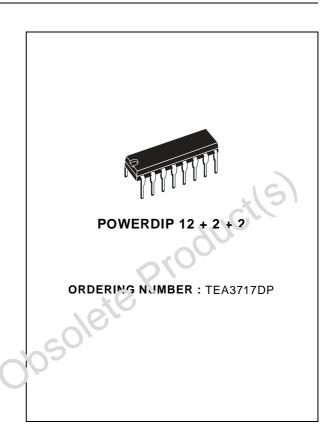




STEPPER MOTOR DRIVER

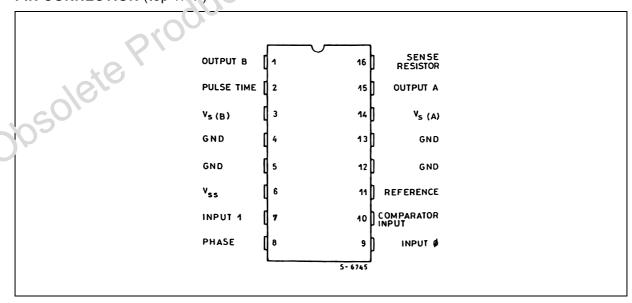
- HALF-STEP AND FULL-STEP MODE
- BIPOLAR DRIVE OF STEPPER MOTOR FOR MAXIMUM MOTOR PERFORMANCE
- BUILT-IN PROTECTION DIODES
- WIDE RANGE OF CURRENT CONTROL 5 TO 1000 mA
- WIDE VOLTAGE RANGE 10 TO 45 V
- DESIGNED FOR UNSTABILIZED MOTOR SUPPLY VOLTAGE
- CURRENT LEVELS CAN BE SELECTED IN STEPS OR VARIED CONTINUOUSLY



DESCRIPTION

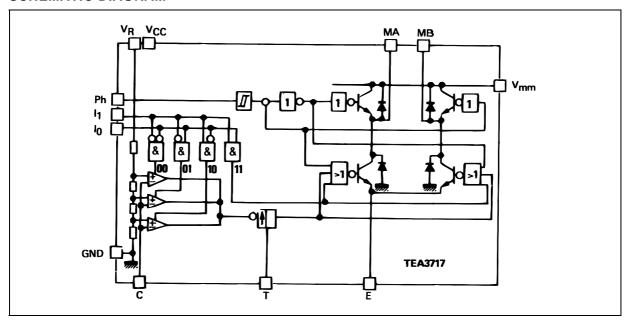
The TEA3717 is a bipolar monolithic integrated circuit intended to control and drive the current in one winding of a bipolar stepper motor. The circuit consists of an LS-TTL compatible logic input, a current sensor, a monostable and an output stage with built-in protection diodes. Two TEA3717 and a few external components form a complete control and drive unit for LS-TTL or microprocessor-controlled stepper motor systems.

PIN CONNECTION (top view)



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SCHEMATIC DIAGRAM



ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V _{mm}	Power Supply Voltage (pins 14, 3)	45	V
V _{CC}	Logic Supply Voltage (pin 6)	7	V
V _{in} V _{in} V _V	Input Voltage Logic Inputs Analog Inputs Reference Input	- 0.5 to 6 V _{CC} 15	V
I _{in} I _{in}	Input Current Logic Inputs Analog Inputs	- 10 - 10	mA
lo	Output Current	± 1	Α
Tj	Junction Temperature	+ 150	°C
T _{stg}	Storage Temperature Range	- 55 to + 150	°C
T _{oper}	Operating Ambiant Temperature Range	0 to + 70	°C

THERMAL DATA

Symbol	Parameter	Value	Unit
R _{th (j-c)}	Maximum Junction-pins Thermal Resistance	11	°C/W
R _{th (j-a)}	Maximum Junction-ambient Thermal Resistance	45*	°C/W

^{*} Soldered on a 35 mm thick 20 cm³ PC board copper area

RECOMMENDED OPERATING CONDITIONS

Symbol	Parameter	Min.	Тур.	Max.	Unit
V _{CC}	Supply Voltage	4.75	5	5.25	V
V_{mm}	Supply Voltage	10	-	40	V
lo	Output Current	0.020	_	0.8	Α
T _{amb}	Ambient Temperature	0	_	70	°C
t _r	Rise Time, Logic Inputs	_	_	3	μS
t _f	Fall Time, Logic Inputs	_	_	3	μs

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ELECTRICAL CHARACTERISTICS

 V_{CC} = 5V, $\pm 5\%$, V_{mm} = + 10V to + 40V, T_{amb} = 0°C to + 70°C (unless otherwise specified)

Symbol	Parameter	Min.	Тур.	Max.	Unit
Icc	Supply Current	-	-	25	mA
V _{IH}	High Level Input Voltage - Logic Inputs	2.0	-	-	V
V_{IL}	Low Level Input Voltage - Logic Inputs	_	_	0.8	V
I _{IH}	High Level Input Current - Logic Input $(V_1 = +2.4V)$	-	_	20	μΑ
I _{IL}	Low Level Input Current - Logic Inputs (V _I = + 0.4V)	- 0.4	_	-	mA
V _{CH} V _{CM} V _{CL}	Comparator Threshold Voltage ($V_R = +5.0V$), $I_0 = 0$, $I_1 = 0$ $I_0 = 1$, $I_1 = 0$ $I_0 = 0$, $I_1 = 1$	390 230 65	420 250 80	440 270 90	mV
Ico	Comparator Input Current	- 20	_	+ 20	μΑ
l _{off}	Output Leakage Current ($I_0 = 1$, $I_1 = 1$) $T_{amb} = + 25^{\circ}C$ $T_{amb} = + 70^{\circ}C$, $V_S = 40V$, $V_{SS} = 5V$		_ 100	100 200	μΑ
V _{sat}	Total Saturation Voltage Drop (I _o = 500mA)	_	_	4.0	V
P _{tot}	Total Power Dissipation $ I_{o} = 500 \text{mA}, f_{s} = 30 \text{kHz} \\ I_{o} = 800 \text{mA}, f_{s} = 30 \text{kHz} $	_ _	1.8 3.7	2.3	W
t _{off}	Cut off Time (see figure 1 and 2, $V_{mm} = + 10V$, $t_{on} \ge 5\mu s$)	25	30	35	μs
t _d	Turn off Delay (see figure 1 and 2, T_{amb} = + 25°C, $dVC/dt \ge 50mV/\mu s$)	_	1.6		μs

Figure 1 (see note)

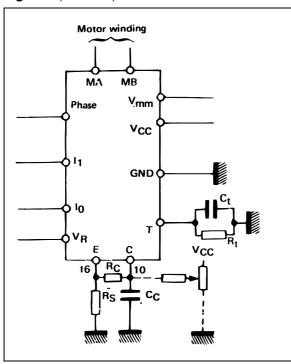
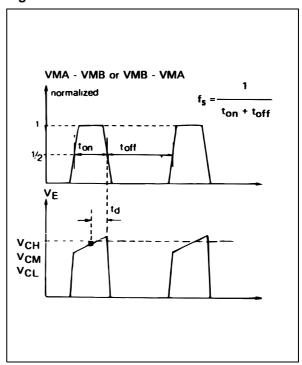


Figure 2.



FUNCTIONAL DESCRIPTION

The circuit is intented to drive a bipolar constant current through one motor winding. The constant current is generated through switch mode regulation.

There is a choice of three different current levels with the two logic inputs l_0 and l_1 . The current can also be switched off completely.

INPUT LOGIC

If any of the logic inputs is left open, the circuit will treat it as a high level input.

I ₀	I ₁	Current Level
H L H L	II	No Current Low Current Medium Current Maximum Current

PHASE – This input determines the direction of current flow in the winding, depending on the motor connections. The signal is fed through a Schmidttrigger for noise immunity, and through a time delay in order to guarantee that no short-circuit occurs in the output stage during phase-shift. High level on the PHASE-input causes the motor current flow from MA through the winding to MB.

 l_0 and l_1 — The current level in the motor winding is selected with these inputs. The values of the different current levels are determined by the reference voltage V_R together with the value of the sensing resistor R_S .

CURRENT SENSOR

This part contains a current sensing resistor (Rs), a low pass filter (Rc, Cc) and three comparators. Only one comparator is active at a time. It is activated by the input logic according to the current level chosen with signals l_0 and $l_1.$ The motor current flows through the sensing resistor $R_S.$ When the current has increased so that the voltage across R_S becomes higher than the reference voltage on the

Note : $R_S = 1 \Omega$, inductance free $R_C = 1 k\Omega$ $C_C = 820 \text{ pF}$, ceramic $R_t = 56 \text{ } k\Omega$

other comparator input, the comparator output goes high, which triggers the pulse generator and its output goes high during a fixed pulse time (toff), thus switching off the power feed to the motor winding, and causing the motor current to decrease during toff

SINGLE-PULSE GENERATOR

The pulse generator is a monostable triggered on the positive going edge of the comparator output. The monostable output is high during the pulse time, $t_{\rm off}$, which is determined by the timing components R_t and C_t .

$$t_{\text{off}} = 0.69 \cdot R \cdot Ct$$

The single pulse switches off the power feed to the motor winding, causing the winding current to decrease during toff.

If a new trigger signal should occur during toff, it is ignored.

OUTPUT STAGE

The output stage contains four Darlington transistors and four diodes, connected in an H-bridge. The two sinking transistors are used to switch the powersupplied to the motor winding, thus driving a constant current through the winding.

It should be noted however, that it is not permitted to short circuit the outputs.

Vcc, V_{mm}, V_R

The circuit will stand any order of turn-on or turn-off of the supply voltages V_{SS} and V_{S} . Normal dV/dt values are then assumed.

Preferably, V_R should be tracking V_{CC} during power-on and power-off.

ANALOG CONTROL

The current levels can be varied continuously either if V_R is varied or with a circuit varying the voltage fed into the comparator terminal (see fig.1).

Figure 3

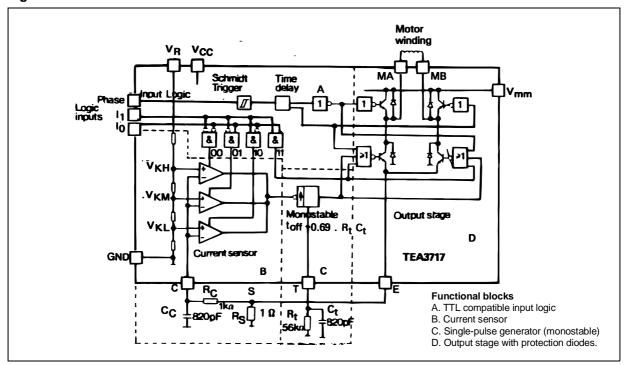


Figure 4: Typical Sink Saturation Voltage versus Output Current

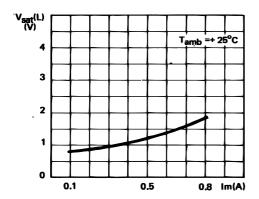


Figure 6: Typical Power Losses versus Output Current

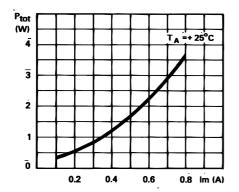
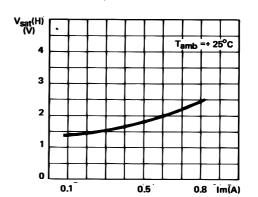


Figure 5: Typical Source Saturation Voltage versus Output Current



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TYPICAL APPLICATION

Figure 7: Serial Printer Carriage Drive.

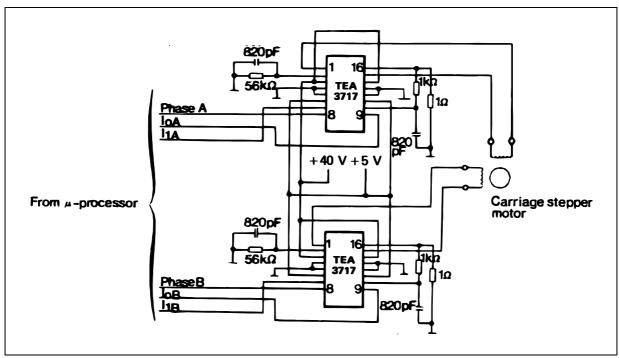
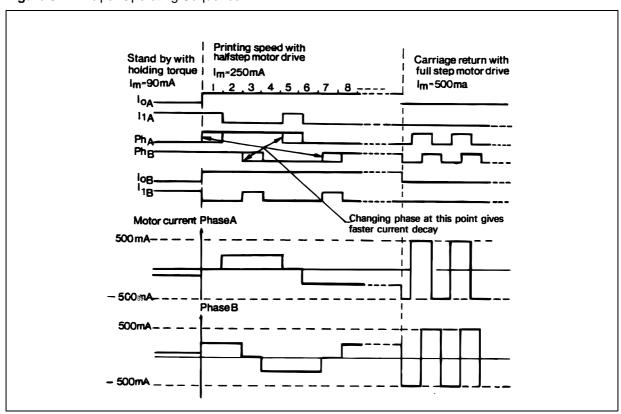


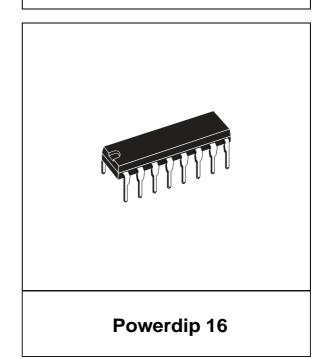
Figure 8 : Principal Operating Sequence.

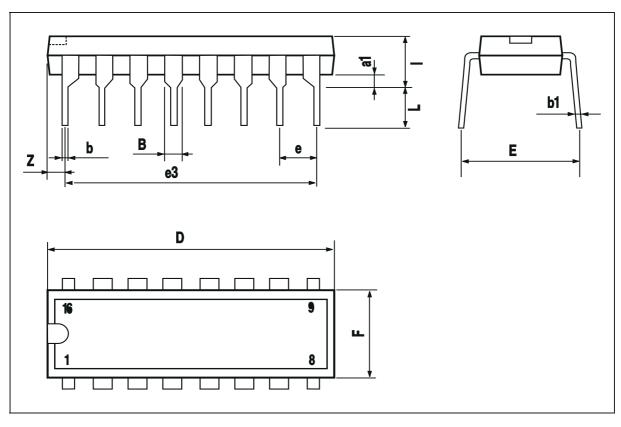


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DIM.		mm		inch			
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
a1	0.51			0.020			
В	0.85		1.40	0.033		0.055	
b		0.50			0.020		
b1	0.38		0.50	0.015		0.020	
D			20.0			0.787	
Е		8.80			0.346		
е		2.54			0.100		
e3		17.78			0.700		
F			7.10			0.280	
I			5.10			0.201	
L		3.30			0.130		
Z			1.27			0.050	

OUTLINE AND MECHANICAL DATA





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