



#### 2.5 GHz 1:4 LVDS Fanout Buffer with Internal Termination

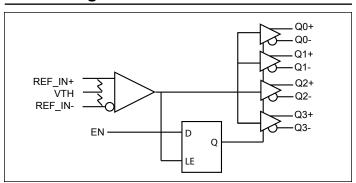
#### **Features**

- $F_{MAX} < 2.5GHz$
- 4 pairs of differential LVDS outputs
- Low additive jitter, < 0.05ps (max)
- Input CLK accepts: LVDS, LVDS, CML, SSTL input level
- Output to Output skew: <20ps
- Operating Temperature: -40°C to 85°C
- Power supply:  $3.3V \pm 10\%$  or  $2.5V \pm 5\%$
- Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)
- Halogen and Antimony Free. "Green" Device (Note 3)
- For automotive applications requiring specific change control (i.e. parts qualified to AEC-Q100/101/200, PPAP capable, and manufactured in IATF 16949 certified facilities), please contact us or your local Diodes representative.
  - https://www.diodes.com/quality/product-definitions/
- Packaging (Pb-free & Green)
  - □ 16-pin, TQFN (ZH)

## **Description**

The PI6C5922504 is a high-performance low-skew 1-to-4 LVDS fanout buffer. The CLK inputs accept LVPECL, LVDS, CML and SSTL signals. PI6C5922504 is ideal for clock distribution applications such as providing fanout for low noise Diodes oscillators.

## **Block Diagram**



www.diodes.com

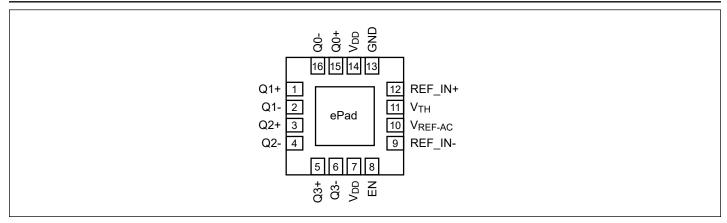
#### Notes:

- 1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant.
- 2. See https://www.diodes.com/quality/lead-free/ for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
- 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.





# **Pin Configuration**



# **Pin Description**

Pin #	Name	Type	Description	
1, 2	Q1+, Q1-	Output	Differential output pair, LVDS interface level.	
3, 4	Q2+, Q2-	Output	Differential output pair, LVDS interface level.	
5, 6	Q3+, Q3-	Output	Differential output pair, LVDS interface level.	
7, 14	V <sub>DD</sub>	Power	Core Power Supply	
8	EN	Input	Synchronous Output Enable, with internal $25k\Omega$ pull-up resistor. Logic high selects enable, and logic low selects disable.	
9	REF_IN-	Input	Differential IN negative input, AC and DC coupled	
10	V <sub>REF-AC</sub>	Output	Reference Voltage: Biased to $V_{\mathrm{DD}}$ -1.4V. Used when AC coupling inputs	
11	$V_{TH}$	Output	Differential pair IN center-tap node. Tie to VREF-AC for AC Coupled inputs.	
12	REF_IN+	Input	Differential IN positive input, AC and DC coupled	
13	GND	Power	Ground	
15, 16	Q0+, Q0-	Output	Differential output pair, LVDS interface level.	
_	ePad	Power	Connect to GND	

# **Functional Description**

REF_IN+	REF_IN-	EN	Q+	Q-
0	1	1	0	1
1	0	1	1	0
X	X	0	0	1





## **Maximum Ratings**

(Over operating free-air temperature range)

Storage Temperature	65°C to+155°C
Junction Temperature	Max. 125°C
3.3V Core Supply Voltage	-0.5 to +4.6V
ESD Protection (HBM)	2000V

#### Note:

Stresses greater than those listed under MAXIMUM RATINGS may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

## **DC Characteristics**

Symbol	Parameter	Conditions	Min	Тур	Max	Units
37	D. C. L.W.I.		3.0		3.6	V
$V_{\mathrm{DD}}$	Power Supply Voltage		2.375		2.625	V
$T_{A}$	Ambient Temperature		-40		85	°С
т	P. C. I.C.	@3.3V± 10%, loaded		88	105	<u> </u>
$I_{DD}$	Power Supply Current	@2.5V± 5% loaded		58	75	mA
R <sub>DIFF_IN</sub>	Differential Input Resistance (IN+ to IN-)		90	100	110	Ω
V <sub>IH</sub>	Input High Voltage		1.2		V <sub>DD</sub> - 0.9	V
V <sub>IL</sub>	Input Low Voltage		0.4		V <sub>IH</sub> -0.1	V
V <sub>IN</sub>	Input Voltage Swing		0.1		$V_{\mathrm{DD}}$	V
V <sub>DIFF_IN</sub>	Differential Input Swing		0.2			V
V <sub>REF-AC</sub>	Output Reference Voltage		V <sub>DD</sub> -1.5	V <sub>DD</sub> -1.3	V <sub>DD</sub> -1.15	V

# **LVCMOS/LVTTL DC Characteristics** ( $T_A = -40$ °C to +85 °C, $V_{DD} = 2.5$ V $\pm 5$ % to 3.3V $\pm 10$ %)

Symbol	Parameter	Conditions	Min	Тур	Max	Units
$V_{IH}$	Input High Voltage		2.0		$V_{\mathrm{DD}}$	V
$V_{\rm IL}$	Input Low Voltage		0		0.8	V
$I_{IH}$	Input High Current		-125		20	μΑ
$I_{\mathrm{IL}}$	Input Low Current		-300			μΑ

## LVDS DC Characteristics ( $T_A = -40$ °C to +85°C, $V_{DD} = 3.3$ V $\pm 10$ %, 2.5V $\pm 5$ %)

Symbol	Parameter	Conditions	Min	Тур	Max	Units
V <sub>OUT</sub>	Output Voltage Swing		250	325	400	mV
V <sub>DIFF_OUT</sub>	Differential Output Voltage Swing		500	650	800	mV
V <sub>OCM</sub>	Output Common Mode Voltage		1.15		1.35	V
Delta V <sub>OCM</sub>	Change in Common mode Voltage		-100		100	mV





# AC Characteristics ( $T_A = -40$ °C to +85°C, $V_{DD} = 3.3$ V $\pm 10$ %, 2.5V $\pm 5$ %)

Symbol	Parameter	Conditions	Min	Тур	Max	Units
f <sub>max</sub>	Output Frequency			2.5		GHz
4 -	D (1 D 1	V <sub>IN</sub> < 400mV	370	470	570	
t <sub>pd</sub>	Propagation Delay	$V_{IN} \ge 400 \text{mV}$	300	410	500	ps
т.	Output-to-output Skew <sup>(2)</sup>				20	ps
$T_{sk}$	Device to Device skew				200	ps
T <sub>s</sub>	Setup time			150		ps
T <sub>h</sub>	Hold time			150		ps
t <sub>r</sub> /t <sub>f</sub>	Output Rise/Fall time	20% - 80%	55		200	ps
4	0 1 1	f≤1 GHz	48		52	%
$t_{odc}$	Output duty cycle	$1 \text{ GHz} \le f < 2.5 \text{ GHz}$	40		60	%
V <sub>PP</sub>	Output Swing	LVDS outputs	250		800	mV
t <sub>j</sub>	Buffer additive jitter RMS	156.25MHz with 12KHz to 20MHz integration range <sup>(3)</sup>		30		fs

#### Notes:

- 1. Measured from the differential input to the differential output crossing point
- 2. Defined as skew between outputs at the same supply voltage and with equal loads. Measured at the output differential crossing point
- 3. Input source phase noise similar to phase noise plot in page 5

## **Thermal Information**

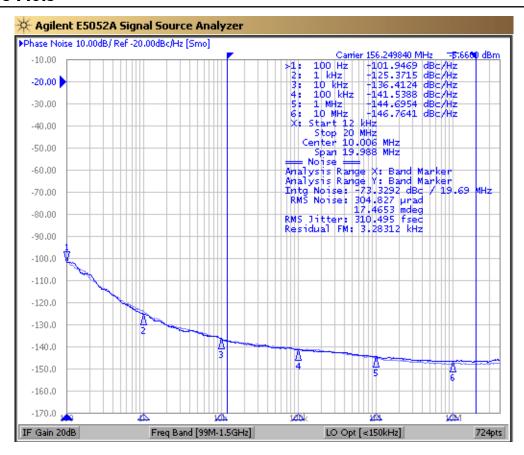
Symbol	Description	Condition	
$\Theta_{\mathrm{JA}}$	Junction-to-ambient thermal resistance	Still air	57.7 °C/W
$\Theta_{ m JC}$	Junction-to-case thermal resistance		32.2 °C/W

Note: Thermal data accounts for ePad being connected to GND.

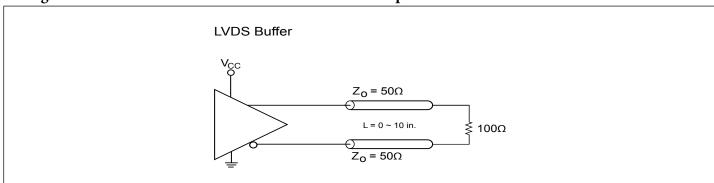




## **Phase Noise Plots**



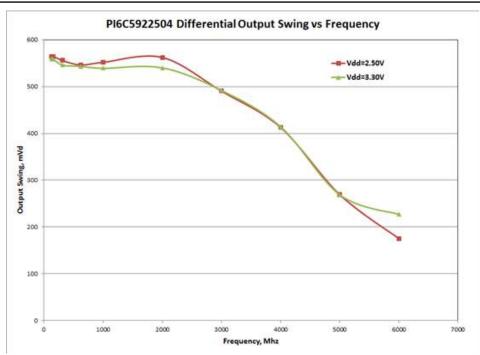
## **Configuration Test Load Board Termination for LVDS Outputs**



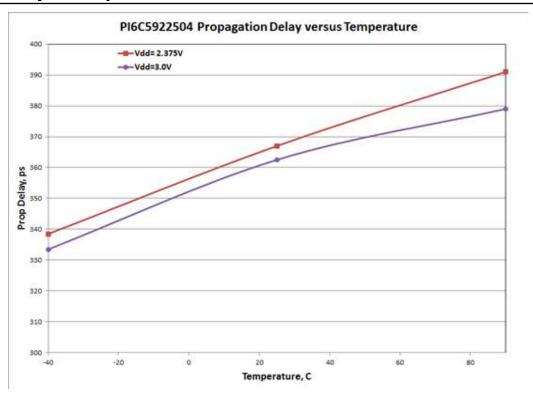




# **Output Swing vs Frequency**



# **Propagation Delay vs Temperature**



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## **Application Information**

## **Suggest for Unused Inputs and Outputs**

## **LVCMOS Input Control Pins**

It is suggested to add pull-up=4.7k and pull-down=1k for LVCMOS pins even though they have internal pull-up/down but with much higher value (>=50k) for higher design reliability.

#### REF IN=/ REF IN- Input Pins

They can be left floating if unused. For added reliability, connect  $1k\Omega$  to GND.

#### Outputs

All unused outputs are suggested to be left open and not connected to any trace. This can lower the IC power supply power.

## **Power Decoupling & Routing**

#### **VDD Pin Decoupling**

As general design rule, each VDD pin must have a 0.1uF decoupling capacitor. For better decoupling, 1uF can be used. Locating the decoupling capacitor on the component side has better decoupling filter result as shown in Fig. 1.

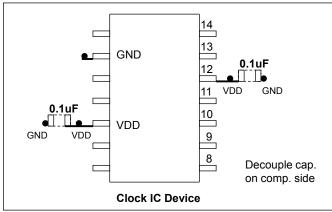
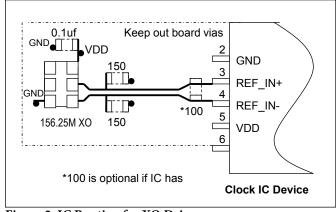


Figure 1: Placement of Decoupling Caps

#### **Differential Clock Trace Routing**

Always route differential signals symmetrically, make sure there is enough keep-out space to the adjacent trace (>20mil.). In 156.25MHz XO drives IC example, it is better routing differential trace on component side as the following Fig. 2.



**Figure 2: IC Routing for XO Drive** 

Clock timing is the most important component in PCB design, so its trace routing must be planned and routed as a first priority in manual routing. Some good practices are to use minimum vias (total trace vias count <4), use independent layers with good reference plane and keep other signal traces away from clock traces (>20mil.) etc.





## **LVPECL** and **LVDS** Input Interface

#### LVPECL and LVDS DC Input

LVPECL and LVDS clock input to this IC is connected as shown in the Fig. 3.

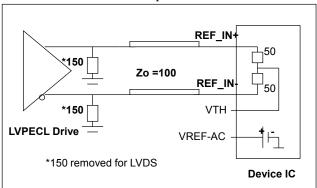


Figure 3: LVPECL/ LVDS Input

#### LVPECL and LVDS AC Input

LVPECL and LVDS AC drive to this clock IC requires the use of the VREF-AC output to recover the DC bias for the IC input as shown in Fig. 4

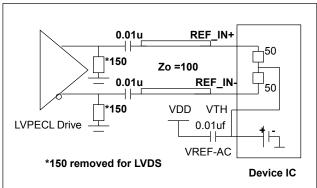


Figure 4: LVPECL/ LVDS AC Coupled Input

## **CML AC-Coupled Input**

CML AC-coupled drive requires a connection to VREF-AC as shown in Fig. 5. The CML DC drive is not recommended as different vendors have different CML DC voltage level. CML is mostly used in AC coupled drive configuration for data and clock signals.

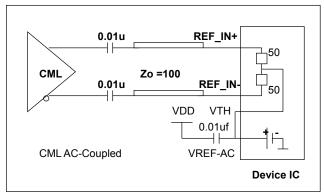


Figure 5: CML AC-Coupled Input Interface





#### **HCSL AC-Coupled Input**

It is suggested to use AC coupling to buffer PCIe HCSL 100MHz clock since its V\_cm is relatively low at about 0.4V, as shown in Fig. 6.

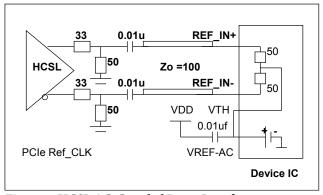


Figure 6: HCSL AC-Coupled Input Interface

#### **CMOS Clock DC Drive Input**

LVCMOS clock has voltage Voh levels such as 3.3V, 2.5V, 1.8V. CMOS drive requires a Vcm design at the input: Vcm=  $\frac{1}{2}$  (CMOS V) as shown in Fig. 7. Rs =22 ~33 $\Omega$  typically.

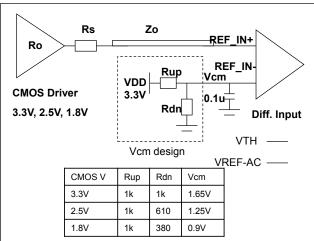


Figure 7: CMOS DC Input Vcm Design

### **Device LVPECL Output Terminations**

#### **LVPECL Output Popular Termination**

The most popular LVPECL termination is  $150\Omega$  pull-down bias and  $100\Omega$  across at RX side. Please consult ASIC datasheet if it already has  $100\Omega$  or equivalent internal termination. If so, do not connect external  $100\Omega$  across as shown in Fig. 8. This popular termination's advantage is that it does not allow any bias through from  $V_{DD}$ . This prevents  $V_{DD}$  system noise coupling onto clock trace.





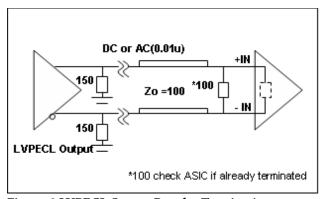


Figure. 8 LVPECL Output Popular Termination

#### LVPECL Output Thevenin Termination

Fig. 9 shows LVPECL output Thevenin termination which is used for shorter trace drive (<5in.), but it takes  $V_{DD}$  bias current and  $V_{DD}$  noise can get onto clock trace. It also requires more component count. So it is seldom used today.

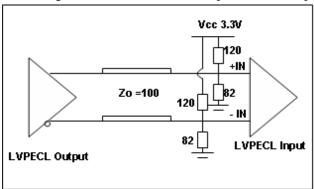


Figure. 9 LVPECL Thevenin Output Termination

## **LVPECL Output AC Thevenin Termination**

LVPECL AC Thevenin terminations require a  $150\Omega$  pull-down before the AC coupling capacitor at the source as shown in Fig. 10. Note that pull-up/down resistor value is swapped compared to Fig. 9. This circuit is good for short trace (<5in.) application only.

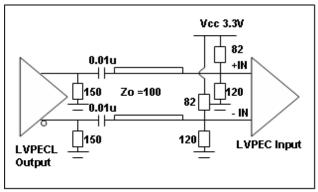


Figure. 10 LVPECL Output AC Thenvenin Termination





#### LVPECL Output Drive HCSL Input

Using the LVPECL output to drive a HCSL input can be done using a typical LVPECL AC Thenvenin termination scheme. Use pull-up/down  $450/60\Omega$  to generate Vcm=0.4V for the HCSL input clock. This termination is equivalent to  $50\Omega$  load as shown in Fig. 11.

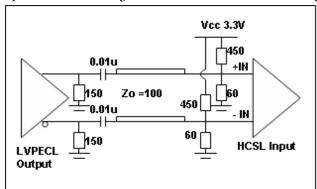


Figure. 11 LVPECL Output Drive HCSL Termination

#### LVPECL Output V\_swing Adjustment

It is suggested to add another cross  $100\Omega$  at TX side to tune the LVPECL output V\_swing without changing the optimal  $150\Omega$  pull-down bias in Fig. 12. This form of double termination can reduce the V\_swing in ½ of the original at the RX side. By fine tuning the  $100\Omega$  resistor at the TX side with larger values like 150 to  $200\Omega$ , one can increase the V\_swing by > 1/2 ratio.

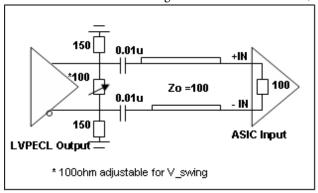


Figure. 12 LVPECL Output V\_swing Adjustment

#### **Clock Jitter Definitions**

#### Total jitter= RJ + DJ

Random Jitter (RJ) is unpredictable and unbounded timing noise that can fit in a Gaussian math distribution in RMS. RJ test values are directly related with how long or how many test samples are available. Deterministic Jitter (DJ) is timing jitter that is predictable and periodic in fixed interference frequency. Total Jitter (TJ) is the combination of random jitter and deterministic jitter: , where is a factor based on total test sample count. JEDEC std. specifies digital clock TJ in 10k random samples.

### Phase Jitter

Phase noise is short-term random noise attached on the clock carrier and it is a function of the clock offset from the carrier, for example dBc/Hz@10kHz which is phase noise power in 1-Hz normalized bandwidth vs. the carrier power @10kHz offset. Integration of phase noise in plot over a given frequency band yields RMS phase jitter, for example, to specify phase jitter <=1ps at 12k to 20MHz offset band as SONET standard specification.





#### **Device Thermal Calculation**

Fig. 13 shows the JEDEC thermal model in a 4-layer PCB.

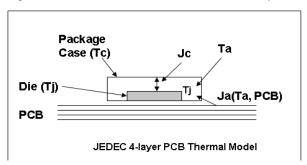


Figure. 13 JEDEC IC Thermal Model

Important factors to influence device operating temperature are:

- 1) The power dissipation from the chip (P\_chip) is after subtracting power dissipation from external loads. Generally it can be the no-load device Idd
- 2) Package type and PCB stack-up structure, for example, 1oz 4 layer board. PCB with more layers and are thicker has better heat dissipation
- 3) Chassis air flow and cooling mechanism. More air flow M/s and adding heat sink on device can reduce device final die junction temperature Tj

The individual device thermal calculation formula:

#### Tj =Ta + Pchip x Ja

#### $Tc = Tj - Pchip \times Jc$

Ja \_\_\_ Package thermal resistance from die to the ambient air in C/W unit; This data is provided in JEDEC model simulation. An air flow of 1m/s will reduce Ja (still air) by 20~30%

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Jc \_\_\_\_ Package thermal resistance from die to the package case in C/W unit

Tj \_\_\_ Die junction temperature in C (industry limit <125C max.)

Ta \_\_\_\_ Ambiant air température in C

Tc \_\_\_ Package case temperature in C

Pchip\_\_\_ IC actually consumes power through Iee/GND current





#### Thermal calculation example

To calculate Tj and Tc of PI6CV304 in an SOIC-8 package:

Step 1: Go to Diodes web to find Ja=157 C/W, Jc=42 C/W

https://www.diodes.com/design/support/packaging/pericom-packaging/packaging-mechanicals-and-thermal-characteristics/

Step 2: Go to device datasheet to find Idd=40mA max.

		C <sub>L</sub> :	C <sub>L</sub> = 33pF/33MHz	20	
•		C <sub>L</sub> = 33pF/66MHz	40		
		C <sub>L</sub> = 22pF/80MHz	35	200	
ID	Supply Current	C <sub>L</sub> = 15pF/100MHz	32	mA	
		C <sub>L</sub> = 10pF/125MHz	28		
		C <sub>L</sub> = 10pF/155MHz	41		

Step 3: P\_total= 3.3Vx40mA=0.132W

Step 4: If Ta=85C

 $Tj = 85 + Ja \times P_{total} = 85 + 25.9 = 105.7C$ 

 $Tc = Tj + Jc \times P_{total} = 105.7 - 5.54 = 100.1C$ 

#### Note:

The above calculation is directly using Idd current without subtracting the load power, so it is a conservative estimation. For more precise thermal calculation, use P\_unload or P\_chip from device Iee or GND current to calculate Tj, especially for LVPECL buffer ICs that have a 150 $\Omega$  pull-down and equivalent 100 $\Omega$  differential RX load.

# **Part Marking** PI6C592 2504ZHIE **YYWWXX** YY: Year WW: Workweek 1st X: Assembly Code 2nd X: Fab Code

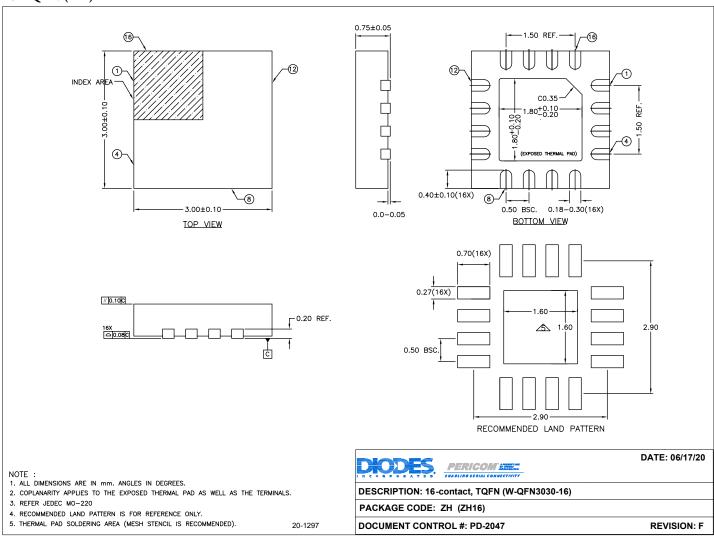
13





## **Packaging Mechanical**

## 16-TQFN (ZH)



#### For latest package info.

please check: http://www.diodes.com/design/support/packaging/pericom-packaging/packaging-mechanicals-and-thermal-characteristics/

# **Ordering Information**

Ordering Code	Package Code	Package Description
PI6C5922504ZHIEX	ZH	16-Contact, Thin Fine Pitch Quad Flat No-Lead (TQFN)

#### Notes

- 1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant.
- 2. See https://www.diodes.com/quality/lead-free/ for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
- 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.
- 4. I = Industrial
- 5. E = Pb-free and Green
- 6. X suffix = Tape/Reel





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