

PFE3000-12-069RA

3000 W AC-DC Front-End Power Supply

The **PFE3000-12-069RA** is a 3000 Watt AC/DC power-factor-corrected (PFC) and DC-DC power supply that converts standard AC mains power or high voltage DC bus voltages into a main output of 12 VDC for powering intermediate bus architectures (IBA) in high performance and reliability servers, routers, and network switches.

The PFE3000-12-069RA meets international safety standards and displays the CE-Mark for the European Low Voltage Directive (LVD).



Key Features & Benefits

- Best-in-class, Platinum efficiency
- Wide input voltage range: 90-300 VAC
- AC input with power factor correction
- DC input voltage range: 192-400 VDC
- Hot-plug capable
- Parallel operation with active current sharing thru analog bus
- Full digital controls for improved performance
- High density design: 30.5 W/in³
- Small form factor: 555 x 69 x 42 mm (21.85 x 2.72 x 1.65 in)
- I2C communication interface with Power Management Bus protocol for monitoring, control, and firmware update via bootloader
- Overtemperature, output overvoltage and overcurrent protection
- RoHS Compliant
- 2 Status LEDs: AC OK and DC OK with fault signaling
- Safety-approved to IEC/EN 60950-1 and UL/CSA 60950-1 2nd ed.
- US Patent Pending

Applications

- High Performance Servers
- Routers
- Switches



1. ORDERING INFORMATION

PFE	3000		12		069	R	Α	Option Code
Product Family	Power Level	Dash	V1 Output	Dash	Width	Airflow	Input	Blank: Standard model
PFE Front-Ends	3000 W		12 V		69 mm	R: Reversed ¹	A: AC	\$366: Screw for Key-in feature is installed.

Front to Rear

2. OVERVIEW

The PFE3000-12-069RA is a fully DSP controlled, highly efficient front-end power supply. It incorporates resonant-soft-switching technology and interleaved power trains to reduce component stresses, providing increased system reliability and very high efficiency. With a wide input operating voltage range and minimal linear derating of output power with respect to ambient temperature, the PFE3000-12-069RA maximizes power availability in demanding server, switch, and router applications. The power supply is fan cooled and ideally suited for server integration with a matching airflow path.

The PFC stage is digitally controlled using a state-of-the-art digital signal processing algorithm to guarantee best efficiency and unity power factor over a wide operating range.

The DC-DC stage uses soft switching resonant techniques in conjunction with synchronous rectification. An active OR-ing device on the output ensures no reverse load current and renders the supply ideally suited for operation in redundant power systems. The always-on +12V standby output provides power to external power distribution and management controllers. Its protection with an active OR-ing device provides for maximum reliability.

Status information is provided with front-panel LEDs. In addition, the power supply can be monitored and controlled (i.e. fan speed setpoint) via I²C communication interface with Power Management Bus protocol. It allows full monitoring of the supply, including input and output voltage, current, power, and inside temperatures. The same I²C bus supports the bootloader to allow field update of the firmware in the DSP controllers.

Cooling is managed by a fan, controlled by the DSP controller. The fan speed is adjusted automatically depending on the actual power demand and supply temperature and can be overridden through the I²C bus.

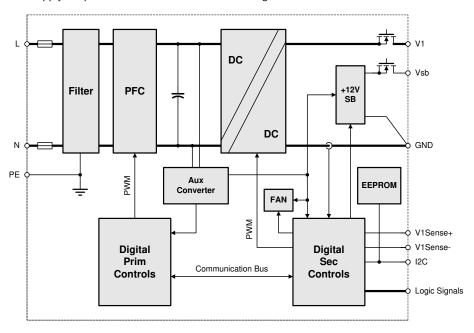


Figure 1 - PFE3000-12-0069RA Block Diagram



3. ABSOLUTE MAXIMUM RATINGS

Stresses in excess of the absolute maximum ratings may cause performance degradation, adversely affect long-term reliability, and cause permanent damage to the supply.

PARAMETER		CONDITIONS / DESCRIPTION	MIN	MAX	UNITS
Vi maxc	Maximum Input	Continuous		300	VAC

4. INPUT

General Condition: $T_A = 0...45$ °C unless otherwise noted.

PARAME	TER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
V_{inom}	AC Nominal Input Voltage		100	230	277	VAC
V i	AC Input Voltage Ranges	Normal operating (V_{imin} to V_{imax})	90		300	VAC
Vinom DC	DC Nominal input voltage		240		380	VDC
V_{iDC}	DC Input voltage ranges	Normal operating ($V_{i min}$ to $V_{i max}$)	192		400	VDC
Vi red	Derated Input Voltage Range	See Figure 20 and Figure 33	90		180	VAC
I _{i max}	Max Input Current	V _i > 200 VAC, >100 VAC			17	A_{rms}
I_{ip}	Inrush Current Limitation	$V_{i min}$ to $V_{i max}$, 0 ° T_{NTC} = 25°C (Figure 5)			50	A_p
F _i	Input Frequency		47	50/60	63	Hz
PF	Power Factor	<i>Vi nom</i> , 50Hz, > 0.3 <i>I</i> 1 nom	0.96			W/VA
Vi on	Turn-on Input Voltage ²	Ramping up	80		87	VAC
V _{i off}	Turn-off Input Voltage ²	Ramping down	73		85	VAC
		V_{nom} , $0.1 \cdot I_{\text{x nom}}$, $V_{\text{x nom}}$, $T_{\text{A}} = 25^{\circ}\text{C}$	90.0	91.85		
,	Efficiency without Fan	$V_{\text{i nom}}$, 0.2· $I_{\text{x nom}}$, $V_{\text{x nom}}$, $T_{\text{A}} = 25^{\circ}\text{C}$	93.0	94.40		%
η	Efficiency without Fair	$V_{1 \text{ nom}}$, $0.5 \cdot I_{2 \text{ nom}}$, $V_{2 \text{ nom}}$, $T_{A} = 25^{\circ}\text{C}$	94.5	94.95		70
		$V_{1 \text{ nom}}$, $I_{2 \text{ nom}}$, $V_{2 \text{ nom}}$, $T_{A} = 25^{\circ}\text{C}$	93.0	93.75		
T _{hold}	Hold-up Time	After last AC zero point, $V_1 > 10.8 \text{ V}$, V_{SB} within regulation, $V_2 = 230 \text{ VAC}$, $P_{x \text{ nom}}$	12			ms

² The Front-End is provided with a minimum hysteresis of 3 V during turn-on and turn-off within the ranges

4.1 INPUT FUSE

Quick-acting 25 A input fuses (6.3×32 mm) in series with both the L- and N-line inside the power supply protect against severe defects. The fuses are not accessible from the outside and are therefore not serviceable parts.

4.2 INRUSH CURRENT

The AC-DC power supply exhibits an X capacitance of only 4.3μ F, resulting in a low and short peak current, when the supply is connected to the mains. The internal bulk capacitor will be charged through an NTC which will limit the inrush current.

NOTE:

Do not repeat plug-in / out operations below 90sec interval time at maximum input, high temperature condition, or else the internal in-rush current limiting device (NTC) may not sufficiently cool down and excessive inrush current or component failure(s) may result.



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4.3 INPUT UNDER-VOLTAGE

If the RMS value of input voltage (either AC or DC) stays below the input undervoltage lockout threshold Vi on, the supply will be inhibited. Once the input voltage returns within the normal operating range, the supply will return to normal operation again.

4.4 POWER FACTOR CORRECTION

Power factor correction (PFC) (see *Figure 4*) is achieved by controlling the input current waveform synchronously with the input voltage. A fully digital controller is implemented giving outstanding PFC results over a wide input voltage and load ranges. The input current will follow the shape of the input voltage. If for instance the input voltage has a trapezoidal waveform, then the current will also show a trapezoidal waveform. At DC input voltage the PFC is still in operation, but the input current will be DC in this case.

4.5 EFFICIENCY

The high efficiency (see *Figure 2*) is achieved by using state-of-the-art silicon power devices in conjunction with soft-transition topologies minimizing switching losses and a full digital control scheme. Synchronous rectifiers on the output reduce the losses in the high current output path. The rpm of the fan is digitally controlled to keep all components at an optimal operating temperature regardless of the ambient temperature and load conditions.

Figure 3 shows efficiency when input voltage is supplied from a high voltage DC source.

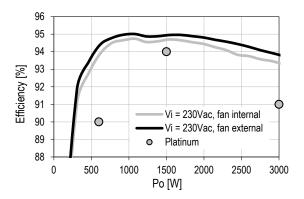


Figure 2 – AC Input Efficiency vs. Load current (ratio metric loading)

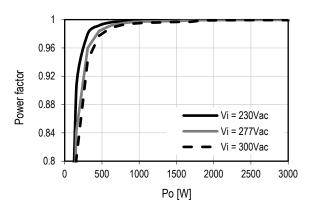


Figure 4 - Power factor vs. Load current

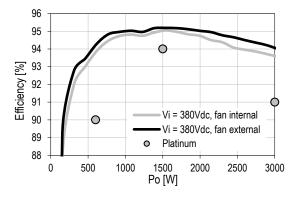


Figure 3 - DC Input Efficiency vs. load current (ratio metric loading)



Figure 5 - Inrush current, Vin = 230 Vac, 0°phase angle CH4: Vin (200 V/div), CH3: Iin (10 A/div)



5. OUTPUT

General Condition: $T_A = 0...45$ °C unless otherwise noted.

Main Output V1 V1 nom Nominal Output Voltage Output Setpoint Accuracy 0.5 ⋅ h nom, Tamb = 25 °C 12.3 VDC V1 set Output Setpoint Accuracy 0.5 ⋅ h nom, Tamb = 25 °C -0.5 +0.5 % Vn nom dV1 set Total Regulation V min to V max, 0 to 100% h nom, Ta min to Ta max -1 +1 % Vn nom P1 nom Nominal Output Power V = 12.3 VDC, Vin < 180 VAC	PARAME	TER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
V1 set Output Setpoint Accuracy 0.5 ⋅ Λ nom, Tamb = 25 °C -0.5 +0.5 % V nom dV1 tot Total Regulation V min to V max. 0 to 100% Λ nom, Ta min to Ta max -1 +1 % V nom P1 nom Nominal Output Power Vi = 12.3 VDC, Vin < 180 VAC	Main Out	put V1					
P1 nom Nominal Output Power $V_1 = 12.3 \text{ VDC}$, Vin < 180 VAC 1400 W $I_{I nom}$ Nominal Output Current $V_1 = 12.3 \text{ VDC}$, Vin < 180 VAC		·	0.5 · / _{1 nom} , $T_{amb} = 25 ^{\circ}\text{C}$	-0.5	12.3	+0.5	
$I_{I nom}$ Nominal Output Current $V_1 = 12.3 \text{ VDC}$, $V \text{in} < 180 \text{ VAC}$ 114ADC $P_{I nom}$ Nominal Output Power $V_1 = 12.3 \text{ VDC}$, $V \text{in} > 180 \text{ VAC}$ 3000W $I_{I nom}$ Nominal Output Current $V_1 = 12.3 \text{ VDC}$, $V \text{in} > 180 \text{ VAC}$ 244ADC $I_{I nom}$ Nominal Output Current $V_1 = 12.3 \text{ VDC}$, $V \text{in} > 180 \text{ VAC}$ 244ADC $I_{I nom}$ Short time over load current $V_1 = 12.3 \text{ VDC}$, $V \text{in} > 180 \text{ VAC}$ 244ADC $I_{I nom}$ Short time over load current $I_{I nom}$, $I_{I nom}$, max, maximum duration 20 ms (See Section 5.2)292A $V_{I pp}$ Output Ripple Voltage $V_{I nom}$, $I_{I nom}$ 160mVpp $dV_{I load}$ Load Regulation $V_{I I I nom}$, $V_{I nom}$, $V_{I nom}$, $V_{I nom}$ 170mV $dV_{I Line}$ Line Regulation $V_{I I I Nom}$, $V_{I nom}$, $V_{I nom}$ 120127 $V_{I I I I I I I I I I I I I I I I I I I $	dV _{1 tot}	Total Regulation	$V_{i min}$ to $V_{i max}$, 0 to 100% $I_{i nom}$, $T_{a min}$ to $T_{a max}$	-1		+1	% V _{1 nom}
$P_{1 \ nom}$ Nominal Output Power $V_1 = 12.3 \ \text{VDC}$, $V_1 > 180 \ \text{VAC}$ 3000W $I_{1 \ nom}$ Nominal Output Current $V_1 = 12.3 \ \text{VDC}$, $V_1 > 180 \ \text{VAC}$ 244ADC $V_{1 \ nom}$ Short time over load current $V_1 = 12.3 \ \text{VDC}$, $V_1 > 180 \ \text{VAC}$ 244ADC $V_{1 \ nom}$ Short time over load current $V_1 = 12.3 \ \text{VDC}$, $V_2 > 180 \ \text{VAC}$ 292A $V_{1 \ nom}$ Output Ripple Voltage $V_1 \ nom$, $V_2 > 100 \ \text{MHz}$ BW (See Section 5.1)160mVpp $V_{1 \ nom}$ Load Regulation $V_1 = V_2 \ \text{Min}$ $V_2 = V_2 = 100 \ \text{MeV}$ 170mV $V_1 \ nom$ $V_2 = V_2 = 100 \ \text{MeV}$ 170mV $V_1 \ nom$ $V_2 = V_2 = 100 \ \text{MeV}$ 120127 $V_1 \ nom$ $V_2 = V_2 = 100 \ \text{MeV}$ 120127 $V_2 \ nom$ $V_3 = 180 \ \text{VAC}$, $V_4 = 250 \ \text{C}$ 248274 $V_3 \ nom$ $V_4 = 180 \ \text{VAC}$, $V_4 = 250 \ \text{C}$ 186212 $V_1 \ nom$ $V_2 \ nom$ 180 \ NAC, $V_3 = 250 \ \text{C}$ 186212 $V_3 \ nom$ $V_4 \ nom$ 180 \ NAC, $V_4 \ nom$ 180 \ NA	P _{1 nom}	Nominal Output Power	V ₁ = 12.3 VDC, Vin < 180 VAC		1400		W
$I_{1 \text{ nom}}$ Nominal Output Current $V_1 = 12.3 \text{ VDC}$, $V_{1n} > 180 \text{ VAC}$ 244ADC $k_{1 \text{ ol}}$ Short time over load current $V_{1 \text{ amin to}}$ $T_{2 \text{ amax}}$, maximum duration 20 ms (See Section 5.2)292A $v_{1 \text{ pp}}$ Output Ripple Voltage $V_{1 \text{ nom}}$, $A_{\text{ nom}}$, 20 MHz BW (See Section 5.1)160mVpp $dV_{1 \text{ Load}}$ Load Regulation $V_{1 \text{ Nom}}$, $A_{\text{ nom}}$, 20 MHz BW (See Section 5.1)170mV $dV_{1 \text{ Line}}$ Line Regulation $V_{1 \text{ Nom}}$, $V_{1 \text{ max}}$ 0mV $V_{1 \text{ oll lim}}$ $V_{1 \text{ oll lim}}$ $V_{1 \text{ ell}}$, $V_{1 \text{ oll lim}}$, $V_{2 $	I _{1 nom}	Nominal Output Current	V ₁ = 12.3 VDC, Vin < 180 VAC		114		ADC
	P _{1 nom}	Nominal Output Power	V ₁ = 12.3 VDC, Vin > 180 VAC		3000		W
$k_{1 \text{ ol}}$ Short time over load current $T_{a \text{ min to }} T_{a \text{ max}}$, maximum duration 20 ms (See Section 5.2)292A $V_{1 \text{ pp}}$ Output Ripple Voltage $V_{1 \text{ nom}}$, $A_{1 \text{ nom}}$, 20 MHz BW (See Section 5.1)160mVpp $dV_{1 \text{ Line}}$ Load Regulation $V_{1 \text{ Him}}$, $V_{1 \text{ max}}$ 0mV $dV_{1 \text{ Line}}$ Line Regulation $V_{1 \text{ Him}}$, $V_{1 \text{ max}}$ 0127 $V_{1 \text{ line}}$ Current limitation $V_{1 \text{ 80}}$ VAC, $V_{1 \text{ a}} < 45^{\circ}\text{C}$ $V_{1 \text{ 80}}$ VAC, $V_{1 \text{ a}} < 55^{\circ}\text{C}^{-3}$ $V_{1 \text{ 80}}$ VAC, $V_{1 \text{ a}} < 45^{\circ}\text{C}$ $V_{1 \text{ 80}}$ VAC, $V_{1 \text{ a}} < 45^{\circ}\text{C}$ $V_{2 \text{ 80}}$ VAC, $V_{2 \text{ a}} < 45^{\circ}\text{C}$ $V_{2 \text{ 80}}$ VAC $V_{2 \text{ 80}}$ VAC <br< td=""><td>I_{1 nom}</td><td>Nominal Output Current</td><td>V₁ = 12.3 VDC, Vin > 180 VAC</td><td></td><td>244</td><td></td><td>ADC</td></br<>	I _{1 nom}	Nominal Output Current	V ₁ = 12.3 VDC, Vin > 180 VAC		244		ADC
$dV_{1 Load}$ Load Regulation $V_{1 lom}$ $V_{1 lom}$ $V_{2 lom}$ $V_$	∕v₁ ol	Short time over load current	$T_{a \text{ min to}} T_{a \text{ max}}$, maximum duration 20 ms			292	Α
$dV_{1 Line}$ Line Regulation $V_{1} = V_{1 min} V_{1 max}$ 0 mV $V_{1 Line}$ Line Regulation $V_{1 Rin} V_{1 max}$ 0 mV $V_{1 Line}$ Line Regulation $V_{1 Rin} V_{1 max}$ 120 127 $V_{1 Rin} V_{1 Rin$	V _{1 pp}	Output Ripple Voltage	$V_{1 \text{ nom}}$, $I_{1 \text{ nom}}$, 20 MHz BW (See Section 5.1)			160	mVpp
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	dV _{1 Load}	Load Regulation	$V_i = V_{i \text{ nom}}$, 0 - 100 % $h_{i \text{ nom}}$		170		mV
$k_{1 \text{ ol lim}}$ Current limitation $V_{1} < 180 \text{ VAC}, T_{a} < 45^{\circ}\text{C}$ 248 274 274 $180 \text{ VAC}, T_{a} < 45^{\circ}\text{C}$ 248 274 186 212 ADC dl_{share} Current Sharing Deviation from $h_{10}l$ / $N, h > 25\% h_{nom}$ -5% $+5\%$ A A dV_{clyn} Dynamic Load Regulation Deviation from $h_{10}l$ / h_{10m} , $h_{10}l$ h_{1	$dV_{1 Line}$	Line Regulation			0		mV
$ \frac{dV_{dyn}}{dV_{dyn}} \text{Dynamic Load Regulation} \frac{\Delta h_1 = 50\% \ h_{\text{nom}}, \ h_1 = 5 \dots 100\% \ h_{\text{nom}}, \ dh/dt = 1A/\mu_{\text{N}}, \ f_{\Delta h} = 0.05 \dots 10 \ \text{kHz}, \ Duty_{\Delta h} = 10 \dots 90\%, \text{ recovery within 1\% of } V_1 \text{ final steady state} $	√ 1 ol lim	Current limitation	<i>I</i> /< 180 VAC, <i>T</i> _a = 55 °C ³⁾ <i>I</i> /> 180 VAC, <i>T</i> _a < 45°C	92 248		99 274	ADC
dV_{clyn} Dynamic Load Regulation $\frac{dh}{dt} = 1A/\mu_{S}$, $f_{\Delta /l} = 0.0510$ kHz, Duty $_{\Delta /l} = 1090\%$, recovery within 1% of V_1 final steady state-0.6+0.6V T_{rec} Recovery Time0.5ms $t_{AC\ VI}$ Start-up Time from AC $V_1 = 10.8\ VDC$ (see Figure 7)3sec $t_{V1\ rise}$ Rise Time $V_1 = 1090\%\ V_1\ nom$ (see Figure 8)2.5ms	dlshare	Current Sharing	Deviation from h_{tot} / N, $h > 25\%$ h_{nom}	-5%		+5%	Α
$t_{AC\ VI}$ Start-up Time from AC $V_1 = 10.8\ \text{VDC}$ (see Figure 7)3sec $t_{VI\ rise}$ Rise Time $V_1 = 1090\%\ V_1\ \text{nom}$ (see Figure 8)2.5ms	dV _{dyn}	Dynamic Load Regulation	$dh/dt = 1A/\mu s$, $f_{\Delta h} = 0.0510$ kHz, Duty $_{\Delta h} = 1090\%$, recovery within 1% of V_1 final	-0.6		+0.6	V
$t_{V1 \text{ rise}}$ Rise Time $V_1 = 1090\% \ V_{1 \text{ nom}} \text{ (see Figure 8)}$ 2.5 ms	T _{rec}	Recovery Time				0.5	ms
	t _{AC V1}	Start-up Time from AC	V₁ = 10.8 VDC (see <i>Figure 7</i>)			3	sec
C_{Load} Capacitive Loading $T_a = 25^{\circ}\text{C}$ 30000 μF	t _{V1 rise}	Rise Time	$V_1 = 1090\% \ V_{1 \text{ nom}} \text{ (see Figure 8)}$		2.5		ms
	C _{Load}	Capacitive Loading	$T_a = 25$ °C			30000	μF

³ See *Figure 20* for linear derating > 45°C

Stanby O	utput V sB					
V _{SB nom}	Nominal Output Voltage	&B nom. Tamb = 25°C		12		VDC
V _{SB set}	Output Setpoint Accuracy	ASB nom, Tamb = 25 C	-0.5		+0.5	% V _{SBnom}
dV _{SB tot}	Total Regulation	$V_{i min} to V_{i max}$, $k_{SB nom}$, $T_{a min} to T_{a max}$	-1		+1	% V _{SBnom}
PSB nom	Nominal Output Power	V _{SB} = 12 VDC		60		W
I _{SB nom}	Nominal Output Current	V _{SB} = 12 VDC		5		ADC
V _{SB pp}	Output Ripple Voltage	V _{SB nom} , J _{SB nom} , 20 MHz BW (See Section 5.1)			300	mVpp
dVsB	Droop	0 - 100 % <i>I</i> sB nom		400		mV
√ SB lim	Current Limitation		6		9	ADC
dV _{SBdyn}	Dynamic Load Regulation	$\Delta k_{\rm B} = 50\%$ $k_{\rm B\ nom}$, $k_{\rm B} = 5$ 100% $k_{\rm B\ nom}$, $dk/dt = 1A/\mu s$, $f_{\Delta M} = 0.0510 kHz$, Duty $\Delta M = 1090\%$, recovery within 1% of $V_{\rm SB}$ final steady state	-0.6		+0.6	V_{SBnom}
Trec	Recovery Time				0.5	ms
t _{AC VSB}	Start-up Time from AC	V _{SB} = 90% V _{SB nom} (see Figure 7)			3	sec
t∕\SB rise	Rise Time	V _{SB} = 1090% V _{SB nom} (see <i>Figure 9</i>)		10		ms
C_{Load}	Capacitive Loading	$T_{\text{amb}} = 25^{\circ}\text{C}$			3000	μF



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5.1 OUTPUT VOLTAGE RIPPLE

The internal output capacitance at the power supply output (behind OR-ing element) is minimized to prevent disturbances during hot plug. In order to provide low output ripple voltage in the application, external capacitors should be added close to the power supply output.

The setup of *Figure 6* has been used to evaluate suitable capacitor types. The capacitor combinations of Table 1 and Table 2 should be used to reduce the output ripple voltage.

The ripple voltage is measured with 20 MHz BWL, close to the external capacitors.

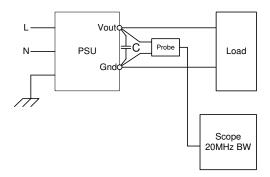


Figure 6 - Output Ripple Test Setup

NOTE: Care must be taken when using ceramic capacitors with a total capacitance of 1 μ F to 50 μ F on output V1, due to their high quality factor the output ripple voltage may be increased in certain frequency ranges due to resonance effects.

External Capacitor V1	dV1max	Unit
2Pcs 47μF/16V/X5R/1210	160	mVpp
1Pcs 1000µF/16V/Low ESR Aluminum/ø10x20	160	mVpp
1Pcs 270µF/16V/Conductive Polymer/ø8x12	160	mVpp
2Pcs 47µF/16V/X5R/1210 plus 1Pcs 270µF Conductive Polymer OR 1Pcs 1000µF Low ESR AlCap	90	mVpp

External capacitor VSB	dVSBmax	Unit
1Pcs 10µF/16 V/X7R/1206	300	mVpp

Table 1 - Suitable Capacitors for V1

Table 2 - Suitable Capacitors for VSB

The output ripple voltage on VSB is influenced by the main output V1. Evaluating VSB output ripple must be done when maximum load is applied to V1.

5.2 SHORT TIME OVERLOAD

The main output has the capability to allow load current up to 20% above the nominal output current rating for a maximum duration of 20 ms. This allows the system to consume extended power for short time dynamic processes.

5.3 OUTPUT ISOLATION

Main and standby output and all signals are isolated from the chassis and protective earth connection, although the applied voltage must not exceed 100 Vpeak to prevent any damage of the supply.

Internal to the supply the main output ground, standby output ground and signal ground are interconnected through 10Ω resistors to prevent any circulating current within the supply. In order to prevent any potential difference in outputs or signals within the application these 3 grounds must be directly interconnected at system level. See also section 14 for pins to be interconnected.



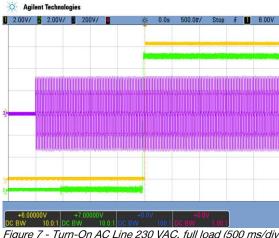


Figure 7 - Turn-On AC Line 230 VAC, full load (500 ms/div) CH1: V1 (2 V/div); CH2: VSB (2 V/div); CH3: Vin (200 V/div)



Figure 9 - Turn-On AC Line 230 VAC, full load (5 ms/div) CH2: VSB (2 V/div)



Figure 11 - Short circuit on V1 (50ms/div) CH1: V1 (2V/div) CH2: VSB (2V/div) CH4: I1 (200A/div)



Figure 8 - Turn-On AC Line 230 VAC, full load (1 ms/div) CH1: V1 (2 V/div)



Figure 10 - Turn-Off AC Line 230 VAC, full load (20 ms/div) CH1: V1 (2 V/div); CH2: VSB (2 V/div); CH3: Vin (200 V/div) Agilent Technologies

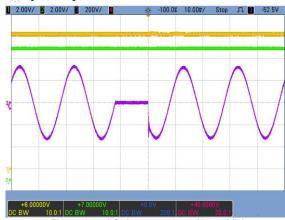


Figure 12 - AC drop out 12ms (10ms/div) CH1: V1 (2V/div) CH2: VSB (2V/div) CH3: Vin (200V/div)



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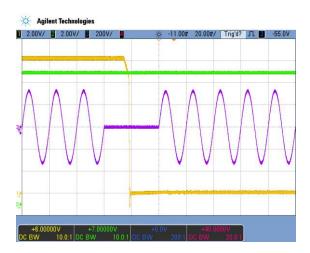
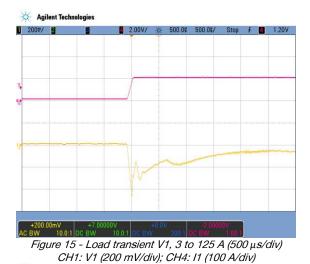


Figure 13 - AC drop out 40 ms, full load (20 ms/div) CH1: V1 (2 V/div); CH2: VSB (2 V/div); CH3: Vin (200 V/div)



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Figure 17 - Load transient V1, 122 to 244 A (500 µs/div) CH1: V1 (200 mV/div); CH4: I1 (100 A/div)

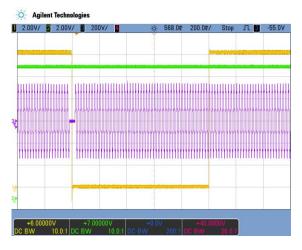


Figure 14 - AC drop out 40 ms, full load (200 ms/div),V1 restart after 1 sec CH1: V1 (5 V/div); CH2: VSB (2 V/div); CH3: I1 (200 V/div)



Figure 16 - Load transient V1, 125 to 3 A (500 μs/div) CH1: V1 (200 mV/div); CH4: I1 (100 A/div)

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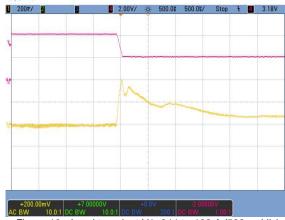


Figure 18 - Load transient V1, 244 to 122 A (500 μs/div) CH1: V1 (200 mV/div); CH4: I1 (100 A/div)



6. PROTECTION

PARAME	ΓER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
F	Input Fuses (L+N)	Not user accessible, quick-acting (F)		25		Α
V ₁ ov	OV Threshold 1/1		13.6	14.2	14.8	VDC
<i>t</i> ov v1	OV Latch Off Time 1/1				1	ms
V _{SB OV}	OV Threshold V _{SB}		13.3	13.9	14.5	VDC
t _{OV VSB}	OV Latch Off Time V _{SB}				1	ms
ん 1 lim	Current limitation	$\it K$ < 180 VAC, $\it T_a$ < 45°C $\it K$ < 180 VAC, $\it T_a$ = 55°C $\it ^4$ $\it K$ > 180 VAC, $\it T_a$ < 45°C $\it K$ > 180 VAC, $\it T_a$ = 55°C $\it ^4$	120 92 248 186		127 99 274 212	А
t√₁ lim	Current limit blanking time	Time to latch off when in over current	20	22	24	ms
1 ∕1 ol lim	Current limit during short time overload V_1	Maximum duration 20 ms	292	300	308	Α
l∕₁ sc	Max Short Circuit Current 1/1	$V_1 < 3 \text{ V}$			350 ⁵	Α
t√1 SC off	Short circuit latch off time	Time to latch off when in short circuit		10		ms
√ SB lim	Current limitation V _{SB}		6		9	Α
t√SB lim	Current limit blanking time	Time to hit hiccup when in over current			1	ms
T _{SD}	Over temperature on critical points	Inlet Ambient Temperature PFC Primary Heatsink Temperature Secondary Sync Mosfet Temperature Secondary OR-ing Mosfet Temperature			60 80 115 125	°C

See Figure 20 for linear derating > 45°

6.1 AUTOMATIC RETRY

For all fault conditions except current limitation on Standby output, the supply will shut down for 10sec and restart automatically. The supply will auto-restart from a fault up to 5 times, after that it will latch off. The latch and restart counter can be cleared by recycling the input voltage or the PSON_L input. A failure on the Standby output will shut down both Main and Standby outputs. A failure on the Main output will shut down only the Main output, while Standby continues to operate.

6.2 OVERVOLTAGE PROTECTION

The PFE front-ends provide a fixed threshold overvoltage (OV) protection implemented with a HW comparator. Once an OV condition has been triggered, the supply will shut down and latch the fault condition.

6.3 UNDERVOLTAGE DETECTION

Both main and standby outputs are monitored. LED and PWOK_L pin signal if the output voltage exceeds ±7% of its nominal voltage.

Output undervoltage protection is provided on both outputs. When either V1 or VSB falls below 93% of its nominal voltage, the output is inhibited.



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Limit set don't include effects of main output capacitive discharge.

6.4 CURRENT LIMITATION

MAIN OUTPUT

Two different over current protection features are implemented on the main output.

A static over current protection will shut down the output, if the output current does exceed I_{V1 lim} for more than 20 ms. If the output current is increased slowly this protection will shut down the supply. The main output current limitation level I_{V1 lim} will decrease if the ambient (inlet) temperature increases beyond 45 °C (see *Figure 20*). Note that the actual current limitation on V1 will kick in at a current level approximately 20 A higher than what is shown in *Figure 20* (see also section 9 for additional information).

The 2^{nd} profection is a substantially rectangular output characteristic controlled by a software feedback loop. This protects the power supply and system during the 20ms blanking time of the static over current protection. If the output current is rising fast and reaches k_{10} lim, the supply will immediately reduce its output voltage to prevent the output current from exceeding k_{10} lim. When the output current is reduced below k_{10} lim, the output voltage will return to its nominal value.

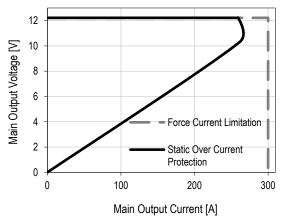


Figure 19 - Current Limitation on V_1 ($V_i = 230VAC$)

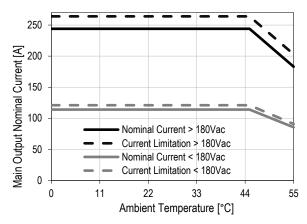


Figure 20 - Derating on V1 vs. Ta

STANDBY OUTPUT

On the standby output a hiccup type over current protection is implemented. This protection will shut down the standby output immediately when standby current reaches or exceeds $k_{SB \text{ lim}}$. After an off-time of 1s the output automatically tries to restart. If the overload condition is removed the output voltage will reach again its nominal value. At continuous overload condition the output will repeatedly trying to restart with 1s intervals.

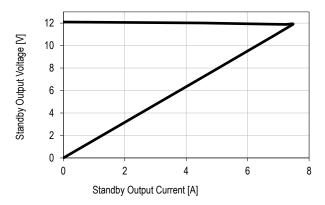


Figure 21 - Current Limitation on VSB



7. MONITORING

I_{mon} Input RMS Current $I_{c} > 4$ A _{rms} $I_{c} < 5$ $I_{c} < 5$ $I_{c} < 5$ $I_{c} < 6$ $I_{$	PARAME	TER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
I_{mon} Input RMS Current $I_{l} \le 4 A_{rms}$ -0.2 $+0.2$ A_{rms} $R_{l} > 700 W$ -5 $+5$ 96 $R_{l} \le 700 W$ -35 $+35$ 100 $I_{l} = I_{l} = I_{$	V₁ mon	Input RMS Voltage	$V_{i \min} \leq V_{i} \leq V_{i \max}$	-2.5		+2.5	%
$R_{\text{mon}} = \frac{1}{2} \left\{ \begin{array}{c} A_{\text{rms}} \\ A_{\text{rms}} A_{\text{rms}} \\ A_{\text{rms}} \\ A_{\text{rms}} \\ A_{\text{rms}} \\ A_{\text{rms}} \\ A$	1	Input PMS Current	$h > 4 \text{ A}_{rms}$	-5		+5	%
R mon True Input Power R ≤ 700 W -35 +35 W E_{imon} Total Input Energy R ≤ 700 W -5 +5 96 V_{imon} V1 Voltage -2 +2 96 I_{imon} V1 Current I1 ≤ 30 A -2 +2 96 I_{imon} Po > 200 W -5 +5 96 I_{imon} Po ≥ 200 W -10 +10 W I_{imon} Total Output Energy Po ≥ 200 W -5 +5 96 I_{imon} Po ≥ 200 W -10 +10 W	/i mon	input nivio Guirent	$I_i \le 4 A_{rms}$	-0.2		+0.2	A _{rms}
$R \le 700 \text{ W} \qquad -35 \qquad +35 \qquad \text{W}$ $E_{i mon} \qquad \text{Total Input Energy} \qquad \begin{array}{c} P_i > 700 \text{ W} \\ P_i \le 700 \text{ W} \\ P_i \le 700 \text{ W} \\ \end{array} \qquad \begin{array}{c} -5 \\ -35 \\ \end{array} \qquad \begin{array}{c} +5 \\ 96 \\ \end{array} \qquad \begin{array}{c} 96 \\ 96 \\ \end{array} \qquad \begin{array}{c} 11 > 30 \text{ A} \\ P_0 > 200 \text{ W} \\ \end{array} \qquad \begin{array}{c} -2 \\ -2 \\ \end{array} \qquad \begin{array}{c} +2 \\ -2 \\ \end{array} \qquad \begin{array}{c} 96 \\ -2 \\ \end{array} \qquad \begin{array}{c} +2 \\ 96 \\ -2 \\ \end{array} \qquad \begin{array}{c} 96 \\ -2 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ \end{array} \qquad \begin{array}{c} 96 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ -10 \\ $	D	True Input Power	$P_{\rm i} > 700 {\rm W}$	-5		+5	%
Eimon Total Input Energy $P_1 \le 700 \text{ W}$ -35 +35 W Vi mon V1 Voltage -2 +2 96 $I_1 = 30 \text{ A}$ -2 +2 96 $I_1 \le 30 \text{ A}$ -0.6 +0.6 A Po nom Total Output Power Po > 200 W -5 +5 96 Po ≈ 200 W -10 +10 W Eo mon Total Output Energy Po ≤ 200 W -10 +10 W	r i mon	True input Fower	<i>P</i> _i ≤ 700 W	-35		+35	W
$R \le 700 \text{ W}$ -35 +35 W $R \le 700 \text{ W}$ -35 +35 W $R \le 700 \text{ W}$ -35 +25 W $R \le 700 \text{ W}$ -2 +2 % $R \ge 700 \text{ W}$ -2 +2 % $R \ge 700 \text{ W}$ -10 +10 W $R \ge 700 \text{ W}$ -5 +5 % $R \ge 700 \text{ W}$ -10 +10 W $R \ge 700 \text{ W}$ -10 W $R \ge 700 \text{ W}$ -1	E .	Total Input Energy	<i>P</i> i > 700 W	-5		+5	%
h_{100} V ₁ Current $\begin{array}{c} \text{II} > 30 \text{ A} \\ \text{II} ≤ 30 \text{ A} \\ \end{array}$ $\begin{array}{c} -2 \\ \text{II} ≤ 30 \text{ A} \\ \end{array}$ $\begin{array}{c} -0.6 \\ \text{A} \\ \end{array}$ $\begin{array}{c} +0.6 \\ \text{A} \\ \end{array}$ $\begin{array}{c} +0.6 \\ \text{A} \\ \end{array}$ $\begin{array}{c} P_{0} > 200 \text{ W} \\ P_{0} ≤ 200 \text{ W} \\ \end{array}$ $\begin{array}{c} -5 \\ \text{10} \\ \end{array}$ $\begin{array}{c} +10 \\ \text{H} \\ \end{array}$ $\begin{array}{c} 0.6 \\ \text{H} \\ $	⊏ i mon	Total Input Energy	<i>P</i> _i ≤ 700 W	-35		+35	Wh
h_{mon} V1 Current I1 ≤ 30 A -0.6 +0.6 A $P_{O nom}$ Total Output Power Po > 200 W -5 +5 % $P_{O nom}$ Po ≤ 200 W -10 +10 W $E_{O mon}$ Total Output Energy Po ≤ 200 W -5 +5 % $P_{O nom}$ Po ≤ 200 W -10 +10 W	V₁ mon	V ₁ Voltage		-2		+2	%
$P_{\text{O nom}}$ Total Output Power $P_{\text{O nom}}$ 11 ≤ 30 A -0.6 +0.6 A Po > 200 W -5 +5 % Po ≤ 200 W -10 +10 W Po ≤ 200 W -5 +5 % Po > 200 W -5 +5 % Po > 200 W -5 +5 % Po ≤ 200 W -10 +10 W Po ≤ 200 W -10 W Po ≤ 200 W Po ≤ 200 W -10 W Po ≤ 200 W Po ≤		V. Current	I1 > 30 A	-2		+2	%
Po nom Total Output Power Po ≤ 200 W -10 +10 W Po > 200 W -5 +5 % Fo mon Po ≤ 200 W -10 +10 W	/1 mon	V1 Gurrent	I1 ≤ 30 A	-0.6		+0.6	Α
Po ≤ 200 W -10 +10 W Eo mon Total Output Energy Po ≤ 200 W -5 +5 % Po ≤ 200 W -10 +10 W	D	Total Output Power	Po > 200 W	-5		+5	%
E _{o mon} Total Output Energy Po ≤ 200 W -10 +10 W	ro nom	Total Output Fower	Po ≤ 200 W	-10		+10	W
Po ≤ 200 W -10 +10 W	_	Total Output Engrav	Po > 200 W	-5		+5	%
I Standby Voltago	Eo mon	rotal Output Energy	Po ≤ 200 W	-10		+10	Wh
VSB mon Standby Voltage -2 +2 /d	VSB mon	Standby Voltage		-2		+2	%
$k_{\rm B\;mon}$ Standby Current $k_{\rm B} \le k_{\rm B\;nom}$ -0.3 +0.3 A	SB mon	Standby Current	lsB ≤ lsB nom	-0.3		+0.3	Α



8. SIGNALING AND CONTROL

8.1 ELECTRICAL CHARACTERISTICS

PARAMETER	DESCRIPTION / CONDITION		MIN	NOM	MAX	UNIT
PSKILL / PSON_L	inputs					
ИL	Input low level voltage		-0.2		0.8	V
Ин	Input high level voltage		2.0		3.6	V
∕IL, H	Maximum input sink or source current		0		1	mA
<i>R</i> _{puPSKILL}	Internal pull up resistor on PSKILL			10		kΩ
R _{puPSON_L}	Internal pull up resistor on PSON_L			10		kΩ
PWOK_L output						
V _{OL}	Output low level voltage	/ _{sink} < 4 mA	-0.2		0.4	V
$V_{ m puPWOK_L}$	External pull up voltage				12	V
$R_{ m puPWOK_L}$	Recommended external pull up resistor on PWOK_L at \$\mu_{PuPWOK_L} = 3.3 \text{ V}			10		kΩ
Low level output	All outputs are turned on and within regulation					
High level output	In standby mode or V ₁ /V _{SB} have triggered a fault condition					
INOK_L output						
V o∟	Output low level voltage	$I_{sink} < 4 \text{ mA}$	-0.2		0.4	V
$V_{ m pulNOK_L}$	External pull up voltage				12	V
$R_{ m pulNOK_L}$	Recommended external pull up resistor on INOK_L at V_{pulNoK_L} = 3.3 V			10		kΩ
Low level output	Input voltage is within range for PSU to operate					
High level output	Input voltage is not within range for PSU to operate					
SMB_ALERT_L out	'put					
V OL	Output low level voltage	/ _{sink} < 4 mA	-0.2		0.4	V
$V_{ m puSMB_ALERT_L}$	External pull up voltage				12	V
R _{puSMB_ALERT_L}	Recommended external pull up resistor on SMB_ALERT_L at \(V_{pusMB_ALERT_L} = 3.3 \) V			10		kΩ
Low level output	PSU in warning or failure condition					
High level output	PSU is ok					

8.2 INTERFACING WITH SIGNALS

A 15V zener diode is added on all signal pins versus signal ground SGND to protect internal circuits from negative and high positive voltage. Signal pins of several supplies running in parallel can be interconnected directly. A supply having no input power will not affect the signals of the paralleled supplies.

ISHARE pins must be interconnected without any additional components. This in-/output also has a 15 V zener diode as a protection device and is disconnected from internal circuits when the power supply is switched off.

8.3 FRONT LEDs

The front-end has 2 front LEDs showing the status of the supply. LED number one is green and indicates AC power is on or off, while LED number two is bi-colored: green and yellow, and indicates DC power presence or fault situations. For the position of the LEDs see *Table 3* listing the different LED status.



OPERATING CONDITION	LED SIGNALING
AC LED	
AC Line within range	Solid Green
AC Line UV condition	Off
DC LED*	
Normal Operation	Solid Green
PSON_L High	Blinking Yellow (1:1)
V_1 or V_{SB} out of regulation	
Over temperature shutdown	
Output over voltage shutdown (V_1 or V_{SB})	Solid Yellow
Output under voltage shutdown (V_1 or V_{SB})	
Output over current shutdown (V_1 or V_{SB})	
Over temperature warning	Blinking Yellow/Green (2:1)
Minor fan regulation error (>5%, <15%)	Blinking Yellow/Green (1:1)

^{*} The order of the criteria in the table corresponds to the testing precedence in the controller.

Table 3 - LED Status

8.4 PRESENT L

The PRESENT_L is normally a trailing pin within the connector and will contact only once all other connector contacts are closed. This active-low pin is used to indicate to a power distribution unit controller that a supply is plugged in. The maximum sink current on PRESENT_L pin should not exceed 10 mA.

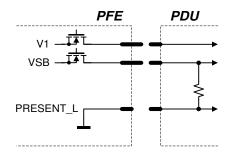


Figure 22 - PRESENT_L signal pin

8.5 PSKILL INPUT

The PSKILL input is an active-low and normally a trailing pin in the connector and is used to disconnect the main output as soon as the power supply is being plugged out. This pin should be connected to SGND in the power distribution unit. The standby output will remain on regardless of the PSKILL input state.

8.6 AC TURN-ON / DROP-OUTS / INOK L

The power supply will automatically turn-on when connected to the AC line under the condition that the PSON_L signal is pulled low and the AC line is within range. The INOK_L is an open collector output that requires an external pull-up to a maximum of 12 V indicating whether the input is within the range the power supply can use and turn on. The INOK_L signal is active-low. The timing diagram is shown in *Figure 23* and referenced in Table 4.



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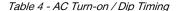
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OPERATIN	G CONDITION	MIN	MAX	UNIT
t _{AC VSB}	AC Line to 90% V _{VSB}		3	sec
<i>t</i> AC V1	AC Line to 90% V ₁		3	sec
tNOK_L on1	INOK_L signal on delay (start-up)		1800	ms
tNOK_L on2	INOK_L signal on delay (dips)	0	100	ms
t∕v1 holdup	Effective V ₁ holdup time	12	300	ms
t√SB holdup	Effective V _{SB} holdup time	40	300	ms
t _{INOK_L V1}	INOK_L to 1/1 holdup	7		ms
t _{INOK_L VSB}	INOK_L to V _{SB} holdup	27		ms
t√1 off	Minimum V_1 off time	1000	1200	ms
t/SB off	Minimum V _{SB} off time	1000	1200	ms
t∕/1dropout	Minimum V ₁ dropout time	12		ms
t√SBdropout	Minimum V _{SB} dropout time	40		ms



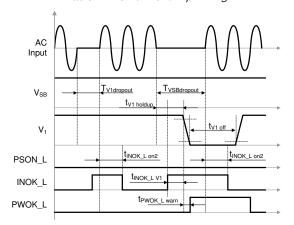


Figure 24 - AC short dips

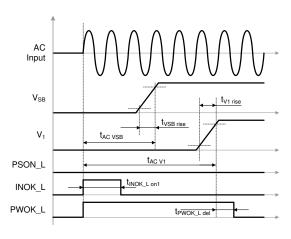


Figure 23 - AC turn-on timing

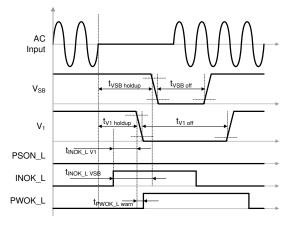


Figure 25 - AC long dips

8.7 PSON_L INPUT

The PSON_L is an internally pulled-up (3.3 V) input signal to enable / disable the main output V1 of the front-end. This active-low pin is also used to clear any latched fault condition. The timing diagram is given in *Figure 26* and the parameters in *Table 5*.

OPERATIN	IG CONDITION	MIN	MAX	UNIT
tPSON_L V1on	PSON_L to 1/1 delay (on)	190	220	ms
tPSON_L V1off	PSON_L to V ₁ delay (off)	0	100	ms

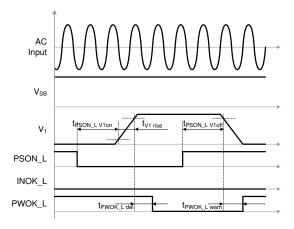
Table 5 - PSON_L timing

8.8 PWOK_L SIGNAL

The PWOK_L is an open collector output that requires an external pull-up to a maximum of 12 V indicating whether both VSB and V1 outputs are within regulation. This pin is active-low.

The timing diagram is shown in Figure 26 and referenced in Table 6.





OPERATING	G CONDITION	MIN	MAX	UNIT
tpwok_L del	V_1 to PWOK_L delay (on)	250	350	ms
tpwok_L warn	V₁ to PWOK_L delay (off)	0	5	ms

Figure 26 - PSON_L turn-on/off timing

Table 6 - PWOK_L timing

8.9 CURRENT SHARE

The PFE front-ends have an active current share scheme implemented for V1. All the ISHARE current share pins need to be interconnected in order to activate the sharing function. If a supply has an internal fault or is not turned on, it will disconnect its ISHARE pin from the share bus. This will prevent dragging the output down (or up) in such cases.

The current share function uses an analog bus. The controller implements a Master/Slave current share function. The power supply providing the largest current among the group is automatically the Master. The other supplies will operate as Slaves and increase their output current to a value close to the Master by slightly increasing their output voltage. The voltage increase is limited to +250 mV.

The standby output uses a passive current share method (droop output voltage characteristic).

No of paralleled PSUs	Maximum available power on main 12 V without redundancy	Maximum available power on main 12 V with n+1 redundancy	Maximum available power on standby output
1	3000 W	-	60 W
2	5850 W	3000 W	60 W
3	8700 W	5850 W	60 W
4	11550 W	8700 W	60 W
5	14400 W	11550 W	60 W
6	17250 W	14400 W	60 W

Table 7 - Power available when PSU in redundant operation

8.10 SENSE INPUTS

Main output has sense lines implemented to compensate for voltage drop on load wires. The maximum allowed voltage drop is 200 mV on the positive rail and 100 mV on the PGND rail.

With open sense inputs the main output voltage will rise by 250 mV. Therefore, if not used, these inputs should be connected to the power output and PGND close to the power supply connector. The sense inputs are protected against short circuit. In this case the power supply will shut down.

8.11 I2C / POWER MANAGEMENT BUS COMMUNICATION

The interface driver in the PFE supply is referenced to the SGND. The PFE supply is a communication slave device only; it never initiates messages on the I²C bus by itself. The communication bus voltage and timing is defined in Table 8 and further characterized through:



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- There are 100 $k\Omega$ internal pull-up resistors
- The SDA/SCL IOs must be pull-up externally to 3.3 ± 0.3 V
- Pull-up resistor should be 2 5 kΩ to ensure SMBUS compliant signal rise times
- I²C clock speed up to 100 kbps
- · Clock stretching limited to 1 ms
- SCL low time-out of >25 ms with recovery within 10 ms
- Recognizes any time Start/Stop bus conditions

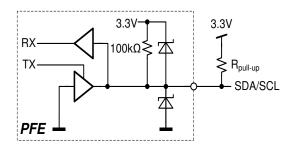


Figure 27 - Physical layer of communication interface

The SMB_ALERT_L signal indicates that the power supply is experiencing a problem that the system agent should investigate. This is a logical OR of the Shutdown and Warning events.

Communication to the DSP or the EEPROM will be possible as long as the input AC (DC) voltage is provided. If no AC (DC) is present, communication to the unit is possible as long as it is connected to a live VSB output (provided e.g. by the redundant unit). If only V1 is provided, communication is not possible.

PARAMETER	DESCRIPTION	CONDITION	MIN	MAX	UNIT
VíL	Input low voltage		-0.2	0.4	V
V iH	Input high voltage		2.1	3.6	V
V_{hys}	Input hysteresis		0.15		V
V ₀L	Output low voltage	4 mA sink current	0	0.4	V
$t_{\rm r}$	Rise time for SDA and SCL		20+0.1C _b *	300	ns
$t_{ m of}$	Output fall time ViHmin → ViLmax	$10 \ pF < C_b{}^{\star} < 400 \ pF$	20+0.1C _b *	250	ns
k	Input current SCL/SDA	0.1 VDD < Vi < 0.9 VDD	-10	10	μΑ
Ci	Capacitance for each SCL/SDA			10	pF
fscl	SCL clock frequency		0	100	kHz
<i>R</i> _{pu}	External pull-up resistor	f _{SCL} ≤ 100 kHz		1000 ns / C _b *	Ω
<i>t</i> hdsta	Hold time (repeated) START	f _{SCL} ≤ 100 kHz	4.0		μS
<i>t</i> Low	Low period of the SCL clock	f _{SCL} ≤ 100 kHz	4.7		μs
<i>t</i> HIGH	High period of the SCL clock	f _{SCL} ≤ 100 kHz	4.0		μs
<i>t</i> susta	Setup time for a repeated START	f _{SCL} ≤ 100 kHz	4.7		μS
<i>t</i> _{HDDAT}	Data hold time	f _{SCL} ≤ 100 kHz	0	3.45	μS
<i>t</i> sudat	Data setup time	f _{SCL} ≤ 100 kHz	250		ns
<i>t</i> susto	Setup time for STOP condition	f _{SCL} ≤ 100 kHz	4.0		μS
<i>t</i> BUF	Bus free time between STOP and START	f _{SCL} ≤ 100 kHz	4.7		μs
EEPROM_WP					
V í∟	Input low voltage		-0.2	0.4	V
V_{iH}	Input high voltage		2.1	3.6	V
<i>I</i> ₁	Input sink or source current		-1	1	mA
<i>R</i> _{pu}	Internal pull-up resistor to 3.3V			10k	Ω

^{*} Cb = Capacitance of bus line in pF, typically in the range of 10...400 pF

Table 8 - FC / SMBus Specification



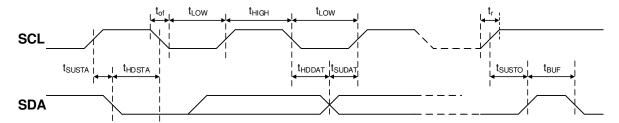


Figure 28 - FC / SMBus Timing

8.12 ADDRESS

The supply supports Power Management Bus communication protocol. Its address is fixed to 0x20. The EEPROM is at fixed address = 0xA0.

8.13 CONTROLLER AND EEPROM ACCESS

The controller and the EEPROM in the power supply share the same I²C bus physical layer (see *Figure 29*). In order to write to the EEPROM, the write protection needs to be disabled by setting EEPROM_WP input correctly. If EEPROM_WP is High, write is not allowed to the EEPROM and if Low, write is allowed. The EEPROM provides 2k bytes of user memory. None of the bytes are used for the operation of the power supply.

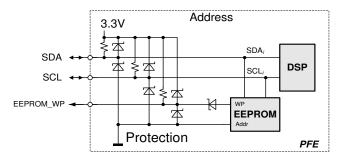


Figure 29 - f C Bus to DSP and EEPROM

8.14 EEPROM PROTOCOL

The EEPROM follows the industry communication protocols used for this type of device. Even though page write / read commands are defined, it is recommended to use the single byte write / read commands.

WRITE

The write command follows the SMBus 1.1 Write Byte protocol. After the device address with the write bit cleared a first byte with the data address to write to is sent followed by the data byte and the STOP condition. A new START condition on the bus should only occur after 5ms of the last STOP condition to allow the EEPROM to write the data into its memory.



READ

The read command follows the SMBus 1.1 Read Byte protocol. After the device address with the write bit cleared the data address byte is sent followed by a repeated start, the device address and the read bit set. The EEPROM will respond with the data byte at the specified location.



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8.15 POWER MANAGEMENT BUS PROTOCOL

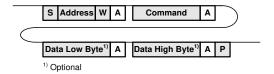
The Power Management Bus is an open standard protocol that defines means of communicating with power conversion and other devices. For more information, please see the System Management Interface Forum web site at: www.powerSIG.org.

Power Management Bus command codes are not register addresses. They describe a specific command to be executed. PFE3000-12-069RA supply supports the following basic command structures:

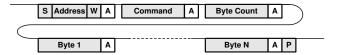
- Clock stretching limited to 1 ms
- SCL low time-out of >25 ms with recovery within 10 ms
- · Recognized any time Start/Stop bus conditions

WRITE

The write protocol is the SMBus 1.1 Write Byte/Word protocol. Note that the write protocol may end after the command byte or after the first data byte (Byte command) or then after sending 2 data bytes (Word command).

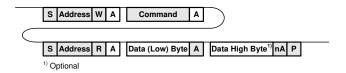


In addition, Block write commands are supported with a total maximum length of 255 bytes. See PFE3000-12-069RA Power Management Bus Communication Manual BCA.00070 for further information.

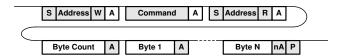


READ

The read protocol is the SMBus 1.1 Read Byte/Word protocol. Note that the read protocol may request a single byte or word.



In addition, Block read commands are supported with a total maximum length of 255 bytes. See PFE3000-12-069RA Power Management Bus Communication Manual BCA.00070 for further information.





8.16 GRAPHICAL USER INTERFACE

Bel Power Solutions I²C Utility provides a Windows® Vista/Win7/8 compatible graphical user interface allowing the programming and monitoring of the PFE3000-12-069RA Front-End. The utility can be downloaded on belfuse.com/power-solutions and supports the Power Management Bus protocol.

The GUI allows automatic discovery of the units connected to the communication bus and will show them in the navigation tree. In the monitoring view the power supply can be controlled and monitored.

If the GUI is used in conjunction with the PFE3000-12-069RA Evaluation Kit it is also possible to control the PSON_L pin(s) of the power supply.

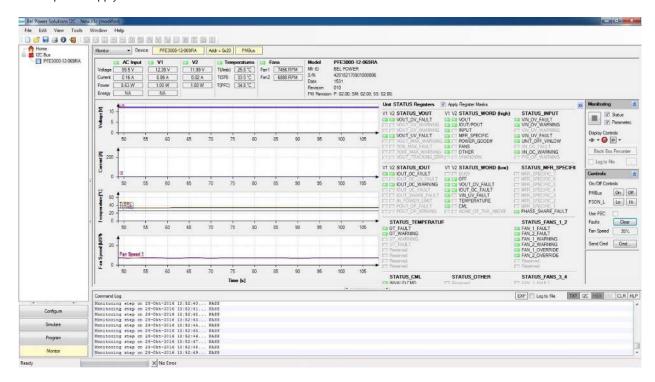


Figure 30 - Monitoring dialog of the I2C Utility

9. TEMPERATURE AND FAN CONTROL

To achieve best cooling results sufficient airflow through the supply must be ensured. Do not block or obstruct the airflow at the rear of the supply by placing large objects directly at the output connector. The PFE3000-12-069RA is provided with a reverse airflow, which means the air enters through the front of the supply and leaves at the rear. PFE supplies have been designed for horizontal operation.

The fan inside of the supply is controlled by a microprocessor. The rpm of the fan is adjusted to ensure optimal supply cooling and is a function of output power and the inlet temperature.



Figure 31 - Airflow Direction



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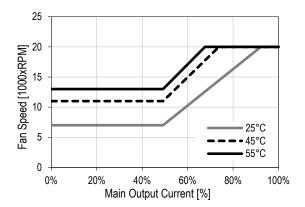


Figure 32 - Fan speed vs. main output load for PFE3000-12-069RA

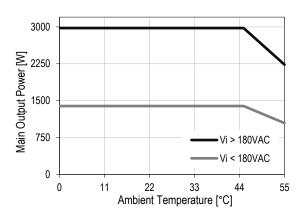


Figure 33 - Thermal derating for PFE3000-12-069RA

10. ELECTROMAGNETIC COMPATIBILITY

10.1 IMMUNITY

NOTE: Most of the immunity requirements are derived from EN 55024:1998/A2:2003.

PARAMETER	DESCRIPTION / CONDITION	CRITERION
ESD Contact Discharge	IEC / EN 61000-4-2, ±8 kV, 25+25 discharges per test point (metallic case, LEDs, connector body)	А
ESD Air Discharge	IEC / EN 61000-4-2, ±15 kV, 25+25 discharges per test point (non-metallic user accessible surfaces)	Α
Radiated Electromagnetic Field	IEC / EN 61000-4-3, 10 V/m, 1 kHz/80% Amplitude Modulation, 1 μs Pulse Modulation, 10 kHz2 GHz	Α
Burst	IEC / EN 61000-4-4, level 3 AC port ±2 kV, 1 minute DC port ±1 kV, 1 minute	А
Surge	IEC / EN 61000-4-5 Line to earth: level 3, ±2 kV Line to line: level 2, ±1 kV	А
RF Conducted Immunity	IEC/EN 61000-4-6, Level 3, 10 Vrms, CW, 0.1 80 MHz	Α
Voltage Dips and Interruptions	IEC/EN 61000-4-11 1: Vi 230Volts, 100% Load, Dip 100%, Duration 12ms 2: Vi 230Volts, 100% Load, Dip 100%, Duration < 150 ms 3. Vi 230Volts, 100% Load, Dip 100%, Duration > 150 ms	A V1: B, VSB: A B

10.2 EMISSION

PARAMETER	DESCRIPTION / CONDITION	CRITERION
Conducted Emission	EN55022 / CISPR 22: 0.15 30 MHz, QP and AVG	Class A
Radiated Emission	EN55022 / CISPR 22: 30 MHz 1 GHz, QP	Class A
Harmonic Emissions	IEC61000-3-2, Vin = 115/230 VAC, 50 Hz, 100% Load	Class A
Acoustical Noise	Sound power statistical declaration (ISO 9296, ISO 7779, IS9295) @ 50% load	60 dBA
AC Flicker	IEC / EN 61000-3-3, d _{max} < 3.3%	PASS



11. SAFETY / APPROVALS

Maximum electric strength testing is performed in the factory according to IEC/EN 60950, and UL 60950. Input-to-output electric strength tests should not be repeated in the field. Bel Power Solutions will not honor any warranty claims resulting from electric strength field tests.

PARAMETER	DESC	CRIPTION / CONDITION	MIN	NOM	MAX	UNIT
Agency App	• IEC6 • EN6 • UL/0 • CNS • EAC • BIS,	ved to the latest edition of the following 60950-1 2nd edition (CB) 0950-1 2nd Edition (Nemko) CSA0950-1 2nd Edition (cCSAus) 614336-1, CNS13438 (BSMI) 6, TR-CU (Russia) (India) C Safety/EMC, (South Korea)	standards:			
Isolation Str	ength Input	(L/N) to case (PE) (L/N) to output It to case (PE)		Basic Reinforced Functional		
dc Creepage /	Jearance	ry (L/N) to protective earth (PE) ry to secondary				
Electrical St	ranath Last '	to case to output (tested by manufacturer only)	2121 4242			VDC

12. ENVIRONMENTAL

PARA	METER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
T. Ambient Temperareture		V_{1min} to V_{1max} , V_{1nom} , V_{2Bnom} at 4000 m	0		+35	°C
\mathcal{T}_A	Ambient Temperature	$V_{1 min}$ to $V_{1 max}$, $V_{1 nom}$, $V_{2 B nom}$ at 1800 m	0		+45	°C
<i>T</i> _{Aext}	Extended Temp. Range	Derated output (see Figure 20 and Figure 33) at 1800 m	+45		+55	°C
T_S	Storage Temperature	Non-operational	-40		+70	°C
	Altitude	Operational, above Sea Level (see derating)	-		4000	m
<i>N</i> _a	Audible Noise	$V_{\text{i nom}}$, 50% $I_{\text{o nom}}$, $T_{\text{A}} = 25^{\circ}\text{C}$		60		dBA
	Cooling	System Back Pressure			0.5	in-H₂0

13. MECHANICAL

PARA	METER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
		Width		69		mm
	Dimensions	Heigth		42		mm
		Depth		555		mm
m	Weight			2.60		kg

NOTE: A 3D step file of the power supply casing is available on request.



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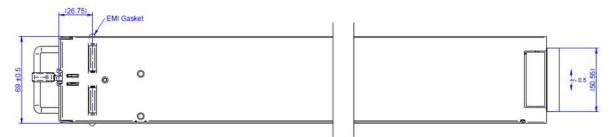


Figure 34 - Bottom view

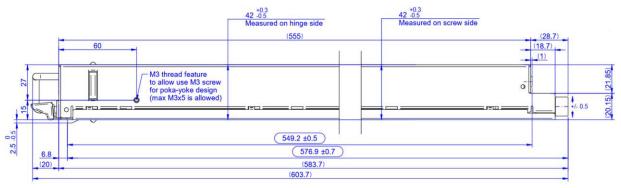


Figure 35 - Side view

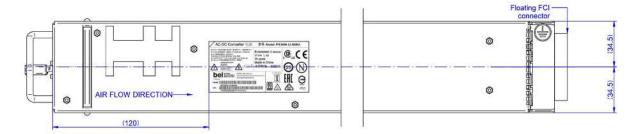


Figure 36 - Top view

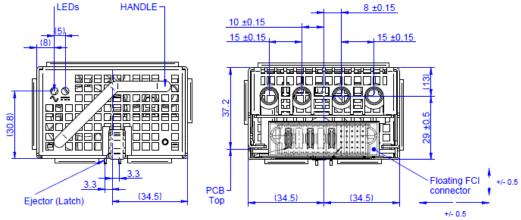
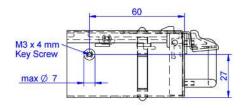


Figure 37 - Front and Rear view





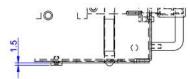
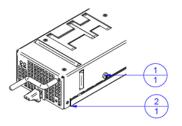


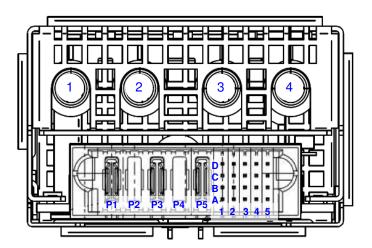
Figure 38 - PFE3000-12-069RA with Key-in screw dimension (Option code S366)



Item Number	Document Number
1	XFM.00479
2	PFE3000-12-069RA

Figure 39 - PFE3000-12-069RA with Key-in screw (Option code S366)

14. CONNECTORS



Unit: FCI Connectors P/N 51939-768LF Counterpart: FCI Connectors P/N 51915-401LF

For Main Output Pins, see section 15

Note: A1 and A2 are Trailing Pin (short pins)



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PIN	NAME	DESCRIPTION
Output		
3,4	V1	+12 VDC main output
1,2	PGND	+12 VDC main output ground
Input Pins		
P1	LIVE	AC Live Pin
P2	N.C	No metal pin connection
P3	NEUTRAL	AC Neutral Pin
P4	N.C.	No metal pin connection
P5	P.E.	Protective Earth Pin
Control Pins		
A1	PSKILL	Power supply kill (trailing pin): active-high
B1	PWOK_L	Power OK signal output: active-low
C1	INOK_L	Input OK signal: active-low
D1	PSON_L	Power supply on input: active-low
A2	PRESENT_L	Power supply present (trailing pin): active-low
B2	SGND	Signal ground* (return)
C2	SGND	Signal ground* (return)
D2	SGND	Signal ground* (return)
A3	SCL	I ² C clock signal line
B3	SDA	I ² C data signal line
C3	SMB_ALERT_L	SMB Alert signal output: active-low
D3	ISHARE	V ₁ Current share bus
A4	EEPROM_WP	EEPROM write protect
B4	RESERVED	Reserved
C4	V1_SENSE_R	Main output negative sense
D4	V1_SENSE	Main output positive sense
A5	VSB	Standby positive output
B5	VSB	Standby positive output
C5	VSB_GND	Standby Ground*
D5	VSB_GND	Standby Ground*

^{*} These pins should be connected to PGND on the system. See Section 8 for pull up resistor settings of signal pins. All signal pins are referred to SGND

Table 9 – Pin assignment



15. SHELF LEVEL CONFIGURATION (PROVISIONAL)

The recommended pin configuration below is based on company's own Shelf design and provided here as reference. Customer pin lengths within the range indicated is acceptable.

Shelf level recommendations Max 31.3 Pin GND info optional: Ø5.71 ±0.025mm Material: C14500 or eq. Plating: 4-8 micro Ag over 2-4 micro Ni ENTRY (GND) Min 21.3 ENTRY (GND) Finish: min 0.8 micro Α GND 0.5 RADIUS TANGENT TO THE Ø5.71 DIA WITHIN 3 DEGREES. NO SHARP EDGE AT TRANSITION PERMISSIBLE 15 ± 0.125 Pin 12V info optional: В Ø5.71 ±0.025mm 18 ± 0.25 12V В Material: C14500 or eq. 1 Plating: 4-8 micro Ag over 2-4 micro Ni 15 ±0.125 Finish: min 0.8 micro * 1 x30° *Min 12.8 ENTRY (12V) -0.5 RADIUS TANGENT TO THE Ø5.71 DIA WITHIN 3 DEGREES. NO SHARP EDGE AT TRANSITION PERMISSIBLE ** recommended * option ** 10.5 or *8.5 req. OFFSET 24.2 ± 0.25 FCI female shelf level connector Shelf level PCB 0.5 (0) 34.5 Shelf level Bay 34.5 (0) *0.5 569.8 -0.25 Latch slot (□9mm)



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16. ACCESSORIES

ITEM	DESCRIPTION	ORDERING PN	SOURCE
	I ² C Utility Windows Vista/7/8 compatible GUI to program, control and monitor PFE Front-Ends (and other I ² C units)	N/A	belfuse.com/power-solutions
	Single Connector Board Connector board to operate PFE3000-12-069RA unit. Includes an on-board USB to I ² C converter (use I ² C Utility as desktop software).	YTM.U0M00.0	Bel Power Solutions
	AC Can Filter Recommended AC can filter	C20F.0011	Schurter Inc.
	used on system side.	20GENG3E-R	Delta Electronics
	Key-in Screw Screw for PSU Orientation.	XFM.00479	Focus Metal



17. REVISION HISTORY

REV	DESCRIPTION	PSU PRODUCT VERSION	DATE	AUTHOR
AA	Initial Release of Datasheet.	V001 V004 V007	11-27-2013	GS
АВ	 Handle position and size has changed to a diagonal format to allow better handling/grip. +12VSB turn-on delay is changed from 2 seconds to 3 seconds. Main output will only turn on (if enabled by PSKILL and PSON) once +12VSB is in regulation. Datasheet format was changed to Bel Power Solution. 	V008	10-22-2014	GS
AC	 Added option code model in ordering information. \$101 denotes Screw for Key-in feature is added. 	V009	12-22-2014	GS
AD	 +12VSB parameter change in output ripple voltage, droop, and current read back accuracy. PSU Fans is supplied only from Internal Auxiliary. Option code is changed from S101 to S366. Added Revision History. 	V010	09-09-2015	GS
AE	 PSU Revision on product label was incremented due to internal documentation. Clarification on Dynamic Load Regulation, Mechanical Drawing and Key-in Screw accessory for option code S366. 	V011	10-28-2016	GS
AE	 Passed EAC certification and added EAC logo on product label. 	V204	04-06-2017	GS
AF	 PSKILL and SMB_ALERT_L pin active state description on section 14 was corrected but no functional change. PSU firmware was updated to support calibration of MFR Model suffix. Passed BIS certification and added BIS logo on product label. Transfer 80plus platinum logo on product label. Mechanical update on section 13 for PSU height tolerance. 	V205	05-09-2017	GS
AF1	 Mechanical update on section 13. PSU height tolerance on hinge side was adjusted to 42 +0.3/-0.5mm. Removed "80plus optional coloured label" on PSU drawing. 	V205	08-14-2017	GS
	PSU firmware was updated to improve I2C during hot plug.	V206	11-28-2017	GS
AG	 Passed KCC certification and added KCC logo on product label. A disclaimer added to the first page Figure 28. I²C / SMBus Timing updated 	V207	01-09-2018	GS

For more information on these products consult: tech.support@psbel.com

NUCLEAR AND MEDICAL APPLICATIONS - Products are not designed or intended for use as critical components in life support systems, equipment used in hazardous environments, or nuclear control systems.

TECHNICAL REVISIONS - The appearance of products, including safety agency certifications pictured on labels, may change depending on the date manufactured. Specifications are subject to change without notice.



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