

[Sample &](http://www.ti.com/product/CSD87384M?dcmp=dsproject&hqs=sandbuy&#samplebuy)

SLPS415D –SEPTEMBER 2013–REVISED MARCH 2015

CSD87384M Synchronous Buck NexFET™ Power Block II

Technical [Documents](http://www.ti.com/product/CSD87384M?dcmp=dsproject&hqs=td&#doctype2)

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- Optimized for 5 V Gate Drive driver.
- Low Switching Losses
- **TEXT ADDED FOR SPACING PACKAGE FOR SPACING SPACING SPACING CONSUMING PROPERTIES Ordering Information**^{[\(1\)](#page-0-0)}
- **RoHS Compliant**
-
-

 V_{DD}

ENABLE PWM

2 Applications

Synchronous Buck Converters

ENABLE PWM

- High Frequency Applications
- High Current, Low Duty Cycle Applications
- Multiphase Synchronous Buck Converters
- POL DC-DC Converters

1 Features 3 Description

Tools & **[Software](http://www.ti.com/product/CSD87384M?dcmp=dsproject&hqs=sw&#desKit)**

Half-Bridge Power Block

1 is a • The CSD87384M NexFET™ Power Block II is a

highly optimized design for synchronous buck highly optimized design for synchronous buck
applications offering high current and high efficiency
Up to 30 A Operation of the synchronous offering high current and high efficiency capability in a small $5.0 \, \text{mm} \times 3.5 \, \text{mm}$ outline.
Optimized for 5 V gate drive applications, this product • High Density – 5 mm x 3.5 mm LGA Footprint Optimized for 5 V gate drive applications, this product Double-Side Cooling Capability

Double-Side Cooling Capability

Ultra-Low Profile – 0.48 mm Max

with any 5 V gate drive from an external controller or with any 5 V gate drive from an external controller or

Support & [Community](http://www.ti.com/product/CSD87384M?dcmp=dsproject&hqs=support&#community)

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(1) For all available packages, see the orderable addendum at

VDD GND BOOT DRVH LL VIN VSW TG

DRVL

Driver IC CSD87384M

PGND

BG

Typical Circuit Typical Power Block Efficiency and Power Loss

 V_{IN}

 $\overline{V}_{\text{OUT}}$

Table of Contents

4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

• Changed VGS(th) from 1.0 V to 1.1 V in the Electrical Characteristics table.. [4](#page-3-2)

5 Specifications

5.1 Absolute Maximum Ratings

 $T_A = 25^{\circ}$ C (unless otherwise noted)⁽¹⁾

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated is not implied. Exposure to absolutemaximum-rated conditions for extended periods may affect device reliability.

(2) Pulse Duration ≤50 µs, duty cycle ≤0.01

 (3) Device mounted on FR4 material with 1 inch² (6.45 cm²) Cu

5.2 Recommended Operating Conditions

 $T_A = 25^{\circ}$ C (unless otherwise noted)

5.3 Power Block Performance

 $T_A = 25^{\circ}$ C (unless otherwise noted)

(1) Measurement made with six 10 µF (TDK C3216X5R1C106KT or equivalent) ceramic capacitors placed across V_{IN} to P_{GND} pins and using a high current 5 V driver IC.

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5.4 Thermal Information

 $T_A = 25^{\circ}$ C (unless otherwise stated)

(1) $R_{\theta JC}$ is determined with the device mounted on a 1 inch² (6.45 cm²), 2 oz. (0.071 mm thick) Cu pad on a 1.5 inch × 1.5 inch (3.81 cm × 3.81 cm), 0.06 inch (1.52 mm) thick FR4 board. R_{θJC} is specified by design while R_{θJA} is determined by the user's board design.

(2) Device mounted on FR4 material with 1 inch² (6.45 cm²) Cu.

5.5 Electrical Characteristics

 $T_A = 25^{\circ}$ C (unless otherwise stated)

(1) Specified by design

5.6 Typical Power Block Device Characteristics

 $T_{\rm J}$ = 125°C, unless stated otherwise. For [Figure 3](#page-4-1) and [Figure 4,](#page-4-1) the Typical Power Block System Characteristic curves are based on measurements made on a PCB design with dimensions of 4.0 inches (W) \times 3.5 inches (L) \times 0.062 inch (H) and 6 copper layers of 1 oz. copper thickness. See [Application and Implementation](#page-9-1) for detailed explanation.

Typical Power Block Device Characteristics (continued)

 T_J = 125°C, unless stated otherwise. For [Figure 3](#page-4-1) and [Figure 4,](#page-4-1) the Typical Power Block System Characteristic curves are based on measurements made on a PCB design with dimensions of 4.0 inches (W) \times 3.5 inches (L) \times 0.062 inch (H) and 6 copper layers of 1 oz. copper thickness. See [Application and Implementation](#page-9-1) for detailed explanation.

5.7 Typical Power Block MOSFET Characteristics

 $T_A = 25^{\circ}$ C, unless stated otherwise.

Typical Power Block MOSFET Characteristics (continued)

 $T_A = 25^{\circ}$ C, unless stated otherwise.

Typical Power Block MOSFET Characteristics (continued)

 $T_A = 25^{\circ}$ C, unless stated otherwise.

EXAS NSTRUMENTS

6 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI ' s customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

6.1 Application Information

The CSD87384M NexFET™ power block is an optimized design for synchronous buck applications using 5 V gate drive. The Control FET and Sync FET silicon are parametrically tuned to yield the lowest power loss and highest system efficiency. As a result, a new rating method is needed which is tailored toward a more systemscentric environment. System level performance curves such as Power Loss, Safe Operating Area, and normalized graphs allow engineers to predict the product performance in the actual application.

6.2 Power Loss Curves

MOSFET-centric parameters such as $R_{DS(ON)}$ and Q_{gd} are needed to estimate the loss generated by the devices. To simplify the design process for engineers, TI has provided measured power loss performance curves. [Figure 1](#page-4-1) plots the power loss of the CSD87384M as a function of load current. This curve is measured by configuring and running the CSD87384M as it would be in the final application (see [Figure 27](#page-10-0)). The measured power loss is the CSD87384M loss and consists of both input conversion loss and gate drive loss. [Equation 1](#page-9-5) is used to generate the power loss curve.

$$
(\mathsf{V}_{\mathsf{IN}} \times \mathsf{I}_{\mathsf{IN}}) + (\mathsf{V}_{\mathsf{DD}} \times \mathsf{I}_{\mathsf{DD}}) - (\mathsf{V}_{\mathsf{SW_AVG}} \times \mathsf{I}_{\mathsf{OUT}}) = \mathsf{Power}\ \mathsf{Loss} \tag{1}
$$

The power loss curve in [Figure 1](#page-4-1) is measured at the maximum recommended junction temperatures of 125°C under isothermal test conditions.

6.3 Safe Operating Curves (SOA)

The SOA curves in the CSD87384M data sheet provide guidance on the temperature boundaries within an operating system by incorporating the thermal resistance and system power loss. [Figure 3](#page-4-1) to [Figure 4](#page-4-1) outline the temperature and airflow conditions required for a given load current. The area under the curve dictates the safe operating area. All the curves are based on measurements made on a PCB design with dimensions of 4 inches (W) \times 3.5 inches (L) \times 0.062 inch (T) and 6 copper layers of 1 oz. copper thickness.

6.4 Normalized Curves

The normalized curves in the CSD87384M data sheet provide guidance on the Power Loss and SOA adjustments based on their application-specific needs. These curves show how the power loss and SOA boundaries adjust for a given set of systems conditions. The primary y-axis is the normalized change in power loss and the secondary y-axis is the change in system temperature required to comply with the SOA curve. The change in power loss is a multiplier for the Power Loss curve and the change in temperature is subtracted from the SOA curve.

Normalized Curves (continued)

Figure 27. Typical Application

CSD87384M

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6.5 Calculating Power Loss and SOA

The user can estimate product loss and SOA boundaries by arithmetic means (see [Design Example](#page-11-1)). Though the Power Loss and SOA curves in this data sheet are taken for a specific set of test conditions, the following procedure outlines the steps the user should take to predict product performance for any set of system conditions.

6.5.1 Design Example

Operating Conditions:

- Output Current = 20 A
- Input Voltage $=$ 4 V
- Output Voltage = 1 V
- Switching Frequency = 800 kHz
- Inductor = 0.2μ H

6.5.2 Calculating Power Loss

- Power Loss at 20 A = 3.5 W [\(Figure 1](#page-4-1))
- Normalized Power Loss for input voltage \approx 1.18 ([Figure 6\)](#page-4-2)
- Normalized Power Loss for output voltage \approx 0.94 ([Figure 7\)](#page-4-2)
- Normalized Power Loss for switching frequency ≈ 1.15 ([Figure 5\)](#page-4-2)
- Normalized Power Loss for output inductor \approx 1.02 [\(Figure 8](#page-4-2))
- **Final calculated Power Loss = 3.5 W × 1.18 × 0.94 × 1.15 × 1.02 ≈ 4.6 W**

6.5.3 Calculating SOA Adjustments

- SOA adjustment for input voltage ≈ 1.5^oC [\(Figure 6](#page-4-2))
- SOA adjustment for output voltage \approx –0.5 \degree C ([Figure 7\)](#page-4-2)
- SOA adjustment for switching frequency ≈ 1.2 °C [\(Figure 5](#page-4-2))
- SOA adjustment for output inductor $≈ 0.2°C$ ([Figure 8\)](#page-4-2)
- **Final calculated SOA adjustment = 1.5 + (–0.5) + 1.2 + 0.2 ≈ 2.4ºC**

In the previous design example, the estimated power loss of the CSD87384M would increase to 4.6 W. In addition, the maximum allowable board and/or ambient temperature would have to decrease by 2.4ºC. [Figure 28](#page-11-2) graphically shows how the SOA curve would be adjusted accordingly.

- 1. Start by drawing a horizontal line from the application current to the SOA curve.
- 2. Draw a vertical line from the SOA curve intercept down to the board or ambient temperature.
- 3. Adjust the SOA board/ambient temperature by subtracting the temperature adjustment value.

In the design example, the SOA temperature adjustment yields a reduction in allowable board/ambient temperature of 2.4ºC. In the event the adjustment value is a negative number, subtracting the negative number would yield an increase in allowable board or ambient temperature.

Figure 28. Power Block SOA

7 Layout

7.1 Layout Guidelines

7.1.1 Recommended PCB Design Overview

There are two key system-level parameters that can be addressed with a proper PCB design: electrical and thermal performance. Properly optimizing the PCB layout yields maximum performance in both areas. A brief description on how to address each parameter is provided.

7.1.2 Electrical Performance

The CSD87384M has the ability to switch voltages at rates greater than 10 kV/µs. Take special care with the PCB layout design and placement of the input capacitors, inductor, and output capacitors.

- The placement of the input capacitors relative to VIN and PGND pins of CSD87384M device should have the highest priority during the component placement routine. It is critical to minimize these node lengths. As such, ceramic input capacitors need to be placed as close as possible to the VIN and PGND pins (see [Figure 29\)](#page-12-3). The example in [Figure 29](#page-12-3) uses 1 \times 10 nF 0402 25 V and 4 \times 10 µF 1206 25 V ceramic capacitors (TDK part number C3216X5R1C106KT or equivalent). Notice there are ceramic capacitors on both sides of the board with an appropriate amount of vias interconnecting both layers. In terms of priority of placement next to the Power Stage C21, C5, C8, C19, and C18 should follow in order.
- The switching node of the output inductor should be placed relatively close to the Power Block II CSD87384M VSW pins. Minimizing the VSW node length between these two components will reduce the PCB conduction losses and actually reduce the switching noise level. See [Figure 29.](#page-12-3) (1)

7.1.3 Thermal Performance

The CSD87384M has the ability to utilize the PGND planes as the primary thermal path. As such, the use of thermal vias is an effective way to pull away heat from the device and into the system board. Concerns of solder voids and manufacturability problems can be addressed by the use of three basic tactics to minimize the amount of solder attach that wicks down the via barrel:

- Intentionally space out the vias from each other to avoid a cluster of holes in a given area.
- Use the smallest drill size allowed in your design. The example in [Figure 29](#page-12-3) uses vias with a 10 mil drill hole and a 16 mil capture pad.
- Tent the opposite side of the via with solder-mask.

The number and drill size of the thermal vias should align with the end user's PCB design rules and manufacturing capabilities.

7.2 Layout Example

Figure 29. Recommended PCB Layout (Top Down View)

(1) Keong W. Kam, David Pommerenke, "EMI Analysis Methods for Synchronous Buck Converter EMI Root Cause Analysis", University of Missouri – Rolla

EXAS STRUMENTS

8 Device and Documentation Support

8.1 Trademarks

NexFET is a trademark of Texas Instruments. All other trademarks are the property of their respective owners.

8.2 Electrostatic Discharge Caution

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

8.3 Glossary

[SLYZ022](http://www.ti.com/lit/pdf/SLYZ022) — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

9 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation

9.1 CSD87384M Package Dimensions

Bottom View

Pin Configuration

Texas **INSTRUMENTS**

9.2 Land Pattern Recommendation

9.3 Stencil Recommendation (100 µm)

9.4 Stencil Recommendation (125 µm)

For recommended circuit layout for PCB designs, see application note [SLPA005](http://www.ti.com/lit/pdf/SLPA005) – Reducing Ringing Through PCB Layout Techniques.

9.5 Pin Drawing

9.6 CSD87384M Embossed Carrier Tape Dimensions

(1) Pin 1 is oriented in the top-left quadrant of the tape enclosure (closest to the carrier tape sprocket holes).

PACKAGING INFORMATION

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

TEXAS

ISTRUMENTS

TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

Pack Materials-Page 1

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PACKAGE MATERIALS INFORMATION

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*All dimensions are nominal

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