

LOW DROP DUAL POWER OPERATIONAL AMPLIFIERS

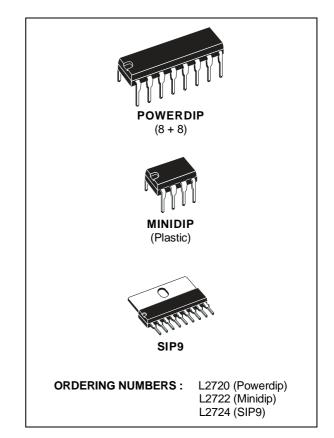
- OUTPUT CURRENT TO 1 A
- OPERATES AT LOW VOLTAGES
- SINGLE OR SPLIT SUPPLY
- LARGE COMMON-MODE AND DIFFEREN-TIAL MODE RANGE
- LOW INPUT OFFSET VOLTAGE
- GROUND COMPATIBLE INPUTS
- LOW SATURATION VOLTAGE
- THERMAL SHUTDOWN
- CLAMP DIODE

DESCRIPTION

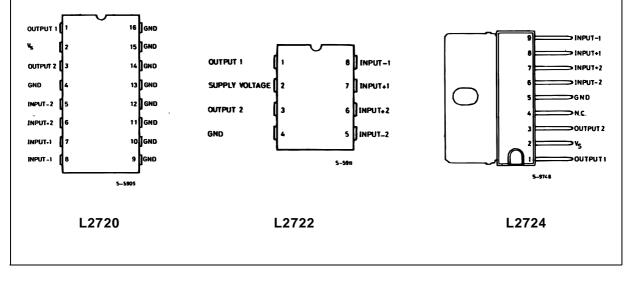
The L2720, L2722 and L2724 are monolithic integrated circuits in powerdip, minidip and SIP-9 packages, intended for use as power operational amplifiers in a wide range of applications including servo amplifiers and power supplies.

They are particularly indicated for driving, inductive loads, as motor and finds applications in compactdisc VCR automotive, etc.

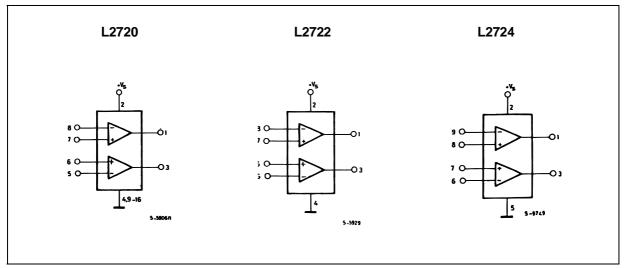
The high gain and high output power capability provide superior performance whatever an operational amplifier/power booster combination is required.



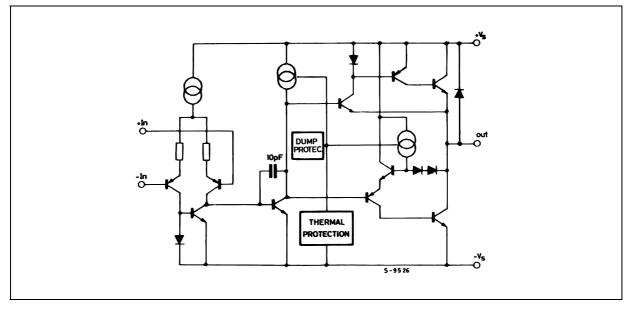
PIN CONNECTIONS (top views)



BLOCK DIAGRAM



SCHEMATIC DIAGRAM (one section)



ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit	
Vs	Supply Voltage	28	V	
Vs	Peak Supply Voltage (50ms)	50	V	
Vi	Input Voltage	Vs		
Vi	Differential Input Voltage	±Vs		
lo	DC Output Current	1	Α	
Ι _p	Peak Output Current (non repetitive)	1.5	Α	
P _{tot}	Power Dissipation at $T_{amb} = 80^{\circ}C$ (L2720), $T_{amb} = 50^{\circ}C$ (L2722) $T_{case} = 75^{\circ}C$ (L2720) $T_{case} = 50^{\circ}C$ (L2724)	1 5 10	W	
T _{stg} , T _j	Storage and Junction Temperature	-40 to 150	°C	

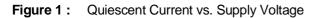
THERMAL DATA

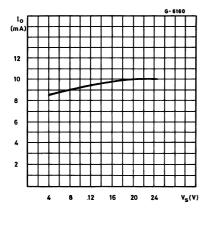
			SIP-9	Powerdip	Minidip
R _{th j-case}	Thermal Resistance Junction-case	Max.	10°C/W	15°C/W	70°C/W
R _{th j-amb}	Thermal Resistance Junction-ambient	Max.	70°C/W	70°C/W	100°C/W

ELECTRICAL CHARACTERISTICS

 $V_s = 24V$, $T_{amb} = 25^{\circ}C$ unless otherwise specified

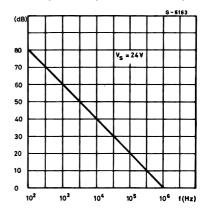
Symbol	Parameter Test Conditions				Тур.	Max.	Unit
Vs	Single Supply Voltage		4		28	V	
Vs	Split Supply Voltage			± 2		± 14	V
I _s	Quiescent Drain Current	$V_o = \frac{V_s}{2}$	$V_s = 24V$		10	15	mA
			$V_s = 8V$		9	15	
l _b	Input Bias Current				0.2	1	μA
Vos	Input Offset Voltage					10	mV
l _{os}	Input Offset Current					100	nA
SR	Slew Rate				2		V/μs
В	Gain-bandwidth Product				1.2		MHz
Ri	Input Resistance			500			kΩ
Gv	O.L. Voltage Gain	f = 100Hz f = 1kHz		70	80 60		dB
e _N	Input Noise Voltage	B = 22Hz to $22kHz$			10		μV
I _N	Input Noise Voltage				200		pА
CMR	Common Mode Rejection	f = 1kHz		66	84		dB
SVR	Supply Voltage Rejection	$ f = 100Hz \\ R_G = 10k\Omega \\ V_R = 0.5V $	$V_s = 24V$ $V_s = \pm 12V$ $V_s = \pm 6V$	60	70 75 80		dB
V _{DROP(HIGH)}		$V_s = \pm 2.5V$ to $\pm 12V$	$I_p = 100mA$ $I_p = 500mA$		0.7 1	1.5	V
V _{DROP(LOW)}		$V_s = \pm 2.5V$ to $\pm 12V$	$I_p = 100 \text{mA}$ $I_p = 500 \text{mA}$		0.3 0.5	1	V
Cs	Channel Separation		$V_s = 24V$ $V_s = 6V$		60 60		dB
T_{sd}	Thermal Shutdown Junction Temperature				145		°C





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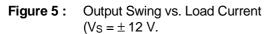
Figure 2: Open Loop Gain vs. Frequency



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V_S=24V (dB) πш 90 ┼┼╀ 70 Ш 50 30 Ш 10² 10³ 10 10⁴ f (Hz)

Figure 3 : Common Mode Rejection vs. Frequency



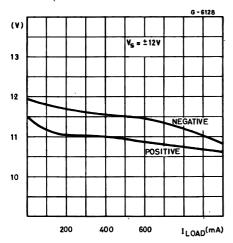


Figure 7 : Channel Separation vs. Frequency

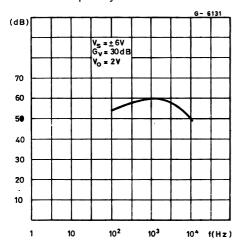


Figure 4 : Output Swing vs. Load Current (Vs = \pm 5 V.

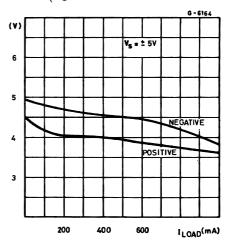
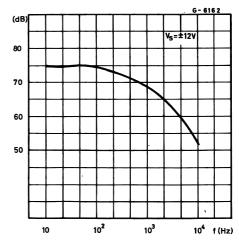


Figure 6 : Supply Voltage rejection vs. Frequency

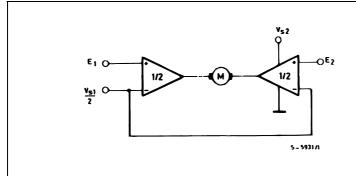


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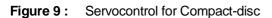
APPLICATION SUGGESTION

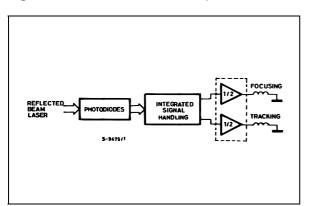
In order to avoid possible instability occuring into final stage the usual suggestions for the linear power stages are useful, as for instance :

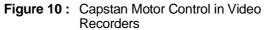
- layout accuracy ;
- A 100nF capacitor connected between supply _ pins and ground;



- Bidirectional DC Motor Control with µP Compatible Inputs Figure 8 :
- boucherot cell (0.1 to 0.2 μF + 1 Ω series) between outputs and ground or across the load. With single supply operation, a resistor $(1k\Omega)$ between the output and supply pin can be necessary for stability.







V_{S1} = logic supply voltage

Must be V_{S2} > V_{S1}

E1, E2 = logic inputs

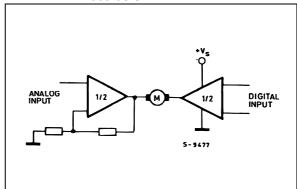


Figure 11: Motor Current Control Circuit

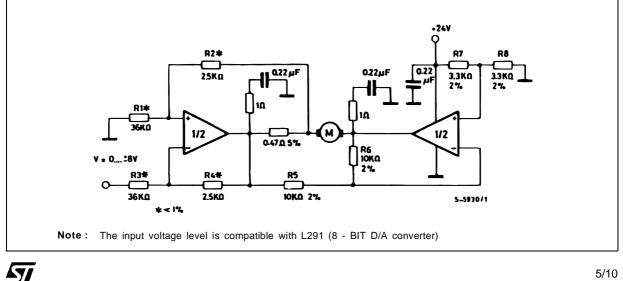


Figure 12: Bidirectional Speed Control of DC Motors

For circuit stability ensure that $R_X > \frac{2R3 \cdot R1}{RM}$ where R_M = internal resistance of motor. The voltage available at the terminals of the motor is $V_M = 2(V_I - \frac{V_S}{2}) + |R_O| \cdot I_M$ where $|R_O| = \frac{2R3 \cdot R1}{R_X}$

and I_M is the motor current.

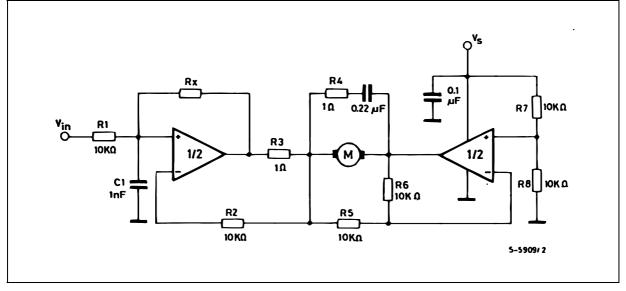
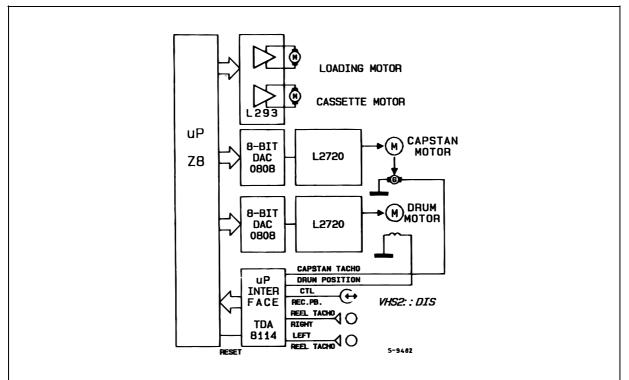
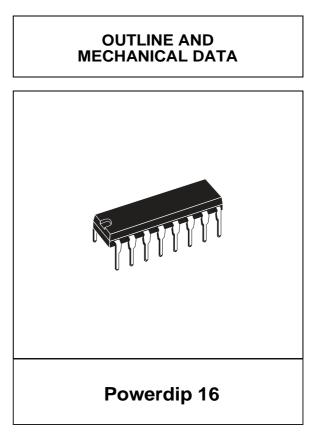
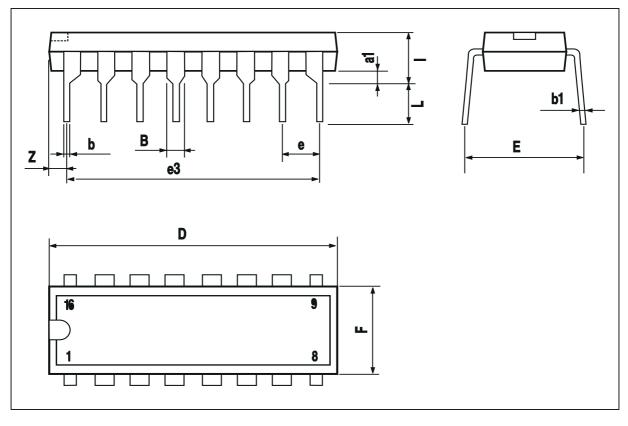


Figure 13: VHS-VCR Motor Control Circuit



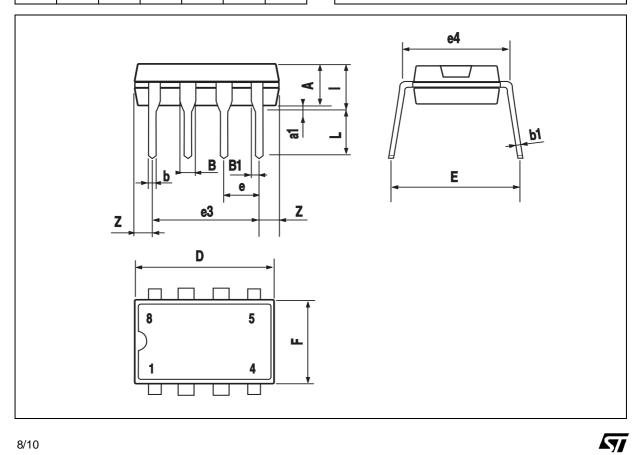
DIM.		mm			inch	
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
a1	0.51			0.020		
В	0.85		1.40	0.033		0.055
b		0.50			0.020	
b1	0.38		0.50	0.015		0.020
D			20.0			0.787
Е		8.80			0.346	
е		2.54			0.100	
e3		17.78			0.700	
F			7.10			0.280
I			5.10			0.201
L		3.30			0.130	
Z			1.27			0.050





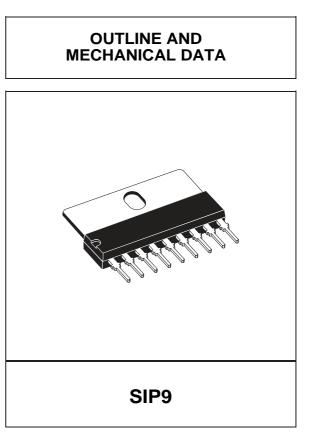
DIM.	mm			inch			
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
А		3.32			0.131		
a1	0.51			0.020			
В	1.15		1.65	0.045		0.065	
b	0.356		0.55	0.014		0.022	
b1	0.204		0.304	0.008		0.012	
D			10.92			0.430	
Е	7.95		9.75	0.313		0.384	
е		2.54			0.100		
e3		7.62			0.300		
e4		7.62			0.300		
F			6.6			0.260	
I			5.08			0.200	
L	3.18		3.81	0.125		0.150	
Z			1.52			0.060	

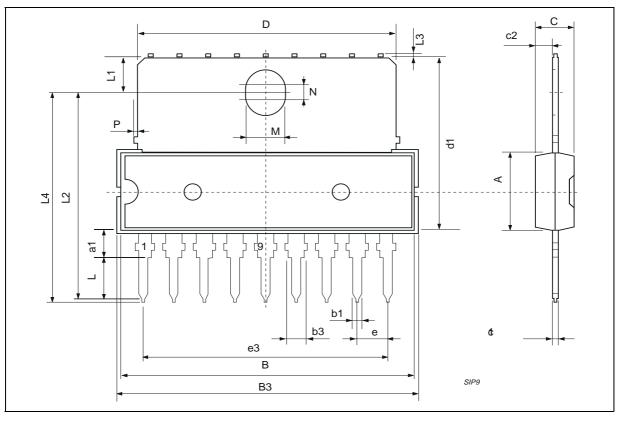
OUTLINE AND **MECHANICAL DATA** Minidip



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DIM.		mm			inch	
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
А			7.1			0.280
a1	2.7		3	0.106		0.118
В			23			0.90
B3			24.8			0.976
b1		0.5			0.020	
b3	0.85		1.6	0.033		0.063
С		3.3			0.130	
c1		0.43			0.017	
c2		1.32			0.052	
D			21.2			0.835
d1		14.5			0.571	
е		2.54			0.100	
e3		20.32			0.800	
L	3.1			0.122		
L1		3			0.118	
L2		17.6			0.693	
L3			0.25			0.010
L4	17.4		17.85	0.685		0,702
М		3.2			0.126	
Ν		1			0.039	
Р			0.15			0.006





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