# Product **Document**

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### **Datasheet**

DS000511

# **AS5715A/AS5715R**

### **On-/Off-Axis Inductive Position Sensor with Sin/Cos Output**

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## **Content Guide**





## <span id="page-3-0"></span>**1 General Description**

The AS5715A/AS5715R is an inductive sensor IC with differential SIN/COS output for high speed position sensing applications. To fulfill its purpose, the AS5715A/AS5715R device has to be connected to a coil system. The coil system consists of a TX (transmitting) and two RX (receiving) coils typically integrated on a PCB (Printed Circuit Board). The device excites the TX coil with a high frequency voltage. A target above the coil system effects the amount of voltage induced into the RX coils depending on the position of the target. The device senses, demodulates, and amplifies this voltage. If If the coil system is designed properly, then the device output delivers SIN and COS shaped voltages. An ECU (Electronic Control Unit) can calculate the target position by applying an atan2() function to the SIN and COS output signal. The AS5715A/AS5715R is defined as SEooC (Safety Element out of Context) according ISO26262.

### 1.1 Key Benefits & Features

<span id="page-3-1"></span>The benefits and features of AS5715R, On-/Off-Axis Inductive Position Sensor with Sin/Cos Output, are listed below:

#### **Figure 1: Added Value of Using AS5715R**



### 1.2 Applications

<span id="page-3-2"></span>Typical applications are: BLDC motor control, traction motors for electric vehicles, electric power steering, brake boosters, replacement of expensive and bulky resolvers, mirror LIDAR and closed loop servo motor systems.

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### 1.3 Block Diagram

<span id="page-4-0"></span>The functional blocks of this device are shown below:

**Figure 2:** 

**Functional Blocks of AS5715R** 



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## <span id="page-5-0"></span>**2 Ordering Information**



The AS5715A is a preprogrammed derivative of the AS5715R. This means that the AS5715A is not programmable. The table below shows how the registers of the AS5715A are programmed:



The tolerance of the preprogrammed AS5715A gain factor is  $\pm$  1dB.

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## <span id="page-6-0"></span>**3 Pin Assignment**

### 3.1 Pin Diagram

<span id="page-6-1"></span>**Figure 3:** 

**TSSOP-14 Pin Assignment** 



### 3.2 Pin Description

#### <span id="page-7-0"></span>**Figure 4:**

**Pin Description of AS5715R (TSSOP14)** 



(1) Explanation of abbreviations:

DI Digital Input<br>DIO Digital Input

DIO Digital Input/Output<br>Al Analog Input

AI Analog Input<br>AO Analog Outpu

AO Analog Output<br>S Supply Pin Supply Pin

## <span id="page-8-0"></span>**4 Absolute Maximum Ratings**

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated under "Operating Conditions" is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### **Figure 5**

#### **Absolute Maximum Ratings of AS5715R**



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(1) The reflow peak soldering temperature (body temperature) is specified according to IPC/JEDEC J-STD-020E "Moisture/Reflow Sensitivity Classification for Nonhermetic Solid State Surface Mount Devices." The lead finish for Pbfree leaded packages is "Matte Tin" (100% Sn)

### <span id="page-10-0"></span>**5 Electrical Characteristics**

All limits are guaranteed. The parameters with Min and Max values are guaranteed with production tests or SQC (Statistical Quality Control) methods.

All in this datasheet defined tolerances for external components need to be assured over the whole operation conditions range and also over lifetime.

<span id="page-10-1"></span>Overall condition:  $T_A = -40$  °C to 160 °C, VCC = 4.5 V to 5.5 V; components specification; unless otherwise noted.

### 5.1 Power Supply

**Figure 6: Power Supply** 



### 5.2 Analog Signal Chain

#### <span id="page-11-0"></span>**Figure 7:**

**Analog Signal Chain** 



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(1) This error includes noise and gain mismatch as main error contributor but also errors caused by other effects. The error caused by noise can be decreased by an input signal with higher amplitude. The error caused by gain mismatch can be reduced by gain error compensation in the ECU. Guaranteed between 10% to 90% of VCC peak to peak differential output.

(2) The parameter RX\_ERR<sub>UNCOMP</sub> specifies the overall uncompensated angular error produced by the IC only, assuming ideal RX input signals. This parameter includes INL<sub>RX</sub>, Delay\_error, V<sub>OFFSET\_RX</sub>, OPN and gain\_mm.

<span id="page-12-0"></span>(3) Typical calculation:  $f_{IN} = (rotation\_speed_in\_rpm / 60) * num_of_pole_pairs$ 

### 5.3 LC - Oscillator

**Figure 8: LC - Oscillator** 



### 5.4 Analog Output

#### <span id="page-13-0"></span>**Figure 9:**

**Analog Output** 



(1) Guaranteed by design

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- (2) The parameter VOUT specifies the general possible output voltage range under the assumption, that the ECU is detecting a voltage level out of this range as a failureband. The final amplitude of the output voltage signal depends only on the RX input signal and the AGC gain factor and should fall into this specified range.
- (3) The parameter  $I_{\text{OUTHIGH-Z}}$  specifies the maximum output leakage current at the output pin in high-Z configuration to fulfill the 96% of VCC level, assuming minimum VCC and minimum external pull-up. In the end application, the output level clamping high depends on the external pull-up resistor  $(R_{PU})$ , which must be connected between each output pin and the VCC level.

The voltages  $V_{SIN+}$ ,  $V_{SIN-}$ ,  $V_{COS+}$  and  $V_{COS-}$  are measured from the pins SIN+, SIN-, COS+ and COS- to GND. OP refers to  $OP_{SIN}$  and  $OP_{COS}$ .

$$
OP_{SIN} = \frac{V_{SIN+} + V_{SIN-}}{2} \approx 2.5 \text{ V}
$$
\n
$$
OP_{COS} = \frac{V_{COS+} + V_{COS-}}{2} \approx 2.5 \text{ V}
$$
\n
$$
OP_{SIN\%} = \frac{OP_{SIN}}{VCC} \times 100 \approx 50 \text{ % of VCC}
$$
\n
$$
OP_{COS\%} = \frac{OP_{COS}}{VCC} \times 100 \approx 50 \text{ % of VCC}
$$
\n
$$
OP_{SIN} \neq OP_{COS\%}
$$
\n
$$
OP_{SIN\%} \neq OP_{COS\%}
$$

$$
\frac{|OP_{SIN}-OP_{COS}|}{VCC} * 100 < OP_{DIFF}
$$

### 5.5 Power Management

#### <span id="page-15-0"></span>**Figure 10:**

**Power Management** 



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### 5.6 Off-Chip Components

#### <span id="page-17-0"></span>**Figure 11:**

#### **Off-Chip Components**



 $(1)$  Due to the specified output capacitor and pull-up resistor range, full  $l^2C$  specification compliance is not quaranteed. (2) There is no max limit specified, as a higher cap value always leads to better stabilization behavior, but with increasing cap value, a longer startup time must be taken into account. After reaching the internal POR level (~2.8 V), the outputs remain per default in high-Z configuration for 3 ms. If the nominal VCC level is reached within this 3 ms, the outputs are already settled. If the startup takes longer than 3 ms, the output signals will provide immediately the SIN/COS signal according to the RX input signal multiplied with the gain factor. With active AGC regulation, where the thresholds are

#### **Information**

All specified tolerances for external components need to be assured over the whole operation conditions range and also over lifetime.

ratiometric to VCC, a regulation can still occur, as long as the final VCC level is not reached.

## <span id="page-18-0"></span>**6 Functional Description**

<span id="page-18-1"></span>The AS5715A/AS5715R is an inductive-based rotary or linear position sensor using CMOS technology.

### 6.1 Power Supply

#### **6.1.1 5 V VCC Supply**

The device has an integrated overvoltage and undervoltage detection. [Figure 12](#page-18-2) describes the transition behavior between overvoltage, undervoltage and operational mode condition. An undefined zone is not possible, the device is either in operational mode or in safe mode.

<span id="page-18-2"></span>**Figure 12: Reset Thresholds for 5 V VCC Supply** 



The internal LDO block regulates the 5 V supply voltage (VCC) down to a 3.3 V level  $(V_{\text{LDO}})$ .



### 6.2 Oscillator

<span id="page-19-0"></span>The AS5715A/AS5715R device and an external LC-tank form an LC-oscillator. The LC-tank consists of the components LTX, CLC1 and CLC2. The minimum and maximum specified range of LTX and CLC-TANK must not be exeeded. Furtermore the parasitic components of L<sub>TX</sub> and C<sub>LC-TANK</sub> must be within the specified range.

**Figure 13: TX Resonator Signal**  **Figure 14: LC-Tank** 



<span id="page-19-1"></span>**Equation 1:** 

$$
C_{LC-TANK} = \frac{C_{LC1} * C_{LC2}}{C_{LC1} + C_{LC2}}
$$

<span id="page-19-2"></span>**Equation 2:** 

$$
f_{TX} = \frac{1}{2 * \pi * \sqrt{L_{TX} * C_{LC-TANK}}}
$$

The oscillation frequency of the LC-oscillator is  $f_{TX}$  and depends on the LC-tank. [Equation 1](#page-19-1) and [Equation 2](#page-19-2) show how to calculate  $f_{TX}$ . Note that  $f_{TX}$  has to be within the specified range.

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### 6.3 Automatic Gain Control (AGC)

<span id="page-20-0"></span>By default, the OTP register *AGC\_disable* is set to "0", therfore the internal AGC algorithm is active and sets the gain for the two AM moduleted RX signals (SIN and COS) as described in [Figure 15.](#page-21-0) Depending on the coil system the AM modulated RX signal may have a high initial offset coming from the coil system. This offsets may result in frequent gain changes over an electrical 360 deg rotation. EMC events that change voltages on the outputs may change the gain and may cause that the device does permanently change it's currently used gain until the next power on reset.

Alternativly *AGC\_disable* can be set to "1". Then the device permanently operates with the gain selected by the bits *AGC\_gain\_factor <6:0>*.

The startup procedure of the AGC is described in [Figure 15.](#page-21-0) This startup procedure is executed each time after the device is powered on.

#### **dis\_AGC\_boost**

When the *dis* AGC boost bit is "0" then the device does increase the gain until one of the differential outputs (SIN or COS) does reach a positive or negative voltage of  $V_{AGC}$  REG  $H/2$  and then does additionaly increase the gain by 4 steps. The maximum possible gain is 109 steps.

#### **dis\_AGC\_3dBred**

When the *dis* AGC 3dBred bit is "0" then gain is reduced by 18 steps. The purpose of this reduction is that the gain does not change when the target is rotated after the startup procedure, assuming that the differential output signals have no offset and that the distance between the coil sytem and the target stays constant. Without this gain reduction the gain would change after the completion of the startup procedure when the target is rotated and when the target position before the startup is not 0, 90, 180 or 270 degree. The minimum possible gain is 0 steps.

#### **gain\_freeze**

When the gain freeze bit is "0" then the device does operate in fixed gain operation after the startup procedure. The gain will not change until the next startup.

#### **Normal AGC algorithm regulation**

The normal AGC algoritm regulation is described in [Figure 16](#page-22-0) and [Figure 17.](#page-22-1) The logic gates decide by how many counts the gain will be increased or decreased. A logic "1" at the outputs of the gates causes a gain increase or decrease. A logic "0" at the outputs does not cause any gain change. The gain always changes for both channels together. The inputs of the gates are connected to comparators that compare the absolute value of the differential SIN and COS output signals to four different comperator levels. This comperator levels are derived from VAGC\_LIMIT\_L, VAGC\_REG\_L, VAGC REG H and VAGC LIMIT H. The regulation mode and step mode in [Figure 16](#page-22-0) and [Figure 17](#page-22-1) is executed periodically with the times tree and tLIMIT.



#### <span id="page-21-0"></span>**Figure 15: AGC Startup Procedure**





#### <span id="page-22-0"></span>**Figure 16:**

**Digital Implementation of Regulation Mode** 



<span id="page-22-1"></span>**Figure 17: Digital Implementation of Step Mode** 





#### **Figure 18:**

**AGC Regulation and Step Mode** 



#### **Information**

- Whenever the AS5715R moves into safe state (valid for all safety mechanism initiations except the OTP signature check), the AGC gain freezes (assuming that the AGC algorithm is active) for this safe state condition duration.
- As soon as the safe state condition expires and the AS5715R recovers into normal operation mode, the AGC gain unfreezes and it starts regulating with the same gain value as right before entering the safe state condition.
- If the AGC algorithm is disabled, a well defined fixed gain factor must be chosen with respect to the expected maximum RX signal amplitude occurring in certain applications. If the gain factor is set too high, the RX input signal get over-amplified and starts to clip, which may lead into a wrong output signal.

### <span id="page-24-0"></span>6.4 Output

#### **6.4.1 Output Signals**

The differential SIN output signal is the voltage between the pins SIN+ and SIN-. The differential COS output signal is the voltage between the pins COS+ and COS-.

[Figure 19](#page-24-1) and [Figure 20](#page-25-0) show an example of the SIN+ and SIN- signal measured signgle ended (against GND):

- Red signal with peak2peak amplitude of 4.5% of VCC
- Yellow signal with peak2peak amplitude of 17.5% of VCC
- Green signal with peak2peak amplitude of 35% of VCC
- **●** Blue signal with peak2peak amplitude of 45% of VCC

Beside this, the failureband is visible at the low and high side of the output voltage range.

<span id="page-24-1"></span>**Figure 19 : SIN+ Single-Ended Output Signal** 





<span id="page-25-0"></span>**Figure 20 : SIN- Single-Ended Output Signal** 



The corresponding differential SIN output signals are drawn in the figure below.

The differential signals are calculated from the single-ended signals (SIN+ minus SIN-):

- Red SIN signal  $\rightarrow$  peak2peak amplitude of 9% of VCC  $\rightarrow$  VAGC LIMIT L threshold
- Yellow SIN signal  $\rightarrow$  peak2peak amplitude of 35% of VCC  $\rightarrow$  V<sub>AGC\_REG\_L</sub> threshold
- Green SIN signal  $\rightarrow$  peak2peak amplitude of 70% of VCC  $\rightarrow$  V<sub>AGC</sub> REG H threshold
- Blue SIN signal  $\rightarrow$  peak2peak amplitude of 90% of VCC  $\rightarrow$  V<sub>AGC</sub> LIMIT H threshold



**Figure 21 : SIN Differential Output Signal** 



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### 6.5 Diagnostic and Functional Safety

<span id="page-27-0"></span>AS5715A/AS5715R can be used in safety critical applications. For this reason, AS5715A/AS5715R is developed as SEooC (Safety Element out of Context) according the ISO26262, with assumed safety goals and assumed ASIL level.

The assumption of use (AoU) and the embedded self-diagnostic, to achieve particular ASIL level in the application, are described in the AS5715A/AS5715R safety manual.

For additional information regarding the ISO26262 flow at ams and the SEooC relevant documents (e.g. FMEDA, safety manual) please contact the technical support of ams.

**Figure 22: Diagnostic Table** 





#### **Figure 23: Safe State Definition**



(1) Enable/disable by customer OTP bit setting

The PWM frequency is 500 Hz. The PWM voltage may not be rectangular, depending on the used pullup resistor and output capacitor.

e.g. PWM (20:80) stands for 20% high level and 80% low level

### 6.6 I<sup>2</sup>C Interface and Communication Procedure

<span id="page-29-0"></span>The AS5715R will start up in operational mode after powering up. In operational mode the 4 analog outputs of the device are operational.

The AS5715R features and OTP (One Time Programmable) memory. This memory can be used to programm custom settings. The OTP can be programmed over an I<sup>2</sup>C interface which is applicable at the pins SIN+/SDA and COS+/SCL. To activate the I<sup>2</sup>C interface, the so called customer rma procedure is performed. The customer rma procedure allows to switch the analog outputs into I<sup>2</sup>C mode in order to allow programming over I<sup>2</sup>C. To switch the outputs back into operational mode without burning the OTP, the pass2funct command can be used.

#### **6.6.1 Customer rma Procedure**

- **1.** On the VCC pin, an overvoltage condition must be applied by supplying with a voltage between 5.95 V and 20 V.
- **2.** The OV detection takes ~50 µs followed by a 1 ms wait time. After this, a 1 ms window is open to execute the customer rma procedure.
- **3.** To execute the customer rma procedure it is necessary to apply 2 pulses (= 2 x rising edge) on  $SN+/SDA$  and independent from that, 16 pulses (= 16 x rising edge) on  $COS+/SCL$  during the rma window. The pulses must be applied with a frequency <1 MHz.
- **4.** After this, the I<sup>2</sup>C port is open and the VCC supply voltage must be reduced to the operationg voltage range before I<sup>2</sup>C communication is possible.

#### **Figure 24: Customer rma Procedure**



#### **6.6.2 pass2funct Command**

The pass2funct command is usefull to check the effect of changed memory settings before burning them into the OTP. The pass2funct switches the device from I<sup>2</sup>C mode into functional mode without the need to perform a power on reset or burning the OTP.



If a register bit gets changed, it is important also to adjust the signature byte as well. Even if a correct signature is written, the signature error is triggered, because the error is latched. To clean the signature error it's necessary to write dsp\_rst before performing a pass2funct command.

The pass2funct is executed when the register **PASS2FUNCT** (address f1h) is written with the value 01010011b (53h). After the pass2funct command, the I<sup>2</sup>C interface is disabled and the only way to enable again the communication mode is an execution of the customer rma procedure.

#### **6.6.3 Customer Signature Calculation**

After the desired bit settings of the customer OTP registers **P2RAM\_BYTE\_13**, **P2RAM\_BYTE\_14**  and **P2RAM\_BYTE\_15**, the correct setting of the customer signature bits must be applied into register **P2RAM\_BYTE\_16.** The OTP of AS5715R uses a BIST technique with Multiple Input Signature Register circuits. To activate this Built-In-Self-Test, a calculation of the signature byte is necessary which has to be stored in the OTP during programming. For calculating the signature byte, the content of the whole memory has to be read out. Out of this information, the following calculation has to be done.

```
@content =($byte0, $byte1, $byte2, $byte3, $byte4, $byte5, $byte6, $byte7, $byte8, 
$byte9, $byte10, $byte11, $byte12, $byte13, $byte14, $byte15);
```

```
$missr = 0;for($i=12; $i< 15; $i++) { 
$missr shift = ($missr<<1);
$misr xor = ($misr shift \land $content[$i]) & 0xFF;
$missr_msb = $missr/(128);if ($misr_msb eq 0) { 
$misr = $misr_xor; 
} 
else 
{ 
$missr = ($missr_xor ^ 29) & 0xFF;}
```
#### **6.6.4 Burning Procedure for Customer OTP Section**

<span id="page-30-0"></span>Before to activate the burning procedure of the customer OTP section, the customer has to write first the unlock command 01h to the register **P2RAM\_CONTROL** (address 03h), otherwise the burning function is locked. Now the OTP is ready for burning.

The BURN command has to be triggered by writing 08h to the register **P2RAM\_CONTROL**. The status of the burning procedure can be seen by polling the BURN bit. During burn operation it stays high and get released to 0 after the burning procedure has finished. The content of the register **P2RAM\_CONTROL** will be 09h during burning and 01h when burning procedure is finished.

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To check the quality of the burned fuses, right after the burning process, the whole customer OTP section (register **P2RAM\_BYTE\_13** up to register **P2RAM\_BYTE\_16**) should be set to 00h. Using the LOAD operation by writing 04h to the register **P2RAM\_CONTROL** will load the content of the burned fuses into the customer OTP section and the correct content can be checked.

The same routine has to be done with the GLOAD command!

Set the whole customer OTP section (register **P2RAM\_BYTE\_13** up to register **P2RAM\_BYTE\_16**) to 00h again. Use the GLOAD + LOAD operation by writing 44h to the register **P2RAM\_CONTROL** and this will load again the content of the burned fuses into the customer OTP section and check, if the content is correct.



#### **Information**

GLOAD Test:

Restricted to temperature range: 25 °C ± 20 °C Right after the programming procedure (max. 1 hour with same conditions 25 °C  $\pm$  20 °C), same VCC voltage. The GLOAD test is only for the verification of the burned OTP fuses during the programming sequence. A use of the GLOAD in other cases is not allowed.



#### **6.6.5 Programming Flowchart**

#### **Figure 25:**

**Programming Flowchart** 



## <span id="page-33-0"></span>**7 Register Description**

### 7.1 Register Overview

<span id="page-33-1"></span>**Figure 26: Register Overview** 



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### <span id="page-34-0"></span>7.2 Detailed Register Description

### **7.2.1 P2RAM\_BYTE\_13 Register (Address 1ch)**



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### **7.2.2 P2RAM\_BYTE\_14 Register (Address 1dh)**



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### **7.2.3 P2RAM\_BYTE\_15 Register (Address 1eh)**



#### **7.2.4 P2RAM\_BYTE\_16 Register (Address 1fh)**



### **7.2.5 AGC\_VALUE Register (Address f2h)**

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## <span id="page-37-0"></span>**8 Functional Safety**

### 8.1 Safety Manual

<span id="page-37-1"></span>The Safety Manual, available upon request, contains all the necessary information for the system integrator, to integrate AS5715A/AS5715R in a safety related item.

The sensor is supporting the ISO26262 as Safety Element out of Context (SEooC).

The Safety Manual includes the following information:

- **●** Product development lifecycle
- **•** Description of the technical safety concept on system level
- Detailed information of Assumption of Use of the element with respect to its intended use, which includes
	- **●** System Safe State information
	- **●** Fault Tolerant Time Interval
	- **●** Coverage information

As part of the Safety Manual, the Verification and Safety Analysis Report includes following information:

- **HW architectural metric results (Single Point Fault Metric)**
- Description of verifications based on the ISO26262
- **●** Detailed FMEDA

## <span id="page-38-0"></span>**9 Application Information**

### 9.1 Typical Application Circuits

<span id="page-38-1"></span>**Figure 27:** 

**Typical Differential Output Application for 5 V VCC Supply System** 



**Figure 28:** 

**Typical Single Ended Output Application for 5 V VCC Supply System** 





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## <span id="page-39-0"></span>**10 Package Drawings & Markings**

#### **Figure 29:**

**TSSOP14 Package Outline Drawing** 



(1) All dimensions are in millimeters. Angles in degrees.

- (2) Dimensioning and tolerancing conform to ASME Y14.5-2009.
- (3) N is the total number of terminals.



**Figure 30: TSSOP14 Package Marking/Code for AS5715R** 



- YY Manufacturing Year<br>WW Manufacturing Week WW Manufacturing Week<br>M Assembly Plant Ident M Assembly Plant Identifier<br>ZZ Assembly Traceability Co  $ZZ$  Assembly Traceability Code<br>  $\omega$  Sublot Identifier
	- Sublot Identifier

**Figure 31: TSSOP14 Package Marking/Code for AS5715A** 



YY Manufacturing Year<br>WW Manufacturing Week WW Manufacturing Week<br>M Assembly Plant Ident M Assembly Plant Identifier ZZ Assembly Traceability Code @ Sublot Identifier

## <span id="page-41-0"></span>**11 Revision Information**





● Page and figure numbers for the previous version may differ from page and figure numbers in the current revision.

• Correction of typographical errors is not explicitly mentioned.

## <span id="page-42-0"></span>**12 Legal Information**

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