

# **TDC1044**

# Monolithic Video A/D Converter

The TDC1044A is a 25 Msps (Megasample per second) full-parallel analog-to-digital converter, capable of converting an analog signal with full-power frequency components up to 12.5 MHz into 4-bit digital words. Use of a sample-and-hold circuit is not necessary for operation of the TDC 1044A. All digital inputs and outputs are TTL compatible.

The TDC1044A consists of 15 latching comparators, encoding logic, and an output register. A single convert signal controls the conversion operation. Output formats are true/inverted binary or true/inverted offset two's complement codes.

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Parts are tested using original factory test programs or Rochester developed test solutions to guarantee product meets or exceeds the OCM data sheet.

## **Quality Overview**

AS9120 certification

• ISO-9001

- Qualified Manufacturers List (QML) MIL-PRF-35835
  - Class Q Military
  - Class V Space Level
- Qualified Suppliers List of Distributors (QSLD)
  - Rochester is a critical supplier to DLA and meets all industry and DLA standards.

Rochester Electronics, LLC is committed to supplying products that satisfy customer expectations for quality and are equal to those originally supplied by industry manufacturers.

The original manufacturer's datasheet accompanying this document reflects the performance and specifications of the Rochester manufactured version of this device. Rochester Electronics guarantees the performance of its semiconductor products to the original OCM specifications. 'Typical' values are for reference purposes only. Certain minimum or maximum ratings may be based on product characterization, design, simulation, or sample testing.

# FOR REFERENCE ONLY

# TDC1044A Monolithic Video A/D Converter 4-Bit, 25 Msps

### Features

- 4-bit resolution
- 1/4 LSB non-linearity
- Sample-and-hold circuit not required
- 25 Msps conversion rate
- Selectable output format
- 16-lead DIP and 20-lead PLCC packages

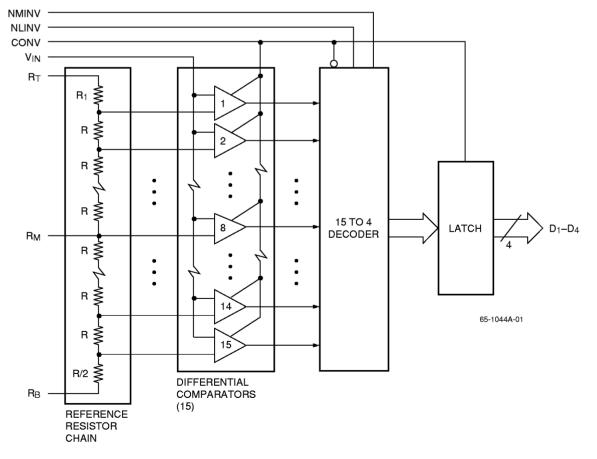
### **Applications**

- Digital communications
- Video special effects
- Radar data conversion
- Medical imaging

### Description

The TDC1044A is a 25 Msps (Megasample per second) fullparallel analog-to-digital converter, capable of converting an analog signal with full-power frequency components up to 12.5 MHz into 4-bit digital words. Use of a sample-and-hold circuit is not necessary for operation of the TDC1044A. All digital inputs and outputs are TTL compatible.

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### **Block Diagram**

### **Functional Description**

#### **General Information**

The TDC1044A has three functional sections: a comparator array, encoding logic, and an output register. The comparator array compares the input signal with 15 reference voltages to produce an N-of-15 thermometer code. All the comparators referred to voltages more positive than the input signal will be off, and those referred to voltages more negative than the input signal will be on. Encoding logic converts the N-of-15 code into binary or two's complement coding and can invert either output code. This coding function is controlled by DC signals on pins NMINV and NLINV. The output register holds the output constant between updates.

#### Power

The TDC1044A operates from two power supply voltages, +5.0V and -5.2V. The return for I<sub>CC</sub> (the current drawn from the +5.0V supply) is D<sub>GND</sub>. The return for I<sub>EE</sub> (the current drawn from the -5.2V supply) is A<sub>GND</sub>. All power and ground pins must be connected.

#### Reference

The TDC1044A converts analog signals in the range V<sub>RB</sub>  $\leq$  V<sub>IN</sub>  $\leq$  V<sub>RB</sub> into digital form. V<sub>RB</sub> (the voltage applied to R<sub>B</sub> at the bottom of the reference resistor chain) and V<sub>RT</sub> (the voltage applied to R<sub>B</sub> at the top of the reference resistor chain) should be between +0.1V and -1.1V. V<sub>RT</sub> should be more positive than V<sub>RB</sub> within that range. The voltage applied across the reference resistor chain (V<sub>RT</sub> - V<sub>RB</sub>) must be between 0.4V and 1.3V.

Nominal voltages are  $V_{RT} = 0.00V$  and  $V_{RB} = -1.00V$ . These voltages may be varied dynamically up to 10MHz. Due to slight variation in the reference currents with clock and input signals, RT and RB should be low-impedance points. For circuits in which the reference is not varied, a bypass capacitor to ground is recommended. If the reference inputs are varied dynamically (as in an Automatic Gain Control circuit), a low-impedance reference source is required.

A reference middle,  $R_M$ , is also provided; this may be used as an input to adjust the mid-scale point in order to improve integral linearity. This point may also be used as a tap to supply a mid-scale voltage to offset the analog input. If  $V_{RM}$  is used as an output, it must be connected to a high input impedance device which has small input current. Noise at this point may adversely affect the performance of this device.

#### Controls

Two function control pins, NMINV and NLINV, set the output format to be either straight binary or offset two's complement, in either true or inverted sense, according to Table 1. These pins are active LOW as signified by the prefix "N" in the signal name. They may be tied to VCC for a logic "1" and DGND for a logic "0."

NMINV controls the MSB, D<sub>1</sub>; NLINV controls the three LSBs: D<sub>2</sub>, D<sub>3</sub> and D<sub>4</sub>.

#### Convert

The TDC1044A requires a CONVert (CONV) signal. A sample is taken (the comparators are latched) within tSTO after a rising edge of CONV. The coded result is translated to the output latches on the next rising edge. The outputs hold the previous data a minimum time (tHO) after the rising edge of the CONV signal. New data becomes valid after a maximum delay time, tD.

#### Analog Input

The TDC1044A uses latching comparators which cause the input impedance to vary slightly with the signal level. For optimal performance, the source impedance of the driving circuit must less than 25 Ohms. Within the range of  $V_{EE}$  to +0.5V, the input signal will not damage the device. If the input signal is at a voltage between V<sub>RT</sub> and V<sub>RB</sub>, the output will be a binary code between 0 and 15 inclusive. A signal outside this range will indicate either full-scale positive or full-scale negative, depending on whether the signal is off-scale in the positive or negative direction.

#### Outputs

TDC1044A outputs are TTL compatible, and capable of driving four low-power Schottky TTL (54/74 LS) unit loads. The outputs hold the previous data a minimum time (tHO) after the rising edge of the CONV signal. Data becomes valid after a maximum delay time (tD) after the rising edge of CONV. For optimum performance, 2.2 kOhm pull-up resistors are recommended.

#### No Connects

Pin 3 of the TDC1044A is labeled No Connect (NC), and has no connection to the chip. Connect this pin to AGND for best noise performance.

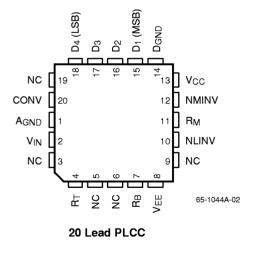
	Bir	nary	Offset Two's	o's Complement	
Range	True	Inverted	True	Inverted	
-1.00V FS	NMINV = 1	0	0	1	
	NLINV = 1	0	1	0	
0.000V	0000	1111	1000	0111	
-0.067V	0001	1110	1001	0110	
-0.133V	0010	1101	1010	0101	
-0.200V	0011	1100	1011	0100	
-0.267V	0100	1011	1100	0011	
-0.333V	0101	1010	1101	0010	
-0.400V	0110	1001	1110	0001	
-0.467V	0111	1000	1111	0000	
-0.533V	1000	0111	0000	1111	
-0.600V	1001	0110	0001	1110	
-0.667V	1010	0101	0010	1101	
-0.733V	1011	0100	0011	1100	
-0.800V	1100	0011	0100	1011	
-0.867V	1101	0010	0101	1010	
-0.933V	1110	0001	0110	1001	
-1.000V	1111	0000	0111	1000	

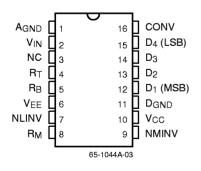
#### Table 1. Output Coding<sup>1</sup>

#### Note:

1. Input voltages are at code centers.

### **Pin Assignments**







# **Pin Descriptions**

	Pin N	umber		
Pin Name	DIP	PLCC	Value	Description
Power				·
Vcc	10	13	+5.0V	Positive Supply Voltage
VEE	6	8	-5.2V	Negative Supply Voltage
Dgnd	11	14	0.0V	Digital Ground
Agnd	1	1	0.0V	Analog Ground
Reference				
RT	4	4	0.0V	Reference Resistor, Top
RM	8	11	-0.5V	Reference Resistor, Middle
RB	5	7	-1.0V	Reference Resistor, Bottom
Control				
NMINV	9	12	TTL	Not MSB Invert
NLINV	7	10	TTL	Not LSB Invert
Convert				
CONV	16	20	TTL	Convert
Analog Input				
VIN	2	2	0V to -1V	Analog Input Signal
Output				·
D1	12	15	TTL	MSB Output
D2	13	16	TTL	
D3	14	17	TTL	
D4	15	18	TTL	LSB Output
NC	3	3, 5, 6, 9, 19	Agnd	No Connect

#### **Absolute Maximum Ratings**

(beyond which the device may be damaged)<sup>1</sup>

Туре	Parameter	Min	Max	Unit
Supply Voltages	V <sub>CC</sub> (measured to D <sub>GND</sub> )		7.0	V
	VEE (measured to AGND)	+0.5	-7.0	V
	AGND (measured to DGND)	-0.5	+0.5	V
Input Voltages	CONV, NMINV, NLINV (measured to DGND)	-0.5	+5.5	V
	VIN, VRT, VRB (measured to AGND)		VEE	V
	VRT (measured to VRB)		+2.2	V
Output	Applied voltage (measured to DGND) <sup>2</sup>	-0.5	+5.5	V
	Applied current, externally forced <sup>3,4</sup>	-1.0	+6.0	mA
	Short circuit duration (single output in high state to ground)		1	sec
Temperature	Operating, ambient	-55	+125	°C
	Operating, junction		+150	°C
	Lead, soldering (10 seconds)		+300	°C
	Storage	-65	+150	°C

Notes:

1. Absolute maximum ratings are limiting values applied individually while all other parameters are within specified operating conditions. Functional operation under any of these conditions is NOT implied.

- 2. Applied voltage must be current limited to specified range.
- 3. Forcing voltage must be limited to specified range.
- 4. Current is specified as positive when flowing into the device.

Parameter		Min.	Nom.	Max.	Units
Vcc	Positive Supply Voltage (measured to DGND)	4.75	5.0	5.25	V
VEE	Negative Supply Voltage (measured to AGND)	-4.9	-5.2	-5.5	v
Vagnd	Analog Ground Voltage (measured to DGND)	-0.1	0.0	0.1	v
tpwL	CONV Pulse Width, LOW	17			ns
tP <b>W</b> H	CONV Pulse Width, HIGH	17			ns
VIL	Input Voltage, Logic LOW			0.8	V
VIH	Input Voltage, Logic HIGH	2.0			v
IOL	Output Current, Logic LOW			4.0	mA
ЮН	Output Current, Logic HIGH			-400	μA
VRT	Most Positive Reference	-1.9	0.0	0.1	V
VRB	Most Negative Reference	-2.1	-1.0	-0.1	v
Vrt – Vrb	Reference Differential	0.2	1.0	2.0	v
VIN	Input Voltage	VRB		VRT	V
Та	Ambient Temperature, Still Air	0		70	°C

# **Operating Conditions**

### **Electrical Characteristics**

Within specified operating conditions

Param	eter	Test Conditions	Min.	Max.	Units
lcc	Positive Supply Current	V <sub>CC</sub> = Max, static <sup>1</sup>		15	mA
IEE	Negative Supply Current	VEE = Max, static	•		
		$T_A = 0^{\circ}C$ to $70^{\circ}C$		-50	mA
		TA = 70°C		-40	mA
IREF	Reference Current	VRT, VRB = Nom		2	mA
RREF	Total Reference Resistance	7	500		Ohms
RIN	Input Equivalent Resistance	VRT, VRB = Nom, VIN = VRB	250		Kohms
CIN	Input Capacitance	7		25	pF
ICB	Input Constant Bias Current	VEE = Max		40	μA
ΙL	Input Current, Logic LOW	Vcc = Max, VI - 0.5V			
		CONV		-0.8	mA
		NMINV, NLINV		-0.8	mA
Ιн	Input Current, Logic HIGH	Vcc = Max, VI = 2.4V		200	μA
lj –	Input Current, Max Input Voltage	V <sub>CC</sub> = Max, V <sub>I</sub> = 5.5V		1.0	mA
Vol	Output Voltage, Logic LOW	V <sub>C</sub> C = Min, I <sub>O</sub> L = Max		0.5	V
Vон	Output Voltage, Logic HIGH	V <sub>C</sub> C = Min, I <sub>O</sub> H = Max	2.4		V
los	Short Circuit Output Current	V <sub>CC</sub> = Max, One pin to ground, one second duration, Output HIGH		-300	mA
CI	Digital Input Capacitance	TA = 25°C, F = 1 MHz		15	pF

Note:

1. Worst case: all digital inputs and outputs LOW.

### **Switching Characteristics**

Within specified operating conditions

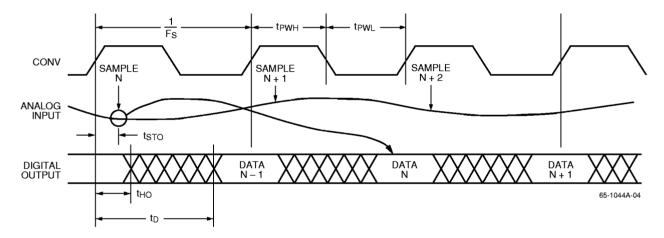
Parameter		Test Conditions	Min.	Max.	Units
Fs	Maximum Conversion Rate	V <sub>C</sub> C = Min, V <sub>E</sub> E = Min	25		Msps
tsto	Sampling Time Offset	V <sub>C</sub> C = Min, V <sub>E</sub> E = Min		10	ns
tD	Digital Output Delay	V <sub>CC</sub> = Min, V <sub>EE</sub> = Min, Load 1		30	ns
tHO	Digital Output Hold Time	V <sub>CC</sub> = Max, V <sub>EE</sub> = Max, Load 1	5		ns

### **System Performance Characteristics**

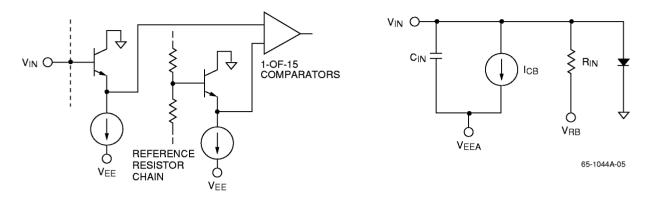
Within specified operating conditions

Param	leter	Test Conditions	Min.	Max.	Units
Eli	Linearity Error Integral Independent	VRB = Nom		1.6	%
Eld	Linearity Error Differential			1.6	%
CS	Code Size	VRT, VRB = Nom	75	125	% Nominal
Еот	Offset Error Top	VIN = VRT		+30	mV
Еов	Offset Error Bottom	VIN = VRB		+40	mV
Тсо	Offset Error Temperature Coefficient			<b>±</b> 20	μV/°C
BW	Bandwidth, Full Power Input		12.5		MHz
tTR	Transient Response, Full Scale			10	ns
Eap	Aperture Error			30	ps

# **Timing Diagram**



### **Equivalent Circuits**



#### Figure 1. Simplified Analog Input Equivalent Circuit

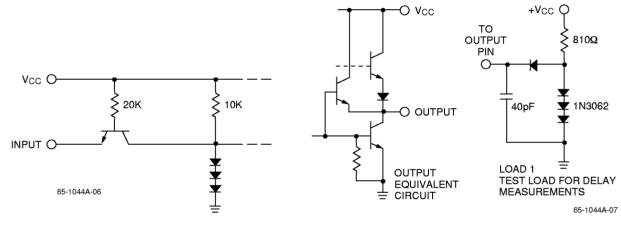




Figure 3. Output Circuits

### **Applications Discussion**

#### Calibration

To calibrate the TDC1044A, adjust  $V_{RT}$  and  $V_{RB}$  to set the 1st and 15th thresholds to the desired voltages. Assuming a 0V to -1V desired range, continuously strobe the converter with -0.0033V (1/2 LSB from 0.000V) on the analog input, and adjust  $V_{RT}$  for output toggling between codes 0000 and 0001. Then apply -0.976V (1/2 LSB from -1.000V) and adjust  $V_{RB}$  for toggling between codes 1110 and 1111. Instead of adjusting  $V_{RT}$ ,  $R_T$  can be connected to analog ground and the 0V end of the range calibrated with an amplifier offset control.  $R_B$  is a convenient point for gain adjustment that is not in the analog signal path.

#### **Typical Interface Circuit**

The TDC1044A does not require a special input buffer amplifier to drive the analog input because of its low input capacitance. A terminated low-impedance transmission line (<100 Ohms) connected to the VIN terminal of the device is sufficient if the input voltage levels match those of the A/D converter.

However, many driver circuits lack sufficient offset control, drive current, or gain stability. The typical interface circuit in Figure 4 shows a simple amplifier and voltage reference circuit that may be used with the device. U2 is a wide-band operational amplifier with a gain factor of -1. As the video input increases from zero to one volt, VIN of the TDC1044A decreases from zero to -1 volt. With true binary selected (NMINV = 1 and NLINV = 1), output codes increase from 0000 to 1111.

A small value resistor, R12, serves to isolate the small input capacitance of the A/D converter from the amplifier output and insure frequency stability. Pulse and frequency response of the amplifier are optimized by variable capacitor C12. The reference voltage for the TDC1044A is generated by amplifier U3. System gain is adjusted by varying R9, which controls the reference voltage level to the A/D converter.

Input voltage range and input impedance for the circuit are determined by resistors R1 and R2. Formulas for calculating values for these input resistors are:

$$\mathbf{R1} = \frac{1}{\left(\frac{2\mathbf{VR}}{\mathbf{Z}_{\mathrm{IN}}}\right) - \frac{1}{1000}}$$

and

$$R2 = Z_{IN} - \left(\frac{1000 R1}{1000 + R1}\right)$$

where  $V_R$  is the input voltage range of the circuit,  $Z_{IN}$  is the input impedance of the circuit, and the constant 1000 comes from the value of R3. As shown, the circuit is set up for  $1V_{P-P}$  75 Ohm video input.

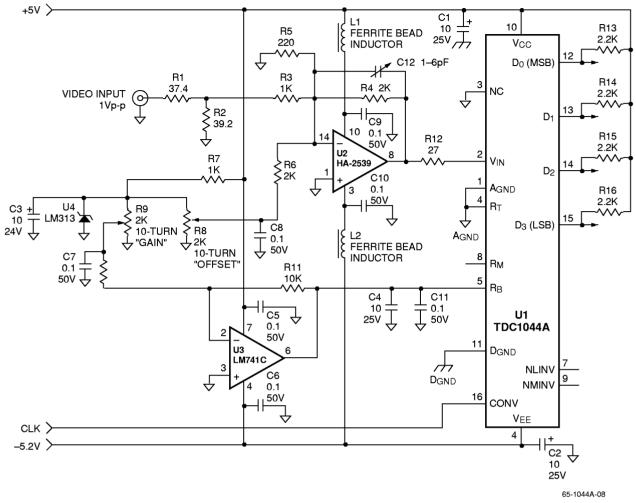


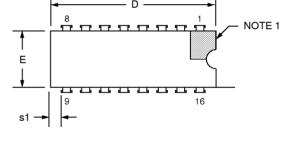
Figure 4. Typical Interface Circuit

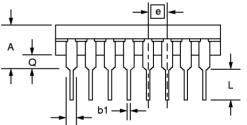
### **Mechanical Dimensions**

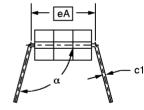
#### **16-Lead Ceramic DIP Package**

Sumbol	Inc	hes	Millim	neters	Notes
Symbol	Min.	Max.	Min.	Max.	Notes
А	_	.200	_	5.08	
b1	.014	.023	.36	.58	8
b2	.050	.065	1.27	1.65	2
c1	.008	.015	.20	.38	8
D	.745	.840	18.92	21.33	4
Е	.220	.310	5.59	7.87	4
е	.100	BSC	2.54 BSC		5, 9
eA	.300	BSC	7.62	BSC	7
L	.115	.160	2.92	4.06	
Q	.015	.060	.38	1.52	3
s1	.005	—	.13	—	6
α	90°	105°	90°	105°	

- 1. Index area: a notch or a pin one identification mark shall be located adjacent to pin one. The manufacturer's identification shall not be used as pin one identification mark.
- 2. The minimum limit for dimension "b2" may be .023 (.58mm) for leads number 1, 8, 9 and 16 only.
- 3. Dimension "Q" shall be measured from the seating plane to the base plane.
- 4. This dimension allows for off-center lid, meniscus and glass overrun.
- The basic pin spacing is .100 (2.54mm) between centerlines. Each pin centerline shall be located within ±.010 (.25mm) of its exact longitudinal position relative to pins 1 and 16.
- 6. Applies to all four corners (leads number 1, 8, 9, and 16).
- 7. "eA" shall be measured at the center of the lead bends or at the centerline of the leads when " $\alpha$ " is 90°.
- 8. All leads Increase maximum limit by .003 (.08mm) measured at the center of the flat, when lead finish applied.
- 9. Fourteen spaces.







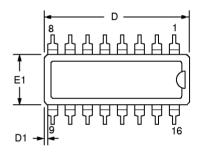
#### PRODUCT SPECIFICATION

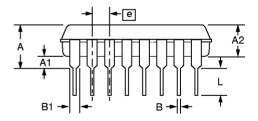
#### Mechanical Dimensions (continued)

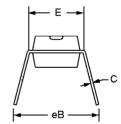
#### 16-Lead Plastic DIP Package

Cumhal	Inc	hes	Millim	neters	Notes
Symbol	Min.	Max.	Min.	Max.	Notes
A	_	.210	_	5.33	
A1	.015		.38	—	
A2	.115	.195	2.93	4.95	
В	.014	.022	.36	.56	
B1	.045	.070	1.14	1.78	
С	.008	.015	.20	.38	4
D	.745	.840	18.92	21.33	2
D1	.005		.13	_	
E	.300	.325	7.62	8.26	
E1	.240	.280	6.10	7.11	2
е	.100	BSC	2.54 BSC		
eB	l	.430		10.92	
L	.115	.160	2.92	4.06	
Ν	1	6	1	6	5

- 1. Dimensioning and tolerancing per ANSI Y14.5M-1982.
- "D" and "E1" do not include mold flashing. Mold flash or protrusions shall not exceed .010 inch (0.25mm).
- 3. Terminal numbers are shown for reference only.
- 4. "C" dimension does not include solder finish thickness.
- 5. Symbol "N" is the maximum number of terminals.



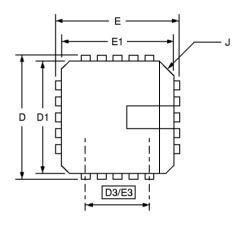


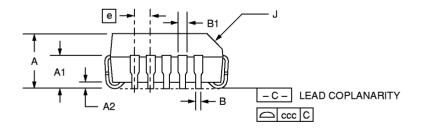


#### Mechanical Dimensions (continued)

#### 20-Lead PLCC Package

Symbol	Inc	hes	Millim	neters	Notes
Symbol	Min.	Max.	Min.	Max.	Notes
А	.165	.180	4.19	4.57	
A1	.090	.120	2.29	3.05	
A2	.020	_	.51		
В	.013	.021	.33	.53	
B1	.026	.032	.66	.81	
D/E	.385	.395	9.78	10.03	
D1/E1	.350	.356	8.89	9.04	3
D3/E3	.200	BSC	5.08 BSC		
е	.050	BSC	1.27	BSC	
J	.042	.048	1.07	1.22	2
ND/NE	5		5		
Ν	2	0	2	0	
CCC	_	.004	_	0.10	





- 1. All dimensions and tolerances conform to ANSI Y14.5M-1982
- 2. Corner and edge chamfer  $(J) = 45^{\circ}$
- Dimension D1 and E1 do not include mold protrusion. Allowable protrusion is .245" (.101mm)

#### **Ordering Information**

Product Number	Temperature Range	Screening	Package	Package Marking
TDC1044AB9C	0°C to 70°C	Commercial	16-Lead Ceramic DIP	1044 <b>A</b> B9C
TDC1044AN9C	0°C to 70°C	Commercial	16-Lead Plastic DIP	1044 <b>AN</b> 9C
TDC1044AR4C	0°C to 70°C	Commercial	20-Lead PLCC	1044 <b>A</b> R4C

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- 2. A critical component in any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

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