Crystal-to-LVPECL Clock Synthesizer

ICS843312

General Description

The ICS843312 is a high frequency clock generator. The ICS843312 uses an external 25MHz crystal to synthesize 312.5MHz and 125MHz clocks. The ICS843312 has excellent RMS period jitter performance.

The ICS843312 operates at full 3.3V and 2.5V, or mixed 3.3V/2.5V supply modes and is available in a fully RoHS compliant 32-lead VFQFN package.

Features

- Ten differential LVPECL outputs for 312.5MHz and 125MHz
- Crystal oscillator interface designed for 18pF, 25MHz parallel resonant crystal
- RMS phase jitter at 125MHz (1.875MHz - 20MHz): 0.45ps (typical)
- RMS phase jitter at 312.5MHz (1.875MHz - 20MHz): 0.31ps (typical)
- Output duty cycle: 45% – 55%, at 125MHz
- Full 3.3V and 2.5V, or mixed 3.3V/2.5V supply modes
- 0°C to 70°C ambient operating temperature
- Available in lead-free (RoHS 6) package

Output Frequency Table

Crystal Frequency (MHz)	Feedback Divider	VCO Frequency (MHz)	Output Divider	Output Frequency (MHz)
25	25	625	÷2	312.5
25	25	625	÷5	125

Block Diagram
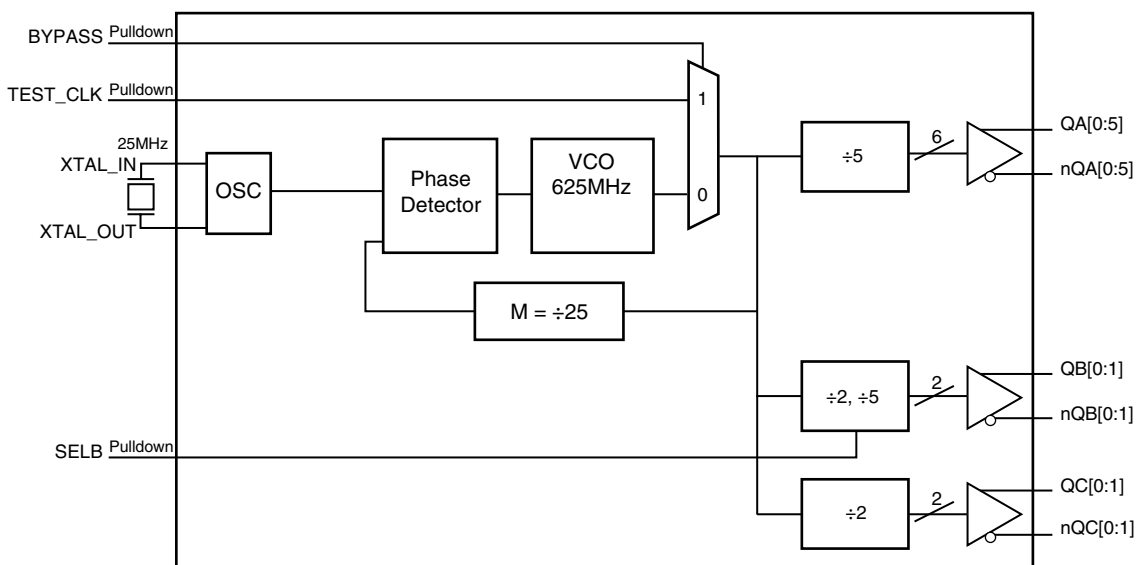
Pin Assignment
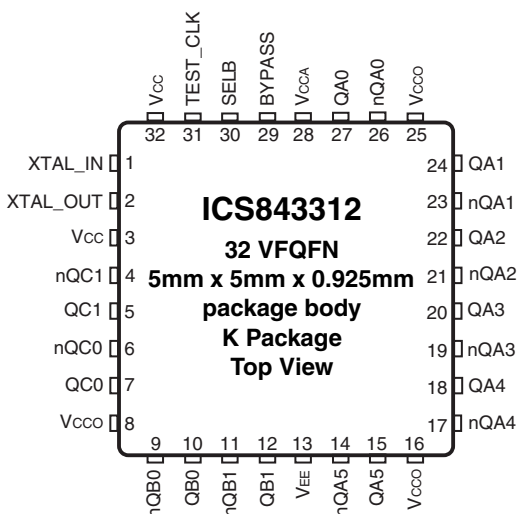
Table 1. Pin Descriptions

Number	Name	Type		Description
1, 2	XTAL_IN XTAL_OUT	Input		Crystal oscillator interface. XTAL_IN is the input, XTAL_OUT is the output.
3, 32	V_{CC}	Power		Core supply pins.
4, 5	nQC1/QC1	Output		Differential output pair. LVPECL interface levels.
6, 7	nQC0/QC0	Output		Differential output pair. LVPECL interface levels.
8, 16, 25	V_{CCO}	Power		Output supply pins.
9, 10	nQB0/QB0	Output		Differential output pair. LVPECL interface levels.
11, 12	nQB1/QB1	Output		Differential output pair. LVPECL interface levels.
13	V_{EE}	Power		Negative supply pin.
14, 15	nQA5/QA5	Output		Differential output pair. LVPECL interface levels.
17, 18	nQA4/QA4	Output		Differential output pair. LVPECL interface levels.
19, 20	nQA3/QA3	Output		Differential output pair. LVPECL interface levels.
21, 22	nQA2/QA2	Output		Differential output pair. LVPECL interface levels.
23, 24	nQA1/QA1	Output		Differential output pair. LVPECL interface levels.
26, 27	nQA0/QA0	Output		Differential output pair. LVPECL interface levels.
28	V_{CCA}	Power		Analog supply pin.
29	BYPASS	Input	Pulldown	A HIGH on BYPASS signal allows TEST_CLK to propagate to output dividers and bypass the PLL. a LOW on BYPASS signal allows VCO frequency to propagate to the output dividers. See Table 3B. LVCMOS/LVTTL interface levels.
30	SELB	Input	Pulldown	Selects the output divider value. See Table 3A. LVCMOS/LVTTL interface levels.
31	TEST_CLK	Input	Pulldown	Single-ended input test clock. LVCMOS/LVTTL interface levels.

NOTE: *Pulldown* refers to internal input resistors. See Table 2, *Pin Characteristics,* for typical values.

Table 2. Pin Characteristics

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C_{IN}	Input Capacitance			4		pF
$R_{PULLDOWN}$	Input Pulldown Resistor			51		kΩ

Function Tables

Table 3A. SELB Function Table

Input	
SELB	Bank B Output Divider
0	÷5
1	÷2

Table 3B. Bypass Function Table

Input	
BYPASS	Device Configuration
0	PLL Mode
1	Bypass the PLL

Absolute Maximum Ratings

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device.
These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond
those listed in the *DC Characteristics or AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for
extended periods may affect product reliability.

Item	Rating
Supply Voltage, V_{CC}	4.6V
Inputs, V_I	-0.5V to V_{CC} + 0.5V
Outputs, I_O Continuous Current Surge Current	50mA 100mA
Package Thermal Impedance, θ_{JA}	37°C/W (0 mps)
Storage Temperature, T_{STG}	-65°C to 150°C

DC Electrical Characteristics

Table 4A. Power Supply DC Characteristics, $V_{CC} = V_{CCO} = 3.3V \pm 5\%$, $V_{EE} = 0V$, $T_A = 0°C$ to $70°C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{CC}	Core Supply Voltage		3.135	3.3	3.465	V
V_{CCA}	Analog Supply Voltage		$V_{CC} - 0.33$	3.3	V_{CC}	V
V_{CCO}	Output Supply Voltage		3.135	3.3	3.465	V
I_{EE}	Power Supply Current				179	mA
I_{CCA}	Analog Supply Current				33	mA

Table 4B. Power Supply DC Characteristics, $V_{CC} = V_{CCO} = 2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = 0°C$ to $70°C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{CC}	Core Supply Voltage		2.375	2.5	2.625	V
V_{CCA}	Analog Supply Voltage		$V_{CC} - 0.23$	2.5	V_{CC}	V
V_{CCO}	Output Supply Voltage		2.375	2.5	2.625	V
I_{EE}	Power Supply Current				168	mA
I_{CCA}	Analog Supply Current				23	mA

Table 4C. Power Supply DC Characteristics, $V_{CC} = 3.3V \pm 5\%$, $V_{CCO} = 2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = 0°C$ to $70°C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{CC}	Core Supply Voltage		3.135	3.3	3.465	V
V_{CCA}	Analog Supply Voltage		$V_{CC} - 0.33$	3.3	V_{CC}	V
V_{CCO}	Output Supply Voltage		2.375	2.5	2.625	V
I_{EE}	Power Supply Current				164	mA
I_{CCA}	Analog Supply Current				33	mA

Table 4D. LVCMOS/LVTTL DC Characteristics, T_A = 0°C to 70°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V_{IH}	Input High Voltage		V_{CC} = 3.3V	2		V_{CC} + 0.3	V
			V_{CC} = 2.5V	1.7		V_{CC} + 0.3	V
V_{IL}	Input Low Voltage		V_{CC} = 3.3V	-0.3		0.8	V
			V_{CC} = 2.5V	-0.3		0.7	V
I_{IH}	Input High Current	TEST_CLK, BYPASS, SELB	V_{CC} = V_{IN} = 3.465V or 2.625V			150	µA
I_{IL}	Input Low Current	TEST_CLK, BYPASS, SELB	V_{CC} = 3.465V or 2.625V, V_{IN} = 0V	-5			µA

Table 4E. LVPECL DC Characteristics, V_{CC} = V_{CCO} = 3.3V ± 5%, V_{EE} = 0V, T_A = 0°C to 70°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{OH}	Output High Voltage		V_{CCO} - 1.4		V_{CCO} - 0.9	V
V_{OL}	Output Low Voltage		V_{CCO} - 2.0		V_{CCO} - 1.7	V
V_{SWING}	Peak-to-Peak Output Voltage Swing		0.6		1.0	V

Table 4F. LVPECL DC Characteristics, V_{CC} = 3.3V ± 5% or 2.5V ± 5%, V_{CCO} = 2.5V ± 5%, V_{EE} = 0V, T_A = 0°C to 70°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V_{OH}	Output High Voltage		V_{CC} – 1.4		V_{CC} – 0.9	V
V_{OL}	Output Low Voltage		V_{CC} – 2.0		V_{CC} – 1.5	V
V_{SWING}	Peak-to-Peak Output Voltage Swing		0.4		1.0	V

Table 5. Crystal Characteristics

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation			Fundamental		
Frequency			25		MHz
Equivalent Series Resistance (ESR)				50	Ω
Shunt Capacitance				7	pF

4

AC Electrical Characteristics

Table 6A. AC Characteristics, $V_{CC} = V_{CCO} = 3.3V \pm 5\%$, $V_{EE} = 0V$, $T_A = 0°C$ to $70°C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
f_{OUT}	Output Frequency	Output divider = ÷2		312.5		MHz
		Output divider = ÷5		125		MHz
$f_{jit(Ø)}$	RMS Phase Jitter, (Random); NOTE 1	312.5MHz, Integration Range: 1.875MHz - 20MHz		0.31		ps
		125MHz, Integration Range: 1.875MHz - 20MHz		0.45		ps
t_R / t_F	Output Rise/Fall Time	20% to 80%	200		700	ps
odc	Output Duty Cycle	f_{OUT} = 125MHz	45		55	%
		f_{OUT} = 312.5MHz	40		60	%
t_{LOCK}	PLL Lock Time				100	ms

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.
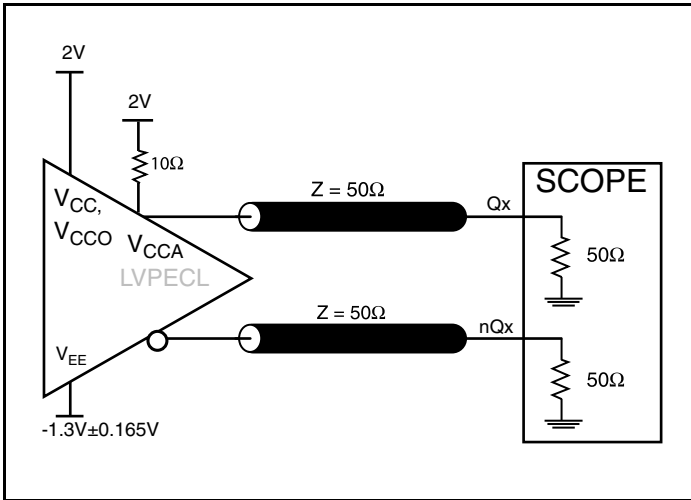NOTE 1: Refer to phase noise plots.

Table 6B. AC Characteristics, $V_{CC} = V_{CCO} = 2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = 0°C$ to $70°C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
f_{OUT}	Output Frequency	Output divider = ÷2		312.5		MHz
		Output divider = ÷5		125		MHz
$f_{jit(Ø)}$	RMS Phase Jitter, (Random)	312.5MHz, Integration Range: 1.875MHz - 20MHz		0.45		ps
		125MHz, Integration Range: 1.875MHz - 20MHz		0.55		ps
t_R / t_F	Output Rise/Fall Time	20% to 80%	200		700	ps
odc	Output Duty Cycle	f_{OUT} = 125MHz	45		55	%
		f_{OUT} = 312.5MHz	40		60	%
t_{LOCK}	PLL Lock Time				100	ms

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

Table 6C. AC Characteristics, $V_{CC} = 3.3V \pm 5\%$, $V_{CCO} = 2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = 0°C$ to $70°C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
f_{OUT}	Output Frequency	Output divider = ÷2		312.5		MHz
		Output divider = ÷5		125		MHz
$f_{jit(Ø)}$	RMS Phase Jitter, (Random)	312.5MHz, Integration Range: 1.875MHz - 20MHz		0.31		ps
		125MHz, Integration Range: 1.875MHz - 20MHz		0.45		ps
t_R / t_F	Output Rise/Fall Time	20% to 80%	200		700	ps
odc	Output Duty Cycle	f_{OUT} = 125MHz	45		55	%
		f_{OUT} = 312.5MHz	40		60	%
t_{LOCK}	PLL Lock Time				100	ms

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.
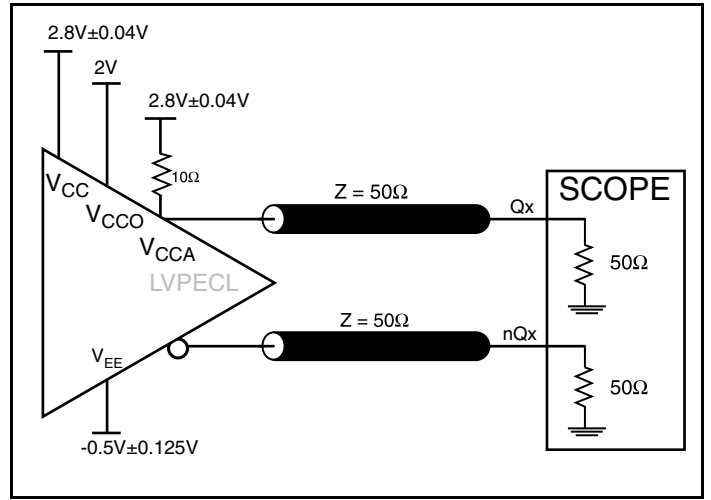
Typical Phase Noise at 125MHz (3.3V)

125MHz
RMS Phase Jitter (Random)
1.875MHz to 20MHz = 0.45ps (typical)
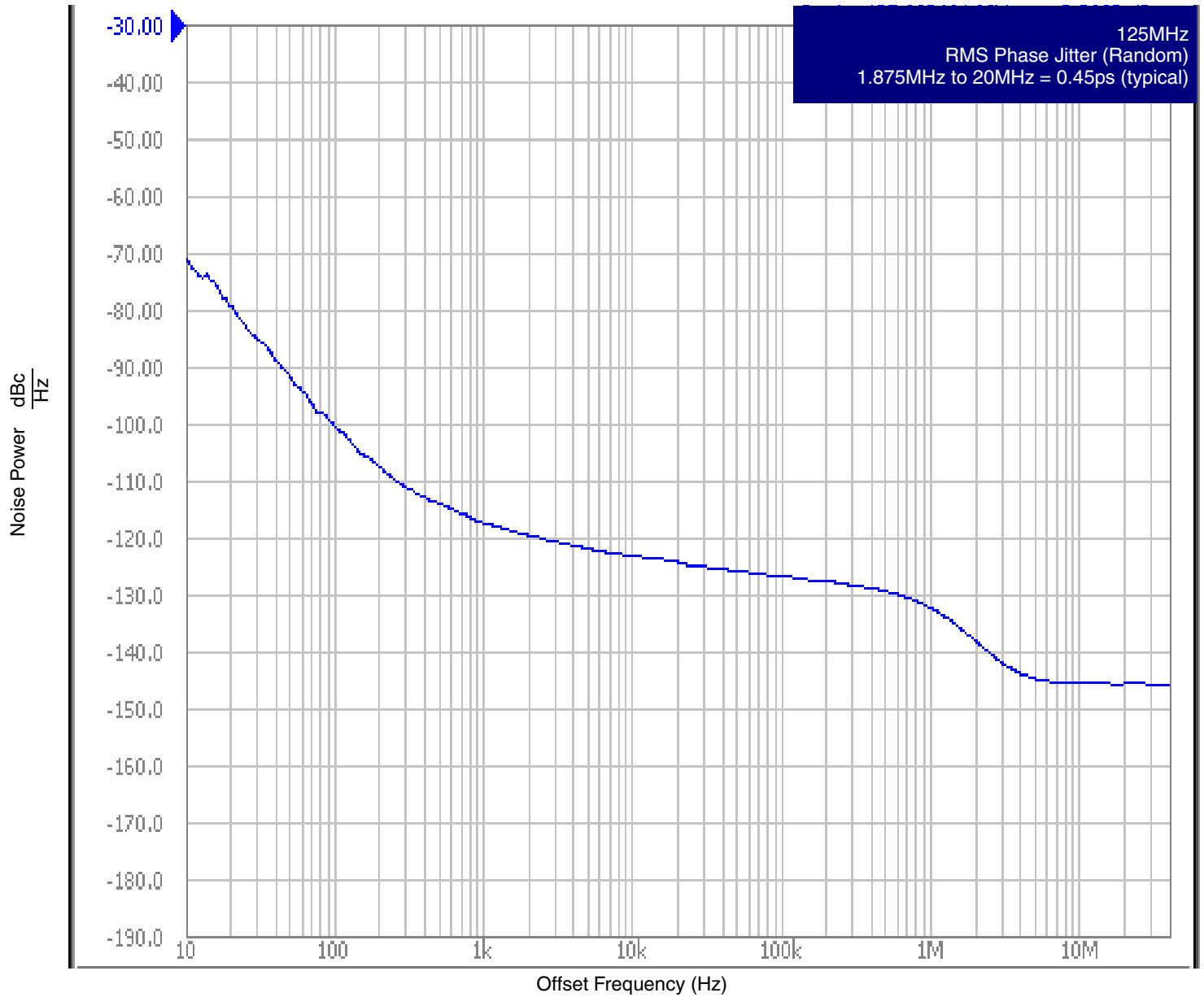
Typical Phase Noise at 312.5MHz (3.3V)

312.5MHz
RMS Phase Jitter (Random)
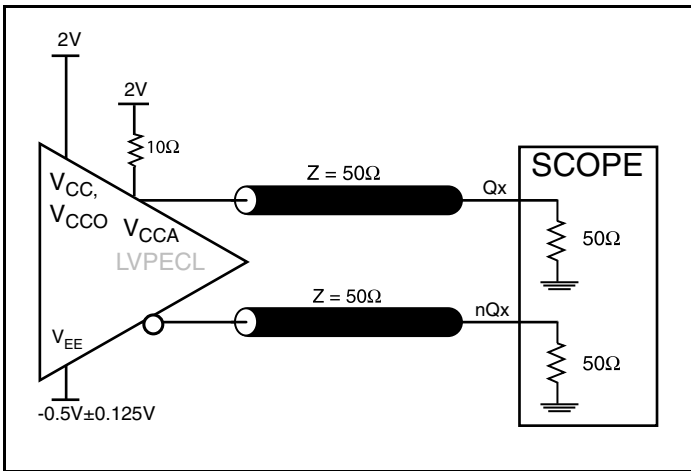1.875MHz to 20MHz = 0.31ps (typical))
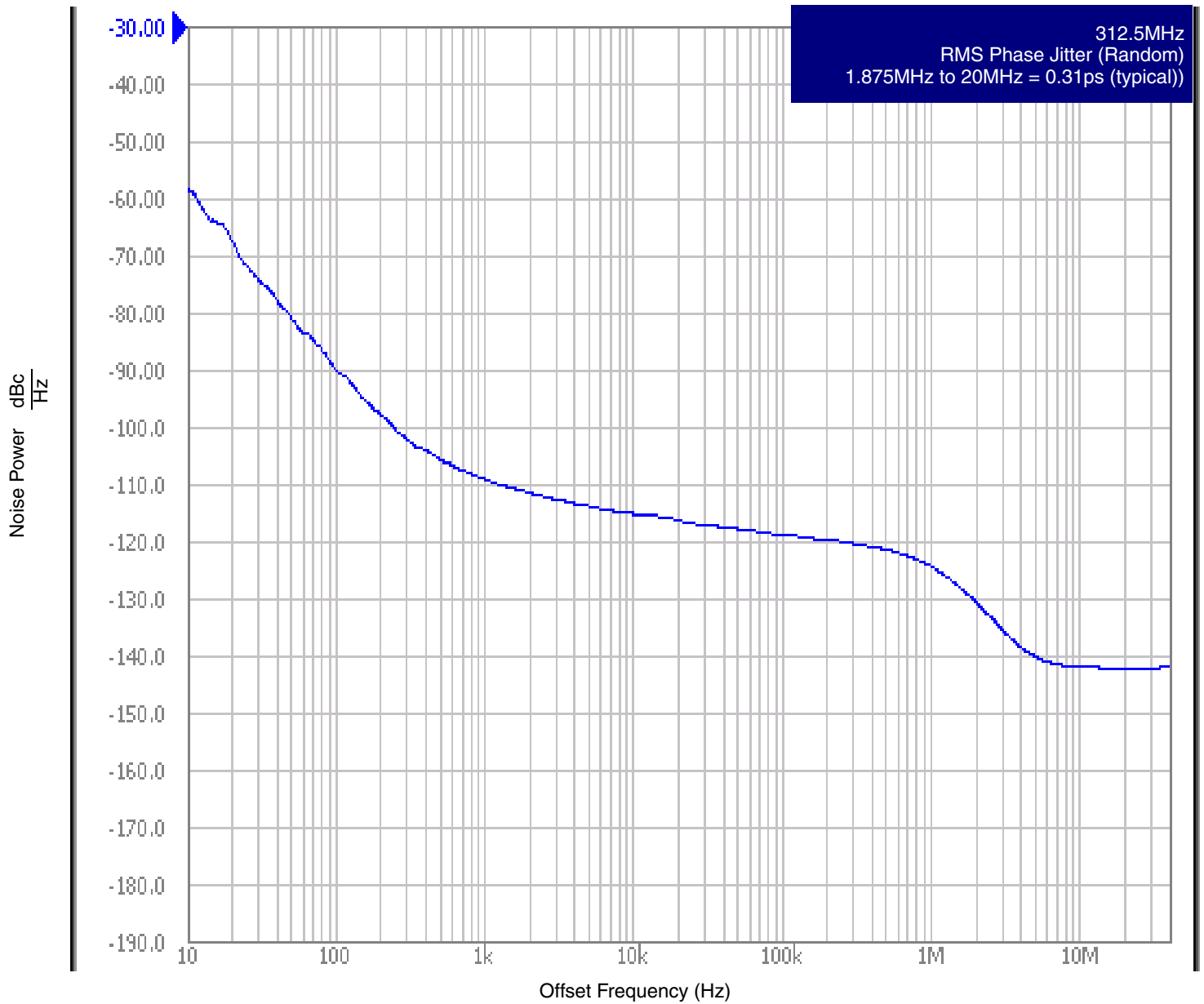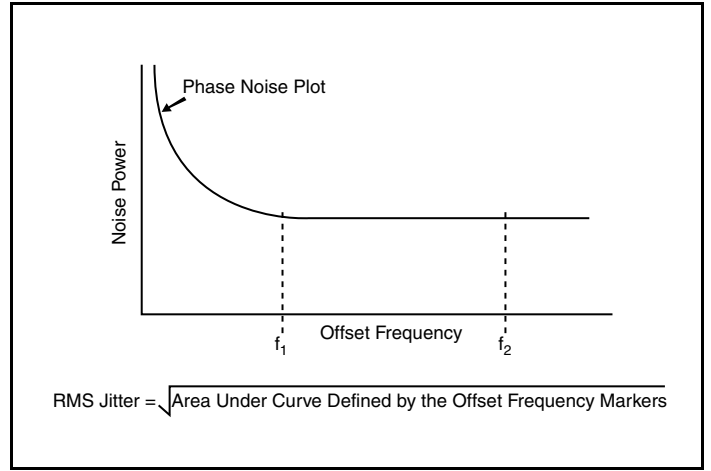
Parameter Measurement Information

3.3V Core/ 3.3V LVPECL Output Load AC Test Circuit
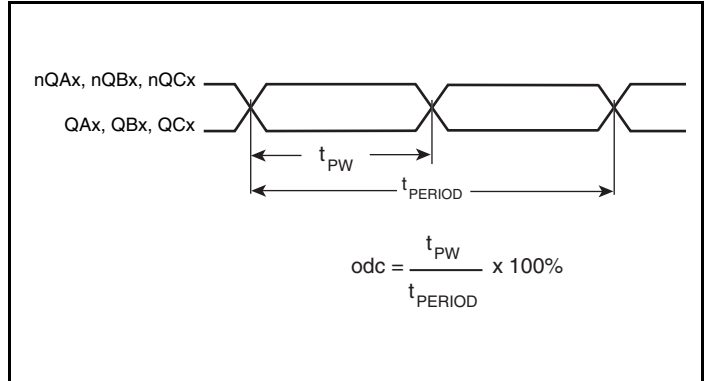
3.3V Core/ 2.5V LVPECL Output Load AC Test Circuit

2.5V Core/ 2.5V LVPECL Output Load AC Test Circuit

$$\text{RMS Jitter} = \sqrt{\text{Area Under Curve Defined by the Offset Frequency Markers}}$$

RMS Phase Jitter
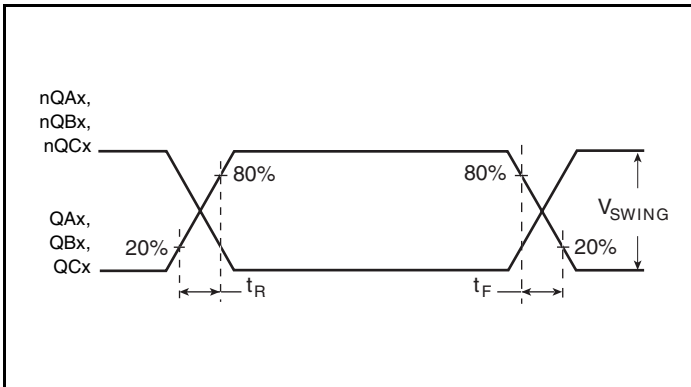
Output Rise/Fall Time

$$odc = \frac{t_{PW}}{t_{PERIOD}} \times 100\%$$

Output Duty Cycle/Pulse Width/Period

8

Applications Information

Power Supply Filtering Technique

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. The ICS843312 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL. V_{CC}, V_{CCA} and V_{CCO} should be individually connected to the power supply plane through vias, and 0.01µF bypass capacitors should be used for each pin. *Figure 1* illustrates this for a generic V_{CC} pin and also shows that V_{CCA} requires that an additional 10Ω resistor along with a 10µF bypass capacitor be connected to the V_{CCA} pin.
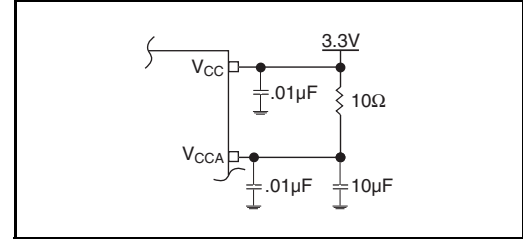
Figure 1. Power Supply Filtering

Recommendations for Unused Input and Output Pins

Inputs:

TEST_CLK Input

For applications not requiring the use of the test clock, it can be left floating. Though not required, but for additional protection, a 1kΩ resistor can be tied from the TEST_CLK to ground.

Crystal Inputs

For applications not requiring the use of the crystal oscillator input, both XTAL_IN and XTAL_OUT can be left floating. Though not required, but for additional protection, a 1kΩ resistor can be tied from XTAL_IN to ground.

LVCMOS Control Pins

All control pins have internal pulldowns; additional resistance is not required but can be added for additional protection. A 1kΩ resistor can be used.

Outputs:

LVPECL Outputs

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

Crystal Input Interface

The ICS843312 has been characterized with 18pF parallel resonant crystals. The capacitor values, C1 and C2, shown in *Figure 2* below were determined using a 25MHz, 18pF parallel resonant crystal and were chosen to minimize the ppm error. The optimum C1 and C2 values can be slightly adjusted for different board layouts.
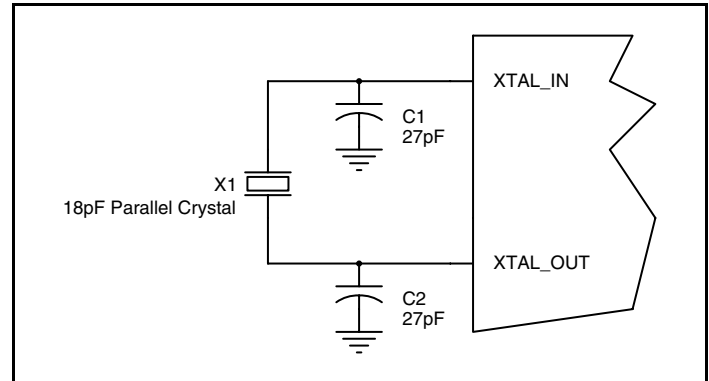
Figure 2. Crystal Input Interface

Overdriving the XTAL Interface

The XTAL_IN input can accept a single-ended LVCMOS signal through an AC coupling capacitor. A general interface diagram is shown in *Figure 3A.* The XTAL_OUT pin can be left floating. The maximum amplitude of the input signal should not exceed 2V and the input edge rate can be as slow as 10ns. This configuration requires that the output impedance of the driver (Ro) plus the series resistance (Rs) equals the transmission line impedance. In addition,

matched termination at the crystal input will attenuate the signal in half. This can be done in one of two ways. First, R1 and R2 in parallel should equal the transmission line impedance. For most 50Ω applications, R1 and R2 can be 100Ω. This can also be accomplished by removing R1 and making R2 50Ω. By overdriving the crystal oscillator, the device will be functional, but note, the device performance is guaranteed by using a quartz crystal.
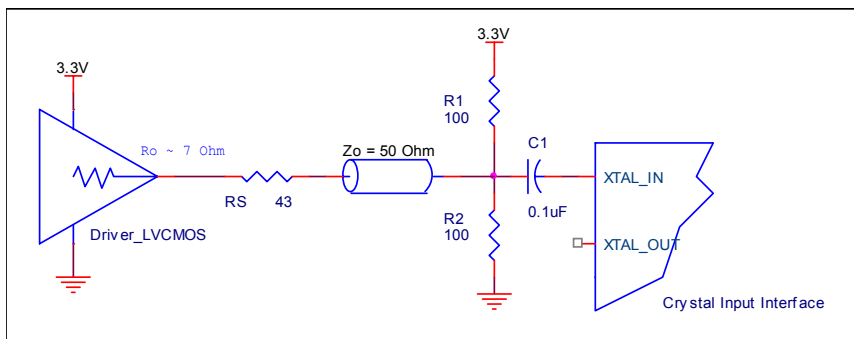
Figure 3A. General Diagram for LVCMOS Driver to XTAL Input Interface
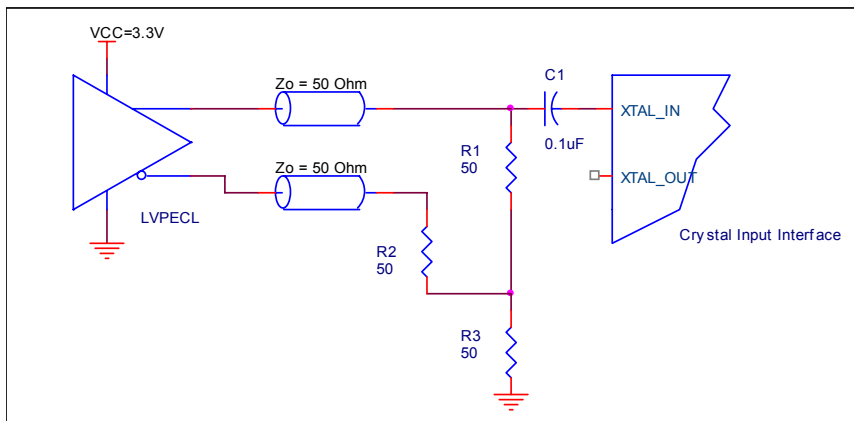
Figure 3B. General Diagram for LVPECL Driver to XTAL Input Interface

Termination for 3.3V LVPECL Outputs

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

The differential outputs are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive 50Ω

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 4A and 4B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.
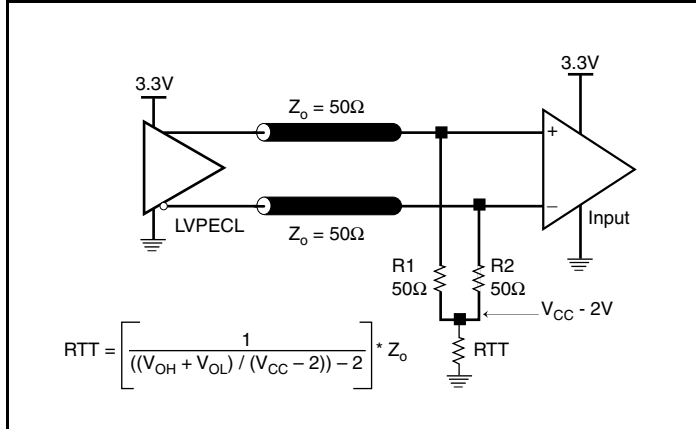
Figure 4A. 3.3V LVPECL Output Termination
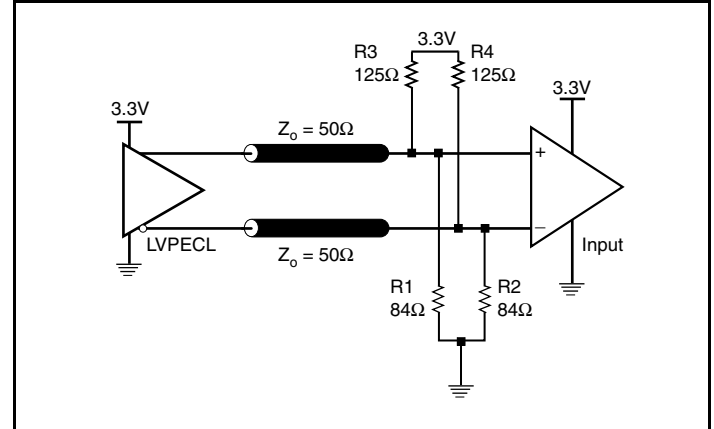
Figure 4B. 3.3V LVPECL Output Termination

Termination for 2.5V LVPECL Outputs

Figure 5A and *Figure 5B* show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating 50Ω to V_{CC} – 2V. For V_{CCO} = 2.5V, the V_{CCO} – 2V is very close to ground

level. The R3 in Figure 5B can be eliminated and the termination is shown in *Figure 5C*.
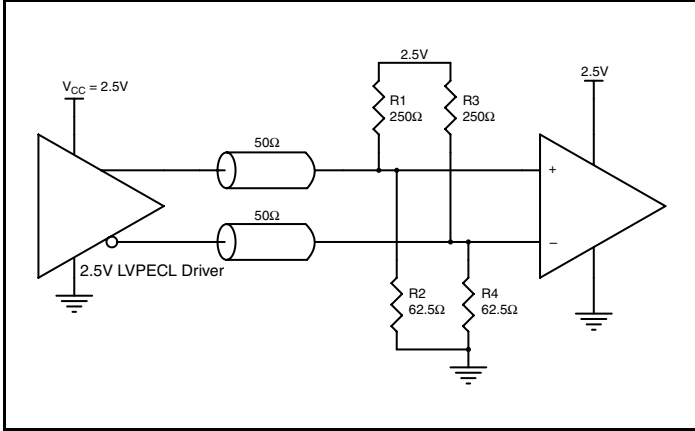
Figure 5A. 2.5V LVPECL Driver Termination Example
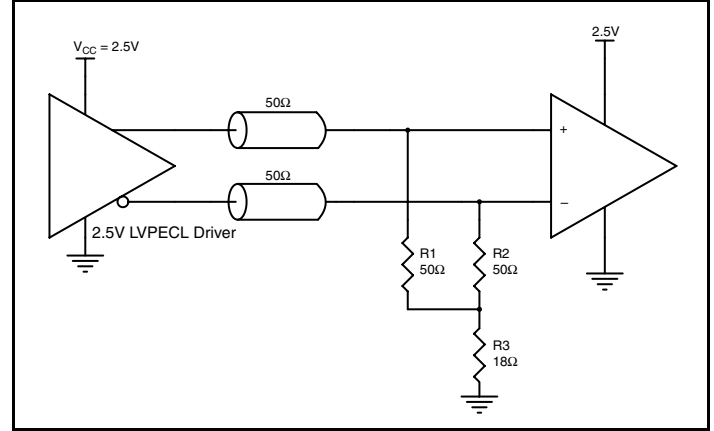
Figure 5B. 2.5V LVPECL Driver Termination Example
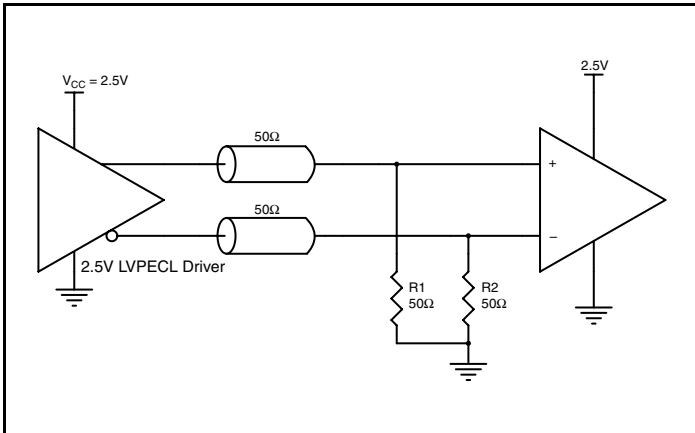
Figure 5C. 2.5V LVPECL Driver Termination Example

VFQFN EPAD Thermal Release Path

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in *Figure 6.* The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The vias act as "heat pipes". The number of vias (i.e. "heat pipes") are application specific

and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, please refer to the Application Note on the Surface Mount Assembly of Amkor's Thermally/ Electrically Enhance Leadframe Base Package, Amkor Technology.
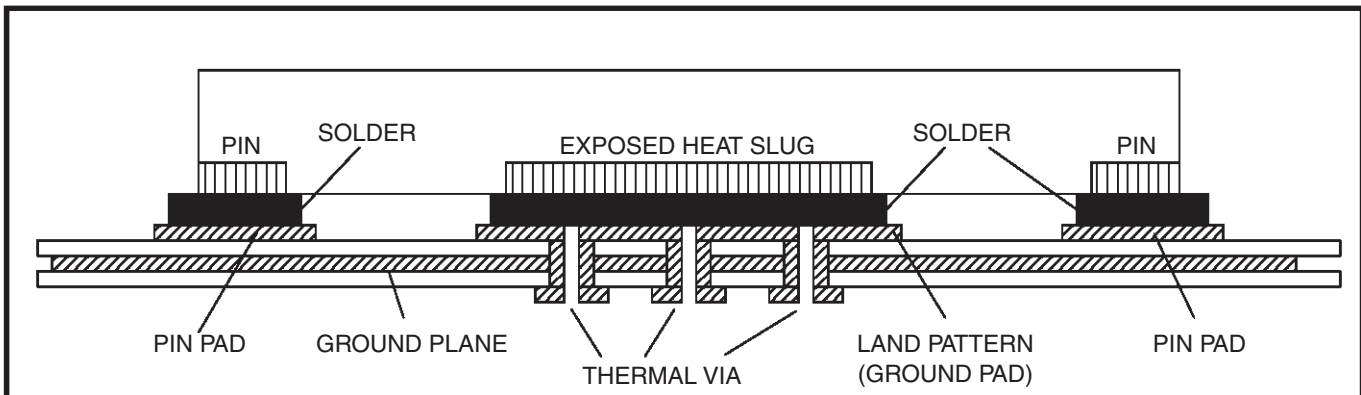

Figure 6. P.C. Assembly for Exposed Pad Thermal Release Path – Side View (drawing not to scale)

Power Considerations

This section provides information on power dissipation and junction temperature for the ICS843312.
Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the ICS843312 is the sum of the core power plus the power dissipation in the load(s).
The following is the power dissipation for V_{CCO} = 3.3V + 5% = 3.465V, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipation in the load.

- Power (core)$_{MAX}$ = V_{CC_MAX} * I_{EE_MAX} = 3.465V * 179mA = **620.235mW**

- Power (outputs)$_{MAX}$ = **30mW/Loaded Output pair**
  If all outputs are loaded, the total power is 10 * 30mW = **300mW**

Total Power$_{\_MAX}$ (3.3V, with all outputs switching) = 620.235mW + 300mW = **920.235mW**

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad, and directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj = θ_{JA} * Pd_total + T_A

Tj = Junction Temperature

θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 37°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 70°C with all outputs switching is:

70°C + 0.920W * 37°C/W = 104°C. This is well below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 7. Thermal Resistance θ_{JA} for 32 Lead VFQFN, Forced Convection

θ_{JA} by Velocity			
Meters per Second	**0**	**1**	**2.5**
Multi-Layer PCB, JEDEC Standard Test Boards	37.0°C/W	32.4°C/W	29.0°C/W

3. Calculations and Equations.

The purpose of this section is to calculate the power dissipation for the LVPECL output pairs.

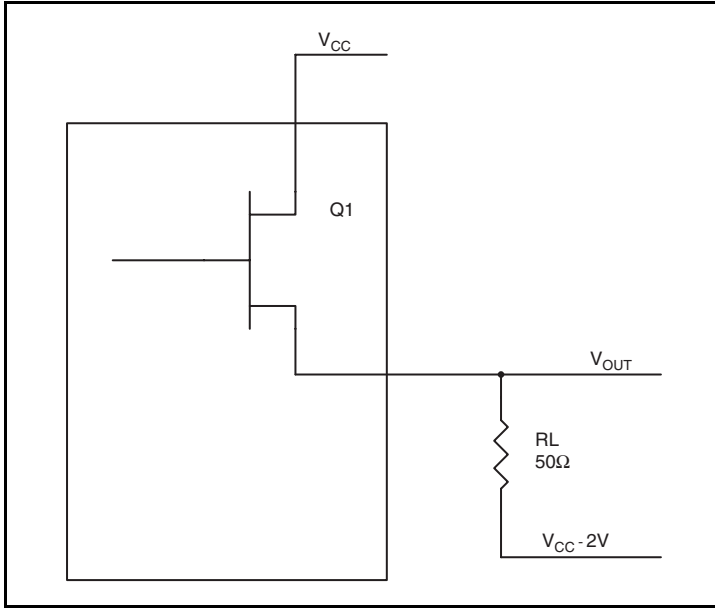The LVPECL output driver circuit and termination are shown in *Figure 7.*

Figure 7. LVPECL Driver Circuit and Termination

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of $V_{CCO} - 2V$.

- For logic high, $V_{OUT} = V_{OH\_MAX} = \mathbf{V_{CCO\_MAX} - 0.9V}$
  $(V_{CC\_MAX} - V_{OH\_MAX}) = \mathbf{0.9V}$

- For logic low, $V_{OUT} = V_{OL\_MAX} = \mathbf{V_{CCO\_MAX} - 1.7V}$
  $(V_{CCO\_MAX} - V_{OL\_MAX}) = \mathbf{1.7V}$

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

$Pd\_H = [(V_{OH\_MAX} - (V_{CCO\_MAX} - 2V))/R_L] * (V_{CCO\_MAX} - V_{OH\_MAX}) = [(2V - (V_{CCO\_MAX} - V_{OH\_MAX}))/R_L] * (V_{CCO\_MAX} - V_{OH\_MAX}) =$
$[(2V - 0.9V)/50\Omega] * 0.9V = \mathbf{19.8mW}$

$Pd\_L = [(V_{OL\_MAX} - (V_{CCO\_MAX} - 2V))/R_L] * (V_{CCO\_MAX} - V_{OL\_MAX}) = [(2V - (V_{CCO\_MAX} - V_{OL\_MAX}))/R_L] * (V_{CCO\_MAX} - V_{OL\_MAX}) =$
$[(2V - 1.7V)/50\Omega] * 1.7V = \mathbf{10.2mW}$

Total Power Dissipation per output pair = Pd_H + Pd_L = **30mW**

Reliability Information

Table 8. θ_{JA} vs. Air Flow Table for a 32Lead VFQFN

θ_{JA} vs. Air Flow			
Meters per Second	**0**	**1**	**2.5**
Multi-Layer PCB, JEDEC Standard Test Boards	37.0°C/W	32.4°C/W	29.0°C/W

Transistor Count

The transistor count for ICS843312 is: 3059

Package Outline and Package Dimensions

Package Outline - K Suffix for 32 Lead VFQFN

Bottom View w/Type A ID

Bottom View w/Type B ID

Bottom View w/Type C ID
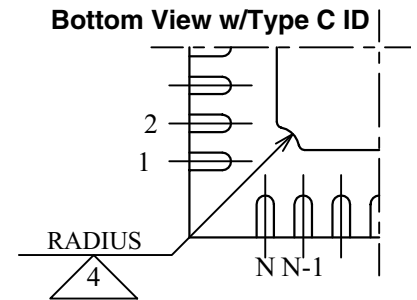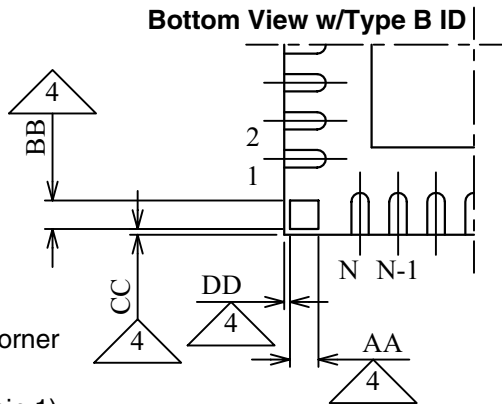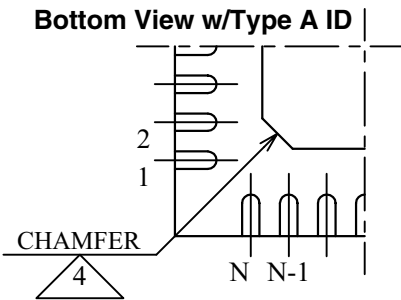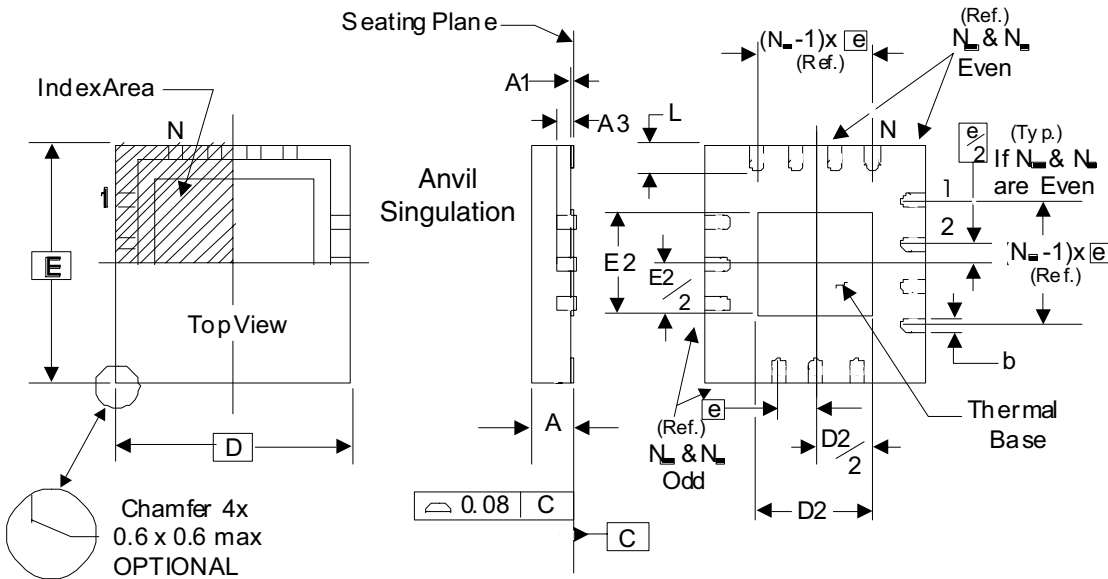
There are 3 methods of indicating pin 1 corner at the back of the VFQFN package are:

1. Type A: Chamfer on the paddle (near pin 1)

2. Type B: Dummy pad between pin 1 and N.

3. Type C: Mouse bite on the paddle (near pin 1)

NOTE: The following package mechanical drawing is a generic drawing that applies to any pin count VFQFN package. This drawing is not intended to convey the actual pin count or pin layout of this device. The pin count and pinout are shown on the front page. The package dimensions are in Table 9.

Table 9. Package Dimensions

JEDEC Variation: VHHD-2/-4 All Dimensions in Millimeters			
Symbol	Minimum	Nominal	Maximum
N		32	
A	0.80		1.00
A1	0		0.05
A3		0.25 Ref.	
b	0.18	0.25	0.30
N_D & N_E			8
D & E		5.00 Basic	
D2 & E2	3.0		3.3
e		0.50 Basic	
L	0.30	0.40	0.50

Reference Document: JEDEC Publication 95, MO-220

Ordering Information

Table 10. Ordering Information

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
843312AKLF	ICS843312AL	"Lead-Free" 32 Lead VFQFN	Tray	0°C to 70°C
843312AKLFT	ICS843312AL	"Lead-Free" 32 Lead VFQFN	2500 Tape & Reel	0°C to 70°C

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

Revision History Sheet

Rev	Table	Page	Description of Change	Date
B	T6A - T6C	1 5 10 17	Block Diagram - added outline to distinguish crystal oscillator is inside the part. AC Characteristics Tables - added Output Duty Cycle spec for 312.5MHz. Updated Overdriving the XTAL Interface. Updated Package Outline.	9/28/10

We've Got Your Timing Solution

IDT
www.IDT.com

6024 Silver Creek Valley Road
San Jose, California 95138

Sales
800-345-7015 (inside USA)
+408-284-8200 (outside USA)
Fax: 408-284-2775
www.IDT.com/go/contactIDT

Technical Support
netcom@idt.com
+480-763-2056