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# LM158/LM258/LM358/LM2904 Low Power Dual Operational Amplifiers

## **General Description**

The LM158 series consists of two independent, high gain, internally frequency compensated operational amplifiers which were designed specifically to operate from a single power supply over a wide range of voltages. Operation from split power supplies is also possible and the low power supply current drain is independent of the magnitude of the power supply voltage.

Application areas include transducer amplifiers, dc gain blocks and all the conventional op amp circuits which now can be more easily implemented in single power supply systems. For example, the LM158 series can be directly operated off of the standard  $\pm 5V$  power supply voltage which is used in digital systems and will easily provide the required interface electronics without requiring the additional  $\pm 15V$  power supplies.

The LM358 and LM2904 are available in a chip sized package (8-Bump micro SMD) using National's micro SMD package technology.

## **Unique Characteristics**

- In the linear mode the input common-mode voltage range includes ground and the output voltage can also swing to ground, even though operated from only a single power supply voltage.
- The unity gain cross frequency is temperature compensated.
- The input bias current is also temperature compensated.

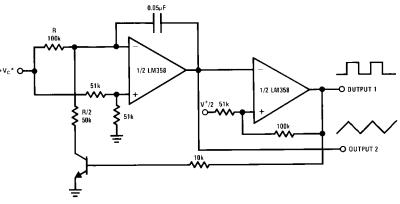
## **Advantages**

- Two internally compensated op amps
- Eliminates need for dual supplies
- Allows direct sensing near GND and V<sub>OUT</sub> also goes to GND
- Compatible with all forms of logic
- Power drain suitable for battery operation

## **Features**

- Available in 8-Bump micro SMD chip sized package, (See AN-1112)
- Internally frequency compensated for unity gain
- Large dc voltage gain: 100 dB
- Wide bandwidth (unity gain): 1 MHz (temperature compensated)
- Wide power supply range:
  - Single supply: 3V to 32V
  - or dual supplies: ±1.5V to ±16V
- Very low supply current drain (500 µA)—essentially independent of supply voltage
- Low input offset voltage: 2 mV
- Input common-mode voltage range includes ground
- Differential input voltage range equal to the power supply voltage
- Large output voltage swing

## **Voltage Controlled Oscillator (VCO)**



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## **Absolute Maximum Ratings** (Note 9)

Distributors for availability and specifications.

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/

	LM158/LM258/LM358	LM2904
	LM158A/LM258A/LM358A	
Supply Voltage, V <sup>+</sup>	32V	26V
Differential Input Voltage	32V	26V
Input Voltage	-0.3V to +32V	-0.3V to $+26V$
Power Dissipation (Note 1)		
Molded DIP	830 mW	830 mW
Metal Can	550 mW	
Small Outline Package (M)	530 mW	530 mW
micro SMD	435mW	
Output Short-Circuit to GND		
(One Amplifier) (Note 2)		
$V^+ \le 15V$ and $T_A = 25^{\circ}C$	Continuous	Continuous
Input Current ( $V_{IN} < -0.3V$ ) (Note 3)	50 mA	50 mA
Operating Temperature Range		
LM358	0°C to +70°C	-40°C to +85°C
LM258	−25°C to +85°C	
LM158	−55°C to +125°C	
Storage Temperature Range	−65°C to +150°C	-65°C to +150°C
Lead Temperature, DIP		
(Soldering, 10 seconds)	260°C	260°C
Lead Temperature, Metal Can		
(Soldering, 10 seconds)	300°C	300°C
Soldering Information		
Dual-In-Line Package		
Soldering (10 seconds)	260°C	260°C
Small Outline Package		
Vapor Phase (60 seconds)	215°C	215°C
Infrared (15 seconds)	220°C	220°C
See AN-450 "Surface Mounting Methods and Their I	Effect on Product Reliability" for other methods	of soldering
surface mount devices.		
ESD Tolerance (Note 10)	250V	250V

## **Electrical Characteristics**

 $V^+ = +5.0V$ , unless otherwise stated

Parameter	Conditions	LM158A			LM358A			LM	.M258	Units	
		Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	
Input Offset Voltage	(Note 5), T <sub>A</sub> = 25°C		1	2		2	3		2	5	mV
Input Bias Current	$I_{IN(+)}$ or $I_{IN(-)}$ , $T_A = 25^{\circ}C$ ,		20	50		45	100		45	150	nA
	V <sub>CM</sub> = 0V, (Note 6)										
Input Offset Current	$I_{IN(+)} - I_{IN(-)}, V_{CM} = 0V, T_A = 25^{\circ}C$		2	10		5	30		3	30	nA
Input Common-Mode	V <sup>+</sup> = 30V, (Note 7)	0		V+-1.5	0		V+-1.5	0		V+-1.5	V
Voltage Range	(LM2904, $V^+ = 26V$ ), $T_A = 25$ °C										
Supply Current	Over Full Temperature Range										
	R <sub>L</sub> = ∞ on All Op Amps										
	$V^{+} = 30V \text{ (LM2904 } V^{+} = 26V)$		1	2		1	2		1	2	mA
	V <sup>+</sup> = 5V		0.5	1.2		0.5	1.2		0.5	1.2	mA

## **Electrical Characteristics**

 $V^+ = +5.0V$ , unless otherwise stated

Parameter	Conditions		LM358			Units		
		Min	Тур	Max	Min	Тур	Max	
Input Offset Voltage	(Note 5) , T <sub>A</sub> = 25°C		2	7		2	7	mV
Input Bias Current	$I_{IN(+)}$ or $I_{IN(-)}$ , $T_A = 25^{\circ}C$ ,		45	250		45	250	nA
	V <sub>CM</sub> = 0V, (Note 6)							
Input Offset Current	$I_{IN(+)} - I_{IN(-)}, V_{CM} = 0V, T_A = 25^{\circ}C$		5	50		5	50	nA
Input Common-Mode	V <sup>+</sup> = 30V, (Note 7)	0		V+-1.5	0		V+-1.5	V
Voltage Range	(LM2904, $V^+ = 26V$ ), $T_A = 25^{\circ}C$							
Supply Current	Over Full Temperature Range							
	R <sub>L</sub> = ∞ on All Op Amps							
	V <sup>+</sup> = 30V (LM2904 V <sup>+</sup> = 26V)		1	2		1	2	mA
	V <sup>+</sup> = 5V		0.5	1.2		0.5	1.2	mA

## **Electrical Characteristics**

 $V^+ = +5.0V$ , (Note 4), unless otherwise stated

Paramete	<b></b>	Conditions		LM158	Α	ı	LM358	A	LM <sup>-</sup>	Units		
		Conditions		Тур	Max	Min	Тур	Max	Min	Тур	Max	
Large Signal Vol	tage	V <sup>+</sup> = 15V, T <sub>A</sub> = 25°C,										
Gain		$R_L \ge 2 \text{ k}\Omega$ , (For $V_O = 1V$	50	100	0	25	100		50	100		V/mV
		to 11V)										
Common-Mode		$T_A = 25^{\circ}C$ ,	70	85		65	85		70	0.E		dB
Rejection Ratio		$V_{CM} = 0V \text{ to } V^{+}-1.5V$	10	00		05	00		70	85		uБ
Power Supply		V <sup>+</sup> = 5V to 30V										
Rejection Ratio		(LM2904, V <sup>+</sup> = 5V	65	100		65	100		65	100		dB
		to 26V), T <sub>A</sub> = 25°C										
Amplifier-to-Amp	lifier	$f = 1 \text{ kHz to } 20 \text{ kHz}, T_A = 25^{\circ}\text{C}$		100			100			100		-ID
Coupling		(Input Referred), (Note 8)		-120			-120			-120		dB
Output Current	Source	$V_{IN}^{+} = 1V,$										
		$V_{IN}^- = 0V$ ,		40			40			40		
		$V^{+} = 15V,$	20	40		20	40		20	40		mA
		$V_{O} = 2V, T_{A} = 25^{\circ}C$										
	Sink	$V_{IN}^{-} = 1V, V_{IN}^{+} = 0V$										
		$V^{+} = 15V, T_{A} = 25^{\circ}C,$	10	20		10	20		10	20		mA
		$V_O = 2V$										
		$V_{IN}^- = 1V$ ,										
		$V_{IN}^+ = 0V$	1.0									
		$T_A = 25^{\circ}C, V_O = 200 \text{ mV},$	12	50		12	50		12	50		μΑ
		V <sup>+</sup> = 15V										
Short Circuit to 0	Ground	T <sub>A</sub> = 25°C, (Note 2),		40	00		40			40		
		V <sup>+</sup> = 15V		40	60		40	60		40	60	mA
Input Offset Volta	age	(Note 5)			4			5			7	mV
Input Offset Volta	age	$R_S = 0\Omega$			45							1100
Drift				7	15		7	20		7		μV/°C
Input Offset Curr	ent	$I_{IN(+)} - I_{IN(-)}$			30			75			100	nA
Input Offset Curr	ent	$R_S = 0\Omega$		40	000		40	000		40		A /º C
Drift				10	200		10	300		10		pA/°C
Input Bias Curre	nt	I <sub>IN(+)</sub> or I <sub>IN(-)</sub>		40	100		40	200		40	300	nA
Input Common-N		V <sup>+</sup> = 30 V, (Note 7)			144 -	_					14.	
Voltage Range		$(LM2904, V^{+} = 26V)$	0		V <sup>+</sup> –2	0		V+-2	0		V+-2	V

## **Electrical Characteristics** (Continued) $V^+ = +5.0V$ , (Note 4), unless otherwise stated

Parameter		Conditions		L	LM158A			LM358A			LM158/LM258			
				Min	Тур	Max	Min	Тур	Max	Min	Тур	Max		
Large Signal Voltage		V <sup>+</sup> = +15V												
Gain		$(V_O = 1V \text{ to } 11V)$		25			15			25			V/mV	
		$R_L \ge 2 k\Omega$												
Output	V <sub>OH</sub>	V <sup>+</sup> = +30V	$R_L = 2 k\Omega$	26			26			26			V	
Voltage		$(LM2904, V^+ = 26V)$	$R_L = 10 \text{ k}\Omega$	27	28		27	28		27	28		V	
Swing	V <sub>OL</sub>	$V^{+} = 5V, R_{L} = 10 \text{ k}\Omega$			5	20		5	20		5	20	mV	
Output Current	Source	$V_{IN}^{+} = +1V, V_{IN}^{-} = 0V$	<b>′</b> ,	10	20		10	20		10	20		mA	
		$V^{+} = 15V, V_{O} = 2V$		10	20		10	20		10	20		IIIA	
	Sink	$V_{IN}^- = +1V, V_{IN}^+ = 0V,$		10	15		5	8		5	8		mA	
		$V^{+} = 15V, V_{O} = 2V$		10	13		3	o		٦	o		IIIA	

## **Electrical Characteristics**

 $V^+ = +5.0V$ , (Note 4), unless otherwise stated

Parameter		Conditions		LM358			Units		
Parameter		Conditions	Min	Тур	Max	Min	Тур	Max	
Large Signal Voltage		$V^{+} = 15V, T_{A} = 25^{\circ}C,$							
Gain		$R_L \ge 2 \text{ k}\Omega$ , (For $V_O = 1V$	25	100		25	100		V/mV
		to 11V)							
Common-Mode		$T_A = 25^{\circ}C$ ,	65	0.5		50	70		4D
Rejection Ratio		$V_{CM} = 0V \text{ to } V^{+}-1.5V$	05	85		50	70		dB
Power Supply		V <sup>+</sup> = 5V to 30V							
Rejection Ratio		(LM2904, V <sup>+</sup> = 5V	65	100		50	100		dB
		to 26V), T <sub>A</sub> = 25°C							
Amplifier-to-Amplifier		f = 1 kHz to 20 kHz, T <sub>A</sub> = 25°C		-120			100		4D
Coupling		(Input Referred), (Note 8)		-120			-120		dB
Output Current	Source	$V_{IN}^{+} = 1V,$							
		$V_{IN}^- = 0V,$	00	40		00	40		Л
		$V^{+} = 15V,$	20	40		20	40		mA
		$V_{O} = 2V, T_{A} = 25^{\circ}C$							
	Sink	$V_{IN}^{-} = 1V, V_{IN}^{+} = 0V$							
		V <sup>+</sup> = 15V, T <sub>A</sub> = 25°C,	10	20		10	20		mA
		$V_O = 2V$							
		$V_{IN}^- = 1V$ ,							
		$V_{IN}^{+} = 0V$	10	50		10	F0		
		$T_A = 25^{\circ}C, V_O = 200 \text{ mV},$	12	50		12	50		μΑ
		V <sup>+</sup> = 15V							
Short Circuit to Groun	nd	T <sub>A</sub> = 25°C, (Note 2),		40	60		40	60	Л
		V <sup>+</sup> = 15V		40	60		40	60	mA
Input Offset Voltage		(Note 5)			9			10	mV
Input Offset Voltage		$R_S = 0\Omega$		7			7		μV/°C
Drift				/			7		μν/ Ο
Input Offset Current		$I_{IN(+)} - I_{IN(-)}$			150		45	200	nA
Input Offset Current		$R_S = 0\Omega$		10			10		~ A /° C
Drift				10			10		pA/°C
Input Bias Current		I <sub>IN(+)</sub> or I <sub>IN(-)</sub>		40	500		40	500	nA
Input Common-Mode		V <sup>+</sup> = 30 V, (Note 7)			\/+ C	_		\/+ C	\/
Voltage Range		(LM2904, V <sup>+</sup> = 26V)	0		V <sup>+</sup> –2	0		V <sup>+</sup> –2	V

## **Electrical Characteristics** (Continued)

 $V^+ = +5.0V$ , (Note 4), unless otherwise stated

Parameter		Conditions			LM358				Units	
		Condition	Min	Тур	Max	Min	Тур	Max		
Large Signal Voltage		V <sup>+</sup> = +15V								
Gain		$(V_O = 1V \text{ to } 11V)$		15			15			V/mV
		$R_L \ge 2 k\Omega$								
Output	V <sub>OH</sub>	V <sup>+</sup> = +30V	$R_L = 2 k\Omega$	26			22			V
Voltage		(LM2904, V <sup>+</sup> = 26V)	$R_L = 10 \text{ k}\Omega$	27	28		23	24		V
Swing	$V_{OL}$	$V^{+} = 5V, R_{L} = 10 \text{ k}\Omega$			5	20		5	100	mV
Output Current	Source	$V_{IN}^{+} = +1V, V_{IN}^{-} = 0V$	,	10	20		10	20		mA
		$V^{+} = 15V, V_{O} = 2V$		10	20		10	20		111/4
	Sink	$V_{IN}^- = +1V, V_{IN}^+ = 0V$	$V_{IN}^- = +1V, V_{IN}^+ = 0V,$		8		5	8		mA
		$V^{+} = 15V, V_{O} = 2V$		5	0		3	0		111/4

Note 1: For operating at high temperatures, the LM358/LM358A, LM2904 must be derated based on a +125°C maximum junction temperature and a thermal resistance of 120°C/W for MDIP, 182°C/W for Metal Can, 189°C/W for Small Outline package, and 230°C/W for micro SMD, which applies for the device soldered in a printed circuit board, operating in a still air ambient. The LM258/LM258A and LM158/LM158A can be derated based on a +150°C maximum junction temperature. The dissipation is the total of both amplifiers—use external resistors, where possible, to allow the amplifier to saturate or to reduce the power which is dissipated in the integrated circuit.

**Note 2:** Short circuits from the output to V<sup>+</sup> can cause excessive heating and eventual destruction. When considering short circuits to ground, the maximum output current is approximately 40 mA independent of the magnitude of V<sup>+</sup>. At values of supply voltage in excess of +15V, continuous short-circuits can exceed the power dissipation ratings and cause eventual destruction. Destructive dissipation can result from simultaneous shorts on all amplifiers.

**Note 3:** This input current will only exist when the voltage at any of the input leads is driven negative. It is due to the collector-base junction of the input PNP transistors becoming forward biased and thereby acting as input diode clamps. In addition to this diode action, there is also lateral NPN parasitic transistor action on the IC chip. This transistor action can cause the output voltages of the op amps to go to the V<sup>+</sup>voltage level (or to ground for a large overdrive) for the time duration that an input is driven negative. This is not destructive and normal output states will re-establish when the input voltage, which was negative, again returns to a value greater than -0.3V (at 25°C).

Note 4: These specifications are limited to  $-55^{\circ}C \le T_{A} \le +125^{\circ}C$  for the LM158/LM158A. With the LM258/LM258A, all temperature specifications are limited to  $-25^{\circ}C \le T_{A} \le +85^{\circ}C$ , the LM358/LM358A temperature specifications are limited to  $0^{\circ}C \le T_{A} \le +70^{\circ}C$ , and the LM2904 specifications are limited to  $-40^{\circ}C \le T_{A} \le +85^{\circ}C$ .

Note 5:  $V_O \approx 1.4V$ ,  $R_S = 0\Omega$  with V<sup>+</sup> from 5V to 30V; and over the full input common-mode range (0V to V<sup>+</sup> -1.5V) at 25°C. For LM2904, V<sup>+</sup> from 5V to 26V.

Note 6: The direction of the input current is out of the IC due to the PNP input stage. This current is essentially constant, independent of the state of the output so no loading change exists on the input lines.

Note 7: The input common-mode voltage of either input signal voltage should not be allowed to go negative by more than 0.3V (at 25°C). The upper end of the common-mode voltage range is V<sup>+</sup> –1.5V (at 25°C), but either or both inputs can go to +32V without damage (+26V for LM2904), independent of the magnitude of V<sup>+</sup>

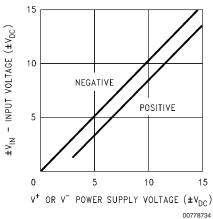
Note 8: Due to proximity of external components, insure that coupling is not originating via stray capacitance between these external parts. This typically can be detected as this type of capacitance increases at higher frequencies.

Note 9: Refer to RETS158AX for LM158A military specifications and to RETS158X for LM158 military specifications.

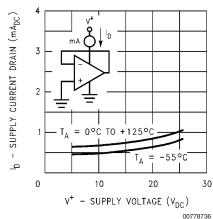
Note 10: Human body model, 1.5 k $\Omega$  in series with 100 pF.

## **Typical Performance Characteristics**

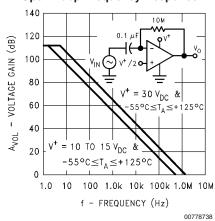




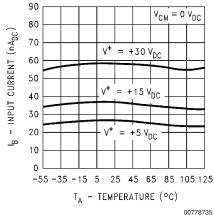
#### **Supply Current**



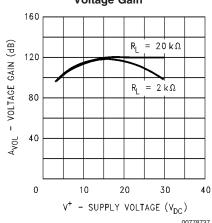
## **Open Loop Frequency Response**



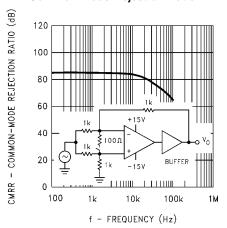
## **Input Current**



#### Voltage Gain



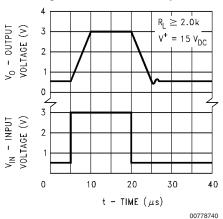
#### Common-Mode Rejection Ratio



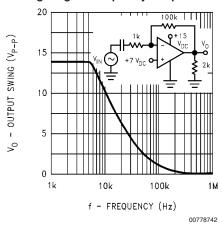
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## **Typical Performance Characteristics** (Continued)

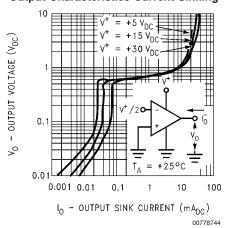
#### Voltage Follower Pulse Response



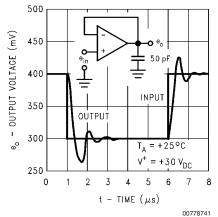
#### Large Signal Frequency Response



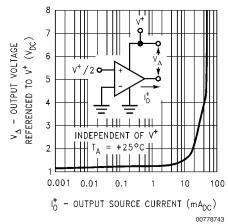
#### **Output Characteristics Current Sinking**



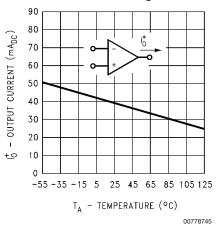
## Voltage Follower Pulse Response (Small Signal)



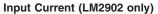
## **Output Characteristics Current Sourcing**

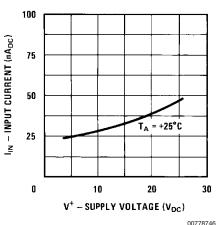


## **Current Limiting**

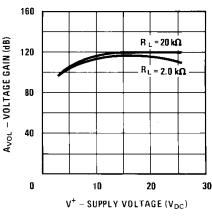


## **Typical Performance Characteristics** (Continued)





#### Voltage Gain (LM2902 only)



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## **Application Hints**

The LM158 series are op amps which operate with only a single power supply voltage, have true-differential inputs, and remain in the linear mode with an input common-mode voltage of 0  $V_{\rm DC}.$  These amplifiers operate over a wide range of power supply voltage with little change in performance characteristics. At  $25\,^{\circ}\text{C}$  amplifier operation is possible down to a minimum supply voltage of  $2.3~V_{\rm DC}.$ 

Precautions should be taken to insure that the power supply for the integrated circuit never becomes reversed in polarity or that the unit is not inadvertently installed backwards in a test socket as an unlimited current surge through the resulting forward diode within the IC could cause fusing of the internal conductors and result in a destroyed unit.

Large differential input voltages can be easily accomodated and, as input differential voltage protection diodes are not needed, no large input currents result from large differential input voltages. The differential input voltage may be larger than V+ without damaging the device. Protection should be provided to prevent the input voltages from going negative more than  $-0.3~V_{\rm DC}$  (at  $25^{\circ}{\rm C}$ ). An input clamp diode with a resistor to the IC input terminal can be used.

To reduce the power supply current drain, the amplifiers have a class A output stage for small signal levels which converts to class B in a large signal mode. This allows the amplifiers to both source and sink large output currents. Therefore both NPN and PNP external current boost transistors can be used to extend the power capability of the basic amplifiers. The output voltage needs to raise approximately 1 diode drop above ground to bias the on-chip vertical PNP transistor for output current sinking applications.

For ac applications, where the load is capacitively coupled to the output of the amplifier, a resistor should be used, from the output of the amplifier to ground to increase the class A bias current and prevent crossover distortion. Where the load is directly coupled, as in dc applications, there is no crossover distortion.

Capacitive loads which are applied directly to the output of the amplifier reduce the loop stability margin. Values of 50 pF can be accommodated using the worst-case non-inverting unity gain connection. Large closed loop gains or resistive isolation should be used if larger load capacitance must be driven by the amplifier.

The bias network of the LM158 establishes a drain current which is independent of the magnitude of the power supply voltage over the range of 3  $V_{\rm DC}$  to 30  $V_{\rm DC}$ .

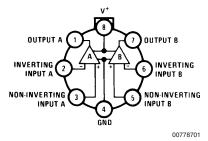
Output short circuits either to ground or to the positive power supply should be of short time duration. Units can be destroyed, not as a result of the short circuit current causing metal fusing, but rather due to the large increase in IC chip dissipation which will cause eventual failure due to excessive function temperatures. Putting direct short-circuits on more than one amplifier at a time will increase the total IC power dissipation to destructive levels, if not properly protected with external dissipation limiting resistors in series with the output leads of the amplifiers. The larger value of output source current which is available at 25°C provides a larger output current capability at elevated temperatures (see typical performance characteristics) than a standard IC op amp.

The circuits presented in the section on typical applications emphasize operation on only a single power supply voltage. If complementary power supplies are available, all of the standard op amp circuits can be used. In general, introducing a pseudo-ground (a bias voltage reference of V+/2) will allow operation above and below this value in single power supply systems. Many application circuits are shown which take advantage of the wide input common-mode voltage range which includes ground. In most cases, input biasing is not required and input voltages which range to ground can easily be accommodated.

## **Connection Diagrams**

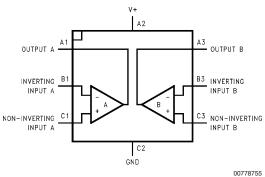
# OUTPUT A 1 0 UTPUT B INVERTING INPUT A 2 OUTPUT B NON-INVERTING INPUT B GND 4 INVERTING INPUT B 00778702

#### Metal Can Package



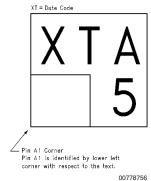
**Top View** 

#### 8-Bump micro SMD



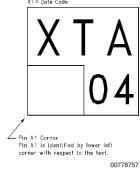
Top View (Bump Side Down)

## LM358BP micro SMD Marking Orientation



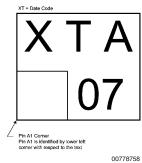
**Top View** 

#### LM2904IBP micro SMD Marking Orientation



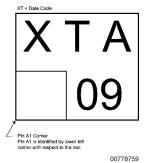
**Top View** 

## LM358TP micro SMD Marking Orientation



**Top View** 

## LM2904ITP micro SMD Marking Orientation



**Top View** 

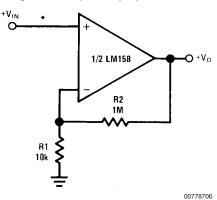
#### **Ordering Information Temperature Range NSC Drawing Package** -25°C to 85°C -55°C to 125°C 0°C to 70°C -40°C to 85°C SO-8 LM358AM LM2904M LM2904MX LM358AMX A80M LM358M LM358MX 8-Pin Molded DIP LM2904N LM358AN N08E LM358N 8-Pin Ceramic DIP LM158AJ/883(Note 11) LM158J/883(Note 11) J08A LM158J LM158AJLQML(Note 12) LM158AJQMLV(Note 12) TO-5, 8-Pin Metal LM258H LM358H LM158AH/883(Note 11) Can LM158H/883(Note 11) LM158AH H08C LM158H LM158AHLQML(Note 12) LM158AHLQMLV(Note 12) 8-Bump micro LM358BP LM2904IBP BPA08AAB SMD LM358BPX LM2904IBPX 0.85 mm Thick 8-Bump micro LM358TP LM2904ITP TPA08AAA SMD LM358TPX LM2904ITPX 0.50 mm Thick Lead Free 14-Pin Ceramic LM158AWG/883 WG10A SOIC

**Note 11:** LM158 is available per SMD #5962-8771001 LM158A is available per SMD #5962-8771002

Note 12: See STD Mil DWG 5962L87710 for Radiation Tolerant Devices

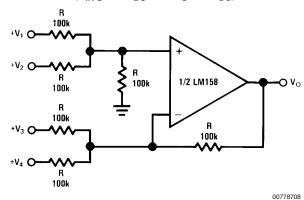
# Typical Single-Supply Applications (V+ = $5.0 \text{ V}_{DC}$ )

Non-Inverting DC Gain (0V Output)

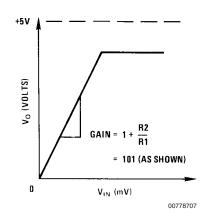


\*R not needed due to temperature independent I<sub>IN</sub>

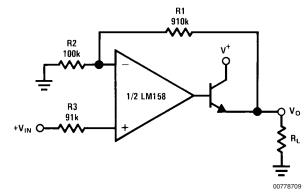
## **DC Summing Amplifier** (V<sub>IN'S</sub> $\geq$ 0 V<sub>DC</sub> and V<sub>O</sub> $\geq$ 0 V<sub>DC</sub>)



Where:  $V_0 = V_1 + V_2 - V_3 - V_4$  $(V_1 + V_2) \ge (V_3 + V_4)$  to keep  $V_O \ge 0$   $V_{DC}$ 



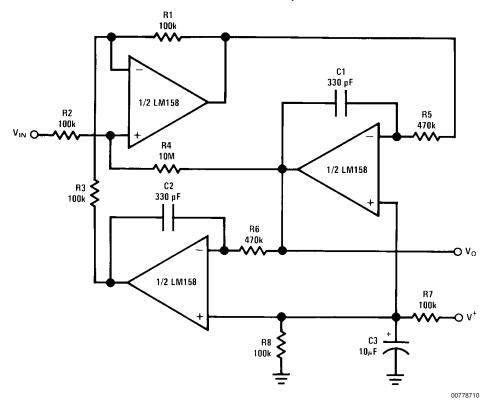
#### **Power Amplifier**



 $V_O = 0 V_{DC}$  for  $V_{IN} = 0 V_{DC}$  $A_{V} = 10$ 

## Typical Single-Supply Applications (V<sup>+</sup> = 5.0 V<sub>DC</sub>) (Continued)

## "BI-QUAD" RC Active Bandpass Filter



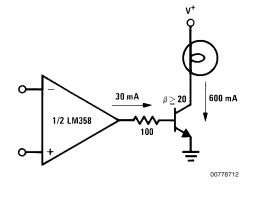
 $f_0 = 1 \text{ kHz}$  Q = 50  $A_V = 100 (40 \text{ dB})$ 

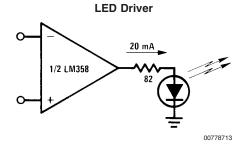
#### **Fixed Current Sources**

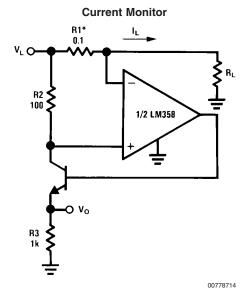
# 

 $I_2 = \left(\frac{R1}{R2}\right)I_1$ 

## Lamp Driver

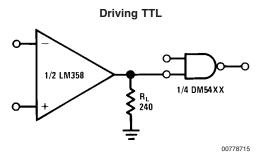




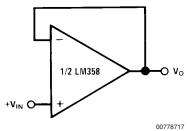


 $V_O = \frac{1V \, (I_{\underline{L}})}{1A}$ 

\*(Increase R1 for  $I_L$  small)  $V_L \leq V^+ - 2V$ 

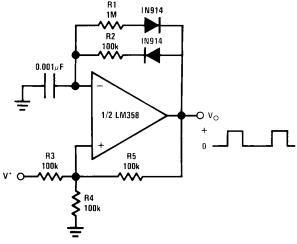


## Voltage Follower



 $V_{O} = V_{IN}$ 

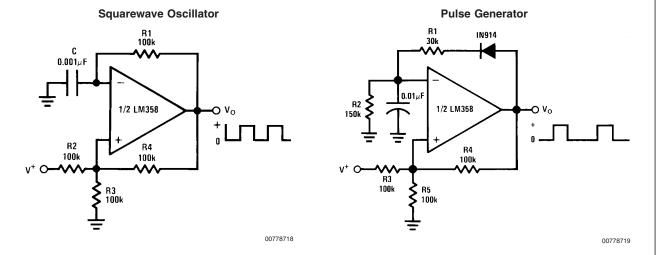
#### **Pulse Generator**



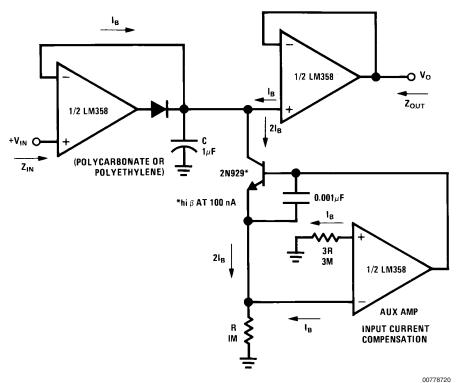
13

00778716

## Typical Single-Supply Applications (V<sup>+</sup> = 5.0 V<sub>DC</sub>) (Continued)

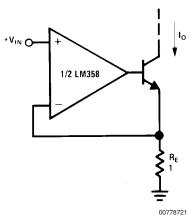


## **Low Drift Peak Detector**

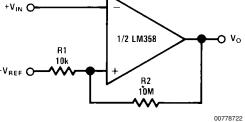


HIGH  $Z_{IN}$ LOW  $Z_{OUT}$ 

## **High Compliance Current Sink**

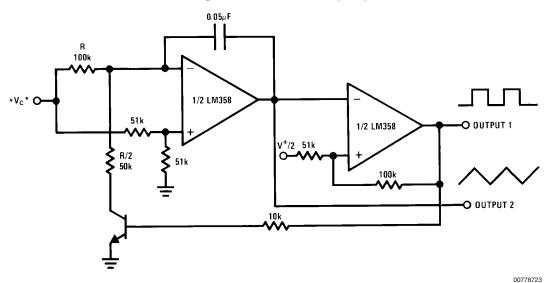


# **Comparator with Hysteresis**



 $I_O = 1 \text{ amp/volt } V_{IN}$ (Increase  $R_E$  for  $I_O$  small)

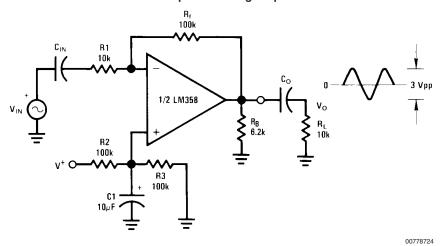
#### **Voltage Controlled Oscillator (VCO)**



\*WIDE CONTROL VOLTAGE RANGE: 0  $V_{DC} \le V_C \le 2$  (V+ -1.5V  $_{DC}$ )

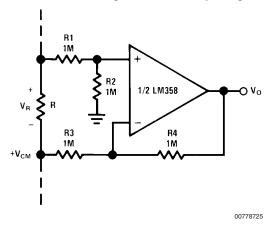
## Typical Single-Supply Applications (V<sup>+</sup> = 5.0 V<sub>DC</sub>) (Continued)

## **AC Coupled Inverting Amplifier**

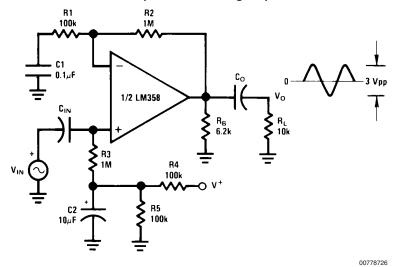


 $A_V = \frac{R_f}{R_1}$  (As shown,  $A_V = 10$ 

## **Ground Referencing a Differential Input Signal**



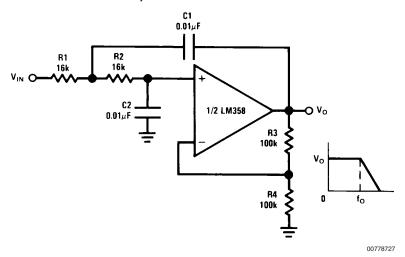
## **AC Coupled Non-Inverting Amplifier**



 $A_V = 1 + \frac{R2}{R1}$ 

 $A_v = 11$  (As Shown)

#### DC Coupled Low-Pass RC Active Filter

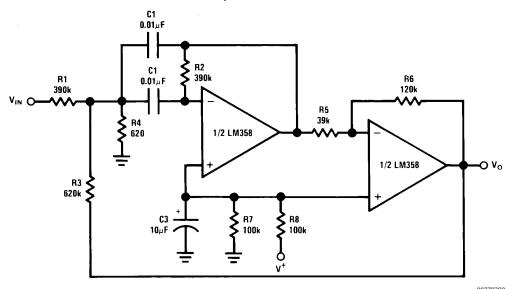


 $f_0 = 1 \text{ kHz}$ 

Q = 1

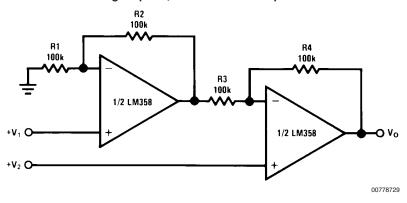
A<sub>V</sub> = 2

#### **Bandpass Active Filter**



 $f_0 = 1 \text{ kHz}$ Q = 25

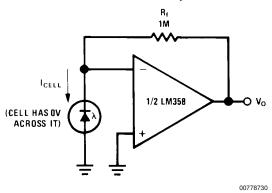
## High Input Z, DC Differential Amplifier



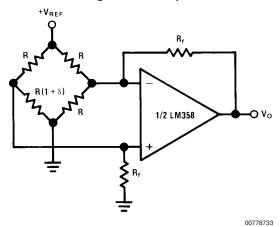
For 
$$\frac{R1}{R2} = \frac{R4}{R3}$$
 (CMRR depends on this resistor ratio match)
$$V_O = 1 + \frac{R4}{R3} (V_2 - V_1)$$
As Shown:  $V_O = 2 (V_2 - V_1)$ 

$$V_0 = 1 + \frac{R4}{R3} (V_2 - V_1)$$

#### Photo Voltaic-Cell Amplifier

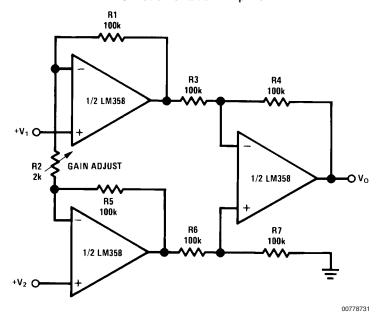


#### **Bridge Current Amplifier**



For  $\delta << 1$  and  $R_f >> R$   $V_O \cong V_{REF} \left(\frac{\delta}{2}\right) \frac{R_f}{R}$ 

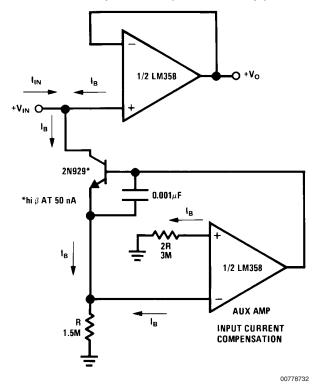
## High Input Z Adjustable-Gain DC Instrumentation Amplifier



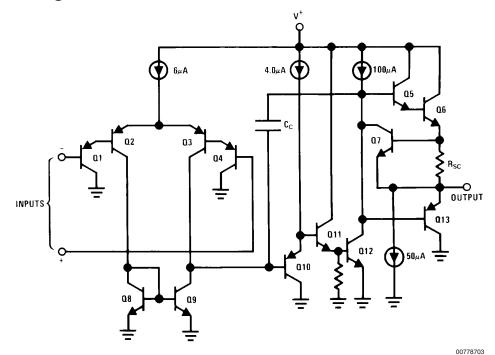
If R1 = R5 & R3 = R4 = R6 = R7 (CMRR depends on match)  $V_O=1+\frac{2R1}{R2}\;(V_2-V_1)$  As shown  $V_O=$  101 ( $V_2-V_1$ )

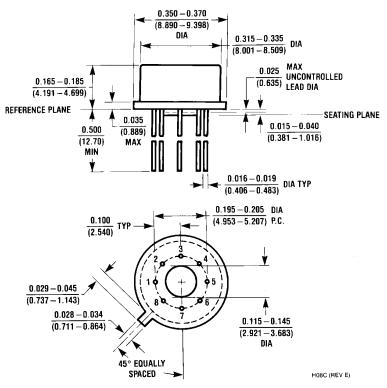
## Typical Single-Supply Applications (V<sup>+</sup> = 5.0 V<sub>DC</sub>) (Continued)

Using Symmetrical Amplifiers to Reduce Input Current (General Concept)

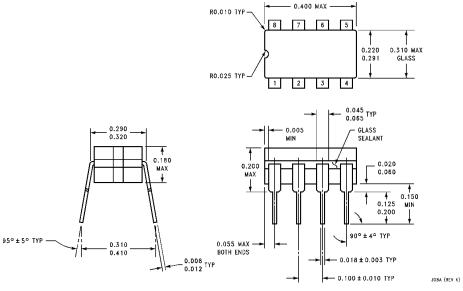


## Schematic Diagram (Each Amplifier)

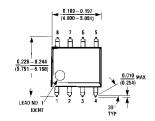


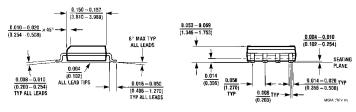


Metal Can Package (H) NS Package Number H08C

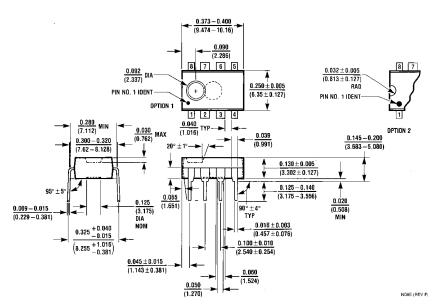


Cerdip Package (J) NS Package Number J08A

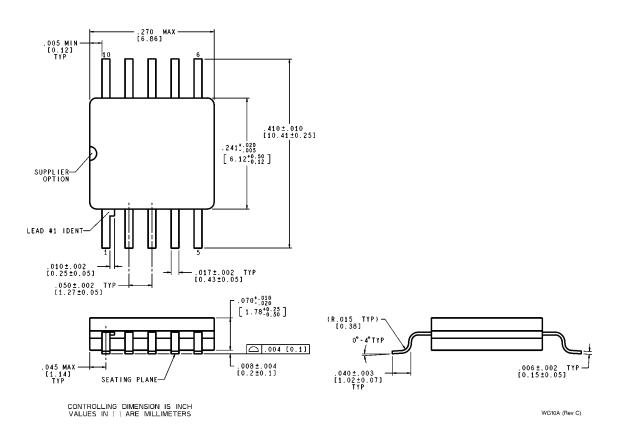




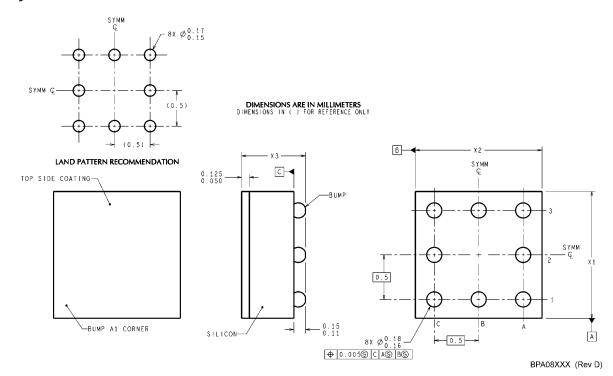
SOIC Package (M)
NS Package Number M08A



Molded Dip Package (N) NS Package Number N08E



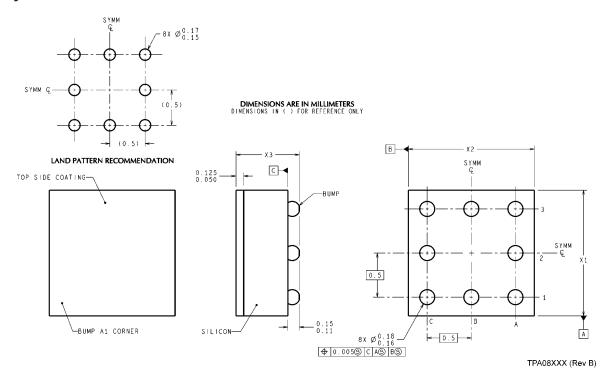
Order Number LM158AWG/883 NS Package Number WG10A



NOTES: UNLESS OTHERWISE SPECIFIED

- 1. EPOXY COATING
- 2. 63Sn/37Pb EUTECTIC BUMP
- 3. RECOMMEND NON-SOLDER MASK DEFINED LANDING PAD.
- 4. PIN A1 IS ESTABLISHED BY LOWER LEFT CORNER WITH RESPECT TO TEXT ORIENTATION REMAINING PINS ARE NUMBERED COUNTERCLOCKWISE.
- 5. XXX IN DRAWING NUMBER REPRESENTS PACKAGE SIZE VARIATION WHERE  $X_1$  IS PACKAGE WIDTH,  $X_2$  IS PACKAGE LENGTH AND  $X_3$  IS PACKAGE HEIGHT.
- 6. REFERENCE JEDEC REGISTRATION MO-211, VARIATION BC.

8-Bump micro SMD NS Package Number BPA08AAB  $X_1 = 1.285$   $X_2 = 1.285$   $X_3 = 0.850$ 



NOTES: UNLESS OTHERWISE SPECIFIED

- 1. EPOXY COATING
- 2 RECOMMEND NON-SOLDER MASK DEFINED LANDING PAD
- 3. PIN A1 IS ESTABLISHED BY LOWER LEFT CORNER WITH RESPECT TO TEXT ORIENTATION REMAINING PINS ARE NUMBERED COUNTERCLOCKWISE.
- 4. XXX IN DRAWING NUMBER REPRESENTS PACKAGE SIZE VARIATION WHERE X1 IS PACKAGE WIDTH, X2 IS PACKAGE LENGTH AND X3 IS PACKAGE HEIGHT.
- 5. REFERENCE JEDEC REGISTRATION MO-211, VARIATION BC.

8-Bump micro SMD Lead Free **NS Package Number TPA08AAA**  $X_1 = 1.285$   $X_2 = 1.285$   $X_3 = 0.500$ 

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- 2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

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