16-BIT, 2-MSPS, LVDS SERIAL INTERFACE, SAR ANALOG-TO-DIGITAL CONVERTER

Check for Samples: ADS8410

FEATURES

- **2-MHz Sample Rate**
- **16-Bit Resolution**
- **SNR 87.5 dB at 10 kHz I/P**
- **THD –98 dB at 10 kHz I/P**
- **±1 LSB Typ, ±2.5 LSB INL Max**
- **+0.8/–0.5 LSB Typ, +1.5/–1 LSB DNL Max**
- **Unipolar Differential Input Range: 0 V to 4 V**
- **Internal Reference**
- **Internal Reference Buffer**
- **200-Mbps LVDS Serial Interface**
- **Optional 200-MHz Internal Interface Clock**
- **16-/8-Bit Data Frame**
- **Zero Latency at Full Speed**
- **Power Dissipation: 290 mW at 2 MSPS**
- **Nap Mode (125 mW Power Dissipation)**
- **Power Down (5 µW)**
- **48-Pin QFN Package**

APPLICATIONS

- **Medical Instrumentation**
- **HIgh-Speed Data Acquisiton Systems**
- **High-Speed Close-Loop Systems**
- **Communication**

DESCRIPTION

The ADS8410 is a 16-bit, 2-MSPS, analog-to-digital (A/D) converter with 4-V internal reference. The device includes a capacitor based SAR A/D converter with inherent sample and hold.

The ADS8410 also includes a 200-Mbps, LVDS, serial interface. This interface is designed to support daisy chaining or cascading of multiple devices. A selectable 16-/8-bit data frame mode enables the use of a single shift register chip (SN65LVDS152) for converting the data to parallel format.

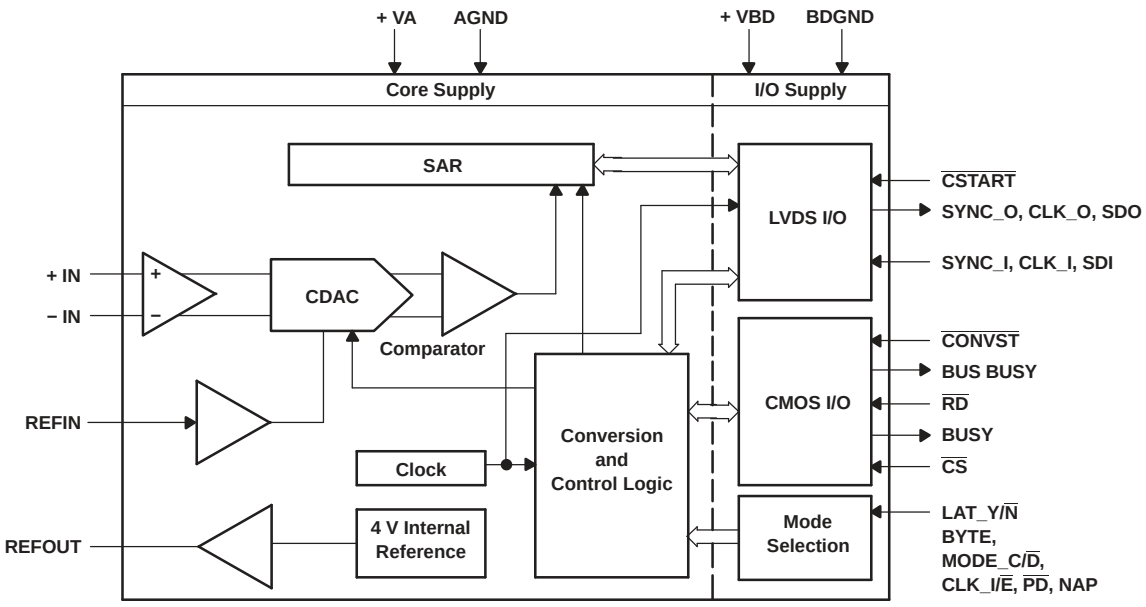
The ADS8410 unipolar single-ended input range supports a differential input swing of 0 V to +V_{ref}.

The nap feature provides substantial power saving when used at lower conversion rates.

The ADS8410 is available in a 48-pin QFN package.

Table 1. High-Speed SAR Converter Family

Type/Speed	500 kHz	~ 600 kHz	750 kHZ	1 MHz	1.25 MHz	2 MHz	3 MHz	4 MHz
18-Bit Pseudo-Diff	ADS8383	ADS8381						
		ADS8380 (S)						
18-Bit Pseudo-Bipolar, Fully Diff		ADS8382 (S)						
16-Bit Pseudo-Diff		ADS8370 (S)	ADS8371		ADS8401/05	ADS8411		
						ADS8410 (S-LVDS)		
16-Bit Pseudo-Bipolar, Fully Diff		ADS8372 (S)			ADS8402/06	ADS8412		
						ADS8413 (S-LVDS)		
14-Bit Pseudo-Diff					ADS7890 (S)		ADS7891	
12-Bit Pseudo-Diff								ADS7881



ORDERING INFORMATION(1)

MODEL	MAXIMUM INTEGRAL LINEARITY (LSB)	MAXIMUM DIFFERENTIAL LINEARITY (LSB)	NO MISSING CODES AT RESOLUTION (BIT)	PACKAGE TYPE	PACKAGE DESIGNATOR	TEMPERATURE RANGE	ORDERING INFORMATION	TRANSPORT MEDIA QUANTITY
ADS8410IB	±2.5	1.5/–1	16	48 pin QFN	RGZ	–40°C to 85°C	ADS8410IBRGZT	250
							ADS8410IBRGZR	2000
ADS8410I	±4	3/–1	16	48 pin QFN	RGZ	–40°C to 85°C	ADS8410IRGZT	250
							ADS8410IRGZR	2000

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)(1)

		UNIT
+IN to AGND		–0.3 V to +VA + 0.3 V
-IN to AGND		–0.3 V to +0.3 V
+VA to AGND		–0.3 to 7 V
+VBD to BDGND		–0.3 to 7 V
Digital input voltage to GND		–0.3 V to (+VBD + 0.3 V)
Digital output to GND		–0.3 V to (+VBD + 0.3 V)
Operating temperature range		–40°C to 85°C
Storage temperature range		–65°C to 150°C
Junction temperature ($T_J$max)		150°C
QFN package	Power dissipation	(T_J Max – T_A)/ θ_{JA}
	θ_{JA} Thermal impedance	86°C/W
Lead temperature, soldering	Vapor phase (60 sec)	215°C
	Infrared (15 sec)	220°C

(1) Stresses beyond those listed under *absolute maximum ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *recommended operating conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

SPECIFICATIONS

T_A = –40°C to 85°C, +VA = 5 V,+VBD = 5 V or 3.3 V, V_{ref} = 4.096 V, f_{sample} = 2 MHz (unless otherwise noted)

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
ANALOG INPUT							
	Full-scale input voltage span (1)		+IN – (–IN)	0		V_{ref}	V
	Absolute input voltage range		+IN	–0.2		V_{ref} + 0.2	V
			–IN	–0.2		+0.2	
C_i	Input capacitance				25		pF
	Input leakage current				500		pA
SYSTEM PERFORMANCE							
	Resolution				16		Bits
	No missing codes	ADS8410IB		16			Bits
		ADS8410I		16			
INL	Integral linearity(2)	ADS8410IB		–2.5	±1	2.5	LSB(3)
		ADS8410I		–4.0	±2.5	4.0	
DNL	Differential linearity	ADS8410IB		–1	0.8/–0.5	1.5	LSB(3)
		ADS8410I		–1.0	1.5/–0.8	3	
E_O	Offset error	ADS8410IB	External reference	–0.75	±0.1	0.75	mV
		ADS8410I		–1.5	±0.75	1.5	
E_G	Gain error(4)	ADS8410IB	External reference	–0.05	±0.01	0.05	% of FS
		ADS8410I		–0.15	±0.05	0.15	
CMMR	Common-mode rejection ratio		With common mode input signal = 200 mV_{p-p} at 1 MHz		60		dB
PSRR	Power supply rejection ratio		At $FFF0_H$ output code		80		dB
SAMPLING DYNAMICS							
	Conversion time		+VBD = 5 V		360	391	ns
			+VBD = 3 V			391	
	Acquisition time		+VBD = 5 V	100			ns
			+VBD = 3 V	100			
	Maximum throughput rate with or without latency					2.0	MHz
	Aperture delay				20		ns
	Aperture jitter				10		psec
	Step response				50		ns
	Overvoltage recovery				50		ns
DYNAMIC CHARACTERISTICS							
THD	Total harmonic distortion(5)		V_{IN} 0.5 dB below FS at 10 kHz		–98		dB
			V_{IN} 0.5 dB below FS at 100 kHz		–92.5		
SNR	Signal-to-noise ratio		V_{IN} 0.5 dB below FS at 10 kHz		87.5		dB
			V_{IN} 0.5 dB below FS at 100 kHz		86		
SINAD	Signal-to-noise and distortion		V_{IN} 0.5 dB below FS at 10 kHz		87		dB
			V_{IN} 0.5 dB below FS at 100 kHz		85		
SFDR	Spurious free dynamic range		V_{IN} 0.5 dB below FS at 10 kHz		–101		dB
			V_{IN} 0.5 dB below FS at 100 kHz		–93		
	–3 dB Small signal bandwidth				37.5		MHz
EXTERNAL REFERENCE INPUT							
	Input voltage range, V_{REF}			3.9	4.096	4.2	V
	Resistance(6)		To internal reference voltage		500		kΩ
INTERNAL REFERENCE OUTPUT							

(1) Ideal input span; does not include gain or offset error.
(2) This is endpoint INL, not best fit.
(3) Least significant bit
(4) Measured relative to actual measured reference.
(5) Calculated on the first nine harmonics of the input frequency.
(6) Can vary ±20%

SPECIFICATIONS (continued)

T_A = –40°C to 85°C, +VA = 5 V,+VBD = 5 V or 3.3 V, V_{ref} = 4.096 V, f_{sample} = 2 MHz (unless otherwise noted)

PARAMETER			TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Start-up time		From 95% (+VA), with 1-µF storage capacitor on REFOUT to AGND			25	ms
	Reference voltage range, V_{ref}		At room temperature	4.080	4.096	4.112	V
	Source current		Static load			10	µA
	Line regulation		+VA = 4.75 V to 5.25 V		0.6		mV
	Drift		IOUT = 0 V		36		PPM/°C
POWER SUPPLY REQUIREMENTS							
	Power supply voltage	+VBD		2.7	3.3	5.25	V
		+VA		4.75	5	5.25	
	Supply current, 2-MHz sample rate	+VA			58	64	mA
	Power dissipation, 2-MHz sample rate		+VA = 5 V		290	320	mW
NAP MODE							
	Supply current	+VA			25		mA
POWER DOWN							
	Supply current	+VA			1	2.5	µA
	Powerdown time				10		µs
	Powerup time		With 1-µF storage capacitor on REFOUT to AGND		25		ms
	Invalid conversions after power up or reset				3		Numbers
TEMPERATURE RANGE							
	Operating free air			–40		85	°C
LOGIC FAMILY CMOS							
V_{IH}	High-level input voltage		I_{IH} = 5 µA	+VBD –1		+VBD +0.3	V
V_{IL}	Low-level input voltage		I_{IL} = 5 µA	–0.3		0.8	V
V_{OH}	High-level output voltage		I_{OH} = 2 TTL loads	+VBD – 0.6		+VBD	V
V_{OL}	Low-level output voltage		I_{OL} = 2 TTL loads	0		0.4	V
LOGIC FAMILY LVDS(7)							
DRIVER							
$\|V_{OD(SS)}\|$	Steady-state differential output voltage magnitude		R_L = 100 Ω, See Figure 52, Figure 53	247	340	454	mV
$\Delta\|V_{OD(SS)}\|$	Change in steady-state differential output voltage magnitude between logic states			-50		50	
$V_{OC(SS)}$	Steady-state common-mode output voltage		See Figure 54	1.125	1.2	1.375	V
$\Delta\|V_{OC(SS)}\|$	Change in steady-state common-mode output voltage between logic states			–50		50	mV
$V_{OC(pp)}$	Peak to peak change in common-mode output voltage				50	150	
I_{OS}	Short circuit output current		V_{OY} or V_{OZ} = 0 V		3	10	mA
			V_{OD} = 0 V		3	10	
I_{OZ}	High impedance output current		VO = 0 V or +VBD	–5		5	µA

(7) All min max values ensured by design.

SPECIFICATIONS (continued)

T_A = –40°C to 85°C, +VA = 5 V,+VBD = 5 V or 3.3 V, V_{ref} = 4.096 V, f_{sample} = 2 MHz (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
RECEIVER						
V_{ITH+}	Positive going differential voltage threshold				50	mV
V_{ITH-}	Negative going differential voltage threshold		–50			
V_{IC}	Common mode input voltage		0.2	1.2	2.2	V
C_I	Input capacitance			5		pF

TIMING REQUIREMENTS

T_A = –40°C to 85°C, +VA = 5 V, +VBD = 5 V or 3.3 V (unless otherwise noted)

PARAMETER			MIN	TYP	MAX	UNIT	REF
SAMPLING AND CONVERSION RELATED							
t_{acq}	Acquisition time		100			ns	Figure 1, Figure 2
t_{cnv}	Conversion time				391	ns	Figure 1, Figure 2
t_{w1}	Pulse duration, $\overline{\text{CONVST}}$ high		100			ns	Figure 1
t_{w2}	Pulse duration, $\overline{\text{CONVST}}$ low		40			ns	Figure 1, Figure 2
t_{d1}	Delay time, $\overline{\text{CONVST}}$ rising edge to sample start				5	ns	Figure 1
t_{d2}	Delay time, $\overline{\text{CONVST}}$ falling edge to conversion start				5	ns	Figure 1, Figure 2
t_{d3}	Delay time, $\overline{\text{CONVST}}$ falling edge to busy high	+VBD = 3.3 V			14	ns	Figure 1, Figure 2
		+VBD = 5 V			13		
t_{d4}	Delay time, conversion end to busy low	+VBD = 3.3 V			8	ns	Figure 1, Figure 2
		+VBD = 5 V			7		
t_{w3}	Pulse duration, $\overline{\text{CSTART}}$ high		100			ns	Figure 1, Table 3
t_{w4}	Pulse duration, $\overline{\text{CSTART}}$ low		45			ns	Figure 1, Figure 2, Table 3
t_{d5}	Delay time, $\overline{\text{CSTART}}$ rising edge to sample start				7.5	ns	Figure 1, Table 3
t_{d6}	Delay time, $\overline{\text{CSTART}}$ falling edge to conversion start				7.5	ns	Figure 1, Figure 2, Table 3
t_{d7}	Delay time, $\overline{\text{CSTART}}$ falling edge to busy high	+VBD = 3.3 V			16.5	ns	Figure 1, Figure 2, Table 3
		+VBD = 5 V			15.5		
I/O RELATED							
t_{d8}	Delay time, $\overline{\text{RD}}$ falling edge while $\overline{\text{CS}}$ low to BUS_BUSY high				16	ns	Figure 5
t_{d9}	Delay time, $\overline{\text{RD}}$ falling edge while $\overline{\text{CS}}$ low to SYNC_O and SDO out of 3-state condition (for device with LAT_Y/$\overline{\text{N}}$ pulled low)	+VBD = 3.3 V			29	ns	Figure 5
		+VBD = 5 V			28		
t_{d10}	Delay time, pre_conversion end (point A) to SYNC_O and SDO out of 3-state condition				22	ns	Figure 6
t_{d11}	Delay time, pre_conversion end (point A) to BUS_BUSY high	+VBD = 3.3 V			8	ns	Figure 6
		+VBD = 5 V			7		
t_{d12}	Delay time, conversion phase end to SYNC_O high		6		9 + t_{CLK}	ns	Figure 6
t_{d13}	Delay time, $\overline{\text{RD}}$ falling edge while $\overline{\text{CS}}$ low to SYNC_O high	+VBD = 3.3 V	5.5 + 4*t_{CLK}		8.5 + 5*t_{CLK}	ns	Figure 5
		+VBD = 5 V	5 + 4*t_{CLK}		8 + 5*t_{CLK}		
t_{w5}	Pulse duration, $\overline{\text{RD}}$ low for device in no latency mode		5			ns	Figure 11
t_{d14}	Delay time, CLK_O rising edge to data valid	+VBD = 3.3 V			1.4	ns	Figure 5, Figure 6
		+VBD = 5 V			1.3		
t_{d15}	Delay time, BUS_BUSY low to SYNC_O high in daisy chain mode indicating receiving device to output data	+VBD = 3.3 V	4*t_{CLK} – 6.5		4*t_{CLK} – 3	ns	Figure 7, Figure 12
		+VBD = 5 V	4*t_{CLK} – 6		4*t_{CLK} – 2.5		

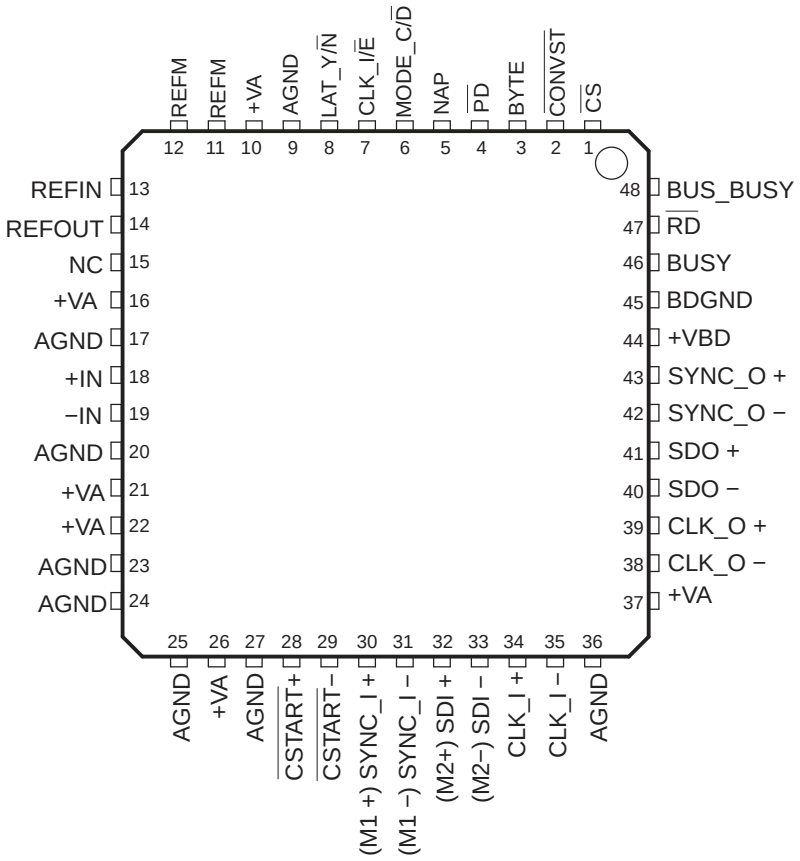
TIMING REQUIREMENTS (continued)

T_A = –40°C to 85°C, +VA = 5 V, +VBD = 5 V or 3.3 V (unless otherwise noted)

PARAMETER			MIN	TYP	MAX	UNIT	REF
t_{d16}	Delay time, CLK_O to SDO and SYNC_O 3-state				4	ns	Figure 7, Figure 8, Figure 12, Figure 15
t_{pd1}	Propagation delay time, SYNC_I to SYNC_O in daisy chain mode				11 + 0.5*t_{CLK}	ns	Figure 12
t_{d17}	Delay time, SYNC_O and SDO 3-state to BUS_BUSY low in cascade mode		0		2	ns	Figure 8
t_{d18}	Delay time, $\overline{RD}$ rising edge to BUS_BUSY high for device with LAT_Y/$\overline{N}$ = 1	+VBD = 3.3 V			8	ns	Figure 11, Figure 14
		+VBD = 5 V			7		
t_{d19}	Delay time, point A indicating clear for bus 3-state release to BUSY falling edge	+VBD = 3.3 V			40.5	ns	Figure 6
		+VBD = 5 V			40		
t_r	Rise time, differential LVDS output signal				950	ps	Figure 53
t_f	Fall time, differential LVDS output signal				950	ps	Figure 53
	CLK frequency (serial data rate)		190		210	MHz	
t_{d20}	Delay time, from $\overline{PD}$ falling edge to SDO 3-state				10	ns	Figure 22, Figure 23
t_{d21}	Delay time, from $\overline{PD}$ falling edge to device powerdown				10	µs	Figure 22, Figure 23
t_{d22}	Delay time, from $\overline{PD}$ rising edge to device powerup				25	ms	Figure 22, Figure 23
t_{s1}	Settling time, internal reference after first three conversions				4	ms	Figure 22
t_{d23}	Delay time, $\overline{CONVST}$ falling edge to start of restricted zone for start of data read cycle		335			ns	Figure 9
t_{d24}	Delay time, $\overline{CONVST}$ falling edge to end of restricted zone for start of data read cycle				406	ns	Figure 9

DEVICE INFORMATION

RGZ PACKAGE
(TOPVIEW)



NC − No internal connection

TERMINAL FUNCTIONS

TERMINAL		I/O	DESCRIPTION
NO.	NAME		
ANALOG PINS			
11, 12	REFM	I	Reference ground. Connect to analog ground plane.
13	REFIN	I	Reference (positive) input. Decouple with REFM pin using 0.1-µF bypass capacitor and 1-µF storage capacitor.
14	REFOUT	O	Internal reference output. Short to REFIN pin when the internal reference is used. Do not connect to the REFIN pin when an external reference is used. Always decouple with AGND using a 0.1-µF bypass capacitor.
18	+IN	I	Noninverting analog input channel
19	–IN	I	Inverting analog input channel (supports ±0.2 V i/p range)
LVDS I/O PINS(1)			
28, 29	$\overline{\text{CSTART}}$+ $\overline{\text{CSTART}}$–	I	Device sample and convert control input. Device enters sample phase with the rising edge of $\overline{\text{CSTART}}$ and the conversion phase starts with the falling edge of $\overline{\text{CSTART}}$ (provided other conditions are satisfied). Set $\overline{\text{CSTART}}$ = 0 when the $\overline{\text{CONVST}}$ input is used.

(1) All LVDS inputs and outputs are differential with signal+ and signal– lines. Whenever only the 'signal' is mentioned it refers to the signal+ line and the signal– line is the compliment. For example CLK_O refers to CLK_O+ and CLK_O is the compliment.

TERMINAL FUNCTIONS (continued)

TERMINAL		I/O	DESCRIPTION
NO.	NAME		
30, 31	SYNC_I + SYNC_I–	I Dasiy Chain	Connect to previous device SYNC_O with same polarity, while the device is selected to operate in daisy chain mode.
	M1+ M1–	I Cascade	Mode 1 (valid in cascade mode only). CLK_O is available while M1=1 (LVDS) or M1+ is pulled up to +VBD and M1– is grounded (AGND). CLK_O o/p goes to 3-state when M1 = 0 (LVDS) or M1+ is grounded (AGND) and M1– is pulled up to +VBD. Do not allow these pins to float.
32, 33	SDI+ SDI–	I Daisy Chain	Serial data input. Connect to previous device SDO with same polarity, while the device is selected to operate in daisy chain mode.
	M2+ M2–	I Cascade	Mode 2 (valid in cascade mode only). Doubles LVDS o/p current while M2 = 1 (LVDS) or M2+ is pulled up to +VBD and M2– is grounded (AGND). LVDS o/p current is normal (3.4 mA typ) when M2 = 0 (LVDS) or M2+ is grounded (AGND) and M2 – is pulled up to +VBD. Do not allow these pins to float.
34, 35	CLK_I+ CLK_I–	I	Serial external clock input. Set CLK_I/$\overline{E}$ (pin 7) = 0 to select an external clock source.
38, 39	CLK_O– CLK_O+	O	Serial clock out. Data is latched out on the rising edge of CLK_O and can be captured on the next falling edge.
40, 41	SDO– SDO+	O	Serial data out. Data is latched out on the rising edge of CLK_O with MSB first format.
42, 43	SYNC_O – SYNC_O +	O	Synchronizes the data frame. (2)
CMOS I/O PINS			
1	$\overline{CS}$	I	Chip select, active low signal. All of the LVDS o/p except CLK_O are 3-state if this pin is high.
2	$\overline{CONVST}$	I	CMOS equivalent of $\overline{CSTART}$ input. So functionality is the same as the $\overline{CSTART}$ input. Set $\overline{CONVST}$ = 0 when the $\overline{CSTART}$ input is used.
3	BYTE	I	Controls the data frame(3) duration. The frame duration is 16 CLKs if BYTE = 0 or 8 CLKs if BYTE = 1.
4	$\overline{PD}$	I	Active low input, acts as device power down.
5	NAP	I	Selects nap mode while high. Device enters the nap state at conversion end and remains so until the next acquisition phase begins.
6	MODE_C/$\overline{D}$	I	Selects cascade (MODE_C/$\overline{D}$ = 1) or daisy chain mode (MODE_C/$\overline{D}$ = 0).
7	CLK_I/$\overline{E}$	I	Selects the source of the I/O clock. CLK_I/$\overline{E}$ = 1 selects internally generated clock with 200-MHz typ frequency. CLK_I/$\overline{E}$ = 0 selects CLK_I as the I/O clock.
8	LAT_Y/$\overline{N}$	I	Controls the data read with latency (LAT_Y/$\overline{N}$ = 1) or without latency ((LAT_Y/$\overline{N}$ = 0). It is essential to set LAT_Y/$\overline{N}$ = 0 for the first device in daisy chain or cascade.
46	BUSY	O	Active high signal, indicates a conversion is in progress.
47	$\overline{RD}$	I	Data read request to the device, also acts as a handshake signal for daisy chain and cascade operation.
48	BUS_BUSY	O	Status output. Indicates that the bus is being used by the device. Connect to $\overline{RD}$ of the next device for daisy chain or cascade operation.
POWER SUPPLY PINS			
10, 16, 21, 22, 26, 37	+VA	–	Analog power supply and LVDS input buffer power supply.
9, 17, 20, 23, 24, 25, 27, 36	AGND	–	Analog ground pins. Short to the analog ground plane below the device.
44	+VBD	–	Digital power supply for all CMOS digital inputs and CMOS LVDS outputs.
45	BDGND	–	Digital ground for all digital inputs and outputs. Short to the analog ground plane below the device.

(2) The duration from the first rising edge of SYNC_O to the second rising edge of SYNC_O is one data frame. The data frame duration is 16 CLKs if BYTE = 0 or 8 CLKs if BYTE = 1.
(3) The duration from the first rising edge of SYNC_O to the second rising edge of SYNC_O is one data frame. The data frame duration is 16 CLKs if BYTE = 0 or 8 CLKs if BYTE = 1.

TERMINAL FUNCTIONS (continued)

TERMINAL		I/O	DESCRIPTION
NO.	NAME		
NOT CONNECTED PINS			
15	NC	–	No connection pins

Table 2. Device Configuration for Various Modes of Operation

OPERATION MODE		DEVICE PINS AND RECOMMENDED LOGIC LEVELS							COMMENTS	REFERENCE FIGURES	
		MODE_C/$\overline{D}$	CLK_I/$\overline{E}$	LAT_Y/$\overline{N}$	M1+	M1–	M2+	M2–		FOR SAMPLING AND CONVERSION	FOR DATA READ
Single device		1	1 or 0	0	+VBD	AGND	AGND	+VBD	Recommended configuration	1 or 2	See Figures 3,4 and 5,6,8 for more details
					or M1 = 1 LVDS		or M2 = 0 LVDS				
		0	1 or 0	0	See comments		See comments		Set SYNC_I and SDI to logic 0 or + terminal to AGND and –ve terminal to +VBD	1 or 2	See Figures 3,4 and 5,6,7 for more details
Multiple devices in daisy chain	1st Device	0	1 or 0	0	See comments		See comments		Set SYNC_I and SDI to logic 0 or + terminal to AGND and –ve terminal to +VBD	1 or 2	See Figures 3,4,11 and 6,12 for more details
	2nd To last device	0	0	1	See comments		See comments		Maximum 4 devices supported at 2 MSPS with 200-MHz CLK	1 or 2	
Multiple devices in cascade	1st Device	1	0	0	+VBD	AGND	AGND	+VBD	Maximum 3 devices supported at 2 MSPS	1 or 2	See Figures 3,4,14 and 6,15 for more details
					or M1 = 1 LVDS		or M2 = 0 LVDS(1)				
	2nd To last device	1	0	1	+VBD	AGND	AGND	+VBD			
					or M1 = 0 LVDS		or M2 = 0 LVDS(1)				

(1) Specified polarity is suitable for a 100-Ω differential load across the LVDS outputs. However, polarity can be reversed to double the output current in order to support two 100-Ω loads on both ends of the transmission lines, resulting in 50-Ω net load.

DETAILED DESCRIPTION

SAMPLE AND CONVERT

The sampling and conversion process is controlled by the $\overline{\text{CSTART}}$ (LVDS) or $\overline{\text{CONVST}}$ (CMOS) signal. Both signals are functionally identical. The following diagrams show control with $\overline{\text{CONVST}}$. The rising edge of $\overline{\text{CONVST}}$ (or $\overline{\text{CSTART}}$) starts the sample phase, if the conversion has completed and the device is in the wait state. Figure 2 shows the case when the device is in the conversion phase at the rising edge of $\overline{\text{CONVST}}$. In this case, the sample phase starts immediately at the end of the conversion phase and there is no wait state.

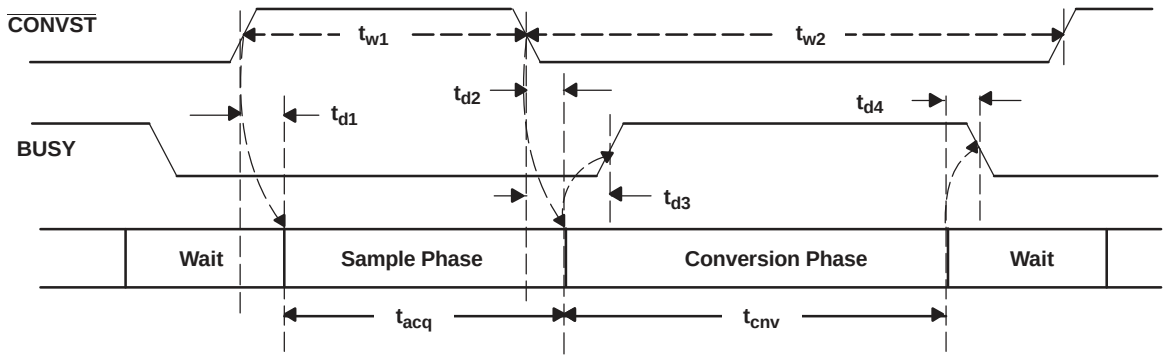


Figure 1. Sample and Convert With Wait (Less Than 2 MSPS Throughput)

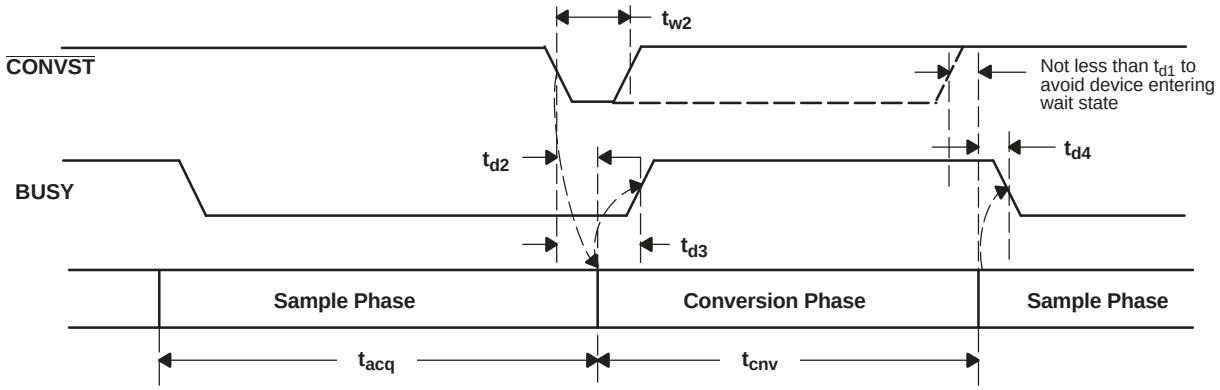


Figure 2. Sample and Convert With No Wait or *Back to Back* (2 MSPS Throughput)

The device ends the sample phase and enters the conversion phase on the falling edge of $\overline{\text{CONVST}}$ ($\overline{\text{CSTART}}$). A high level on the BUSY output indicates an ongoing conversion. The device conversion time is fixed. The falling edge of $\overline{\text{CONVST}}$ ($\overline{\text{CSTART}}$) during the conversion phase aborts the ongoing conversion. A data read after a conversion abort fetches invalid data. Valid data is only available after a sample phase and a conversion phase has completed. The timing diagram for control with $\overline{\text{CSTART}}$ is similar to Figure 1 and Figure 2. Table 3 shows the equivalent timing for control with $\overline{\text{CONVST}}$ and $\overline{\text{CSTART}}$.

Table 3. $\overline{\text{CONVST}}$ and $\overline{\text{CSTART}}$ Timing Control

TIMING CONTROL WITH $\overline{\text{CONVST}}$	TIMING CONTROL WITH $\overline{\text{CSTART}}$
t_{w1}	t_{w3}
t_{w2}	t_{w4}
t_{d1}	t_{d5}
t_{d2}	t_{d6}
t_{d3}	t_{d7}

DATA READ OPERATION

The ADS8410 supports a 200-MHz serial LVDS interface for data read operation. The three signal LVDS interface (SDO, CLK_O, and SYNC_O) is well suited for high-speed data transfers. An application with a single device or multiple devices can be implemented with a daisy chain or cascade configuration. The following sections discuss data read timing when a single device is used.

DATA READ FOR A SINGLE DEVICE (See Table 1 for Device Configuration)

For a single device, there are two possible read cycle starts: a data read cycle start during a wait or sample phase or a data read cycle start at the end of a conversion phase. Read cycle end conditions can change depending on MODE C/$\overline{\text{D}}$ selection. Figure 3 explains the data read cycle. The details of a read frame start with the two previous listed conditions and a read cycle end with MODE C/$\overline{\text{D}}$ selection are explained in Figure 5 and Figure 6 and Figure 7 and Figure 8, respectively.

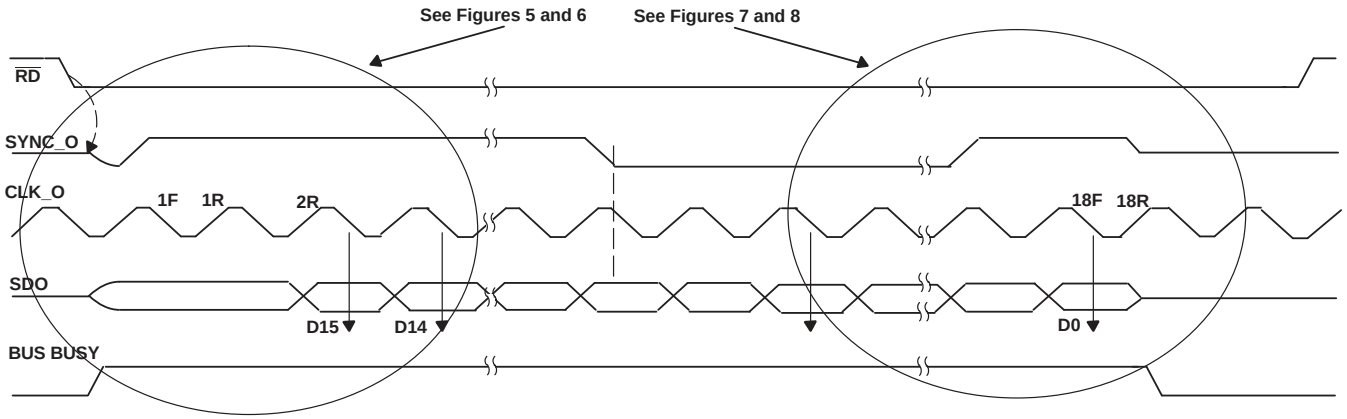


Figure 3. Data Read With $\overline{CS}$ Low and BYTE = 0

As shown in Figure 3, a new data read cycle is initiated with the falling edge of $\overline{RD}$, if $\overline{CS}$ is low and the device is in a wait or sample phase. The device releases the LVDS o/p (SYNC_O, SDO) from 3-state and sets BUS_BUSY high at the start of the read cycle. The SYNC_O cycle is 16 clocks wide (rising edge to rising edge) if BYTE i/p is held low and can be used to synchronize a data frame. The clock count begins with the first CLK_O falling edge after a SYNC_O rising edge. The MSB is latched out on the second rising edge (2R) and each subsequent data bit is latched out on the rising edge of the clock. The receiver can shift data bits on the falling edges of the clock. The next rising edge of SYNC_O coincides with the 16th rising edge of the clock. D0 is latched out on the 17th rising edge of the clock. The receiver can latch the de-serialized 16-bit word on the 18th rising edge (18R, or the second rising edge after a SYNC_O rising edge).

$\overline{CS}$ high during a data read 3-states SYNC_O and SDO. These signals remain in 3-state until the start of the next data read cycle.

DATA READ IN BYTE MODE

Byte mode is selected by setting BYTE = 1; this mode is allowed for any condition listed in Table 2. Figure 4 shows a data read operation in byte mode.

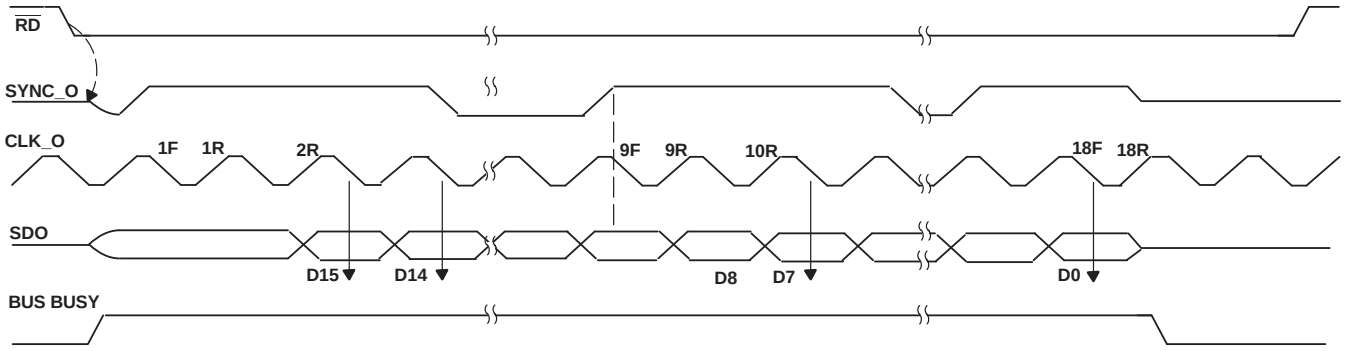


Figure 4. Data Read Timing Diagram with $\overline{CS}$ Low and BYTE = 1

Similar to Figure 3, a new data read cycle is initiated with the falling edge of $\overline{RD}$, if $\overline{CS}$ is low and the device is in a wait or sample phase. The device releases the LVDS o/p (SYNC_O, SDO) from 3-state and sets BUS_BUSY high at the start of the read cycle. The SYNC_O cycle is 8 clocks wide (rising edge to rising edge) if BYTE i/p is held high and can be used to synchronize a data frame. The clock count begins with the first falling edge of CLK_O after the rising edge of SYNC_O. The MSB is latched out on the second rising edge (2R) and each subsequent data bit is latched out on the rising edge of the clock. The receiver can shift data bits on the falling edges of clock. The next rising edge of SYNC_O coincides with the 8th rising edge of the clock. D8 is latched out on the 9th rising edge of the clock. The receiver can latch the de-serialized higher byte on the 10th rising edge (10R, or second rising edge after a SYNC_O rising edge). The de-serialized lower byte can be latched on the 18th rising edge (18R).

$\overline{CS}$ high during a data read 3-states SYNC_O and SDO. These signals remain in 3-state until the start of the next data read cycle.

DATA READ CYCLE START DURING WAIT OR SAMPLE PHASE

As shown in Figure 5, the falling edge of $\overline{RD}$, with $\overline{CS}$ low and the device is in a wait or sample phase, triggers the start of a read cycle. The cycle starts when BUS_BUSY goes high and SYNC_O and SDO are released from 3-state. SYNC_O is low at the start and rises to a high level t_{d13} ns after the falling edge of $\overline{RD}$. As shown in Figure 5, the MSB is shifted on the 2nd rising edge of the clock (2R). Other details about the data read cycle are discussed in the previous section (see Figure 3).

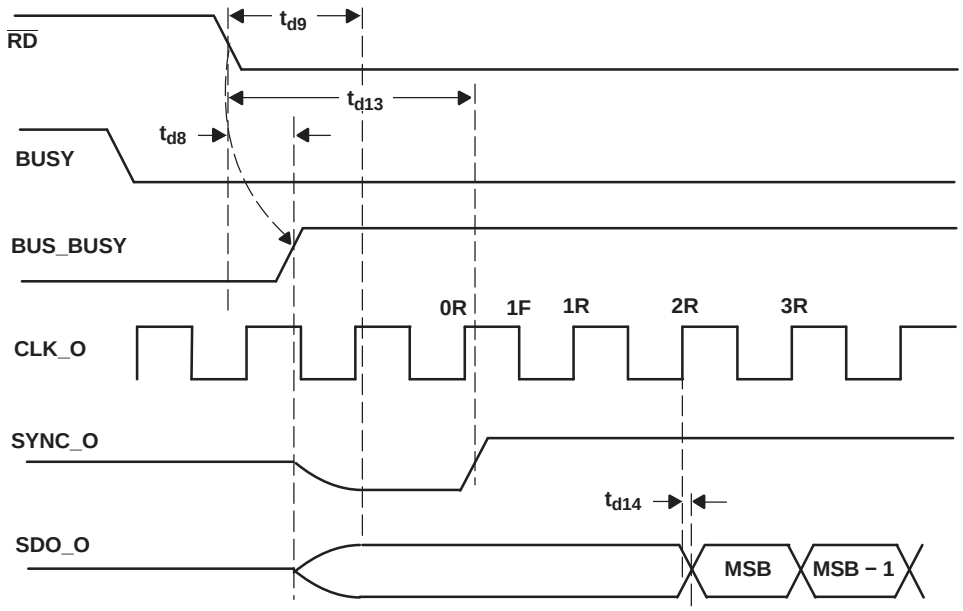


Figure 5. Start of Data Read Cycle with $\overline{RD}$ with $\overline{CS}$ Low and Device in Wait or Sample Phase

DATA READ CYCLE START AT END OF CONVERSION PHASE (Read Without Latency, Back-to-Back)

This mode is optimized for a data read immediately after the end of a conversion phase and ensures the data read is complete before the sample end while running at 2 MSPS. Point A in Figure 6 indicates a 'pre_conversion_end'; it occurs t_{d19} ns before the falling edge of BUSY or $[(t_{d2} + t_{cnv} + t_{d4}) – t_{d19}]$ ns after the falling edge of $\overline{CONVST}$. A read cycle is initiated at point A if $\overline{RD}$ is issued before point A while $\overline{CS}$ is low. Alternately, $\overline{RD}$ and $\overline{CS}$ can be held low. At the start of the read cycle, BUS_BUSY rises to a high level and the LVDS outputs are released from 3-state. The rising edge of SYNC_O occurs t_{d12} ns after the conversion end. As shown in Figure 6, the MSB is shifted on the 2nd rising edge of the clock (2R). Other details about the data read cycle are discussed in the previous section (see Figure 3).

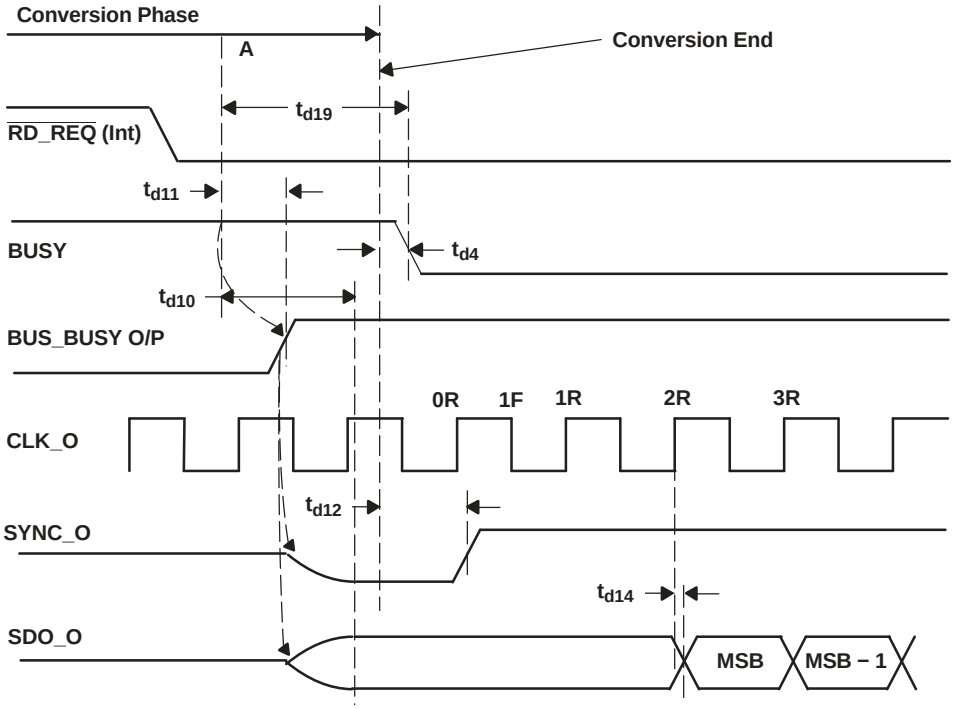


Figure 6. Start of Data Read Cycle with End of Conversion

DATA READ CYCLE END (With MODE C/$\overline{D}$ = 0)

A data read cycle ends after all 16 bits have been serially latched out. Figure 7 shows the timing of the falling edge of BUS_BUSY and the rising edge of SYNC_O with respect to SDO. SYNC_O rises on the 16th rising edge of CLK_O. As shown in Figure 5 and Figure 6, the MSB is shifted out on the 2nd rising edge of CLK_O. Therefore, the LSB-1 is shifted out on the 16th rising edge of CLK_O.

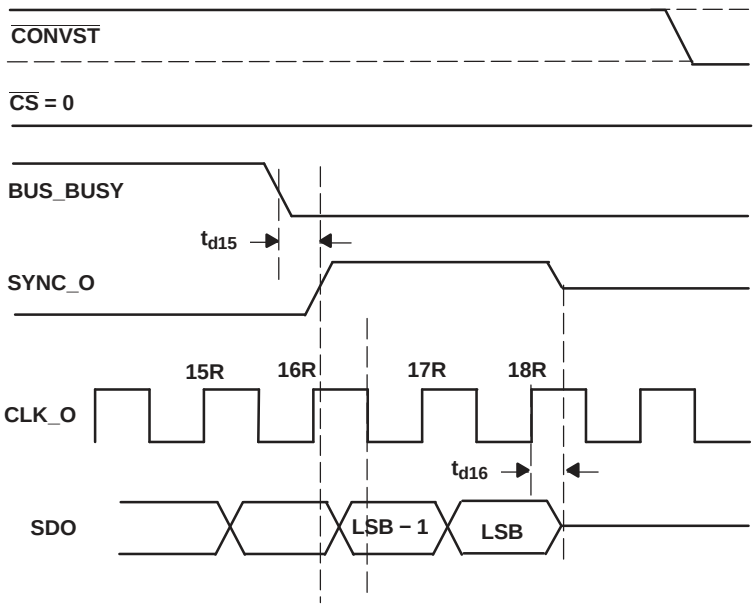


Figure 7. Data Read Cycle End with MODE C/$\overline{D}$ = 0

The next two rising edges of CLK_O are shown as 17R and 18R in Figure 7. On 17R the LSB is latched out, and on 18R SDO and SYNC_O go to 3-state. Note that BUS_BUSY falls t_{d15} ns before the rising edge of SYNC_O when MODE C/$\overline{D}$ = 0. Care must be taken not to allow LVDS bus usage by any other device until the end of the read cycle or (t_{d15} + 2/f_{clk} + t_{d16}) ns after the falling edge of BUS_BUSY.

DATA READ CYCLE END (With MODE C/$\overline{D}$ = 1)

A data read cycle ends after all 16 bits have been serially latched out. Figure 8 shows the timing of the falling edge of BUS_BUSY and the rising edge of SYNCO with respect to SDO. SYNC_O rises on the 16th rising edge of CLK_O. As shown in Figure 5 and Figure 6, the MSB is shifted out on the 2nd rising edge of CLK_O. Therefore, the LSB-1 is shifted out on the 16th rising edge of CLK_O.

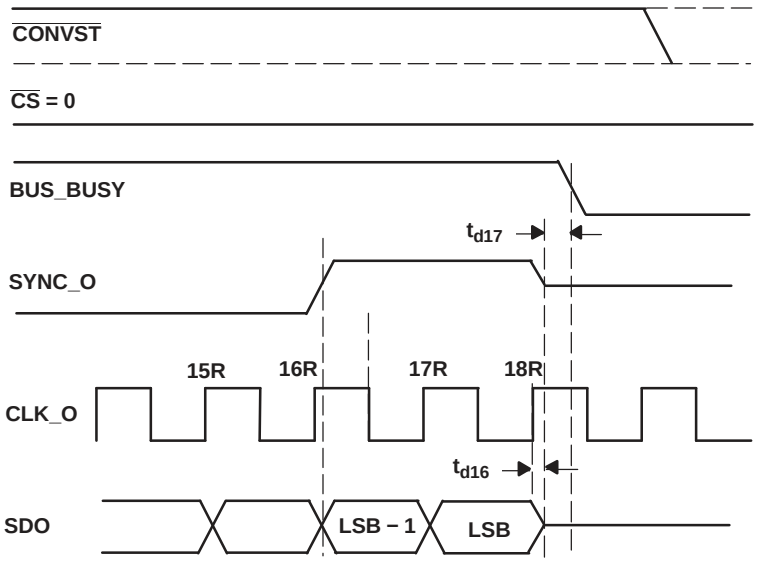


Figure 8. Data Read Cycle End with MODE C/$\overline{D}$ = 1

The next two rising edges of CLK_O are shown as 17R and 18R in Figure 8. On 17R the LSB is latched out and on 18R SDO and SYNC_O go to 3-state. In cascade mode (with MODE C/$\overline{D}$ = 1), unlike daisy chain mode, the falling edge of BUS_BUSY occurs after the LVDS outputs are 3-stated. One can use the falling edge of BUS_BUSY to allow LVDS bus usage by any other device.

RESTRICTIONS ON READ CYCLE START

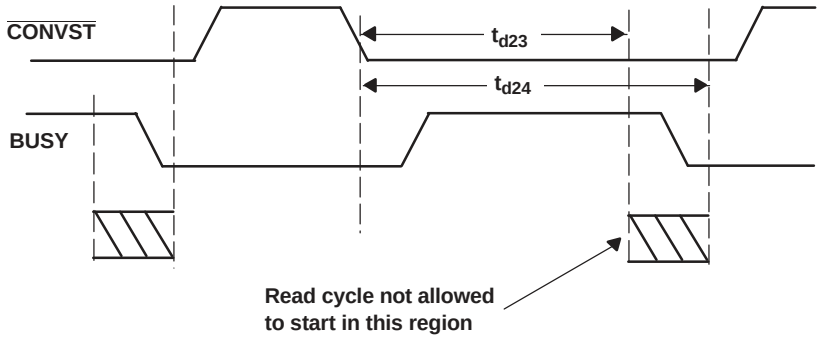


Figure 9. Read Cycle Restriction Region

The start of a data read cycle is not allowed in the region bound by t_{d23} and t_{d24}. Previous conversion results are available for a data read cycle start before this region, and current conversion results are available for a read cycle start after this region.

MULTIPLE DEVICES IN DAISY CHAIN OR CASCADE

Multiple devices can be connected in either a daisy chain or cascade configuration. The following sections describe detailed timing diagrams and electrical connections. The ADS8410 provides all of the handshake signals required for both of these modes. CONVST or CSTART is the only external signal needed for operation.

DAISY CHAIN

Figure 10 shows the first two devices in daisy chain. The signals shown by double lines are LVDS and the others are CMOS. Daisy chain mode is selected by setting MODE_C/$\overline{D}$ = 0. The first device in the chain is identified by selecting LAT_Y/$\overline{N}$ = 0.

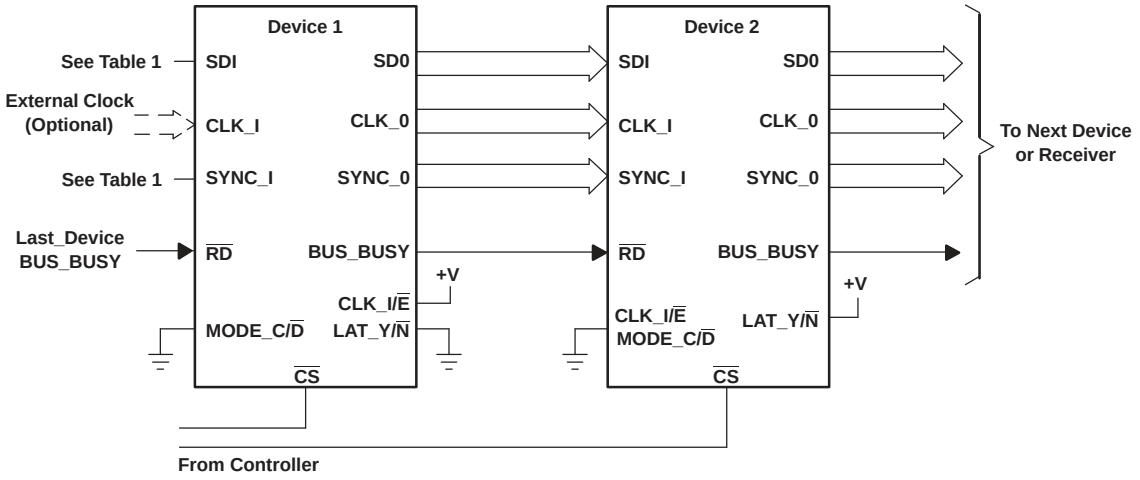


Figure 10. Connecting Multiple Devices in Daisy Chain

For all other devices in the chain LAT_Y/$\overline{N}$ = 1. See Table 2 for more details on device configurations. SDO, CLK_O, and SYNC_O of device *n* are connected to SDI, CLK_I, and SYNC_I of device *n+1*. SDO, CLK_O, and SYNC_O of the last device in the chain go to the receiver. BUS_BUSY of device *n* is connected to $\overline{RD}$ of device *n+1* and so on. Finally, BUS_BUSY of the last device in the chain is connected to $\overline{RD}$ of device 1. This ensures the necessary handshake to seamlessly propagate the data of all devices through the chain (it is also allowed to tie $\overline{RD}$ = 0 for device 1).

TIMING DIAGRAMS FOR DAISY CHAIN OPERATION

The conversion speed for n devices in the chain must be selected such that:

1/conversion speed > read startup delay + n*(data frame duration) + t_{d16}

Read startup delay = 10 ns + (t_{d19} - t_{d4}) + t_{d12} + 2/f_{CLK}

Data frame duration = 16/f_{CLK} Note that it is not necessary for all devices in the chain to sample the data simultaneously. But all of the devices must operate with the same exact conversion speed.

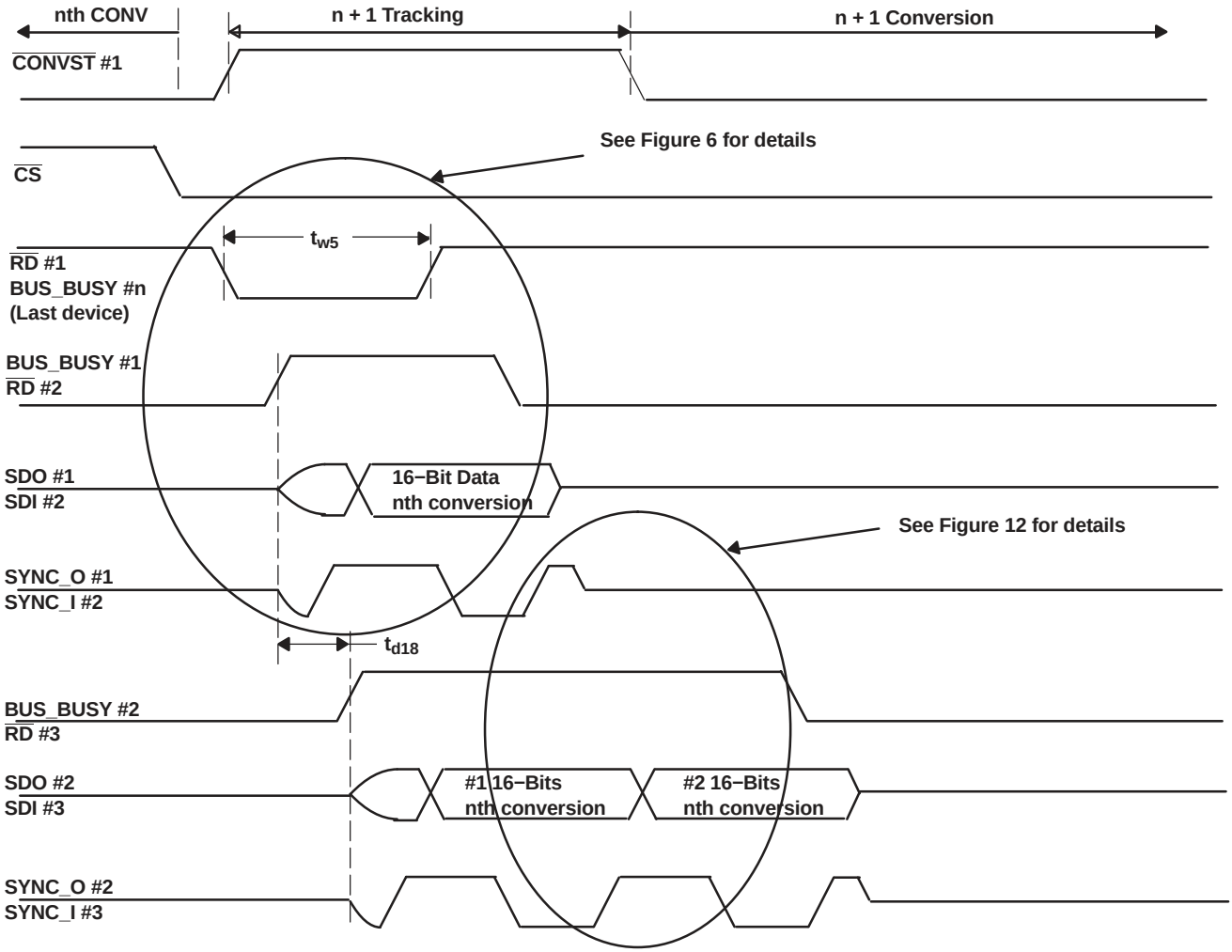


Figure 11. Data Read Operation for Devices in Daisy Chain

DATA READ OPERATION

On power up, BUS_BUSY of all of the devices is low. The devices receive $\overline{\text{CONVST}}$ or $\overline{\text{CSTART}}$ to sample and start the conversion. The first device in the chain starts the data read cycle at the end of its conversion. BUS_BUSY of device 1 (connected to $\overline{\text{RD}}$ of device 2) goes high on the read cycle start. Device 2 BUS_BUSY goes high on the rising edge of $\overline{\text{RD}}$. This propagates until the last device in the chain. Device 2 receives CLK_I, SDI, and SYNC_I from device 1 and it passes all of these signals to the next device. Device 2 (and every subsequent device in the chain) passes the received signals to its output until it sees the falling edge of $\overline{\text{RD}}$ (same as BUS_BUSY of the previous device). In daisy chain mode, BUS_BUSY for any device falls when it has passed all of the previous device data followed by its own data. The falling edge of BUS_BUSY occurs before the rising edge of SYNC_O. This indicates to the receiving device that the previous data chain is over and it is its own turn to output the data. The device outputs the data from the last completed conversion. BUS_BUSY of the last device in the chain is fed back to $\overline{\text{RD}}$ of the first device as shown in Figure 10 (or device 1 $\overline{\text{RD}}$ tied to 0). This makes sure that $\overline{\text{RD}}$ of device 1 is low before its conversion is over. The chain continues with only one external signal ($\overline{\text{CONVST}}$ or $\overline{\text{CSTART}}$) when $\overline{\text{CS}}$ is held low. Every device LVDS output goes to 3-state once all data transfer through the device has been completed.

$\overline{\text{CS}}$ going high during the data read cycle of any device 3-states its SYNC_O and SDO. This halts the propagation of data through the chain. To reset this condition it is necessary to assert $\overline{\text{CS}}$ high for all devices. The new read sequence starts only after $\overline{\text{CS}}$ for all devices is low before point A shown in Figure 6. The high pulse on $\overline{\text{CS}}$ must be at least 20 ns wide. It is better to connect $\overline{\text{CS}}$ of all of the devices together to avoid undesired halting of the daisy chain.

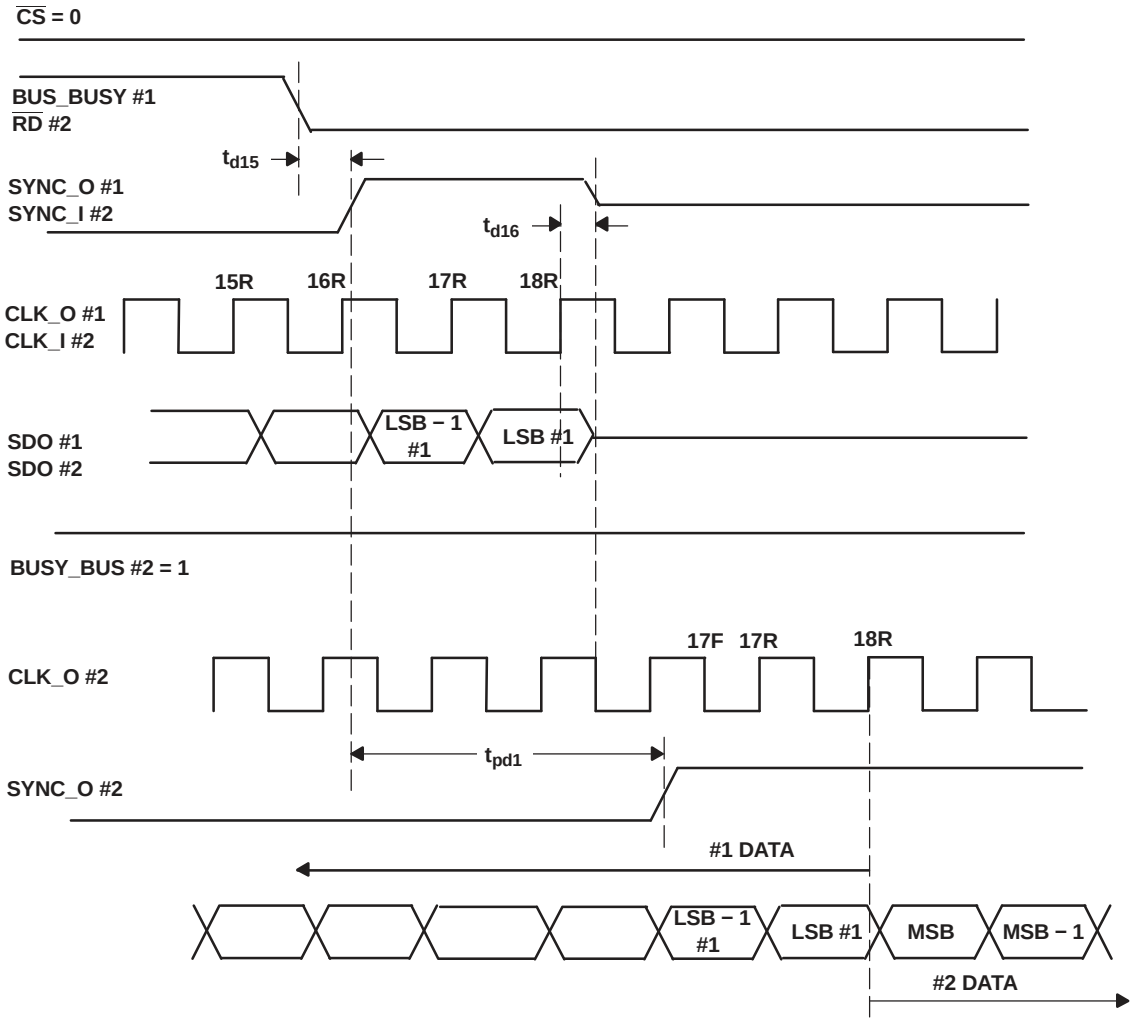


Figure 12. Data Propagation from Device *n* to Device *n+1* in Daisy Chain Mode

As shown in Figure 12 there is a propagation delay of t_{pd1} from SYNC_I to SYNC_O or SDI to SDO. Note that the data frames of all devices in the chain appear seamless at the last device output. The rising edge of SYNC occurs at an interval of 16 clocks (or 8 clocks in BYTE mode); this can be used as a data frame sync. The deserializer at the output of the last device can shift the data on every falling edge of the clock and it can latch the parallel 16-bit word on the second rising edge of CLK_O (shown as 18R) after every rising edge of SYNC_O.

CASCADE

Figure 13 shows the cascade connection. The signals shown by double lines are LVDS and the others are CMOS. Cascade mode is selected by setting MODE_C/$\overline{D}$ = 1. Similar to daisy chain, the first device in the chain is identified by selecting LAT_Y/$\overline{N}$ = 0. For all other devices in the chain LAT_Y/$\overline{N}$ = 1. See Table 2 for more details on device configuration. SDO, CLK_O, and SYNC_O are connected to the common bus. This means only one device occupies the bus at a time while the LVDS drivers for all other devices 3-state. Unlike SDO and SYNC_O, the clock cannot be switched out from device to device as the receiver requires a continuous clock. So only device 1 outputs the clock and CLK_O of all other devices is 3-stated by appropriately setting M1+ and M1- as listed in Table 2.

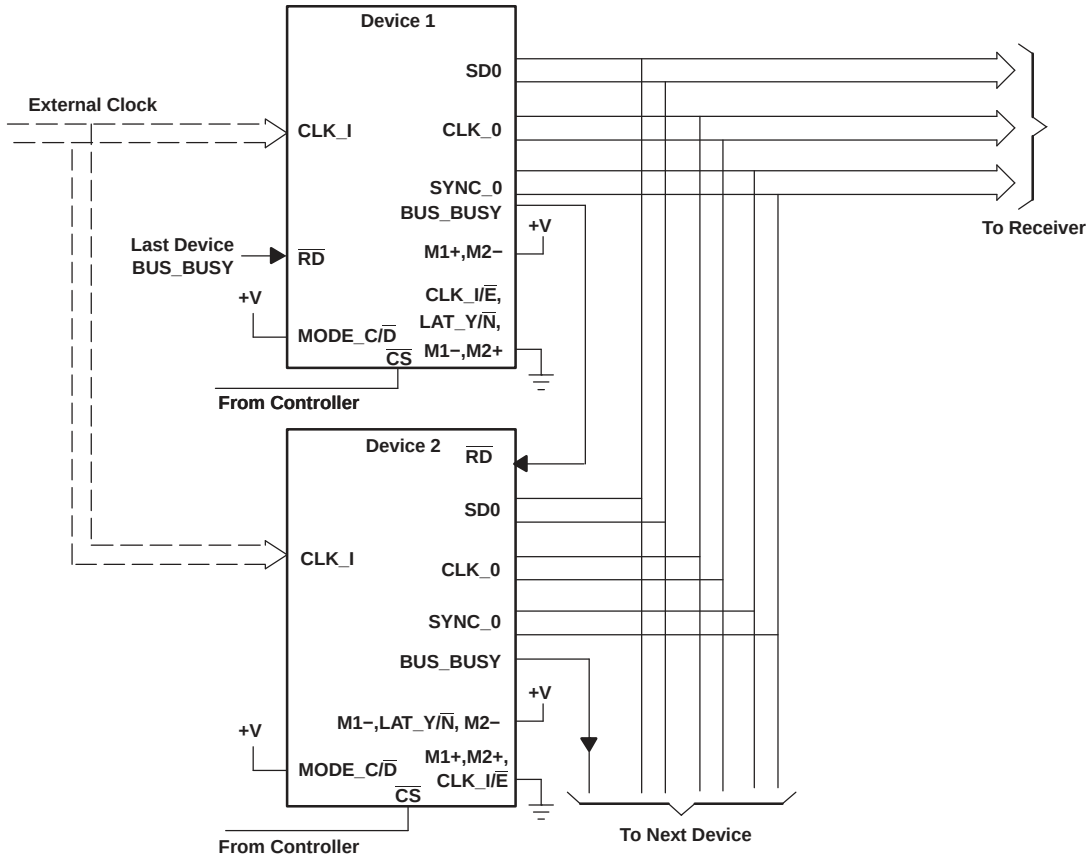


Figure 13. Cascade Connection

CLOCK SOURCE

In this mode it is very critical to control the skew between the three LVDS o/p signals. It is recommended to use external clock mode only for all of the devices in cascade. BUS_BUSY of device *n* is connected to $\overline{RD}$ of device *n + 1* and so on. Finally BUS_BUSY of the last device in the chain is to be connected to $\overline{RD}$ of device 1. This ensures the necessary handshake to control the sequence of data reads for all of the devices in cascade. (It is also allowed to tie $\overline{RD}$ to 0 for device 1.)

TIMING DIAGRAMS FOR CASCADE OPERATION

The conversion rate for n devices in cascade must be selected such that:

1/conversion speed > first device read cycle duration + (n - 1) next device read cycle duration

First device read cycle duration = read startup delay_1 + data frame duration + (t_{d16} + t_{d17})

Next device read cycle duration = read startup delay_n + data frame duration + (t_{d16} + t_{d17})

Read startup delay_1 = 10 ns + (t_{d19} - t_{d4} + t_{d12}) + 2/f_{clk}

Read startup delay_n = (t_{d13} + 2/f_{clk})

Data frame duration = 16/f_{clk} Note that it is not necessary that all devices in the chain sample the data simultaneously. But all of the devices must operate with the same exact conversion speed.

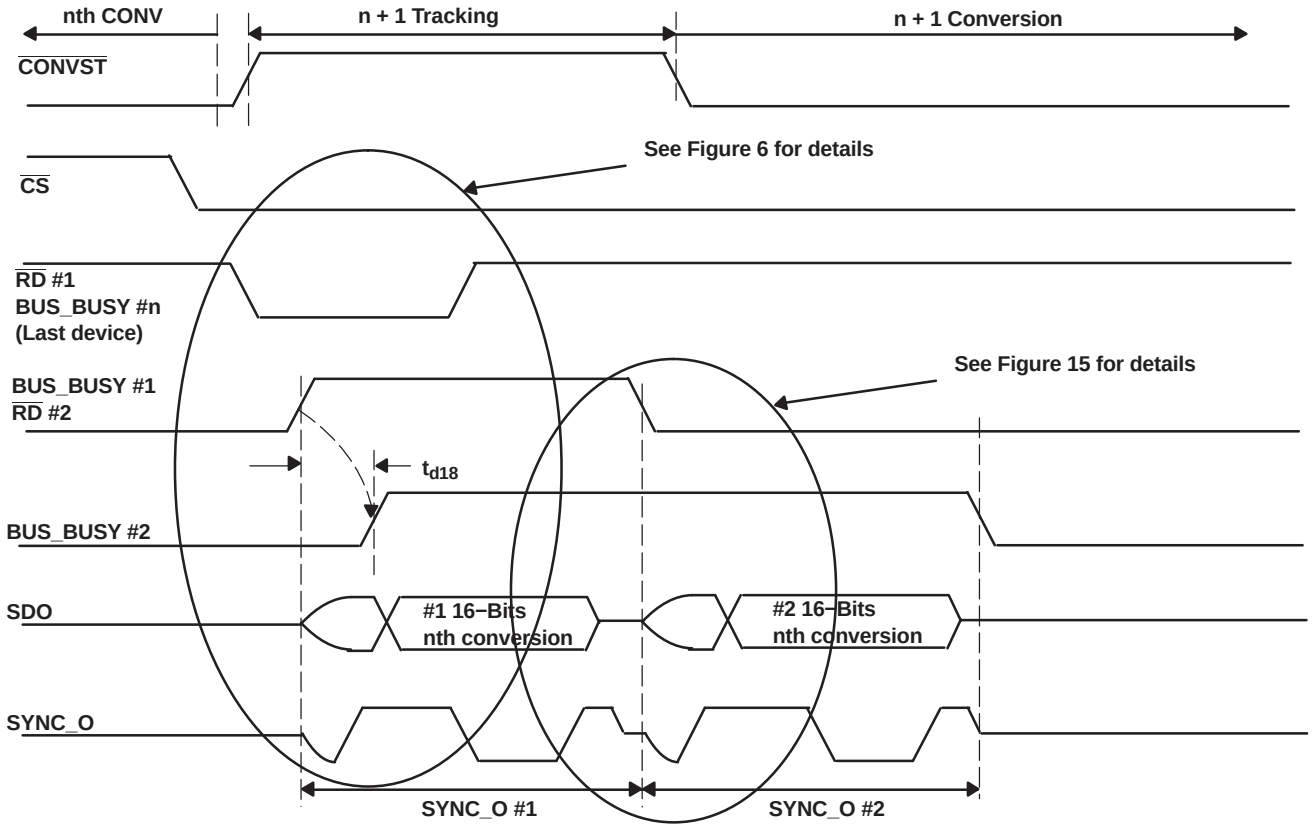


Figure 14. Data Read Operation for Devices in Cascade Mode

DATA READ OPERATION

On power up, BUS_BUSY for all of the devices is low. The devices receive $\overline{\text{CONVST}}$ or $\overline{\text{CSTART}}$ to sample and start the conversion. The first device starts the data read cycle at the end of its conversion. BUS_BUSY of device 1 (connected to $\overline{\text{RD}}$ of device 2) goes high on the read cycle start indicating that it wants to occupy the bus. Device 2 BUS_BUSY goes high on the rising edge of $\overline{\text{RD}}$. This propagates until the last device.

Device 1 BUS_BUSY goes low after it outputs its data; at this time SDO and SYNC_O for device 1 go to 3-state. The falling edge of BUS_BUSY ($\overline{\text{RD}}$ of the next device) indicates to the next device that it is its turn to output the data. The next device outputs the data from the last completed conversion. BUS_BUSY of the last device goes low and its SYNC_O and SDO go to 3-state after it outputs its data. BUS_BUSY of the last device is fed back to $\overline{\text{RD}}$ of the first device as shown in Figure 13 ($\overline{\text{RD}}$ can also be tied to 0 for device 1). This ensures that $\overline{\text{RD}}$ of device 1 is low before its conversion is over. The data read sequence continues with only one external signal, $\overline{\text{CONVST}}$ or $\overline{\text{CSTART}}$, when $\overline{\text{CS}}$ = 0. For any device, $\overline{\text{CS}}$ high during the data read cycle 3-states SYNC_O and SDO of the device and halts the data read sequence. To reset this condition it is necessary to assert $\overline{\text{CS}}$ high for all of the devices. The new read sequence starts only after $\overline{\text{CS}}$ for all of the devices is low before point A shown in Figure 6. The high pulse on $\overline{\text{CS}}$ must be at least 20 ns wide. It is better to connect $\overline{\text{CS}}$ for all of the devices together to avoid undesired halting of the data read sequence.

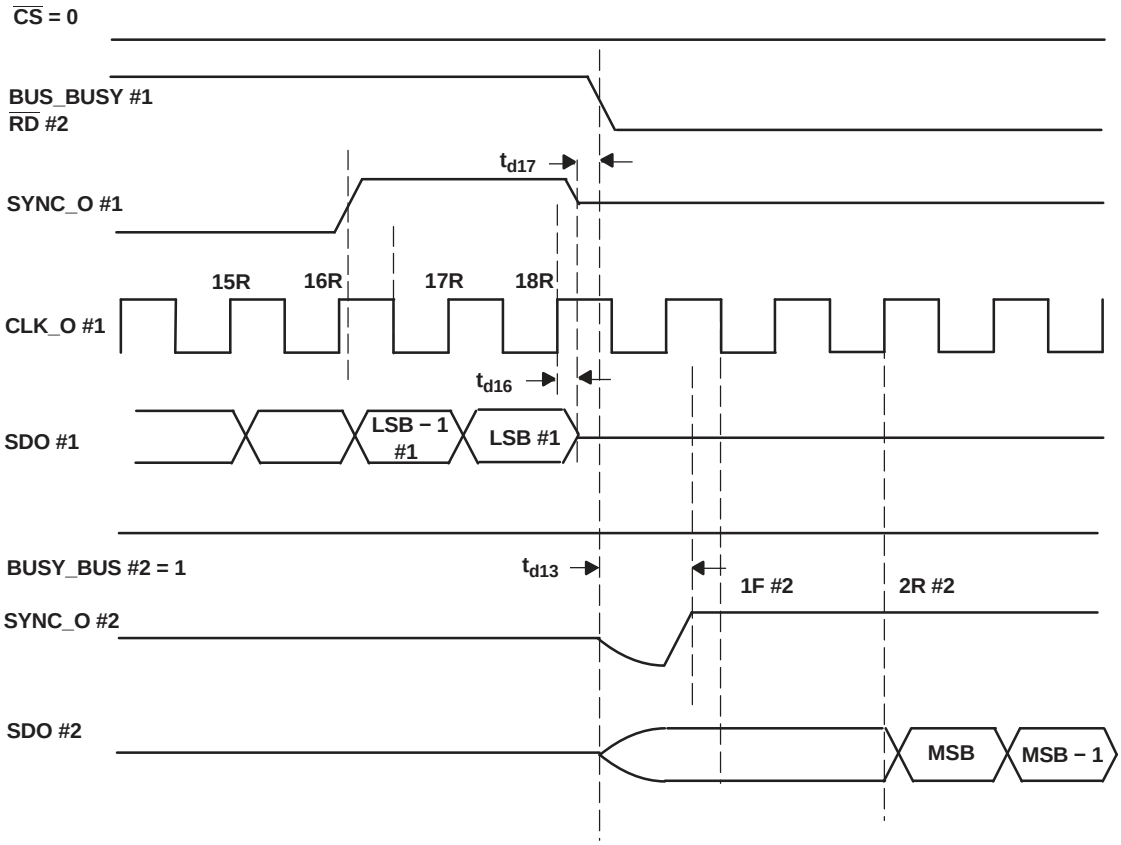


Figure 15. Device *n* Read Cycle End and Device *n+1* Read Cycle Start

Unlike daisy chain, the data frames of all the devices in cascade are not seamless and there is a loss of time between one device 3-state to other device data valid due to wakeup time from 3-state and a two clock phase shift between SYNC and data (see Figure 15 for details). As a result, the number of data frames per second in this mode are less than in daisy chain mode. Also, a maximum of 4 devices can be cascaded on the same bus. But, I/O power per device is considerably lower in cascade as compared to daisy chain as each device LVDS o/p goes to 3-state after its data transfer. The deserializer at the output of the last device can shift the data on every clock falling edge, and it can latch the parallel 16-bit word on the second CLK_O rising edge (shown as 18R) after every SYNC_O rising edge.

THEORY OF OPERATION

The ADS8410 is a member of the high-speed successive approximation register (SAR) analog-to-digital converters family. The architecture is based on charge redistribution, which inherently includes a sample/hold function. The device includes a built-in conversion clock, internal reference, and 200-MHz LVDS serial interface. The device can be operated at maximum throughput of 2 MSPS.

ANALOG INPUT

An analog input is provided to two input pins: +IN and -IN. When a conversion is initiated, the voltage difference between these pins is sampled on the internal capacitor array. While a conversion is in progress, both inputs are disconnected from any internal function.

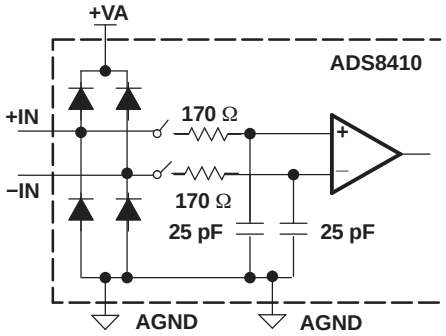


Figure 16. Simplified Input Circuit

When the converter enters hold mode, the voltage difference between the +IN and -IN inputs is captured on the internal capacitor array. The input current on the analog inputs depends upon a number of factors: sample rate, input voltage, signal frequency, and source impedance. Essentially, the current into the ADS8410 charges the internal capacitor array during the sample period. After this capacitance has been fully charged, there is no further input current (this may not happen when the signal is moving continuously). The source of the analog input voltage must be able to charge the input capacitance (25 pF) to better than a 16-bit settling level with a step input within the acquisition time of the device. For calculation, the step size can be selected equal to the maximum voltage difference between two consecutive samples at the maximum signal frequency (see the TYPICAL ANALOG INPUT CIRCUIT section). When the converter goes into hold mode, the input impedance is greater than 1GΩ.

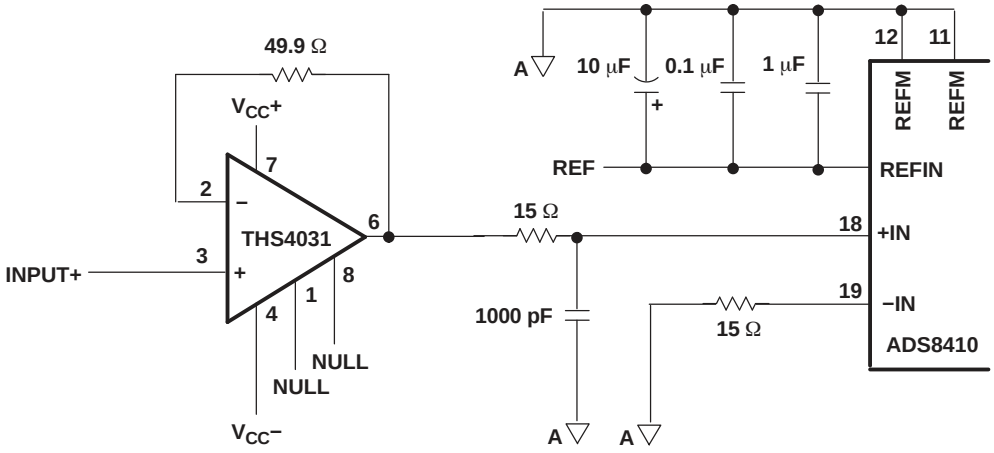


Figure 17. Typical Analog Input Schematic

Care must be taken regarding the absolute analog input voltage. To maintain the linearity of the converter, both -IN and +IN inputs should be within the limits specified. Outside of these ranges, the converter linearity may not meet specifications. Care should be taken to ensure that +IN and -IN see the same impedance to the respective sources. If this is not observed, the two inputs could have different setting times. This may result in offset error, gain error, and linearity error which changes with temperature and input voltage.

REFERENCE

The ADS8410 has a built-in 4.096-V (nominal value) reference. The ADS8410 can also operate with an external reference. When the internal reference is used, pin 14 (REFOUT) should be connected to pin 13 (REFIN), and a 0.1-µF decoupling capacitor and 1-µF storage capacitor must be connected between pin 14 (REFOUT) and pins 11 and 12 (REFM) (see Figure 18). The internal reference of the converter is buffered.

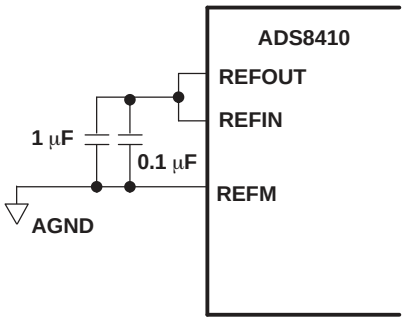


Figure 18. Using Internal Reference

The REFIN pin is also internally buffered. This eliminates the need to put a high bandwidth buffer onboard to drive the ADC reference and saves system area and power. When an external reference is used, the reference must be low noise, which can be achieved by the additional bypass capacitor from the REFIN pin to the REFM pin (see Figure 19). REFM must be connected to the analog ground plane.

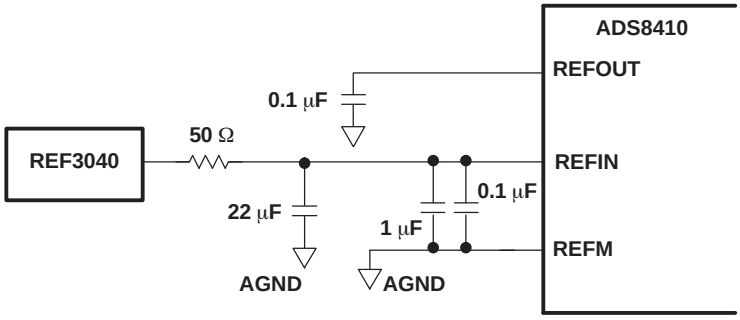


Figure 19. Using External Reference

DIGITAL INTERFACE

TIMING AND CONTROL

Refer to the timing diagrams and TIMING REQUIREMENTS table for detailed information.

SAMPLING AND CONVERSION

Sampling and conversion is controlled by the $\overline{\text{CONVST}}$ pin. For higher noise performance it is essential to have low jitter on the falling edge of $\overline{\text{CONVST}}$. The device uses the internally generated clock for conversion, hence it has a fixed conversion time.

READING DATA

The ADS8410 includes a high-speed LVDS serial interface. As discussed prior, an external clock (CLK_I, less than 200 MHz) or an internal 200-MHz clock can be used for a data read. The device outputs data in two's compliment format. Table 4 lists the ideal output codes.

Table 4. Ideal Input Voltages and Output Codes

DESCRIPTION	ANALOG VALUE (+IN – (–IN))	HEX CODE
Full-scale range	$+V_{ref}$	–
Least significant bit (LSB)	$+V_{ref}/2^{16}$	–
Full scale	V_{ref} – 1 LSB	FFFF
Midscale	$+V_{ref}/2$	8000
Midscale – 1 LSB	V_{ref} /2 – 1 LSB	7FFF

Table 4. Ideal Input Voltages and Output Codes (continued)

DESCRIPTION	ANALOG VALUE (+IN – (–IN))	HEX CODE
Zero	0 V	0000

The restrictions on read cycle start are described in the section RESTRICTIONS ON READ CYCLE START (see Figure 9).

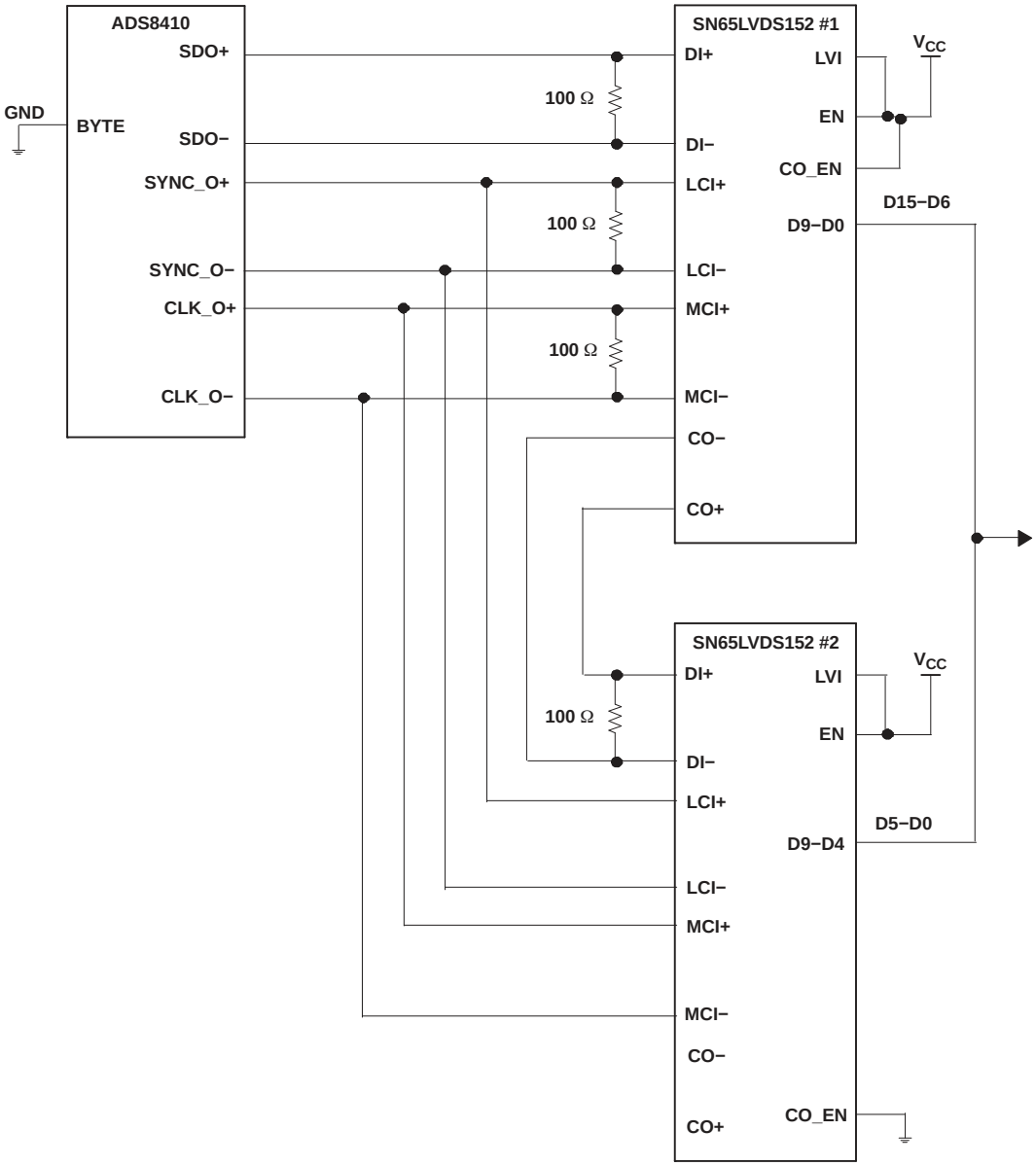


Figure 20. 16-Bit Data De-Serialization While BYTE = 0

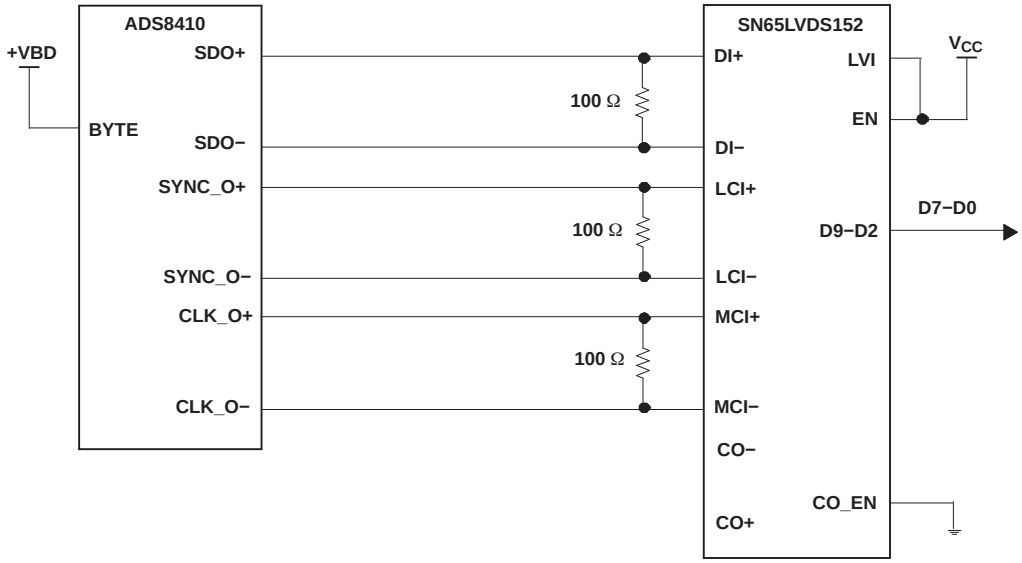


Figure 21. 8-Bit Data De-Serialization While BYTE = 1, Data

POWER SAVING

The converter provides two power saving modes, full powerdown and nap. Table 5 lists information on the activation/deactivation and resumption times for both modes.

Table 5. Powerdown Modes

POWERDOWN MODE	SDO	POWER CONSUMPTION	ACTIVATED BY	ACTIVATION TIME	RESUME POWER BY
Normal operation	Refer to DATA READ OPERATION section	58 mA	NA	NA	NA
Full powerdown (internal reference)	3 Stated	1 µA	$\overline{PD}$ = 0	t_{d21}	$\overline{PD}$ = 1
Full powerdown (external reference)	3 Stated	1 µA	$\overline{PD}$ = 0	t_{d21}	$\overline{PD}$ = 1
Nap powerdown	Not 3 stated	25 mA	Nap = 1	150 ns	Sample start

FULL POWERDOWN MODE

Full powerdown mode is activated by deasserting $\overline{PD}$ = 0; the device takes t_{d21} ns to reach the full powerdown state. The device can return to normal mode from full powerdown by asserting $\overline{PD}$ = 1. The powerup sequence is different for device operation with an internal reference or external reference as shown in Figure 22 and Figure 23.

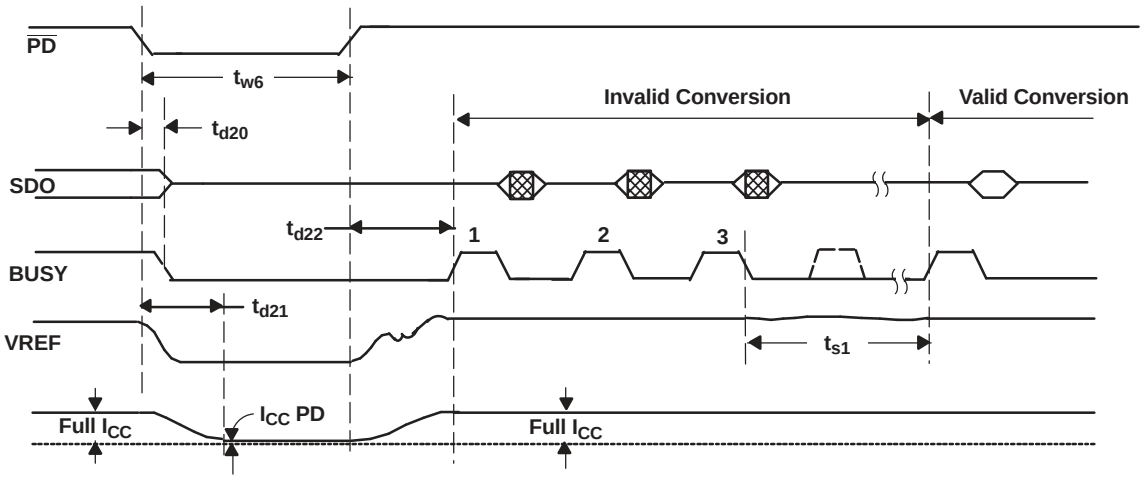


Figure 22. Device Full Powerdown and Powerup Sequence with Device Operation in Internal Reference Mode

When an internal reference is used, a conversion can be started t_{d22} ns after asserting $\overline{PD}$ = 1. After the first three conversions, t_{s1} ns are required for reference voltage settling to the trimmed value. Any conversions after this provide data at the specified accuracy.

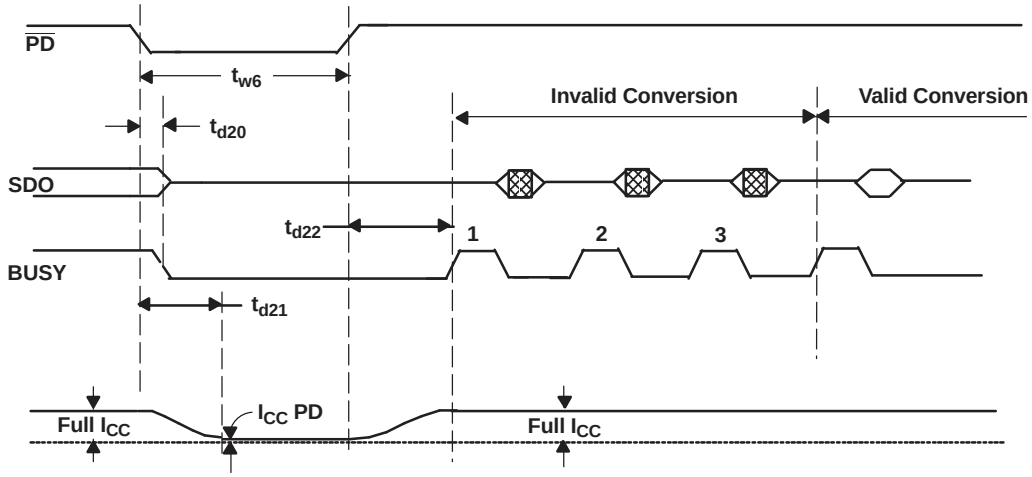


Figure 23. Device Full Powerdown and Powerup Sequence with Device Operation in External Reference Mode

When an external reference is used, a conversion can be started t_{d22} n after asserting $\overline{PD}$ = 1. The first three conversions are required for internal circuit stabilization. Any conversions after this provide data at the specified accuracy.

NAP MODE

The device automatically enters the nap state if nap = 1 at end of a conversion, and it remains in the nap state until the start of the sampling phase. A minimum of 150 ns is required after a sample start for the device to come out of the nap state and to perform normal sampling. So the minimum sampling time needed for nap mode is $t_{acq(min)}$ + 150 ns, or the maximum conversion speed in nap mode is 1.5 MHz.

LAYOUT

For optimum performance, care should be taken with the physical layout of the ADS8410 circuitry. The device offers single-supply operation, and it is often used in close proximity with digital logic, FPGA, microcontrollers, microprocessors, and digital signal processors. The more digital logic present in the design and the higher the switching speed, the more difficult it is to achieve good performance from the converter.

The basic SAR architecture is sensitive to glitches or sudden changes on the power supply, reference, ground connections, and digital inputs that occur just prior to the end of sampling and just prior to latching the output of the analog comparator during the conversion phase. Such glitches might originate from switching power supplies, nearby digital logic, or high power devices. Noise during the end of sampling and the later half of a conversion must be kept to a minimum (the former half of a conversion is not very sensitive since the device uses a proprietary error correction algorithm to correct for transient errors during this period).

The degree of error in the digital output depends on the reference voltage, layout, and the exact timing of the external event. On average, the device draws very little current from an external reference as the reference voltage is internally buffered. If the reference voltage is external and originates from an op amp, make sure that it can drive the bypass capacitor or capacitors without oscillation. A 0.1-µF bypass capacitor and 1-µF storage capacitor are recommended from REFIN directly to REFM.

The AGND and BDGND pins should be connected to a clean ground point. In all cases, this should be the analog ground. Avoid connections that are too close to the grounding point of a microcontroller or digital signal processor. If required, run a ground trace directly from the converter to the power supply entry point. The ideal layout consists of an analog ground plane dedicated to the converter and associated analog circuitry.

As with the AGND connections, +VA should be connected to a +5-V power supply plane that is separate from the connection for +VBD and digital logic until they are connected at the power entry point onto the PCB. Power to the ADC should be clean and well bypassed. A 0.1-µF ceramic bypass capacitor should be placed as close to the device as possible. See Table 6 for the placement of the capacitor. In addition to the 0.1-µF capacitor, a 1-µF capacitor is recommended. In some situations, additional bypassing may be required, such as a 100-µF electrolytic capacitor or even a Pi filter made up of inductors and capacitors; all designed to essentially low-pass filter the +5-V supply, thus removing the high frequency noise.

Table 6. Power Supply Decoupling Capacitor Placement

POWER SUPPLY PLANE SUPPLY PINS	CONVERTER ANALOG SIDE	CONVERTER DIGITAL SIDE
Pair of pins require a shortest path to decoupling capacitors	(9,10) (16,17) (20,21) (22,23) (26,27 or 25,26) (36,37)	(44,45)

TYPICAL CHARACTERISTICS

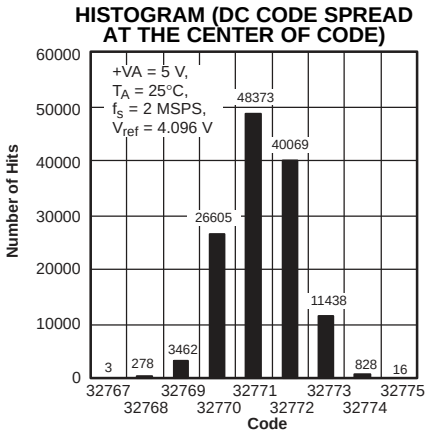


Figure 24.

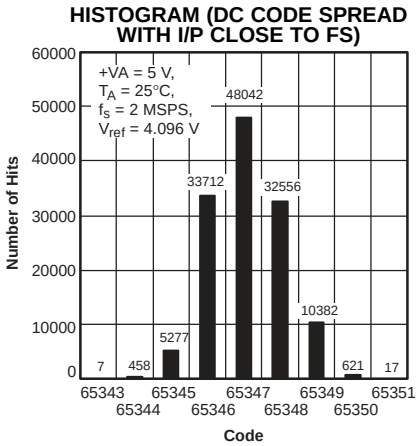


Figure 25.

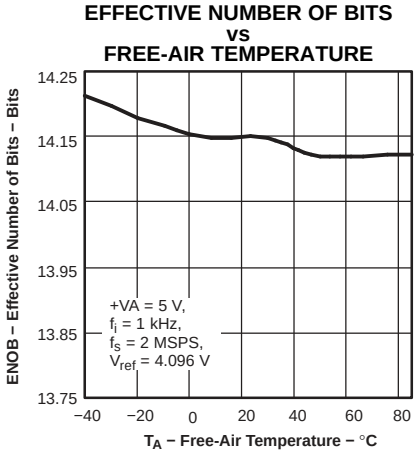


Figure 26.

TYPICAL CHARACTERISTICS (continued)

SIGNAL-TO-NOISE AND DISTORTION vs FREE-AIR TEMPERATURE

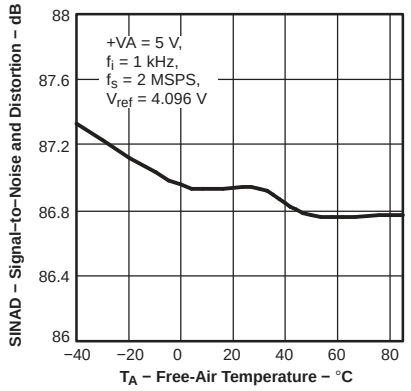


Figure 27.

SIGNAL-TO-NOISE RATIO vs FREE-AIR TEMPERATURE

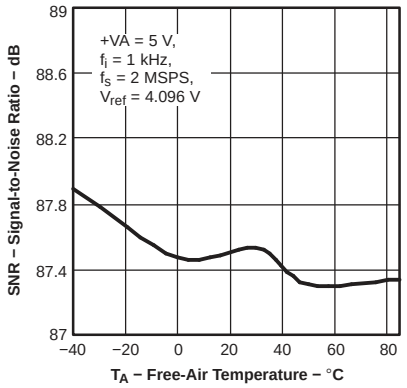


Figure 28.

SPURIOUS FREE DYNAMIC RANGE vs FREE-AIR TEMPERATURE

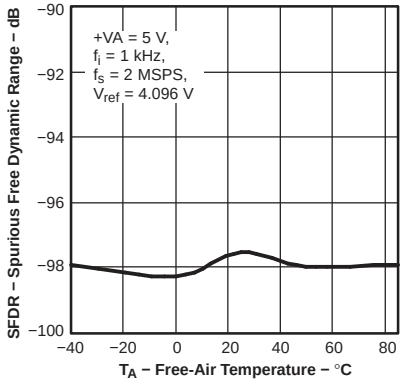


Figure 29.

TOTAL HARMONIC DISTORTION vs FREE-AIR TEMPERATURE

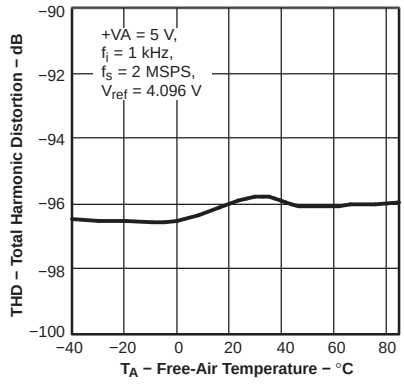


Figure 30.

EFFECTIVE NUMBER OF BITS vs INPUT FREQUENCY

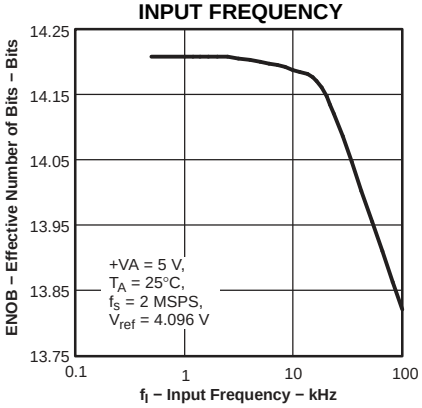


Figure 31.

SIGNAL-TO-NOISE AND DISTORTION vs INPUT FREQUENCY

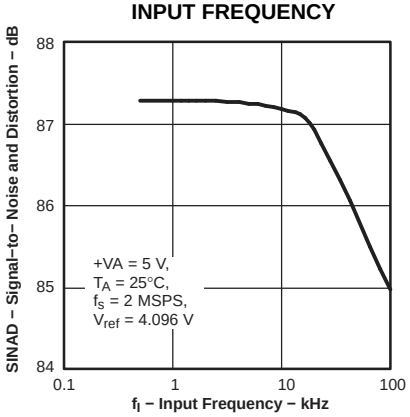


Figure 32.

SIGNAL-TO-NOISE RATIO vs INPUT FREQUENCY

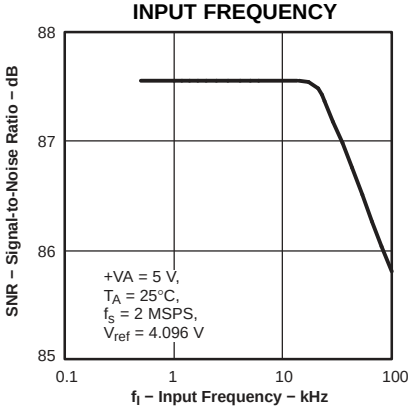


Figure 33.

SPURIOUS FREE DYNAMIC RANGE vs INPUT FREQUENCY

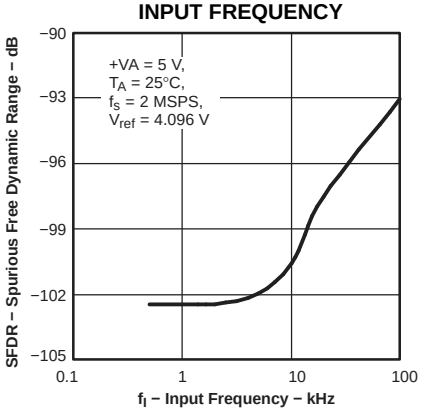


Figure 34.

TOTAL HARMONIC DISTORTION vs INPUT FREQUENCY

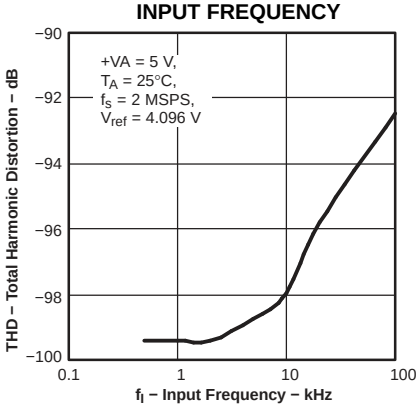


Figure 35.

TYPICAL CHARACTERISTICS (continued)

OFFSET ERROR vs SUPPLY VOLTAGE

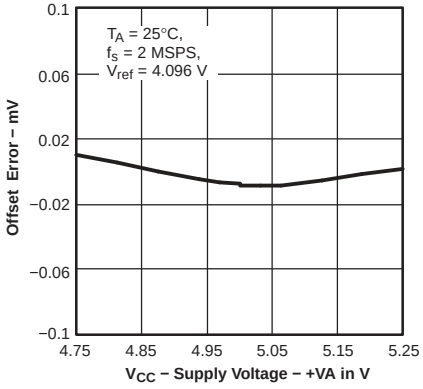


Figure 36.

GAIN ERROR vs SUPPLY VOLTAGE

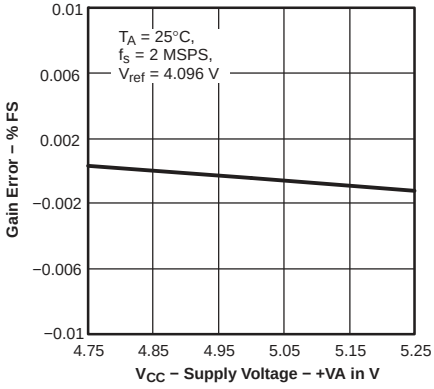


Figure 37.

OFFSET ERROR vs FREE-AIR TEMPERATURE

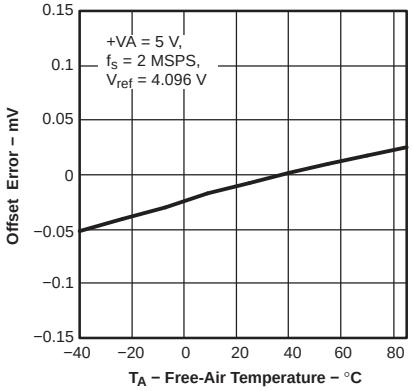


Figure 38.

GAIN ERROR vs FREE-AIR TEMPERATURE

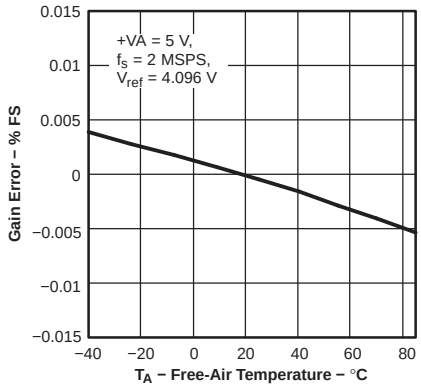


Figure 39.

POWER DISSIPATION vs SAMPLE RATE (NAP MODE)

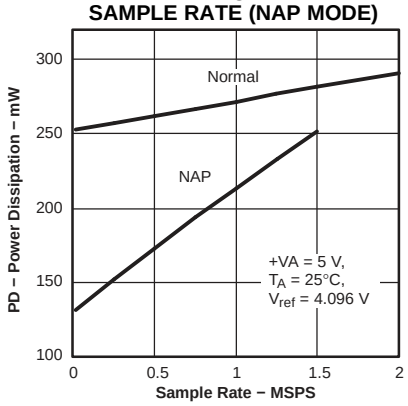


Figure 40.

POWER DISSIPATION vs SUPPLY VOLTAGE

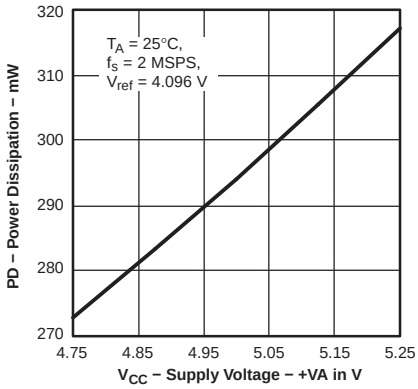


Figure 41.

POWER DISSIPATION vs FREE-AIR TEMPERATURE

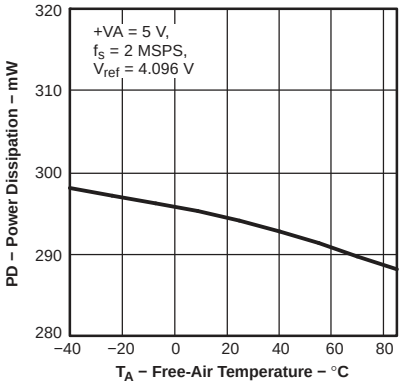


Figure 42.

DIFFERENTIAL NONLINEARITY vs FREE-AIR TEMPERATURE

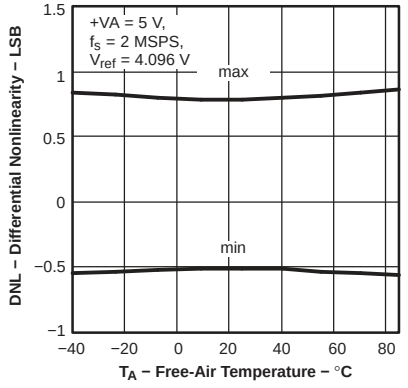


Figure 43.

INTEGRAL NONLINEARITY vs FREE-AIR TEMPERATURE

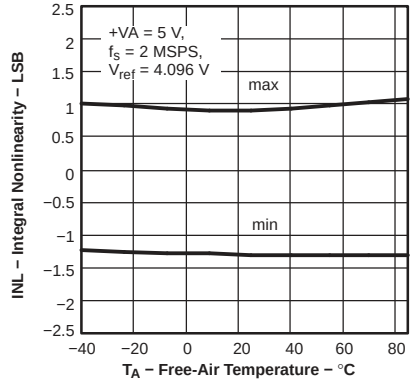


Figure 44.

TYPICAL CHARACTERISTICS (continued)

POSITIVE INTEGRAL NONLINEARITY DISTRIBUTION OVER 25 UNITS

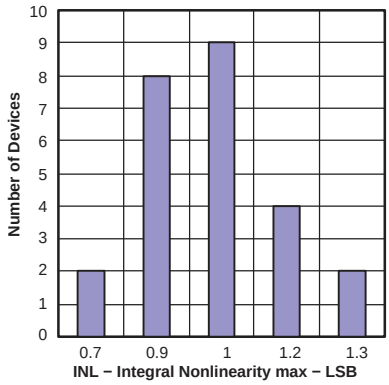


Figure 45.

NEGATIVE INTEGRAL NONLINEARITY DISTRIBUTION OVER 25 UNITS

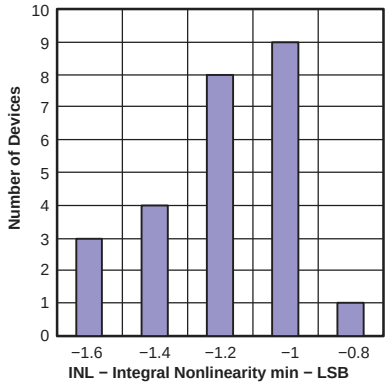


Figure 46.

INTERNAL REFERENCE OUTPUT vs SUPPLY VOLTAGE

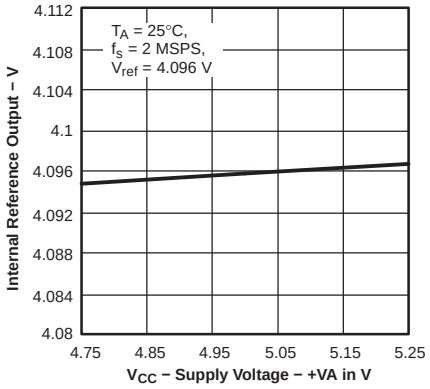


Figure 47.

INTERNAL REFERENCE OUTPUT vs FREE-AIR TEMPERATURE

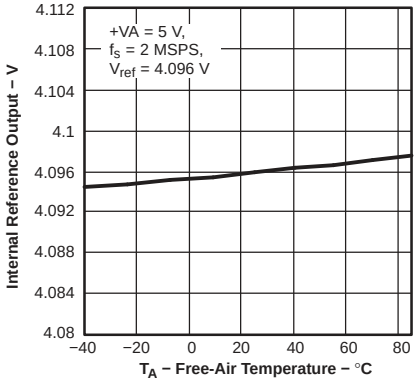


Figure 48.

PARAMETER MEASUREMENT INFORMATION

DRIVER

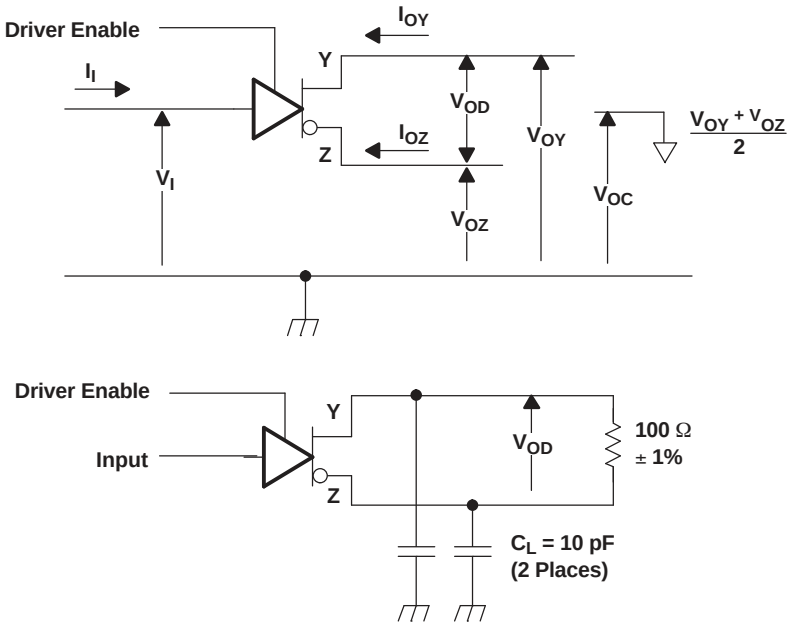


Figure 52. Driver Voltage and Current Definitions

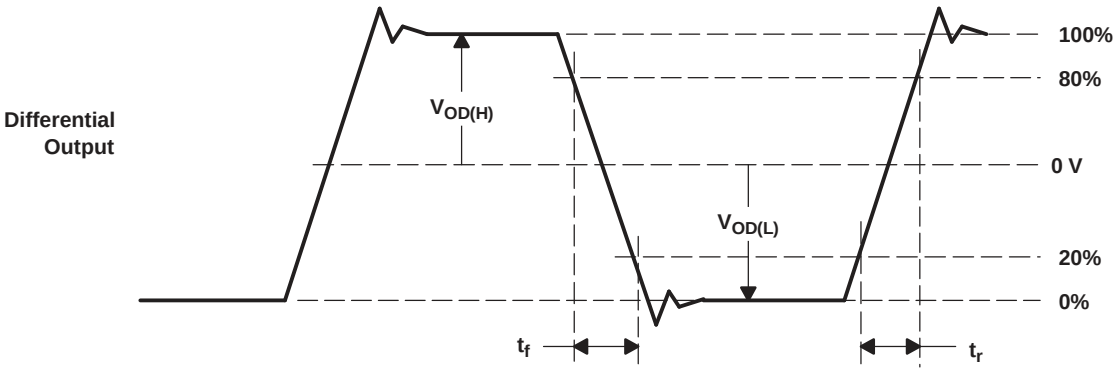


Figure 53. Timing and Voltage Definitions of the Differential Output Signal

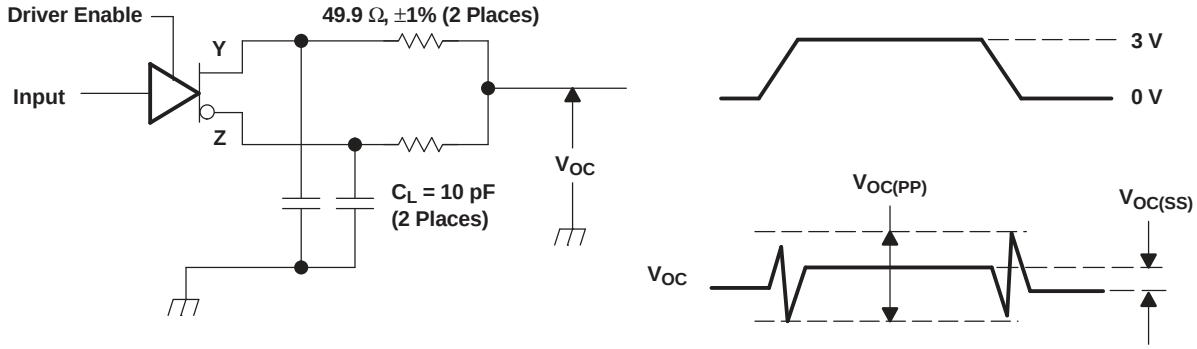


Figure 54. Test Circuit and Definitions for the Driver Common-Mode Output Voltage

PARAMETER MEASUREMENT INFORMATION (continued)

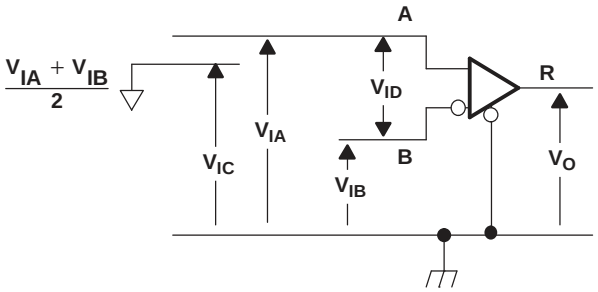


Figure 55. Receiver Voltage Definitions

REVISION HISTORY

Changes from Original (October 2005) to Revision A **Page**

- Added min spec for t_{d23} paramenter in the *Timing Requirements* table (used to be max spec for this parameter) 6
- Deleted max spec for t_{d23} paramenter in the *Timing Requirements* table (this spec is now the min spec for this parameter) 6

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
ADS8410IRGZT	NRND	VQFN	RGZ	48	250	Green (RoHS & no Sb/Br)	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	ADS8410I	

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

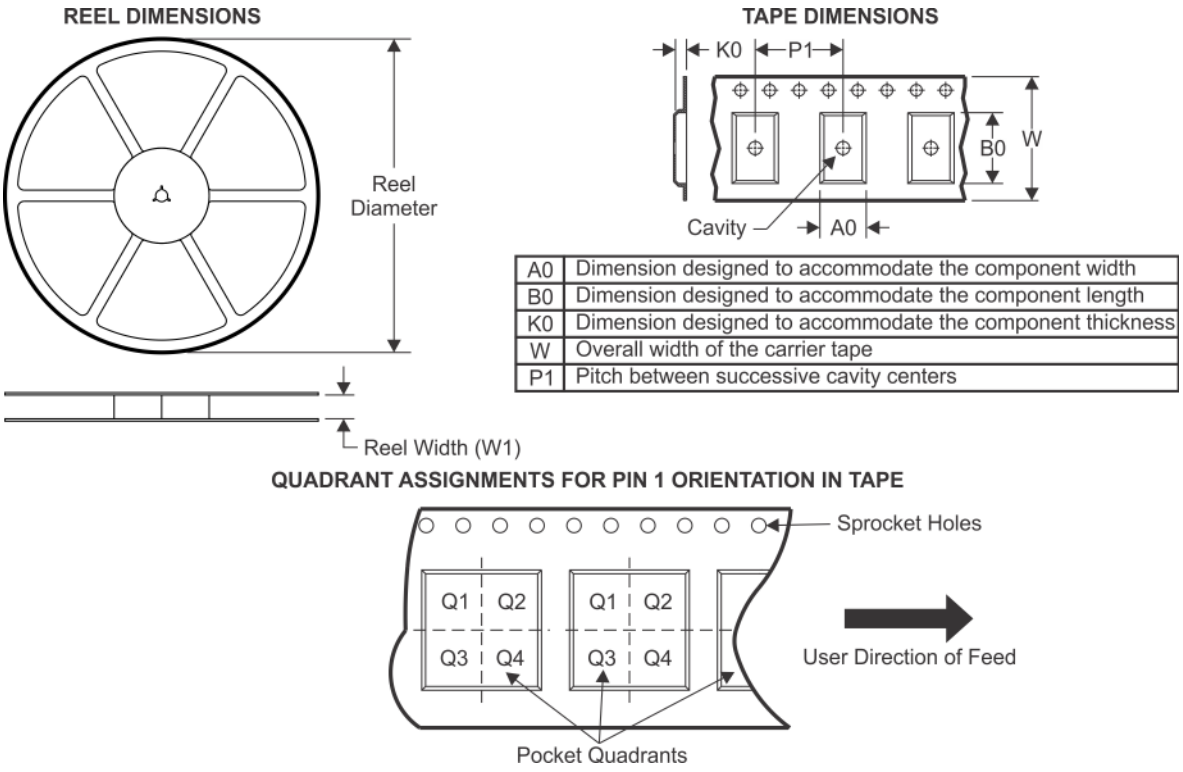
(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

TAPE AND REEL INFORMATION

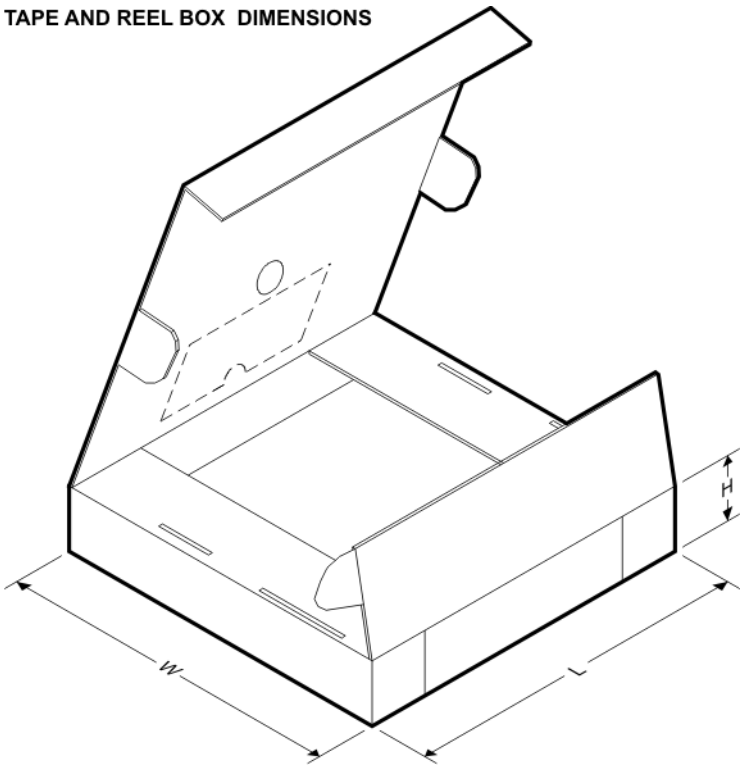


A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ADS8410IRGZT	VQFN	RGZ	48	250	180.0	16.4	7.3	7.3	1.5	12.0	16.0	Q2

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ADS8410IRGZT	VQFN	RGZ	48	250	213.0	191.0	55.0

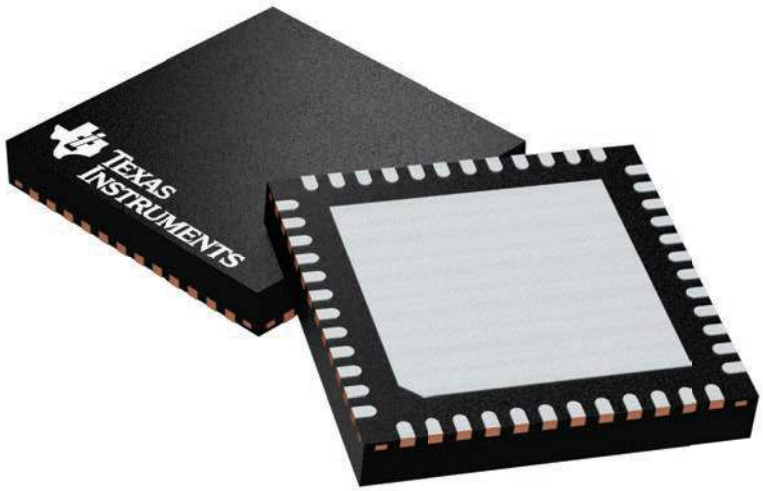
GENERIC PACKAGE VIEW

RGZ 48

VQFN - 1 mm max height

7 x 7, 0.5 mm pitch

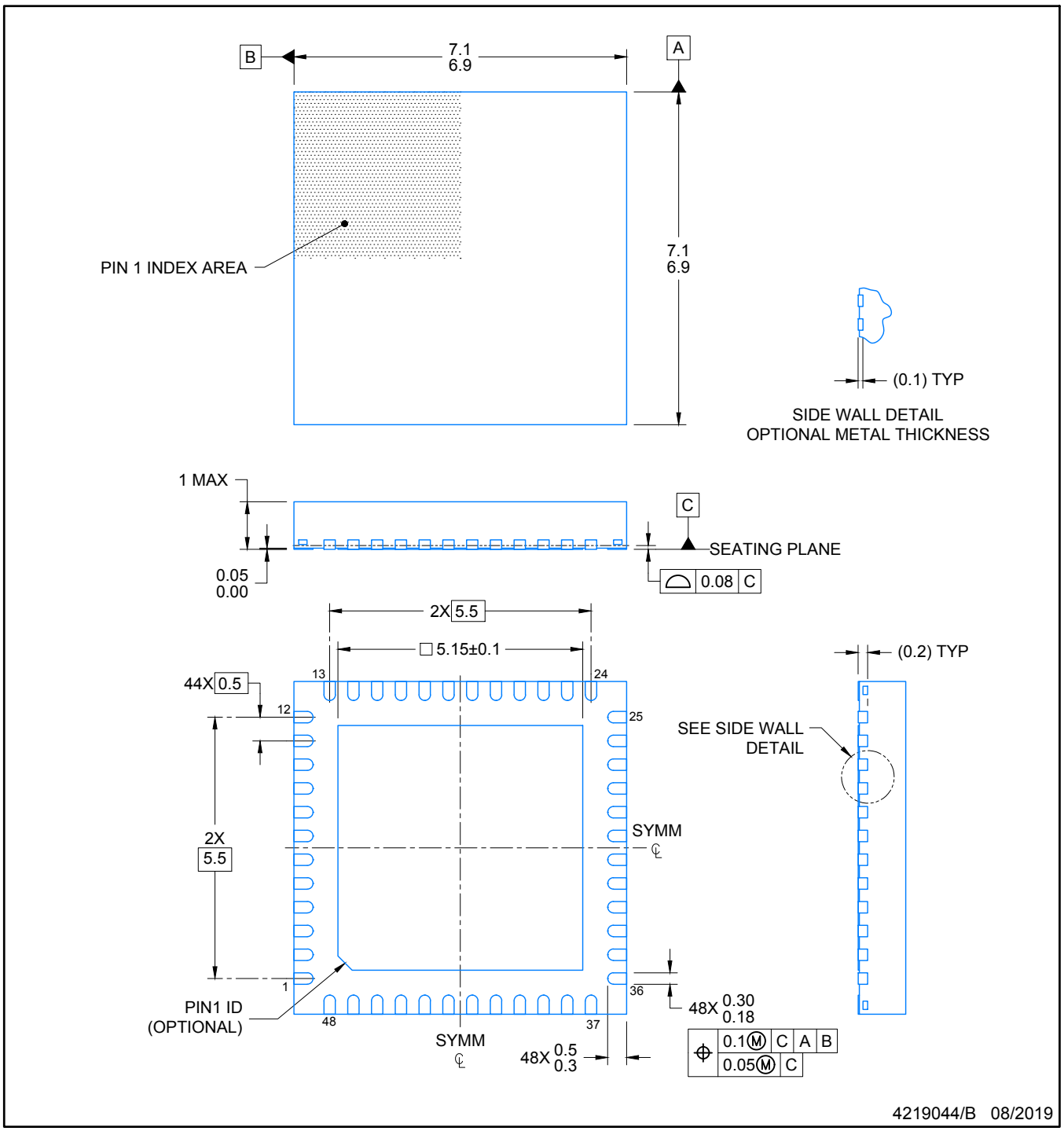
PLASTIC QUADFLAT PACK- NO LEAD



Images above are just a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.

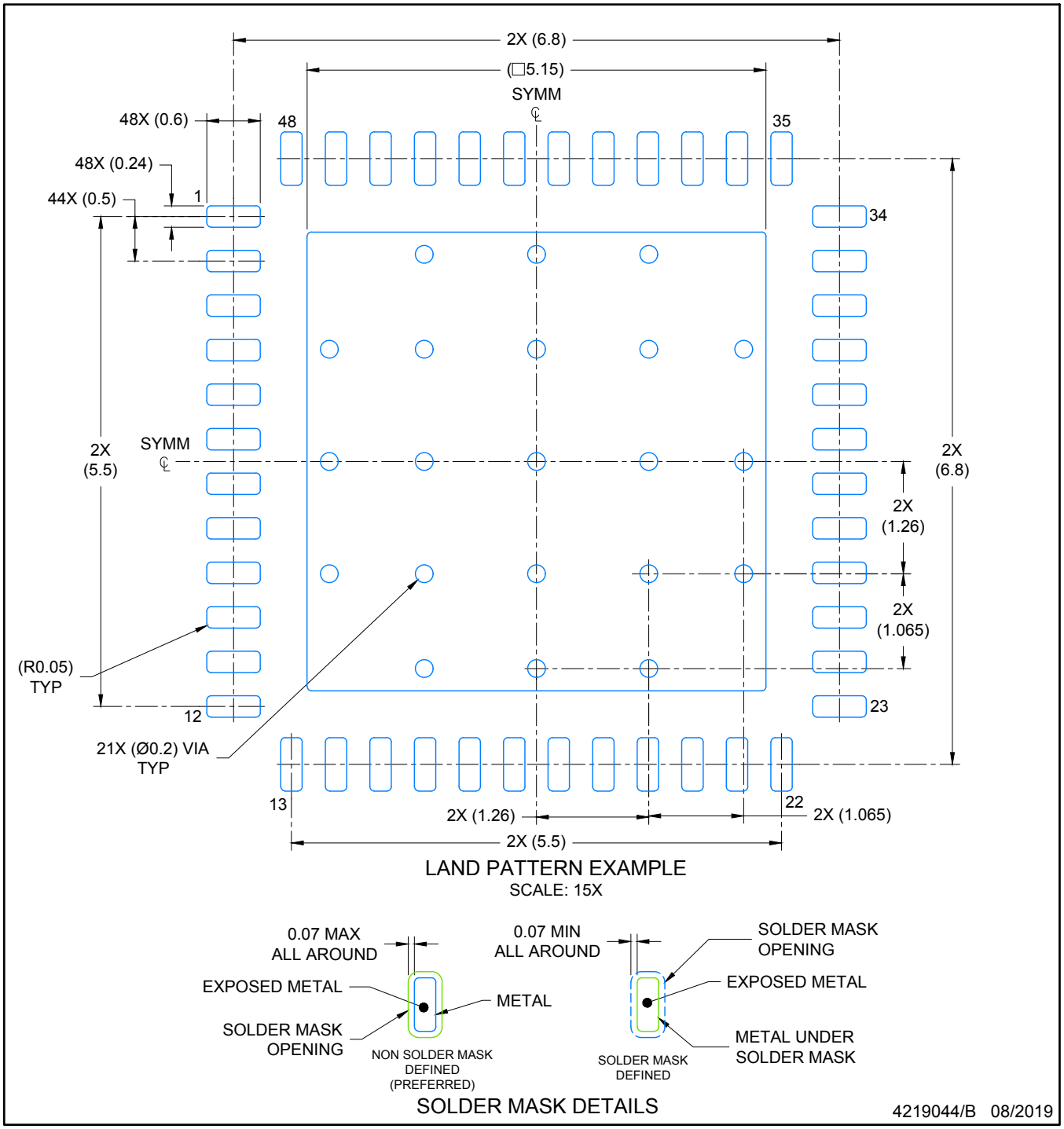
4224671/A

PLASTIC QUADFLAT PACK- NO LEAD



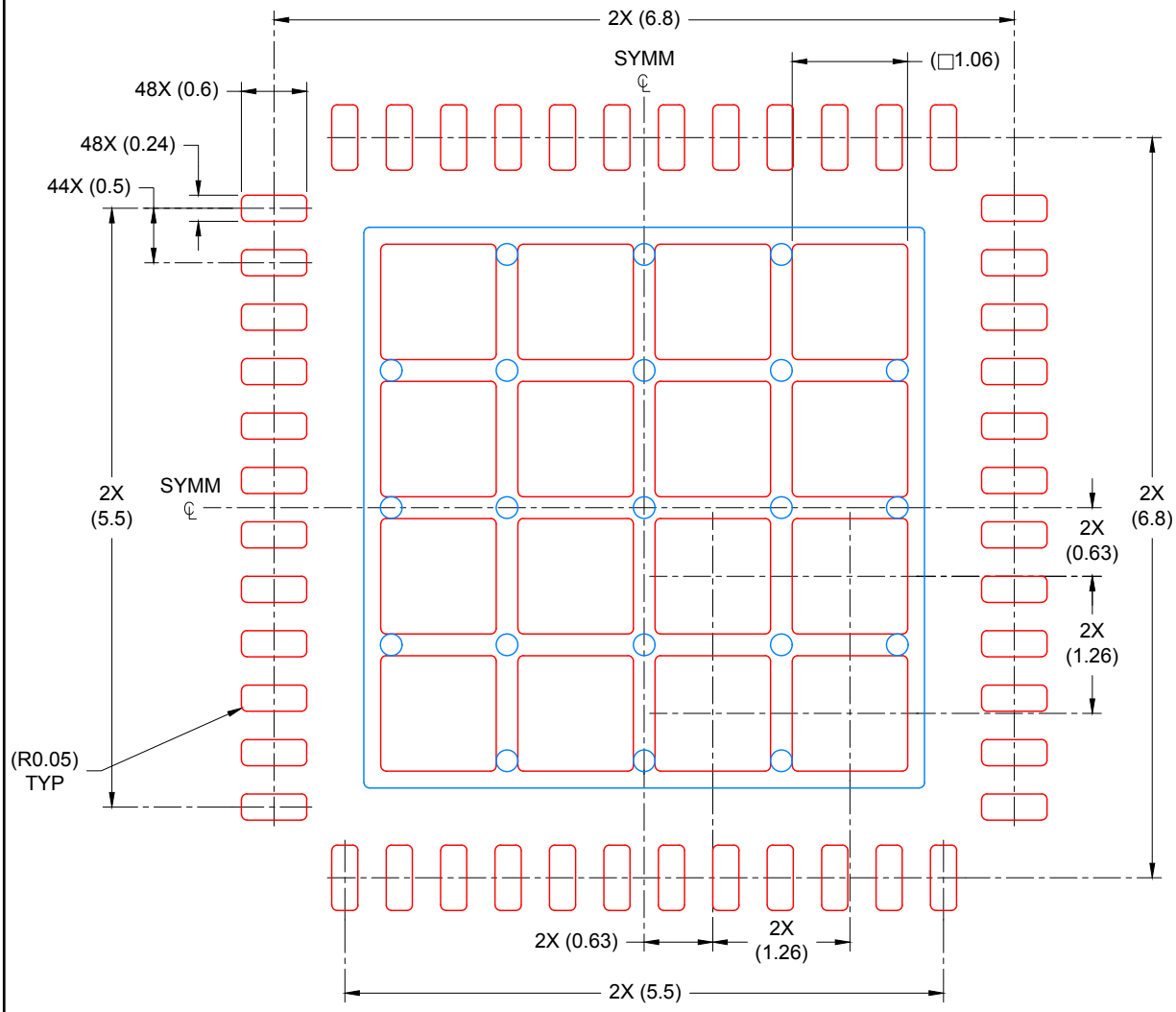
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.



NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD
67% PRINTED COVERAGE BY AREA
SCALE: 15X

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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