FemtoClock[®] NG Crystal-to-3.3V, 2.5V LVPECL/LVCMOS Clock Generator with Fanout Buffer

8T49N012

DATA SHEET

General Description

The 8T49N012 is a high performance Clock Generator with selectable LVPECL or Single-ended outputs. The 8T49N012 can generate selectable frequencies from a crystal or a single-ended reference clock. The frequency is selected from the Frequency Selection Table.

Excellent phase noise performance is maintained with IDT's Fourth Generation FemtoClock[®] NG PLL technology.

Pin Assignment

56-Lead VFQFN, 8.0mm x 8.0mm x 0.9mm

Features

- Fourth Generation FemtoClock NG PLL technology
- Three differential output banks: **Bank A:** selectable between four pairs of LVPECL or four pairs of complementary LVCMOS/LVTTL outputs **Bank B:** two pairs of LVPECL outputs **Bank C:** six pairs of LVPECL outputs
- Selectable clock input or crystal input.
- Supports 25MHz fundamental crystal or 25MHz, 50MHz, 66.67MHz clock input
- Selectable 156.25MHz, 125MHz, 100MHz clock for Bank B and Bank C outputs
- Selectable 156.25MHz, 125MHz, 100MHz, 250MHz, 312.5MHz, 50MHz, 25MHz, 62.5MHz or 78.125MHz for Bank A outputs
- PLL lock indication (LVCMOS output)
- RMS phase jitter at 156.25MHz (12kHz 20MHz): 0.199ps (typical)
- Power supply modes: Core / Output 3.3V / 3.3V 3.3V / 2.5V 2.5V / 2.5V
- -40°C to 85°C ambient operating temperature
- 56-Lead VFQFN
- Lead-free (RoHS 6) packaging

RENESAS

Block Diagram

Pin Description and Pin Characteristic Tables

Table 1. Pin Descriptions¹

Table 1. Pin Descriptions¹ (Continued)

NOTE 1: Pulldown refers to internal input resistors. See Table 2, Pin Characteristics, for typical values.

Table 2. Pin Characteristics

Function Configuration Tables

Table 3A. FB_DIV Function Table

Table 3B. SE Function Table

Table 3D. LOCK indicaiotn Table

٦

Table 3E. nQE_B Function Table

High High Impedance

Table 3F. nOE_C Function Table

Table 3C. XTAL_SEL Function Table nOE_C Bank C Output Low (default) $\qquad \qquad$ LVPECL Mid (Reserved) and Reserved

Table 3G. Bank B and C Output Divider Table

Table 3H. Bank A Output Divider table

Absolute Maximum Ratings

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of the product at these conditions or any conditions beyond those listed in the DC [Electrical](#page-6-0) [Characteristics](#page-6-0) or AC Electrical [Characteristics](#page-9-0) is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

DC Electrical Characteristics

 $\bm{\textsf{Table 4A.} }$ Power Supply DC Characteristics, $\bm{\mathsf{V}}_{\bm{\text{CC}}}$ = $\bm{\mathsf{V}}_{\bm{\text{CC}}-X}$ = $\bm{\mathsf{V}}_{\bm{\text{CCO}}-X}$ 1 = 3.3V \pm 5%, $\bm{\mathsf{V}}_{\bm{\text{EE}}}$ = 0V, $\bm{\mathsf{T}}_{\bm{\mathsf{A}}}$ = -40°C to 85°C

NOTE 1: $V_{\text{CCO_X}}$ denotes $V_{\text{CCO_A}}$, $V_{\text{CCO_B}}$ and $V_{\text{CCO_C}}$,

$\bm{\rm Table}$ 4B. Power Supply DC Characteristics, $\rm V_{CC}$ = V \rm_{CC_X} = V $\rm_{CCO_X}^{-1}$ = 2.5V \pm 5%, V \rm_{EE} = 0V, T_A = -40°C to 85°C

NOTE 1: $V_{\text{CCO_X}}$ denotes $V_{\text{CCO_A}}$, $V_{\text{CCO_B}}$ and $V_{\text{CCO_C}}$,

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{\rm CC}$	Core Supply Voltage		3.135	3.3	3.465	v
$V_{CC_}$	XTAL Supply Voltage		3.135	3.3	3.465	v
V_{CCA}	Analog Supply Voltage		3.135	3.3	3.465	v
V_{CCO_X}	Output Supply Voltage		2.375	2.5	2.625	v
ICCA	Analog Supply Current			42	50	mA
LEE	Power Supply Current	Unterminated Outputs		323	373	mA

<code>Table 4C. Power Supply DC Characteristics, V $_{\rm CC}$ = V $_{\rm CC_X}$ = 3.3V \pm 5%, V $_{\rm CCO_X}$ ¹ = 2.5V \pm 5%, V $_{\rm EE}$ = 0V, T $_{\sf A}$ = -40°C to 85°C</code>

NOTE 1: V_{CCO_X} denotes V_{CCO_A} , V_{CCO_B} and V_{CCO_C} ,

Table 4D. 2-Level LVCMOS/LVTTL DC Characteristics, T_A = -40°C to 85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
	Input High Voltage		$V_{\text{CC}} = 3.3V$	$\overline{2}$		$V_{\rm CC}$ + 0.3	V
V_{IH}			V_{CC} = 2.5V	1.7		$V_{\rm CC}$ + 0.3	V
	Input Low Voltage		$V_{\rm CC}$ = 3.3V	-0.3		0.8	v
V_{IL}			$V_{\rm CC}$ = 2.5V	-0.3		0.7	v
I_{IH}	Input High Current	CLK_IN, XTAL_SEL	$V_{\text{CC}} = V_{\text{IN}} = 3.465V$ or 2.625V			150	μA
$I_{\rm IL}$	Input Low Current	CLK_IN, XTAL_SEL	V_{CC} = 3.465V or 2.625V, V_{IN} = 0V	-5			μA
V_{OH}	Output High Voltage	QA[0:3]_CMOS, nQA[0:3]_CMOS, LOCK	$V_{CCO A}$ = 3.465V; I_{OH} = -8mA	2.6			v
V_{OL}	Output Low Voltage	QA[0:3]_CMOS, nQA[0:3]_CMOS, LOCK	$V_{CCO A}$ = 3.465V; I_{OL} = 8mA			0.6	v

Table 4E. 3-Level LVCMOS/LVTTL DC Characteristics, T_A = -40°C to 85°C

Table 4F. LVPECL DC Characteristics, $\rm V_{CCO_X}^1$ = 3.3V±5%, V_{EE} = 0V, T_A = -40°C to 85°C

NOTE 1: V_{CCO_X} denotes V_{CCO_A} , V_{CCO_B} and V_{CCO_C} ,

NOTE 2: Outputs termination with 50 Ω to V_{CCO} $_X$ – 2V.

Table 4G. LVPECL DC Characteristics, $V_{\text{CCO_X}}{}^1$ = 2.5V \pm 5%, V_{EE} = 0V, T_A = -40°C to 85°C

NOTE 1: $V_{\text{CCO_X}}$ denotes $V_{\text{CCO_A}}$, $V_{\text{CCO_B}}$ and $V_{\text{CCO_C}}$,

NOTE 2: Outputs termination with 50 Ω to V_{CCO_X} – 2V.

. Table 5. Crystal Characteristics

AC Electrical Characteristics

Table 6. AC Characteristics, $V_{CC} = V_{CC_X} = V_{CCO_X}^{-1} = 3.3V \pm 5\%$ or 2.5V \pm 5%, or $V_{CC} = V_{CC_X} = 3.3V \pm 5\%, V_{CCO_X}^{-1} = 2.5V \pm 5\%, V_{EE} = 0V, T_A = -40^{\circ}\text{C}$ to 85°C

NOTE 1: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 2: Characterized using XTAL input.

NOTE 3: Defined as skew between outputs at the same supply voltage and with equal load conditions. Measured at the differential crosspoints.

NOTE 4: These parameters are guaranteed by characterization.

Typical Phase Noise at 156.25MHz

Parameter Measurement Information

LVCMOS Output Duty Cycle/Pulse Width/Period

2.5V Core/2.5V QAx_CMOS LVPECL Output Load Test Circuit

2.5V/2.5V QAx_CMOS Output Load Test Circuit

Output Rise/Fall Time

Parameter Measurement Information, continued

LVPECL Output Rise/Fall Time

Differential Output Duty Cycle

Applications Information

Recommendations for Unused Input and Output Pins

Inputs:

Crystal Inputs

For applications not requiring the use of the crystal oscillator input, both XTAL_IN and XTAL_OUT can be left floating. Though not required, but for additional protection, a $1k\Omega$ resistor can be tied from XTAL_IN to ground.

CLK Input

For applications not requiring the use of a clock input, it can be left floating. Though not required, but for additional protection, a $1k\Omega$ resistor can be tied from the CLK input to ground.

LVCMOS Control Pins

All control pins have internal pullup or pulldown resistors; additional resistance is not required but can be added for additional protection. A 1 $k\Omega$ resistor can be used.

Outputs:

LVPECL Outputs

All unused LVPECL output pairs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

LVCMOS Outputs

All unused LVCMOS outputs can be left floating. There should be no trace attached.

VFQFN EPAD Thermal Release Path

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in *[Figure](#page-14-0) 1*. The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The vias act as "heat pipes". The number of vias (i.e. "heat pipes") are application specific

and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, please refer to the Application Note on the Surface Mount Assembly of Amkor's Thermally/ Electrically Enhance Lead frame Base Package, Amkor Technology.

Figure 1. P.C. Assembly for Exposed Pad Thermal Release Path – Side View (drawing not to scale)

Overdriving the XTAL Interface

The XTAL_IN input can be overdriven by an LVCMOS driver or by one side of a differential driver through an AC coupling capacitor. The XTAL_OUT pin can be left floating. The amplitude of the input signal should be between 500mV and 1.8V and the slew rate should not be less than 0.2V/ns. For 3.3V LVCMOS inputs, the amplitude must be reduced from full swing to at least half the swing in order to prevent signal interference with the power rail and to reduce internal noise. [Figure](#page-15-0) 2A shows an example of the interface diagram for a high speed 3.3V LVCMOS driver. This configuration requires that the sum of the output impedance of the driver (Ro) and the series resistance (Rs) equals the transmission line impedance. In addition, matched termination at the crystal input will attenuate the signal in half. This

can be done in one of two ways. First, R1 and R2 in parallel should equal the transmission line impedance. For most 50Ω applications, R1 and R2 can be 100 Ω . This can also be accomplished by removing R1 and changing R2 to 50 Ω . The values of the resistors can be increased to reduce the loading for a slower and weaker LVCMOS driver. [Figure](#page-15-1) 2B shows an example of the interface diagram for an LVPECL driver. This is a standard LVPECL termination with one side of the driver feeding the XTAL_IN input. It is recommended that all components in the schematics be placed in the layout. Though some components might not be used, they can be utilized for debugging purposes. The datasheet specifications are characterized and guaranteed by using a quartz crystal as the input.

Figure 2A. General Diagram for LVCMOS Driver to XTAL Input Interface

Figure 2B. General Diagram for LVPECL Driver to XTAL Input Interface

Termination for 3.3V LVPECL Outputs

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

The differential outputs are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive 50Ω

Figure 3A. 3.3V LVPECL Output Termination

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. Figure 3A shows a layout that is recommended only as a guideline. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

RENESAS

Termination for 2.5V LVPECL Outputs

[Figure](#page-17-0) 4A and [Figure](#page-17-1) 4B show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating 50Ω to $V_{\text{CCO}} - 2V$. For $V_{\text{CCO}} = 2.5V$, the $V_{\text{CCO}} - 2V$ is very close to ground

Figure 4A. 2.5V LVPECL Driver Termination Example

Figure 4B. 2.5V LVPECL Driver Termination Example

level. The R3 in [Figure](#page-17-1) 4B can be eliminated and the termination is shown in [Figure](#page-17-2) 4C.

Figure 4C. 2.5V LVPECL Driver Termination Example

Schematic Example

[Figure](#page-19-0) ⁵ (next page) is an 8T49N012 application example schematic. The schematic focuses on functional connections and is not configuration specific. Refer to the pin description and functional tables in the datasheet to ensure that the logic control inputs are properly set.

In this example the device is operated at all V_{CC} power pins = 3.3V. The Bank A outputs are configured for LVCMOS by pulling SE high. Three different examples of LVPECL terminations are shown for the outputs to demonstrate less common termination options.

Crystal layout is very important to minimize capacitive coupling between the crystal pads and leads and other metal in the circuit board. Capacitive coupling to other conductors has two adverse effects. The first is that it reduces the oscillator frequency, leaving less tuning margin. Second, noise on power planes and logic transitions on signal traces can pull the phase of voltages on the XTAL_IN and XTAL_OUT pins of the oscillator.

Using a crystal on the top layer as an example, void all signal and power layers under the crystal, XTAL_IN, XTAL_OUT and the input pins of the 8T49N012 between the top layer and the ground plane for the 8T49N012. If the ground plane for the 8T49N012 is the first layer under the crystal and the parasitic capacity of the traces and pads is excessive, then void enough power and signal planes to minimize the coupling capacity to the first ground plane. Ensure that the ground under the crystal is the same ground as used for the tuning caps and the oscillator.

As with any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. The 8T49N012 provides separate power supply pins to isolate any high switching noise from coupling into the internal PLL.

In order to achieve the best possible filtering, it is recommended that the placement of the filter components be on the device side of the PCB as close to the power pins as possible. The 0.1µF capacitors in each power pin filter must be placed on the device side. If space is limited, the other components can be on the opposite side of the PCB. Power supply filter recommendations are a general guideline to be used for reducing external noise from coupling into the devices.

The V_{CC} and V_{CCO} LC filters start to attenuate noise at approximately 10kHz. If a specific frequency noise component is known, such as switching power supplies frequencies, it is recommended that component values be adjusted and if required, additional filtering be added. Additionally, good general design practices for power plane voltage stability suggests adding bulk capacitance in the local area of all devices.

Figure 5. 8T49N012 Schematic Example

Power Considerations

This section provides information on power dissipation and junction temperature for the 8T49N012I. Equations and example calculations are also provided.

1. Power Dissipation

The total power dissipation for the8T49N012 is the sum of the core power plus the output power dissipated due to the loading. The following is the power dissipation for $V_{CC} = 3.3V + 5\% = 3.465V$, which gives worst case results.

 I_{FF} = 344mA with QA Bank in LVCMOS mode is the worst case scenario.

- Power (core)_{MAX} = $V_{\text{CC_MAX}}$ * I_{EE} = 3.465V * 0.345A = **1.195W**
- Power $(LVPECL)_{MAX} = 31$ mW per output
- Total Power (LVPECL) $_{MAX} = 31$ mW $* 8 = 0.248W$

LVCMOS Dynamic Power Dissipation

• Dynamic Power at 312.5MHz:

 $P = C_{PD}$ * Freq * $(V_{DDO})^2 = 6.5pF$ * 312.5MHz * $(3.465V)^2$ $P = 0.024W$ per output Total Power (CPD) = $0.024W * 8 = 0.195W$

Total Power Dissipation

- **Total Powe**r
	- = Power (core) + Total Power (LVPECL) + Total Power (CPD)
	- $= 1.195W + 0.248W + 0.195W$
	- = **1.638W**

2. Junction Temperature

Junction temperature, T_J, is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, T_J, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: $T_J = \theta_{JA} * Pd_total + T_A$

 T_J = Junction Temperature

 θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

 T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{AB} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 23.1 per [Table](#page-20-0) 7 below. °C/W per Table 6 below.

Therefore, T_J for an ambient temperature of 85°C with all outputs switching is:

85°C + 1.638W * 23.1°C/W = 122.8°C. This is below the limit of 125°C.

This calculation is only an example. T_J will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 7. Thermal Resistance θ_{JA} for 56-Lead VFQFN, Forced Convection

3. Calculations and Equations.

The purpose of this section is to calculate the power dissipation for the LVPECL output pair.

The LVPECL output driver circuit and termination are shown in [Figure](#page-21-0) 6.

Figure 6. LVPECL Driver Circuit and Termination

To calculate power dissipation per output pair due to loading, use the following equations which assume a 50Ω load, and a termination voltage of V_{CCO} – 2V.

- For logic high, $V_{\text{OUT}} = V_{\text{OH_MAX}} = V_{\text{CCO_MAX}} 0.7V$ $(V_{\text{CCO_MAX}} - V_{\text{OH_MAX}}) = 0.7V$
- \bullet For logic low, $\mathsf{V}_{\mathsf{OUT}}$ = $\mathsf{V}_{\mathsf{OL_MAX}}$ = $\mathsf{V}_{\mathsf{CCO_MAX}}$ **1.6V** (VCCO_MAX **–** VOL_MAX) = **1.6V**

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

Pd_H = [(V_{OH_MAX} – (V_{CCO_MAX} – 2V))/R_L] * (V_{CCO_MAX} – V_{OH_MAX}) = [(2V – (V_{CCO_MAX} – V_{OH_MAX}))/R_L] * (V_{CCO_MAX} – V_{OH_MAX}) = $[(2V – 0.7V)/50 Ω] * 0.7V = **18.2mW**$

Pd_L = [(V_{OH_MAX} – (V_{CCO_MAX} – 2V))/R_L] * (V_{CCO_MAX} – V_{OL_MAX}) = [(2V – (V_{CCO_MAX} – V_{OL_MAX}))/R_L] * (V_{CCO_MAX} – V_{OL_MAX}) = $[(2V-1.6V)/50\Omega] * 1.6V = 12.8mW$

Total Power Dissipation per output pair = Pd_H + Pd_L = **31.00mW**

Reliability Information

Table 8. θ_{JA} vs. Air Flow Table for a 56-Lead VFQFN

Transistor Count

The transistor count for 8T49N012 is 176,638

RENESAS

56-Lead VFQFN Package Outline and Package Dimension

RENESAS

56-Lead VFQFN Package Outline and Package Dimensions, continued

56-Lead VFQFN Package Outline and Package Dimensions, continued

Ordering Information

Table 9. Ordering Information

IMPORTANT NOTICE AND DISCLAIMER

RENESAS ELECTRONICS CORPORATION AND ITS SUBSIDIARIES ("RENESAS") PROVIDES TECHNICAL SPECIFICATIONS AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS OR IMPLIED, INCLUDING, WITHOUT LIMITATION, ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE, OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for developers skilled in the art designing with Renesas products. You are solely responsible for (1) selecting the appropriate products for your application, (2) designing, validating, and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, or other requirements. These resources are subject to change without notice. Renesas grants you permission to use these resources only for development of an application that uses Renesas products. Other reproduction or use of these resources is strictly prohibited. No license is granted to any other Renesas intellectual property or to any third party intellectual property. Renesas disclaims responsibility for, and you will fully indemnify Renesas and its representatives against, any claims, damages, costs, losses, or liabilities arising out of your use of these resources. Renesas' products are provided only subject to Renesas' Terms and Conditions of Sale or other applicable terms agreed to in writing. No use o any Renesas resources expands or otherwise alters any applicable warranties or warranty disclaimers for these products.

(Disclaimer Rev.1.0 Mar 2020)

Corporate Headquarters

TOYOSU FORESIA, 3-2-24 Toyosu,

Koto-ku, Tokyo 135-0061, Japan

Koto-ku, Tokyo 135-0061, Japan www.renesas.com office, please visit:

Trademarks

Renesas and the Renesas logo are trademarks of Renesas Electronics Corporation. All trademarks and registered trademarks are the property of their respective owners.

For further information on a product, technology, the most up-to-date version of a document, or your nearest sales www.renesas.com/contact/