Features

- **80C52 Compatible**
	- **Four 8-bit I/O Ports**
	- **Three 16-bit Timer/Counters**
	- **256 Bytes Scratch Pad RAM**
	- **8 Interrupt Sources with 4 Priority Levels**
	- **Dual Data Pointer**
- **Variable Length MOVX for Slow RAM/Peripherals**
- **High-speed Architecture**
- **10 to 40 MHz in Standard Mode**
- **16K/32K Bytes On-Chip ROM Program**
- **AT80C51RD2 ROMless Versions**
- **On-Chip 1024 bytes Expanded RAM (XRAM)**
	- **Software Selectable Size (0, 256, 512, 768, 1024 bytes)**
	- **256 Bytes Selected at Reset**
- **Keyboard Interrupt Interface on Port P1**
- **8-bit Clock Prescaler**
- **64K Program and Data Memory Spaces**
- **Improved X2 Mode with Independant Selection for CPU and Each Peripheral**
- **Programmable Counter Array 5 Channels with:**
	- **– High-speed Output**
	- **– Compare/Capture**
	- **– Pulse Width Modulator**
	- **– Watchdog Timer Capabilities**
- **Asynchronous Port Reset**
- **Full Duplex Enhanced UART**
- **Dedicated Baud Rate Generator for UART**
- **Low EMI (Inhibit ALE)**
- **Hardware Watchdog Timer (One-time Enabled with Reset-out)**
- **Power Control Modes**
	- **Idle Mode**
	- **Power-down Mode**
	- **Power-off Flag**
- **Power Supply: 2.7V to 5.5V**
- **Temperature Ranges: Commercial (0 to +70**°**C) and Industrial (-40**°**C to +85**°**C)**
- **Packages: PDIL40, PLCC44, VQFP44**

80C51 High Performance ROM 8-bit Microcontroller

AT80C51RD2

1. Description

AT80C51RD2 microcontrollers are high performance versions of the 80C51 8-bit microcontrollers.

The microcontrollers retain all features of the Atmel 80C52 with 256 bytes of internal RAM, a 7 source 4-level interrupt controller and three timer/counters.

In addition, the microcontrollers have a Programmable Counter Array, an XRAM of 1024 byte, a Hardware Watchdog Timer, a Keyboard Interface, a more versatile serial channel that facilitates multiprocessor communication (EUART) and a speed improvement mechanism (X2 mode).

The microcontrollers have 2 software-selectable modes of reduced activity and 8 bit clock prescaler for further reduction in power consumption. In Idle mode, the CPU is frozen while the peripherals and the interrupt system are still operating. In the Power-down mode, the RAM is saved and all other functions are inoperative.

Table 1. Memory Size

2. Block Diagram

3. Pin Configurations

*NIC: No Internal Connection

Table 3-1. Pin Description

Table 3-1. Pin Description (Continued)

4. SFR Mapping

The Special Function Registers (SFRs) of the microcontroller fall into the following categories:

- C51 core registers: ACC, B, DPH, DPL, PSW, SP
- I/O port registers: P0, P1, P2, P3
- Timer registers: T2CON, T2MOD, TCON, TH0, TH1, TH2, TMOD, TL0, TL1, TL2, RCAP2L, RCAP2H
- Serial I/O port registers: SADDR, SADEN, SBUF, SCON
- PCA (Programmable Counter Array) registers: CCON, CCAPMx, CL, CH, CCAPxH, CCAPxL (x: 0 to 4)
- Power and clock control registers: PCON
- Hardware Watchdog Timer registers: WDTRST, WDTPRG
- Interrupt system registers: IE0, IPL0, IPH0, IE1, IPL1, IPH1
- Keyboard Interface registers: KBE, KBF, KBLS
- BRG (Baud Rate Generator) registers: BRL, BDRCON
- Clock Prescaler register: CKRL
- Others: AUXR, AUXR1, CKCON0, CKCON1

Table 3 shows all SFRs with their address and their reset value.

Table 4-1. SFR Mapping

Reserved

▊

5. Oscillators

5.1 Overview

One oscillator is available for CPU:

• OSC used for high frequency (3 MHz to 40 MHz)

In order to optimize the power consumption and the execution time needed for a specific task, an internal prescaler feature has been implemented between the selected oscillator and the CPU.

5.2 Registers

Table 5-1. Clock Reload Register

Reset Value = 1111 1111b Not bit addressable

5.2.1 Prescaler Divider

A hardware RESET puts the prescaler divider in the following state:

- CKRL = FFh: $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSC}/2$ (Standard C51 feature) KS signal selects OSC: $F_{CLK\,OUT} = F_{OSC}$
- Any value between FFh down to 00h can be written by software into CKRL register in order to divide frequency of the selected oscillator:
	- CKRL = 00h: minimum frequency
		- $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSC}/1020$ (Standard Mode)
		- $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSC} / 510$ (X2 Mode)
	- CKRL = FFh: maximum frequency
		- $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSC}/2$ (Standard Mode)
		- $F_{CLK CPU} = F_{CLK PERIPH} = F_{OSC}$ (X2 Mode)
	- $F_{CLK CPU}$ and $F_{CLK PERIPH}$ In X2 mode:

$$
F_{CPU} = F_{CLKPERIPH} = \frac{F_{OSC}}{2 \times (255 - CKRL)}
$$

In X1 mode:
$$
F_{CPU} = F_{CLKPERIPH} = \frac{F_{OSCA}}{4 \times (255 - CKRL)}
$$

6. Enhanced Features

In comparison to the original 80C52, the microcontrollers implement the following new features:

- X2 option
- Dual Data Pointer
- Extended RAM
- Programmable Counter Array (PCA)
- Hardware Watchdog
- 4-level Interrupt Priority System
- Power-off Flag
- Power On Reset
- ONCE mode
- ALE disabling
- Some enhanced features are also located in the UART and the Timer 2

6.1 X2 Feature and OSC Clock Generation

The microcontroller core needs only 6 clock periods per machine cycle. This feature called "X2" provides the following advantages:

- Divides frequency crystals by 2 (cheaper crystals) while keeping same CPU power.
- Saves power consumption while keeping same CPU power (oscillator power saving).
- Saves power consumption by dividing dynamically the operating frequency by 2 in operating and idle modes.
- Increases CPU power by 2 while keeping same crystal frequency.

In order to keep the original C51 compatibility, a divider by 2 is inserted between the XTAL1 signal and the main clock input of the core (phase generator). This divider may be disabled by software.

6.1.1 Description

The clock for the whole circuit and peripherals is first divided by two before being used by the CPU core and the peripherals.

This allows any cyclic ratio to be accepted on XTAL1 input. In X2 mode, as this divider is bypassed, the signals on XTAL1 must have a cyclic ratio between 40 to 60%.

Figure 6-1 shows the clock generation block diagram. X2 bit is validated on the rising edge of the XTAL1 \div 2 to avoid glitches when switching from X2 to standard mode. Figure 6-2 shows the switching mode waveforms.

Figure 6-1. Clock Generation Diagram

Figure 6-2. Mode Switching Waveforms

The X2 bit in the CKCON0 register (see Table 6-1) allows to switch from 12 clock periods per instruction to 6 clock periods and vice versa. At reset, the speed is set according to X2 bit of Hardware Config Byte (HCB). By default, Standard mode is activated. Setting the X2 bit activates the X2 feature (X2 mode).

The T0X2, T1X2, T2X2, UARTX2, PCAX2 and WDX2 bits in the CKCON0 register (Table 6-1) allow to switch from standard peripheral speed (12 clock periods per peripheral clock cycle) to fast peripheral speed (6 clock periods per peripheral clock cycle). These bits are active only in X2 mode.

Table 6-1. CKCON0 Register CKCON0 - Clock Control Register (8Fh)

	6	5		3	2		0
	WDX2	PCAX ₂	SIX ₂	T2X2	T1X2	T0X2	X2
Bit Number	Bit Mnemonic	Description					
7		Reserved Do not set this bit.					
6	WDX2	Watchdog clock (This control bit is validated when the CPU clock X2 is set; when X2 is low, this bit has no effect). Cleared to select 6 clock periods per peripheral clock cycle. Set to select 12 clock periods per peripheral clock cycle.					

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Reset Value = 0000 000'HCB.X2'b (see Hardware Config Byte)

Not bit addressable

7. Dual Data Pointer Register

The additional data pointer can be used to speed up code execution and reduce code size.

The dual DPTR structure is a way by which the chip will specify the address of an external data memory location. There are two 16-bit DPTR registers that address the external memory, and a single bit called DPS = $AUXR1.0$ (see Table 7-1) that allows the program code to switch between them (Refer to Figure 7-1).

Table 7-1. AUXR1 Register AUXR1- Auxiliary Register 1(0A2h)

Reset Value: XXXX XXXX0b Not bit addressable

Note: 1. Bit 2 stuck at 0; this allows to use INC AUXR1 to toggle DPS without changing GF3.

7.1 Assembly Language

; Block move using dual data pointers ; Modifies DPTR0, DPTR1, A and PSW ; note: DPS exits opposite of entry state ; unless an extra INC AUXR1 is added ; 00A2 AUXR1 EQU 0A2H ; 0000 909000MOV DPTR,#SOURCE ; address of SOURCE 0003 05A2 INC AUXR1 ; switch data pointers 0005 90A000 MOV DPTR,#DEST ; address of DEST 0008 LOOP: 0008 05A2 INC AUXR1 ; switch data pointers 000A E0 MOVX A,@DPTR ; get a byte from SOURCE 000B A3 INC DPTR ; increment SOURCE address 000C 05A2 INC AUXR1 ; switch data pointers 000E F0 MOVX @DPTR,A ; write the byte to DEST 000F A3 INC DPTR ; increment DEST address 0010 70F6JNZ LOOP ; check for 0 terminator 0012 05A2 INC AUXR1 ; (optional) restore DPS

INC is a short (2 bytes) and fast (12 clocks) way to manipulate the DPS bit in the AUXR1 SFR. However, note that the INC instruction does not directly force the DPS bit to a particular state, but simply toggles it. In simple routines, such as the block move example, only the fact that DPS is toggled in the proper sequence matters, not its actual value. In other words, the block move routine works the same whether DPS is '0' or '1' on entry. Observe that without the last instruction (INC AUXR1), the routine will exit with DPS in the opposite state.

8. Expanded RAM (XRAM)

The AT80C51RD2 devices provide additional Bytes of Random Access Memory (RAM) space for increased data parameter handling and high level language usage.

The devices have expanded RAM in external data space; maximum size and location are described in Table 8-1.

Table 8-1. Expanded RAM

The AT80C51RD2 has internal data memory that is mapped into four separate segments.

The four segments are:

- 1. The Lower 128 bytes of RAM (addresses 00h to 7Fh) are directly and indirectly addressable.
- 2. The Upper 128 bytes of RAM (addresses 80h to FFh) are indirectly addressable only.
- 3. The Special Function Registers (SFRs) (addresses 80h to FFh) are directly addressable only.
- 4. The expanded RAM bytes are indirectly accessed by MOVX instructions, and with the EXTRAM bit cleared in the AUXR register (see Table 8-1).

The lower 128 bytes can be accessed by either direct or indirect addressing. The Upper 128 bytes can be accessed by indirect addressing only. The Upper 128 bytes occupy the same address space as the SFR. That means they have the same address, but are physically separate from SFR space.

Figure 8-1. Internal and External Data Memory Address

When an instruction accesses an internal location above address 7Fh, the CPU knows whether the access is to the upper 128 bytes of data RAM or to SFR space by the addressing mode used in the instruction.

• Instructions that use direct addressing access SFR space. For example: MOV 0A0H, # data, accesses the SFR at location 0A0h (which is P2).

- Instructions that use indirect addressing access the Upper 128 bytes of data RAM. For example: MOV @R0, # data where R0 contains 0A0h, accesses the data byte at address 0A0h, rather than P2 (whose address is 0A0h).
- The XRAM bytes can be accessed by indirect addressing, with EXTRAM bit cleared and MOVX instructions. This part of memory which is physically located on-chip, logically occupies the first bytes of external data memory. The bits XRS0 and XRS1 are used to hide a part of the available XRAM as explained in Table 8-1. This can be useful if external peripherals are mapped at addresses already used by the internal XRAM.
- With **EXTRAM = 0,** the XRAM is indirectly addressed, using the MOVX instruction in combination with any of the registers R0, R1 of the selected bank or DPTR. An access to XRAM will not affect ports P0, P2, P3.6 (WR) and P3.7 (RD). For example, with EXTRAM $=$ 0, MOVX @R0, # data where R0 contains 0A0H, accesses the XRAM at address 0A0H rather than external memory. An access to external data memory locations higher than the accessible size of the XRAM will be performed with the MOVX DPTR instructions in the same way as in the standard 80C51, with P0 and P2 as data/address busses, and P3.6 and P3.7 as write and read timing signals. Accesses to XRAM above 0FFH can only be done by the use of DPTR.
- With $EXTRAN = 1$, MOVX @Ri and MOVX @DPTR will be similar to the standard 80C51. MOVX @ Ri will provide an eight-bit address multiplexed with data on Port 0 and any output port pins can be used to output higher order address bits. This is to provide the external paging capability. MOVX @DPTR will generate a sixteen-bit address. Port2 outputs the highorder eight address bits (the contents of DPH) while Port0 multiplexes the low-order eight address bits (DPL) with data. MOVX @ Ri and MOVX @DPTR will generate either read or write signals on P3.6 (\overline{WR}) and P3.7 (\overline{RD}).

The stack pointer (SP) may be located anywhere in the 256 bytes RAM (lower and upper RAM) internal data memory. The stack may not be located in the XRAM.

The M0 bit allows to stretch the XRAM timings; if M0 is set, the read and write pulses are extended from 6 to 30 clock periods. This is useful to access external slow peripherals.

Table 8-2. AUXR Register AUXR - Auxiliary Register (8Eh)

Reset Value = XX0X 00'HSB.XRAM'0b (see Table 8-1)

Not bit addressable

9. Timer 2

The Timer 2 in the AT80C51RD2 is the standard C52 Timer 2.

It is a 16-bit timer/counter: the count is maintained by two eight-bit timer registers, TH2 and TL2 are cascaded. It is controlled by T2CON (Table 9-1) and T2MOD (Table 9-2) registers. Timer 2 operation is similar to Timer 0 and Timer 1. $C/T2$ selects $F_{\rm osc}/12$ (timer operation) or external pin T2 (counter operation) as the timer clock input. Setting TR2 allows TL2 to be incremented by the selected input.

Timer 2 has 3 operating modes: capture, auto-reload and Baud Rate Generator. These modes are selected by the combination of RCLK, TCLK and CP/RL2 (T2CON).

Refer to the Atmel 8-bit Microcontroller Hardware description for Capture and Baud Rate Generator Modes.

Timer 2 includes the following enhancements:

- Auto-reload mode with up or down counter
- Programmable clock-output

9.1 Auto-reload Mode

The auto-reload mode configures Timer 2 as a 16-bit timer or event counter with automatic reload. If DCEN bit in T2MOD is cleared, Timer 2 behaves as in 80C52 (refer to the Atmel 8-bit Microcontroller Hardware description). If DCEN bit is set, Timer 2 acts as an Up/down timer/counter as shown in Figure 9-1. In this mode the T2EX pin controls the direction of count.

When T2EX is high, Timer 2 counts up. Timer overflow occurs at FFFFh which sets the TF2 flag and generates an interrupt request. The overflow also causes the 16-bit value in RCAP2H and RCAP2L registers to be loaded into the timer registers TH2 and TL2.

When T2EX is low, Timer 2 counts down. Timer underflow occurs when the count in the timer registers TH2 and TL2 equals the value stored in RCAP2H and RCAP2L registers. The underflow sets TF2 flag and reloads FFFFh into the timer registers.

The EXF2 bit toggles when Timer 2 overflows or underflows according to the direction of the count. EXF2 does not generate any interrupt. This bit can be used to provide 17-bit resolution.

Figure 9-1. Auto-Reload Mode Up/Down Counter (DCEN = 1)

9.2 Programmable Clock-Output

In the clock-out mode, Timer 2 operates as a 50% duty-cycle, programmable clock generator (see Figure 9-2). The input clock increments TL2 at frequency $F_{\text{Cl K PFRIPH}}/2$. The timer repeatedly counts to overflow from a loaded value. At overflow, the contents of RCAP2H and RCAP2L registers are loaded into TH2 and TL2. In this mode, Timer 2 overflows do not generate interrupts. The formula gives the clock-out frequency as a function of the system oscillator frequency and the value in the RCAP2H and RCAP2L registers:

$$
Clock-OutFrequency = \frac{F_{CLKPERIPH}}{4 \times (65536 - RCAP2H/RCAP2L)}
$$

For a 16 MHz system clock, Timer 2 has a programmable frequency range of 61 Hz $(F_{CLK PERIPH}/2^{16})$ to 4 MHz ($F_{CLK PERIPH}/4$). The generated clock signal is brought out to T2 pin (P1.0).

Timer 2 is programmed for the clock-out mode as follows:

- Set T2OE bit in T2MOD register.
- Clear C/T2 bit in T2CON register.
- Determine the 16-bit reload value from the formula and enter it in RCAP2H/RCAP2L registers.
- Enter a 16-bit initial value in timer registers TH2/TL2. It can be the same as the reload value or a different one depending on the application.
- To start the timer, set TR2 run control bit in T2CON register.

It is possible to use Timer 2 as a baud rate generator and a clock generator simultaneously. For this configuration, the baud rates and clock frequencies are not independent since both functions use the values in the RCAP2H and RCAP2L registers.

Table 9-1. T2CON Register T2CON - Timer 2 Control Register (C8h)

Reset Value = 0000 0000b Bit addressable

J.

Table 9-2. T2MOD Register

T2MOD - Timer 2 Mode Control Register (C9h)

Reset Value = XXXX XX00b Not bit addressable

10. Programmable Counter Array (PCA)

The PCA provides more timing capabilities with less CPU intervention than the standard timer/counters. Its advantages include reduced software overhead and improved accuracy. The PCA consists of a dedicated timer/counter which serves as the time base for an array of five compare/capture modules. Its clock input can be programmed to count any one of the following signals:

- Peripheral clock frequency ($F_{CLK\,PERIPH}$) ÷ 6
- Peripheral clock frequency ($F_{CLK\,PERIPH}$) ÷ 2
- Timer 0 overflow
- External input on ECI (P1.2)

Each compare/capture modules can be programmed in any one of the following modes:

- Rising and/or falling edge capture
- Software timer
- High-speed output
- Pulse width modulator

Module 4 can also be programmed as a Watchdog Timer (see Section "PCA Watchdog Timer", page 33).

When the compare/capture modules are programmed in the capture mode, software timer, or high-speed output mode, an interrupt can be generated when the module executes its function. All five modules plus the PCA timer overflow share one interrupt vector.

The PCA timer/counter and compare/capture modules share Port 1 for external I/O. These pins are listed below. If the port is not used for the PCA, it can still be used for standard I/O.

The PCA timer is a common time base for all five modules (see Figure 10-1). The timer count source is determined from the CPS1 and CPS0 bits in the CMOD register (Table 10-1) and can be programmed to run at:

- 1/6 the peripheral clock frequency $(F_{\text{C-K PERIPH}})$
- $1/2$ the peripheral clock frequency ($F_{CLK\,PERIPH}$)
- The Timer 0 overflow
- The input on the ECI pin (P1.2)

Table 10-1. CMOD Register

CMOD - PCA Counter Mode Register (D9h)

Reset Value = 00XX X000b Not bit addressable

The CMOD register includes three additional bits associated with the PCA (see Figure 10-4 and Table 10-1).

- The CIDL bit which allows the PCA to stop during idle mode.
- The WDTE bit which enables or disables the watchdog function on module 4.
- The ECF bit which when set causes an interrupt and the PCA overflow flag CF (in the CCON SFR) to be set when the PCA timer overflows.

The CCON register contains the run control bit for the PCA and the flags for the PCA timer (CF) and each module (see Table 10-2).

- Bit CR (CCON.6) must be set by software to run the PCA. The PCA is shut off by clearing this bit.
- Bit CF: The CF bit (CCON.7) is set when the PCA counter overflows and an interrupt will be generated if the ECF bit in the CMOD register is set. The CF bit can only be cleared by software.

• Bits 0 through 4 are the flags for the modules (bit 0 for module 0, bit 1 for module 1, etc.) and are set by hardware when either a match or a capture occurs. These flags can only be cleared by software.

Table 10-2. CCON Register

CCON - PCA Counter Control Register (D8h)

Reset Value = 000X 0000b Not bit addressable

The watchdog timer function is implemented in module 4 (see Figure 10-4).

The PCA interrupt system is shown in Figure 10-2.

Figure 10-2. PCA Interrupt System

PCA Modules: each one of the five compare/capture modules has six possible functions. It can perform:

- 16-bit Capture, positive-edge triggered
- 16-bit Capture, negative-edge triggered
- 16-bit Capture, both positive and negative-edge triggered
- 16-bit Software Timer
- 16-bit High-speed Output
- 8-bit Pulse Width Modulator

In addition, module 4 can be used as a Watchdog Timer.

Each module in the PCA has a special function register associated with it. These registers are: CCAPM0 for module 0, CCAPM1 for module 1, etc. (see Table 10-3). The registers contain the bits that control the mode that each module will operate in.

- The ECCF bit (CCAPMn.0 where $n = 0, 1, 2, 3$, or 4 depending on the module) enables the CCF flag in the CCON SFR to generate an interrupt when a match or compare occurs in the associated module.
- PWM (CCAPMn.1) enables the pulse width modulation mode.
- The TOG bit (CCAPMn.2) when set causes the CEX output associated with the module to toggle when there is a match between the PCA counter and the module's capture/compare register.
- The match bit MAT (CCAPMn.3) when set will cause the CCFn bit in the CCON register to be set when there is a match between the PCA counter and the module's capture/compare register.
- The next two bits CAPN (CCAPMn.4) and CAPP (CCAPMn.5) determine the edge that a capture input will be active on. The CAPN bit enables the negative edge, and the CAPP bit enables the positive edge. If both bits are set both edges will be enabled and a capture will occur for either transition.
- The last bit in the register ECOM (CCAPMn.6) when set enables the comparator function.

Table 10-3 shows the CCAPMn settings for the various PCA functions.

Table 10-3. CCAPMn Registers $(n = 0-4)$

CCAPM0 - PCA Module 0 Compare/Capture Control Register (0DAh)

CCAPM1 - PCA Module 1 Compare/Capture Control Register (0DBh)

CCAPM2 - PCA Module 2 Compare/Capture Control Register (0DCh)

CCAPM3 - PCA Module 3 Compare/Capture Control Register (0DDh)

CCAPM4 - PCA Module 4 Compare/Capture Control Register (0DEh)

Reset Value = X000 0000b Not bit addressable

Table 10-4. PCA Module Modes (CCAPMn Registers)

There are two additional registers associated with each of the PCA modules. They are CCAPnH and CCAPnL and these are the registers that store the 16-bit count when a capture occurs or a compare should occur. When a module is used in the PWM mode these registers are used to control the duty cycle of the output (see Table 10-5 and Table 10-6).

Table 10-5. CCAPnH Registers $(n = 0-4)$

- CCAP0H PCA Module 0 Compare/Capture Control Register High (0FAh)
- CCAP1H PCA Module 1 Compare/Capture Control Register High (0FBh)
- CCAP2H PCA Module 2 Compare/Capture Control Register High (0FCh)
- CCAP3H PCA Module 3 Compare/Capture Control Register High (0FDh)
- CCAP4H PCA Module 4 Compare/Capture Control Register High (0FEh)

Reset Value = 0000 0000b Not bit addressable

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Table 10-6. CCAPnL Registers $(n = 0-4)$

CCAP0L - PCA Module 0 Compare/Capture Control Register Low (0EAh)

CCAP1L - PCA Module 1 Compare/Capture Control Register Low (0EBh)

CCAP2L - PCA Module 2 Compare/Capture Control Register Low (0ECh)

CCAP3L - PCA Module 3 Compare/Capture Control Register Low (0EDh)

CCAP4L - PCA Module 4 Compare/Capture Control Register Low (0EEh)

Reset Value = 0000 0000b

Not bit addressable

Table 10-7. CH Register

CH - PCA Counter Register High (0F9h)

Reset Value = 0000 0000b Not bit addressable

Table 10-8. CL Register

CL - PCA Counter Register Low (0E9h)

Reset Value = 0000 0000b Not bit addressable

10.1 PCA Capture Mode

To use one of the PCA modules in the capture mode either one or both of the CCAPM bits CAPN and CAPP for that module must be set. The external CEX input for the module (on port 1) is sampled for a transition. When a valid transition occurs the PCA hardware loads the value of the PCA counter registers (CH and CL) into the module's capture registers (CCAPnL and CCA-PnH). If the CCFn bit for the module in the CCON SFR and the ECCFn bit in the CCAPMn SFR are set then an interrupt will be generated (see Figure 10-3).

Figure 10-3. PCA Capture Mode

10.2 16-bit Software Timer/ Compare Mode

The PCA modules can be used as software timers by setting both the ECOM and MAT bits in the modules CCAPMn register. The PCA timer will be compared to the module's capture registers and when a match occurs an interrupt will occur if the CCFn (CCON SFR) and the ECCFn (CCAPMn SFR) bits for the module are both set (see Figure 10-4).

Figure 10-4. PCA Compare Mode and PCA Watchdog Timer

Before enabling ECOM bit, CCAPnL and CCAPnH should be set with a non zero value, otherwise an unwanted match could happen. Writing to CCAPnH will set the ECOM bit.

Once ECOM set, writing CCAPnL will clear ECOM so that an unwanted match doesn't occur while modifying the compare value. Writing to CCAPnH will set ECOM. For this reason, user software should write CCAPnL first, and then CCAPnH. Of course, the ECOM bit can still be controlled by accessing to CCAPMn register.

10.3 High-speed Output Mode

In this mode, the CEX output (on port 1) associated with the PCA module will toggle each time a match occurs between the PCA counter and the module's capture registers. To activate this mode the TOG, MAT, and ECOM bits in the module's CCAPMn SFR must be set (see Figure 10-5).

A prior write must be done to CCAPnL and CCAPnH before writing the ECOMn bit.

Figure 10-5. PCA High-speed Output Mode

Before enabling ECOM bit, CCAPnL and CCAPnH should be set with a non zero value, otherwise an unwanted match could occur.

Once ECOM is set, writing CCAPnL will clear ECOM so that an unwanted match doesn't occur while modifying the compare value. Writing to CCAPnH will set ECOM. For this reason, user software should write CCAPnL first, and then CCAPnH. Of course, the ECOM bit can still be controlled by accessing the CCAPMn register.

10.4 Pulse Width Modulator Mode

All of the PCA modules can be used as PWM outputs. Figure 10-6 shows the PWM function. The frequency of the output depends on the source for the PCA timer. All of the modules will have the same frequency of output because they all share the PCA timer. The duty cycle of each module is independently variable using the module's capture register CCAPLn. When the value of the PCA CL SFR is less than the value in the module's CCAPLn SFR the output will be low, when it is equal to or greater than the output will be high. When CL overflows from FF to 00, CCAPLn is reloaded with the value in CCAPHn. This allows updating the PWM without glitches. The PWM and ECOM bits in the module's CCAPMn register must be set to enable the PWM mode.

Figure 10-6. PCA PWM Mode

10.5 PCA Watchdog Timer

An on-board watchdog timer is available with the PCA to improve the reliability of the system without increasing chip count. Watchdog timers are useful for systems that are susceptible to noise, power glitches, or electrostatic discharge. Module 4 is the only PCA module that can be programmed as a watchdog. However, this module can still be used for other modes if the watchdog is not needed. Figure 10-4 shows a diagram of how the watchdog works. The user pre-loads a 16-bit value in the compare registers. Just like the other compare modes, this 16-bit value is compared to the PCA timer value. If a match is allowed to occur, an internal reset will be generated. This will not cause the RST pin to be driven high.

In order to hold off the reset, the user has three options:

- 1. Periodically change the compare value so it will never match the PCA timer.
- 2. Periodically change the PCA timer value so it will never match the compare values.
- 3. Disable the watchdog by clearing the WDTE bit before a match occurs and then reenable it.

The first two options are more reliable because the watchdog timer is never disabled as in option #3. If the program counter ever goes astray, a match will eventually occur and cause an internal reset. The second option is also not recommended if other PCA modules are being used. Remember, the PCA timer is the time base for all modules; changing the time base for other modules would not be a good idea. Thus, in most applications the first solution is the best option.

This watchdog timer won't generate a reset out on the reset pin.

11. Serial I/O Port

The serial I/O port in the AT80C51RD2 is compatible with the serial I/O port in the 80C52. It provides both synchronous and asynchronous communication modes. It operates as a Universal Asynchronous Receiver and Transmitter (UART) in three full-duplex modes (Modes 1, 2 and 3). Asynchronous transmission and reception can occur simultaneously and at different baud rates

Serial I/O port includes the following enhancements:

- Framing error detection
- Automatic address recognition

11.1 Framing Error Detection

Framing bit error detection is provided for the three asynchronous modes (modes 1, 2 and 3). To enable the framing bit error detection feature, set SMOD0 bit in PCON register (see Figure 11- 1).

When this feature is enabled, the receiver checks each incoming data frame for a valid stop bit. An invalid stop bit may result from noise on the serial lines or from simultaneous transmission by two CPUs. If a valid stop bit is not found, the Framing Error bit (FE) in SCON register (see Table 11-4) bit is set.

Software may examine FE bit after each reception to check for data errors. Once set, only software or a reset can clear FE bit. Subsequently, received frames with valid stop bits cannot clear FE bit. When FE feature is enabled, RI rises on stop bit instead of the last data bit (see Figure 11-2 and Figure 11-3).

Figure 11-3. UART Timings in Modes 2 and 3

11.2 Automatic Address Recognition

The automatic address recognition feature is enabled when the multiprocessor communication feature is enabled (SM2 bit in SCON register is set).

Implemented in hardware, automatic address recognition enhances the multiprocessor communication feature by allowing the serial port to examine the address of each incoming command frame. Only when the serial port recognizes its own address, the receiver sets RI bit in SCON register to generate an interrupt. This ensures that the CPU is not interrupted by command frames addressed to other devices.

If desired, you may enable the automatic address recognition feature in mode 1. In this configuration, the stop bit takes the place of the ninth data bit. Bit RI is set only when the received command frame address matches the device's address and is terminated by a valid stop bit.

To support automatic address recognition, a device is identified by a given address and a broadcast address.

Note: The multiprocessor communication and automatic address recognition features cannot be enabled in mode 0 (i.e. setting SM2 bit in SCON register in mode 0 has no effect).

11.2.1 Given Address

Each device has an individual address that is specified in SADDR register; the SADEN register is a mask byte that contains don't care bits (defined by zeros) to form the device's given address. The don't care bits provide the flexibility to address one or more slaves at a time. The following example illustrates how a given address is formed.

To address a device by its individual address, the SADEN mask byte must be 1111 1111b. For example:

SADDR0101 0110b SADEN1111 1100b Given0101 01XXb

The following is an example of how to use given addresses to address different slaves:

Slave A:SADDR1111 0001b SADEN1111 1010b Given1111 0X0Xb

```
Slave B:SADDR1111 0011b
   SADEN1111 1001b
```


Given1111 0XX1b

```
Slave C:SADDR1111 0010b
   SADEN1111 1101b
   Given1111 00X1b
```
The SADEN byte is selected so that each slave may be addressed separately.

For slave A, bit 0 (the LSB) is a don't-care bit; for slaves B and C, bit 0 is a 1. To communicate with slave A only, the master must send an address where bit 0 is clear (e.g. 1111 0000b).

For slave A, bit 1 is a 1; for slaves B and C, bit 1 is a don't care bit. To communicate with slaves B and C, but not slave A, the master must send an address with bits 0 and 1 both set (e.g. 1111 0011b).

To communicate with slaves A, B and C, the master must send an address with bit 0 set, bit 1 clear, and bit 2 clear (e.g. 1111 0001b).

11.2.2 Broadcast Address

A broadcast address is formed from the logical OR of the SADDR and SADEN registers with zeros defined as don't-care bits, e.g.:

SADDR0101 0110b SADEN1111 1100b Broadcast = SADDR OR SADEN1111 111Xb

The use of don't-care bits provides flexibility in defining the broadcast address, however in most applications, a broadcast address is FFh. The following is an example of using broadcast addresses:

```
Slave A:SADDR1111 0001b
   SADEN1111 1010b
   Broadcast1111 1X11b,
Slave B:SADDR1111 0011b
   SADEN1111 1001b
   Broadcast1111 1X11B,
Slave C:SADDR = 1111 0010b
   SADEN1111 1101b
   Broadcast1111 1111b
```
For slaves A and B, bit 2 is a don't care bit; for slave C, bit 2 is set. To communicate with all of the slaves, the master must send an address FFh. To communicate with slaves A and B, but not slave C, the master can send and address FBh.
11.2.3 Reset Addresses

On reset, the SADDR and SADEN registers are initialized to 00h, i.e. the given and broadcast addresses are XXXX XXXXb (all don't-care bits). This ensures that the serial port will reply to any address, and so, that it is backwards compatible with the 80C51 microcontrollers that do not support automatic address recognition.

Table 11-1. SADEN Register

SADEN - Slave Address Mask Register (B9h)

Reset Value = 0000 0000b Not bit addressable

Table 11-2. SADDR Register SADDR - Slave Address Register (A9h)

Reset Value = 0000 0000b

Not bit addressable

11.3 Baud Rate Selection for UART for Mode 1 and 3

The Baud Rate Generator for transmit and receive clocks can be selected separately via the T2CON and BDRCON registers.

Figure 11-4. Baud Rate selection

TCLK (T2CON)	RCLK (T2CON)	TBCK (BDRCON)	RBCK (BDRCON)	Clock Source UART Tx	Clock Source UART Rx
0	$\mathbf 0$	0	0	Timer 1	Timer 1
	0	0	0	Timer 2	Timer 1
$\mathbf 0$		0	$\mathbf 0$	Timer 1	Timer 2
		0	0	Timer 2	Timer 2
X	$\mathbf 0$		0	INT_BRG	Timer 1
X			0	INT_BRG	Timer 2
0	X	0		Timer 1	INT_BRG
	X	0		Timer 2	INT_BRG
X	X			INT_BRG	INT_BRG

Table 11-3. Baud Rate Selection Table UART

11.3.1 Internal Baud Rate Generator (BRG)

When the internal Baud Rate Generator is used, the Baud Rates are determined by the BRG overflow depending on the BRL reload value, the value of SPD bit (Speed Mode) in BDRCON register and the value of the SMOD1 bit in PCON register.

Figure 11-5. Internal Baud Rate

• The baud rate for UART is token by formula:

$$
\textit{BaudRate} = \frac{2_{SMOD} \times F_{CLKPERIPH}}{2 \times 2 \times 6 \langle 1 - SPD \rangle \times 16 \times [256 - (BRL)]}
$$

$$
(BRL) = 256 - \frac{2_{SMOD1} \times F_{CLKPERIPH}}{2 \times 2 \times 6_{(1-SPD)} \times 16 \times BaudRate}
$$

Table 11-4. SCON Register SCON - Serial Control Register (98h)

Reset Value = 0000 0000b

Bit addressable

Table 11-5. Example of Computed Value when $X2 = 1$, SMOD1 = 1, SPD = 1

Table 11-6. Example of Computed Value when $X2 = 0$, SMOD1 = 0, SPD = 0

Baud Rates	F_{osc} =16.384 MHz		$F_{\rm osc}$ =24 MHz		
	BRL	Error $(\%)$	BRL	Error $(\%)$	
4800	247	1.23	243	0.16	
2400	238	1.23	230	0.16	
1200	220	1.23	202	3.55	
600	185	0.16	152	0.16	

The baud rate generator can be used for mode 1 or 3 (see Figure 11-4.), but also for mode 0 for UART, thanks to the bit SRC located in BDRCON register (Table 11-13.)

11.4 UART Registers

Table 11-7. SADEN Register

SADEN - Slave Address Mask Register for UART (B9h)

Reset Value = 0000 0000b

Table 11-8. SADDR Register

SADDR - Slave Address Register for UART (A9h)

Reset Value = 0000 0000b

Table 11-9. SBUF Register

SBUF - Serial Buffer Register for UART (99h)

Reset Value = XXXX XXXXb

Table 11-10. BRL Register

BRL - Baud Rate Reload Register for the internal baud rate generator, UART (9Ah)

Reset Value = 0000 0000b

Table 11-11. T2CON Register

T2CON - Timer 2 Control Register (C8h)

Reset Value = 0000 0000b Bit addressable

Table 11-12. PCON Register PCON - Power Control Register (87h)

Reset Value = 00X1 0000b Not bit addressable

Power-off flag reset value will be 1 only after a power on (cold reset). A warm reset doesn't affect the value of this bit.

Table 11-13. BDRCON Register

BDRCON - Baud Rate Control Register (9Bh)

Reset Value = XXX0 0000b Not bit addressable

12. Interrupt System

The AT80C51RD2 have a total of 8 interrupt vectors: two external interrupts ($\overline{\text{INT0}}$ and $\overline{\text{INT1}}$), three timer interrupts (timers 0, 1 and 2), the serial port interrupt, Keyboard interrupt and the PCA global interrupt. These interrupts are shown in Figure 12-1.

Each of the interrupt sources can be individually enabled or disabled by setting or clearing a bit in the Interrupt Enable register (Table 12-5 and Table 12-3). This register also contains a global disable bit, which must be cleared to disable all interrupts at once.

Each interrupt source also can be individually programmed to one out of four priority levels by setting or clearing a bit in the Interrupt Priority register (Table 12-6) and in the Interrupt Priority High register (Table 12-4 and Table 12-5) shows the bit values and priority levels associated with each combination.

12.1 Registers

The PCA interrupt vector is located at address 0033H, the Keyboard interrupt vector is located at address 004BH. All other vectors addresses are the same as standard C52 devices.

A low-priority interrupt can be interrupted by a high priority interrupt, but not by another low-priority interrupt. A high-priority interrupt can't be interrupted by any other interrupt source.

If two interrupt requests of different priority levels are received simultaneously, the request of higher priority level is serviced. If interrupt requests of the same priority level are received simultaneously, an internal polling sequence determines which request is serviced. Thus within each priority level there is a second priority structure determined by the polling sequence.

Table 12-2. IEO Register

IE0 - Interrupt Enable Register (A8h)

Reset Value = 0000 0000b Bit addressable

Table 12-3. IPL0 Register

IPL0 - Interrupt Priority Register (B8h)

Reset Value = X000 0000b Bit addressable

Table 12-4. IPH0 Register

IPH0 - Interrupt Priority High Register (B7h)

Reset Value = X000 0000b Not bit addressable

Table 12-5. IE1 Register

IE1 - Interrupt Enable Register (B1h)

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Reset Value = XXXX XXX0b

Bit addressable

Table 12-6. IPL1 Register

IPL1 - Interrupt Priority Register (B2h)

Reset Value = XXXX XXX0b Bit addressable

Table 12-7. IPH1 Register

IPH1 - Interrupt Priority High Register (B3h)

Reset Value = XXXX XXX0b Not bit addressable

12.2 Interrupt Sources and Vector Addresses

Table 12-8. Interrupt Sources and Vector Addresses

 \blacksquare

13. Keyboard Interface

The AT80C51RD2 implement a keyboard interface allowing the connection of a 8 x n matrix keyboard. It is based on 8 inputs with programmable interrupt capability on both high or low level. These inputs are available as alternate function of P1 and allow to exit from idle and powerdown modes.

The keyboard interfaces with the C51 core through 3 special function registers: KBLS, the Keyboard Level Selection register (Table 13-3), KBE, The Keyboard Interrupt Enable register (Table 13-2), and KBF, the Keyboard Flag register (Table 13-1).

13.0.1 Interrupt

The keyboard inputs are considered as 8 independent interrupt sources sharing the same interrupt vector. An interrupt enable bit (KBD in IE1) allows global enable or disable of the keyboard interrupt (see Figure 13-1). As detailed in Figure 13-2 each keyboard input has the capability to detect a programmable level according to KBLS.x bit value. Level detection is then reported in interrupt flags KBF.x that can be masked by software using KBE.x bits.

This structure allow keyboard arrangement from 1 x n to 8 x n matrix and allows usage of P1 inputs for other purpose.

KBLS.x

1

Figure 13-1. Keyboard Interface Block Diagram

Internal Pull-up

KBE.x

13.0.2 Power Reduction Mode

P1 inputs allow exit from idle and power-down modes as detailed in Section "Power-down Mode", page 56.

13.1 Registers

Table 13-1. KBF Register

KBF - Keyboard Flag Register (9Eh)

Reset Value = 0000 0000b

▊

Table 13-2. KBE Register

KBE - Keyboard Input Enable Register (9Dh)

Reset Value = 0000 0000b

Table 13-3. KBLS Register **KBLS -** Keyboard Level Selector Register (9Ch)

Reset Value = 0000 0000b

٦

14. Power Management

14.1 Idle Mode

An instruction that sets PCON.0 indicates that it is the last instruction to be executed before going into Idle mode. In Idle mode, the internal clock signal is gated off to the CPU, but not to the interrupt, Timer, and Serial Port functions. The CPU status is preserved in its entirety: the Stack Pointer, Program Counter, Program Status Word, Accumulator and all other registers maintain their data during idle. The port pins hold the logical states they had at the time Idle was activated. ALE and PSEN hold at logic high level.

There are two ways to terminate the Idle mode. Activation of any enabled interrupt will cause PCON.0 to be cleared by hardware, terminating the Idle mode. The interrupt will be serviced, and following RETI the next instruction to be executed will be the one following the instruction that put the device into idle.

The flag bits GF0 and GF1 can be used to give an indication if an interrupt occurred during normal operation or during idle. For example, an instruction that activates idle can also set one or both flag bits. When idle is terminated by an interrupt, the interrupt service routine can examine the flag bits.

The other way of terminating the Idle mode is with a hardware reset. Since the clock oscillator is still running, the hardware reset needs to be held active for only two machine cycles (24 oscillator periods) to complete the reset.

14.2 Power-down Mode

To save maximum power, a power-down mode can be invoked by software (refer to Table 11- 12, PCON register).

In power-down mode, the oscillator is stopped and the instruction that invoked power-down mode is the last instruction executed. The internal RAM and SFRs retain their value until the power-down mode is terminated. V_{CC} can be lowered to save further power. Either a hardware reset or an external interrupt can cause an exit from power-down. To properly terminate powerdown, the reset or external interrupt should not be executed before V_{CC} is restored to its normal operating level and must be held active long enough for the oscillator to restart and stabilize.

Only external interrupts INTO, INT1 and Keyboard Interrupts are useful to exit from power-down. Thus, the interrupt must be enabled and configured as level - or edge - sensitive interrupt input. When Keyboard Interrupt occurs after a power-down mode, 1024 clocks are necessary to exit to power-down mode and enter in operating mode.

Holding the pin low restarts the oscillator but bringing the pin high completes the exit as detailed in Figure 14-1. When both interrupts are enabled, the oscillator restarts as soon as one of the two inputs is held low and power-down exit will be completed when the first input is released. In this case, the higher priority interrupt service routine is executed. Once the interrupt is serviced, the next instruction to be executed after RETI will be the one following the instruction that put AT80C51RD2 into power-down mode.

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Figure 14-1. Power-down Exit Waveform

Exit from power-down by reset redefines all the SFRs, exit from power-down by external interrupt does no affect the SFRs.

Exit from power-down by either reset or external interrupt does not affect the internal RAM content.

Note: If idle mode is activated with power-down mode (IDL and PD bits set), the exit sequence is unchanged, when execution is vectored to interrupt, PD and IDL bits are cleared and idle mode is not entered.

Table 14-1 shows the state of ports during idle and power-down modes.

Table 14-1. State of Ports

Mode	Program Memory	ALE	PSEN	PORTO	PORT ₁	PORT ₂	PORT ₃
Idle	Internal			Port Data ⁽¹⁾	Port Data	Port Data	Port Data
Idle	External			Floating	Port Data	Address	Port Data
Power-down	Internal	0		Port Dat ⁽¹⁾	Port Data	Port Data	Port Data
Power-down	External	υ		Floating	Port Data	Port Data	Port Data

Note: 1. Port 0 can force a 0 level. A "one" will leave port floating.

15. Hardware Watchdog Timer

The WDT is intended as a recovery method in situations where the CPU may be subjected to software upset. The WDT consists of a 14-bit counter and the WatchDog Timer Reset (WDTRST) SFR. The WDT is by default disabled from exiting reset. To enable the WDT, user must write 01EH and 0E1H in sequence to the WDTRST, SFR location 0A6H. When WDT is enabled, it will increment every machine cycle while the oscillator is running and there is no way to disable the WDT except through reset (either hardware reset or WDT overflow reset). When WDT overflows, it will drive an output RESET HIGH pulse at the RST-pin.

15.1 Using the WDT

To enable the WDT, user must write 01EH and 0E1H in sequence to the WDTRST, SFR location 0A6H. When WDT is enabled, the user needs to service it by writing to 01EH and 0E1H to WDTRST to avoid WDT overflow. The 14-bit counter overflows when it reaches 16383 (3FFFH) and this will reset the device. When WDT is enabled, it will increment every machine cycle while the oscillator is running. Therefore, the user must reset the WDT at least every 16383 machine cycles. To reset the WDT the user must write 01EH and 0E1H to WDTRST. WDTRST is a write only register. The WDT counter cannot be read or written. When WDT overflows, it will generate an output RESET pulse at the RST-pin. The RESET pulse duration is 96 x $T_{CLK PERIPH}$, where $T_{\text{CLK PERIPH}}$ = 1/ $F_{\text{CLK PERIPH}}$. To make the best use of the WDT, it should be serviced in those sections of code that will periodically be executed within the time required to prevent a WDT reset.

To have a more powerful WDT, a 2⁷ counter has been added to extend the Time-out capability, ranking from 16 ms to 2s \circledR F_{osc} = 12 MHz. To manage this feature, refer to WDTPRG register description, Table 15-1.

Table 15-1. WDTRST Register

WDTRST - Watchdog Reset Register (0A6h)

Reset Value = XXXX XXXXb

Write only, this SFR is used to reset/enable the WDT by writing 01EH then 0E1H in sequence.

Table 15-2. WDTPRG Register WDTPRG - Watchdog Timer Out Register (0A7h)

Reset Value = XXXX X000

15.2 WDT During Power-down and Idle

In Power-down mode the oscillator stops, which means the WDT also stops. While in Powerdown mode the user does not need to service the WDT. There are 2 methods of exiting Powerdown mode: by a hardware reset or via a level activated external interrupt which is enabled prior to entering Power-down mode. When Power-down is exited with hardware reset, servicing the WDT should occur as normal, whenever the AT80C51RD2 is reset. Exiting Power-down with an interrupt is significantly different. The interrupt is held low long enough for the oscillator to stabilize. When the interrupt is brought high, the interrupt is serviced. To prevent the WDT from resetting the device while the interrupt pin is held low, the WDT is not started until the interrupt is pulled high. It is suggested that the WDT be reset during the interrupt service routine.

To ensure that the WDT does not overflow within a few states of exiting of power-down, it is better to reset the WDT just before entering power-down.

In the Idle mode, the oscillator continues to run. To prevent the WDT from resetting the AT80C51RD2 while in Idle mode, the user should always set up a timer that will periodically exit Idle, service the WDT, and re-enter Idle mode.

16. Power-off Flag

The Power-off flag allows the user to distinguish between a "cold start" reset and a "warm start" reset.

A cold start reset is the one induced by V_{CC} switch-on. A warm start reset occurs while V_{CC} is still applied to the device and could be generated for example by an exit from power-down.

The Power-off flag (POF) is located in PCON register (Table 16-1). POF is set by hardware when V_{CC} rises from 0 to its nominal voltage. The POF can be set or cleared by software allowing the user to determine the type of reset.

Table 16-1. PCON Register

PCON - Power Control Register (87h)

Reset Value = 00X1 0000b Not bit addressable

17. Reduced EMI Mode

The ALE signal is used to demultiplex address and data buses on port 0 when used with external program or data memory. Nevertheless, during internal code execution, ALE signal is still generated. In order to reduce EMI, ALE signal can be disabled by setting AO bit.

The AO bit is located in AUXR register at bit location 0. As soon as AO is set, ALE is no longer output but remains active during MOVX and MOVC instructions and external fetches. During ALE disabling, ALE pin is weakly pulled high.

Table 17-1. AUXR Register AUXR - Auxiliary Register (8Eh)

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17. Electrical Characteristics

Table 17-1. Absolute Maximum Ratings

17.1 DC Parameters for Standard Voltage

 T_A = 0°C to +70°C; V_{SS} = 0V; V_{CC} = 4.5V to 5.5V; F = 10 to 40 MHz $T_{\rm A}$ = -40°C to +85°C; V_{SS} = 0V; V_{CC} =4.5V to 5.5V; F = 10 to 40 MHz

17.2 DC Parameters for Standard Voltage (2)

TA = 0° C to +70 $^{\circ}$ C; V_{SS} = 0 V; V_{CC} = 2.7V to 5.5V; F = 10 to 40 MHz TA = -40°C to +85°C; V_{SS} = 0 V; V_{CC} = 2.7V to 5.5V; F = 10 to 40 MHz

Notes: 1. Operating I_{CC} is measured with all output pins disconnected; XTAL1 driven with T_{CLCH} , $T_{CHCL} = 5$ ns (see Figure 17-4.), $V_{IL} = V_{SS} + 0.5V$,

 $V_{IH} = V_{CC}$ - 0.5V; XTAL2 N.C.; \overline{EA} = RST = Port 0 = V_{CC} . I_{CC} would be slightly higher if a crystal oscillator used (see Figure 17-1).

- 2. Idle I_{CC} is measured with all output pins disconnected; XTAL1 driven with T_{CLCH} , $T_{CHCL} = 5$ ns, $V_{IL} = V_{SS} + 0.5V$, $V_{IH} = V_{CC}$ - 0.5V; XTAL2 N.C; Port 0 = V_{CC} ; \overline{EA} = RST = V_{SS} (see Figure 17-2).
- 3. Power-down I_{CC} is measured with all output pins disconnected; $\overline{EA} = V_{SS}$, PORT 0 = V_{CC}; XTAL2 NC.; RST = V_{SS} (see Figure 17-3).
- 4. Capacitance loading on Ports 0 and 2 may cause spurious noise pulses to be superimposed on the V_{O} s of ALE and Ports 1 and 3. The noise is due to external bus capacitance discharging into the Port 0 and Port 2 pins when these pins make 1 to 0 transitions during bus operation. In the worst cases (capacitive loading 100pF), the noise pulse on the ALE line may exceed 0.45V with maxi V_{OL} peak 0.6V. A Schmitt Trigger use is not necessary.
- 5. Typical are based on a limited number of samples and are not guaranteed. The values listed are at room temperature and 5V.

6. Under steady state (non-transient) conditions, I_{OL} must be externally limited as follows: Maximum I_{OL} per port pin: 10 mA Maximum I_{OL} per 8-bit port: Port 0: 26 mA Ports 1, 2 and 3: 15 mA Maximum total I_{OL} for all output pins: 71 mA If I_{OL} exceeds the test condition, V_{OL} may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test conditions.

7. For other values, please contact your sales office.

All other pins are disconnected.

Figure 17-2. I_{CC} Test Condition, Idle Mode

All other pins are disconnected.

Figure 17-3. I_{CC} Test Condition, Power-down Mode

All other pins are disconnected.

Figure 17-4. Clock Signal Waveform for I_{CC} Tests in Active and Idle Modes

17.3 AC Parameters

17.3.1 Explanation of the AC Symbols

Each timing symbol has 5 characters. The first character is always a "t" (stands for time). The other characters, depending on their positions, stand for the name of a signal or the logical status of that signal. The following is a list of all the characters and what they stand for.

Example: T_{AVLL} = Time for Address Valid to ALE Low. T_{LLPL} = Time for ALE Low to \overline{PSEN} Low.

(Load Capacitance for port 0, ALE and PSEN = 100 pF; Load Capacitance for all other outputs = 80 pF.)

Table 17-2 Table 17-5, and Table 17-7 give the description of each AC symbols.

Table 17-4, Table 17-6 and Table 17-8 give for each range the AC parameter.

Table 17-3, Table 17-4 and Table 17-9 gives the frequency derating formula of the AC parameter for each speed range description. To calculate each AC symbols. take the x value in the correponding column and use this value in the formula.

Example: $T_{LL|U}$ for 20 MHz, Standard clock. $x = 35$ ns $T = 50$ ns $T_{\text{CCIV}} = 4T - x = 165 \text{ ns}$

17.3.2 External Program Memory Characteristics

Table 17-2. Symbol Description

Table 17-3. AC Parameters for a Fix Clock

Symbol	-M		
	Min	Max	
$\sf T$	25		ns
$\mathsf{T}_{\mathsf{LHLL}}$	35		ns
T_{AVLL}	5		ns
$\mathsf{T}_{\mathsf{LLAX}}$	5		ns
$\mathsf{T}_{\mathsf{LLIV}}$		65	ns
T_{LLPL}	5		ns
$\mathsf{T}_{\mathsf{PLPH}}$	50		ns
$\mathsf{T}_{\mathsf{PLIV}}$		30	ns
$\mathsf{T}_{\mathsf{PXIX}}$	0		ns
$\mathsf{T}_{\mathsf{PXIZ}}$		10	ns
T_{AVIV}		80	ns
T _{PLAZ}		10	ns

17.3.3 External Program Memory Read Cycle

17.3.4 External Data Memory Characteristics

Table 17-5. Symbol Description

Table 17-6. AC Parameters for a Fix Clock

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17.3.5 External Data Memory Write Cycle

17.3.6 External Data Memory Read Cycle

17.3.7 Serial Port Timing - Shift Register Mode Table 17-7. Symbol Description

Table 17-8. AC Parameters for a Fix Clock

17.3.8 Shift Register Timing Waveforms

17.3.9 External Clock Drive Waveforms

17.3.10 AC Testing Input/Output Waveforms

AC inputs during testing are driven at V_{CC} - 0.5 for a logic "1" and 0.45V for a logic "0". Timing measurement are made at V_{IH} min for a logic "1" and V_{II} max for a logic "0".

17.3.11 Float Waveforms

For timing purposes as port pin is no longer floating when a 100 mV changes from load voltage occurs and begins to float when a 100 mV change from the loaded V_{OH}/V_{OL} level occurs. I_{OL}/I_{OH} $\geq \pm 20$ mA.

17.3.12 Clock Waveforms

Valid in normal clock mode. In X2 mode XTAL2 must be changed to XTAL2/2.

Figure 17-5. Internal Clock Signals

This diagram indicates when signals are clocked internally. The time it takes the signals to propagate to the pins, however, ranges from 25 to 125 ns. This propagation delay is dependent on variables such as temperature and pin loading. Propagation also varies from output to output and component. Typically though (T_A = 25°C fully loaded) RD and WR propagation delays are approximately 50 ns. The other signals are typically 85 ns. Propagation delays are incorporated in the AC specifications.

18. Ordering Information

Table 18-1. Ordering Information

19. Package Information

19.1 PDIL40

40 PINS PLASTIC .600

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19.2 PLCC44

STANDARD NOTES FOR PLCC

1/ CONTROLLING DIMENSIONS : INCHES

2/ DIMENSIONING AND TOLERANCING PER ANSI Y 14.5M - 1982.

3/ "D" AND "E1" DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTUSIONS. MOLD FLASH OR PROTUSIONS SHALL NOT EXCEED 0.20 mm (.008 INCH) PER SIDE.

19.3 VQFP44

STANDARD NOTES FOR PQFP/ VQFP / TQFP / DQFP

1/ CONTROLLING DIMENSIONS : INCHES

2/ ALL DIMENSIONING AND TOLERANCING CONFORM TO ANSI Y 14.5M - 1982.

3/ "D1 AND E1" DIMENSIONS DO NOT INCLUDE MOLD PROTUSIONS. MOLD PROTUSIONS SHALL NOT EXCEED 0.25 mm (0.010 INCH). THE TOP PACKAGE BODY SIZE MAY BE SMALLER THAN THE BOTTOM PACKAGE BODY SIZE BY AS MUCH AS 0.15 mm.

4/ DATUM PLANE "H" LOCATED AT MOLD PARTING LINE AND COINCIDENT WITH LEAD, WHERE LEAD EXITS PLASTIC BODY AT BOTTOM OF PARTING LINE.

5/ DATUM "A" AND "D" TO BE DETERMINED AT DATUM PLANE H.

6/ DIMENSION " f " DOES NOT INCLUDE DAMBAR PROTUSION ALLOWABLE DAMBAR PROTUSION SHALL BE 0.08mm/.003" TOTAL IN EXCESS OF THE " f " DIMENSION AT MAXIMUM MATERIAL CONDITION . DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OR THE FOOT.

20. Datasheet Change Log

20.1 Changes from 4113A - 09/02 to 4113B -03/05

1. Added Green product ordering information.

20.2 Changes from 4113B -03/05 to 4113C -01/08

- 1. Removed AT80C51RD2 product offering Table 18-1 on page 63.
- 2. Updated Package Drawings.

20.3 Changes from 4113C -01/08 to 4113D -01/09

1. Removed AT83C51RD2 product offering

