

Primary-Side Control LED Driver with High PFC

FEATURES

- Primary Feedback Control Without Secondary feedback Circuit
- High Power Factor >0.9 Over the Universal Input
- High Current Accuracy ±3%
- Boundary Conduction Mode Operation
- Ultra-low Start Up Current
- Cycle-by-cycle Current Limit
- Over-voltage Protection
- Short-circuit Protection
- Under Voltage Lock Out
- Available in a SOP8/DIP8 Package

APPLICATIONS

- Isolated, Solid State Lighting
- Industrial and Commercial Lighting
- Residential Lighting

DESCRIPTION

The FT825 is a primary-side-control offline LED lighting controller which can achieve high power factor and accurate LED current for an isolated lighting application in a single stage converter.

The real current control method can control the LED current accurately from the primary side information. It can significantly simplify the LED lighting system design by eliminating the secondary side feedback components and the opto-coupler.

The FT825 works in boundary conduction mode for reducing the MOSFET switching losses. The extremely low start-up current and the quiescent current can reduce the power consumption thus lead to an excellent efficiency performance.

The multi-protection function of FT825 can greatly enhance the system reliability and safety. The FT825 features over-voltage protection, short-circuit protection, cycle-by-cycle current limit, VCC UVLO protection.

The FT825 is available in SOP8/DIP8 Package.



TYPICAL APPLICATION CIRCUIT

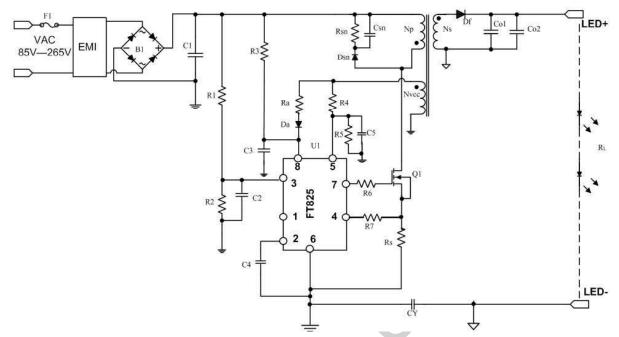


Figure 1: Typical Application Circuit

ABSOLUTE MAXIMUM RATINGS

VCC to GND	0.3V to +27V
Analog Inputs and Outputs	0.3V to 7V
ZCD Pin Maximum Current	5mA (source) / 5mA(sink)
Max. Gate Source Current	1.0A
Max. Gate Sink Current	1.2A
Operating Temperature Range	40 °C to +125 °C
Junction Temperature	40 °C to +150 °C
Storage Temperature Range	60°C to +150°C
ESD Protection HBM	2000V
ESD Protection MM	200V

^{*} Stresses exceed those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Functional operation of the device at conditions beyond those listed in the specification is not guaranteed. Prolonged exposure to extreme conditions may affect device reliability or functionality.



PIN CONFIGURATION

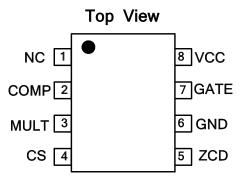


Figure 2: Pin Assignments

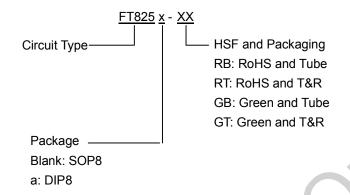
TERMINAL DESCRIPTION

No.	PIN	FUNCTION		
1	NC	This pin can be NC or connects to GND.		
2	СОМР	Loop Compensation pin. Connects a compensation network to stabilize the LED		
	driver and get an accurate LED current of the LED driver.			
3	MULT Main input to the multiplier. This pin is connected to the rectified main			
3	WIOLI	a resistor divider and provides the sinusoidal reference to the current loop.		
		Current sense pin. The MOSFET current is sensed via a resistor, the resulting		
4	cs	voltage compared to the internal sinusoid shaped current reference signal to		
		determine when the MOSFET turns off.		
		Zero current detection pin. A negative going edge triggers the turn on signal of the		
		external MOSFET, connects this pin through a resistor divider from the auxiliary		
5	ZCD	winding to GND. Over-voltage condition is detected through ZCD, if ZCD voltage is		
		higher than the Over-voltage protection (OVP) threshold after a blanking time 1us,		
		the over-voltage condition is detected.		
6	GND	Ground.		
		Gate drive output pin. The totem pole output stage is able to drive high power		
47	GATE	MOSFET with a peak current of 1A source capability and 1.2A sink capability. The		
	GATE	high level voltage of this pin is clamped to 13V to avoid excessive gate drive		
		voltage.		
8	VCC	Supply Voltage of both the signal part of the IC and the gate driver.		

Table1



ORDERING INFORMATION



Package	Temperature Range	HSF	Packaging	Ordering Code
SOP8	-40°C-125°C	RoHS	Tube	FT825-RB
			T&R	FT825-RT
		Green	Tube	FT825-GB
			T&R	FT825-GT
DIP8	-40℃-125℃	RoHS	Tube	FT825a-RB
		Green	Tube	FT825a-GB

Table 2

MARKING RULE

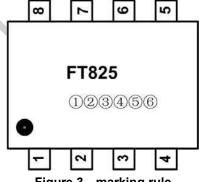
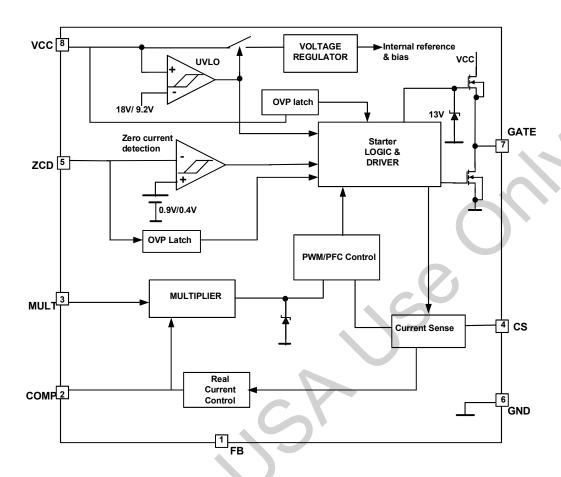


Figure 3 marking rule

123456 for internal reference



BLOCK DIAGRAM





ELECTRICAL CHARACTERISTICS

(Tj = 25°C, VCC = 16V, unless otherwise specified)

Symbol	Parameter	Test Condition	Min	Тур	Max	Unit
SUPPLY VOLTAGE						
VCC	Operating Range	After turn-on	9.5		20	V
VCC _{on}	Turn-on Threshold			17.8		V
VCC _{off}	Turn-off Threshold			9.2		V
VCC _{OVP}	VCC over Voltage			24		V
SUPPLY CURF	RENT					13
Istart-up	Start-up Current	Before turn-on,VCC= 16V		6		uA
Iq	Quiescent Current	After turn-on, No switch		1.3		mA
Icc	Operating Supply Current	@ 70 kHz		1.8		mA
MULTIPLIER II	NPUT			5		
I _{MULT}	Input Bias Current	V _{MULT} = 0 to 3 V			-1	uA
V _{MULT}	Linear Operation Range		0	0 to 3		V
$\triangle V_{CS} / \triangle V_{MULT}$	Output Max Slope	V _{MULT} = 0 to 1V V _{COMP} = Upper clamp		2.4		V/V
К	Gain (1)	$V_{MULT} = 1 V$, $V_{COMP} = 3 V$		0.72		1/V
ERROR AMPL	IFIER				•	
V _{FB}	Voltage Feedback Input Threshold	(3)		0.4		V
V _{EA}	Voltage Gain			400		V/V
G _{EA}	Transconductance			100		100uA/V
	Source Current			45		uA
I _{COMP}	Sink Current			-65		uA
V	Upper Clamp Voltage			5.2		V
V_{COMP}	Lower Clamp Voltage			1.2		V
CURRENT SE	NSE COMPARATOR					
t _{d (H-L)}	Delay to Output			100		ns
V _{CS clamp}	Current Sense Reference Clamp			2.4		V



ELECTRICAL CHARACTERISTICS

(Tj = 25°C, VCC = 16V, unless otherwise specified)

Symbol	Parameter	Test Condition	Min	Тур	Max	Unit		
ZERO CURRENT DETECTOR								
V _{ZCD_T}	Zero Current Detect Threshold	(2)		0.4		٧		
V_{ZCD_Hy}	Zero Current Detect Hysterestic	(2)		0.5		V		
V _{ZCD_OVP}	Output over Voltage Threshold	1us delay after turn-off		5.3		V		
LOGIC								
t _{START}	Start Timer Period			130		us		
t _{LEB}	Leading Edge Blanking Time			300		ns		
T _{off_min}	Minimum off Time			3.5		us		
GATE DRIVER								
Igate-source	Max Source Current			1		Α		
lgate-sink	Max Sink Current			-1.2		Α		
Voclamp	Output Clamp Voltage			12		V		

Table 4

- (1) The multiplier output is given by: Vcs = $K * V_{MULT} * (V_{COMP} 1.2)$
- (2) Parameters guaranteed by design, functionality tested in production.



FUNCTIONAL DESCRIPTION

Under-Voltage Lockout(UVLO)

The turn-on and turn-off threshold voltages are fixed internally at 17.8V and 9.2V respectively for FT825. This hysteresis behavior will guarantee a one shot start-up, as long as a proper start-up resistor and hold-up capacitor are used.

Line Compensation

An offset current proportional to instantaneous mains voltage streams out CS pin. A compensation resistor inverse proportional to the offset current is added between ZCD Pin and transformer auxiliary winding to achieve extremely low THD and line compensation, the compensation resistor can be adjusted externally.

Zero current switching

The IC is switched on when the transformer is demagnetized after MOSFET's minimum off time about 3.5us typically. Internal circuitry connected to the ZCD pin detects the end of the secondary stroke. After MOSFET's minimum off time, if no demagnetization signal is detected on the ZCD pin, the controller generates a startup signal 130us typically after the last GATE signal.

Leading Edge Blanking (LEB)

A turn on spike on CS pin will inevitably appear when the power MOSFET is switched on. At the beginning of each switching pulse, the current-limit comparator is disabled for around 300nsec to avoid premature termination. The gate driver output cannot be switched off during the blanking period. However, it is strongly recommended to add a small RC filter for higher power application to avoid the CS pin from damage by the negative turn-on spike.

Output Driver

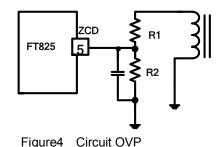
With a low ON-resistance and a high current driving capability, the output driver can easily driver a big MOSFET or IGBT. Cross conduction currents are avoided to minimize heat dissipation, allowing the efficiency and reliability to be improved.

Over Voltage Protection (OVP)

Under steady-state condition, system output voltage is controlled by load connection, the voltage of ZCD pin set by R1, R2 and auxiliary winding (see figure 4) does not exceeds 5.3V, When system output is open, the voltage of output increases, due to the coupling polarity between auxiliary winding and secondary winding of a transformer, ZCD voltage will increase also. Once the ZCD voltage exceeds 5.3V and last for about 1us to guarantee the leakage inductance ringing has been fully damped, OVP of IC is triggered, the IC stops switching, It can be reset by re-starting the voltage on pin VCC.

The OVP voltage can be adjusted by setting R1, R2 and auxiliary winding.





Power Factor Correction

The MULT pin is connected to the tap of the resistor divider from the rectified instantaneous line voltage and fed as one input of the Multiplier. The output of the multiplier will be shaped as sinusoid too. This signal provides the reference for the current comparator and comparing with the primary side inductor current which sets the primary peak current shaped as sinusoid with the input line voltage, High power factor can be achieved. The maximum voltage of the multiplier output to the current comparator is clamped to 2.4V to get a cycle-by-cycle current limitation.

Minimum off time

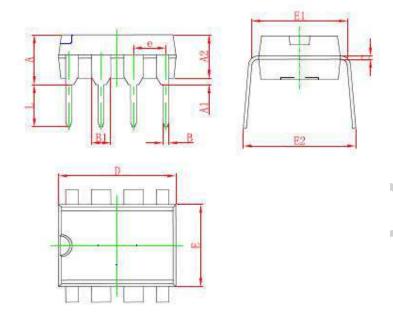
The FT825 operates with variable switching frequency, the frequency is changing with the input instantaneous line voltage. To limit the maximum frequency and get a good EMI performance, FT825 employs an internal minimum off time limiter—3.5µs.

Output Short Circuit Protection

When the output short circuit happens, the positive plateau of auxiliary winding voltage is also near zero, the VCC can not be held on and it will drop below VCC UVLO. The IC will shut down and restart again.



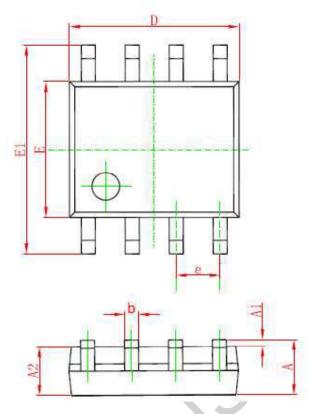
DIP8 PACKAGE OUTLINE DIMENSIONS

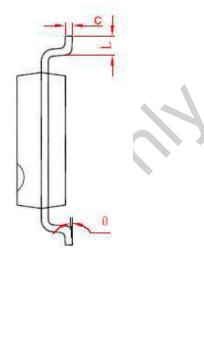


Symbol	Dimensions In Millimeters		Dimensions In Inches		
	Min	Max	Min	Max	
А	3.710	4.310	0.146	0.170	
A1	0.510		0.020		
A2	3.200	3.600	0.126	0.142	
В	0.380	0.570	0.015	0.022	
B1	1.524 (BSC)		0.060 (BSC)		
С	0.204	0.360	0.008	0.014	
D	9.000	9.400	0.354	0.370	
Е	6.200	6.600	0.244	0.260	
E1 .	7.320	7.920	0.288	0.312	
е	2.540 (BSC)		0.100 (BSC)		
L	3.000	3.600	0.118	0.142	
E2	8.400	9.000	0.331	0.354	



SOP8 PACKAGE OUTLINE DIMENSIONS





Symbol	Dimensions In Millimeters		Dimensions In Inches		
Syllibol	Min	Max	Min	Max	
Α	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350	1.550	0.053	0.061	
b	0.330	0.510	0.013	0.020	
С	0.170	0.250	0.006	0.010	
D	4.700	5.100	0.185	0.200	
E	3.800	4.000	0.150	0.157	
E1	5.800	6.200	0.228	0.244	
е	1.270 (BSC)		0.050	(BSC)	
Ĺ	0.400	1.270	0.016	0.050	
θ	0°	8°	0°	8°	



Fremont Micro Devices (SZ) Limited

#5-8, 10/F, Changhong Building, Ke-Ji Nan 12 Road, Nanshan District, Shenzhen

Tel: (86 755) 86117811 Fax: (86 755) 86117810

Fremont Micro Devices (Hong Kong) Limited

#16, 16/F, Blk B, Veristrong Industrial Centre, 34-36 Au Pui Wan Street, Fotan, Shatin, Hong Kong

Tel: (852) 27811186 Fax: (852) 27811144

Fremont Micro Devices (USA), Inc.

42982 Osgood Road Fremont, CA 94539

Tel: (1-510) 668-1321 Fax: (1-510) 226-9918

Web Site: http://www.fremontmicro.com/

* Information furnished is believed to be accurate and reliable. However, Fremont Micro Devices, Incorporated (BVI) assumes no responsibility for the consequences of use of such information or for any infringement of patents of other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent rights of Fremont Micro Devices, Incorporated (BVI). Specifications mentioned in this publication are subject to change without notice. This publication supersedes and replaces all information previously supplied. Fremont Micro Devices, Incorporated (BVI) products are not authorized for use as critical components in life support devices or systems without express written approval of Fremont Micro Devices, Incorporated (BVI). The FMD logo is a registered trademark of Fremont Micro Devices, Incorporated (BVI). All other names are the property of their respective owners.