

6 A Synchronous DC-DC Step down Regulator ($V_{IN} = 4.5\text{ V to }28\text{ V}$, $V_{OUT} = 0.75\text{ V to }5.5\text{ V}$)

FEATURES

- High-Speed Response DC-DC Step Down Regulator Circuit that employs Hysteretic Control System
- Hi-side 20 mΩ (Typ) and Low-side 10 mΩ (Typ) MOSFETs for High Efficiency at 6 A
- Skip (discontinuous) Mode for Light Load Efficiency
- Maximum Output Current : 6 A
- Input Voltage Range : $AV_{IN} = 4.5\text{ V to }28\text{ V}$,
 $PV_{IN} = 4.5\text{ V to }28\text{ V}$
- Output Voltage Range : 0.75 V to 5.5 V
- Selectable Switching Frequency
: 210 kHz, 430 kHz, 650 kHz
- Adjustable Soft Start
- Low Operating and Standby Quiescent Current
- Power Good Indication for Output Over and Under Voltage
- Built-in Under Voltage Lockout (UVLO), Thermal Shut Down (TSD), Over Voltage Detection (OVD), Under Voltage Detection (UVD), Over Current Protection (OCP), Short Circuit Protection (SCP)
- 24 pin Plastic Quad Flat Non-leaded Package Heat Slug Down (QFN Type)
(Size : 4 mm × 4 mm × 0.7 mm, 0.5 mm pitch)

DESCRIPTION

NN30321A is a synchronous DC-DC Step down Regulator (1-ch) comprising of a Controller IC and two power MOSFETs and employs the hysteretic control system.

By this system, when load current changes suddenly, it responds at high speed and minimizes the changes of output voltage.

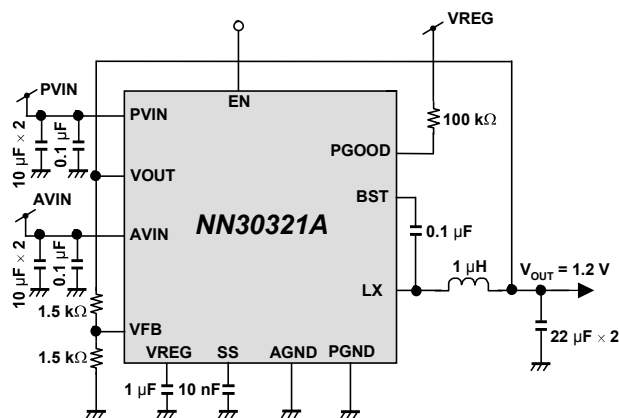
Since it is possible to use capacitors with small capacitance and it is unnecessary to add external parts for system phase compensation, this IC realizes downsizing of set and reducing in the number of external parts. Output voltage is adjustable by user. Maximum current is 6 A.

APPLICATIONS

High Current Distributed Power Systems such as

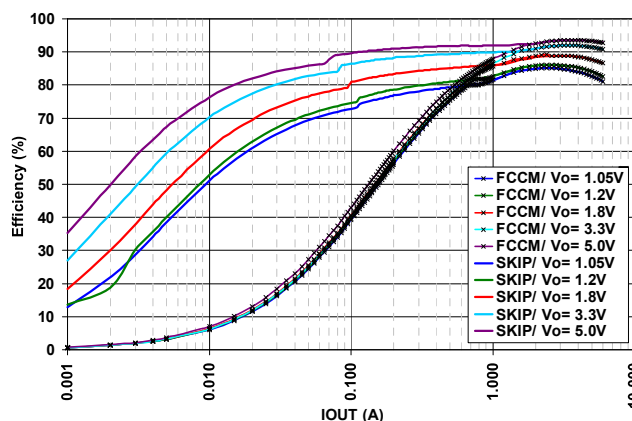
- HDDs (Hard Disk Drives)
- SSDs (Solid State Drives)
- PCs
- Game consoles
- Servers
- Security Cameras
- Network TVs
- Home Appliances
- OA Equipment etc.

APPLICATION CIRCUIT EXAMPLE



Note : The application circuit is an example. The operation of the mass production set is not guaranteed. Sufficient evaluation and verification is required in the design of the mass production set. The Customer is fully responsible for the incorporation of the above illustrated application circuit in the design of the equipment.

EFFICIENCY CURVE



Condition :

$V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.05 V, 1.2 V, 1.8 V, 3.3 V, 5.0 V,
Switching Frequency = 650 kHz, FCCM / Skip Mode,
 $L_O = 1\text{ }\mu\text{H}$, $C_O = 44\text{ }\mu\text{F}$ (22 $\mu\text{F} \times 2$)

ORDERING INFORMATION

Order Number	Feature	Package	Output Supply
NN30321A-VB	Maximum Output Current : 6 A	24 pin HQFN	Emboss Taping

ABSOLUTE MAXIMUM RATINGS

Parameter	Symbol	Rating	Unit	Notes
Supply voltage	V_{IN}	30	V	*1
Operating free-air temperature	T_{opr}	- 40 to + 85	°C	*2
Operating junction temperature	T_j	- 40 to + 150	°C	*2
Storage temperature	T_{stg}	- 55 to + 150	°C	*2
Input Voltage Range	$V_{MODE}, V_{FSEL}, V_{OUT}, V_{FB}$	- 0.3 to ($V_{REG} + 0.3$)	V	*1 *3
	V_{EN}	- 0.3 to 6.0	V	*1
Output Voltage Range	V_{PGOOD}	- 0.3 to ($V_{REG} + 0.3$)	V	*1 *3
	V_{LX}	- 0.3 to ($V_{IN} + 0.3$)	V	*1 *4
ESD	HBM	2	kV	—

Notes : This product may sustain permanent damage if subjected to conditions higher than the above stated absolute maximum rating. This rating is the maximum rating and device operating at this range is not guaranteed as it is higher than our stated recommended operating range.
When subjected under the absolute maximum rating for a long time, the reliability of the product may be affected.
 V_{IN} is voltage for AVIN, PVIN. $V_{IN} = AV_{IN} = PV_{IN}$.
Do not apply external currents and voltages to any pin not specifically mentioned.

- *1 : The values under the condition not exceeding the above absolute maximum ratings and the power dissipation.
- *2 : Except for the power dissipation, operating ambient temperature, and storage temperature, all ratings are for $T_a = 25$ °C.
- *3 : ($V_{REG} + 0.3$) V must not exceed 6 V.
- *4 : ($V_{IN} + 0.3$) V must not exceed 30 V.

POWER DISSIPATION RATING

Package	θ_{J-a}	θ_{J-C}	PD (Ta = 25 °C)	PD (Ta = 85 °C)	Notes
24 pin Plastic Quad Flat Non-leaded Package Heat Slug Down (QFN Type)	58.0 °C / W	7.2 °C / W	2.155 W	1.120 W	*1
	35.7 °C / W	4.7 °C / W	3.501 W	1.820 W	*2

Notes : For the actual usage, please follow the power supply voltage, load and ambient temperature conditions to ensure that there is enough margin and the thermal design does not exceed the allowable value.

*1:Glass Epoxy Substrate (4 Layers) [50 × 50 × 0.8 t (mm)]

*2:Glass Epoxy Substrate (4 Layers) [50 × 50 × 1.57 t (mm)]



CAUTION

Although this IC has built-in ESD protection circuit, it may still sustain permanent damage if not handled properly. Therefore, proper ESD precautions are recommended to avoid electrostatic damage to the MOS gates.

RECOMMENDED OPERATING CONDITIONS

Parameter	Symbol	Min	Typ	Max	Unit	Notes
Supply voltage range	AV _{IN}	4.5	12	28	V	—
	PV _{IN}	4.5	12	28	V	—
Input Voltage Range	V _{MODE}	-0.3	—	V _{REG} + 0.3	V	*1
	V _{FSEL}	-0.3	—	V _{REG} + 0.3	V	*1
	V _{EN}	-0.3	—	5.0	V	—
Output Voltage Range	V _{PGOOD}	-0.3	—	V _{REG} + 0.3	V	*1
	V _{LX}	-0.3	—	V _{IN} + 0.3	V	*2

Notes : Voltage values, unless otherwise specified, are with respect to GND.

GND is voltage for AGND, PGND. AGND = PGND

V_{IN} is voltage for AVIN, PVIN. V_{IN} = AV_{IN} = PV_{IN}.

Do not apply external currents or voltages to any pin not specifically mentioned.

*1 : (V_{REG} + 0.3) V must not exceed 6 V.

*2 : (V_{IN} + 0.3) V must not exceed 30 V.

ELECTRICAL CHARACTERISTICS

$C_O = 22 \mu\text{F} \times 2$, $L_O = 1 \mu\text{H}$, V_{OUT} Setting = 1.2 V, $V_{IN} = AV_{IN} = PV_{IN} = 12 \text{ V}$,
Switching Frequency = 650 kHz, $V_{MODE} = V_{REG}$ (FCCM)
 $T_a = 25 \text{ }^\circ\text{C} \pm 2 \text{ }^\circ\text{C}$ unless otherwise noted.

Parameter	Symbol	Condition	Limits			Unit	Note
			Min	Typ	Max		
Current Consumption							
Consumption current at active	I_{OPR}	$V_{EN} = 5 \text{ V}$, $I_{OUT} = 0 \text{ A}$ $R_{FB1} = 1.5 \text{ k}\Omega$ $R_{FB2} = 1.5 \text{ k}\Omega$ $V_{MODE} = \text{GND}$ (Skip Mode)	—	650	1000	μA	—
Consumption current at standby	I_{STB}	$V_{EN} = 0 \text{ V}$	—	2	4	μA	—
Logic Pin Characteristics							
EN pin Low level input voltage	V_{ENL}	—	—	—	0.3	V	—
EN pin High level input voltage	V_{ENH}	—	1.5	—	5.0	V	—
EN pin leak current	I_{leakEN}	$V_{EN} = 5 \text{ V}$	—	6.25	12.5	μA	—
MODE pin Low level input voltage	V_{MDL}	—	—	—	$V_{REG} \times 0.3$	V	—
MODE pin High level input voltage	V_{MDH}	—	$V_{REG} \times 0.7$	—	V_{REG}	V	—
MODE pin leak current	I_{leakMD}	$V_{MODE} = 5 \text{ V}$	—	6.25	12.5	μA	—
FSEL pin Low level input voltage	V_{FSL}	—	—	—	0.3	V	—
FSEL pin High level input voltage	V_{FSH}	—	—	$V_{REG} - 0.1$	—	V	*1
FSEL pin leak current	I_{leakFS}	$V_{FSEL} = 5 \text{ V}$	—	15.0	25.0	μA	—
VREG Characteristics							
Output voltage	V_{REG}	$I_{VREG} = 6 \text{ mA}$	5.4	5.7	6.0	V	—
Line regulation	V_{REGLIN}	$V_{REGLIN} = V_{REG} (V_{IN} = 12 \text{ V})$ $- V_{REG} (V_{IN} = 6 \text{ V})$ $I_{VREG} = 6 \text{ mA}$	—	—	200	mV	—
Drop out voltage	V_{REGDO}	$V_{IN} = 4.5 \text{ V}$ $I_{VREG} = 6 \text{ mA}$	4.11	—	—	V	—

Note : *1 : Typical design value

ELECTRICAL CHARACTERISTICS (Continued)

$C_O = 22 \mu\text{F} \times 2$, $L_O = 1 \mu\text{H}$, V_{OUT} Setting = 1.2 V, $V_{IN} = AV_{IN} = PV_{IN} = 12 \text{ V}$,

Switching Frequency = 650 kHz, $V_{MODE} = V_{REG}$ (FCCM)

$T_a = 25 \text{ }^\circ\text{C} \pm 2 \text{ }^\circ\text{C}$ unless otherwise noted.

Parameter	Symbol	Condition	Limits			Unit	Note
			Min	Typ	Max		
VFB Characteristics							
VFB comparator threshold	V_{FBTH}	—	0.594	0.600	0.606	V	—
VFB pin leak current 1	I_{leakF1}	$V_{FB} = 0 \text{ V}$	-1	—	1	μA	—
VFB pin leak current 2	I_{leakF2}	$V_{FB} = 6 \text{ V}$	-1	—	1	μA	—
Under Voltage Lockout (UVLO)							
UVLO detection voltage	V_{UVLODE}	$V_{IN} = 5 \text{ V to } 0 \text{ V}$	3.5	3.8	4.1	V	—
UVLO recover voltage	V_{UVLORE}	$V_{IN} = 0 \text{ V to } 5 \text{ V}$	3.9	4.2	4.5	V	—
PGOOD Characteristics							
PGOOD Threshold 1 (V_{FB} ratio for UVD detect)	V_{PGUV}	V_{PGOOD} : High to Low	77	85	93	%	—
PGOOD Hysteresis 1 (V_{FB} ratio for UVD release)	ΔV_{PGUV}	V_{PGOOD} : Low to High	3.5	5.0	6.5	%	—
PGOOD Threshold 2 (V_{FB} ratio for OVD detect)	V_{PGOV}	V_{PGOOD} : High to Low	107	115	123	%	—
PGOOD Hysteresis 2 (V_{FB} ratio for OVD release)	ΔV_{PGOV}	V_{PGOOD} : Low to High	3.5	5.0	6.5	%	—
PGOOD ON resistance	R_{PGOOD}	—	—	10	15	Ω	—

ELECTRICAL CHARACTERISTICS (Continued)

$C_O = 22 \mu\text{F} \times 2$, $L_O = 1 \mu\text{H}$, $V_{\text{OUT}} \text{ Setting} = 1.2 \text{ V}$, $V_{\text{IN}} = AV_{\text{IN}} = PV_{\text{IN}} = 12 \text{ V}$,

Switching Frequency = 650 kHz, $V_{\text{MODE}} = V_{\text{REG}}$ (FCCM)

$T_a = 25 \text{ }^\circ\text{C} \pm 2 \text{ }^\circ\text{C}$ unless otherwise noted.

Parameter	Symbol	Condition	Limits			Unit	Note
			Min	Typ	Max		
DC-DC Characteristics							
Line regulation	V_{LIN}	$V_{\text{IN}} = 6 \text{ V to } 28 \text{ V}$ $I_{\text{OUT}} = 0.5 \text{ A}$	—	0.25	0.75	%/V	—
Load regulation	V_{LOA}	$I_{\text{OUT}} = 10 \text{ mA to } 6 \text{ A}$	—	3.5	—	%	*1
Output ripple voltage 1	V_{R1}	$I_{\text{OUT}} = 10 \text{ mA}$	—	20	—	mV [p-p]	*1
Output ripple voltage 2	V_{R2}	$I_{\text{OUT}} = 3 \text{ A}$	—	20	—	mV [p-p]	*1
Load transient response 1	ΔV_{TR1}	$I_{\text{OUT}} = 100 \text{ mA to } 3 \text{ A}$ $\Delta t = 0.5 \text{ A} / \mu\text{s}$	—	20	—	mV	*1
Load transient response 2	ΔV_{TR2}	$I_{\text{OUT}} = 3 \text{ A to } 100 \text{ mA}$ $\Delta t = 0.5 \text{ A} / \mu\text{s}$	—	20	—	mV	*1
High Side Power MOSFET ON resistance	R_{ONH}	$V_{\text{GS}} = 5.7 \text{ V}$	—	20	40	m Ω	—
Low Side Power MOSFET ON resistance	R_{ONL}	$V_{\text{GS}} = 5.7 \text{ V}$	—	10	20	m Ω	—
MIN input and output voltage difference	V_{diff}	$V_{\text{diff}} = V_{\text{IN}} - V_{\text{OUT}}$	—	2.5	—	V	*1

Note : *1 : Typical design value

ELECTRICAL CHARACTERISTICS (Continued)

$C_O = 22 \mu\text{F} \times 2$, $L_O = 1 \mu\text{H}$, V_{OUT} Setting = 1.2 V, $V_{IN} = AV_{IN} = PV_{IN} = 12 \text{ V}$,

Switching Frequency = 650 kHz, $V_{MODE} = V_{REG}$ (FCCM)

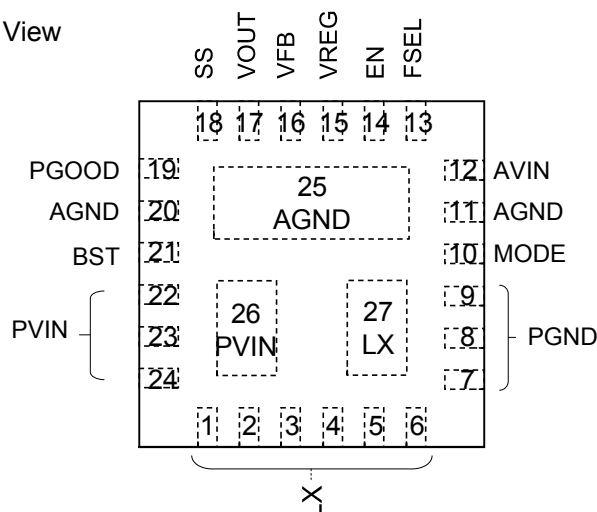
$T_a = 25 \text{ }^\circ\text{C} \pm 2 \text{ }^\circ\text{C}$ unless otherwise noted.

Parameter	Symbol	Condition	Limits			Unit	Note
			Min	Typ	Max		
Protection							
DC-DC Over Current Protection Limit	I_{LMT}	—	—	9	—	A	*1
DC-DC Short Circuit Protection Threshold	I_{short}	$V_{FB} = 0.6 \text{ V to } 0.0 \text{ V}$	50	60	70	%	—
Thermal Shut Down (TSD) Threshold	T_{TSDTH}	—	—	130	—	$^\circ\text{C}$	*1
Thermal Shut Down (TSD) Hysteresis	T_{TSDHYS}	—	—	34	—	$^\circ\text{C}$	*1
Soft Start Timing							
SS Charge Current	I_{SSCH}	$V_{SS} = 0.3 \text{ V}$	1.1	2.3	3.5	μA	—
SS Discharge Resistance (Shut down)	R_{SSDCH}	$V_{EN} = 0 \text{ V}$	—	5	10	$\text{k}\Omega$	—
Switching Frequency							
DC-DC Switching Frequency 1	f_{sw1}	$I_{OUT} = 3 \text{ A}$	—	210	—	kHz	*1
DC-DC Switching Frequency 2	f_{sw2}	$I_{OUT} = 3 \text{ A}$	—	430	—	kHz	*1
DC-DC Switching Frequency 3	f_{sw3}	$I_{OUT} = 3 \text{ A}$	—	650	—	kHz	*1

Note : *1 : Typical design value

PIN CONFIGURATION

Top View



PIN FUNCTIONS

Pin No.	Pin name	Type	Description
1	LX	Output	Power MOSFET output pin
2			An inductor is connected and switching operation is carried out between V_{IN} and GND.
3			Due to high current and large amplitude at this terminal, the parasitic inductance and impedance of the routing path can cause an increase in noise and a degradation in the efficiency. Routing path should be kept as short as possible.
4			
5			
6			
7	PGND	Ground	Ground pin for Power MOSFET
8			
9			
10	MODE	Input	Skip (discontinuous) Mode / FCCM (Forced Continuous Conduction Mode) select pin Skip Mode is set at Low level input, FCCM is set at High level input.
11	AGND	Ground	Ground pin
20			
12	AVIN	Power supply	Power supply pin Recommended rise time (time to reach 90 % of set value) setting is greater than or equal to 10 μ s and less than or equal to 1 s.
13	FSEL	Input	Frequency selection pin This is set to 430 kHz at Low level input, 210 kHz at High level input, and 650 kHz at open.
14	EN	Input	ON / OFF control pin DC-DC is stopped at Low level input, and it is started at High level input.

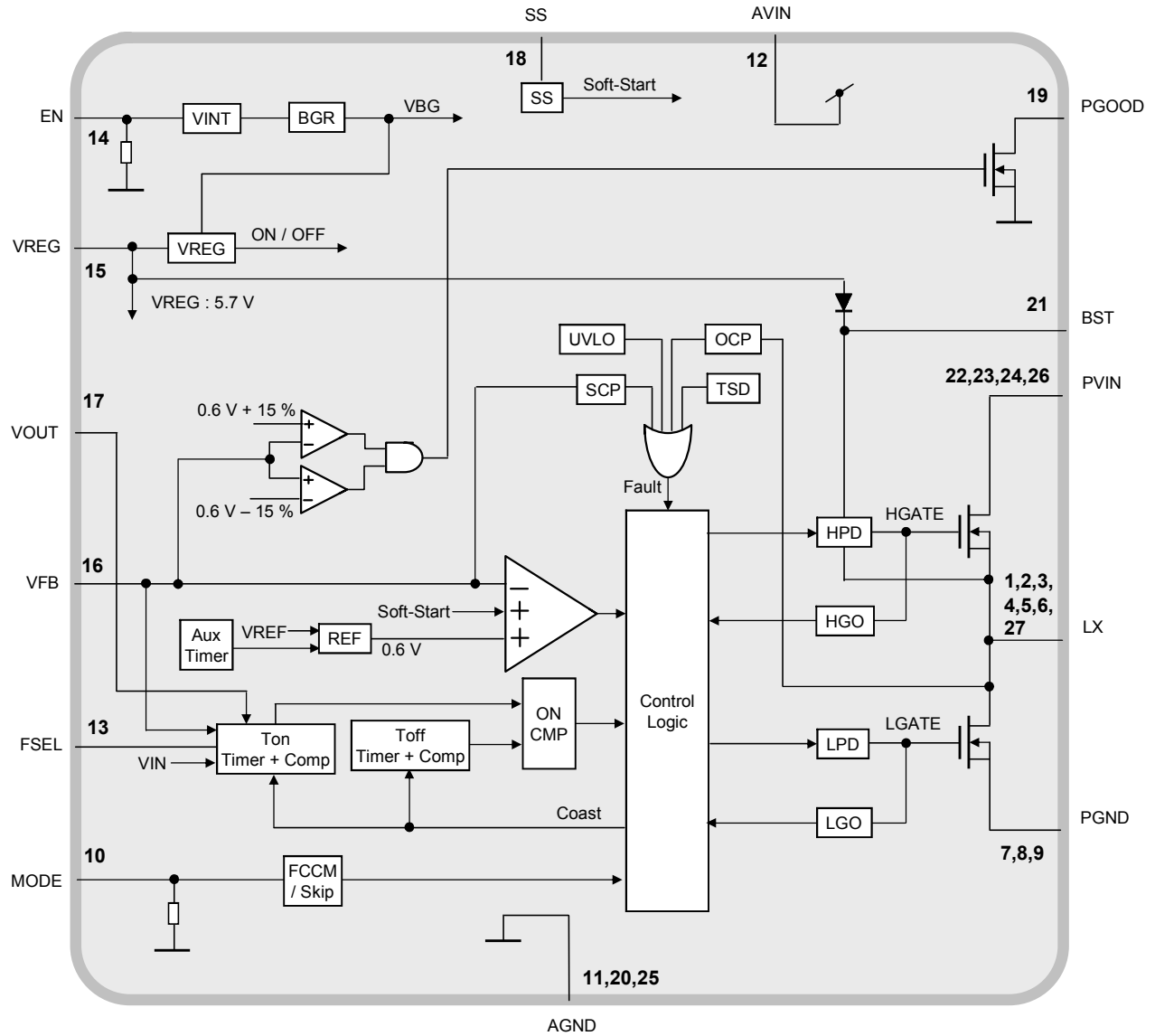
Note : Detailed pin descriptions are provided in the OPERATION and APPLICATION INFORMATION section.

PIN FUNCTIONS (Continued)

Pin No.	Pin name	Type	Description
15	VREG	Output	LDO output pin This is Output pin of Power supply (LDO) for internal control circuit. Please connect capacitor between VREG and GND.
16	VFB	Input	Comparator negative input pin VFB terminal voltage is regulated to REF output (internal reference voltage). Since VFB is a high impedance terminal, it should not be routed near other noisy path (LX, BST, etc.) or an inductor Routing path should be kept as short as possible.
17	VOUT	Input	Output voltage sense pin Switching frequency is controlled by monitoring output voltage.
18	SS	Output	Soft start capacitor connect pin The output voltage at a start up is smoothly controlled by adjusting Soft Start time. Please connect capacitor between SS and GND.
19	PGOOD	Output	Power good open drain pin A pull up resistor between PGOOD and VREG terminal is necessary. Output is low during Over or Under Voltage Detection conditions.
21	BST	Output	High side Power MOSFET gate driver pin Bootstrap operation is carried out in order to drive the gate voltage of High side Power MOSFET. Please connect a capacitor between BST and LX. Routing path should be kept as short as possible to minimize noise.
22	PVIN	Power supply	Power supply pin for Power MOSFET Recommended rise time (time to reach 90 % of set value) setting is greater than or equal to 10 μs and less than or equal to 1 s.
23			
24			
25	AGND	Ground	Ground pin for heat radiation
26	PVIN	Power supply	Power supply pin for heat radiation
27	LX	Output	Power MOSFET output pin for heat radiation

Note : Detailed pin descriptions are provided in the OPERATION and APPLICATION INFORMATION section.

FUNCTIONAL BLOCK DIAGRAM



Note : This block diagram is for explaining functions. Part of the block diagram may be omitted, or it may be simplified.

OPERATION

1. Protection

(1) Over Current Protection (OCP) and Short Circuit Protection (SCP)

- 1) The Over Current Protection is activated at about 9 A (Typ). This device uses pulse – by – pulse valley current protection method. When the low side power MOSFET is turned on, the voltage across the drain and source is monitored which is proportional to the inductor current. The high side power MOSFET is only allowed to turn on when the current flowing in the low side power MOSFET falls below the OCP level. Hence, during the OCP, the output voltage continues to drop at the specified current. OCP is a non – latch type protection.
- 2) The Short Circuit Protection function is implemented when the output voltage decreases and the VFB pin reaches to 60 % of the set voltage (0.6 V). If the VFB voltage stays below 60 % of the set voltage over 250 μ s after SCP triggers, both high side and low side power MOSFET will be latched off and the output will be discharged by internal MOSFET. Power reset or EN pin reset is necessary to activate the device again.

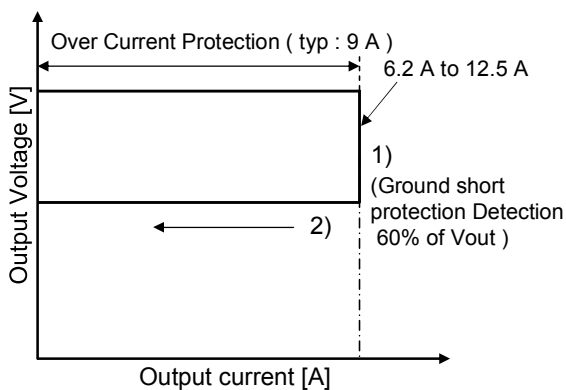


Figure : OCP and SCP Operation

(2) Over Voltage Detection (OVD)

If the VFB pin voltage exceeds 115 % of the set voltage (0.6 V) and lasts more than 10 ns, Over Voltage Detection will be triggered and PGOOD pin will be pulled down. Furthermore, in an over voltage condition, high side power MOSFET is turned off to stop PWM operation, and low side power MOSFET is turned on and held on until the inductor current starts to flow back to the device (for both Skip Mode and FCCM settings). If the VFB pin voltage drops below 110 % of the set voltage within 30 μ s after Over Voltage Detection triggers, PGOOD pin will be pulled up again and PWM operation will resume. Otherwise, both high side and low side MOSFET will be latched off and the output will be discharged by internal MOSFET. Power reset or EN pin reset is necessary to activate the device again.

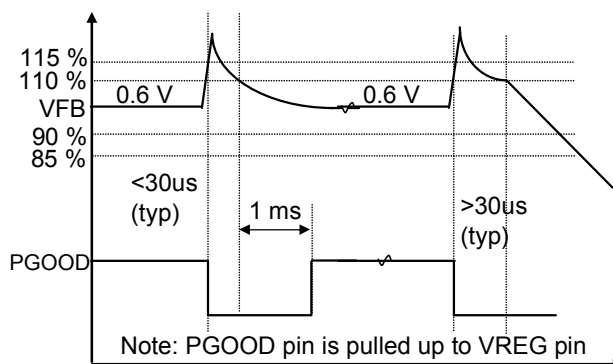


Figure : OVD Operation

(3) Output discharging function

When EN is low, the output is discharged by an internal MOSFET that is connected to VOUT pin. When EN is high, if the controller is turned off by Under Voltage Lock Out, or the controller is latched off by Over Voltage Detection or Short Circuit Protection, the output is discharged by the above said internal MOSFET.

The on resistance of the internal MOSFET is 50 Ω .

1. Protection (Continued)

(4) Under Voltage Detection (UVD)

During the operation, if the output voltage drops and the VFB pin voltage reaches 85 % of the set voltage (0.6 V), the MOSFET, the drain of which is connected to PGOOD pin, will turn on and pull the voltage of PGOOD to be low.

If the output voltage continues to drop and VFB pin voltage reaches 60 % of the set voltage (0.6 V), Short Circuit Protection (SCP) will be triggered. If the output voltage returns to 90 % of the set voltage (0.6 V) before triggering Short Circuit Protection, the MOSFET that is connected to PGOOD pin will turn off after 1 ms and PGOOD voltage will become logic high.

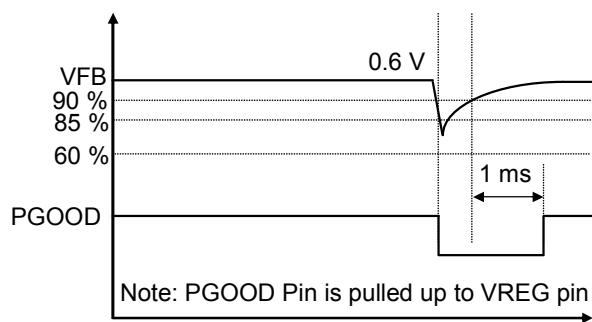


Figure : UVD Operation

(5) Thermal Shut Down (TSD)

When the IC internal temperature becomes more than about 130°C, TSD operates and DC-DC turns off.

2. Pin Setting

(1) Operating Mode Setting

The IC can operate at two different modes :

Skip (discontinuous) Mode and Forced Continuous Conduction Mode (FCCM).

In Skip Mode, the IC is working under pulse skipping mechanism to improve efficiency at light load condition.

In FCCM mode, the IC is working at fixed frequency to avoid EMI issues.

The Operating Mode can be set by MODE pin as follows.

MODE pin	Mode
Low	Skip Mode
High	FCCM

(2) Switching Frequency Setting

The IC can operate at three different frequency : 650 kHz, 430 kHz and 210 kHz.

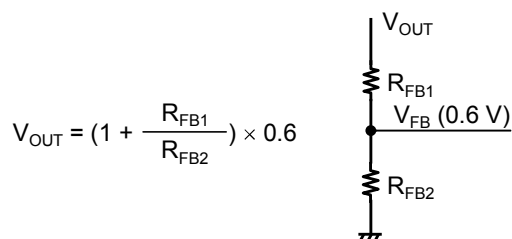
The Switching Frequency can be set by FSEL pin as follows.

FSEL pin	Frequency [kHz]
Low	430
High	210
Open	650

OPERATION (Continued)

3. Output Voltage Setting

The Output Voltage can be set by external resistance of VFB pin, and its calculation is as follows.



Below resistors are recommended for following popular output voltage.

V _{OUT} [V]	R _{FB1} [Ω]	R _{FB2} [Ω]
5.0	11.0 k	1.5 k
3.3	4.5 k	1.0 k
1.8	2.0 k	1.0 k
1.2	1.5 k	1.5 k
1.0	1.0 k	1.5 k

Note : R_{FB2} can be set to a maximum value of 10 kΩ.
A larger R_{FB2} value will be more susceptible to noise.

VFB comparator threshold is adjusted to ± 1 %, but The actual output voltage accuracy becomes more than ± 1 % due to the influence from the circuits other than VFB comparator.

In the case of V_{OUT} setting = 3.3 V, the actual output voltage accuracy becomes ± 2.5 %.

(FCCM, V_{IN} = 12 V, I_{OUT} = 0 A, Switching Frequency = 650 kHz)

4. Soft Start Setting

Soft Start function maintains the smooth control of the output voltage during start up by adjusting soft start time. When the EN pin becomes High, the current (2 μA) begin to charge toward the external capacitor (C_{SS}) of SS pin, and the voltage of SS pin increases straightly.

Because the voltage of VFB pin is controlled by the voltage of SS pin during start up, the voltage of VFB increase straightly to the regulation voltage (0.6 V) together with the voltage of SS pin and keep the regulation voltage after that. On the other hand, the voltage of SS pin increase to about 2.8 V and keep the voltage. The calculation of Soft Start Time is as follows.

$$\text{Soft Start Setting [s]} = \frac{0.6}{2\mu} \times C_{SS}$$

C_{SS} : External capacitor value of SS pin

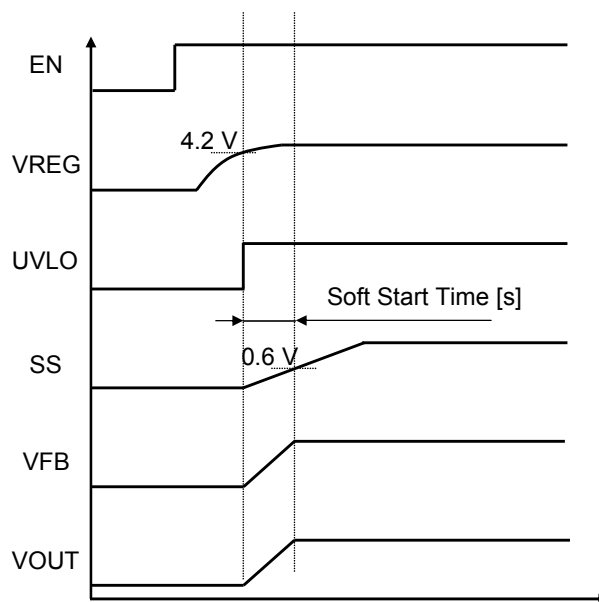


Figure : Soft Start Operation

OPERATION (Continued)

5. Start Up / Shut Down Settings

The Start up / Shut down is enabled by the EN pin.
The EN pin can be set by either applying voltage from an external voltage source or through a resistor connected to the AVIN pin.

Case 1 : Setting up the EN pin using an external voltage source. When an external voltage source is used, the EN pin input voltage (V_{ENH} , V_{ENL}) should satisfy the conditions as defined in the electrical characteristics.

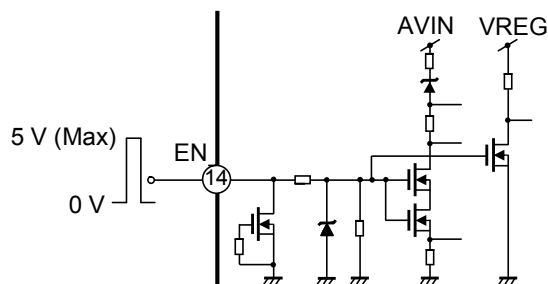


Figure : Internal circuit with EN pin

Case 2 : Setting up the EN pin through a resistor connected to AVIN pin. When setting up the EN pin through a resistor connected to the AVIN pin, refer to equations (1) and (2) to calculate the optimal resistor settings.

[Equation]

$$(1) : R_{EN1} > \frac{AV_{IN} - V_d}{I_d}$$

$$(2) : R_{EN1} < \frac{(AV_{IN} - V_{ENH}) \times R_{EN2}}{V_{ENH}}$$

[Example]

$$(1) : R_{EN1} > \frac{12\text{ V} - 6\text{ V}}{100\ \mu\text{A}} = 60\ \text{k}\Omega$$

$$(2) : R_{EN1} < \frac{(12\text{ V} - 5\text{ V}) \times 400\ \text{k}\Omega}{5\text{ V}} = 560\ \text{k}\Omega$$

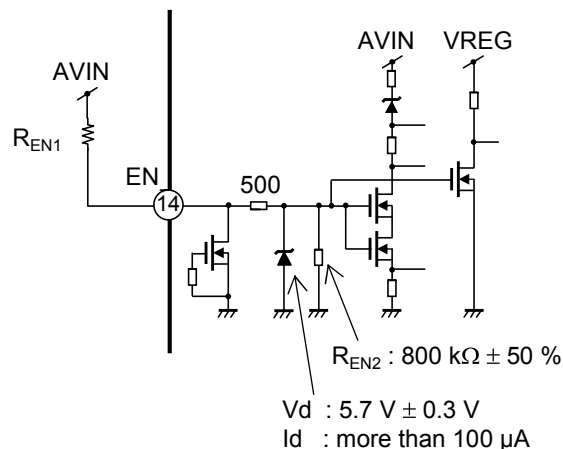


Figure : Internal circuit with EN pin

OPERATION (Continued)

6. Power ON / OFF Sequence

- (1) When the EN pin is set to High after the V_{IN} settles, the BGR and the VREG start up. (Recommended V_{IN} rise time setting is greater than or equal to $10 \mu\text{s}$ and less than or equal to 1 s.)
- (2) When the VREG pin exceeds its threshold value, the UVLO is released and the Soft Start Sequence is enabled.
 The capacitor connected to the SS pin begins to charge and the SS pin voltage increases linearly.
- (3) The VOUT pin (DC-DC Output) voltage increases at the same rate as the SS pin.
 Normal operation begins after the VOUT pin reaches the set voltage.
- (4) When the EN pin is set to Low, the BGR, VREG and UVLO stop operation. The VOUT pin / SS pin Voltage starts to drop and the VOUT pin discharge time depends on the value of the Feedback resistors and the output load current.

Note : The SS pin capacitor should be discharged completely before restarting the startup sequence.
 An incomplete discharge process might result in an overshoot of the output voltage.

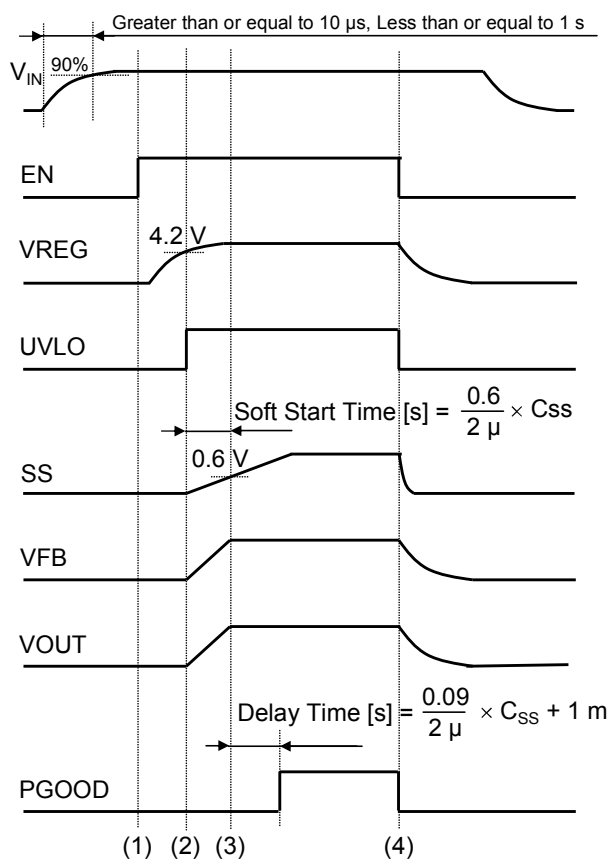
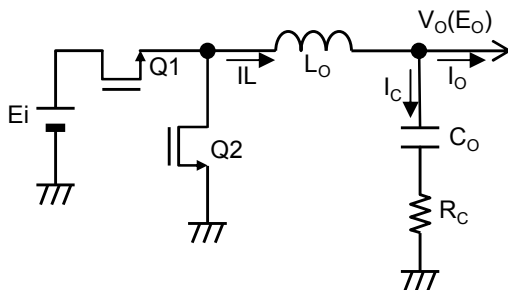
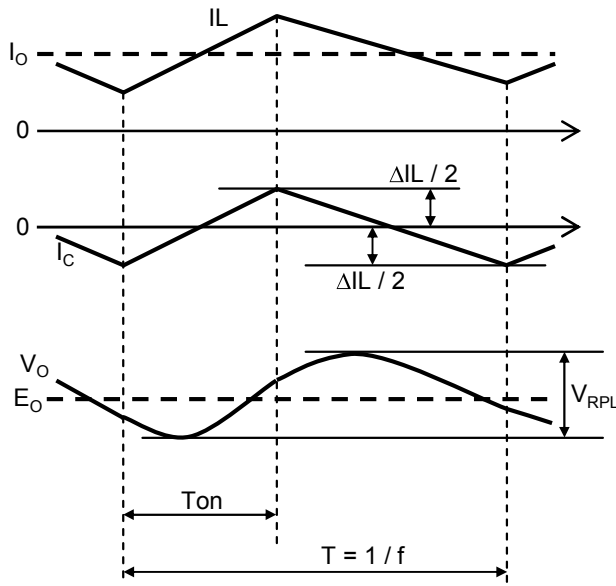


Figure : Power ON / OFF Sequence

OPERATION (Continued)

7. Inductor and Output Capacitor Setting



Given the desired input and output voltages, the inductor value and operating frequency determine the ripple current.

$$\Delta IL = \frac{E_o \cdot (E_i - E_o)}{E_i \cdot L_o \cdot f}$$

$$I_{ox} = \frac{\Delta IL}{2}$$

Highest efficiency operation is obtained at low frequency with small ripple current. However, achieving this requires a large inductor. There is a trade off among component size, efficiency and operating frequency. A reasonable starting point is to choose a ripple current that is about 40 % of I_o (Max). The largest ripple current occurs at the highest E_i . To guarantee that ripple current does not exceed a specified maximum, the inductance should be chosen according to:

$$L_o \geq \frac{E_o \cdot (E_i - E_o)}{2E_i \cdot I_{ox} \cdot f} \quad @ E_i = E_{i_max}$$

And its maximum current rating is

$$I_{L_max} = I_{o_max} + \frac{\Delta IL}{2} \quad @ E_i = E_{i_max}$$

The selection of C_o is primarily determined by the ESR (R_c) required to minimize voltage ripple and load transients. The output ripple V_{RPL} is approximately bounded by:

$$\begin{aligned} V_{rpl} &= V_{op} - V_{ob} = E_i \cdot \frac{C_o \cdot R_c^2}{2L_o} + \frac{\Delta IL}{8C_o \cdot f} \\ &= E_i \cdot \frac{C_o \cdot R_c^2}{2L_o} + \frac{E_o \cdot (E_i - E_o)}{8E_i \cdot L_o \cdot C_o \cdot f^2} \end{aligned}$$

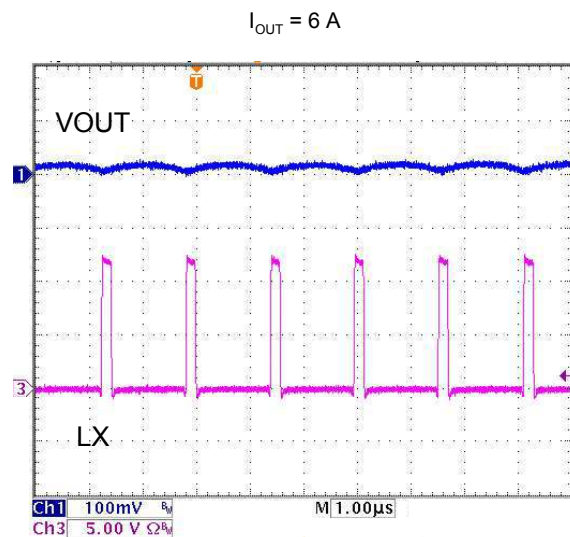
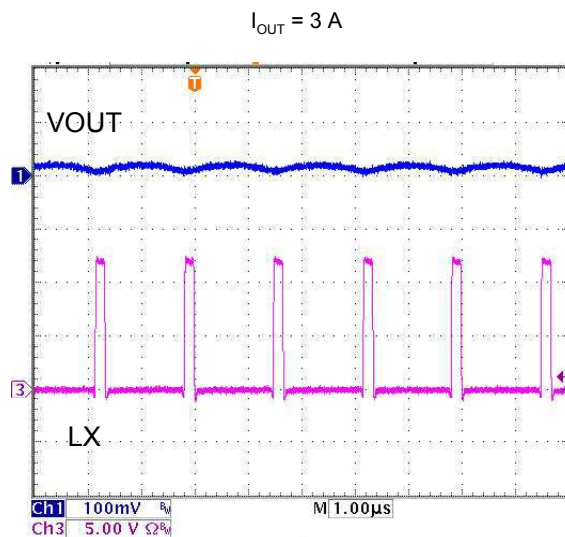
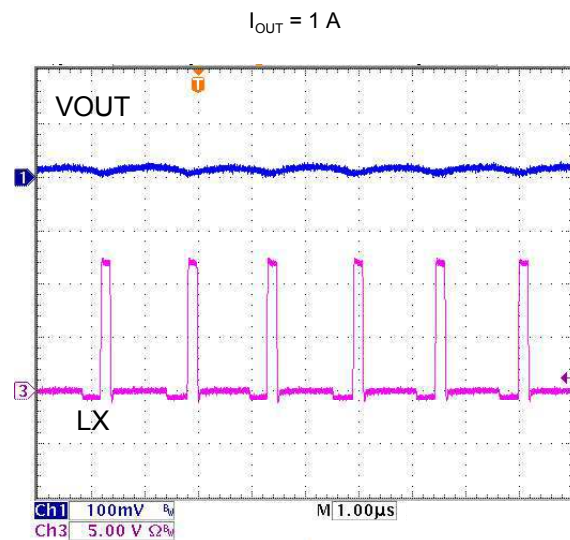
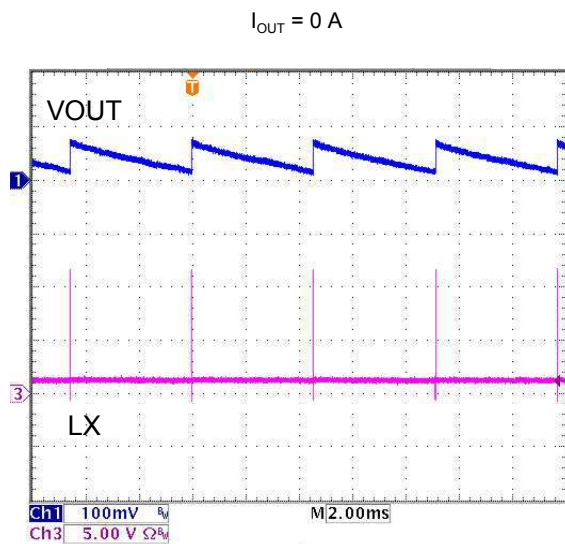
From the above equation, to achieve desired output ripple, low ESR ceramic capacitors are recommended, and its required RMS current rating is:

$$I_{c(rms)_max} = \frac{\Delta IL}{2\sqrt{3}} \quad @ E_i = E_{i_max}$$

TYPICAL CHARACTERISTICS CURVES

1. Output Ripple Voltage

Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V, Switching Frequency = 650 kHz, Skip Mode,
 $L_O = 1\ \mu\text{H}$, $C_O = 44\ \mu\text{F}$ (22 μF x 2)

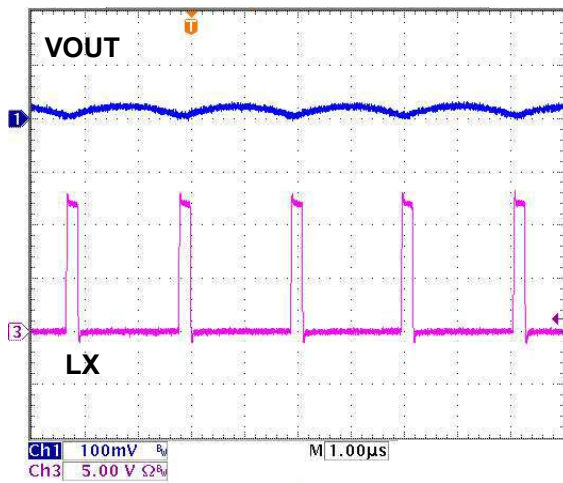


TYPICAL CHARACTERISTICS CURVES (Continued)

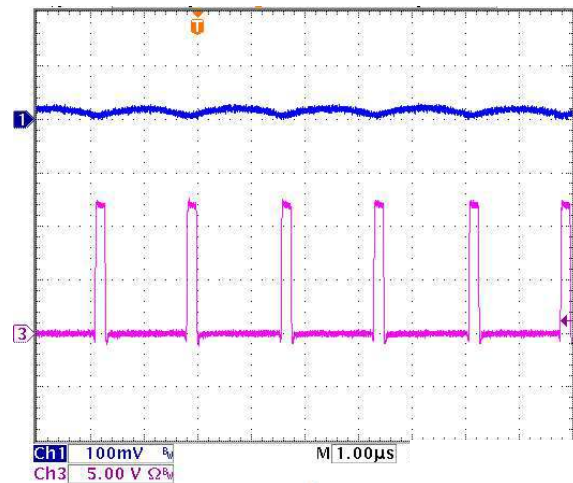
1. Output Ripple Voltage (Continued)

Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V, Switching Frequency = 650 kHz, FCCM,
 $L_O = 1\ \mu\text{H}$, $C_O = 44\ \mu\text{F}$ (22 μF x 2)

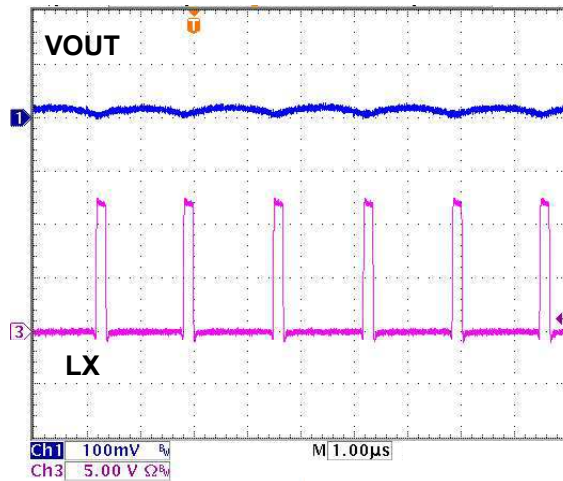
$I_{OUT} = 0\text{ A}$



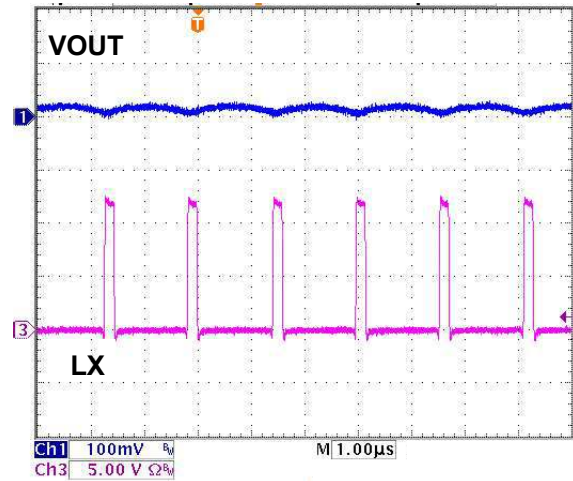
$I_{OUT} = 1\text{ A}$



$I_{OUT} = 3\text{ A}$



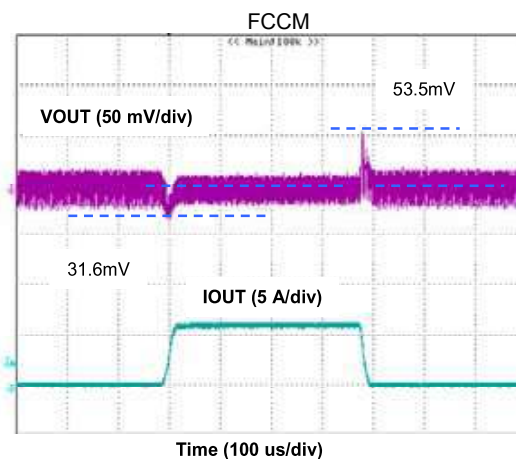
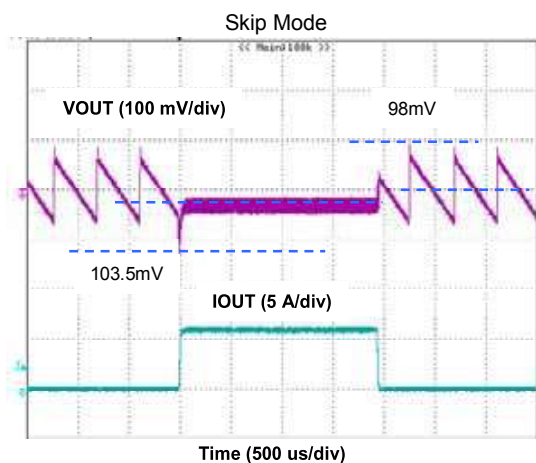
$I_{OUT} = 6\text{ A}$



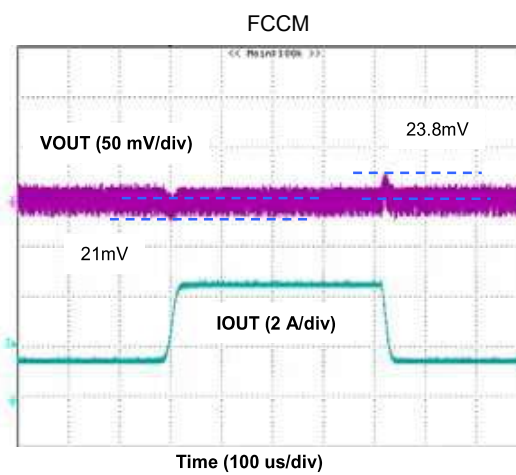
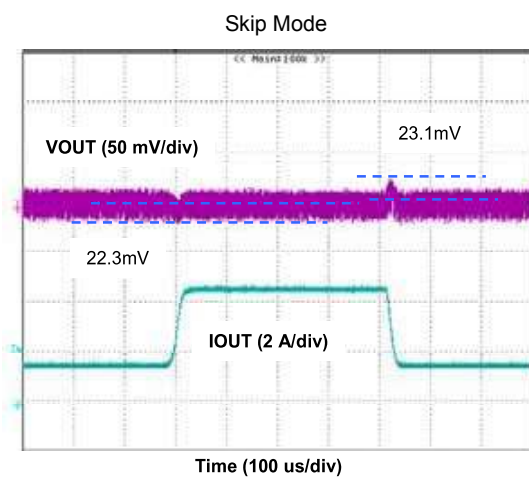
TYPICAL CHARACTERISTICS CURVES (Continued)

2. Load transient response

Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V , Switching Frequency = 430 kHz , $I_{OUT} = 10\text{ mA to }6\text{ A}$ ($0.5\text{ A} / \mu\text{s}$),
 $L_O = 1\text{ }\mu\text{H}$, $C_O = 44\text{ }\mu\text{F}$ ($22\text{ }\mu\text{F} \times 2$)



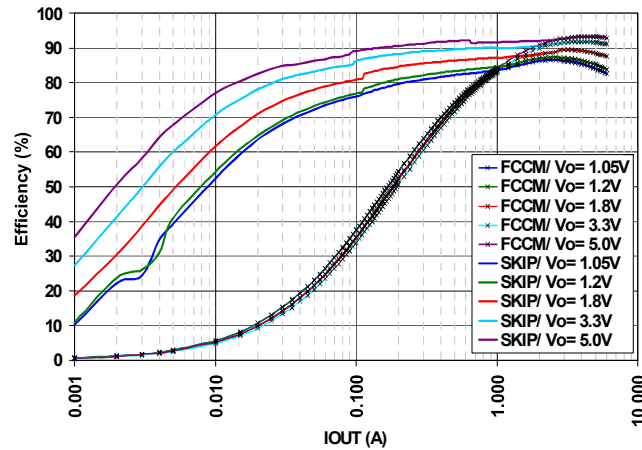
Condition : $V_{IN} = 12\text{ V}$, $V_{OUT} = 1.2\text{ V}$, Switching Frequency = 430 kHz , $I_{OUT} = 1.5\text{ A to }4.5\text{ A}$ ($0.15\text{ A} / \mu\text{s}$),
 $L_O = 1\text{ }\mu\text{H}$, $C_O = 44\text{ }\mu\text{F}$ ($22\text{ }\mu\text{F} \times 2$)



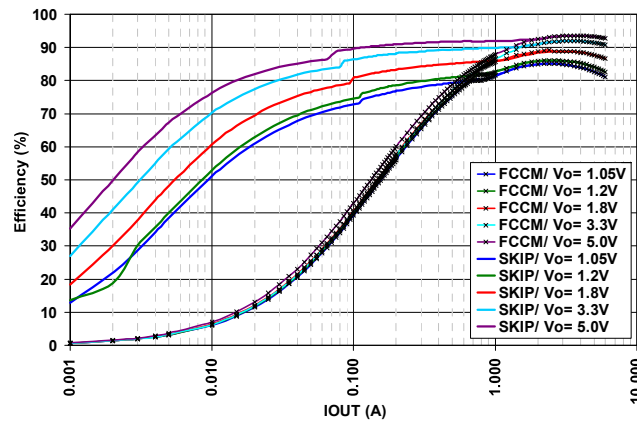
TYPICAL CHARACTERISTICS CURVES (Continued)

3. Efficiency

Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.05 V / 1.2 V / 1.8 V / 3.3 V / 5.0 V, Switching Frequency = 430 kHz,
 $L_O = 1\ \mu\text{H}$, $C_O = 44\ \mu\text{F}$ (22 $\mu\text{F} \times 2$)



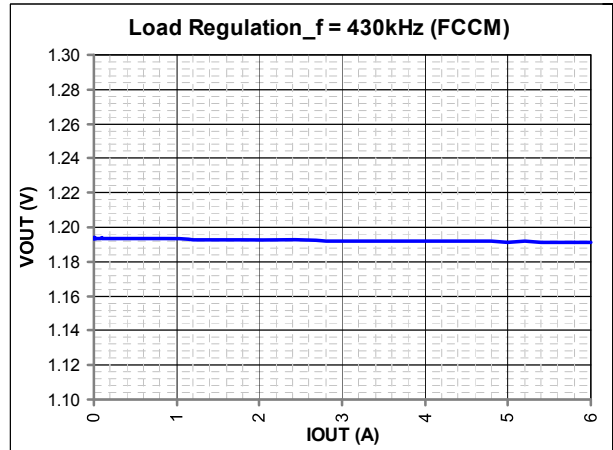
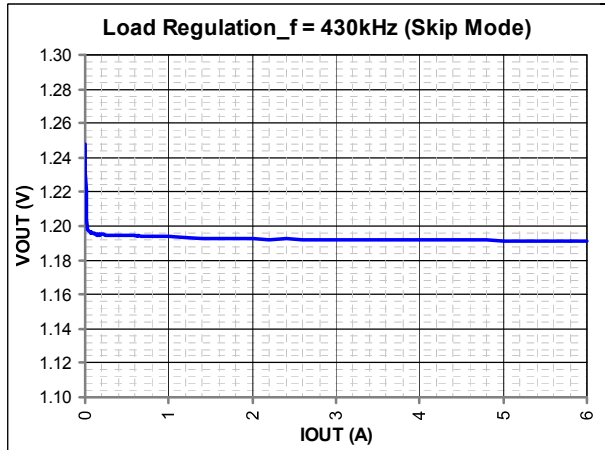
Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V / 1.2 V / 1.8 V / 3.3 V / 5.0 V, Switching Frequency = 650 kHz,
 $L_O = 1\ \mu\text{H}$, $C_O = 44\ \mu\text{F}$ (22 $\mu\text{F} \times 2$)



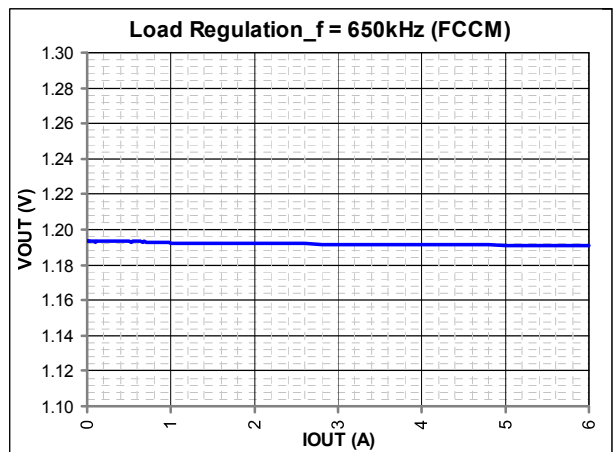
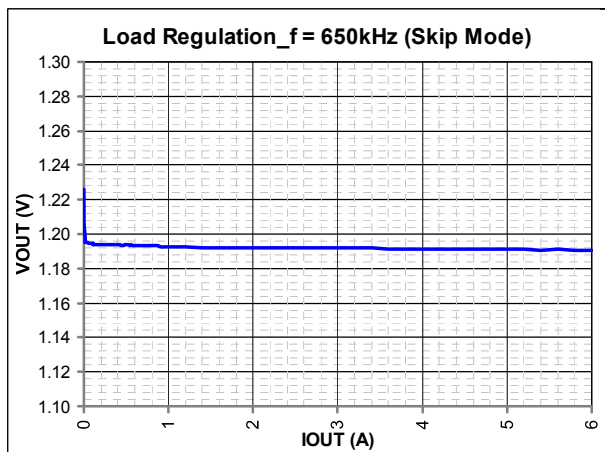
TYPICAL CHARACTERISTICS CURVES (Continued)

4. Load Regulation

Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V, Switching Frequency = 430 kHz, $L_O = 1\ \mu\text{H}$, $C_O = 44\ \mu\text{F}$ (22 μF x 2)



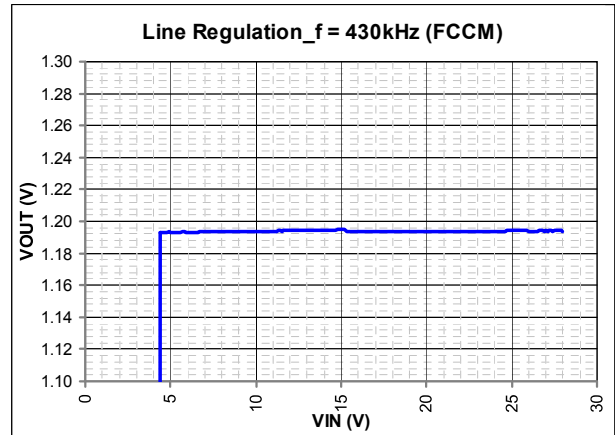
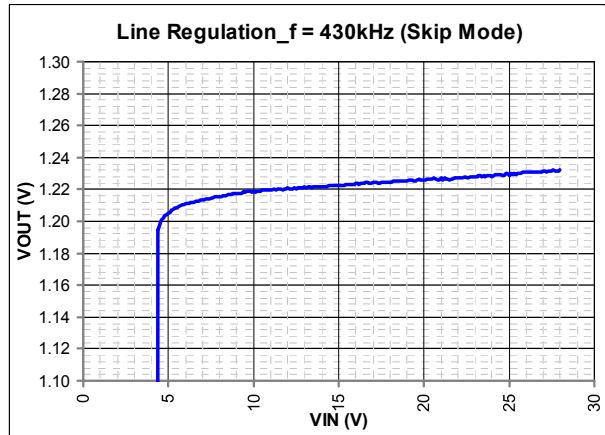
Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V, Switching Frequency = 650 kHz, $L_O = 1\ \mu\text{H}$, $C_O = 44\ \mu\text{F}$ (22 μF x 2)



TYPICAL CHARACTERISTICS CURVES (Continued)

5. Line Regulation

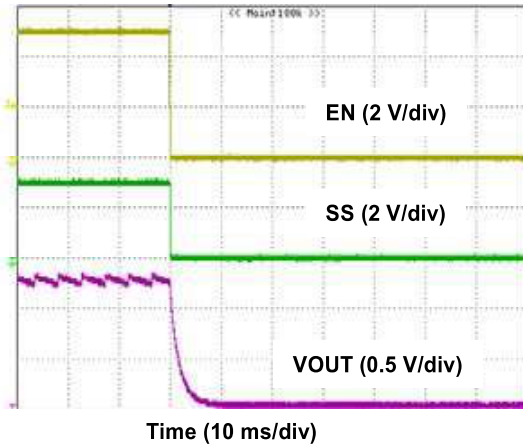
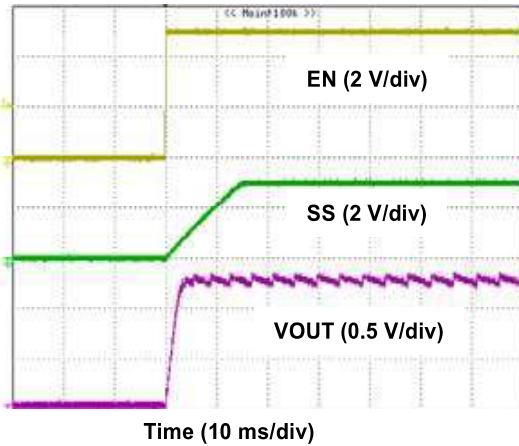
Condition : V_{OUT} Setting = 1.2 V, Switching Frequency = 430 kHz, I_{OUT} = 1.5 A, L_O = 1 μ H, C_O = 44 μ F (22 μ F x 2)



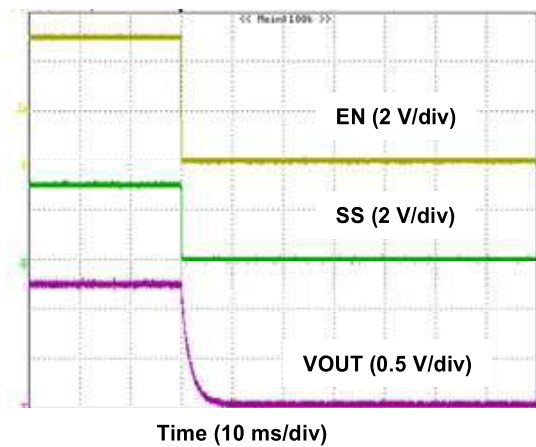
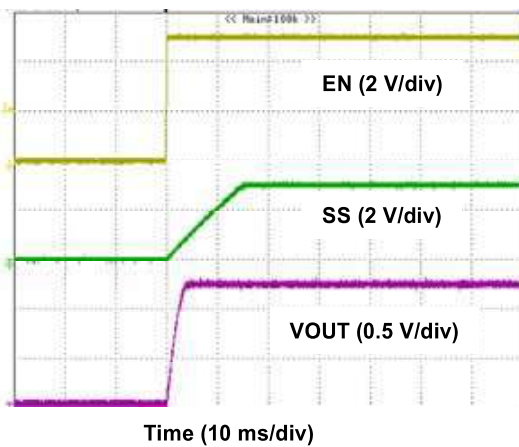
TYPICAL CHARACTERISTICS CURVES (Continued)

6. Start / Shut Down

Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V, Switching Frequency = 430 kHz, Skip Mode, $I_{OUT} = 0\text{ A}$,
 $L_O = 1\text{ }\mu\text{H}$, $C_O = 44\text{ }\mu\text{F}$ (22 μF x 2)



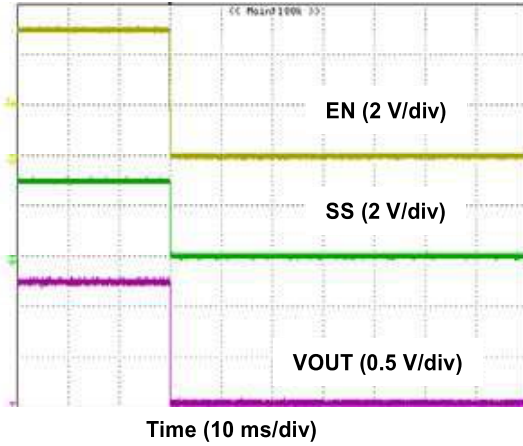
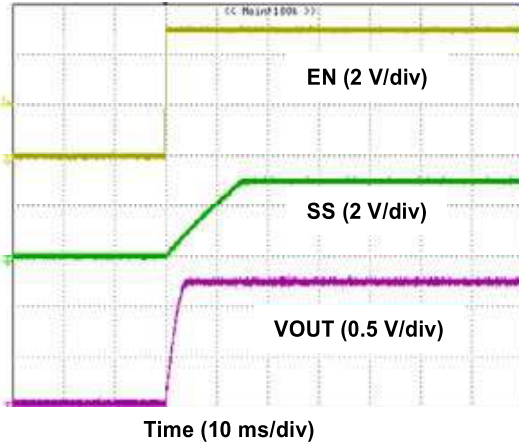
Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V, Switching Frequency = 430 kHz, FCCM, $I_{OUT} = 0\text{ A}$,
 $L_O = 1\text{ }\mu\text{H}$, $C_O = 44\text{ }\mu\text{F}$ (22 μF x 2)



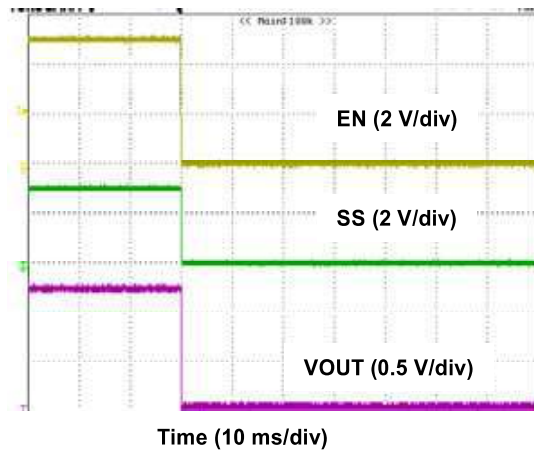
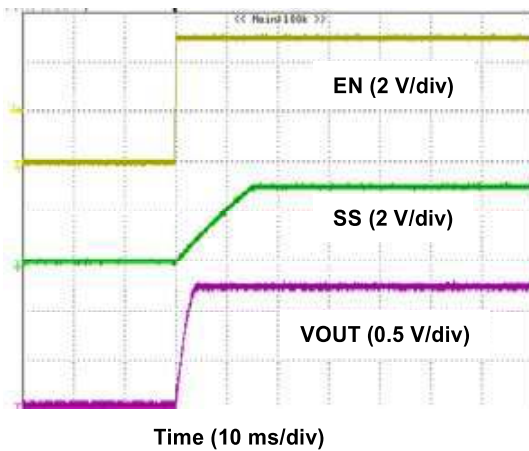
TYPICAL CHARACTERISTICS CURVES (Continued)

6. Start / Shut Down (Continued)

Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V, Switching Frequency = 430 kHz, Skip Mode, $R_{LOAD} = 0.5\ \Omega$,
 $L_O = 1\ \mu\text{H}$, $C_O = 44\ \mu\text{F}$ (22 $\mu\text{F} \times 2$)



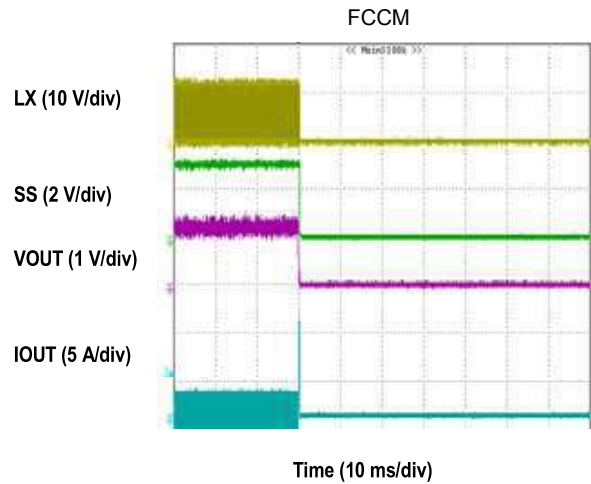
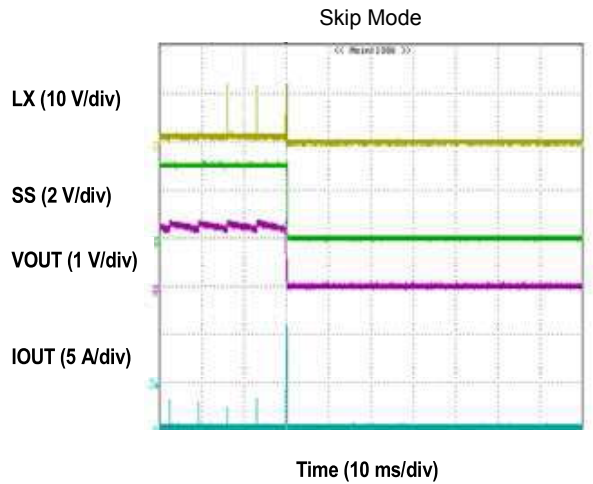
Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V, Switching Frequency = 430 kHz, FCCM, $R_{LOAD} = 0.5\ \Omega$,
 $L_O = 1\ \mu\text{H}$, $C_O = 44\ \mu\text{F}$ (22 $\mu\text{F} \times 2$)



TYPICAL CHARACTERISTICS CURVES (Continued)

7. Short Circuit Protection

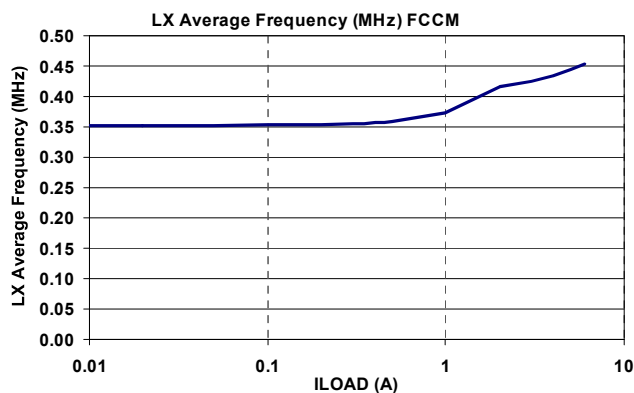
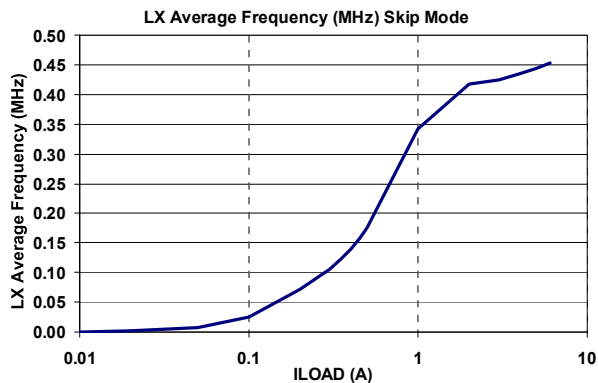
Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V, Switching Frequency = 430 kHz, $L_O = 1\ \mu\text{H}$, $C_O = 44\ \mu\text{F}$ (22 μF x 2)



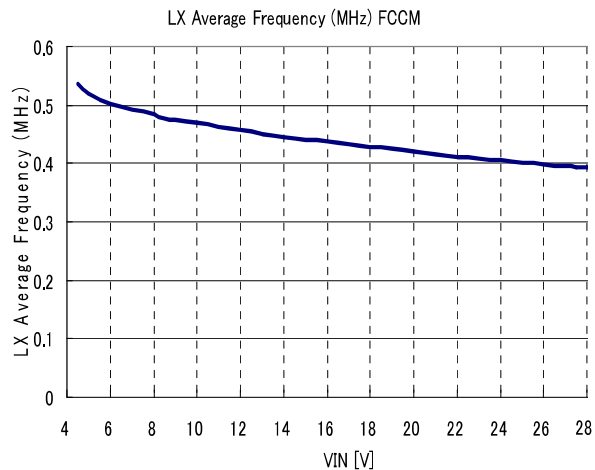
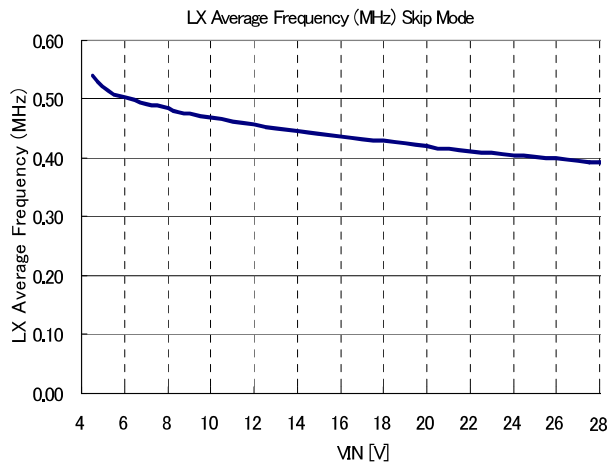
TYPICAL CHARACTERISTICS CURVES (Continued)

8. Switching Frequency

Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V, Switching Frequency = 430 kHz, $I_{OUT} = 10\text{ mA}$ to 6 A,
 $L_O = 1\text{ }\mu\text{H}$, $C_O = 44\text{ }\mu\text{F}$ (22 μF x 2)



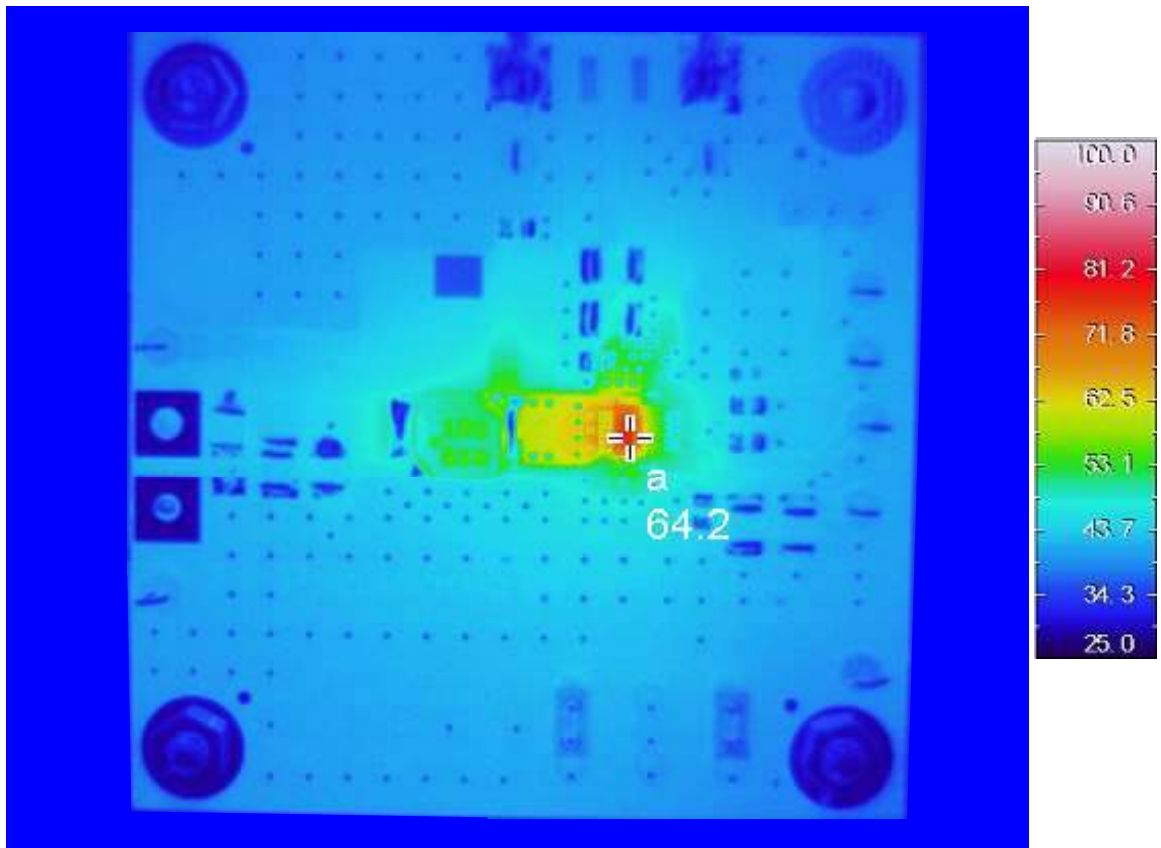
Condition : V_{OUT} Setting = 1.2 V, Switching Frequency = 430 kHz, $I_{OUT} = 6\text{ A}$, $L_O = 1\text{ }\mu\text{H}$, $C_O = 44\text{ }\mu\text{F}$ (22 μF x 2)



TYPICAL CHARACTERISTICS CURVES (Continued)

9. Thermal Performance

Condition : $V_{IN} = 12\text{ V}$, V_{OUT} Setting = 1.2 V, Switching Frequency = 430 kHz, FCCM, $I_{OUT} = 6\text{ A}$,
 $L_O = 1\text{ }\mu\text{H}$, $C_O = 44\text{ }\mu\text{F}$ (22 μF x 2)



APPLICATIONS INFORMATION (Continued)

2. Layout Recommendations

Board layout considerations are necessary for stable operation of the DC-DC regulator. The following precautions must be used when designing the board layout.

- (a) The Input capacitor C_{IN} must be placed in such a way that the distance between PVIN and PGND is minimum, in order to suppress the switching noise. Stray inductance and impedance should be reduced as indicated by loop (1) in the figure below.
- (b) A single point ground connection (2) must be used to connect PGND and AGND to improve operation stability.
- (c) Output current line I_{OUT} and the output sense line VOUT must have small common impedance to reduce output load variations. Output sense line VOUT must be close to the output condenser C_O as indicated by (3) below.
- (d) Power Loss and output ripple voltage can be reduced by placing the inductor L_O and output capacitor C_O such that the stray inductance and the impedance of loop (4) is minimum. This is realized by :
 - i) Minimizing distance between inductor L_O and LX pin.
 - ii) Reducing distance between output capacitor C_O and (2) / (3)
- (e) Thick lines in the application circuit example represent lines with large current flow. These lines should be designed as thick as possible.
- (f) VFB / SS / VREG lines should be placed far away from LX line, BST line and inductor L_O to reduce the effects of switching noise. These lines should be designed as short as possible. This is especially true for the VFB line, which is a high impedance line.
- (g) R_{FB1} / R_{FB2} should also be placed as far away as possible from LX line, BST line and inductor L_O to minimize the effects of switching noise. R_{FB1} / R_{FB2} should be placed close to the VFB pin.
- (h) LX / BST lines are noisy lines. They should be designed as short as possible.

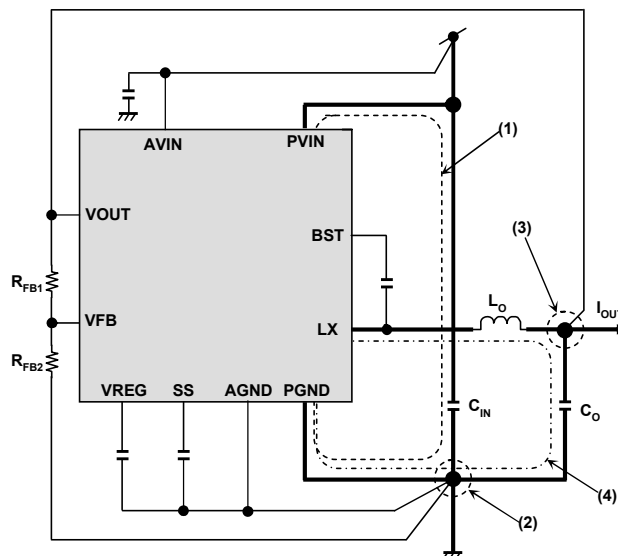


Figure : Application circuit diagram

Note : The application circuit diagram and layout diagram explained in this section, should be used as reference examples. The operation of the mass production set is not guaranteed. Sufficient evaluation and verification is required in the design of the mass production set. The Customer is fully responsible for the incorporation of the above illustrated application circuit and the information attached with it, in the design of the equipment.

APPLICATIONS INFORMATION (Continued)

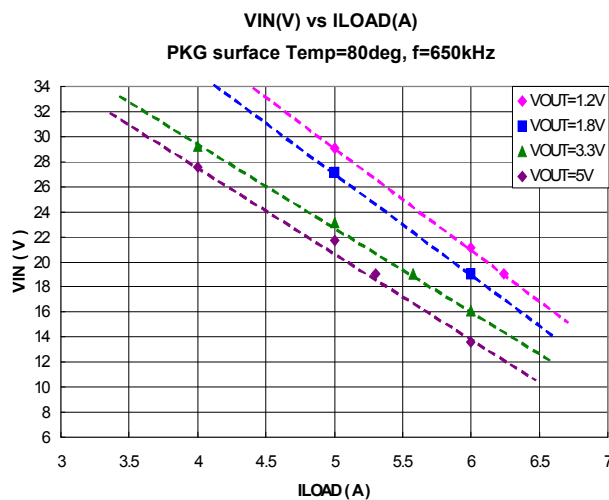
3. Recommended component

Reference Designator	QTY	Value	Manufacturer	Part Number	Note
C-AVIN1	2	10 μ F	TAIYO YUDEN	UMK325AB7106MM-T	—
C-AVIN2	1	0.1 μ F	Murata	GRM188R72A104KA35L	—
C-BST	1	0.1 μ F	Murata	GRM188R72A104KA35L	—
C-DCDCOUT	2	22 μ F	Murata	GRM32ER71E226KE15L	—
C-PVIN5	2	10 μ F	TAIYO YUDEN	UMK325AB7106MM-T	—
C-PVIN6	1	0.1 μ F	Murata	GRM188R72A104KA35L	—
C-SS	1	10 nF	Murata	GRM188R72A103KA01L	—
C-VREG	1	1.0 μ F	Murata	GRM188R71E105KA12L	—
L-LX	1	1.0 μ H	Panasonic	ETQP3W1R0WFN	FSEL : GND (430 kHz) OPEN (650 kHz)
		3.3 μ H	Panasonic	ETQP3W3R3WFN	FSEL : VREG (210 kHz)
R-FB1	1	1.5 k Ω	Panasonic	ERJ3EKF1501V	—
R-FB2	1	0	Panasonic	ERJ3GEY0R00V	—
R-FB3	1	1.5 k Ω	Panasonic	ERJ3EKF1501V	—
R-FB4	1	0	Panasonic	ERJ3GEY0R00V	—
R-PG	1	100 k Ω	Panasonic	ERJ3EKF1003V	—

4. Special attention and precaution in using

This IC is intended to be used for general electronic equipment. Ensure that the IC is used within the recommended safe operating region illustrated by the reference graph below. Do take note that thermal performance may varies with PCB design and PCB materials. You are encourage to use the graph only as a reference for your design and discuss further with our application engineer.

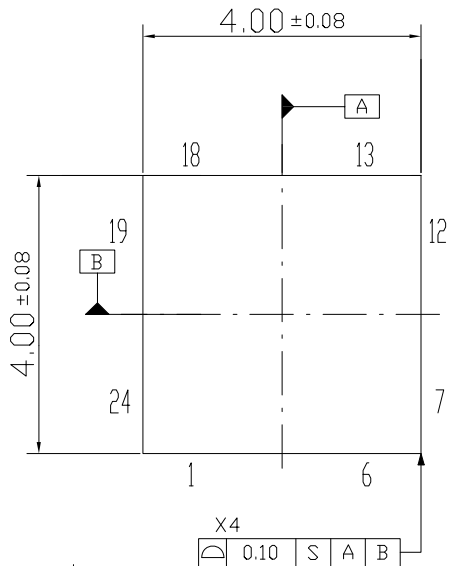
It is to be understood that our company shall not be held responsible for any damage incurred as a result of application beyond the recommended safe operating region.



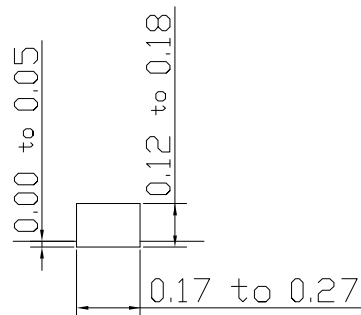
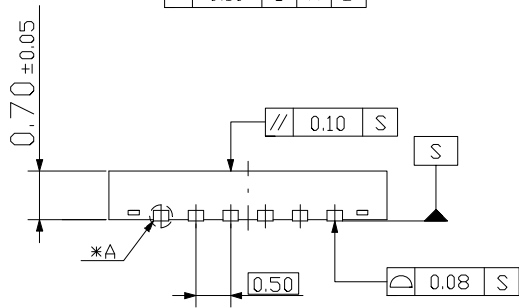
PACKAGE INFORMATION

Outline Drawing

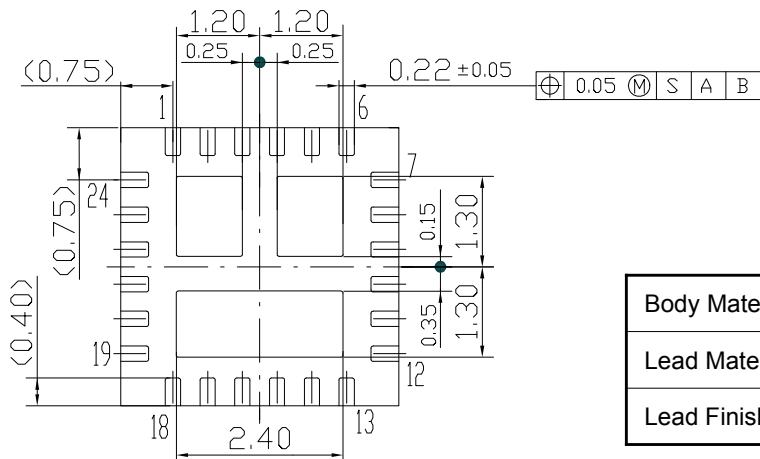
Package Code : HQFN024-A3-0404AZ



Unit : mm



Detail *A (Reference)



Body Material	: Br / Sb Free Epoxy Resin
Lead Material	: Cu Alloy
Lead Finish Method	: Pd Plating

IMPORTANT NOTICE

1. When using the IC for new models, verify the safety including the long-term reliability for each product.
2. When the application system is designed by using this IC, please confirm the notes in this book. Please read the notes to descriptions and the usage notes in the book.
3. This IC is intended to be used for general electronic equipment.
Consult our sales staff in advance for information on the following applications: Special applications in which exceptional quality and reliability are required, or if the failure or malfunction of this IC may directly jeopardize life or harm the human body. Any applications other than the standard applications intended.
 - (1) Space appliance (such as artificial satellite, and rocket)
 - (2) Traffic control equipment (such as for automotive, airplane, train, and ship)
 - (3) Medical equipment for life support
 - (4) Submarine transponder
 - (5) Control equipment for power plant
 - (6) Disaster prevention and security device
 - (7) Weapon
 - (8) Others : Applications of which reliability equivalent to (1) to (7) is requiredOur company shall not be held responsible for any damage incurred as a result of or in connection with the IC being used for any special application, unless our company agrees to the use of such special application. However, for the IC which we designate as products for automotive use, it is possible to be used for automotive.
4. This IC is neither designed nor intended for use in automotive applications or environments unless the IC is designated by our company to be used in automotive applications.
Our company shall not be held responsible for any damage incurred by customers or any third party as a result of or in connection with the IC being used in automotive application, unless our company agrees to such application in this book.
5. Please use this IC in compliance with all applicable laws and regulations that regulate the inclusion or use of controlled substances, including without limitation, the EU RoHS Directive. Our company shall not be held responsible for any damage incurred as a result of our IC being used by our customers, not complying with the applicable laws and regulations.
6. Pay attention to the direction of the IC. When mounting it in the wrong direction onto the PCB (printed-circuit-board), it might be damaged.
7. Pay attention in the PCB (printed-circuit-board) pattern layout in order to prevent damage due to short circuit between pins. In addition, refer to the Pin Description for the pin configuration.
8. Perform visual inspection on the PCB before applying power, otherwise damage might happen due to problems such as solder-bridge between the pins of the IC. Also, perform full technical verification on the assembly quality, because the same damage possibly can happen due to conductive substances, such as solder ball, that adhere to the IC during transportation.
9. Take notice in the use of this IC that it might be damaged when an abnormal state occurs such as output pin-VCC short (Power supply fault), output pin-GND short (Ground fault), or output-to-output-pin short (load short). Safety measures such as installation of fuses are recommended because the extent of the above-mentioned damage will depend on the current capability of the power supply.
10. The protection circuit is for maintaining safety against abnormal operation. Therefore, the protection circuit should not work during normal operation.
Especially for the thermal protection circuit, if the area of safe operation or the absolute maximum rating is momentarily exceeded due to output pin to VCC short (Power supply fault), or output pin to GND short (Ground fault), the IC might be damaged before the thermal protection circuit could operate.
11. Unless specified in the product specifications, make sure that negative voltage or excessive voltage are not applied to the pins because the IC might be damaged, which could happen due to negative voltage or excessive voltage generated during the ON and OFF timing when the inductive load of a motor coil or actuator coils of optical pick-up is being driven.
12. Product which has specified ASO (Area of Safe Operation) should be operated in ASO
13. Verify the risks which might be caused by the malfunctions of external components.
14. Connect the metallic plates (fins) on the back side of the IC with their respective potentials (AGND, PVIN, LX). The thermal resistance and the electrical characteristics are guaranteed only when the metallic plates (fins) are connected with their respective potentials.

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