

72-Mbit Video Frame Buffer

Features

- Memory organization ❐ Density: 72-Mbit ❐ Organization: × 36
- \blacksquare Up to 133-MHz clock operation $^{[1]}$ $^{[1]}$ $^{[1]}$
- Unidirectional operation
- Independent read and write ports
- ❐ Supports simultaneous read and write operations
- ❐ Reads and writes operate on independent clocks, upto a maximum ratio of two, enabling data buffering across clock domains.
- ❐ Supports multiple I/O voltage standard: low voltage complementary metal oxide semiconductor (LVCMOS) 3.3 V and 1.8 V voltage standards.
- Input and output enable control for write mask and read skip operations
- Empty & Full status flags
- Flow-through mailbox register to send data from input to output port, bypassing the Frame Buffer
- Separate serial clock (SCLK) input for serial programming of configuration registers
- Master reset to clear entire Frame Buffer
- Partial reset to clear data but retain programmable settings
- Joint test action group (JTAG) port provided for boundary scan function
- Industrial temperature range: -40 °C to +85 °C

Functional Description

The Video Frame Buffer is a 72-Mbit memory device which operates as a FIFO with a bus width of 36 bits. It has independent read and write ports, which can be clocked up to 133 MHz. The bus size of 36 bits enables a data throughput of 4.8 Gbps. The device also offers a simple and easy-to-use interface to reduce implementation and debugging efforts, improve time-to-market, and reduce engineering costs. This makes it an ideal memory choice for a wide range of applications including video and image processing or any system that needs buffering at high speeds across different clock domains.

The functionality of the Video Frame Buffer is such that the data is read out of the read port in the same sequence in which it was written into the write port. If writes and inputs are enabled (WEN \overline{B} IE), data on the write port gets written into the device at the rising edge of write clock. Enabling reads and outputs (REN & OE) fetches data on the read port at every rising edge of read clock. Both reads and writes can occur simultaneously at different speeds provided the ratio between read and write clock is in the range of 0.5 to 2. Appropriate flags are set whenever the device is empty or full.

The device also supports a flow-through mailbox register to bypass the frame buffer memory

For a complete list of related documentation, [click here](http://www.cypress.com/?rID=86262).

Note

1. For device operating at 150 MHz, Contact Sales.

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Logic Block Diagram

Contents

Pin Configuration

Figure 1. 209-ball FBGA pinout (Top View)

Notes
2. This pin should be tied to V_{SS} preferably or can be left floating to ensure normal operation.

Pin Definitions

Pin Definitions (continued)

^{3.} All V_{SS} pins should be connected to the same ground plane.

Architecture

The video frame buffer consists of a memory array of 72-Mbit along with the logic blocks to implement FIFO functionality and its associated features that are built around this memory array.

The input and output data buses have a maximum width of 36 bits. The input data bus goes to an input register and the data flow from the input register to the memory is controlled by the write control logic. The inputs to the write logic block are WCLK, WEN and IE. When the writes are enabled through WEN and if the inputs are enabled by \overline{E} , then the data on the input bus is written into the memory array at the rising edge of WCLK. This also increments the write pointer. Enabling writes but disabling the data input pins through IE only, increments the write pointer without doing any writes or altering the contents of the memory location.

Similarly, the output register is connected to the data output bus. Transfer of contents from the memory to the output register is controlled by the read control logic. The inputs to the read control logic include RCLK, REN, OE. When reads are enabled by REN and outputs are enabled using OE, the data from the memory pointed by the read pointer is transferred to the output data bus at the rising edge of RCLK along with active low DVal. If the outputs are disabled but the reads enabled, the outputs are in high impedance state, but internally the read pointer is incremented.

During write operation, the number of writes performed is always an even number (i.e., minimum write burst length is two and number of writes always a multiple of two), whereas during read operation, the number of reads performed can be even or odd (i.e., minimum read burst length is one).

Reset Logic

The frame buffer can be reset in two ways: Master Reset (MRS) and Partial Reset (PRS). The MRS initializes the read and write pointers to zero and sets the output register to all zeroes. It also resets empty flag, full flag & the configuration registers to their default values. A Master Reset is required after power-up before accessing the frame buffer.

PRS resets the read pointer, write pointer to the first physical location in the memory array. It also resets the flags to their default values. PRS does not affect the programmed configuration register values.

Data Valid Signal (DVal)

Data valid (DVal) is an active low signal, synchronized to RCLK and is provided to check for valid data on the output bus. When a read operation is performed, the DVal signal goes low along with output data. This helps user to capture the data without keeping track of REN to data output latency. This signal also helps when write and read operations are performed continuously at different frequencies by indicating when valid data is available at the output port Q[35:0].

Write Mask and Read Skip Operation

As mentioned in [Architecture on page 7,](#page-6-0) enabling writes but disabling the inputs (IE HIGH) increments the write pointer without doing any write operations or altering the contents of the location.

This feature is called Write Mask and allows user to move the write pointer without actually writing to the locations. This "write masking" ability is useful in some video applications such as Picture In Picture (PIP).

Similarly, during a read operation, if the outputs are disabled by keeping the OE high, the read data does not appear on the output bus; however, the read pointer is incremented. This feature is referred to as a Read Skip Operation.

Flow-through Mailbox Register

This feature transfers data from input to output directly bypassing the sequential buffer memory. When MB signal is asserted the data present in D[35:0] will be available at Q[35:0] after two WCLK cycles. Normal read and write operations are not allowed during flow-through mailbox operation. Before starting Flow-through mailbox operation reads should be completed to make data valid DVal high to avoid data loss from buffer memory.

Flag Operation

This device provides two flag pins to indicate the condition of the video frame buffer.

Full Flag

The Full Flag (FF) operates on double word (burst length of two) boundaries and goes LOW when the device is full. Write operations are inhibited whenever FF is LOW regardless of the state of WEN. FF is synchronized to WCLK, that is, it is exclusively updated by each rising edge of WCLK. The worst case assertion latency for Full Flag is four. As the user cannot know that the frame buffer is full for four clock cycles, it is possible that user continues writing data during this time. In this case, the four data words written will be stored to prevent data loss and these words have to be read back in order for full flag to get de-asserted. The minimum number of reads required to de-assert full-flag is two and the maximum number of reads required to de-assert full flag is six. The latency associated with Full flag is explained in [Latency Table on page 15.](#page-14-1)

Empty Flag

The Empty Flag (EF) deassertion depends on burst writes and goes LOW when the device is empty. Read operations are inhibited whenever EF is LOW, regardless of the state of REN. EF is synchronized to RCLK, that is, it is exclusively updated by each rising edge of RCLK. The latency associated with Empty flag is explained in [Latency Table on page 15.](#page-14-1)

Programming Configuration Registers

The CYFB0072V has ten 8-bit user configurable registers. The tenth register is the Fast CLK bit which indicates the faster clock domain.

This register can be programmed in one of two ways: serial loading or parallel loading method. The loading method is selected using the SPI_SEN (Serial Enable) pin. A low on the SPI_SEN selects the serial method for writing into the register. For serial programming, there is a separate SCLK and a Serial Input (SI). In parallel mode, a LOW on the load (\overline{LD}) pin causes the write and read operation to these registers. When LD is held LOW, write and read operations happen sequentially from the

first location (0x1) to the last location (0xA). If $\overline{\text{LD}}$ is HIGH, the writes occur to the FIFO.

Register values can be read through the parallel output port regardless of the programming mode selected (serial or parallel). Register values cannot be read serially. The registers may be programmed (and reprogrammed) any time after master reset, regardless of whether serial or parallel programming is selected.

See [Table 1](#page-7-1) and [Table 2 on page 9](#page-8-0) for access to configuration registers in serial and parallel modes.

In parallel mode, the read and write operations loop back when the maximum address location of the configuration registers is reached. Simultaneous read and write operations should be avoided on the configuration registers.

Table 1. Configuration Registers

Table 2. Writing and Reading Configuration Registers in Parallel Mode

SPI SEN	LD	WEN	REN	WCLK	RCLK	SCLK	Operation
Ω		х	X	x	X	Rising edge	Each rising of the SCLK clocks in one bit from the SI (Serial In). Any of the 10 registers can be addressed and written to, following the SPI protocol.
x		0	X	\uparrow Rising edge	х	Χ	Parallel write to Frame Buffer memory.
x		x	0	x	\uparrow Rising edge	Х	IParallel read from Frame Buffer memory.
	0		4	x	х	х	This corresponds to parallel mode (refer to Table 2 on page 9).

Table 3. Writing into Configuration Registers in Serial Mode

Width Expansion Configuration

The width of the frame buffer can be expanded to provide word widths greater than 36 bits. During width expansion mode, all control line inputs are common and all flags are available. Empty (Full) flags are created by ANDing the Empty (Full) flags of every Frame Buffer. This technique avoids reading data from or writing data to the device that is "staggered" by one clock cycle due to the variations in skew between RCLK and WCLK. [Figure 3](#page-10-3) demonstrates an example of 72 bit-word width by using two 36-bit word frame buffers.

Figure 3. Width Expansion

Power Up

The device becomes functional after V_{CC1} , V_{CC2} , V_{CCIO} , and V_{REF} attain minimum stable voltage required as given in [Recommended DC Operating Conditions on page 14](#page-13-2). The device can be accessed t_{PU} time after these supplies attain the minimum required level (see [Switching Characteristics on page](#page-16-0) [17\)](#page-16-0). There is no specific power sequencing required for the device.

Read/Write Clock Requirements

The read and write clocks must satisfy the following requirements:

- Both read (RCLK) and write (WCLK) clocks should be free-running.
- The clock frequency for both clocks should be between the minimum and maximum range given in [Electrical](#page-13-3) [Characteristics on page 14](#page-13-3).
- The WCLK to RCLK ratio should be in the range of 0.5 to 2.

For proper frame buffer operation, the device must determine which of the input clocks $-$ RCLK or WCLK $-$ is faster. This is evaluated using counters after the MRS cycle. The device uses two 9-bit counters (one running on RCLK and other on WCLK), which count 256 cycles of read and write clocks after MRS. The clock of the counter which reaches its terminal count first is used as master clock inside the frame buffer.

When there is change in the relative frequency of RCLK and WCLK during normal operation of Frame Buffer, user can specify it by using "Fast CLK bit" in the configuration register (0xA).

- "1" indicates f_{req} (WCLK) > f_{req} (RCLK)
- "0" indicates f_{req} (WCLK) < f_{req} (RCLK)

The fast clock bit configuration register(0xA), can be accessed by keeping LD low for 10 clock cycles. The result of counter evaluated frequency is available in this register bit. User can override the counter evaluated frequency for faster clock by changing this bit.

Whenever there is a change in this bit value, user must wait t_{PIL} time before issuing the next read or write to buffer memory.

JTAG Operation

The video frame buffer has two devices connected internally in a JTAG chain, as shown in figure [Figure 4](#page-11-3).

Test Access Port

Test Clock (TCK)

The test clock is used only with the TAP controller. All inputs are captured on the rising edge of TCK. All outputs are driven on the falling edge of TCK.

Test Mode Select (TMS)

The TMS input is used to give commands to the TAP controller and is sampled on the rising edge of TCK. This pin may be left unconnected if the TAP is not used. The pin is pulled up internally, resulting in a logic HIGH level.

Test Data-In (TDI)

The TDI pin is used to serially input information into the registers and can be connected to the input of any of the registers. The register between TDI and TDO is chosen by the instruction that is loaded into the TAP instruction register. For information on loading the instruction register, see [Figure 5](#page-12-5). TDI is internally pulled up and can be left unconnected if the TAP is unused in an application. TDI is connected to the most significant bit (MSB) on any of the TAP registers.

Test Data-Out (TDO)

The TDO output pin is used to serially clock data out from the registers. The output is active, depending upon the current state of the TAP state machine. The output changes on the falling edge of TCK. TDO is connected to the least significant bit (LSB) of any of the TAP registers.

Note: Reset is performed by forcing TMS HIGH (V_{DD}) for five rising edges of TCK.

Tap Registers

Registers are connected between the TDI and TDO pins to scan the data in and out of the test circuitry. Only one register can be selected at a time through the instruction registers. Data is serially loaded into the TDI pin on the rising edge of TCK. Data is output on the TDO pin on the falling edge of TCK.

Instruction Register

Three-bit instructions can be serially loaded into the instruction register. This register is loaded when it is placed between the TDI and TDO pins, as shown in [Figure 5](#page-12-5). Upon power-up, the instruction register is loaded with the IDCODE instruction. It is also loaded with the IDCODE instruction if the controller is placed in a Reset state.

Bypass Register

To save time when serially shifting data through registers, it is sometimes advantageous to skip certain chips. The bypass register is a single-bit register that can be placed between TDI and TDO pins. This enables shifting of data through the device with minimal delay.

Boundary Scan Register

The boundary scan register is connected to all of the input and output pins on the device. Several No Connect (NC) pins are also included in the scan register to reserve pins for higher density devices.

The boundary scan register is loaded with the contents of the device input and output ring when the TAP controller is in the Capture-DR state and is then placed between the TDI and TDO pins when the controller is moved to the Shift-DR state. The EXTEST, SAMPLE/PRELOAD instructions can be used to capture the contents of the input and output ring. The MSB of the register is connected to TDI and the LSB is connected to TDO.

Identification (ID) Register

The ID register is loaded with a vendor-specific, 32-bit code during the Capture-DR state when the IDCODE command is loaded in the instruction register. The IDCODE is hardwired into the device and can be shifted out when the TAP controller is in the Shift-DR state. The ID register has a vendor code and other information described in [Table 4.](#page-12-6)

JTAG ID Codes

Table 4. JTAG IDCODES

OPCODES Supported

Table 5. OPCODES Supported

JTAG Instructions

IDCODE

The IDCODE instruction loads a vendor-specific, 32-bit code into the instruction register. It also places the instruction register between the TDI and TDO pins and shifts the IDCODE out of the device when the TAP controller enters the Shift-DR state. The IDCODE instruction is loaded into the instruction register at power-up or whenever the TAP controller is supplied a Test-Logic-RST state.

SAMPLE/PRELOAD

SAMPLE/PRELOAD is an IEEE 1149.1 mandatory instruction. When the SAMPLE/PRELOAD instructions are loaded into the instruction register and the TAP controller is in the Capture-DR state, a snapshot of data on the input and output pins is captured in the boundary scan register.

PRELOAD places an initial data pattern at the latched parallel outputs of the boundary scan register cells before the selection of another boundary scan test operation.

The shifting of data for the SAMPLE and PRELOAD phases can occur concurrently when required; that is, while the data captured is shifted out, the preloaded data can be shifted in.

BYPASS

When the BYPASS instruction is loaded in the instruction register and the TAP is placed in a Shift-DR state, the bypass register is placed between the TDI and TDO pins. The advantage of the BYPASS instruction is that it shortens the boundary scan path when multiple devices are connected together on a board.

HIGHZ

The HIGHZ instruction mode is used to set all the user I/O pins to an inactive drive state. These pins are tri-stated until a new JATG instruction is executed. When this instruction is selected, the bypass register is connected between the TDI and TDO ports.

EXTEST

The EXTEST instruction drives the preloaded data out through the system output pins. This instruction also connects the

boundary scan register for serial access between the TDI and TDO in the Shift-DR controller state.

Instruction Update and Bypass

- 1. Every time an instruction is loaded through JTAG port, BY-PASS command needs to be loaded on device 1. For example, to push PRELOAD command, BYPASS to device-1 i 111["] + PRELOAD to device-2 i 00000001["] needs to be sent.
- 2. When both devices are put on BYPASS, any pattern sent in should be observed on TDO after two TCK delay.

TAP Controller State Diagram

Figure 5. TAP Controller State Diagram

TAP controller is a Finite State Machine with 16 states as shown in [Figure 5](#page-12-5). State change is determined by the state of TMS on rising edge of TCK. [Figure 5](#page-12-5) shows the value of TMS for each state transition.

Maximum Ratings

Exceeding maximum ratings may shorten the useful life of the device. These user guidelines are not tested.

Operating Range

Recommended DC Operating Conditions

Electrical Characteristics

I/O Characteristics

(Over the operating range)

Latency Table

Note 5. These latency values are valid for a clock ratio of 1.

AC Test Load Conditions

Figure 6. AC Test Load Conditions

(a)
$$
V_{\text{CCIO}} = 1.8
$$
 Volt

(b) V_{CCIO} = 3.3 Volt

(c) All Input Pulses

Switching Characteristics

Switching Waveforms

Figure 7. Write Cycle Timing

Figure 12. Initial Data Latency

Figure 17. Empty Flag Assertion

Figure 19. Full Flag Deassertion

Ordering Information

Ordering Code Definitions

Package Diagram

Figure 20. 209-ball FBGA (14 × 22 × 1.76 mm) BB209A Package Outline, 51-85167

PKG WEIGHT: REFER TO PMDD SPEC

51-85167 *C

Acronyms **Document Conventions**

Units of Measure

Document History Page

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