



PAM8302L

2.5W FILTERLESS CLASS-D MONO AUDIO AMPLIFIER

Description

The PAM8302L is a 2.5W Class-D mono audio amplifier. Its low THD+N offers high-quality sound reproduction.

The PAM8302L uses a filterless design that avoids the use of lowpass filters. This new design allows the amplifier to directly drive a speaker, making it cheap and compact. The new design allows the amplifier to be more affordable and take less PCB area.

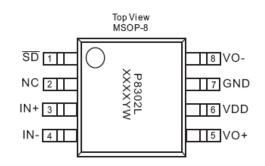
The PAM8302L uses less power than the Class-AB amplifiers. The use of this product can help optimize battery life; it is ideal for portable applications.

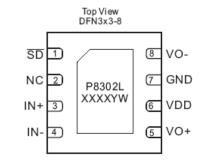
The PAM8302L is available in MSOP-8 and DFN3x3-8 packages.

Features

- Support 2.0V to 5.5V Supply Voltage Range
- 2.5W Output at 10% THD with a 4 Ω Load and 5V Power Supply
- Filterless, Low Quiescent Current and Low EMI
- High Efficiency up to 88%
- Superior Low Noise
- Short Circuit Protection
- Thermal Shutdown
- Few External Components to Save Space and Cost
- MSOP-8 and DFN3x3 Packages Available
- Pb-Free Packages

Pin Assignments

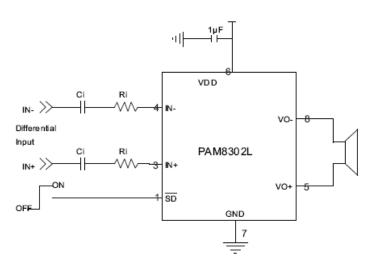




Applications

- PMP/MP4
- GPS
- Portable Speakers
- Walkie Talkie
- Handsfree phones/Speaker Phones
- Cellular Phones

Typical Applications Circuit



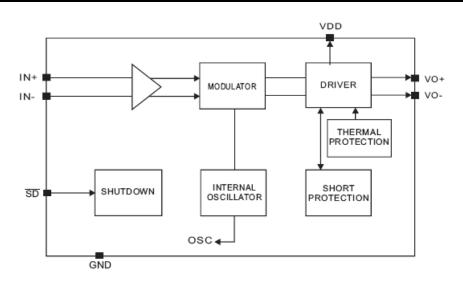




Pin Descriptions

Pin Number	Pin Name	Function
1	SD	Shutdown Terminal (active low)
2	NC	No Connection
3	IN+	Positive Differential Input
4	IN-	Negative Differential Input
5	VO+	Positive BTL Output
6	VDD	Analog Power Supply
7	GND	Ground
8	VO-	Negative BTL Output

Functional Block Diagram



Absolute Maximum Ratings (@T_A = +25°C, unless otherwise specified.)

These are stress ratings only and functional operation is not implied. Exposure to absolute maximum ratings for prolonged time periods may affect device reliability. All voltages are with respect to ground.

Parameter	Rating	Unit	
Supply Voltage at No Input Signal	6.0	N	
Input Voltage	-0.3 to V _{DD} +0.3		
Maximum Junction Temperature	150		
Storage Temperature	-65 to +150	°C	
Soldering Temperature	300, 5sec		

Recommended Operating Conditions (@T_A = +25°C, unless otherwise specified.)

Parameter	Rating	Unit
Suppy Voltage Range	2.0 to 5.5	V
Operation Temperature Range	-40 to +85	°C
Junction Temperature Range	-40 to +125	°C





Thermal Information

Parameter	Package	Symbol	Max	Unit	
Thermal Resistance (Junction to Case)	MSOP-8	0	75	°C/W	
mermar Resistance (Junction to Case)	DFN3x3-8	θ _{JC}	20		
Thermal Resistance (Junction to Ambient)	MSOP-8	0	180	°C/W	
Thermal Resistance (Sunction to Ambient)	DFN3x3-8	θ_{JA}	50	C/VV	
	MSOP-8	D	550	mW	
Internal Power Dissipation @ $T_A = +25^{\circ}C$	DFN3x3-8	PD	2000	mvv	

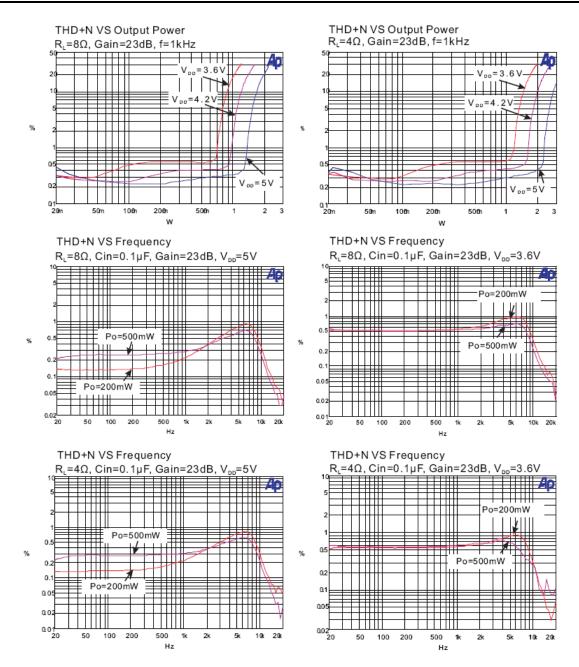
Electrical Characteristics (@T_A = +25°C, V_{IN} = 3.6V, V_O = 1.8V, C_{IN} = 10µF, C_{OUT} = 10µF, L = 4.7µH, unless otherwise specified.)

Parameter	Symbol	Test Conditions		Min	Тур	Max	Units
Supply Voltage Range	V _{DD}			2.5		5.5	V
Quiescent Current	lq	No Load			4	8	mA
Shutdown Current	I _{SHDN}	V _{SHDN} = 0V				1	μA
		f = 1kHz, R _L = 4Ω, THD+N = 10%	V _{DD} = 5V	2.25	2.50		-
			V _{DD} = 3.6V	1.10	1.25		
		$f = 1 kHz$, $R_1 = 4 \Omega$,	V _{DD} = 5V	1.80	2.00		1
	_	THD+N = 1%	V _{DD} = 3.6V	0.86	0.95		
Output Power	Po	$f = 1 kHz$, $R_1 = 8\Omega$,	V _{DD} = 5V	1.35	1.50		W
		THD+N = 10%	V _{DD} = 3.6V	0.72	0.80		
		f = 1kHz, R ₁ = 8Ω,	V _{DD} = 5V	1.15	1.30		-
		THD+N = 1%	V _{DD} = 3.6V	0.6	0.65		
Peak Efficiency	η	f = 1kHz			85	88	%
	·	$R_{L} = 8\Omega, P_{O} = 0.1W, f = 1kHz$			0.30	0.35	%
	THD+N	$R_{L} = 8\Omega, P_{O} = 0.5W, f = 1kHz$			0.45	0.50	
Total Harmonic Distortion Plus Noise		$R_{L} = 4\Omega, P_{O} = 0.1W, f = 1kHz$			0.35	0.40	
		$R_{\rm L} = 4\Omega, P_{\rm O} = 0.5W, f = 1 \text{kHz}$			0.40	0.45	
Gain	Gv			22.5	24.0	25.5	dB
Power Supply Ripple Rejection	PSRR	No Inputs, f = 1kHz, V	No Inputs, f = 1kHz, V _{PP} = 200mV		50		dB
Dynamic Range	DYN	f = 20 to 20kHz		85	90		dB
Signal to Noise Ratio	SNR	f = 20 to 20kHz			80		dB
Naisa		No A-Weighting			180	300	μV
Noise	V _N	A-Weighting	A-Weighting		120	200	
Oscillator Frequency	f _{OSC}			200	250	300	kHz
Drain-Source On-State Resistance		I _{DS} = 100mA	P MOSFET		0.45	0.50	Ω
	R _{DS(ON)}		N MOSFET		0.20	0.25	
SHDN Input High	V _{SH}			1.2			v
SHDN Input Low	V _{SL}					0.4	v
Over Temperature Protection	OTP	Junction Temperautre		120	135		°C
Over Temperature Hysterisis	OTH				30		°C





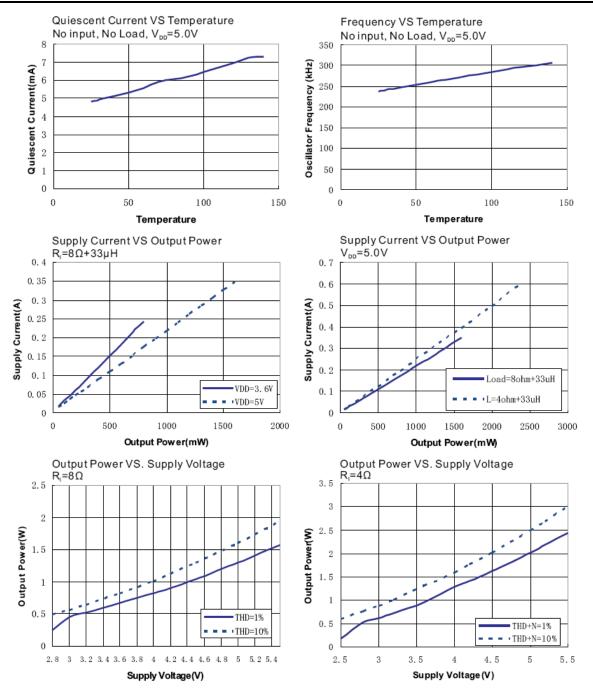
Typical Performance Characteristics (@TA = +25°C, unless otherwise specified.)







Typical Performance Characteristics (cont.) (@TA = +25°C, unless otherwise specified.)







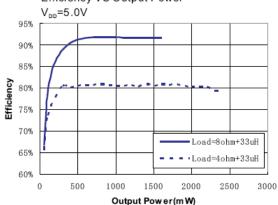
Vin=5V

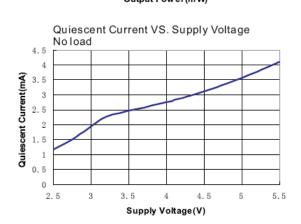
2000

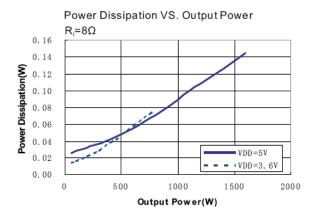
• • Vin=3.6V

1500

Typical Performance Characteristics (cont.) (@T_A = +25°C, unless otherwise specified.) Efficiency VS Output Power V_{DD}=5.0V 95% Efficiency VS Output Power R_i=8Ω+33 µH







Frequency VS. Supply Voltage

1000

Output Power(mW)

500

90%

85%

80%

75%

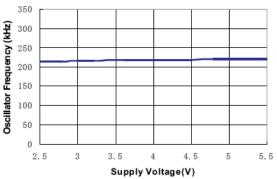
70%

65%

60%

0

Efficiency

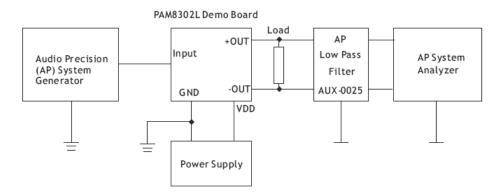


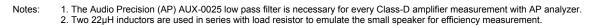




Application Information

Test Setup for Performance Testing





Maximum Gain

As shown in block diagram (Page 2), the PAM8302L has two internal amplifier stages. The first stage's gain is externally con figurable, while the second stage's is internally fixed. The closedloop gain of the first stage is set by selecting the ratio of R_F to R_I while the second stage's gain is fixed at 2x. The output of amplifier one serves as the input to amplifier two, thus the two amplifiers produce signals identical in magnitude, but different in phase by 180°. Consequently, the differential gain for the IC is

A =20*log $[2*(R_F/R_I)]$

The PAM8302L sets maximum R =80k Ω , minimum R_I =10k Ω , so the maximum closed-gain is 24dB.

Input Capacitor (C_I)

Intypical application, an input capacitor, C_I is required to allow the amplifier to bias input signals to a proper DC level for optimum operation. In this case, C_I and the minimum input impedance R_I (10k internal) form a high pass filter with a corner frequeny determind by the following equation:

$$f_{C} = \frac{1}{2\Pi R_{I} C_{I}}$$

It is important to choose the value of C₁ as it directly affects low frequency performance of the circuit, for example, when an application requires a flat bass response as loas as 100Hz. Equation is reconfigured as follows:

$$C_{I} = \frac{1}{2\Pi R_{I} f_{I}}$$

As the input reisitance is varible, for the C_I value of 0.16μ F, one should actually choose the C_I within the range of 0.1μ F to 0.22μ F. A further consideration for this capacitor is the leakage path from the input source through the input network (R_I , R_F , C_I) to the load. This leakage current creates a DC offset voltage at the input to the amplifier that reduces useful headroom, especially in high gain application. For this reason, a low leakage tantalum or ceramic capacitor is the best choice. When a polarized capacitor is used, the positive side of the capacitor should face the amplifier input in most applications as the DC level is held at $V_{DD}/2$, which is likely higher than the source DC level. Please note that it is important to confirm the capacitor polarity in the application.





Application Information (cont.)

Power Supply Decoupling (C_s)

The PAM8302L is a high-performance CMOS audio amplifier that requires adequate power supply decoupling to ensure the output THD and PSRR as low as possible. Power supply decoupling affects low frequency response. Optimum decoupling is achieved by using two capacitors of different types that target different types of noise on the power supply leads. For higher frequency transients, spikes, or digital hash on the line, a good low equivalent-series-resistance (ESR) ceramic capacitor, typicall 1.0µF is good, placing it as close as possible to the device V_{DD} terminal. For filtering lower frequency noise signals, capacitor of 10µF or larger, closely located to near the audio power amplifier is recommended.

Shutdown Operation

In order to reduce shutdown power consumption, the PAM8302L contains shutdown circuitry for turn to turn off the amplifier. This shutdown feature turns the amplifier off when a logic low is apllied on the SD pin. By switching the shutdown pin over to GND, the PAM8302L supply current draw will be minimized inidle mode.

For the best power on/off pop performance, the amplifier should be set in the shutdown mode prior to power on/off operation.

Under Voltage Lock-Out (UVLO)

The PAM8302L incorporates circuitry to detect low on or off voltage. When the supply voltage drops to 2.1V or below, the PAM8302L goes into a state of shutdown, and the device comes out of its shutdown state to normal operation by reset the power supply or SD pin.

How to Reduce EMI (Electro Magnetic Interference)

A simple solution is to put an additional capacitor 1000µF at power supply terminal for power line coupling if the traces from amplifier to speakers are short (< 20CM). Most applications require a ferrite bead filter as shown at Figure 1. The ferrite filter depresses EMI of around 1MHz and higher. When selecting a ferrite bead, choose one with high impedance at high frequencies and low impedance at low frequencies.

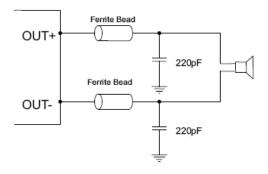
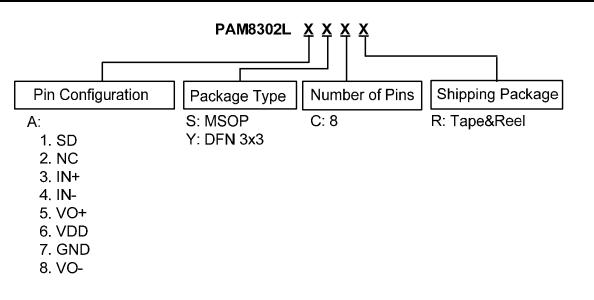


Figure 1. Ferrite Bead Filter to Reduce EMI



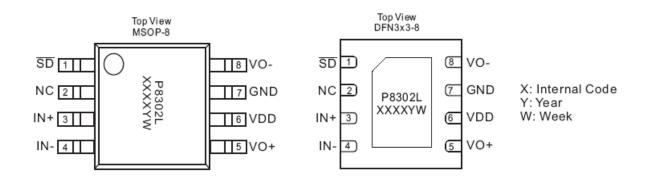


Ordering Information



Part Number	Package Type	Standard Package
PAM8302LASCR	MSOP-8	2500 Units/Tape&Reel
PAM8302LAYCR	DFN3x3-8	3000 Units/Tape&Reel

Marking Information

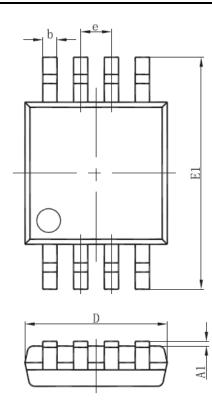


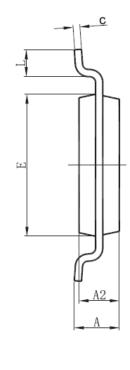




Package Outline Dimensions (All dimensions in mm.)

MSOP-8





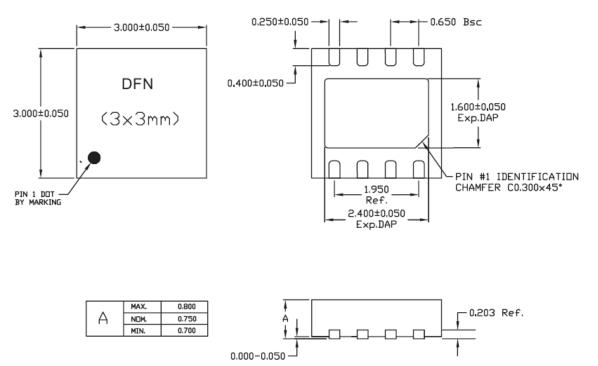
REF	Millimeter		
REF	Min	Max	
A		1.10	
A1	0.05	0.15	
A2	0.78	0.94	
b	0.22	0.38	
с	0.08	0.23	
D	2.90	3.10	
E	2.90	3.10	
E1	4.75	5.05	
е	0.65BSC		
L	0.40	0.70	

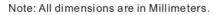




Package Outline Dimensions (cont.) (All dimensions in mm.)

DFN3x3-8









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