

# **MIC184**

# **Local/Remote Thermal Supervisor**

#### **Features**

- Measures Local and Remote Temperatures
- Pin and Software Backward Compatible to LM75
- 9-Bit Sigma-Delta ADC
- 2-Wire I<sup>2</sup>C/SMBus Compatible Interface
- Programmable Thermostatic Settings for Either Internal or External Zone
- Open-Drain Comparator/Interrupt Output Pin
- Interrupt Mask and Status Bits
- Low-Power Shutdown Mode
- Fail-Safe Response to Diode Faults
- 2.7V to 5.5V Power Supply Range
- Up to Eight Devices May Share the Same Bus
- 8-Lead SOIC and MSOP Packages

#### **Applications**

- Desktop, Server, and Notebook Computers
- Printers and Copiers
- Test and Measurement Equipment
- Consumer Electronics

#### **General Description**

The MIC184 is a versatile digital thermal supervisor capable of measuring temperature using either its own internal sensor or an inexpensive external sensor. A 2-wire serial interface is provided to allow communication with either  $I^2C$  or SMBus hosts. This device is a pin-for-pin and software compatible upgrade for the industry standard LM75.

Additional features include remote temperature measurement capability, and interrupt status and mask bits in the chip's configuration register for software polling. The open-drain interrupt output pin can be used as either an overtemperature alarm or thermostatic control signal. Three programmable address pins permit users to multi-drop up to eight devices along the 2-wire bus, allowing simple distributed temperature sensing networks. Superior performance, low power and small size makes the MIC184 an excellent choice for the most demanding thermal management applications.

#### **Package Types**



## **Typical Application Circuit**



# **Functional Block Diagram**



# <span id="page-2-6"></span>**1.0 ELECTRICAL CHARACTERISTICS**

#### **Absolute Maximum Ratings**



# **Operating Ratings †**



**Ü Notice:** Guaranteed by design over the operating temperature range. Not 100% production tested.

<span id="page-2-0"></span>Note 1: Devices are ESD sensitive. Handling precautions recommended. Human body model: 1.5k $\Omega$  in series with 100 pF. Machine model: 200 pF, no series resistance.

# **ELECTRICAL CHARACTERISTICS**



<span id="page-2-1"></span>**1:** Accuracy specification does not include quantization noise, which may be as great as  $\pm\sqrt{2\text{LSB}}$  ( $\pm\sqrt{4}^{\circ}\text{C}$ ).

- <span id="page-2-2"></span>**2:**  $T<sub>D</sub>$  is the temperature of the remote diode junction. Testing is performed using a single unit of one of the transistors listed in [Table 6-1](#page-18-0).
- <span id="page-2-3"></span>**3:** Current into the INT pin will result in self-heating of the MIC184. INT pin current should be minimized for best accuracy.
- <span id="page-2-5"></span>**4:** This is the decimal representation of a binary data value.
- <span id="page-2-4"></span>**5:** Guaranteed by design over the operating temperature range. Not 100% production tested.

# **ELECTRICAL CHARACTERISTICS (CONTINUED)**



**1:** Accuracy specification does not include quantization noise, which may be as great as ±1⁄2LSB (±1/4°C).

**2:**  $T<sub>D</sub>$  is the temperature of the remote diode junction. Testing is performed using a single unit of one of the transistors listed in Table 6-1.

**3:** Current into the INT pin will result in self-heating of the MIC184. INT pin current should be minimized for best accuracy.

- **4:** This is the decimal representation of a binary data value.
- **5:** Guaranteed by design over the operating temperature range. Not 100% production tested.

# **ELECTRICAL CHARACTERISTICS (CONTINUED)**



**1:** Accuracy specification does not include quantization noise, which may be as great as ±1⁄2LSB (±1⁄4°C).

**2:** T<sub>D</sub> is the temperature of the remote diode junction. Testing is performed using a single unit of one of the transistors listed in Table 6-1.

**3:** Current into the INT pin will result in self-heating of the MIC184. INT pin current should be minimized for best accuracy.

**4:** This is the decimal representation of a binary data value.

**5:** Guaranteed by design over the operating temperature range. Not 100% production tested.

# **TEMPERATURE SPECIFICATIONS ([Note 1](#page-5-0))**



<span id="page-5-0"></span>**Note 1:** The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air (i.e., T<sub>A</sub>, T<sub>J</sub>,  $\theta_{\sf JA}$ ). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum +150°C rating. Sustained junction temperatures above +150°C can impact the device reliability.

# **2.0 TYPICAL CHARACTERISTICS**

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.







*FIGURE 2-2: Remote Temperature Measurement Error.*





*Frequency.*

*FIGURE 2-4: Shutdown I<sub>DD</sub> vs.* 



*Temperature.*

*FIGURE 2-5: Shutdown I<sub>DD</sub> vs.* 



*Voltage.*



*FIGURE 2-7: Response to Immersion in 125*°*C Fluid Bath.*



*FIGURE 2-8: Measurement Error vs. PCB Leakage to +5V/+3.3V/GND.*



*FIGURE 2-9: Measurement Error vs. Capacitance on T1.*

# **3.0 PIN DESCRIPTIONS**

The descriptions of the pins are listed in [Table 3-1.](#page-8-0) Pin order and descriptions apply across both package types.



#### <span id="page-8-0"></span>**TABLE 3-1: MSOP AND SOIC PACKAGE PIN FUNCTION TABLE**

# **4.0 FUNCTIONAL DESCRIPTION**

#### **4.1 Pin Descriptions**

#### 4.1.1 VDD

Power supply input. See [Electrical Characteristics.](#page-2-6)

#### 4.1.2 GND

Ground return for all MIC184 functions.

#### 4.1.3 CLK

Clock input to the MIC184 from the two-wire serial bus. The clock signal is provided by the bus host and is shared by all devices on the bus.

#### 4.1.4 DATA

Serial data I/O pin that connects to the two-wire serial bus. DATA is bidirectional and has an open-drain output driver. An external pull-up resistor or current source somewhere in the system is necessary on this line. This line is shared by all devices on the bus.

#### 4.1.5 A2/T1, A1, A0

These inputs set the three least significant bits of the MIC184ís 7-bit client address. Each MIC184 will only respond to its own unique client address, allowing the use of up to eight MIC184s on a single bus. A match between the MIC184's address and the address specified in the serial bit stream must be made to initiate communication. A1 and A0 should be connected directly to VDD or ground. When A2/T1 is used as an address bit input, it should also be tied to VDD alternatively connect to a remote A2/T1 is used for temperature measurements, an off-chip diode junction must be connected between A2/T1 and ground. In this case, internal circuitry will detect A2 as logic low, leaving four possible client addresses. See [Temperature Measurement](#page-9-1) and Power-On sections for more information. A2/T1, A1, and A0 determine the client address as shown in [Table 4-1.](#page-9-0)

#### 4.1.6 INT

Temperature events are indicated to external circuitry via this output. INT may be configured as active-low or active-high by the host. Operation of the INT output is controlled by the MODE and POL bits in the MIC184's configuration register. See [Comparator and Interrupt](#page-13-1) [Modes.](#page-13-1) This output is open-drain and may be wire-ORed with other open-drain signals. Most systems will require a pull-up resistor or current source on this pin. If the IM bit in the configuration register is

set, it prevents the INT output from sinking current. In  $I_2C$  and SMBus systems, the IM bit is therefore an interrupt mask bit.

#### <span id="page-9-0"></span>**TABLE 4-1: MIC184 CLIENT ADDRESS SETTINGS**



#### <span id="page-9-1"></span>4.1.7 TEMPERATURE MEASUREMENT

The temperature-to-digital converter for both internal and external temperature data is built around a switched current source and a 9-bit analog-to-digital converter. The temperature is calculated by measuring the forward voltage of a diode junction at two different bias current levels. An internal multiplexer directs the current source's output to either an internal or external diode junction.

The MIC184 uses two's-complement data to represent temperatures. If the MSB of a temperature value is 0, the temperature is  $\geq 0^{\circ}$ C. If the MSB is 1, the temperature is  $< 0^\circ$ . More detail on this is given in [Temperature Data Format](#page-10-0) below. A temperature event results if the value in the temperature result register (TEMP) is greater than the value in the overtemperature set-point register (T\_SET), or if it is less than the value in the temperature hysteresis register (T\_HYST).

The value of the ZONE bit in the configuration register deter- mines whether readings are taken from the on-chip sensor or from the A2/T1 input. At power-up, the ZONE bit of the configuration register is set to zero. The MIC184 therefore monitors its internal temperature and compares the result against the contents of T\_SET and T\_HYST. Setting the ZONE bit in CONFIG will result in the MIC184 acquiring temperature data from an external diode connected to the A2/T1 pin. This diode may be embedded in an integrated circuit (such as a CPU, ASIC, or graphics processor), or it may be a diode-connected discrete transistor. Once the new

value is written to CONFIG, the A/D converter will begin a new conversion and return temperature data from the external zone. This data will be compared against T\_SET, T\_HYST, and the state of the Fault\_Queue (described below). The internal status bit (STS) and the INT output will then be updated accordingly. See the [Applications Information](#page-17-0) section for more details on switching between zones.

#### <span id="page-10-2"></span>4.1.8 DIODE FAULTS

The MIC184 is designed to respond in a fail-safe manner to hardware faults in the external sensing circuitry. If the connection to the external diode is lost, or the sense line (A2/T1) is shorted to  $V_{DD}$  or ground, the temperature data reported by the A/D converter will be forced to its full-scale value (+127.5°C). This will cause an overtemperature event to occur whenever T\_SET ≤ +127.0°C (0 1111 1110b). An interrupt will be generated if so enabled. The temperature reported for the external zone will remain 0 1111 1111b =  $+127.5^{\circ}$ C until the fault condition is cleared. This fault detection requires that the MIC184 complete the number of conversion cycles specified by Fault\_Queue. The MIC184 may therefore require one or more conversion cycles following power-on or a transition from shutdown to normal operation before reporting an external diode fault.

#### <span id="page-10-3"></span>4.1.9 SERIAL PORT OPERATION

The MIC184 uses standard SMBus WRITE\_BYTE, READ\_ BYTE, WRITE\_WORD, and READ\_WORD operations for communication with its host. The SMBus WRITE\_BYTE\_and WRITE\_WORD operations involve sending the device's client address (with the R/W bit low to signal a write operation), followed by a command byte and one or two data bytes. The SMBus READ BYTE operation is similar, but is a composite write and read operation: the host first sends the device's client address followed by the command byte, as in a write operation. A new "start" bit must then be sent to the MIC184, followed by a repeat of the client address with the R/W bit (LSB) set to the high (read) state. The data to be read from he MIC184 may then be clocked out. There is one exception to this rule: If the location latched in the pointer register from the last write operation is known to be correct (i.e., points to the desired register), then the "RECEIVE\_DATA" procedure may be used. To perform a RECEIVE\_DATA, the host sends an address byte to select the client MIC184, and then retrieves the appropriate number (one or two) of data bytes. [Figure 4-1](#page-11-0) through [Figure 4-5](#page-12-0) show the formats for these data read and data write procedures.

The command byte is 8 bits (1 byte) wide. This byte carries the address of the MIC184 register to be operated upon, and is stored in the MIC184's pointer register. The pointer register is a write-only register, which is implemented for backward compatibility to the National Semiconductor LM75 and similar devices.

The command byte (pointer register) values corresponding to the various MIC184 register addresses are shown in [Table 4-2.](#page-10-1) Command byte values other than  $0000 00$ XXb = 00h through 03h are reserved, and should not be used.

The CONFIG register is 8 bits (1 byte) wide. Therefore, communications with the CONFIG register will at a minimum require a READ\_BYTE, WRITE\_BYTE, or a RECEIVE\_BYTE.

The TEMP, T\_HYST, and T\_SET registers are logically nine bits wide. Note, though, that these registers are physically two bytes (one SMBus word) wide within the MIC184. Properly communicating with the MIC184 involves a 16-bit READ\_WORD or RECEIVE\_WORD from, or WRITE\_WORD to, these registers. This is a requirement of the  $I^2C/SMBus$  serial data protocols, which only allow data transfers to occur in multiples of eight bits.

#### <span id="page-10-1"></span>**TABLE 4-2: MIC184 REGISTER ADDRESSES**



#### <span id="page-10-0"></span>4.1.10 TEMPERATURE DATA FORMAT

The LSB of each 9-bit logical register represents 0.5°C. The values are in a two's complement format, wherein the most significant bit (D8) represents the sign:  $"0"$  for positive temperatures and "1" for negative temperatures. The seven least significant bits of each 16-bit physical register are undefined. Therefore, physical bits D6 through D0 of the data read from these registers must be masked off, and the resulting binary value right justified before using the data received. It is also possible to read only the first byte of any of these three registers, sacrificing 0.5°C of resolution in exchange for somewhat simpler data handling. However, all writes to the T SET and T HYST registers must be in the 16-bit WRITE\_WORD format. [Table 4-3](#page-11-1) shows examples of the data format used by the MIC184 for temperatures.

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<span id="page-11-1"></span>



<span id="page-11-0"></span>*FIGURE 4-1: WRITE\_BYTE Protocol.*



*FIGURE 4-2: READ\_BYTE Protocol.*



*FIGURE 4-3: Write\_WORD Protocol.*





<span id="page-12-0"></span>



<span id="page-12-2"></span>*FIGURE 4-6: A/D Converter Timing.*



<span id="page-12-1"></span>

*FIGURE 4-7: Responding to Interrupts.*

#### 4.1.11 A/D CONVERTER TIMING

Whenever the MIC184 is not in its low power shutdown mode, the internal A/D converter (ADC) attempts to make continuous conversions unless interrupted by a bus transaction accessing the MIC184. When the MIC184 is accessed, the conversion in progress will be halted, and the partial result discarded. When the access of the MIC184 is complete the ADC will begin a new conversion cycle, with results valid  $t_{\text{CONV}}$  after that. [Figure 4-5](#page-12-0) shows this behavior.  $t_{\text{CONV}}$  is twice as long for external conversions as it is for internal conversions. This allows the use of a filter capacitor on the A2/T1 input without a loss of accuracy due to the resulting longer settling times.

#### <span id="page-13-0"></span>4.1.12 POWER-ON

When power is initially applied, the MIC184's internal registers are set to default states which make the MIC184 completely backward compatible with the LM75. At this time, the levels on the address inputs A2, A1, and A0 are read to establish the device's client address. The MIC184's power-up default state can be summarized as follows:

- Normal-mode operation (MIC184 not in shutdown)
- ZONE is set to internal (on-chip temperature sensing)
- INT function is set to comparator mode
- INT output is set to active-low operation
- $\cdot$  Fault Queue depth = 1
- $\cdot$  Interrupts are enabled (IM = 0)
- $\cdot$  T\_SET = +80°C; T\_HYST = +75°C

In order to accommodate the use of A2/T1 as a dual-purpose input, there is a weak pull-down on A2/T1 that will attempt to sink ≈25 µA from the pin to ground for t<sub>POR</sub> following power- up of the MIC184. This allows the MIC184 to pull A2/T1 to a low state when a diode junction is connected from that pin to ground, and latch a zero as the A2 address value. If A2 is not to be used as a diode connection, it should be connected to  $V_{DD}$ or ground. Note that a fault in the external temperature sensor (if used) may not be reported until one or more conversion cycles have been completed following power-on. See [Diode Faults](#page-10-2).

#### 4.1.13 SHUTDOWN MODE

Setting the SHDN bit in the configuration register halts the otherwise continuous conversions by the A/D converter. The MIC184ís power consumption drops to 1 µA typical in shutdown mode. All registers may be read from, or written to, while in shutdown mode. Serial bus activity will slightly increase the MIC184's power consumption.

Entering shutdown mode will not affect the state of INT when the device is in comparator mode ( $MODE = 0$ ). However, If the device is shut down while in interrupt mode, the INT pin will be de-asserted and the internal latch (STS) holding the interrupt status will be cleared. Therefore, no interrupts will be generated while the MIC184 is in shutdown mode, and the interrupt status will not be retained. It is important to note, however, that the cause of the last temperature event will be retained in the MIC184. This is described further in [Comparator](#page-13-1) [and Interrupt Modes](#page-13-1). The diode fault detection mechanism requires one or more A/D conversion cycles to detect external sensor faults. (See [Diode](#page-10-2) [Faults](#page-10-2)) No diode faults will be detected while the device is in shutdown mode.

#### <span id="page-13-1"></span>4.1.14 COMPARATOR AND INTERRUPT **MODES**

Depending on the setting of the MODE bit in the configuration register, the INT output will behave either as an interrupt request signal or a thermostatic control signal. Thermostatic operation is known as comparator mode. The INT output is asserted whenever the measured temperature, as reported in the TEMP register, exceeds the threshold programmed in the T\_SET register for the number of conversions specified by Fault Queue (described below). In comparator mode, INT will remain asserted unless and until the measured temperature falls below the value in the T\_HYST register for Fault\_Queue conversions. No action on the part of the host is required for operation in comparator mode. Note that entering shutdown mode will not affect the state of INT when the device is in comparator mode.

In interrupt mode, once a temperature event has caused STS to be set, and the INT output to be asserted, they will not be automatically de-asserted when the measured temperature falls below T\_HYST. They can only be de-asserted by reading any of the MIC184's internal registers or by putting the device into SHUTDOWN mode. If the most recent temperature event was an overtemperature condition, STS will not be set again, and INT cannot be reasserted, until the device has detected that TEMP< T\_HYST. Similarly, if the most recent temperature event was an undertemperature condition, STS will in be set again, and INT cannot be reasserted, until the device has detected that TEMP > T\_SET. This keeps the internal logic of the MIC184 backward compatible with that of the LM75 and similar devices. There is a software override for this: while the MIC184 is operating in interrupt mode, the part can be unconditionally set to monitor for an overtemperature condition, regardless of what caused the last temperature event. This is done by clearing the MODE bit, and then immediately resetting it to 1. Following this sequence the next temperature event detected will be an overtemperature condition, regardless of whether the last temperature event was the result of an overtemperature or undertemperature condition.

In both modes, the MIC184 will be responsive to overtemperature events upon power up.

#### 4.1.15 FAULT\_QUEUE

A Fault\_Queue (programmable digital filter) is provided in the MIC184 to prevent false tripping due to thermal or electrical noise. Two bits, CONFIG[4:3], set the depth of Fault Queue. Fault Queue then determines the number of consecutive temperature events (TEMP > T\_SET or TEMP < T\_HYST) which must occur in order for the condition to be considered valid. As an example, assume the MIC184 is in comparator mode, and CONFIG[4:3] is programmed with 10 $_{\rm b}$ . Then the measured temperature would have to exceed T\_SET for four consecutive A/D conversions before INT would be asserted or the status bit set. Similarly, TEMP would have to be less than T HYST for four consecutive conversions before INT would be reset. Like any filter, the Fault\_Queue function also has the effect of delaying the detection of temperature events. In this example, it would take four times the  $t_{\text{CONV}}$  to detect a temperature event. The depth of Fault\_Queue vs. D[4:3] of the configuration register is shown in [Table 4-4.](#page-14-0)

<span id="page-14-0"></span>**TABLE 4-4: FAULT\_QUEUE DEPTH SETTINGS**

<b>CONFIG [4:3]</b>	<b>Fault_Queue Depth</b>
nn	1 conversion (default)
01	2 conversions
10	4 conversions
11	6 conversions

#### 4.1.16 HANDLING INTERRUPTS

The MIC184 may be either polled by the host, or request the host's attention via the INT pin. In the case of polled operation, the host periodically reads the contents of CONFIG to check the state of the status bit. The act of reading CONFIG clears the status bit, STS. If more than one event that sets the status bit occurs before the host polls the MIC184, only the fact that at least one such event has occurred will be apparent to the host.

If TEMP < T\_HYST or TEMP > T\_SET for Fault\_Queue con-versions, the status bit STS will be set in the CONFIG register. This action cannot be masked. However, a temperature event will only generate an interrupt signal on INT if interrupts from the MIC184 are enabled (IM =  $0$  and MODE = 1 in the configuration register). Reading any register following an interrupt will cause INT to be de-asserted, and will clear STS. The host should read the contents of the configuration register after receiving an interrupt to confirm that the MIC184 was the source of the interrupt. This is shown in [Figure 4-7](#page-12-1). As noted above, putting the device into shutdown mode will also de-assert INT and clear STS. Therefore, this usually should not be done before

completing the appropriate interrupt service routine(s). Since temperature-to-digital conversions continue while INT is asserted, it is possible that temperature could change between the MIC184's assertion of its INT output and the host's response to the interrupt. It is good practice when servicing interrupts for the host to read the current temperature to confirm that the condition that caused the interrupt still exists.

#### <span id="page-14-1"></span>4.1.17 INTERRUPT POLARITY SELECTION

The INT output can be programmed to behave as an active-low signal or an active-high signal. The default is active-low. INT polarity is selected by programming the appropriate value into the polarity bit (POL) in the CONFIG register. Clearing POL selects active-low interrupts; setting POL selects active-high interrupts. INT is an open-drain digital output and may be wire-ORed with other open-drain logic signals. Most applications will require a pull-up resistor on this pin.

Whether the CONFIG register's POL bit is set to provide a current-sinking (low) or high-Z (high) state at the INT pin when STS is high, writing a one to IM will put the INT pin into a high-Z state. This meets the requirement of an active-low interrupt for the SMBus, while making IM available as an INT-forcing bit for those applications which employ an active-high INT output (for example, software fan-control routines).

#### 4.1.18 LM75 COMPATIBILITY

The MIC184 can be used interchangeably with the LM75 in existing applications. The MIC184 offers several advantages over the LM75:

- Ability to monitor a second, remote temperature
- Interrupt masking capability
- Status bit for software polling routines
- Lower quiescent current
- Supports single-byte reads from 16-bit registers
- No "inadvertent 8-bit read" bus lock-up issues

The three MSB's of the configuration register (which power up as zeros) are used to access the MIC184's additional functions. These are reserved bits according to the LM75 specification and, for the LM75, must always be written as zeros. The MSB of the MIC184ís status register is a status flag that does not exist in the LM75. This bit will be set to one whenever an overtemperature event occurs. This bit would never be set by an LM75. Software should not depend on this bit being zero when using the MIC184 as an LM75 upgrade. If at power-up the measured temperature is higher than T\_SET, the status bit will be set following the first conversion by the A/D. See "[Applications](#page-17-0) Information" for a method by which host software can use this fact to differentiate between an MIC184 and an LM75.

# **5.0 REGISTER SET AND PROGRAMMERíS MODEL**



#### **TABLE 5-1: INTERNAL REGISTER SET**

<span id="page-15-1"></span><span id="page-15-0"></span>**Note 1:** TEMP will contain measured temperature data for the selected zone after the completion of one conversion.

**2:** After the first Fault\_Queue conversions are complete, the status bit will be set if TEMP < T\_HYST or TEMP > T\_SET.

#### **TABLE 5-2: DETAILED REGISTER DESCRIPTIONS**



<span id="page-15-2"></span>Power-Up Default Value: 0000 0000b = 00h.

- Not in shutdown mode
- Comparator mode
- $\cdot$  INT = active low
- $\cdot$  Fault\_Queue depth = 1
- Local temperature zone
- Interrupts enabled.

CONFIG Command Byte Address: 0000 0001b = 01h

Note that after the first Fault\_Queue conversions are completed, the status bit will be set if TEMP < T\_HYST or TEMP > T\_SET.





**Note 1:** The value in TEMP is 9 logical bits in width, but due to the conventions of I<sup>2</sup>C/SMBus, it is represented by 16 serial bits. System software should ignore undefined bits D[6:0]. See [Serial Port Operation](#page-10-3) and [Tem](#page-10-0)[perature Data Format](#page-10-0) for more details.

<span id="page-16-0"></span>Power-Up Default Value: 0 0000 0000b = 0°C. TEMP will contain measured temperature data for the selected zone after the completion of one conversion. TEMP Command Byte Address: 0000 0000b = 00h.

#### **TABLE 5-4: HYSTERESIS REGISTER**



<span id="page-16-1"></span>**Note 1:** The value in T\_HYST is 9 logical bits in width, but due to the conventions of I<sup>2</sup>C/SMBus, it is represented by 16 serial bits. System software should ignore undefined bits D[6:0] during register reads. Bits [6:0] should be set to zero during register writes. See [Serial Port Operation](#page-10-3) and [Temperature Data Format](#page-10-0) for more details.

T\_HYST Power-Up Default Value: 0 1001 0110b (+75°C) T\_HYST Command Byte Address: 0000 0010b = 02h

#### **TABLE 5-5: TEMPERATURE SETPOINT REGISTER**



**Note 1:** The value in T\_SET is 9 logical bits in width, but due to the conventions of I<sup>2</sup>C/SMBus, it is represented by 16 serial bits. System software should ignore undefined bits D[6:0] during register reads. Bits [6:0] should be set to zero during register writes. See [Serial Port Operation](#page-10-3) and [Temperature Data Format](#page-10-0) for more details.

T\_SET Power-Up Default Value: 0 1010 0000b (+80°C) T\_SET Command Byte Address: 0000 0011b = 03h

# <span id="page-17-0"></span>**6.0 APPLICATIONS INFORMATION**

#### **6.1 Switching Zones**

The recommended procedure for switching between the internal and external zones is as follows:

- 1. Disable interrupts (if used) by setting the IM bit in CONFIG.
- 2. Read the CONFIG register to:
	- a) Verify no masked interrupt was pending  $(D[7] = 0)$
	- b) Clear STS prior to switching zones
	- c) Hold the settings of CONFIG register for the current zone
- 3. Write the appropriate values to T\_SET and T\_HYST for the new zone.
- 4. Write to CONFIG as follows:
	- a) To toggle the ZONE bit  $(1 =$  remote,  $0 =$ internal)
	- b) If interrupts are being used, step 4 should also clear MODE
- 5. If interrupts are being used, MODE must then be set to 1 and IM reset to 0

At the conclusion of the serial bus transaction for step 4, the A/D converter will begin a conversion cycle using the new zone setting. The next conversion cycle completed after the serial bus transaction for step 5 will result in the state of the INT output being updated (if enabled) for the new zone. Generally the MIC184's A/D converter operates continuously, but it will be halted and reset each time the part recognizes its client address on the serial bus. Interrupted conversions will remain halted until the end of the host's communication with the MIC184. After the completion of step 5 and a delay of t<sub>CONV</sub> x Fault\_Queue, STS and INT will contain the results for the new zone. The above routine is extremely unlikely to miss a temperature event, as even one A/D conversion is typically much slower than the  $I<sup>2</sup>C/SMBus$  transactions that control the MIC184. See [Figure 4-6](#page-12-2).

Step 2(c) is recommended because the MIC184 has only one CONFIG register, corresponding to the active zone. In order to preserve data integrity for both zones, 2(c) allows the host to create a virtual CONFIG register for the inactive zone by dedicating one byte of memory to that purpose. Additional virtual registers may be created as needed by inserting additional reads as steps 2(d), 2(e), etc. These could for example correspond to the values in T\_SET and T\_HYST immediately prior to switching zones. Steps 4(b) and 5 ensure that the MIC184 will enter the new zone searching for an overtemperature event.

#### **6.2 Identifying an MIC184 by Software Test**

The MIC184 and the LM75 each have an eight-bit CONFIG register. In LM75-type parts, no more than seven of the eight bits of this register are used, and at least one bit (the MSB) will always return a zero. The MIC184 uses all eight bits of the CONFIG register: the MSB is the part's status bit (STS). A simple test by which the host can determine whether a system has an MIC184 installed, or is using a legacy LM75-type device, is to create a situation which will set the MSB in the MIC184ís CONFIG register and then determine if the MSB is in fact set. Two examples of how this can be done are outlined below. The first is interrupt-driven, the second uses software polling. Note that both procedures generate one or more spurious interrupts. The code for these tests should therefore temporarily disable any affected interrupt routines.

{START Interrupt-Driven Test and Initialization Routine}

- 1. Disable the host's overtemperature and undertemperature interrupt handling routine. Redirect interrupts from the part under test to a handler for the interrupt that will be generated in steps (4) and (7) of this routine.
- 2. Write 0000 0010b (02h) to the CONFIG register. (The assumption is made that the host is an  $I^2C$ or SMBus part, and therefore responds to an active-low interrupt request.)
- 3. Write 1100 1000 1000 0000b = C880h to T\_SET and T\_HYST. This corresponds to -55.5°C.
- 4. When the part has finished its first A/D conversion, INT will be asserted.
- 5. Read out the contents of the CONFIG register:
	- a) If the part is an MIC184, the MSB will be set to one (CONFIG = 1000 0010b = 82h).
	- b) If the part is a conventional LM75-type part, the MSB will be zero (CONFIG = 0000  $0010b = 02h$ ).
- 6. Write 0111 1111 1000 0000b = 7F80h to T\_SET and T\_HYST. This corresponds to +127.5°C.
- 7. When the part has finished its next A/D conversion, INT will be asserted a second time.
- 8. Read CONFIG again, to clear the interrupt request from step (7). This will also clear STS, if the part under test is an MIC184.
- 9. Based on the results of the test in step (4), do the following within 50 ms total:
	- a) Set the CONFIG register as required.
	- b) Load T\_HYST with its operational value.
	- c) Load T\_SET with its operational value.
	- d) Set the host's interrupt handling routine back to overtemperature and undertemperature mode.

{END}

{START Polling-Based Test and Initialization Routine}

- 1. Temporarily disable the host's interrupt input from the device under test.
- 2. Write 0000 0010b (02h) to the CONFIG register.
- 3. Write 1100 1000 1000 0000b = C880h to T\_SET and  $T_HYST$ . This corresponds to  $-55.5^{\circ}$ C.
- 4. Wait  $t_{\text{conv}}$  (160 ms max.) for the part to finish at least one A/D conversion.
- 5. Read the contents of the CONFIG register:
	- a) If the part is an MIC184, the MSB will be set to one (CONFIG = 82h).
	- b) If the part is a conventional LM75-type part, the MSB will be zero (CONFIG = 02h).
- 6. Write 0111 1111 1000 0000b = 7F80h to T\_SET and T\_HYST. This corresponds to +127.5°C.
- 7. Wait an additional  $t_{conv}$  for the part to finish a second conversion.
- 8. Read CONFIG again, to clear the interrupt request from step (7). This will also clear STS, if the part under test is an MIC184.
- 9. Based on the results of the test in step (4), do the following four steps within 50 ms total:
	- a) Set the CONFIG register as required.
	- b) Load T\_HYST with its operational value.
	- c) Load T\_SET with its operational value.
	- d) Re-enable the host's interrupt handling input from the part under test.

#### {END}

These routines force the device under test to generate an overtemperature fault (steps 3 and 4), followed by an undertemperature fault (steps 6 through 8). This sequence causes the device under test to exit the routine prepared to respond to an overtemperature condition. If being immediately prepared to detect an undertemperature condition upon exit is desired, swap steps 3 and 6 in each routine.

#### **6.3 Remote Diode Selection**

Most small-signal PNP transistors with characteristics similar to the JEDEC 2N3906 will perform well as remote temperature sensors. [Table 6-1](#page-18-0) lists several examples of such parts. These parts have been tested and marked with a bullet for use with the MIC184.

#### <span id="page-18-0"></span>**TABLE 6-1: TRANSISTORS SUITABLE FOR REMOTE TEMPERATURE SENSING USE**



#### **6.4 Minimizing Errors**

#### 6.4.1 SELF-HEATING

One concern when using a part with the temperature accuracy and resolution of the MIC184 is to avoid errors induced by self-heating ( $V_{DD} \times I_{DD}$ ). In order to understand what level of error this might represent, and how to reduce that error, the dissipation in the MIC184 must be calculated, and its effects examined as a temperature error.

In most applications, the INT output will be low for at most a few milliseconds before the host sets it back to the high state, making its duty cycle low enough that its contribution to self- heating of the MIC184 is negligible. Similarly, the DATA pin will in all likelihood have a duty cycle of substantially below 25% in the low state. These considerations, combined with more typical device and application parameters, allow the following calculation of typical device self-heating in interrupt-mode:

#### **EQUATION 6-1:**

 $P_D = (I_{DD(typ)} 3.3 V + 25\% I_{OL(data)} 0.3 V$  $+1\%$ **I**<sub>OL(int</sub>)0.3*V*)

#### **EQUATION 6-2:**

 $P_D = (0.3mA \times 3.3V + 25\% \times 1.5mA \times 0.3V)$  $+1\% \times 1.5mA \times 0.3V$ 

#### **EQUATION 6-3:**

 $\Delta T_J = 1.11$  *mW*  $\times$  206° *C*/W

**Note:**  $\Delta T_J$  relative to  $T_A$  is 0.23°C.

If the part is to be used in comparator mode, calculations similar to those shown above (accounting for the expected value and duty cycle of  $I_{OL(int)}$ ) will give a good estimate of the device's self-heating error.

In any application, the best test is to verify performance against calculation in the final application environment. This is especially true when dealing with systems for which some of the thermal data, (for example, PC board thermal conductivity and/or ambient temperature), may be poorly defined or unavailable except by empirical means.

#### **6.5 Series Resistance**

The operation of the MIC184 depends upon sensing the  $\Delta_{VCB-E}$  of a diode-connected PNP transistor ("diode") at two different current levels. For remote temperature measurements, this is done using an external diode connected between A2/T1 and ground.

Because this technique relies upon measuring the relatively small voltage difference resulting from two levels of current through the external diode, any resistance in series with the external diode will cause an error in the temperature reading from the MIC184. A good rule of thumb is this: for each ohm in series with the external transistor, there will be a 0.9°C error in the MIC184's temperature measurement. It is not difficult to keep the series resistance well below an ohm (typically  $0.1\Omega$ ), so in most systems this will not be an issue.

#### **6.6 Filter Capacitor Selection**

When using a remote diode for temperature sensing, it is sometimes desirable to use a filter capacitor between the A2/T1 and GND pins of the MIC184. The use of this capacitor is recommended in environments with a significant high frequency noise (such as digital switching noise), or if long wires are used to connect to the remote diode. The maximum recommended total capacitance from the A2/T1 pin to GND is 2700 pF. This usually suggests the use of a 2200 pF NP0 or C0G ceramic capacitor with a 10% tolerance.

If the remote diode is to be at a distance of more than 6" ~ 12" from the MIC184, using a shielded cable (solid foil shield microphone cable is a good choice) for the connections to the diode can significantly help reduce noise pickup. Remember to subtract the cable's conductor-to-shield capacitance from the 2700 pF maximum total capacitance.

## **6.7 Layout Considerations**

#### **Local Mode Only Applications**

If the MIC184 is not going to be used with an external diode, the best layout is one that keeps it thermally coupled to the subsystem(s) whose temperature it must monitor, while avoiding any strong sources of EMI, RFI, or electrostatically coupled noise. Two of the most common examples of such sources are switching power supply transformers and CRTs.

#### **Remote Mode Application**

- 1. If the remote sensing capability of the MIC184 will be used in an application, place the MIC184 as close to the remote diode as possible, while taking care to avoid severe noise sources (high frequency power transformers, CRTs, memory and data buses, and the like).
- 2. Since any conductance from the various voltages on the PC board and the A2/T1 pin can induce serious errors, it is good practice to guard the remote diode's emitter trace with a pair of ground traces. These ground traces should be returned to the MIC184's own ground pin. They should not be grounded at any other part of their run. However, it is highly desirable to use these guard traces to carry the diode's own ground return back to the ground pin of the MIC184, thereby providing a Kelvin connection for the base of the diode. See [Figure 6-1](#page-20-0).
- 3. When using the MIC184 to sense the temperature of a processor or other device which has an integral on-board "diode" (e.g., Intel's Pentium<sup>®</sup> III), connect the emitter and base of the remote sensor to the MIC184 using the guard traces and Kelvin return shown in [Figure 6-1](#page-20-0). The collector of the remote "diode" is inaccessible to the user on these types of chips. To allow for this, the MIC184 has superb rejection of noise appearing from collector to GND, as long as the base to ground connection is relatively quiet.
- 4. Due to the small currents involved in the measurement of the remote diode's  $\Delta V_{BF}$ , it is important to adequately clean the PC board after soldering. This is most likely to show up as an issue in some situations where water-soluble soldering fluxes are used.
- 5. In general, wider traces for the ground and A2/T1 pins will help reduce susceptibility to radiated noise (wider traces are less inductive). Use trace widths and spacing of 10 mm wherever possible. Wherever possible, place a ground plane under the MIC184, and under the connections from the MIC184 to the remote diode. This will help guard against stray noise pickup.
- 6. Always place a good quality  $V_{DD}$  bypass capacitor directly adjacent to, or underneath, the MIC184. This part should be a 0.1 µF ceramic capacitor. Surface-mount parts provide the best bypassing because of their low inductance.
- 7. When the MIC184 is being powered from particularly noisy power supplies, or from supplies which may have sudden high-amplitude spikes appearing on them, it can be helpful to add additional power supply filtering. This should be implemented as a 100Ω resistor in series with the part's  $V_{DD}$  pin, and a 4.7 µF, 6.3V electrolytic capacitor from  $V_{DD}$  to GND. See [Figure 6-2.](#page-20-1)



<span id="page-20-0"></span>

*FIGURE 6-1: Guard Traces/Kelvin Ground Returns.*



<span id="page-20-1"></span>**FIGURE 6-2:** *V<sub>DD</sub>* Decoupling for Very Noisy Supplies.

# **7.0 PACKAGING INFORMATION**

# **7.1 Package Marking Information**







#### **8-Lead MSOP Package Outline and Recommended Land Pattern**

#### **8-Lead SOIC Package Outline and Recommended Land Pattern**



# **APPENDIX A: REVISION HISTORY**

### **Revision A (November 2020)**

- Converted Micrel data sheet MIC184 to Microchip data sheet DS20006457A.
- Minor grammatical corrections throughout.

# **MIC184**

**NOTES:**

# **PRODUCT IDENTIFICATION SYSTEM**

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# **MIC184**

**NOTES:**

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