

# S-1131 Series

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# HIGH RIPPLE-REJECTION AND LOW DROPOUT MIDDLE OUTPUT CURRENT CMOS VOLTAGE REGULATOR

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Rev.4.1\_02

The S-1131 Series is a positive voltage regulator with a low dropout voltage, high-accuracy output voltage, and low current consumption developed based on CMOS technology.

A built-in low on-resistance transistor provides a low dropout voltage and large output current, and a built-in overcurrent protection circuit prevents the load current from exceeding the current capacitance of the output transistor. An ON/OFF circuit ensures a long battery life, and small SOT-89-3, SOT-89-5 and 6-Pin HSON(A) packages realize high-density mounting.

# Features

- Output voltage:
- Output voltage accuracy:
- Dropout voltage:
- Current consumption:
- Output current:
- Ripple rejection:
- Built-in overcurrent protection circuit:
- Built-in ON/OFF circuit:
- Operation temperature range:
- Lead-free, Sn 100%, halogen-free\*2

1.5 V to 5.5 V, selectable in 0.1 V step  $\pm 1.0\%$ 250 mV typ. (3.0 V output product,  $I_{OUT} = 100$  mA) During operation: 35  $\mu$ A typ., 65  $\mu$ A max. During power-off: 0.1  $\mu$ A typ., 1.0  $\mu$ A max. Possible to output 300 mA ( $V_{IN} \ge V_{OUT(S)} + 1.0$  V)<sup>\*1</sup> 70 dB typ. (f = 1.0 kHz) Limits overcurrent of output transistor. Ensures long battery life. Ta = -40°C to +85°C

\*1. Attention should be paid to the power dissipation of the package when the output current is large. \*2. Refer to "■ Product Name Structure" for details.

# Applications

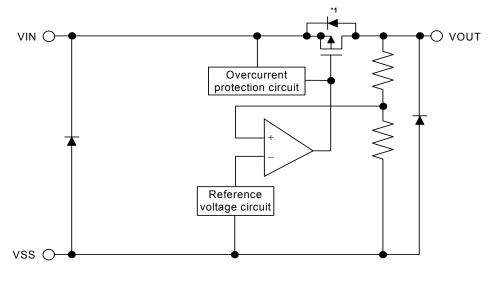
- Constant-voltage power supply for DVD and CD-ROM drive
- Constant-voltage power supply for battery-powered device
- Constant-voltage power supply for personal communication device
- Constant-voltage power supply for note book PC

# Packages

- SOT-89-3
- SOT-89-5
- 6-Pin HSON(A)

# Block Diagrams

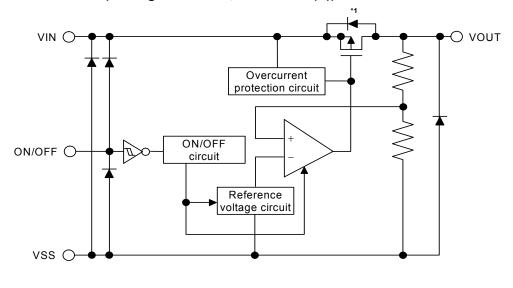
(1) Without ON/OFF circuit (Package: SOT-89-3)



\*1. Parasitic diode

Figure 1

#### (2) With ON/OFF circuit (Package: SOT-89-5, 6-Pin HSON(A))



\*1. Parasitic diode

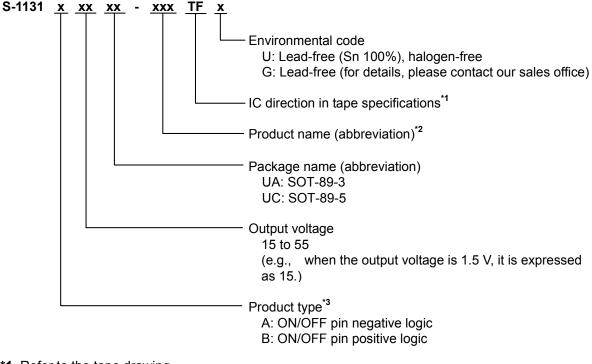
Figure 2

#### Product Name Structure

• Users can select the product type, output voltage, and package type for the S-1131 Series. Refer to "1. **Product name**" regarding the contents of product name, "2. **Packages**" regarding the package drawings, "3. **Product name list**" regarding details of the product name.

#### 1. Product name

(1) SOT-89-3, SOT-89-5

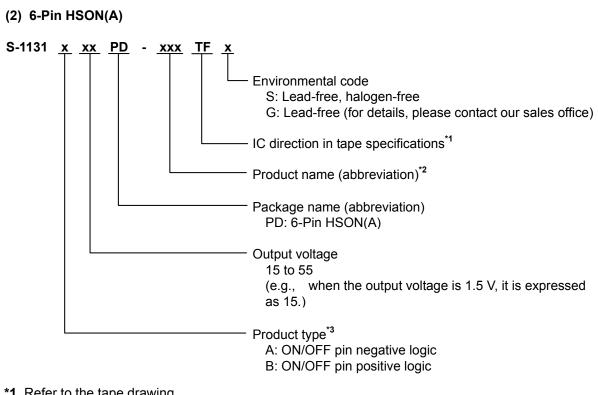


**\*1.** Refer to the tape drawing.

\*2. Refer to the product name list.

\*3. Refer to "3. ON/OFF pin" in "■ Operation" (Expect SOT-89-3).

#### HIGH RIPPLE-REJECTION LOW DROPOUT MIDDLE OUTPUT CURRENT CMOS VOLTAGE REGULATOR S-1131 Series Rev.4.1\_02



- \*1. Refer to the tape drawing.
- \*2. Refer to the product name list.
- \*3. Refer to "3. ON/OFF pin" in "■ Operation"

#### 2. Packages

Daakaga Nama	Drawing Code			
Package Name	Package	Таре	Reel	
SOT-89-3	UP003-A-P-SD	UP003-A-C-SD	UP003-A-R-SD	
SOT-89-5	UP005-A-P-SD	UP005-A-C-SD	UP005-A-R-SD	
6-Pin HSON(A)	PD006-A-P-SD	PD006-A-C-SD	PD006-A-R-SD	

# HIGH RIPPLE-REJECTION LOW DROPOUT MIDDLE OUTPUT CURRENT CMOS VOLTAGE REGULATOR Rev.4.1\_02 S-1131 Series

#### 3. Product name list

Table 1

Output voltage	SOT-89-3	SOT-89-5	6-Pin HSON(A)
1.5V±1.0%	S-1131B15UA-N4ATFx	S-1131B15UC-N4ATFx	S-1131B15PD-N4ATFz
1.6V±1.0%	S-1131B16UA-N4BTFx	S-1131B16UC-N4BTFx	S-1131B16PD-N4BTFz
1.7V±1.0%	S-1131B17UA-N4CTFx	S-1131B17UC-N4CTFx	S-1131B17PD-N4CTFz
1.8V±1.0%	S-1131B18UA-N4DTFx	S-1131B18UC-N4DTFx	S-1131B18PD-N4DTFz
1.9V±1.0%	S-1131B19UA-N4ETFx	S-1131B19UC-N4ETFx	S-1131B19PD-N4ETFz
2.0V±1.0%	S-1131B20UA-N4FTFx	S-1131B20UC-N4FTFx	S-1131B20PD-N4FTFz
2.1V±1.0%	S-1131B21UA-N4GTFx	S-1131B21UC-N4GTFx	S-1131B21PD-N4GTFz
2.2V±1.0%	S-1131B22UA-N4HTFx	S-1131B22UC-N4HTFx	S-1131B22PD-N4HTFz
2.3V±1.0%	S-1131B23UA-N4ITFx	S-1131B23UC-N4ITFx	S-1131B23PD-N4ITFz
2.4V±1.0%	S-1131B24UA-N4JTFx	S-1131B24UC-N4JTFx	S-1131B24PD-N4JTFz
2.5V±1.0%	S-1131B25UA-N4KTFx	S-1131B25UC-N4KTFx	S-1131B25PD-N4KTFz
2.6V±1.0%	S-1131B26UA-N4LTFx	S-1131B26UC-N4LTFx	S-1131B26PD-N4LTFz
2.7V±1.0%	S-1131B27UA-N4MTFx	S-1131B27UC-N4MTFx	S-1131B27PD-N4MTFz
2.8V±1.0%	S-1131B28UA-N4NTFx	S-1131B28UC-N4NTFx	S-1131B28PD-N4NTFz
2.9V±1.0%	S-1131B29UA-N4OTFx	S-1131B29UC-N4OTFx	S-1131B29PD-N4OTFz
3.0V±1.0%	S-1131B30UA-N4PTFx	S-1131B30UC-N4PTFx	S-1131B30PD-N4PTFz
3.1V±1.0%	S-1131B31UA-N4QTFx	S-1131B31UC-N4QTFx	S-1131B31PD-N4QTFz
3.2V±1.0%	S-1131B32UA-N4RTFx	S-1131B32UC-N4RTFx	S-1131B32PD-N4RTFz
3.3V±1.0%	S-1131B33UA-N4STFx	S-1131B33UC-N4STFx	S-1131B33PD-N4STFz
3.4V±1.0%	S-1131B34UA-N4TTFx	S-1131B34UC-N4TTFx	S-1131B34PD-N4TTFz
3.5V±1.0%	S-1131B35UA-N4UTFx	S-1131B35UC-N4UTFx	S-1131B35PD-N4UTFz
3.6V±1.0%	S-1131B36UA-N4VTFx	S-1131B36UC-N4VTFx	S-1131B36PD-N4VTFz
3.7V±1.0%	S-1131B37UA-N4WTFx	S-1131B37UC-N4WTFx	S-1131B37PD-N4WTFz
3.8V±1.0%	S-1131B38UA-N4XTFx	S-1131B38UC-N4XTFx	S-1131B38PD-N4XTFz
3.9V±1.0%	S-1131B39UA-N4YTFx	S-1131B39UC-N4YTFx	S-1131B39PD-N4YTFz
4.0V±1.0%	S-1131B40UA-N4ZTFx	S-1131B40UC-N4ZTFx	S-1131B40PD-N4ZTFz
4.1V±1.0%	S-1131B41UA-N5ATFx	S-1131B41UC-N5ATFx	S-1131B41PD-N5ATFz
4.2V±1.0%	S-1131B42UA-N5BTFx	S-1131B42UC-N5BTFx	S-1131B42PD-N5BTFz
4.3V±1.0%	S-1131B43UA-N5CTFx	S-1131B43UC-N5CTFx	S-1131B43PD-N5CTFz
4.4V±1.0%	S-1131B44UA-N5DTFx	S-1131B44UC-N5DTFx	S-1131B44PD-N5DTFz
4.5V±1.0%	S-1131B45UA-N5ETFx	S-1131B45UC-N5ETFx	S-1131B45PD-N5ETFz
4.6V±1.0%	S-1131B46UA-N5FTFx	S-1131B46UC-N5FTFx	S-1131B46PD-N5FTFz
4.7V±1.0%	S-1131B47UA-N5GTFx	S-1131B47UC-N5GTFx	S-1131B47PD-N5GTFz
4.8V±1.0%	S-1131B48UA-N5HTFx	S-1131B48UC-N5HTFx	S-1131B48PD-N5HTFz
4.9V±1.0%	S-1131B49UA-N5ITFx	S-1131B49UC-N5ITFx	S-1131B49PD-N5ITFz
5.0V±1.0%	S-1131B50UA-N5JTFx	S-1131B50UC-N5JTFx	S-1131B50PD-N5JTFz
5.1V±1.0%	S-1131B51UA-N5KTFx	S-1131B51UC-N5KTFx	S-1131B51PD-N5KTFz
5.2V±1.0%	S-1131B52UA-N5LTFx	S-1131B52UC-N5LTFx	S-1131B52PD-N5LTFz
5.3V±1.0%	S-1131B53UA-N5MTFx	S-1131B53UC-N5MTFx	S-1131B53PD-N5MTFz
5.4V±1.0%	S-1131B54UA-N5NTFx	S-1131B54UC-N5NTFx	S-1131B54PD-N5NTFz
5.5V±1.0%	S-1131B55UA-N5OTFx	S-1131B55UC-N5OTFx	S-1131B55PD-N5OTFz

**Remark1.** Please contact our sales office for type A products.

- **2.** x: G or U
  - z: G or S
- **3.** Please select products of environmental code = U for Sn 100%, halogen-free products.

# ■ Pin Configurations

SOT-89-3

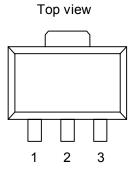


Table 2

Pin No.	Symbol	Description
1	VOUT	Output voltage pin
2	VSS	GND pin
3	VIN	Input voltage pin

Figure 3



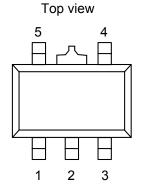


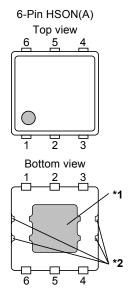
Figure 4

Table 3

Pin No.	Symbol	Description
1	VOUT	Output voltage pin
2	VSS	GND pin
3	NC <sup>*1</sup>	No connection
4	ON/OFF	ON/OFF pin
5	VIN	Input voltage pin

\*1. The NC pin is electrically open.

The NC pin can be connected to the VIN pin or the VSS pin.



\*1. Connect the exposed thermal die pad at shadowed area to the board, and set electric potential open or VSS. However, do not use it as the function of electrode.

\*2. Be careful of the contact with other wires because the pinch lead has the same electric potential as VSS.

#### Figure 5

# ■ Absolute Maximum Ratings

Table 4

Pin No.	Symbol	Description
1	VOUT	Output voltage pin
2	VSS	GND pin
3	NC <sup>*1</sup>	No connection
4	NC <sup>*1</sup>	No connection
5	ON/OFF	ON/OFF pin
6	VIN	Input voltage pin

\*1. The NC pin is electrically open.

The NC pin can be connected to the VIN pin or the VSS pin.

Table 5

			(Ta = 25°C unless other	wise specified)
Item		Symbol	Absolute Maximum Rating	Unit
Input voltage		V <sub>IN</sub>	$V_{SS}$ – 0.3 to $V_{SS}$ + 7	V
input voltage	Input voltage		$V_{SS}$ – 0.3 to $V_{IN}$ + 0.3	V
Output voltag	Output voltage		$V_{SS}$ – 0.3 to $V_{IN}$ + 0.3	V
Power	SOT-89-3		500	mW
dissipation	SOT-89-5	PD	500	mW
uissipation	6-Pin HSON(A)		500	mW
Operation ambient temperature		T <sub>opr</sub>	-40 to +85	°C
Storage temperature		T <sub>stg</sub>	-40 to +125	°C

Caution The absolute maximum ratings are rated values exceeding which the product could suffer physical damage. These values must therefore not be exceeded under any conditions.

# HIGH RIPPLE-REJECTION LOW DROPOUT MIDDLE OUTPUT CURRENT CMOS VOLTAGE REGULATOR S-1131 Series Rev.4.1\_02

Table 6

# Electrical Characteristics

			Table 6	a = 25°C	unless c	otherwise	e spe	cified)
Item	Symbol	Conditions		Min.	Тур.	Max.	Unit	Test Circuit
Output voltage <sup>*1</sup>	V <sub>OUT(E)1</sub>	$V_{IN} = V_{OUT(S)} + 1.0 \text{ V}, I_{OUT} = 30 \text{ mA}$		$\begin{array}{c} V_{OUT(S)} \\ \times \ 0.99 \end{array}$	$V_{\text{OUT}(S)}$	$V_{OUT(S)} \times 1.01$	V	1
	V <sub>OUT(E)2</sub>	$V_{IN}$ = $V_{OUT(S)}$ + 1.0	V, I <sub>OUT</sub> = 80 mA	$\begin{array}{c} V_{OUT(S)} \\ \times \ 0.98 \end{array}$	V <sub>OUT(S)</sub>	$V_{OUT(S)} \times 1.02$	V	1
Output current <sup>*2</sup>	I <sub>OUT</sub>	$V_{IN} \ge V_{OUT(S)} + 1.0$		300 <sup>*5</sup>	—		mA	3
			V <sub>OUT(S)</sub> = 1.5 V		1.00	1.05	V	1
			V <sub>OUT(S)</sub> = 1.6 V		0.90	0.95	V	1
			V <sub>OUT(S)</sub> = 1.7 V		0.80	0.85	V	1
			V <sub>OUT(S)</sub> = 1.8 V		0.70	0.75	V	1
Dropout voltage <sup>*3</sup>	V <sub>drop</sub>	I <sub>OUT</sub> = 100 mA	V <sub>OUT(S)</sub> = 1.9 V	—	0.60	0.65	V	1
Diopoul vollage	• drop	1001 - 100 1114	V <sub>OUT(S)</sub> = 2.0 V	—	0.50	0.60	V	1
			V <sub>OUT(S)</sub> = 2.1 V	—	0.40	0.55	V	1
			$2.2~V \leq V_{OUT(S)} \leq 2.5~V$		0.30	0.49	V	1
			$2.6~V \leq V_{OUT(S)} \leq 3.3~V$		0.25	0.34	V	1
			$3.4 \text{ V} \leq V_{OUT(S)} \leq 5.5 \text{ V}$	_	0.20	0.28	V	1
Line regulation	$\frac{\Delta V \text{OUT1}}{\Delta V \text{IN} \bullet V \text{OUT}}$	$\label{eq:V_OUT(S)} \begin{array}{l} V_{OUT(S)} + 0.5 \ V \leq V_{IN} \leq 6.5 \ V, \\ I_{OUT} = 80 \ mA \end{array}$		_	0.05	0.2	%/V	1
Load regulation	$\Delta V_{OUT2}$	$\label{eq:VIN} \begin{split} V_{\text{IN}} &= V_{\text{OUT}(S)} + 1.0 \text{ V}, \\ 1.0 \text{ mA} \leq I_{\text{OUT}} \leq 80 \text{ mA} \end{split}$		_	20	40	mV	1
Output voltage temperature coefficient*4	$\frac{\Delta V_{OUT}}{\Delta Ta \bullet V_{OUT}}$	$V_{IN} = V_{OUT(S)} + 1.0 \text{ V}, I_{OUT} = 10 \text{ mA}, -40^{\circ}\text{C} \le \text{Ta} \le 85^{\circ}\text{C}$		_	±100		ppm /°C	1
Current consumption during operation	I <sub>SS1</sub>	$V_{IN} = V_{OUT(S)} + 1.0 \text{ V}, \text{ ON/OFF pin} = \text{ON},$ no load		_	35	65	μA	2
Input voltage	V <sub>IN</sub>	_		2.0		6.5	V	_
Ripple rejection	RR	$V_{IN}$ = $V_{OUT(S)}$ + 1.0 V, f = 1.0 kHz, $\Delta V_{rip}$ = 0.5 Vrms, I <sub>OUT</sub> = 80 mA			70		dB	5
Short-circuit current	I <sub>short</sub>	$V_{IN} = V_{OUT(S)} + 1.0 V$ , ON/OFF pin = ON, $V_{OUT} = 0 V$			450		mA	3
Current consumption during power-off	I <sub>SS2</sub>	$V_{IN} = V_{OUT(S)} + 1.0 V$ , ON/OFF pin = OFF, no load			0.1	1.0	μA	2
ON/OFF pin input voltage "H"	$V_{SH}$	$V_{\text{IN}}$ = $V_{\text{OUT}(S)}$ + 1.0 V, R <sub>L</sub> = 1.0 k $\Omega$		1.5			V	4
ON/OFF pin input voltage "L"	V <sub>SL</sub>	$V_{\text{IN}} = V_{\text{OUT}(S)} + 1.0 \text{ V}, \text{ R}_{\text{L}} = 1.0 \text{ k}\Omega$				0.3	V	4
ON/OFF pin input current "H"	I <sub>SH</sub>	V <sub>IN</sub> = 6.5 V, V <sub>ON/OFF</sub> = 6.5 V		-0.1		0.1	μA	4
ON/OFF pin input current "L"	I <sub>SL</sub>	V <sub>IN</sub> = 6.5 V, V <sub>ON/OFF</sub> = 0 V		-0.1	_	0.1	μA	4

\*1. V<sub>OUT(S)</sub>: Set output voltage

V<sub>OUT(E)1</sub>: Actual output voltage

Output voltage when fixing  $I_{\text{OUT}}(\text{= 30 mA})$  and inputting  $V_{\text{OUT}(\text{S})} + 1.0 \text{ V}$ 

V<sub>OUT(E)2</sub>: Actual output voltage

Output voltage when fixing  $I_{OUT}$  (= 80 mA) and inputting  $V_{OUT(S)}$  + 1.0 V

\*2. The output current at which the output voltage becomes 95% of VOUT(E)1 after gradually increasing the output current.

\*3. V<sub>drop</sub> = V<sub>IN1</sub> – (V<sub>OUT3</sub> × 0.98)

 $V_{\text{OUT3}}$  is the output voltage when  $V_{\text{IN}}$  =  $V_{\text{OUT}(S)}$  + 1.0 V and  $I_{\text{OUT}}$  = 100 mA.

 $V_{IN1}$  is the input voltage at which the output voltage becomes 98% of  $V_{OUT3}$  after gradually decreasing the input voltage. \*4. A change in the temperature of the output voltage [mV/°C] is calculated using the following equation.

$$\frac{\Delta V_{\text{OUT}}}{\Delta Ta} \left[ \text{mV/}^{\circ}\text{C} \right]^{*1} = V_{\text{OUT}(S)} \left[ V \right]^{*2} \times \frac{\Delta V_{\text{OUT}}}{\Delta Ta \bullet V_{\text{OUT}}} \left[ \text{ppm/}^{\circ}\text{C} \right]^{*3} \div 1000$$

\*1. Change in temperature of the output voltage

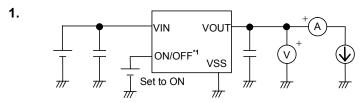
\*2. Set output voltage

\*3. Output voltage temperature coefficient

\*5. The output current can be at least this value.

Due to restrictions on the package power dissipation, this value may not be satisfied. Attention should be paid to the power dissipation of the package when the output current is large. This specification is guaranteed by design.

# Test Circuits





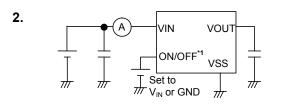


Figure 7

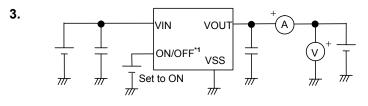
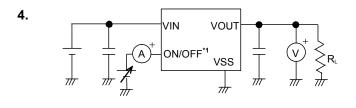


Figure 8





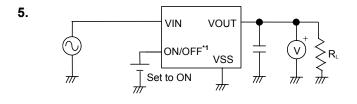
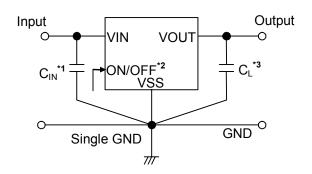


Figure 10

\*1. In case of product with ON/OFF circuit.

# Standard Circuit



- **\*1.**  $C_{IN}$  is a capacitor for stabilizing the input.
- \*2. In case of product with ON/OFF circuit.
- \*3. A tantalum capacitor (2.2  $\mu$ F or more) can be used.

#### Figure 11

#### Caution The above connection diagram and constant will not guarantee successful operation. Perform thorough evaluation using the actual application to set the constant.

# ■ Condition of Application

Input capacitor ( $C_{IN}$ ): 1.0  $\mu$ F or more Output capacitor ( $C_L$ ): 2.2  $\mu$ F or more (tantalum capacitor)

Caution Generally a series regulator may cause oscillation, depending on the selection of external parts. Check that no oscillation occurs with the application using the above capacitor.

# Explanation of Terms

#### 1. Low dropout voltage regulator

This voltage regulator has the low dropout voltage due to its built-in low on-resistance transistor.

#### 2. Output voltage (VOUT)

The accuracy of the output voltage is ensured at  $\pm 1.0\%$  under the specified conditions of fixed input voltage<sup>\*1</sup>, fixed output current, and fixed temperature.

**\*1.** Differs depending the product.

Caution If the above conditions change, the output voltage value may vary and exceed the accuracy range of the output voltage. Refer to "■ Electrical Characteristics" and "■ Characteristics (Typical Data)" for details.

3. Line regulation 
$$\left(\frac{\Delta V_{OUT1}}{\Delta V_{IN} \bullet V_{OUT}}\right)$$

Indicates the dependency of the output voltage on the input voltage. That is, the value shows how much the output voltage changes due to a change in the input voltage with the output current remaining unchanged.

#### 4. Load regulation ( $\Delta V_{OUT2}$ )

Indicates the dependency of the output voltage on the output current. That is, the value shows how much the output voltage changes due to a change in the output current with the input voltage remaining unchanged.

#### 5. Dropout voltage (V<sub>drop</sub>)

Indicates the difference between input voltage (V<sub>IN1</sub>) and the output voltage when; decreasing input voltage (V<sub>IN</sub>) gradually until the output voltage has dropped out to the value of 98% of output voltage (V<sub>OUT3</sub>), which is at  $V_{IN} = V_{OUT(S)} + 1.0 \text{ V}.$ 

 $V_{drop} = V_{IN1} - (V_{OUT3} \times 0.98)$ 

# 6. Output voltage temperature coefficient $\left(\frac{\Delta V_{OUT}}{\Delta Ta \bullet V_{OUT}}\right)$

The shaded area in **Figure 12** is the range where  $V_{OUT}$  varies in the operation temperature range when the output voltage temperature coefficient is  $\pm 100 \text{ ppm/}^{\circ}\text{C}$ .

# V<sub>OUT</sub>[V] V<sub>OUT(E)1</sub>\*1 -40 +25 +85 Ta [°C]

#### Example of S-1131B28 typ. product

\*1.  $V_{OUT(E)1}$  is the value of the output voltage measured at Ta = +25°C.

#### Figure 12

A change in the temperature of the output voltage [mV/°C] is calculated using the following equation.

$$\frac{\Delta V_{OUT}}{\Delta Ta} \left[ mV/^{\circ}C \right]^{*1} = V_{OUT(S)} \left[ V \right]^{*2} \times \frac{\Delta V_{OUT}}{\Delta Ta \bullet V_{OUT}} \left[ ppm/^{\circ}C \right]^{*3} \div 1000$$

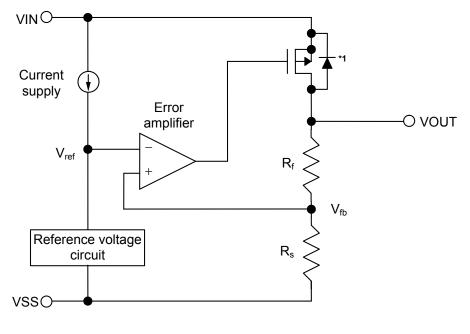
- \*1. Change in temperature of output voltage
- \*2. Set output voltage
- \*3. Output voltage temperature coefficient

# Operation

#### 1. Basic operation

Figure 13 shows the block diagram of the S-1131 Series.

The error amplifier compares the reference voltage ( $V_{ref}$ ) with feedback voltage ( $V_{fb}$ ), which is the output voltage resistance-divided by feedback resistors ( $R_s$  and  $R_f$ ). It supplies the gate voltage necessary to maintain the constant output voltage which is not influenced by the input voltage and temperature change, to the output transistor.



\*1. Parasitic diode

Figure 13

#### 2. Output transistor

In the S-1131 Series, a low on-resistance P-channel MOS FET is used as the output transistor. Be sure that  $V_{OUT}$  does not exceed  $V_{IN} + 0.3$  V to prevent the voltage regulator from being damaged due to reverse current flowing from the VOUT pin through a parasitic diode to the VIN pin, when the potential of  $V_{OUT}$  became higher than  $V_{IN}$ .

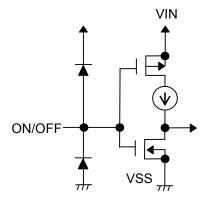
#### 3. ON/OFF pin

In case of product with ON/OFF circuit, this pin starts and stops the regulator.

When the ON/OFF pin is set to OFF level, the entire internal circuit stops operating, and the built-in P-channel MOS FET output transistor between the VIN pin and the VOUT pin is turned off, reducing current consumption significantly. The VOUT pin becomes the Vss level due to the internally divided resistance of several hundreds k $\Omega$  between the VOUT pin and the VSS pin.

The structure of the ON/OFF pin is as shown in **Figure 14**. Since the ON/OFF pin is neither pulled down nor pulled up internally, do not use it in the floating states. In addition, note that the current consumption increases if a voltage of 0.3 V to  $V_{IN} - 0.3$  V is applied to the ON/OFF pin. When not using the ON/OFF pin, connect it to the VSS pin in the product A type, and connect it to the VIN pin in B type.

Table 7				
Product Type	ON/OFF Pin	Internal Circuit	VOUT Pin Voltage	Current Consumption
А	"L": ON	Operate	Set value	I <sub>SS1</sub>
А	"H": OFF	Stop	V <sub>SS</sub> level	I <sub>SS2</sub>
В	"L": OFF	Stop	V <sub>SS</sub> level	I <sub>SS2</sub>
В	"H": ON	Operate	Set value	I <sub>SS1</sub>





# ■ Selection of Output Capacitor (C<sub>L</sub>)

The S-1131 Series performs phase compensation using the internal phase compensator in the IC and the ESR (Equivalent Series Resistance) of the output capacitor to enable stable operation independent of changes in the output load. Therefore, always place a capacitor ( $C_L$ ) of 2.2  $\mu$ F or more between the VOUT pin and the VSS pin.

For stable operation of the S-1131 Series, it is essential to employ a capacitor whose ESR is within an optimum range. Using a capacitor whose ESR is outside the optimum range (approximately  $0.5 \Omega$  to  $5 \Omega$ ), whether larger or smaller, may cause an unstable output, resulting in oscillation. For this reason, a tantalum electrolytic capacitor is recommended.

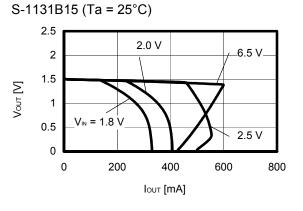
When a ceramic capacitor or an OS capacitor with a low ESR is used, it is necessary to connect an additional resistor that serves as the ESR in series with the output capacitor. The required resistance value is approximately 0.5  $\Omega$  to 5  $\Omega$ , which varies depending on the usage conditions, so perform sufficient evaluation for selection. Ordinarily, around 1.0  $\Omega$  is recommended.

Note that an aluminum electrolytic capacitor may increase the ESR at a low temperature, causing oscillation. When using this kind of capacitor, perform thorough evaluation, including evaluation of temperature characteristics.

### Precautions

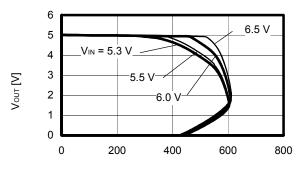
- Wiring patterns for the VIN pin, the VOUT pin and GND should be designed so that the impedance is low. When mounting an output capacitor between the VOUT pin and the VSS pin (C<sub>L</sub>) and a capacitor for stabilizing the input between the VIN pin and the VSS pin (C<sub>IN</sub>), the distance from the capacitors to these pins should be as short as possible.
- Note that generally the output voltage may increase when a series regulator is used at low load current (1.0 mA or less).
- The S-1131 Series performs phase compensation by using an internal phase compensator and the ESR of an output capacitor. Therefore, always place a capacitor of 2.2 μF or more between the VOUT pin and the VSS pin. A tantalum type capacitor is recommended. Moreover, to secure stable operation of the S-1131 Series, it is necessary to employ a capacitor with an ESR within an optimum range (0.5 Ω to 5 Ω). Using a capacitor whose ESR is outside the optimum range (approximately 0.5 Ω to 5 Ω), whether larger or smaller, may cause an unstable output, resulting in oscillation. Perform sufficient evaluation under the actual usage conditions for selection, including evaluation of temperature characteristics.
- The voltage regulator may oscillate when the impedance of the power supply is high and the input capacitance is small or an input capacitor is not connected.
- Overshoot may occur in the output voltage momentarily if the voltage is rapidly raised at power-on or when the power supply fluctuates. Sufficiently evaluate the output voltage at power-on with the actual device.
- The application conditions for the input voltage, the output voltage, and the load current should not exceed the package power dissipation.
- Do not apply an electrostatic discharge to this IC that exceeds the performance ratings of the built-in electrostatic protection circuit.
- In determining the output current, attention should be paid to the output current value specified in Table 6 in "■ Electrical Characteristics" and footnote \*5 of the table.
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# Characteristics (Typical Data)

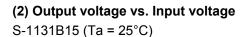


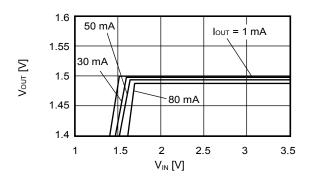
(1) Output voltage vs. Output current (when load current increases)

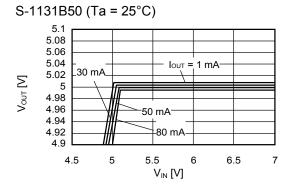
S-1131B50 (Ta = 25°C)

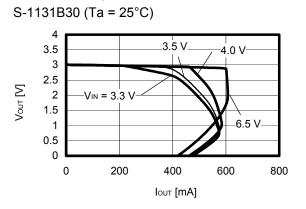


IOUT [mA]





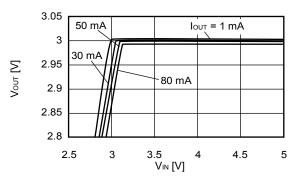




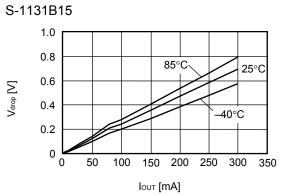
**Remark** In determining the output current, attention should be paid to the following.

- The minimum output current value and footnote \*5 of Table 6 in the "■ Electrical Characteristics"
- 2) The package power dissipation

S-1131B30 (Ta = 25°C)

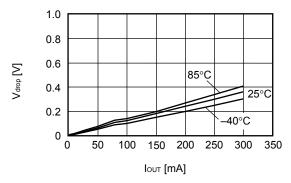


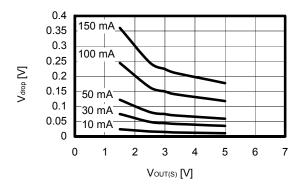
(3) Dropout voltage vs. Output current



S-1131B30 1.0 0.8 0.6 -85°C V<sub>drop</sub> [V] 25°C 0.4 0.2 40°C 0 0 100 200 250 300 50 150 350 IOUT [mA]

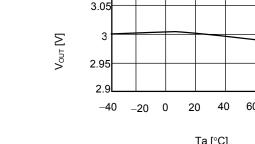
S-1131B50



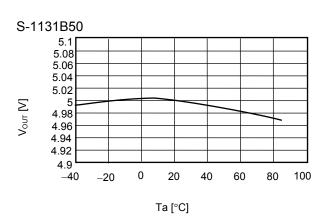


(4) Dropout voltage vs. Set output voltage

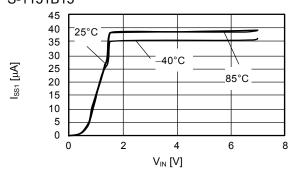
S-1131B15 1.6 1.55 Vour [V] 1.5 1.45 1.4 <u>-40</u> -20 0 60 80 20 40 100 Ta [°C]

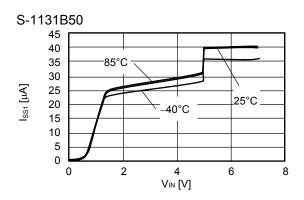


(5) Output voltage vs. Ambient temperature

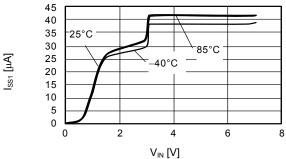


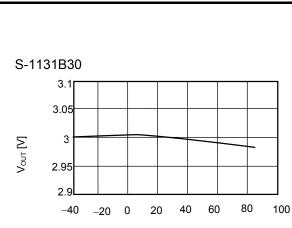
(6) Current consumption vs. Input voltage S-1131B15





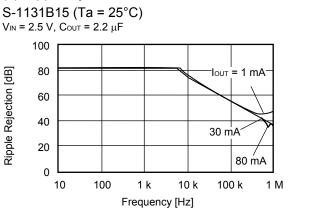






Ta [°C]

#### (7) Ripple rejection



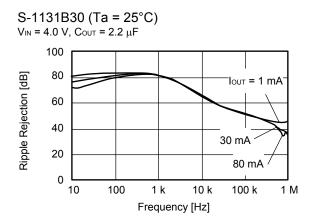
30 mA

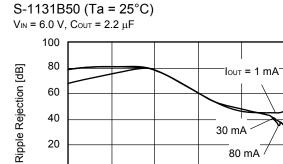
10 k

80 mA

100 k

1 M





100

1 k

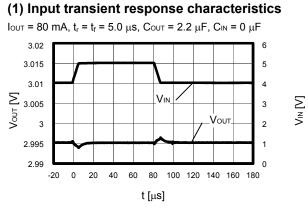
Frequency [Hz]

40

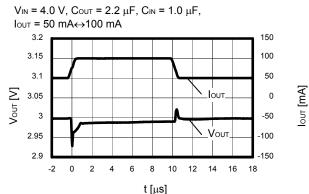
20

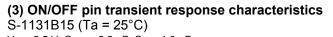
0 10

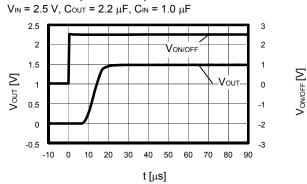
### Reference Data



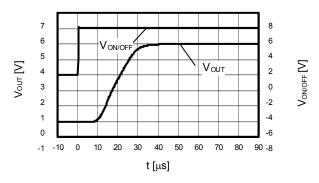
#### (2) Load transient response characteristics

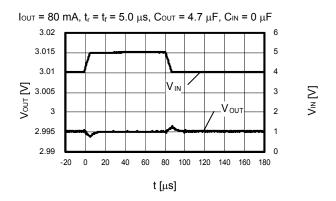




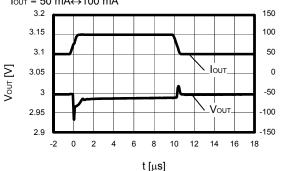


S-1131B50 (Ta = 25°C) VIN = 6.0 V, COUT = 2.2  $\mu$ F, CIN = 1.0  $\mu$ F



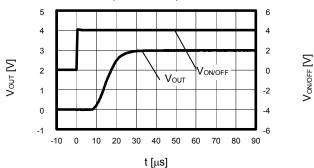


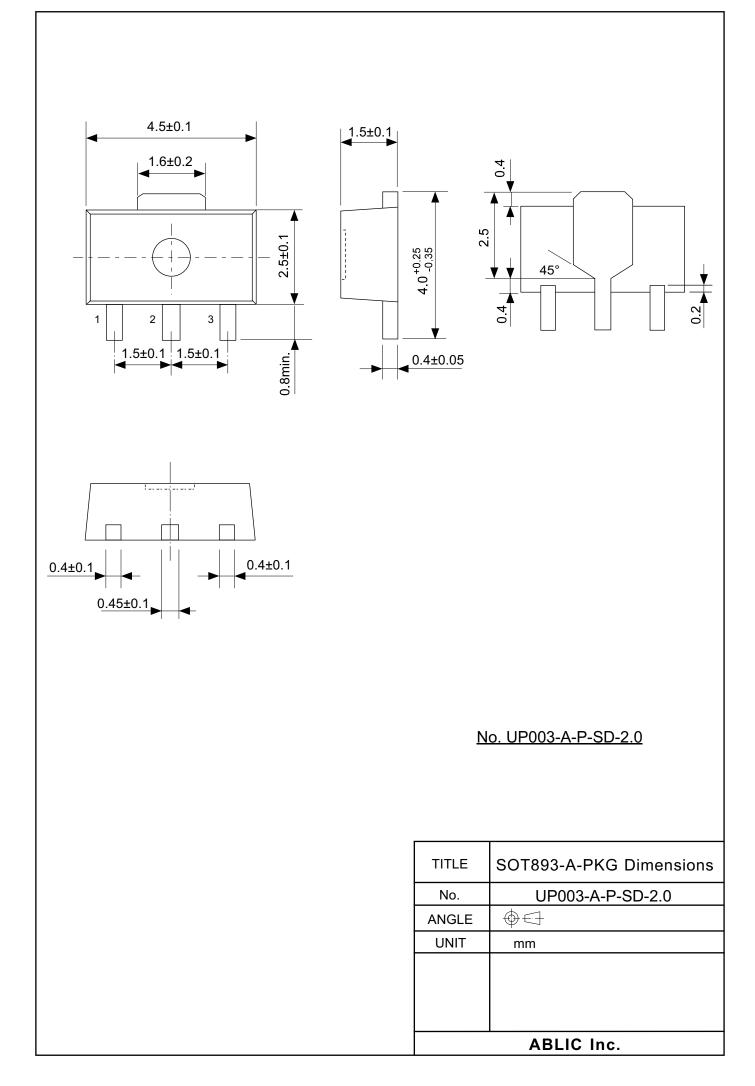
 $\label{eq:Vin} \begin{array}{l} \mathsf{V}_{\mathsf{IN}} = 4.0 \ \mathsf{V}, \ \mathsf{C}_{\mathsf{OUT}} = 4.7 \ \mu\mathsf{F}, \ \mathsf{C}_{\mathsf{IN}} = 1.0 \ \mu\mathsf{F}, \\ \mathsf{I}_{\mathsf{OUT}} = 50 \ \mathsf{mA} {\leftrightarrow} 100 \ \mathsf{mA} \end{array}$ 

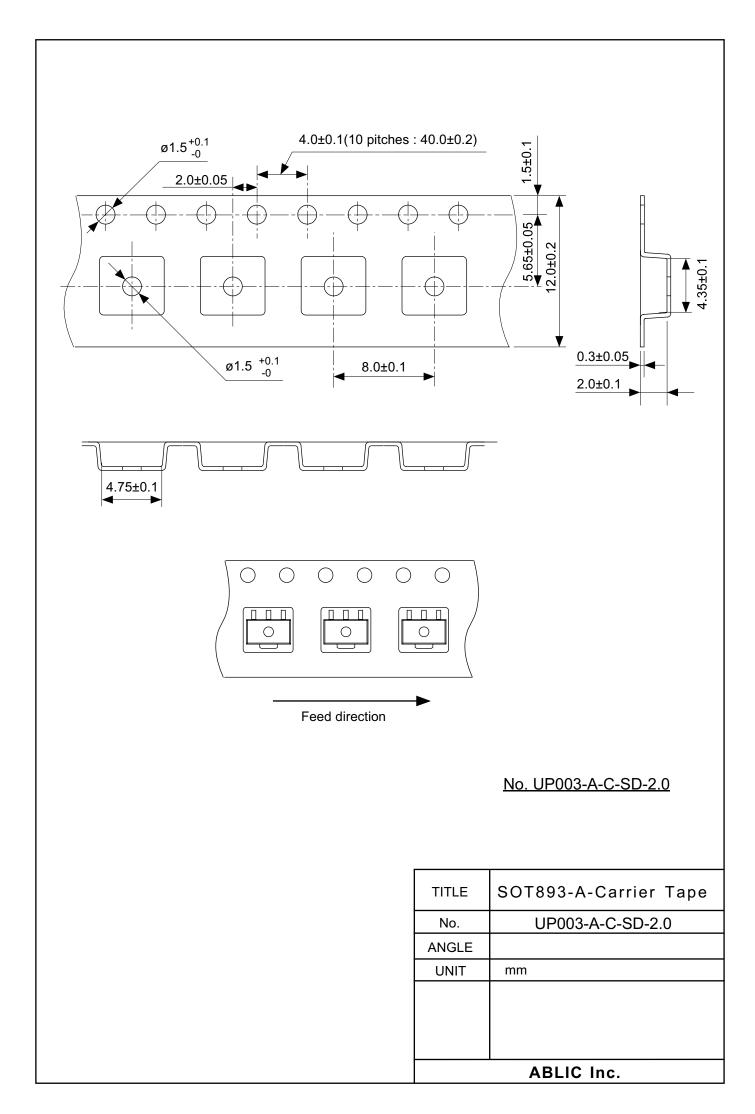


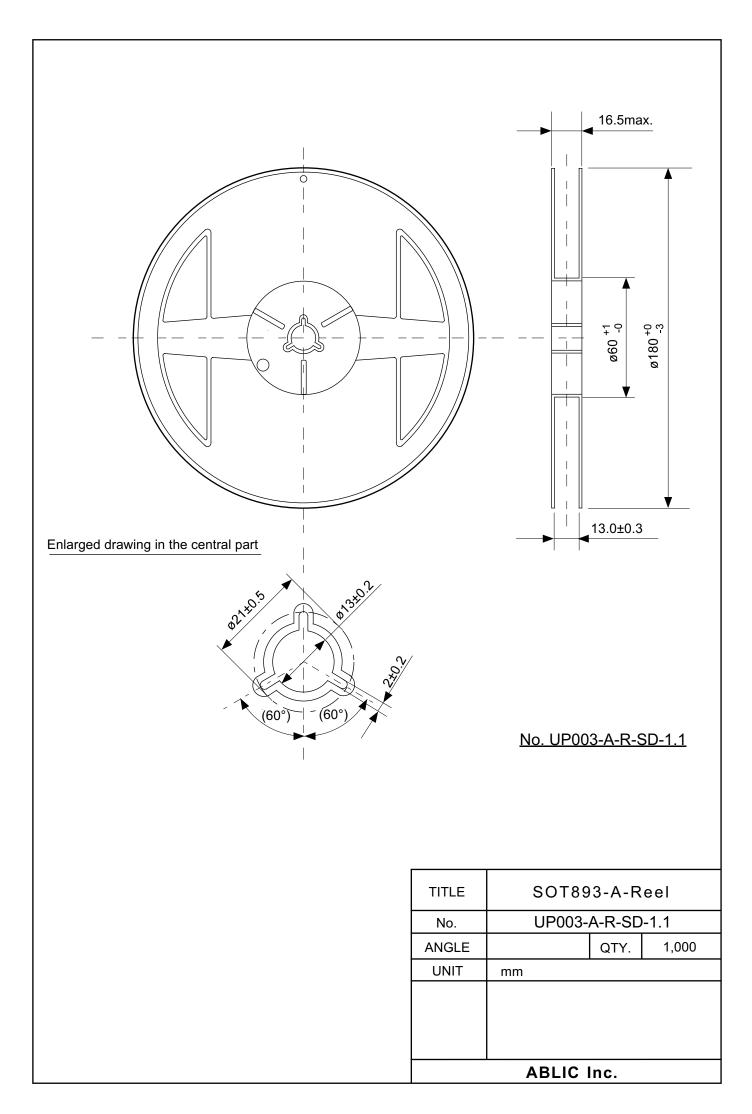
lour [mA]

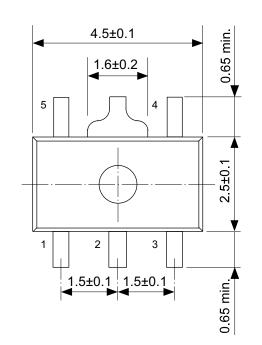
S-1131B30 (Ta = 25°C) V<sub>IN</sub> = 4.0 V, C<sub>OUT</sub> = 2.2 μF, C<sub>IN</sub> = 1.0 μF

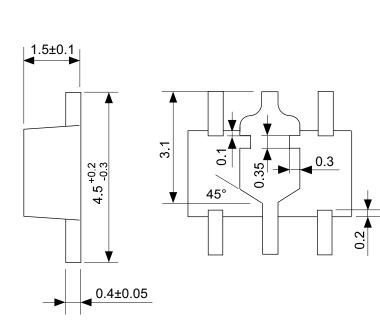


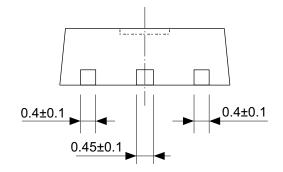






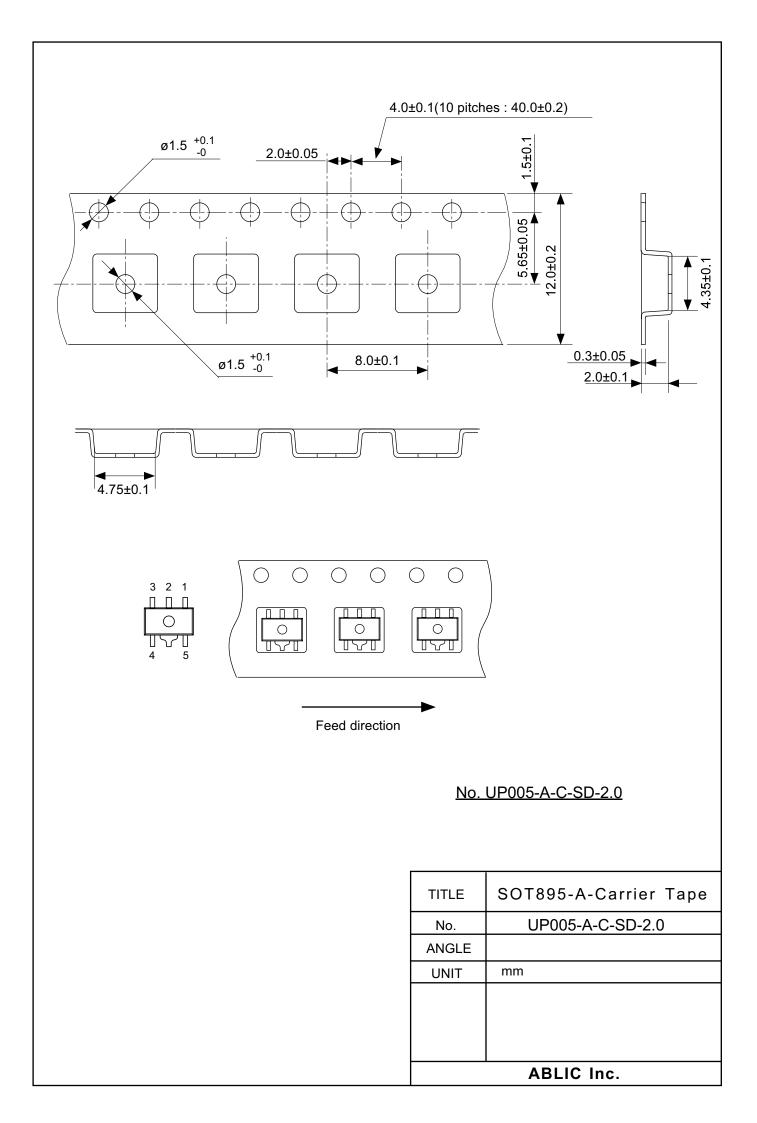


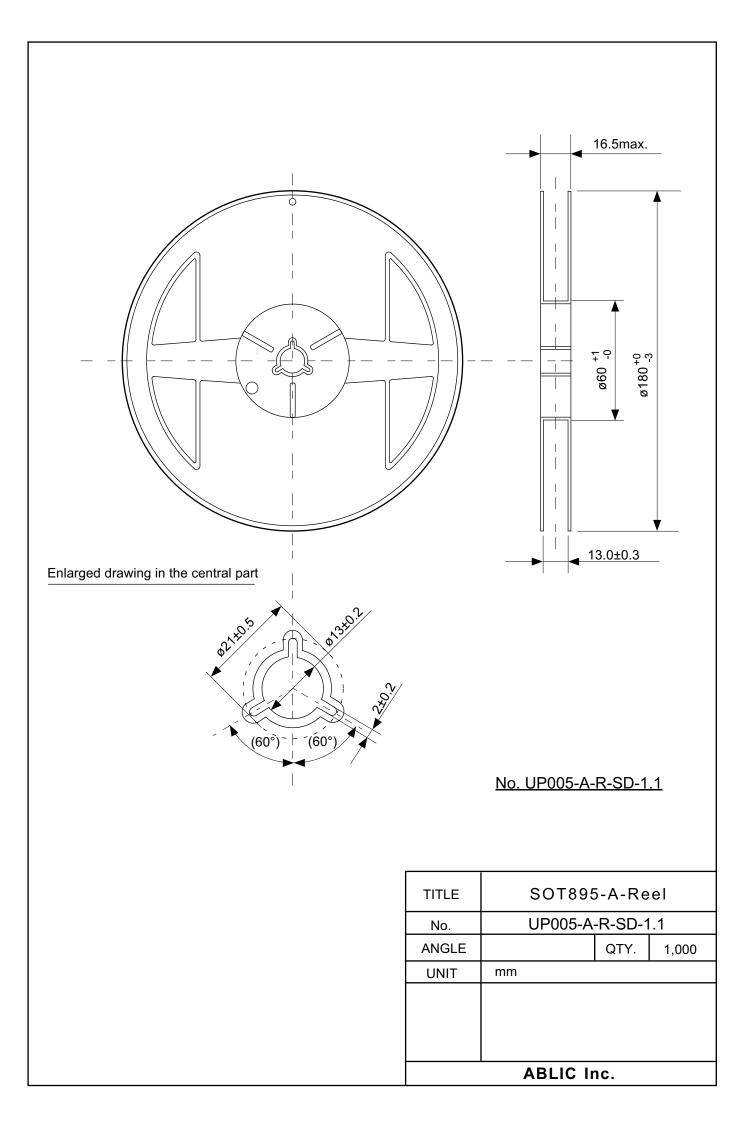


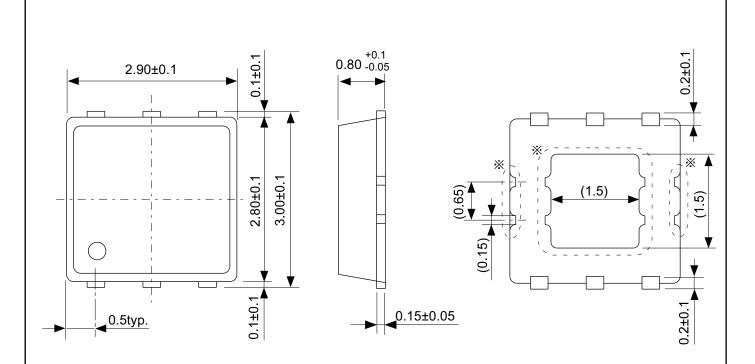


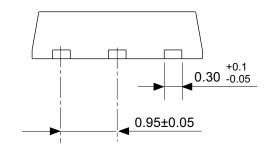
No. UP005-A-P-SD-2.0

TITLE	SOT895-A-PKG Dimensions		
No.	UP005-A-P-SD-2.0		
ANGLE	$\oplus \in \mathbb{R}$		
UNIT	mm		
	ABLIC Inc.		





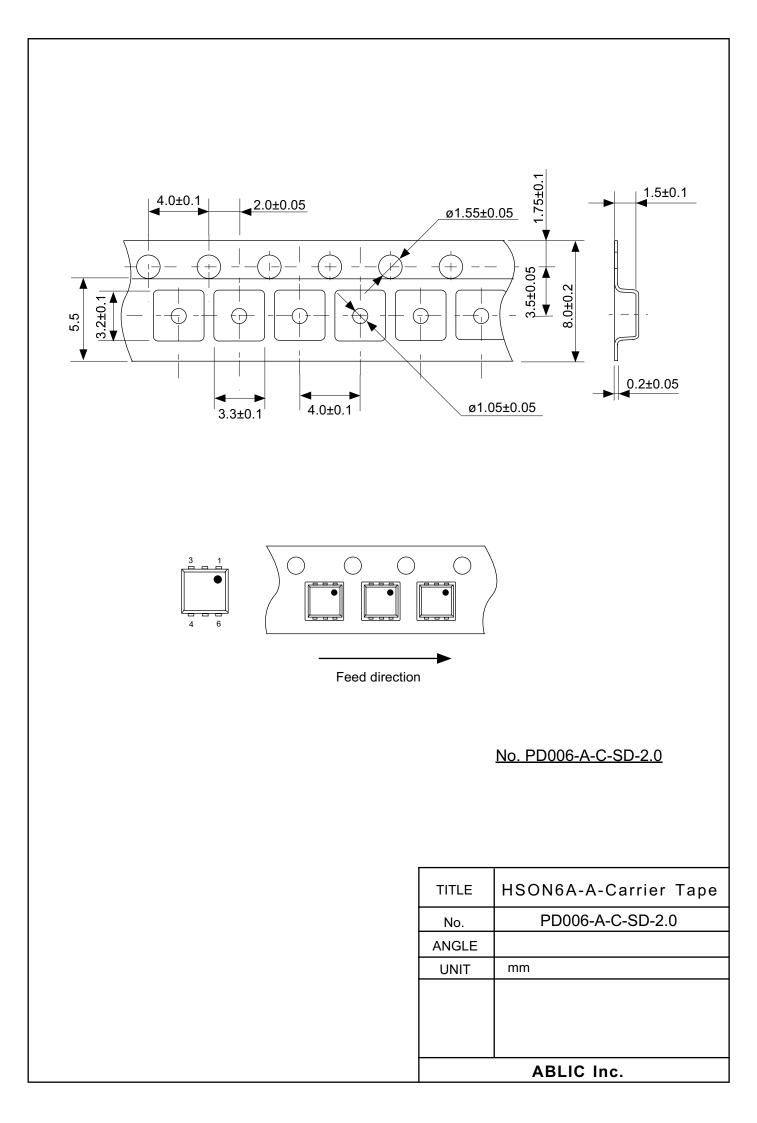


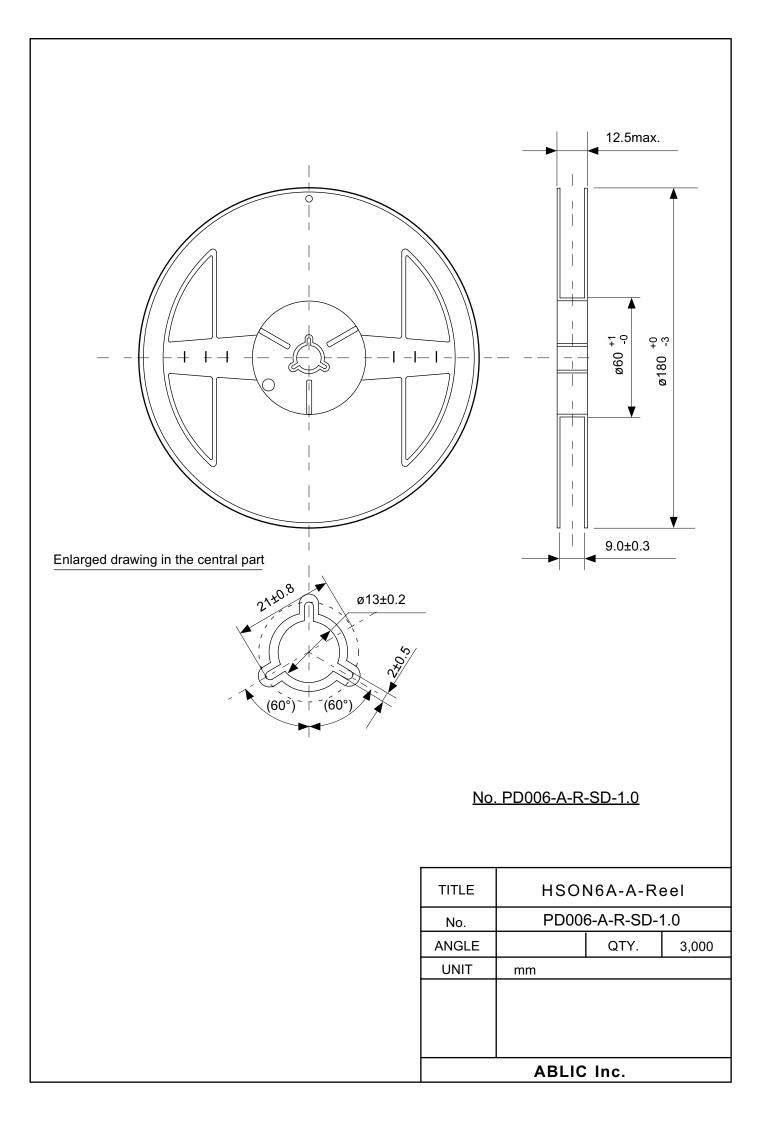


 The exposed thermal die pad has different electric potential depending on the product. Confirm specifications of each product. Do not use it as the function of electrode.

No. PD006-A-P-SD-5.0

TITLE	HSON6A-A-PKG Dimensions		
No.	PD006-A-P-SD-5.0		
ANGLE	$\oplus$		
UNIT	mm		
ABLIC Inc.			





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