## **Features**

- **DC to DC Converter 1.9V / 2.5V (DCDC1)**
- **LDO Regulator 2.7V / 2.8V (LDO1)**
- **LDO Regulator 2.8V (LDO2)**
- **LDO Regulator 2.8V (LDO3)**
- **LDO Regulator 2.47V / 2.66 (LDO4) Backup Battery Supply**
- **LDO Regulator 1.72V / 2.66 (LDO5) RTC Supply**
- **Reset Generator**

## **1. Description**

The AT73C211 is a power management device for digital, analog, interface, and, in some cases, RF and backup sections of add-on modules used as accessories in popular handheld devices like mobile phones, digital still cameras, PDAs and a wide range of multimedia devices. The AT73C211 can also be used to supply the CPU with a high-efficiency DC-DC Converter, a radio frequency transceiver with high power supply rejection ratio (PSRR) and noise performance low-dropout (LDO) regulators, or memories and analog sections with independent LDO channels.

In addition, the AT73C211 integrates LDO regulators to recharge backup elements and convert its voltage to microcontroller RTC supply.

LDO regulators and DC-DC converters output voltage can be programmed by a mask change.



# **Power Management**

# **AT73C211**





## **2. Functional Block Diagram**





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# **3. Pin Description**

**Table 3-1.** Pin Description

<b>Signal</b>	Pin	<b>Type</b>	A/D	<b>Description</b>
<b>VBATT</b>	E <sub>1</sub>	VBATT1		Input supply
ON/OFF	D <sub>5</sub>	<b>IPD</b>	D	Key ON/OFF input, 1.5M Ohm pull-down
UP-ON/OFF	C <sub>6</sub>		D	Hold the Power ON from MCU
<b>RESET-B</b>	F <sub>6</sub>	OD	D	Reset open collector output. Need external pull-up to VBATT
VIN-REG1	G <sub>6</sub>	VBATT2		Input supply for DC/DC converter
<b>LX</b>	F7	O	A	DC/DC converter output inductor
ECO-MODE	G <sub>5</sub>	<b>IPD</b>	D	Eco Mode, from MCU - sets VCORE, V-PAD in low power mode, 1.5M Ohm pull-down
<b>VCORE</b>	G <sub>4</sub>	$\circ$	A	DC/DC converter output (MCU core supply)
GND1	G7	Ground		Ground of DC/DC converter
VIN-REG2	A5	VBATT3		Input supply
EN-ANALOG-B	B <sub>5</sub>	<b>IPD</b>	D	Enable the analog LDO, active at logic 0, 1.5M Ohm pull-down
<b>AVCC</b>	<b>B4</b>	$\Omega$	A	Analog LDO output (MCU chip analog supply)
<b>AGND</b>	Α7	Ground		Ground of AVCC, V-PAD and RTC LDO
V-PAD	B <sub>6</sub>	$\circ$	A	Digital LDO output (MCU chip digital PAD supply)
VCC-RTC	<b>B7</b>	$\circ$	A	MCU RTC supply output
<b>BAT-RTC</b>	A <sub>6</sub>	I/O	A	RTC backup battery charger - must be connected through a 2.2K Ohm resistor to the backup battery
<b>VIN-RF</b>	A <sub>3</sub>	VBATT4		Input supply
AGND <sub>2</sub>	A <sub>2</sub>	Ground		Ground
<b>VIN-VIB</b>	D7	VBATT <sub>5</sub>		Input supply for vibrator LDO
EN-VIB	E <sub>6</sub>	<b>IPD</b>	D	Vibrator driver input (from baseband chip), 1.5M Ohm pull-down
<b>VVIB</b>	E7	$\circ$	A	Vibrator LDO output (Voltage regulator)
<b>GND</b>	D1	Ground		Ground
<b>CREF</b>	C7	$\circ$	A	Bandgap decoupling - 100 nF capacitor must be connected from this pin to ground
BB1	D <sub>4</sub>	$\mathbf{I}$	D	$BB1 = 1 \Rightarrow VCORE = 2.5V$ , $BB1 = 0 \Rightarrow VCORE = 1.9V$
<b>TEST</b>	E <sub>5</sub>	<b>IPD</b>	A	Connect to AGND





## **4. Functional Description**

## **4.1 DC to DC Converter 1.9V/2.5V - 300 mA for Coprocessor Core**

The DC-to-DC converter is a synchronous mode DC-to-DC "buck"-switched regulator using fixed-frequency architecture (PWM) and capable of providing 300 mA of continuous current. It has two levels of voltage programming for the co-processor core (1.9V or 2.5V). The operating supply range is from 3.1V to 5.5V, making it suitable for Li-Ion, Li-polymer or Ni-MH battery applications. The DC-to-DC converter is based on pulse width modulation architecture to control the noise perturbation for switching noise sensitive applications (Wireless). The operating frequency is set to 900 kHz using an internal clock, allowing the use of a small surface inductor and moderate output voltage ripple. The controller consists of a reference ramp generator, a feedback comparator, the logic driver used to drive the internal switches, the feedback circuits used to manage the different modes of operation and the over-current protection circuits. An economic mode has been defined to reduce quiescent current. A low-dropout voltage regulator in parallel to the DC-to-DC converter minimizes standby current consumption during standby mode.





Low undershoot voltage is expected when going from PWM to LDO mode and vice-versa. The circuit is designed in order to avoid any spikes when transition between two modes is enabled.

**Figure 4-2.** Low-power/Full-power DC-to-DC Converter Transition



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[Figure 4-3](#page-4-0) shows typical efficiency levels of the DC-to-DC converter for several input voltages.

<span id="page-4-0"></span>



Note:  $1. L = 10 \mu H$ ,  $ESR = 0.2 \text{ Ohm}$ ,  $c = 22 \mu F$ ,  $@ESR = 0.1 \text{ Ohm}$ 

## **4.2 LDO1, LDO3 Regulators**

The PSRR measures the degree of immunity against voltage fluctuations achieved by a regulator. An example of its importance is in the case of a GSM phone when the antenna switch activates the RF power amplifier (PA). This causes a current peak of up to 2A on the battery, with an important spike on the battery voltage. The voltage regulator must filter or attenuate this spike.







**Figure 4-4.** Functional Diagram of LDO Single Mode

[Figure 4-5](#page-5-0) shows the Power Supply Rejection Ratio as functions of frequency and battery voltage. If a noise signal occurs at 1 kHz when the battery voltage is at 3V, the noise will be attenuated by 70 dB (divided by more than 3000) at the output of the regulator. Consequently, a 2V spike on the battery is attenuated to less than 1 mV, which is low enough to avoid any risk of malfunction by a device supplied by the regulator.

<span id="page-5-0"></span>



**AT73C211** 

## **4.3 LDO2 Regulator**

The first approach to reducing standby current is to decrease the standby current inside the regulators themselves. Atmel achieves this by implementing a dual mode architecture where two output transistors are used in parallel as switches in the regulation loop. [Figure 4-6](#page-6-0) illustrates this architecture.

<span id="page-6-0"></span>



In [Figure 4-6,](#page-6-0) the left-hand output transistor is sized large enough for the required output current under full load, for example, 100 mA. In order to achieve a sufficient margin of stability, the current sensing block uses a bias cell where the current consumption is linked to the required output current. The higher the output current, the higher the bias current needed to stabilize the loop.

The right-hand output transistor delivers a very small output current, typically less than 1 mA, sufficient only to maintain the output voltage with enough current to cover the leakage current of the supplied device. This requires a much smaller bias current and, consequently, a smaller standby current inside the regulator.



## **5. Electrical Characteristics**

## **5.1 Absolute Maximum Ratings**



## **5.2 DC to DC Converter**

**Table 5-1.** DC to DC Converter Electrical Characteristics  $(t_{AMB} = -20^\circ C$  to 85° C, VIN = 3.2V to 4.2V unless otherwise specified)

Symbol	<b>Parameter</b>	<b>Conditions</b>	Min	<b>Typ</b>	Max	Unit
$V_{OUT}$		$BB1 = 0$		1.9		$\vee$
	Output Voltage	$BB1 = 1$		2.5		$\vee$
$I_{OUT}$		PWM Mode (ECO-MODE = $0$ )		150	300	mA
	<b>Output Current</b>	LDO Mode (ECO-MODE = $1$ )			5	mA
$I_{OFF}$	<b>Standby Current</b>			0.1	1	μA
$E_{FF}$	Efficiency	$I_{OUT}$ = 10 mA to 200 mA @ 1.9V		90		$\%$
$\Delta V_{DCLD}$	<b>Static Load Regulation</b>	10% to 90% of $I_{OUT(MAX)}$		$\overline{7}$		mV
$\Delta V$ <sub>TRLD</sub>	<b>Transient Load Regulation</b>	10% to 90% of $I_{\text{OUT}(MAX)}$ , $T_B = T_F = 5 \mu s$		30		mV
$\Delta\rm{V}_{DCLE}$	<b>Static Line Regulation</b>	10% to 90% of $I_{OUT(MAX)}$ , $VIN = 3.2V$ to 4.2V		20		mV
$\Delta V$ TRLE	<b>Transient Line Regulation</b>	10% to 90% of $I_{OUT(MAX)}$ , $VIN = 3.2V$ to 4.2V		35		mV
<b>PSRR</b>	<b>Ripple Rejection</b>	LDO Mode up to 1 KHz	40	45		dB
$\Delta V_{\text{LPFP}}$	Overshoot Voltage	Voltage drop from LDO (ECO- $MODE = 1$ ) to PWM (ECO- $MODE = 0$		$\Omega$	10	mV
$\Delta V_{FPLP}$	Undershoot Voltage	Voltage drop from PWM (ECO- $MODE = 0$ ) to LDO (ECO-MODE $= 1$	$-15$	0		mV





## **5.3 LDO1 Regulator Electrical Characteristics**



## **Table 5-3.** LDO1 Electrical Characteristics  $(t_{AMB} = -20^{\circ}C$  to 85 $^{\circ}C$ , VIN = 3.2V to 4.2V unless otherwise specified)

#### **Table 5-4.** LDO1 External Components







## **5.4 LDO2 Regulator Electrical Characteristics**

<b>TAND</b>								
Symbol	<b>Parameter</b>	<b>Conditions</b>	Min	Typ	Max	<b>Unit</b>		
$V_{OUT}$	Output Voltage			2.8		V		
$I_{\text{OUT}}$		$PWM$ Mode (ECO-MODE = 0)			80	mA		
	<b>Output Current</b>	LDO Mode (ECO-MODE = $1$ )			5	mA		
$I_{\rm QC}$		$PWM$ Mode (ECO-MODE = 0)		100		μA		
	<b>Quiescent Current</b>	LDO Mode (ECO-MODE = $1$ )			10	μA		
$\Delta V_{OUT}$	Line Regulation	$V_{IN}$ : 3V to 3.4V, $I_{OUT}$ = 80 mA		1	$\overline{2}$	mV		
$\Delta V_{PEAK}$	Line Regulation Transient	Same as above, $T_B = T_F = 5 \mu s$		1.5	2.85	mV		
$\Delta V_{\text{OUT}}$	Load Regulation	10% - 90% $I_{OUT}$ VIN = 3V			3	mV		
$\Delta V_{PEAK}$	<b>Load Regulation Transient</b>	Same as above, $T_R = T_F = 5 \,\mu s$		1.2	2.4	mV		
<b>PSRR</b>	Ripple rejection	$F = 217$ Hz; VIN = 3.6V	70	73		dB		
$V_N$	<b>Output Noise</b>	BW: 10 Hz to 100 kHz		29	37	$\mu V_{RMS}$		
$T_R$	<b>Rise Time</b>	100% $I_{\text{OUT}}$ , 10% - 90% $V_{\text{OUT}}$			50	μs		
l <sub>SD</sub>	<b>Shut Down Current</b>					μA		

**Table 5-5.** LDO2 Electrical Characteristics  $(t_{AMB} = -20^{\circ}C$  to 85 $^{\circ}C$ , VIN = 3.2V to 4.2V unless otherwise specified)

#### **Table 5-6.** LDO2 External Components



## **5.5 LDO3 Regulator Electrical Characteristics**



#### **Table 5-7.** LDO3 Electrical Characteristics  $(t_{AMB} = -20^{\circ}C$  to 85°C, VIN = 3.2V to 4.2V unless otherwise specified)

#### **Table 5-8.** LDO3 External Components







## **5.6 LDO4 Regulator Electrical Characteristics**



#### **Table 5-9.** LDO4 Electrical Characteristics  $(t_{AMB} = -20^{\circ}C$  to 85 $^{\circ}C$ , VIN = 3.2V to 4.2V unless otherwise specified)

#### **Table 5-10.** LDO4 External Components



## **5.7 LDO5 Regulator Electrical Characteristics**



## **Table 5-11.** LDO5 Electrical Characteristics  $(t_{AMB} = -20^{\circ}C$  to 85 $^{\circ}C$ , VIN = 3.2V to 4.2V unless otherwise specified)

#### **Table 5-12.** LDO4 External Components







## **5.8 Package Outline (Top view)**



**Figure 5-1.** Forty-nine Ball FBGA Package (Top View)

## **6. Revision History**

**Table 6-1.** Revision History







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