low EMI
- Operates From a Single 3.3-V Supply and 170 mW (Typical) at 75 MHz
- 28 Data Channels Plus Clock In Low-Voltage TTL to 4 Data Channels Plus Clock Out Low-Voltage Differential
- Consumes Less Than 1 mW When Disabled
- Selectable Rising or Falling Clock Edge Triggered Inputs
- ESD: 5-kV HBM
- Supports Spread Spectrum Clocking (SSC)
- Compatible With all OMAP™2x, OMAP3x, and DaVinci™ Application Processors

2 Applications

- LCD Display Panel Drivers
- UMPC and Netbook PCs
- **Digital Picture Frames**

Description

SN65LVDS93A The LVDS SerDes (serializer/deserializer) transmitter contains four 7-bit parallel load serial-out shift registers, a 7 x clock synthesizer, and five low-voltage differential signaling (LVDS) drivers in a single integrated circuit. These functions allow synchronous transmission of 28 bits of single-ended LVTTL data over five balanced-pair conductors for receipt by a compatible receiver, such as the SN65LVDS94 (SLLS928).

When transmitting, data bits D0 through D27 are each loaded into registers upon the edge of the input clock signal (CLKIN). The rising or falling edge of the clock can be selected through the clock select (CLKSEL) pin. The frequency of CLKIN is multiplied seven times and then used to serially unload the data registers in 7-bit slices. The four serial streams and a phase-locked clock (CLKOUT) are then output to LVDS output drivers. The frequency of CLKOUT is the same as the input clock, CLKIN.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
SN65LVDS93A	TSSOP (56)	14.00 mm × 6.10 mm
	BGA MICROSTAR JUNIOR (56)	7.00 mm × 4.50 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

RGB Video System Using Discrete LVDS TX

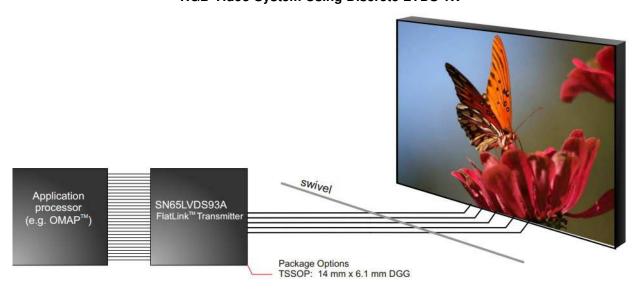




Table of Contents

1	Features 1		9.1 Overview	16
2	Applications 1		9.2 Functional Block Diagram	16
3	Description 1		9.3 Feature Description	17
4	Revision History2		9.4 Device Functional Modes	18
5	Description (continued)	10	Application and Implementation	19
6	Pin Configuration and Functions		10.1 Application Information	19
7	_		10.2 Typical Application	20
1	Specifications 6	11	Power Supply Recommendations	27
	7.1 Absolute Maximum Ratings		Layout	
	7.2 ESD Ratings		12.1 Layout Guidelines	
	7.4 Thermal Information		12.2 Layout Example	
	7.5 Electrical Characteristics	13	Device and Documentation Support	31
	7.6 Timing Requirements 8		13.1 Documentation Support	
	7.7 Switching Characteristics		13.2 Trademarks	31
	7.8 Typical Characteristics		13.3 Electrostatic Discharge Caution	31
8	Parameter Measurement Information		13.4 Glossary	
9	Detailed Description	14	Mechanical, Packaging, and Orderable Information	31

4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision A (August 2011) to Revision B

Page

Changes from Original (August 2009) to Revision A

Page

- Changed t_{en} Enable Time, unit value From: 6 ns To: 6 μs......9

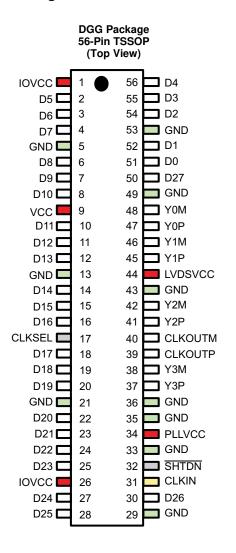


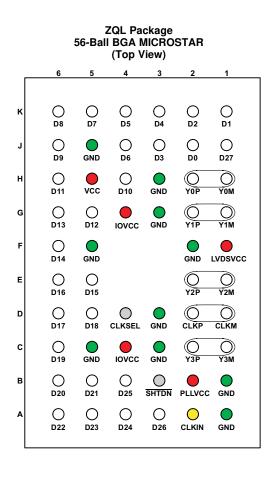
5 Description (continued)

The SN65LVDS93A device requires no external components and little or no control. The data bus appears the same at the input to the transmitter and output of the receiver with the data transmission transparent to the users. The only user intervention is selecting a clock rising edge by inputting a high level to CLKSEL or a falling edge with a low-level input and the possible use of the shutdown/clear (SHTDN) signal. SHTDN is an active-low input to inhibit the clock and shut off the LVDS output drivers for lower power consumption. A low level on this signal clears all internal registers at a low level.

The SN65LVDS93A is characterized for operation over ambient air temperatures of -40°C to 85°C.

6 Pin Configuration and Functions







Pin Functions - TSSOP

	Pin Functions - TSSOP					
	PIN	I/O	DESCRIPTION			
NAME	NO.					
CLKSEL	17	I	Selects between rising edge input clock trigger (CLKSEL = V_{IH}) and falling edge input clock trigger (CLKSEL = V_{IL}).			
CLKIN	31	I	Input pixel clock; rising or falling clock polarity is selectable by Control input CLKSEL.			
CLKOUTM	40	0	Differential LVDS pixel clock output.			
CLKOUTP	39	0	Output is high-impedance when SHTDN is pulled low (de-asserted).			
D0	51					
D1	52					
D2	54					
D3	55					
D4	56					
D5	2					
D6	3					
D7	4		graphic source successfully to a display, the bit assignment of D[27:0] is critical (and no			
D8	6		Data inputs; supports 1.8-V to 3.3-V input voltage selectable by VDD supply. To connect			
D9	7		graphic source successfully to a display, the bit assignment of D[27:0] is critical (and no			
D10	8					
D11	10		Data inputs; supports 1.8-V to 3.3-V input voltage selectable by VDD supply. To conne graphic source successfully to a display, the bit assignment of DI27:01 is critical (and no			
D12	11		graphic source successfully to a display, the bit assignment of D[27:0] is critical (and n			
D13	12	ı	graphic source successfully to a display, the bit assignment of D[27:0] is critical (and not			
D14	14		Note: if application only requires 18-bit color, connect unused inputs D5, D10, D11, D16,			
D15	15		Data inputs; supports 1.8-V to 3.3-V input voltage selectable by VDD supply. To connect graphic source successfully to a display, the bit assignment of D[27:0] is critical (and not necessarily intuitive). Note: if application only requires 18-bit color, connect unused inputs D5, D10, D11, D16, D17, D23, and D27 to GND			
D16	16		graphic source successfully to a display, the bit assignment of D[27:0] is critical (and not necessarily intuitive). Note: if application only requires 18-bit color, connect unused inputs D5, D10, D11, D16,			
D17	18					
D18	19					
D19	20		Data inputs; supports 1.8-V to 3.3-V input voltage selectable by VDD supply. To connect graphic source successfully to a display, the bit assignment of D[27:0] is critical (and not necessarily intuitive). Note: if application only requires 18-bit color, connect unused inputs D5, D10, D11, D16, D17, D23, and D27 to GND Supply Ground for VCC, IOVCC, LVDSVCC, and PLLVCC. I/O supply reference voltage (1.8 V up to 3.3 V matching the GPU data output signal swir 3.3-V LVDS output analog supply 3.3-V LVDS output analog supply			
D20	22					
D21	23					
D22	24					
D23	25					
D24	27					
D25	28					
D26	30					
D27	50					
GND	5, 13, 21, 29, 33, 35, 36, 43, 49, 53	Power	Supply Ground for VCC, IOVCC, LVDSVCC, and PLLVCC.			
IOVCC	1, 26	Supply ⁽¹⁾	I/O supply reference voltage (1.8 V up to 3.3 V matching the GPU data output signal swing)			
LVDSVCC	44		3.3-V LVDS output analog supply			
PLLVCC	34		3.3-V PLL analog supply			
SHTDN	32	I	Device shut down; pull low (de-assert) to shut down the device (low power, resets all registers) and high (assert) for normal operation.			
vcc	9	Power Supply ⁽¹⁾	3.3-V digital supply voltage			

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⁽¹⁾ For a multilayer pcb, TI recommends keeping one common GND layer underneath the device and connecting all ground terminals directly to this plane.



Pin Functions - TSSOP (continued)

Р	PIN	1/0	DECORPORA		
NAME	NO.	I/O	DESCRIPTION		
Y0M	48				
Y1M	46				
Y2M	42	0	Differential LVDS data outputs. Outputs are high-impedance when SHTDN is pulled low (de-asserted)		
Y0P	47		Outputs are high-impedance when SHTDN is pulled low (de-asserted)		
Y1P	45				
Y2P	41				
Y3M	38		Differential LVDS Data outputs.		
Y3P	37	0	Output is high-impedance when SHTDN is pulled low (de-asserted). Note: if the application only requires 18-bit color, this output can be left open.		

Pin Functions - BGA MICROSTAR

BA	ALL					
NAME	NO.	I/O	DESCRIPTION			
CLKIN	A2	CMOS IN with pulldn	Input pixel clock; rising or falling clock polarity is selectable by Control input CLKSEL.			
CLKM	D1	LVDS Out	Differential LVDS pixel clock output.			
CLKP	D2	LVD3 Out	Output is high-impedance when SHTDN is pulled low (de-asserted).			
CLKSEL	D4	CMOS IN with pulldn	Selects between rising edge input clock trigger (CLKSEL = V_{IH}) and falling edge input clock trigger (CLKSEL = V_{IL}).			
D0	J2					
D1	K1					
D2	K2					
D3	J3					
D4	K3					
D5	K4					
D6	J4					
D7	K5					
D8	K6					
D9	J6					
D10	H4					
D11	H6					
D12	G5		Data inputs; supports 1.8-V to 3.3-V input voltage selectable by VDD supply. To			
D13	G6	CMOS IN with	connect a graphic source successfully to a display, the bit assignment of D[27:0] is critical (and not necessarily intuitive).			
D14	F6	pulldn	Note: if application only requires 18-bit color, connect unused inputs D5, D10, D11,			
D15	E5		D16, D17, D23, and D27 to GND.			
D16	E6					
D17	D6					
D18	D5					
D19	C6					
D20	В6					
D21	B5					
D22	A6					
D23	A5					
D24	A4					
D25	B4					
D26	A3					
D27	J1					



Pin Functions - BGA MICROSTAR (continued)

В	ALL	1/0	DECODIDATION		
NAME	NO.	I/O	DESCRIPTION		
GND	A1, B1, C3, C5, F2, F5, J5, D3, G3, H3		Supply Ground for VCC, IOVCC, LVDSVCC, and PLLVCC.		
IOVCC	C4, G4	Power Supply ⁽¹⁾	I/O supply reference voltage (1.8 V up to 3.3 V matching the GPU data output signal swing)		
LVDSVCC	F1		3.3-V LVDS output analog supply		
PLLVCC	B2		3.3-V PLL analog supply		
SHTDN	В3	CMOS IN with pulldn	registers) and high (assert) for normal operation.		
VCC	H5	Power Supply ⁽¹⁾	3.3-V digital supply voltage		
Y0M	H1				
Y1M	G1				
Y2M	E1	LVDC Out	Differential LVDS data outputs.		
Y0P	H2	LVDS Out	Outputs are high-impedance when SHTDN is pulled low (de-asserted)		
Y1P	G2				
Y2P	E2				
Y3M	C1		Differential LVDS Data outputs.		
Y3P	C2	LVDS Out	Output is high-impedance when SHTDN is pulled low (de-asserted). Note: if the application only requires 18-bit color, this output can be left open.		
	E3, E4, F3, F4	_	Not connected		

⁽¹⁾ For a multilayer pcb, it is recommended to keep one common GND layer underneath the device and connect all ground terminals directly to this plane.

7 Specifications

7.1 Absolute Maximum Ratings⁽¹⁾

<u></u>			
	MIN	MAX	UNIT
Supply voltage, VCC, IOVCC, LVDSVCC, PLLVCC ⁽²⁾	-0.5	4	V
Voltage at any output terminal	-0.5	VCC + 0.5	V
Voltage at any input terminal	-0.5	IOVCC + 0.5	V
Continuous power dissipation	See The	rmal Information	
Storage temperature, T _{stg}	-65	150	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltages are with respect to the GND terminals.

7.2 ESD Ratings

			VALUE	UNIT
		Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±5000	
V _(ESD)	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	±500	V

⁽¹⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

PARAMETER		MIN	NOM	MAX	UNIT
Supply voltage, VCC		3	3.3	3.6	
LVDS output supply voltage, LVDSVCC		3	3.3	3.6	
PLL analog supply voltage, PLLVCC		3	3.3	3.6	V
IO input reference supply voltage, IOVCC		1.62	1.8 / 2.5 / 3.3	3.6	
Power supply noise on any VCC terminal				0.1	
	IOVCC = 1.8 V	IOVCC/2 + 0.3 V			
High-level input voltage, V _{IH}	IOVCC = 2.5 V	IOVCC/2 + 0.4 V			V
	IOVCC = 3.3 V	IOVCC/2 + 0.5 V			
	IOVCC = 1.8 V			IOVCC/2 - 0.3 V	
Low-level input voltage, V _{IL}	IOVCC = 2.5 V			IOVCC/2 - 0.4 V	V
	IOVCC = 3.3 V			IOVCC/2 - 0.5 V	
Differential load impedance, Z _L		90		132	Ω
Operating free-air temperature, T _A		-45		85	°C

7.4 Thermal Information

		SN65LVDS93A		
	THERMAL METRIC ⁽¹⁾	ZQL (BGA MICROSTAR)	DGG (TSSOP)	UNIT
		56 PINS	56 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	67.1	62.1	
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	25.2	18.4	
$R_{\theta JB}$	Junction-to-board thermal resistance	31.0	31.1	°C/W
ΨЈТ	Junction-to-top characterization parameter	0.8	0.8	
ΨЈВ	Junction-to-board characterization parameter	30.3	30.8	

⁽¹⁾ For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.

7.5 Electrical Characteristics

over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT
V_{T}	Input voltage threshold			IOVCC/2		٧
$ V_{OD} $	Differential steady-state output voltage magnitude	R_L = 100 Ω , See Figure 7	250		450	mV
$\Delta V_{OD} $	Change in the steady-state differential output voltage magnitude between opposite binary states			1	35	mV
$V_{OC(SS)}$	Steady-state common-mode output voltage	See Figure 7 t _{R/F} (Dx, CLKin) = 1 ns	1.125		1.375	V
V _{OC(PP)}	Peak-to-peak common-mode output voltage				35	mV
I _{IH}	High-level input current	V _{IH} = IOVCC			25	μΑ
I _{IL}	Low-level input current	V _{IL} = 0 V			±10	μΑ
	Chart sireuit sutaut surrent	V _{OY} = 0 V			±24	mA
I _{OS}	Short-circuit output current	V _{OD} = 0 V			±12	mA
l _{OZ}	High-impedance state output current	V _O = 0 V to VCC			±20	μΑ
Б	Input pulldown integrated resistor on all	IOVCC = 1.8 V		200		1.0
R_{pdn}	inputs (Dx, CLKSEL, SHTDN, CLKIN)	IOVCC = 3.3 V		100		kΩ
IQ	Quiescent current	Disabled, all inputs at GND; SHTDN = V _{IL}		2	100	μΑ

(1) All typical values are at VCC = 3.3 V, $T_A = 25 ^{\circ}\text{C}$.



Electrical Characteristics (continued)

over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT
		$I_{(VCC)} + I_{(PLLVCC)} + I_{(LVDSVCC)}$		51.9		
		$I_{(IOVCC)}$ with IOVCC = 3.3 V		0.4		mA
		$I_{(IOVCC)}$ with IOVCC = 1.8 V		0.1		
		$\overline{\text{SHTDN}}$ = V _{IH} , R _L = 100 Ω (5 places), 50% transition density pattern (Figure 8), VCC = 3. 3 V, f _{CLK} = 75 MHz				
		I _(VCC) + I _(PLLVCC) + I _(LVDSVCC)		53.3		
		I _(IOVCC) with IOVCC = 3.3 V		0.6		mA
		I _(IOVCC) with IOVCC = 1.8 V		0.2		
		$\overline{\text{SHTDN}} = \text{V}_{\text{IH}}, \text{ R}_{\text{L}} = 100 \ \Omega \ (5 \text{ places}), \text{ worst-case pattern (Figure 9),} $ VCC = 3.6 V, f _{CLK} = 75 MHz				
I _{CC}	Supply current (average)	I _(VCC) + I _(PLLVCC) + I _(LVDSVCC)		63.7		
		I _(IOVCC) with IOVCC = 3.3 V		1.3		mA
		$I_{(IOVCC)}$ with IOVCC = 1.8 V		0.5		
		$\overline{\text{SHTDN}} = \text{V}_{\text{IH}}, \text{R}_{\text{L}} = 100 \Omega (\text{5 places}), \text{worst-case pattern (Figure 9)}, \\ \text{f}_{\text{CLK}} = 100 \text{MHz}$				
		$I_{(VCC)} + I_{(PLLVCC)} + I_{(LVDSVCC)}$		81.6		
		$I_{(IOVCC)}$ with IOVCC = 3.6 V		1.6		mA
		I _(IOVCC) with IOVCC = 1.8 V		0.6		
		$\overline{\text{SHTDN}} = \text{V}_{\text{IH}}, \text{R}_{\text{L}} = 100 \Omega$ (5 places), worst-case pattern (Figure 9), f _{CLK} = 135 MHz				
		I _(VCC) + I _(PLLVCC) + I _(LVDSVCC)		102.2		
		I _(IOVCC) with IOVCC = 3.6 V		2.1		mA
		I _(IOVCC) with IOVCC = 1.8 V		0.8		
Cı	Input capacitance			2		pF

7.6 Timing Requirements

		MIN	MAX	UNIT
Input clock period, t _c		7.4	100	ns
Input clock modulation	w/ modulation frequency 30 kHz		8%	
	w/ modulation frequency 50 kHz		6%	
High-level input clock pulse widtl	n duration, t _w	0.4 t _c	0.6 t _c	ns
Input signal transition time, t _t			3	ns
Data set up time, D0 through D2	2		ns	
Data hold time, D0 through D27	0.8		ns	



7.7 Switching Characteristics

over operating free-air temperature range (unless otherwise noted)

•		TEST CONDITIONS	MIN	TYP ⁽¹⁾	MAX	UNIT	
t ₀	Delay time, CLKOUT↑ after Yn valid (serial bit position 0, equal D1, D9, D20, D5)		-0.1	0	0.1	ns	
t ₁	Delay time, CLKOUT↑ after Yn valid (serial bit position 1, equal D0, D8, D19, D27)		$^{1}/_{7} t_{c} - 0.1$		$^{1}/_{7} t_{c} + 0.1$	ns	
t ₂	Delay time, CLKOUT↑ after Yn valid (serial bit position 2, equal D7, D18, D26. D23)		$^{2}/_{7} t_{c} - 0.1$		$^{2}/_{7}$ t _c + 0.1	ns	
t ₃	Delay time, CLKOUT↑ after Yn valid (serial bit position 3; equal D6, D15, D25, D17)	See Figure 10, t _C = 10 ns, Input clock jitter < 25 ps ⁽²⁾	$^{3}/_{7} t_{c} - 0.1$		$^{3}/_{7} t_{c} + 0.1$	ns	
t ₄	Delay time, CLKOUT↑ after Yn valid (serial bit position 4, equal D4, D14, D24, D16)		⁴ / ₇ t _c - 0.1		$^{4}/_{7} t_{c} + 0.1$	ns	
t ₅	Delay time, CLKOUT↑ after Yn valid (serial bit position 5, equal D3, D13, D22, D11)		⁵ / ₇ t _c - 0.1		$^{5}/_{7} t_{c} + 0.1$	ns	
t ₆	Delay time, CLKOUT↑ after Yn valid (serial bit position 6, equal D2, D12, D21, D10)		$^{6}/_{7} t_{c} - 0.1$		$^{6}/_{7} t_{c} + 0.1$	ns	
t _{c(o)}	Output clock period			t _c		ns	
		t _C = 10 ns; clean reference clock, see Figure 11		±26			
Λt	Output clock cycle-to-cycle jitter (3)	t _C = 10 ns with 0.05UI added noise modulated at 3 MHz, see Figure 11		±44			
$\Delta t_{c(o)}$	Output clock cycle-to-cycle jittel	t _C = 7.4 ns; clean reference clock, see Figure 11		±35			
		t _C = 7.4 ns with 0.05UI added noise modulated at 3 MHz, see Figure 11		±42			
t _w	High-level output clock pulse duration			4 / $_7$ t_c		ns	
t _{r/f}	Differential output voltage transition time (t _r or t _f)	See Figure 7		225	500	ps	
t _{en}	Enable time, SHTDN↑ to phase lock (Yn valid)	f _(clk) = 135 MHz, See Figure 12		6		μs	
t _{dis}	Disable time, SHTDN↓ to off-state (CLKOUT high-impedance)	f _(clk) = 135 MHz, See Figure 13		7		ns	

All typical values are at $V_{CC}=3.3~V,\,T_A=25^{\circ}C.$ [Input clock jitter] is the magnitude of the change in the input clock period.

The output clock cycle-to-cycle jitter is the largest recorded change in the output clock period from one cycle to the next cycle observed over 15,000 cycles. Tektronix TDSJIT3 Jitter Analysis software was used to derive the maximum and minimum jitter value.



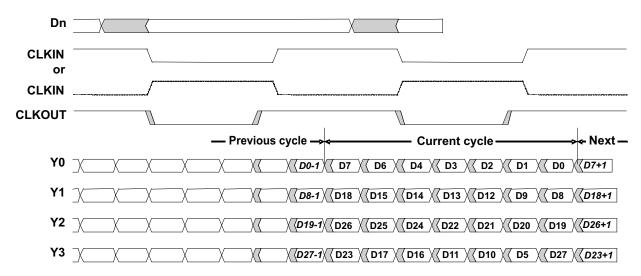
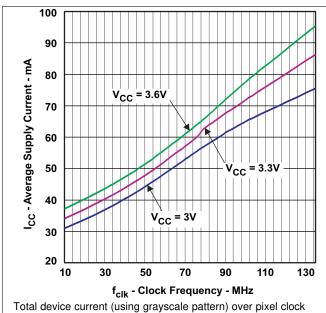


Figure 1. Typical SN65LVDS93A Load and Shift Sequences



7.8 Typical Characteristics



frequency

Figure 2. Average Grayscale ICC vs Clock Frequency

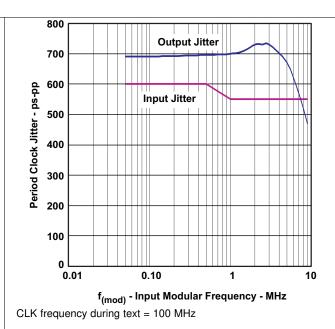
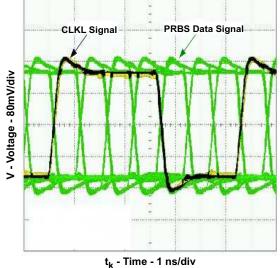


Figure 3. Output Clock Jitter vs Input Clock Jitter



Clock signal = 135 MHz

Figure 4. Typical PRBS Output Signal Over One Clock Period

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8 Parameter Measurement Information

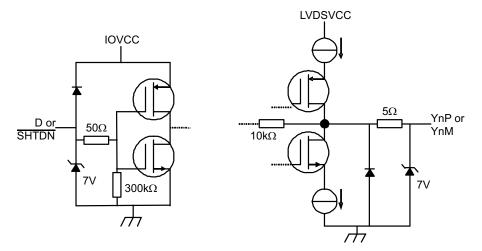
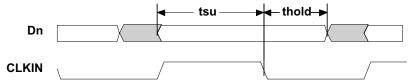


Figure 5. Equivalent Input and Output Schematic Diagrams

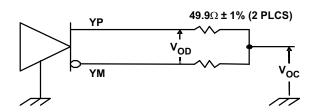


All input timing is defined at IOVDD / 2 on an input signal with a 10% to 90% rise or fall time of less than 3 ns. CLKSEL = 0V.

Figure 6. Setup and Hold Time Definition



Parameter Measurement Information (continued)



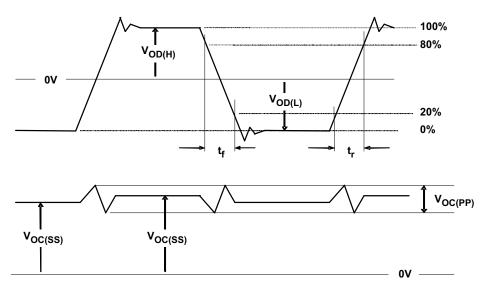
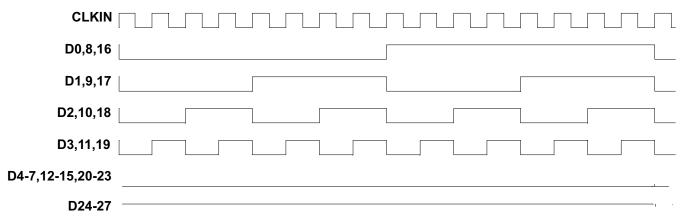


Figure 7. Test Load and Voltage Definitions for LVDS Outputs



The 16 grayscale test pattern test device power consumption for a typical display pattern.

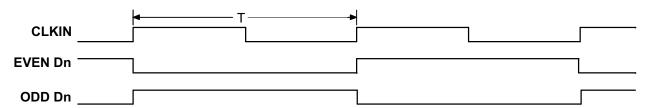
Figure 8. 16 Grayscale Test Pattern

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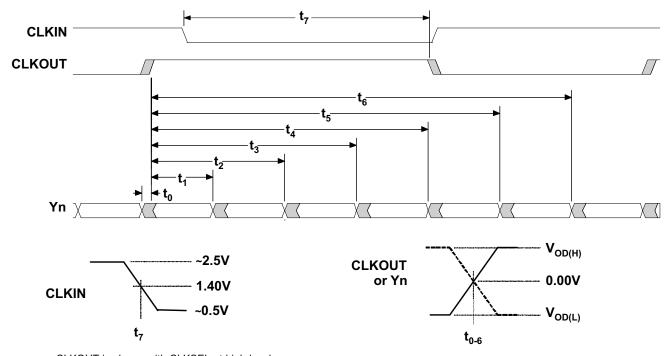


Parameter Measurement Information (continued)



The worst-case test pattern produces nearly the maximum switching frequency for all of the LVDS outputs.

Figure 9. Worst-Case Power Test Pattern



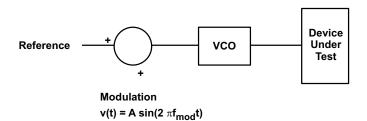
CLKOUT is shown with CLKSEL at high-level. CLKIN polarity depends on CLKSEL input level.

Figure 10. SN65LVDS93A Timing Definitions

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Parameter Measurement Information (continued)



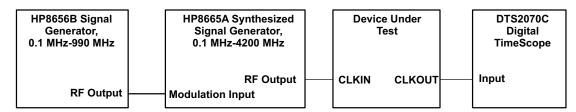


Figure 11. Output Clock Jitter Test Set Up

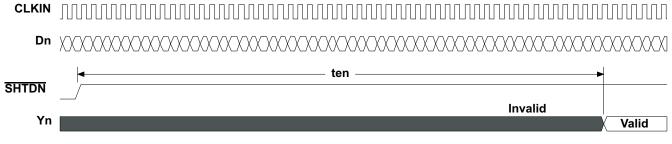


Figure 12. Enable Time Waveforms

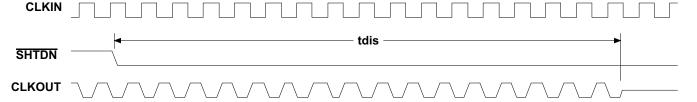


Figure 13. Disable Time Waveforms



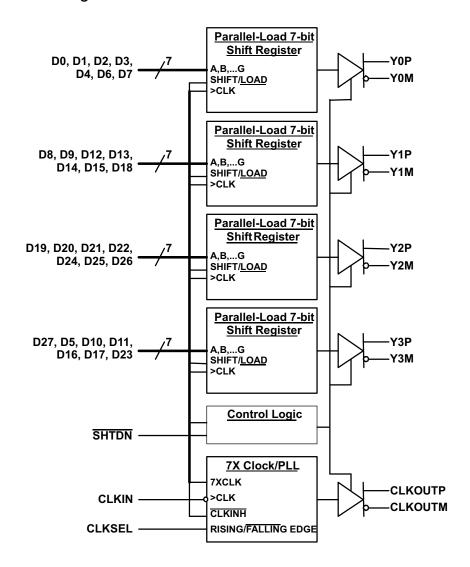
9 Detailed Description

9.1 Overview

FlatLink™ is an LVDS SerDes data transmission system. The SN65LVDS93A takes in three (or four) data words each containing seven single-ended data bits, and converts this to an LVDS serial output. Each serial output runs at seven times that of the parallel data rate. The deserializer (receiver) device operates in the reverse manner. The three (or four) LVDS serial inputs are transformed back to the original 7-bit parallel single-ended data. FlatLink devices are available in 21:3 or 28:4 SerDes ratios.

- The 21-bit devices are designed for 6-bit RGB video for a total of 18 bits in addition to 3 extra bits for horizontal synchronization, vertical synchronization, and data enable.
- The 28-bit devices are intended for 8-bit RGB video applications. Again, the extra 4 bits are for horizontal synchronization, vertical synchronization, data enable, and the remaining is the reserved bit. These 28-bit devices can also be used in 6-bit and 4-bit RGB applications as shown in the subsequent system diagrams.

9.2 Functional Block Diagram



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9.3 Feature Description

9.3.1 TTL Input Data

The data inputs to the transmitter come from the graphics processor and consist of up to 24 bits of video information, a horizontal synchronization bit, a vertical synchronization bit, an enable bit, and a spare bit. The data can be loaded into the registers upon either the rising or falling edge of the input clock selectable by the CLKSEL pin. Data inputs are 1.8 V to 3.3 V tolerant for the SN65LVDS93A and can connect directly to low-power, low-voltage application and graphic processors. The bit mapping is listed in Table 1.

Table 1. Pixel Bit Ordering

	RED	GREEN	BLUE
LSB	R0	G0	В0
	R1	G1	B1
	R2	G2	B2
4-bit MSB	R3	G3	B3
	R4	G4	B4
6-bit MSB	R5	G5	B5
	R6	G6	B6
8-bit MSB	R7	G7	B7

9.3.2 LVDS Output Data

The pixel data assignment is listed in Table 2 for 24-bit, 18-bit, and 12-bit color hosts.

Table 2. Pixel Data Assignment

			8-BIT		6-BIT	4-BIT	
SERIAL CHANNEL	DATA BITS	FORMAT-1	FORMAT-2	FORMAT-3		NON-LINEAR STEP SIZE	LINEAR STEP SIZE
	D0	R0	R2	R2	R0	R2	VCC
	D1	R1	R3	R3	R1	R3	GND
	D2	R2	R4	R4	R2	R0	R0
Y0	D3	R3	R5	R5	R3	R1	R1
	D4	R4	R6	R6	R4	R2	R2
	D6	R5	R7	R7	R5	R3	R3
	D7	G0	G2	G2	G0	G2	VCC
	D8	G1	G3	G3	G1	G3	GND
	D9	G2	G4	G4	G2	G0	G0
	D12	G3	G5	G5	G3	G1	G1
Y1	D13	G4	G6	G6	G4	G2	G2
	D14	G5	G7	G7	G5	G3	G3
	D15	В0	B2	B2	В0	B2	VCC
	D18	B1	В3	В3	B1	В3	GND
	D19	B2	B4	B4	B2	В0	В0
	D20	В3	B5	B5	В3	B1	B1
	D21	B4	В6	B6	B4	B2	B2
Y2	D22	B5	B7	B7	B5	В3	В3
	D24	HSYNC	HSYNC	HSYNC	HSYNC	HSYNC	HSYNC
	D25	VSYNC	VSYNC	VSYNC	VSYNC	VSYNC	VSYNC
	D26	ENABLE	ENABLE	ENABLE	ENABLE	ENABLE	ENABLE



Table 2. Pixel Data Assignment (continued)

SERIAL			8-BIT		6-BIT	4-BIT		
CHANNEL	TO TAKE TO THE TOTAL PROPERTY OF THE TAKE THE TA		FORMAT-2	FORMAT-2 FORMAT-3		NON-LINEAR STEP SIZE	LINEAR STEP SIZE	
	D27	R6	R0	GND	GND	GND	GND	
	D5	R7	R1	GND	GND	GND	GND	
	D10	G6	G0	GND	GND	GND	GND	
Y3	D11	G7	G1	GND	GND	GND	GND	
	D16	В6	В0	GND	GND	GND	GND	
	D17	B7	B1	GND	GND	GND	GND	
	D23	RSVD	RSVD	GND	GND	GND	GND	
CLKOUT	CLKIN	CLK	CLK	CLK	CLK	CLK	CLK	

9.4 Device Functional Modes

9.4.1 Input Clock Edge

The transmission of data bits D0 through D27 occurs as each are loaded into registers upon the edge of the CLKIN signal, where the rising or falling edge of the clock may be selected through CLKSEL. The selection of a clock rising edge occurs by inputting a high level to CLKSEL, which is achieved by populating pullup resistor to pull CLKSEL=high. Inputting a low level to select a clock falling edge is achieved by directly connecting CLKSEL to GND.

9.4.2 Low Power Mode

The SN65LVDS93A can be put in low-power consumption mode by active-low input SHTDN#. Connecting pin SHTDN# to GND will inhibit the clock and shut off the LVDS output drivers for lower power consumption. A low-level on this signal clears all internal registers to a low-level. Populate a pullup to VCC on SHTDN# to enable the device for normal operation.

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10 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

10.1 Application Information

This section describes the power up sequence, provides information on device connectivity to various GPU and LCD display panels, and offers a PCB routing example.

10.1.1 Power

The SN65LVDS93A does not require a specific power-up sequence.

The device is permitted to power up IOVC<u>C</u> while VCC, VCCPLL, and VCCLVDS remain powered down and connected to GND. The input level of the SHTDN during this time does not matter as only the input stage is powered up while all other device blocks are still powered down.

The device is also permitted to power up all 3.3-V power domains while IOVCC is still powered down to GND. The device will not suffer damage. However, in this case, all the I/Os are detected as logic HIGH, regardless of their true input voltage level. Hence, connecting SHTDN to GND will still be interpreted as a logic HIGH; the LVDS output stage will turn on. The power consumption in this condition is significantly higher than standby mode, but still lower than normal mode.

The user experience can be impacted by the way a system powers up and powers down an LCD screen. The following sequence is recommended:

Power-up sequence (SN65LVDS93A SHTDN input initially low):

- 1. Ramp up LCD power (maybe 0.5 ms to 10 ms) but keep backlight turned off.
- 2. Wait for additional 0-200ms to ensure display noise won't occur.
- 3. Enable video source output; start sending black video data.
- 4. Toggle LVDS83B shutdown to SHTDN = V_{IH} .
- 5. Send >1 ms of black video data; this allows the LVDS83B to be phase locked, and the display to show black data first.
- 6. Start sending true image data.
- 7. Enable backlight.

Power-down sequence (SN65LVDS93A SHTDN input initially high):

- 1. Disable LCD backlight; wait for the minimum time specified in the LCD data sheet for the backlight to go low.
- 2. Video source output data switch from active video data to black image data (all visible pixel turn black); drive this for >2 frame times.
- 3. Set SN65LVDS93A input SHTDN = GND; wait for 250 ns.
- 4. Disable the video output of the video source.
- 5. Remove power from the LCD panel for lowest system power.



10.2 Typical Application

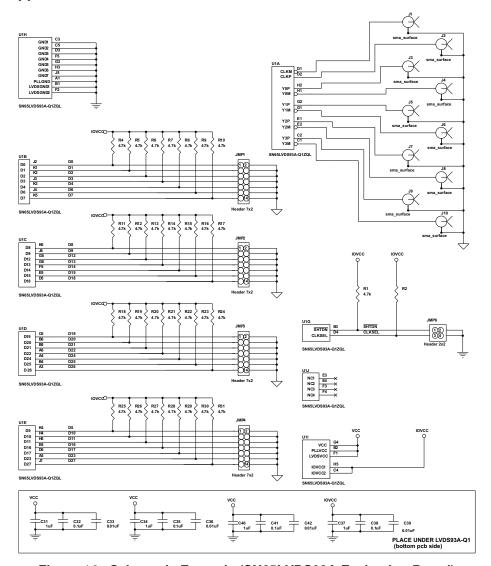


Figure 14. Schematic Example (SN65LVDS93A Evaluation Board)

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Typical Application (continued)

10.2.1 Design Requirements

For this design example, use the parameters listed in Table 3 as the input parameters.

Table 3. Design Parameters

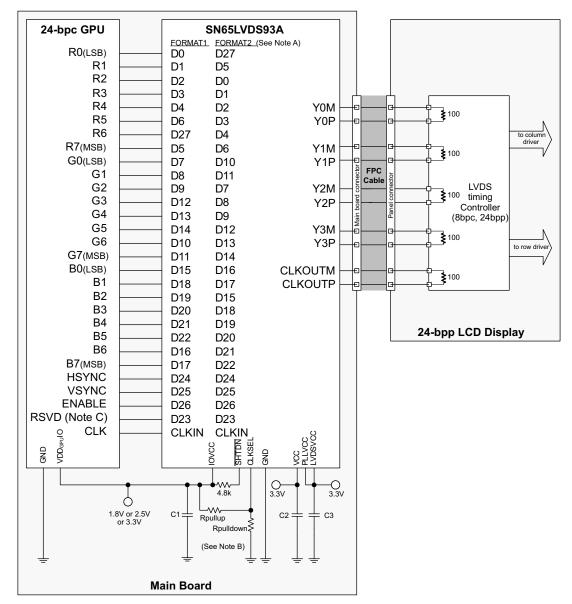
DESIGN PARAMETER	EXAMPLE VALUE
VCC	3.3 V
VCCIO	1.8 V
CLKIN	Falling edge
SHTDN#	High
Format	18-bit GPU to 24-bit LCD

10.2.2 Detailed Design Procedure

10.2.2.1 Signal Connectivity

While there is no formal industry standardized specification for the input interface of LVDS LCD panels, the industry has aligned over the years on a certain data format (bit order). Figure 15 through Figure 18 show how each signal should be connected from the graphic source through the SN65LVDS93A input, output and LVDS LCD panel input. Detailed notes are provided with each figure.





Note A. **FORMAT**: The majority of 24-bit LCD display panels require the two most significant bits (2 MSB) of each color to be transferred over the 4th serial data output Y3. A few 24-bit LCD display panels require the two LSBs of each color to be transmitted over the Y3 output. The system designer needs to verify which format is expected by checking the LCD display data sheet.

- Format 1: use with displays expecting the 2 MSB to be transmitted over the 4th data channel Y3. This is the dominate data format for LCD panels.
- Format 2: use with displays expecting the 2 LSB to be transmitted over the 4th data channel.

Note B. **Rpullup**: install only to use rising edge triggered clocking.

Rpulldown: install only to use falling edge triggered clocking.

- C1: decoupling capacitor for the VDDIO supply; install at least 1x0.01 $\mu\text{F}.$
- C2: decoupling capacitor for the VDD supply; install at least 1x0.1μF and 1x0.01μF.
- C3: decoupling capacitor for the VDDPLL and VDDLVDS supply; install at least 1x0.1μF and 1x0.01μF.

Note C. If RSVD is not driven to a valid logic level, then an external connection to GND is recommended.

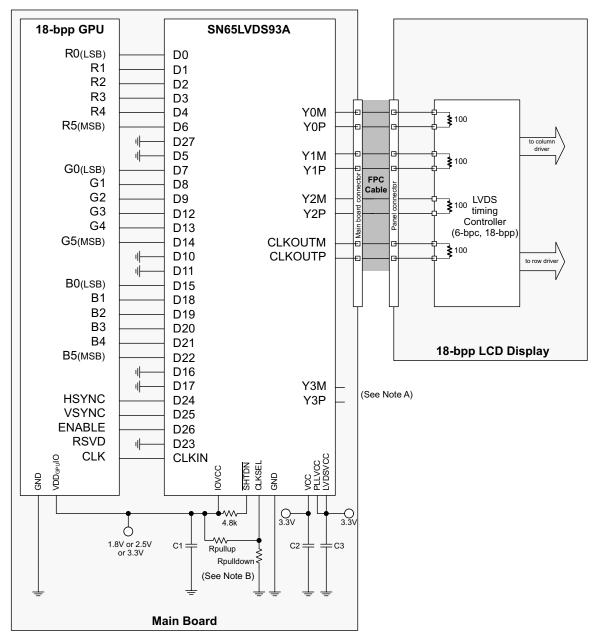
Note D. RSVD must be driven to a valid logic level. All unused SN65LVDS93A inputs must be tied to a valid logic level.

Figure 15. 24-Bit Color Host to 24-Bit LCD Panel Application

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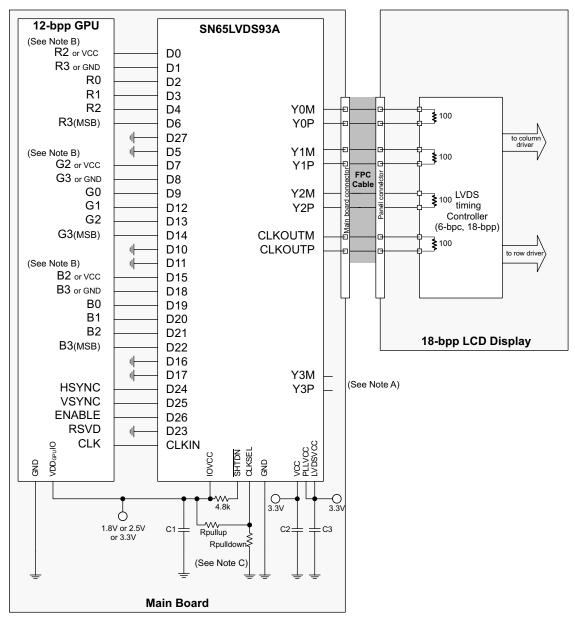
Note A. Leave output Y3 NC.

Note B.Rpullup: install only to use rising edge triggered clocking. **Rpulldown**: install only to use falling edge triggered clocking.

- C1: decoupling capacitor for the VDDIO supply; install at least $1x0.01\mu F$.
- C2: decoupling capacitor for the VDD supply; install at least 1x0.1μF and 1x0.01μF.
- C3: decoupling capacitor for the VDDPLL and VDDLVDS supply; install at least 1x0.1μF and 1x0.01μF.

Figure 16. 18-Bit Color Host to 18-Bit Color LCD Panel Display Application





Note A. Leave output Y3 N.C.

Note B. **R3**, **G3**, **B3**: this MSB of each color also connects to the 5th bit of each color for increased dynamic range of the entire color space at the expense of nonlinear step sizes between each step. For linear steps with less dynamic range, connect D1, D8, and D18 to GND.

R2, **G2**, **B2**: these outputs also connects to the LSB of each color for increased, dynamic range of the entire color space at the expense of nonlinear step sizes between each step. For linear steps with less dynamic range, connect D0, D7, and D15 to VCC.

Note C.Rpullup: install only to use rising edge triggered clocking.

Rpulldown: install only to use falling edge triggered clocking.

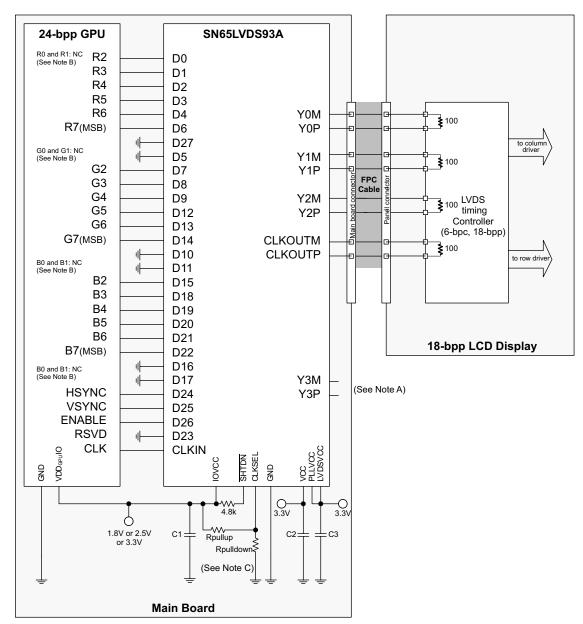
- C1: decoupling capacitor for the VDDIO supply; install at least 1x0.01μF.
- C2: decoupling capacitor for the VDD supply; install at least 1x0.1μF and 1x0.01μF.
- C3: decoupling capacitor for the VDDPLL and VDDLVDS supply; install at least 1x0.1µF and 1x0.01µF.

Figure 17. 12-Bit Color Host to 18-Bit Color LCD Panel Display Application

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Note A. Leave output Y3 NC.

Note B. R0, R1, G0, G1, B0, B1: For improved image quality, the GPU should dither the 24-bit output pixel down to18-bit per pixel.

NoteC.Rpullup: install only to use rising edge triggered clocking.

Rpulldown: install only to use falling edge triggered clocking.

- C1: decoupling capacitor for the VDDIO supply; install at least 1x0.01μF.
- C2: decoupling capacitor for the VDD supply; install at least 1x0.1μF and 1x0.01μF.
- C3: decoupling capacitor for the VDDPLL and VDDLVDS supply; install at least $1x0.1\mu F$ and $1x0.01\mu F$.

Figure 18. 24-Bit Color Host to 18-Bit Color LCD Panel Display Application



10.2.2.2 PCB Routing

Figure 19 shows a possible breakout of the data input and output signals on two layers of a printed-circuit-board.

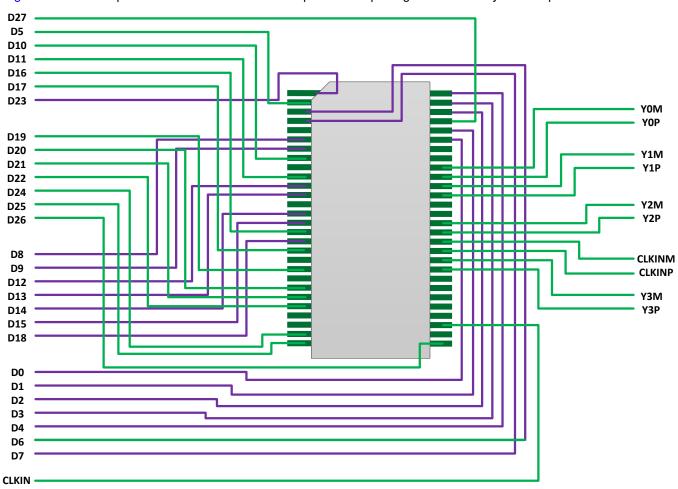


Figure 19. Printed-Circuit-Board Routing Example (See Figure 14 for the Schematic)

10.2.3 Application Curve

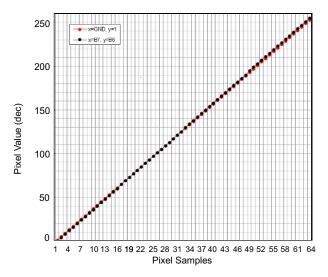


Figure 20. 18b GPU to 24b LCD

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11 Power Supply Recommendations

Power supply PLL, IO, and LVDS pins must be uncoupled from each.

12 Layout

12.1 Layout Guidelines

12.1.1 Board Stackup

There is no fundamental information about how many layers should be used and how the board stackup should look. Again, the easiest way the get good results is to use the design from the EVMs of TI. The magazine *Elektronik Praxis* has published an article with an analysis of different board stackups. These are listed in Table 4. Generally, the use of microstrip traces needs at least two layers, whereas one of them must be a GND plane. Better is the use of a 4-layer PCB, with a GND and a VCC plane and two signal layers. If the circuit is complex and signals must be routed as stripline, because of propagation delay and/or characteristic impedance, a 6-layer stackup should be used.

MODEL 1 MODEL 2 MODEL 3 **MODEL 4** Layer 1 SIG SIG SIG **GND** Layer 2 SIG **GND GND** SIG VCC VCC VCC Layer 3 SIG **GND** SIG VCC SIG Layer 4 Bad Decoupling Good Good Bad **EMC** Bad Bad Bad Bad Bad Signal Integrity Bad Bad Good Self Disturbance Satisfaction Satisfaction Satisfaction High

Table 4. Possible Board Stackup on a Four-Layer PCB

12.1.2 Power and Ground Planes

A complete ground plane in high-speed design is essential. Additionally, a complete power plane is recommended as well. In a complex system, several regulated voltages can be present. The best solution is for every voltage to have its own layer and its own ground plane. But this would result in a huge number of layers just for ground and supply voltages. What are the alternatives? Split the ground planes and the power planes? In a mixed-signal design, for example, using data converters, the manufacturer often recommends splitting the analog ground and the digital ground to avoid noise coupling between the digital part and the sensitive analog part. Take care when using split ground planes because:

- Split ground planes act as slot antennas and radiate.
- A routed trace over a gap creates large loop areas, because the return current cannot flow beside the signal, and the signal can induce noise into the nonrelated reference plane (Figure 21).
- With a proper signal routing, crosstalk also can arise in the return current path due to discontinuities in the ground plane. Always take care of the return current (Figure 22).

For Figure 22, do not route a signal referenced to digital ground over analog ground and vice versa. The return current cannot take the direct way along the signal trace and so a loop area occurs. Furthermore, the signal induces noise, due to crosstalk (dotted red line) into the analog ground plane.



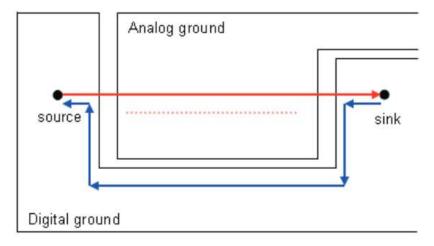


Figure 21. Loop Area and Crosstalk Due to Poor Signal Routing and Ground Splitting

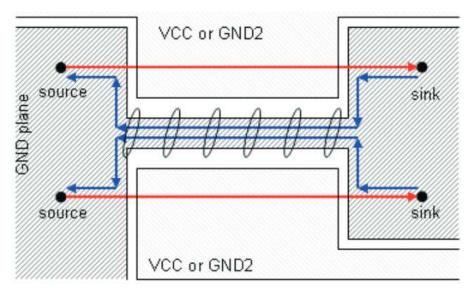


Figure 22. Crosstalk Induced by the Return Current Path

12.1.3 Traces, Vias, and Other PCB Components

A right angle in a trace can cause more radiation. The capacitance increases in the region of the corner, and the characteristic impedance changes. This impedance change causes reflections.

- Avoid right-angle bends in a trace and try to route them at least with two 45° corners. To minimize any impedance change, the best routing would be a round bend (see Figure 23).
- Separate high-speed signals (for example, clock signals) from low-speed signals and digital from analog signals; again, placement is important.
- To minimize crosstalk not only between two signals on one layer but also between adjacent layers, route them with 90° to each other.

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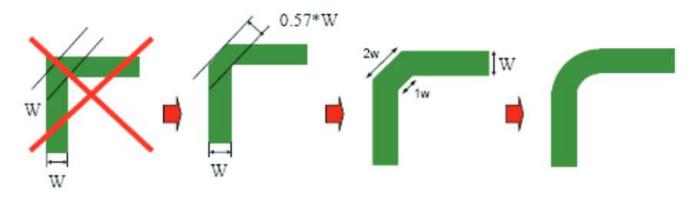


Figure 23. Poor and Good Right-Angle Bends

12.2 Layout Example

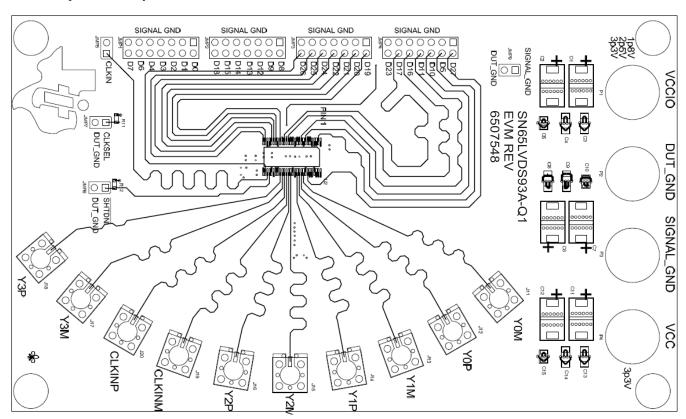


Figure 24. SN65LVDS93A EVM Top Layer - TSSOP Package



Layout Example (continued)

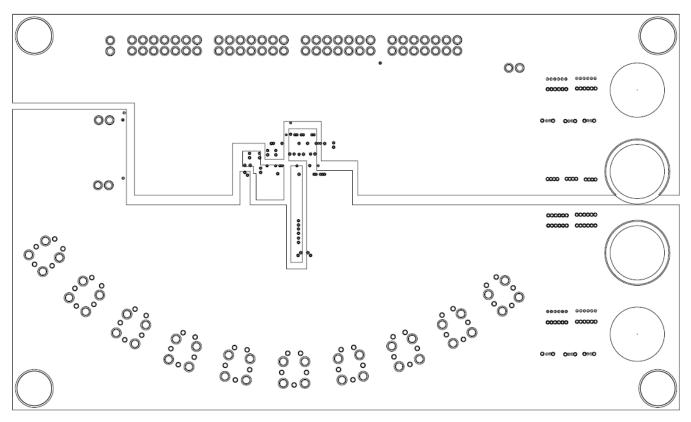


Figure 25. SN65LVDS93A EVM VCC Layer – TSSOP Package

30



13 Device and Documentation Support

13.1 Documentation Support

13.1.1 Related Documentation

For related documentation see the following: LVDS SerDes Receiver, SLLS928

13.2 Trademarks

OMAP, DaVinci, FlatLink are trademarks of Texas Instruments. All other trademarks are the property of their respective owners.

13.3 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

13.4 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



PACKAGE OPTION ADDENDUM

15-Jan-2021

PACKAGING INFORMATION

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Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
SN65LVDS93ADGG	ACTIVE	TSSOP	DGG	56	35	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	LVDS93A	Samples
SN65LVDS93ADGGR	ACTIVE	TSSOP	DGG	56	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	LVDS93A	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.





15-Jan-2021

OTHER QUALIFIED VERSIONS OF SN65LVDS93A:

Automotive: SN65LVDS93A-Q1

NOTE: Qualified Version Definitions:

• Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

PACKAGE MATERIALS INFORMATION

www.ti.com 5-Jan-2022

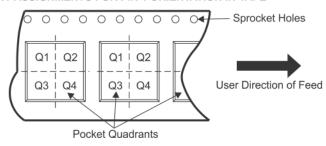
TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

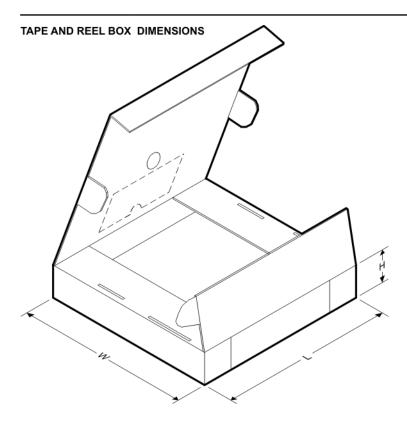
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
SN65LVDS93ADGGR	TSSOP	DGG	56	2000	330.0	24.4	8.6	15.6	1.8	12.0	24.0	Q1

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)	
SN65LVDS93ADGGR	TSSOP	DGG	56	2000	367.0	367.0	45.0	

PACKAGE MATERIALS INFORMATION

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TUBE

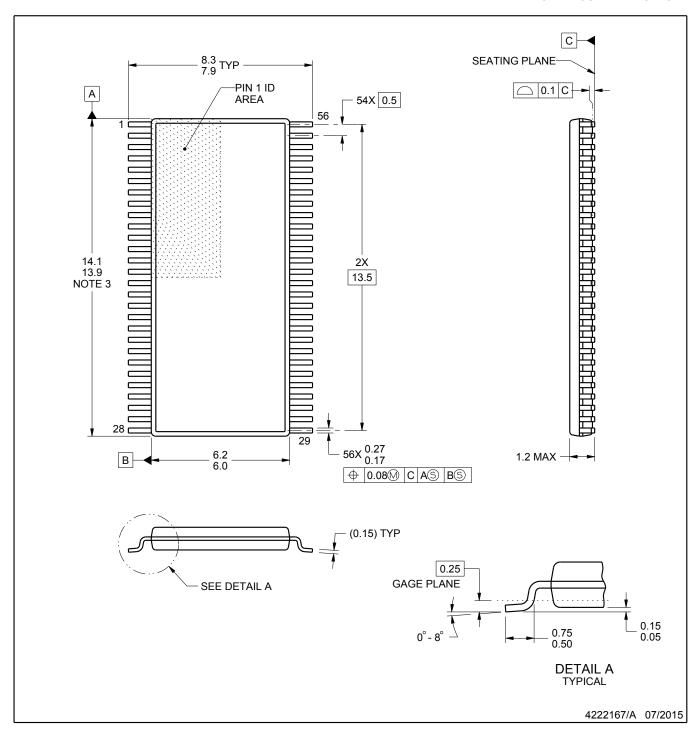


*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
SN65LVDS93ADGG	DGG	TSSOP	56	35	530	11.89	3600	4.9



SMALL OUTLINE PACKAGE



NOTES:

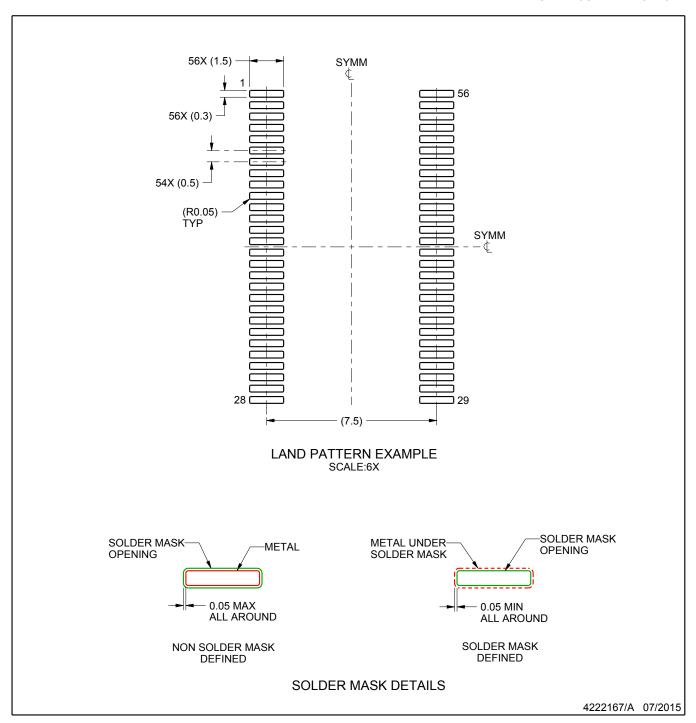
- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

 2. This drawing is subject to change without notice.

 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
 4. Reference JEDEC registration MO-153.



SMALL OUTLINE PACKAGE

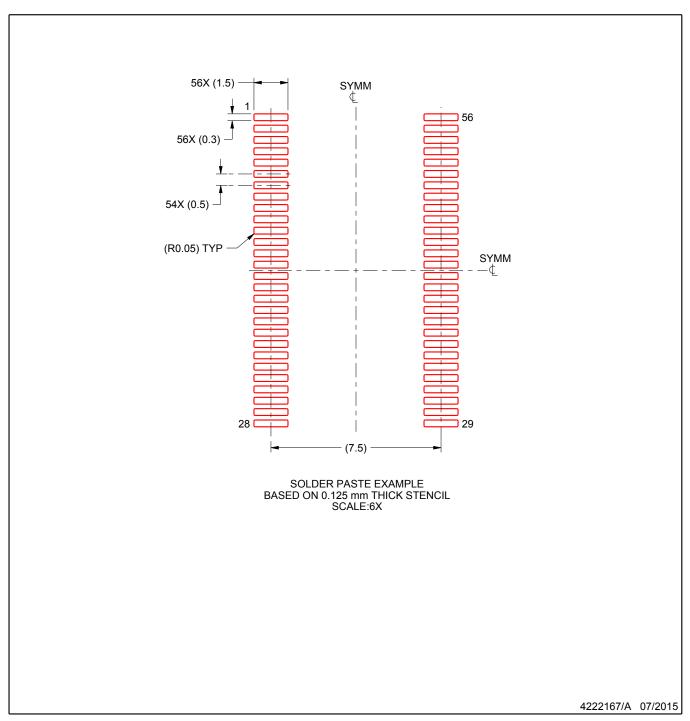


NOTES: (continued)

- 5. Publication IPC-7351 may have alternate designs.
- 6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE PACKAGE



NOTES: (continued)

- Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 8. Board assembly site may have different recommendations for stencil design.



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