General Description

The MAX7651/MAX7652 are complete 12-bit data-acquisition systems featuring an algorithmic, switched-capacitor, analog-to-digital converter (ADC), a pulsewidth-modulated digital-to-analog converter (DAC), three timer/counters, and an industry-standard 8051 microprocessor core with a variety of I/O peripherals. Powerdown capability and full functionality with supply voltages as low as +3V make the MAX7651/MAX7652 suitable for portable and power-sensitive applications.

The MAX7651/MAX7652 perform fully differential voltage measurements with 12-bit resolution, programmable gain, and separate track-and-hold for both positive and negative inputs. The converter accepts versatile input modes consisting of four 2-channel signal pairs or eight 1-channel signals relative to a floating common.

The MAX7651/MAX7652 microprocessor systems feature a CPU, 256 bytes of RAM, two 8kB flash memory, four 8-bit I/O ports, two UARTs, an interrupt controller, and a watchdog timer. Only four clock cycles are required to complete each microprocessor instruction.

The MAX7651/MAX7652 are available in 64-pin TQFP packages.

Applications

Hand-Held Instruments Portable Data-Acquisition Systems Temperature Controllers Smart Transmitters Data Loggers Multi-Channel Data-Acquisition with Data Formatting

Features

- ♦ **12-Bit 53ksps ADC with Fully Differential Inputs**
- ♦ **Dual 8-Bit PWM DAC Outputs**
- ♦ **Three Timers**
- ♦ **4-Clock Cycle 8051-Compatible Instruction Set with Dual Data Pointers**
- ♦ **Programmable Watchdog Supervisor**
- ♦ **Four Parallel I/O Ports**
- ♦ **Dual Serial I/O Ports (up to 375kb)**
- ♦ **+3V or +5V Single-Supply Operation**
- ♦ **DC to 12MHz Clock Speed**
- ♦ **64-Pin TQFP Package**

Ordering Information

Pin configuration appears at end of data sheet.

Functional Diagram

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For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

ABSOLUTE MAXIMUM RATINGS

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(MAX7651 AVDD = VPWMV = DVD = VREF_ + = +4.5V to +5.5V, VREF_ = 0, fXTAL = 12MHz. MAX7652 AVDD = VPWMV = DVD = +2.7V$ to +3.6V, VREF+ = +2.5V, VREF- = 0, ACOM = AVDD/2, fxTAL = 12MHz. TA = TMIN to TMAX, unless otherwise noted. Typical values are at $TA = +25^{\circ}C$.)

ELECTRICAL CHARACTERISTICS (continued)

 $(MAX7651 AVDD = VPWMV = DVD = VREF_ + = +4.5V$ to $+5.5V$, $VREF = 0$, $fXTAL = 12MHz$. $MAX7652 AVDD = VPWMV = DVD = +2.7V$ to +3.6V, VREF+ = +2.5V, VREF- = 0, ACOM = AvDD/2, fxTAL = 12MHz. TA = TMIN to TMAX, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

ELECTRICAL CHARACTERISTICS (continued)

(MAX7651: AVDD = VPWMV = DVDD = VREF+ = +4.5V to +5.5V, VREF- = 0, fXTAL = 12MHz. MAX7652: AVDD = VPWMV = DVDD = +2.7V to +3.6V, VREF+ = +2.5V, VREF- = 0, ACOM = AVDD/2, fXTAL = 12MHz. TA = TMIN to TMAX, unless otherwise noted. Typical values are at TA = +25°C.)

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TIMING CHARACTERISTICS

 $(MAX7651: AVDD = VPWMV = DVD = VREF + = +4.5$ to $+5.5V$, $VREF = 0$, $f_{XTAL} = 12MHz$. $MAX7652$: $AV_{DD} = V_{PWMV} = DV_{DD}$ +2.7V to +3.6V, VREF+ = +2.5V, VREF- = 0, ACOM = AVDD/2, fXTAL = 12MHz. TA = TMIN to TMAX, unless otherwise noted. Typical values are at $TA = +25^{\circ}C$.) (Figure 3)

TIMING CHARACTERISTICS (continued)

 $(MAX7651: AVDD = VPWMV = DVD = VREF + = +4.5$ to +5.5V, VREF- = 0, f_{XTAL} = 12MHz. MAX7652: AV_{DD} = V_{PWMV} = DV_{DD} = $+2.7V$ to $+3.6V$, V_{REF+} = $+2.5V$, V_{REF-} = 0, ACOM = AV_{DD}/2, f_{XTAL} = 12MHz. T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at $TA = +25^{\circ}C$.) (Figure 3)

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TIMING CHARACTERISTICS (continued)

 $(MAX7651: AVDD = VPWMV = DVD = VREF_ + = +4.5$ to $+5.5V$, $VREF_ = 0$, $f_{XTAL} = 12MHz$. $MAX7652: AV_{DD} = V_{PWMV} = DV_{DD}$ +2.7V to +3.6V, VREF+ = +2.5V, VREF- = 0, ACOM = AVDD/2, fXTAL = 12MHz. TA = TMIN to TMAX, unless otherwise noted. Typical values are at $TA = +25^{\circ}C$.) (Figure 3)

Note 1: Relative accuracy is the deviation of the analog value at any code from its theoretical value after the offset and gain errors have been nullified.

Note 2: A_{VDD} = +5.0V, (V_{REF}+) - (V_{REF}-) = +5.0V or A_{VDD} = +3.0V, (V_{REF}+) - (V_{REF}-) = +2.5V.

Note 3: Ground at "ON" channel; 10kHz sine-wave applied to all "off" channels.

Note 4: ALE and PSEN are in reset cycle.

Note 5: All digital inputs are at DGND or DV_{DD}. fx_{TAL} = 0.

Note 6: Table 1. Data Memory Stretch Values.

Note 7: The minimum frequency when writing to the internal flash is 4MHz.

Table 1. Data Memory Stretch Values

Table 2. External Flash Programming Modes

Note 1: To program the lock bits, ALE must be low for duration of "Write Lockbit" cycle.

Note 2: INTO and INTT are open-drain and must either be driven or require a pullup (typically 10kΩ) to DV_{DD}.

Figure 1. FLASH External Mass Erase Waveforms

Figure 2. FLASH External Programming and Verification Waveforms

Figure 3a. External Program Memory Read Cycle

MAX7651/MAX7652

Figure 3b. External Data Memory Read Cycle

Figure 3c. External Program Memory Write Cycle

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MAX7651/MAX7652

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$MAX7652: AVDD = VPWMV = DVD = VREF_+ = 3.0V, VREF_-= 0, V_{COM} = AVDD/2, f_{XTAL} = 12MHz.$ ANALOG SUPPLY CURRENT PWM OUTPUT HIGH vs. INPUT VOLTAGE vs. SOURCE CURRENT PWM OUTPUT LOW vs. SINK CURRENT 3.000 5 250 MAX7651 toc10 MAX7651 toc11 ANALOG SUPPLY CURRENT (mA) ANALOG SUPPLY CURRENT (mA) 4 200 $AV_{DD} = 3V$ 2.950 OUTPUT LOW (mV) PWM OUTPUT LOW (mV) PWM OUTPUT HIGH (V) 3 150 $AV_{DD} = 5V$ 2.900 $AV_{DD} = 3V$ 100 2 **MWd** 2.850 50 1 θ 2.800 $\boldsymbol{0}$ 0 21 3 4 5 2.5 3.53.0 4.0 4.5 5.0 5.5 0 21 3 4 5 SOURCE CURRENT (mA) INPUT VOLTAGE (V) SINK CURRENT (mA) ANALOG SUPPLY CURRENT DIGITAL SUPPLY CURRENT DIGITAL SUPPLY CURRENT vs. TEMPERATURE vs.TEMPERATURE vs. CLOCK FREQUENCY 2.5 5.0 11.75 MAX7651 toc13 MAX7651 toc14 $AV_{DD} = 5V$ 4.5 ANALOG SUPPLY CURRENT (mA) ANALOG SUPPLY CURRENT (mA) OIGITAL SUPPLY CURRENT (mA) 2.4 DIGITAL SUPPLY CURRENT (mA) DIGITAL SUPPLY CURRENT (mA) 4.0 11.50 3.5 2.3 3.0 11.25 2.5 2.0 2.2 11.00 $AV_{DD} = 3V$ **DIGITAL** 1.5 2.1 10.75 1.0 0.5 10.50 0 2.0 0 1 2 3 4 5 6 7 8 9 10 11 12 13 -40 0-20 20 40 60 80 85 -40 10-15 35 60 85 TEMPERATURE (°C) CLOCK FREQUENCY (MHz) TEMPERATURE (°C) IDLE-MODE SUPPLY CURRENT POWER-DOWN CURRENT vs. INPUT VOLTAGE vs. INPUT VOLTAGE 14 5 MAX7651 toc17 MAX7651 toc16 12 IDLE-MODE SUPPLY CURRENT (mA) IDLE-MODE SUPPLY CURRENT (mA) 4 POWER-DOWN CURRENT (µA) POWER-DOWN CURRENT (µA)10 3 8 6 2 4 1 2 $\boldsymbol{0}$ $\boldsymbol{0}$ 3.0 4.03.5 4.5 5.0 5.5 3.0 4.03.5 4.5 5.0 5.5 INPUT VOLTAGE (V) INPUT VOLTAGE (V)

Typical Operating Characteristics (continued)

 $(T_A = +25^{\circ}C$, unless otherwise noted.) $MAX7651: AVDD = VpWMV = DVD = VREF_{+} = 5.0V, VREF_{-} = 0, VCOM = AVDD/2, fXTAL = 12MHz.$

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MAX7651/MAX7652

MAX7651/MAX7652

MAX7651 toc12

MAX7651 toc15

Pin Description

Pin Description (continued)

Pin Description (continued)

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Detailed Description

MAX7651/MAX7652 Architecture

The MAX7651/MAX7652 are complete 12-bit dataacquisition systems featuring an algorithmic, switchedcapacitor, analog-to-digital converter (ADC), dual pulse-width-modulated digital-to-analog converter (DAC), and an industry-standard 8051 microprocessor core with a variety of I/O and timing peripherals.

Using an external oscillator with an operating frequency between 1MHz and 12MHz, the MAX7651/MAX7652 execute the majority of its commands in only four clock periods to yield an average speed improvement of 2.5 times over typical 8051 microprocessors requiring 12 clock periods instructions. See the MAX7651/MAX7652 Programmer's Reference Manual for further details.

On-chip peripherals include four 8-bit parallel ports, two serial ports, three general-purpose timers, and a watchdog timer. The MAX7651/MAX7652 also feature 16kB in two banks of 8kB flash memory and 256 bytes of high-speed random access memory.

Memory Organization

The MAX7651/MAX7652 support up to 64kB of external program (read-only) memory and data (randomaccess) memory in conformance with the 8051 industry standard.

EXTERNAL

FFFFH

Figure 4 shows the program memory organization. When \overline{EA} is high, the CPU has access to two internal 8kB blocks of flash memory beginning at addresses 0000H (lower block) and 2000H (upper block). Addresses 0000H–0002H and 0003H–006AH of the lower block are reserved for the CPU reset vector and a set of interrupt vectors, respectively (see Table 3). Addresses 3FC0H–3FFFH of the upper block are also reserved and cannot be accessed by the CPU. Addresses 4000H–FFFFH are for external ROM. When EA is low, the external ROM must be used for all program addresses (0000H–FFFFH).

Figure 5 shows the data memory (RAM) organization. The first 256 bytes are partitioned between two internal 128-byte blocks. The lower block (addresses 0000H– 007FH) is used for registers or scratchpad memory and can be accessed either directly or indirectly (see the MAX7651/MAX7652 Programmer's Reference Manual). The upper block (addresses 0080H–00FFH) reflects a set of special function registers (SFRs) when accessed directly, and separate scratchpad memory when accessed indirectly. Addresses 0100H–FFFFH are reserved for external RAM.

Table 4 shows the SFR mapping to memory and Table 5 shows the SFR contents on power-up or reset. Unshaded register designations are consistent with the industry standard 8051. Shaded register designations

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Figure 4. Program Memory Organization Figure 5. Data Memory (RAM) Organization

Table 3. Reset and Interrupt Vector Locations

*Lower priority number takes precedence.

Table 4. SFR Memory Organization

Note 1: SFRs in column 0/8 are bit addressable. Other SFRs are not bit addressable.

Note 2: The VERSION SFR contains the silicon ID and will change for future MAX7651/MAX7652 revisions.

Table 5. SFR Contents on Power-Up or Reset

Table 5. SFR Contents on Power-Up or Reset (continued)

are unique to the MAX7651/MAX7652. Subsequent sections of this data sheet explain the SFR functions.

RESERVED SFR addresses are used for MAX7651/ MAX7652 testing and should not be accessed by user software. Undesignated SFR addresses are not implemented and will return indefinite data when read.

Special Function Registers for Microprocessor Operations and Control

Accumulator SFR

The Accumulator SFR is used for arithmetic operations including addition, subtraction, multiplication, division, and Boolean bit manipulation. Accumulator specific instructions designate the accumulator as "A".

B SFR

The B SFR is used for multiply and divide operations. It is otherwise available as a scratchpad register.

Program Status Word SFR

The PSW or Program Status Word SFR contains bits that indicate the state of the microprocessor CPU. Table 6 shows the individual bit functions.

Stack Pointer SFR

The SP or Stack Pointer SFR contains the "top-of-thestack" address in internal RAM. This address increments before data is stored during PUSH and CALL executions. The default value is 07H after reset, so that the stack begins at 08H.

Dual Data Pointer SFRs

The MAX7651/MAX7652 feature dual data pointers to enhance execution times when moving large blocks of data. All DPTR-related instructions use 16 bits contained at SFR pairs DPH0 and DPL0 or DPH1 and DPL1 to address external data RAM or peripherals. Bit 0 (SEL) within the DPS SFR determines the data pointer. No other bits have significance in this register. When SEL= 0, DPTR instructions use DPH0 and DPL0, when SEL=1, DPTR instructions use DPH1 and DPL1. Program code developed for 8051 platforms that use a single data pointer (DPH0 and DPL0) requires no modification if $SEL = 0$ (the default value).

Power Control SFR

The PCON Power Control SFR provides software control over the power modes. In both IDLE and STOP modes, CPU processing is suspended and internal registers maintain their current data. The STOP mode additionally disables the internal clock and analog circuitry. Any enabled CPU interrupt can be used to terminate the IDLE mode. A reset is necessary to terminate the STOP mode and is sufficient to terminate the IDLE mode. Table 7 shows the PCON SFR format.

Instruction Set

The MAX7651/MAX7652 instruction set is compatible with the 8051 industry standard. See the MAX7651/ MAX7652 Programmer's Reference Manual for a complete listing.

Analog-to-Digital Converter

ADC Operation

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Figure 6 shows a simplified model of the converter input structure and the associated switch timing. Once initiated, a voltage conversion requires 224 periods of the external master clock. Capacitor C_{HOLD} charges to the difference between inputs AIN+ and AIN- during eight clock periods of acquisition time that begin on the rising edge of clock cycle 13. This charge sample is subsequently transferred to the ADC (through the action of SW5) during eight clock periods that begin on the rising edge of clock cycle 21. The ADC asserts a conversion complete flag on the rising-edge of clock cycle 225 (see ADC Special Function Registers).

Figure 6. ADC Input Structure and Switch Timing

Since the acquisition time is limited to eight clock periods, the acquired voltage at C_{HOLD} can have significant error if the analog input source impedance (Rs) is large. Limit the worst-case error to 1/2 LSB by ensuring,

Rs < 0.9 tCLK / CHOLD

where t_{CLK} is the clock period. Smaller R_s values may be necessary if an antialiasing filter is used.

The ADC continuously samples the positive and negative difference between the two external reference voltages REF+ and REF- by reconfiguring capacitor CREF over alternate eight clock-period intervals. Switch pairs 1 and 2 are forced off and on, respectively, on the rising edge of clock cycle five to ensure synchronization with conversions. Capacitor C_{HOLD} also charges to the difference between REF+ and REF- on the rising edge of clock cycle 29 and remains charged until the next conversion. Nevertheless, continuous CREF charging requirements dominate loading at the REF+ and REF- inputs.

Analog Inputs

MAX7651/MAX7652

AAX7651/MAX7652

The MAX7651/MAX7652 operate in either single-ended or differential mode. In single-ended mode, one of eight input channels (AIN0–AIN7) is assigned to AIN+, and ACOM is assigned to AIN- (see Figure 6). In differential mode, the eight input channels are assigned to AIN+ and AIN- with four distinct pairings. Table 6 shows the input assignments for different values of bits M3, M2, M1, and M0 in the A/D Control SFR (see ADC Special Function Registers).

Analog Input Protection

Internal protection diodes clamp the analog inputs to AVDD and AGND, so channels can swing within AGND -

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Transfer Function

0.3V and AV_{DD} + 0.3V without damage. For accurate conversions the inputs should not extend beyond the supply rails.

ferential mode reflects voltage increments of VREF/4096 and 2VREF/4096, respectively.

ADC Special Function Registers

The ADCON or A/D Control SFR establishes ADC operating conditions and input configurations. Table 7 shows the individual bit functions. A "write" to ADCON initiates the A/D conversion process.

Figure 7 shows the bipolar two's complement ADC transfer function. The single-ended conversion range extends from -VREF/2 to +VREF/2, where VREF = VREF+ - VREF-. The differential conversion range extends from -VREF to +VREF. Each LSB in the single-ended and dif-

Table 6. Program Status Word (PSW) Format

Table 7. Power Control (PCON) Format

External Reference

The MAX7651/MAX7652 require external reference voltages at VREF+ and VREF-. A single reference voltage can be used at VREF+, when VREF- is connected to AGND. The positive reference voltages must be no greater than the analog supply voltage AV_{DD} and capable of supplying 30µA. Bypass each reference voltage to AGND with a 0.1µF capacitor in parallel with a 10µF low ESR capacitor.

Figure 7a. Single-Ended Mode Transfer Function

Table 8. Analog Input Selection

PWM Digital-to-Analog Converters (DACs)

The MAX7651/MAX7652 provide two pulse-width modulated (PWM) DACs for applications that do not require high conversion accuracy. Figure 8 shows the pulsewidth-modulator block diagram. The clock signal is divided by 2 $(x + 1)$, where x is the content of the Pulse-Width Prescaler (PWPS) SFR register. This reduced frequency signal is used to drive a modulo-255 counter. When the counter value exceeds the value stored in SFRs PWDA (Output A) or PWDB

Figure 7b. Differential Mode Transfer Function

MAX7651/MAX7652

MAX7651/MAX7652

Figure 8. PWM Block Diagram

(Output B), the corresponding output transitions from low to high (Figure 9).

Writing 00H to PWDA or PWDB, yields a waveform with 100% duty cycle (High), and writing FFH to PWDA or PWDB yields a waveform with 0% duty cycle (Low). Writing an intermediate register value y, yields a waveform with duty cycle $(1 - y / 255) \times 100\%$. Tables 10, 11, and 12 show the formats of the PWPS, PWDA, and PWDB SFR's.

External low-pass filters are needed to obtain DC voltages between 0 and DV_{DD} from the PWM outputs. Simple RC filters are preferred. Choose R >2kΩ to avoid excessive loading, and choose C <0.1µF to avoid large transient currents that reflect the PWM switching action. Each filtered PWM output can source or sink up to 2mA. Do not exceed this specification. If larger output capability is required, provide an appropriate buffer such as a unity-gain op amp. PWM circuitry and PWM Outputs A and B are enabled with the Pulse-Width Modulator Control (PWMC) SFR. Table 13 shows the PWMC SFR format.

Watchdog Timer

The MAX7651/MAX7652 features a watchdog timer that resolves irregular software control. The watchdog timer resets the microprocessor if software fails to reset the timer within one of four pre-selected time intervals. The timer generates an optional interrupt after 216, 219, 222, or 225 clock periods of the external oscillator. It generates the reset signal after an additional 512 clock periods. Table 14 indicates specific interrupt and reset times that apply for a 12MHz clock frequency.

Figure 9. PWM Output Waveform

Five watchdog-related control bits and two status flags are located in different special function registers. Table 15 shows the particular functions and SFR locations.

8051-Compatible Peripherals

Parallel I/O Ports

Like other 8051-based systems, the MAX7651/ MAX7652 features four 8-bit parallel ports that support general input and output, address and data lines, and various special functions. Each bidirectional port has a latch register (SFRs P0, P1, P2, and P3), an input buffer, and an output driver.

Port P0 is open-drain. Writing a logic level 1 to a P0 pin establishes a high-impedance input. When used as a general-purpose output, a P0 pin requires an external pull-up resistor to validate a logic level 1. When used as an address/data output, a P0 pin features an internal active high driver. Port 0 is a bidirectional Flash data I/O port during Flash programming and verification.

Port 1: Port 1 is a bidirectional I/O port with internal pullups. Port 1 pins that have 1's written to them are pulled high by the internal pullups and can serve as inputs. Port 1 receives low-order address bytes during Flash programming and verification.

Port 2: Port 2 is a bidirectional I/O port with internal pullups. Port 2 pins that have 1's written to them are pulled high by the internal pullups and can serve as inputs. Port 2 also serves as the high-order address and data bus (for 16-bit operations) during accesses to external memory, using strong internal pullups when emitting 1's.

Port 3: Port 3 is a bidirectional I/O port with internal pullups. Port 3 pins that have 1's written to them are pulled high by the internal pullups and can serve as inputs.

The P1 and P3 ports support the special functions listed in Table 16. Write a "1" to the corresponding bit in the port register to enable the alternative function.

Table 9. A/D Control (ADCON) Format—SFR Address C5H

Note: SFRs ADDAT1 and ADDAT0 contain the results of individual A/D conversions with the formats shown in Tables 8 and 9. A read to ADDAT1 clears the CC and OVRN flags in ADCON.

Table 10. A/D Data-1 (ADDAT1) Format—SFR Address C3H

Table 11. A/D Data-0 (ADDAT0) Format—SFR Address C2H

Table 12. Pulse-Width Prescaler (PWPS) Format—SFR address DAH

Table 13. Pulse-Width Data A (PWDA) Format—SFR address DBH

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Table 14. Pulse-Width Data B (PWDB) Format—SFR address DCH

Table 15. Pulse-Width-Modulator Control (PWMC) Format—SFR Address FEH

Table 16. Watchdog Interrupt and Reset Times (fCK = 12MHz)

Serial Interface Ports

The MAX7651/MAX7652 each have two serial interfaces that operate according to the 8051 industry standard. Serial Port 0 uses SFRs SCON0 and SBUF0 for control and buffer functions. Serial Port 1 uses SFRs SCON1 and SBUF1 with identical bit functionality. See the MAX7651/MAX7652 Programmer's Reference Manual for details concerning serial-port data operations and timing information.

Timers/Counters

The MAX7651/MAX7652 have three timer/counters that function in several different modes for applications such as UART baud-rate control. All three timer/counters operate according to the 8051 industry standard. Specifically, the control (TCON), mode (TMOD), timer-0 parameter (TL0, TH0), Timer1 parameter (TL1, TH1), and Timer-2 parameter (TL2, TH2, RCAP2L, RCAP2H) SFRs have conventional formats. See the MAX7651/ MAX7652 Programmer's Reference Manual for information concerning timer/counter applications.

Crystal Oscillator

The MAX7651/MAX7652 each have a single-stage inverter (Input at XTAL1, Output at XTAL2) that supports a crystal controlled oscillator. The crystal oscillator frequency should be between 1 and 12 MHz.

Note: External flash memory programming requires a minimum crystal oscillator frequency of 4MHz.

Crystal Specification:

An external oscillator can also be used to clock the MAX7651/MAX7652 at frequencies between 1 and 12MHz, provided that the duty cycle is between 40% and 60%. When using an external clock source connect the clock to XTAL1, with XTAL2 unconnected.

Applications Information

Performing a Conversion

An example of a conversion with the MAX7651/ MAX7652 is as follows:

- Write to the ADCON SFR, setting bit CCIE to 1, and bits M3–M0 to appropriate values for the desired differential or single-ended analog input configuration (Tables 6 and 7).
- Wait 224 clock cycles to receive Interrupt 3 as an indication that the A/D conversion is complete.
- Read the conversion data in SFRs ADDAT0 and ADDAT1 as described in Tables 8 and 9.

Using FLASH Memory

The upper and lower 8kB blocks of internal Flash memory are each organized as 128 64-byte pages. Read, write, and page-erase operations cannot be applied to either block while executing program commands from the other block.

Note: Standard MOVC operations are supported.

FLASH Memory Special Function Registers

Tables 17 and 18 show the formats for the EEAH and EEAL SFRs. The EEAH register specifies the applicable Flash memory block (high or low) and the page address within that block. The EEAL register specifies the byte address within the specified page.

Table 19 shows the format for the Flash memory data (EEDAT) SFR that is used for 8-bit read and write transfers from and to a specified address.

Table 20 shows the format for the Flash memory status and command (EESTCMD) SFR. Bits RDYHI and RDYLO are cleared to zero when a read, write, or pageerase operation is applied to the high or low flash memory block. These bits are set to one once the flash

Table 18. Alternate Port Functions

Table 19. Flash Address High (EEAH) Format—SFR Address EBH

Table 20. Flash Address Low (EEAL) Format—SFR Address EAH

memory operation is complete. **Never attempt to execute a flash memory command when either RDYHI or RDYLO are 0 (command action in progress).**

Flash Memory Read

To read Flash memory, load the address into SFRs EEAH and EEAL. Then write AAH to EESTCMD. The results of the read operation will be available in SFR EEDAT in the next CPU instruction cycle.

Flash Memory Write

Erase operations set all bits to "1". After a byte has been programmed it must be erased before it is re-written. To write to Flash memory, load the address into SFRs EEAH and EEAL, and load the data into EEDAT. Then write 55H to EESTCMD. The execution time for flash memory write is 63µs (typ) and is independent of the CPU clock.

Note: Do not write to the same location more than twice before the next page/mass erase operation.

Flash Memory Page Erase

The page erase operation sets all bits within the page to "1"s. To erase a page from Flash memory, load the page address into SFR EEAH, register EEAL is not used. Then write 5AH to EESTCMD. The execution time for page erase is 9.4ms (typ) and is independent of the CPU clock.

Note: Do not attempt to apply read, write, or pageerase operations to the flash memory block in which the CPU is currently executing program instructions.

External Flash Memory Programming

The MAX7651/MAX7652 are normally shipped with the internal Flash memory blocks fully erased (all bits set to 1) and ready for external programming. External write, read (verify), and mass-erase operations are available. Flash memory addresses for either the upper or lower 8-kbyte blocks are specified at Ports 1 and 2.

Before applying any external Flash memory operations, power-up the MAX7651/MAX7652 with RST asserted. ALE, PSEN, and ports P1 –P3 are pulled high with weak resistive pullups. Port P0 requires 10kΩ external pullups. Wait at least 10ms for the oscillator and internal circuitry to stabilize. The program, verify and masserase flash memory programming steps are outlined below.

Note: Failure to follow proper power-up conditions or the specified flash memory programming steps can result in loss of flash data integrity.

External Flash Memory Program (Table 2)

Erase operations. Set all bits to "1". After a byte has been programmed it must be erased before it is re-written.

- 1) Power-up the device with RST asserted and allow ALE and \overline{PSEN} to float to the "1" state (they will be internally pulled-up during RST assertion).
- 2) Wait 10ms for the internal bandgap and oscillator to stabilize.
- 3) Apply the memory location on the address lines at ports 1 and 2.
- 4) Apply data to the data lines at port 0.
- 5) Raise \overline{EA} / Vpp to DV_{DD} and pull PSEN low.
- 6) Set P2.6, P2.7, P3.6, and P3.7 to the levels shown in Table 2.
- 7) Set P2.5 low or high for the lower or higher 8kB Flash memory block.
- 8) Force ALE / PROG low. P3.4 (READY) will go low to indicate a write in progress.
- 9) When P3.4 returns high (write complete after approximately 63µs), set ALE / PROG high.

10)Power-down sequence.

- A) Remove drive from and allow PSEN and ALE/PROG to float high.
- B) Pull EA low.
- C) High-Z all digital pins.
- D) Remove power from all power pins.

Note: Do not write to the same location more than twice before the next page/mass erase operation.

External Flash Memory Verify (Table 2) External Verify:

If lock bits LB1 and LB2 have not been programmed, the programmed flash array(s) can be read back through the address and data lines for verification. The lock bits cannot be verified directly. Verification of the lock bits is achieved by observing that their features are enabled.

External verify (readback) power-up sequence:

- 1) Power-up the MAX7651/MAX7652 with RST asserted, allow ALE and PSEN to float to the "1" state (they will be internally pulled-up during RST assertion). Wait 10ms for the internal bandgap and oscillator to stabilize.
- 2) Pull PSEN LOW, EA HIGH, ALE HIGH, and set P2.6, P2.7, P3.6, P3.7, P2.5, as per Flash Programming Modes (Table 2) for reading either LOWER or UPPER flash memory block.

Note: P2.7 is cycled low/high to perform a FLASH read operation. Minimum low time for P2.7 is ten clock cycles.

External verify power-down sequence:

- 1) Power-down sequence
	- A)Remove drive from and allow PSEN and ALE/ PROG to float high.
	- B) Pull EA low.
	- C) Hi-z all digital pins.
	- D) Remove power from all power pins.

External Flash Memory Mass Erase

A mass erase operation sets all bits, including the lock bits to "1" (Table 22).

External Erase:

Both FLASH arrays can be simultaneously mass-erased electrically by using the proper combination of control signals as shown in Table 2. The erase operation must be executed before either memory can be programmed. Lock bits are also erased (Set to 1).

External chip erase power-up sequence:

- 1) Power-up chip with RST asserted, and allow ALE and PSEN to float to the "1" state (they will be internally pulled-up during RST assertion). Wait 10ms for the internal bandgap and oscillator to stabilize.
- 2) Pull PSEN LOW, EA HIGH, set P2.6, P2.7, P3.6, P3.7, and P2.5, as per Mass Erase mode in the Flash Programming Modes (table 2).
- 3) P3.4 will be LOW during mass erase cycle and return HI at the end of mass erase cycle.

External chip erase power-down sequence:

- 1) Power-down sequence
	- A)Remove drive from and allow PSEN and ALE/ PROG to float high.
	- B) Pull EA low.

C) Hi - z all digital pins.

D) Remove power from all power pins.

Figure 2 shows the timing waveforms that apply for the Flash memory mass erase operation.

Flash Memory Lock Bits

The MAX7651/MAX7652 each contains three lock bits which can be left unprogrammed (logic "1") or can be programmed (logic "0") to obtain the additional features listed in the table below:

When lock bit "1" is programmed (set to logic "0"), the logic level at the EA pin is sampled and latched during RST deassertion. Subsequent changes in logic levels on EA have no effect. If the device is powered-up without a reset (RST), the latch initializes to a random value and holds that value until RST is pulsed high, then low. It is necessary that the latched value of EA be in agreement with the current logic level at that pin in order for the device to function properly.

Signature Bytes

The MAX7651/MAX7652 contain three signature bytes with the information shown in Table 23. Read each byte by following the Flash Memory Read procedure, but set P2.6, P2.7, P3.6, and P3.7 at low. Signature bytes are not affected by mass erase or page erase operations.

Table 21. Flash Memory Data (EEDAT) Format—SFR Address ECH

Table 22. Flash Status and Control (EESTCMD) Format—SFR Address EDH

IVI A XI*IV*I

Interrupt System

The MAX7651/MAX7652 has ten program-assist interrupts that are either external or internal to the 8051 system. Table 24 shows the SFR bit locations for interrupt enable and priority control. Shaded Table regions reflect the 8051 industry standard. Set SFR bit IE.7 high to enable all interrupts. See the MAX7651/MAX7652 Programmer's Reference Manual.

Timers

The MAX7651/MAX7652 feature several modes of timing control through the CKCON special function register. Table 25 shows the CKCON SFR format. The individual control bits can be used to set the number of clock cycles needed (four or twelve) to increment each timer/counter or the number of clock cycles needed to execute the MOVX instruction. See the MAX7651/ MAX7652 Programmer's Reference Manual for further details.

Analog and Digital Supplies

The MAX7651/MAX7652 have multiple power-supply inputs: one analog AV_{DD} and three digital DV_{DD}. The pulse width modulators have their own power supply inputs, PWMV and PWMG. Decouple all supply inputs with a 0.1µF capacitor in parallel with a 10µF low ESR capacitor, with both capacitors as close to the supply pins as possible and with the shortest possible connection to the ground plane.

Table 23

Note: P2.6, P2.7, P3.6, and P3.7 must also meet T_{ASUW} (min) timing specification.

Table 24. Lock Bit Protection Modes

Table 25. MAX7651/MAX7652 Signature Bits

Table 26. MAX7651/MAX7652 Interrupts (Note 1)

Note 1: Shaded areas reflect the 8051 industry standard.

Note 2: Set Enable SFR bit high to enable interrupt.

Note 3: Set Priority SFR bit high to eatablish high priority.

Table 27. CKCON SFR Address 8EH

Power Requirements

MAX7651 operates from +5V while the MAX7652 operates from +3V analog and digital supply voltages. The analog supply current is typically 2mA. The typical digital supply currents (continuous A/D conversions at 12MHz clock frequency) are 5mA and 13mA at +3V and +5V, respectively. Current consumption will vary depending on RAM read/write and flash read/write page erase duty cycle.

Idle Mode

In idle mode, CPU processing is suspended and internal data registers maintain their current data. However, unlike typical 8051 systems, the clock is not disabled internally. Set PCON.0 (IDLE) high to enter the Idle

mode after the instruction is complete. Figure 10 shows the related timing characteristics.

Enable any interrupt to clear PCON.0 and exit the Idle mode (See Figure 11 for the related timing). Assert RST alternately.

Stop Mode

In stop mode, the internal clock and analog circuitry are powered-down. Set PCON.1 (STOP) HIGH to enter the Stop mode after the instruction is complete. Figure 12 shows the related timing characteristics. The only way to exit Stop mode is to assert RST.

Definitions

Integral Nonlinearity (INL)

Integral nonlinearity is the deviation of the values on an actual transfer function from a straight line. This straight line can be either a best straight-line fit or a line drawn between the endpoints of the transfer function, once offset and gain errors have been nullified. The static linearity parameters for the MAX7651/MAX7652 are measured using the best straight-line fit method.

Figure 10. Idle Mode Entry Timing

Figure 11. Idle Mode Exit Timing

Figure 12. Stop Mode Timing

Differential Nonlinearity (DNL)

Differential nonlinearity is the difference between an actual step width and the ideal value of 1LSB. A DNL error specification of less than 1LSB guarantees no missing codes and a monotonic transfer function.

Offset Error

The offset error is the difference between the ideal and the actual offset points. For an ADC, the offset point is the midstep value when the digital output is zero.

Gain Error

The gain or full-scale error is the difference between the ideal and actual gain points on the transfer function, after the offset error has been canceled out. For an ADC the gain point is the midstep value when the digital output is full-scale.

Signal-To-Noise Ratio (SNR)

For a waveform perfectly reconstructed from digital samples, SNR is the ratio of full-scale analog input (RMS value) to the RMS quantization error (residual error). The ideal theoretical minimum analog-to-digital noise is caused by quantization error only and results directly from the ADCs resolution (N Bits):

$$
SNR = (6.02 \times N + 1.76)dB
$$

In reality, there are other noise sources besides quantization noise including thermal noise, reference noise, clock jitter. Therefore, SNR is computed by taking the ratio of the RMS signal to the RMS noise which includes all spectral components minus the fundamental, the first five harmonics, and the DC offset.

Signal-To-Noise Plus Distortion (SINAD)

Signal-To-Noise Plus Distortion is the ratio of the fundamental input frequency's RMS amplitude to RMS equivalent of all other ADC output signals.

 $SINAD$ (dB) = 20 x log (SignalRMS / NoiseRMS)

Effective Number Of Bits (ENOB)

ENOB indicates the global accuracy of an ADC at a specific input frequency and sampling rate. An ideal ADCs error consists of quantization noise only. With an input range equal to the full-scale range of the ADC, calculate the effective number of bits as follows:

$$
ENOB = (SINAD - 1.76) / 6.02
$$

Total Harmonic Distortion (THD)

THD is the ratio of the RMS sum of the first five harmonics of the input signal to the fundamental itself. This is expressed as:

$$
THD = 20 \times \log \left[\frac{\sqrt{(v_2^2 + v_3^2 + v_4^2 + v_5^2)} }{v_1} \right]
$$

where V1 is the fundamental amplitude, and V2 through V5 are the amplitudes of the 2nd- through 5th-order harmonics.

Spurious-Free Dynamic Range (SFDR)

SFDR is the ratio of RMS amplitude of the fundamental maximum signal component to the RMS value of the next largest distortion component.

Chip Information

TRANSISTOR COUNT: 358,000 PROCESS: CMOS

Pin Configuration

Package Information

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

MAX7651/MAX7652

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