











LP358, LP2904

SLOS475A - AUGUST 2005-REVISED MAY 2017

Ultra-Low Power Dual Operational Amplifiers

Features

Low Supply Current: 54 µA (Typical) Low Offset Voltage: 2 mV (Typical) Low Input Bias Current: 2 nA (Typical)

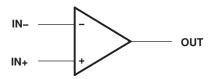
Input Common-Mode to GND

Wide Supply Voltage: 3 V < V_{CC} < 32 V Pin Compatible With LM358 and LM2904

Applications

- LCD Displays
- Portable Instrumentation
- Sensor and Metering Equipment
- Consumer Electronics (MP3 Players, Toys)
- **Power Supplies**

Symbol (Each Amplifier)



3 Description

The LP358 and LP2904 devices are dual low-power operational amplifiers especially suited for batteryoperated applications. Good input specifications and wide supply-voltage range still are achieved despite the ultra-low supply current. Single-supply operation is achieved with an input common-mode range that includes GND.

The LP358 and LP2904 devices are ideal in applications where wide supply voltages and low power are more important than speed and bandwidth. These applications include portable instrumentation, LCD displays, consumer electronics (MP3 players, toys, and so forth), and power supplies.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)	
LP2904	COIC (9)	4.00 2.01	
LP358	SOIC (8)	4.90 × 3.91	

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Schematic (Each Amplifier)

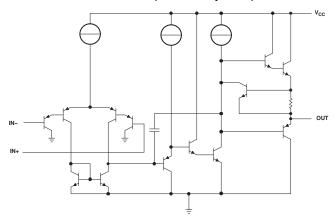




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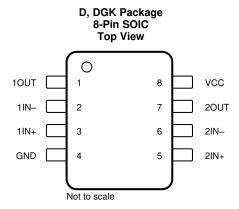
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4 Revision History

CI	nanges from Original (August 2005) to Hevision A	Page
•	Added table of contents and Revision History section	1
•	Deleted Ordering Information table	
•	Changed low supply current in Features list from 85 μA to 54 μA	1
•	Added Device Information table and table note	1
•	pinout image and pinout information in Pin Configuration and Functions section	3
•	Deleted θ_{JA} values and table notes from <i>Absolute Maximum Ratings</i> table and added information to <i>Thermal Information</i> table	4
•	Added Recommended Operating Conditions table	4
•	Added Thermal Information table	4
•	Reformatted Electrical Characteristics table	5
•	Changed typical supply current value ($T_A = 25$ °C) from 85 μ A to 54 μ A in <i>Electrical Characteristics: LP358</i> table	5
•	Changed maximum supply current value (T _A = 25°C) from 150 μA to 75 μA in <i>Electrical Characteristics: LP358</i> table	ə 5
•	Added table note to I _{OS, VCC} parameter in <i>Electrical Characteristics: LP358</i> table	5
•	Changed typical supply current value ($T_A = 25$ °C) from 85 μ A to 54 μ A in <i>Electrical Characteristics: LP2904</i> table	6
•	Changed maximum supply current value (T _A = 25°C) from 150 μA to 75 μA in <i>Electrical Characteristics: LP2904</i> tab	ole 6
•	Added table note to I _{OS, VCC} parameter in <i>Electrical Characteristics: LP2904</i> table	6
•	Added Typical Characteristics graphs	7
•	Added Detailed Description section	8
•	Added Application and Implementation section	9
•	Deleted "of the same magnitude" text from Typical Application section	9
•	Added Power Supply Recommendations section	11
•	Added Layout section	12
•	Added Device and Documentation Support and Mechanical, Packaging, and Orderable Information sections	13
•	Added Related Links table	13



5 Pin Configuration and Functions



Pin Functions

PIN		1/0	DESCRIPTION	
NAME	NO.	I/O	DESCRIPTION	
GND	4	_	Ground or negative power supply	
1IN+	3	I	Channel 1 noninverting input	
1IN-	2	I	Channel 1 inverting input	
1OUT	1	0	Channel 1 output	
2IN+	5	1	Channel 2 noninverting input	
2IN-	6	1	Channel 2 inverting input	
2OUT	7	0	Channel 2 output	
V _{CC}	8	_	Positive power supply	



6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted) (1)

		MIN	MAX	UNIT
V _{CC}	Supply voltage range (2)		±16 or 32	V
V_{ID}	Differential input voltage (3)		±32	V
V_{I}	Input voltage (either input)	-0.3	32	V
	Duration of output short circuit (one amplifier) to ground at (or below) $T_A = 25^{\circ}C$, $V_{CC} \le 15$ $V^{(4)}$		Unlimited	
Operating virtual temperature, T _J		150	°C	
Storage temperature, T _{stg} –65 150		°C		

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- (2) All voltage values (except differential voltages and V_{CC} specified for the measurement of I_{OS}) are with respect to the network GND.
- (3) Differential voltages are at IN+, with respect to IN-.
- 4) Short circuits from outputs to V_{CC} can cause excessive heating and eventual destruction.

6.2 ESD Ratings

			VALUE	UNIT
		Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±2000	
V _(ESD)	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	±1000	V

⁽¹⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process. Manufacturing with less than 500-V HBM is possible with the necessary precautions. Pins listed as ±XXX V may actually have higher performance.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	MAX	UNIT
V _{CC}	Supply voltage		3	32	V
V _{CM}	Common-mode voltage	0	V _{CC} – 1.5 V	V	
T _A		LP358	0	70	00
	Operating free-air temperature	LP2904	-40	85	°C

6.4 Thermal Information

		LP358	LP2904	
THERMAL METRIC ⁽¹⁾		D, DGK (SOIC)	D, DGK (SOIC)	UNIT
		8 PINS	8 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance (2) (3)	118.8	118.8	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	71.7	71.7	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	68.6	68.6	°C/W
ΨЈТ	Junction-to-top characterization parameter	23.3	23.3	°C/W
ΨЈВ	Junction-to-board characterization parameter	67.7	67.7	°C/W

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

(3) The package thermal impedance is calculated in accordance with JESD 51-7.

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⁽²⁾ JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process. Manufacturing with less than 250-V CDM is possible with the necessary precautions. Pins listed as ±YYY V may actually have higher performance.

⁽²⁾ Maximum power dissipation is a function of $T_{J(max)}$, θ_{JA} , and T_A . The maximum allowable power dissipation at any allowable ambient temperature is $P_D = (T_{J(max)} - T_A) / \theta_{JA}$. Operating at the absolute maximum T_J of 150°C can affect reliability.



6.5 Electrical Characteristics: LP358

 $T_A = 25$ °C, $V_{CC} = 5$ V, $V_{IC} = V_{CC} / 2$, $R_L = 100$ k Ω to GND (unless otherwise noted)

PARAMETER	TEST CONDIT	ΓΙΟΝS ^{(1) (2)}	MIN	TYP (3)	MAX	UNIT	
land affect of	T _A = 25°C			2	4	- 17	
Input offset voltage	<u> </u>				9	mV	
	T _A = 25°C			2	10		
Input bias current	T _A = Full range				20	nA	
locate offers to accompany	T _A = 25°C			0.2	2	nΔ	
input offset current	T _A = Full range				4	nA	
Large signal valtage gain	$R_1 = 10 \text{ k}\Omega \text{ to GND},$	T _A = 25°C	50	100		V/mV	
Large-signal voltage gain	$V_{CC} = 30 \text{ V}$	T _A = Full range	40			V/mv	
Common mode rejection ratio	V _{CC} = 30 V,	T _A = 25°C	80	90		٩D	
Common-mode rejection ratio	$V_{IC} = 0 \text{ V to } V_{CC} - 1.5 \text{ V}$	T _A = Full range	75			dB	
Davies aventu salastias satia	V 5 V += 00 V	T _A = 25°C	80	90			
Power-supply rejection ratio	$V_{CC} = 5 V 10 30 V$	T _A = Full range	75			V	
Complex accompant	D	T _A = 25°C		54	75	uА	
Supply current	H _L = ∞	T _A = Full range			125		
Output valtage awing (high)	$I_L = 0.35 \text{ mA to GND}$	T _A = 25°C	3.4	3.6		V	
Output voltage swing (nigh)	$V_{IC} = 0 V$	T _A = Full range	V _{CC} – 1.9				
Output valtage awing (law)	$I_L = 0.35 \text{ mA from } V_{CC}$	T _A = 25°C	0.82	0.7		٧	
Output voltage swing (low)	$V_{IC} = 0 V$	T _A = Full range	1			V	
Output course current	V 2VV 1V	T _A = 25°C	7	10		m Λ	
Output source current	$V_O = 3 V$, $V_{ID} = 1 V$	T _A = Full range	4			mA	
	V 15VV 1V	T _A = 25°C	4	5			
Output aink aurrant	$v_0 = 1.5 \text{ V}, v_{1D} = -1 \text{ V}$	T _A = Full range	3			A	
Output Sink current	$V_{O} = 1.5 \text{ V}, V_{ID} = -1 \text{ V},$	$T_A = 25^{\circ}C$	2	4		mA	
	$V_{IC} = 0 V$	T _A = Full range	1				
Output about to CND	V 1 V	$T_A = 25^{\circ}C$		20	35	A	
Output short to GND	V _{ID} = I V	T _A = Full range			40	mA	
Output short to V (4)	V 1.V	T _A = 25°C		15	30		
Outhor short to ACC (v _{ID} = -1 v	T _A = Full range			45	mA	
Input offset voltage drift		T _A = 25°C		10		μV/°C	
Input offset current drift		T _A = 25°C		10		pA/°C	
	Input offset voltage Input bias current Input offset current Large-signal voltage gain Common-mode rejection ratio Power-supply rejection ratio Supply current Output voltage swing (high) Output voltage swing (low) Output source current Output sink current Output short to GND Output short to V _{CC} (4) Input offset voltage drift	$ \begin{array}{llllllllllllllllllllllllllllllllllll$	$ \begin{array}{l} \text{Input offset voltage} & \begin{array}{l} T_A = 25^{\circ}\text{C} \\ T_A = \text{Full range} \end{array} \\ \text{Input bias current} & \begin{array}{l} T_A = 25^{\circ}\text{C} \\ T_A = \text{Full range} \end{array} \\ \text{Input offset current} & \begin{array}{l} T_A = 25^{\circ}\text{C} \\ T_A = \text{Full range} \end{array} \\ \text{Input offset current} & \begin{array}{l} T_A = 25^{\circ}\text{C} \\ T_A = \text{Full range} \end{array} \\ \text{Large-signal voltage gain} & \begin{array}{l} R_L = 10 \text{ k}\Omega \text{ to GND,} \\ V_{CC} = 30 \text{ V} & \begin{array}{l} T_A = 25^{\circ}\text{C} \\ T_A = \text{Full range} \end{array} \\ \text{Common-mode rejection ratio} & \begin{array}{l} V_{CC} = 30 \text{ V,} \\ V_{IC} = 0 \text{ V to V}_{CC} - 1.5 \text{ V} \end{array} \\ \text{Power-supply rejection ratio} & \begin{array}{l} V_{CC} = 5 \text{ V to } 30 \text{ V} \\ \end{array} \\ \text{R}_L = \infty & \begin{array}{l} T_A = 25^{\circ}\text{C} \\ T_A = \text{Full range} \end{array} \\ \text{Supply current} & \begin{array}{l} R_L = \infty \\ \end{array} \\ \text{R}_L = \infty & \begin{array}{l} T_A = 25^{\circ}\text{C} \\ \end{array} \\ T_A = \text{Full range} \end{array} \\ \text{Output voltage swing (high)} & \begin{array}{l} I_L = 0.35 \text{ mA to GND} \\ V_{IC} = 0 \text{ V} \end{array} \\ \text{Output voltage swing (low)} & \begin{array}{l} I_L = 0.35 \text{ mA from V}_{CC} \\ \end{array} \\ \text{V}_C = 0 \text{ V} & \begin{array}{l} T_A = 25^{\circ}\text{C} \\ \end{array} \\ \text{T}_A = \text{Full range} \end{array} \\ \text{Output source current} & \begin{array}{l} V_O = 3 \text{ V, V}_{ID} = 1 \text{ V} \end{array} \\ \text{Output sink current} & \begin{array}{l} T_A = 25^{\circ}\text{C} \\ \end{array} \\ \text{T}_A = \text{Full range} \end{array} \\ \text{Output short to GND} & \begin{array}{l} V_{ID} = 1 \text{ V} \end{array} \\ \text{Output short to GND} & \begin{array}{l} T_A = 25^{\circ}\text{C} \\ \end{array} \\ \text{T}_A = \text{Full range} \end{array} \\ \text{Output short to V}_{CC} & \begin{array}{l} (4) \\ V_{ID} = 1 \text{ V} \end{array} \\ \text{Output short to V}_{CC} & \begin{array}{l} T_A = 25^{\circ}\text{C} \\ \end{array} \\ \text{T}_A = \text{Full range} \end{array} \\ \text{T}_A = 25^{\circ}\text{C} \end{array} \\ \text{T}_A = \text{Full range} \end{array} \\ \text{T}_A = 25^{\circ}\text{C} \end{array} \\ \text{T}_A = \text{Full range} \end{array} \\ \text{T}_A = 25^{\circ}\text{C} \end{array} \\ \text{T}_A = \text{Full range} \end{array} \\ \text{T}_A = 25^{\circ}\text{C} \end{array} \\ \text{T}_A = \text{Full range} \end{array} \\ \text{T}_A = 25^{\circ}\text{C} \end{array} \\ \text{T}_A = \text{Full range} \end{array} \\ \text{T}_A = 25^{\circ}\text{C} \end{array} \\ \text{T}_A = \text{Full range} \end{array} \\ \text{T}_A = 25^{\circ}\text{C} \end{array} \\ \text{T}_A = \text{Full range} \end{array} \\ \text{T}_A = 25^{\circ}\text{C} \end{array} \\ \text{T}_A = \text{Full range} \end{array} $	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	

⁽¹⁾ For full-range temperature limits: $V_{CC} = 3$ V to 32 V, $V_{ICR} = 0$ V to $V_{CC} - 1.5$ V (unless otherwise noted) (2) Full range is 0°C to 70°C for LP358. (3) All typical values are at $T_A = 25$ °C. (4) Short circuits from outputs to V_{CC} can cause excessive heating and eventual destruction.



6.6 Electrical Characteristics: LP2904

 $T_A = 25$ °C, $V_{CC} = 5$ V, $V_{IC} = V_{CC}$ / 2, $R_L = 100$ k Ω to GND (unless otherwise noted)

	PARAMETER	TEST CONDI	TIONS (1) (2)	MIN	TYP (3)	MAX	UNIT	
\/	Input offeet veltage	T _A = 25°C	T _A = 25°C		2	4	mV	
V_{IO}	Input offset voltage	T _A = Full range				10	IIIV	
	land the annual to	T _A = 25°C			2	20	^	
I _{IB}	Input bias current	T _A = Full range				40	nA	
	land to effect account	T _A = 25°C			0.5	4	^	
I _{IO}	Input offset current	T _A = Full range				8	nA	
^	Large eignel veltege gein	$R_L = 10 \text{ k}\Omega \text{ to GND},$	T _A = 25°C	40	70		V/mV	
A_V	Large-signal voltage gain	$V_{CC} = 30 \text{ V}$	T _A = Full range	30			V/IIIV	
CMRR	Common mode rejection ratio	V _{CC} = 30 V,	T _A = 25°C	80	90		٩D	
CWIRK	Common-mode rejection ratio	$V_{IC} = 0 \text{ V to } V_{CC} - 1.5 \text{ V}$	T _A = Full range	75			dB	
l.	Dower cumply rejection ratio	V 5 V to 20 V	T _A = 25°C	80	90		V	
k _{VSR}	Power-supply rejection ratio	$V_{CC} = 5 \text{ V to } 30 \text{ V}$	T _A = Full range	75			\ \ \ \	
	Complex accompant	D	T _A = 25°C		54	75	пΑ	
I _{CC}	Supply current	R _L = ∞	T _A = Full range			138		
V	Output voltage ewing (high)	$I_L = 0.35 \text{ mA to GND},$	$T_A = 25^{\circ}C$	3.4	3.6		_ v	
V _{OH}	Output voltage swing (high)	$V_{IC} = 0 V$ $T_A = Full range$	V _{CC} – 1.9			V		
V	Output voltage ewing (low)	$I_L = 0.35 \text{ mA from } V_{CC}$	$T_A = 25^{\circ}C$	0.82	0.7		V	
V_{OL}	Output voltage swing (low)	$V_{IC} = 0 V$	T _A = Full range	1			V	
	Output course current	V 2VV 1V	T _A = 25°C	7	10		mA	
lo	Output source current	$V_{O} = 3 V, V_{ID} = 1 V$	T _A = Full range	4			mA	
		V _O = 1.5 V, V _{ID} = -1 V	$T_A = 25^{\circ}C$	4	5			
	Output aink aurrant	$\mathbf{v}_{O} = 1.5 \mathbf{v}, \mathbf{v}_{ID} = -1 \mathbf{v}$	T _A = Full range	3			mA	
I _O	Output sink current	$V_O = 1.5 \text{ V}, V_{ID} = -1 \text{ V},$	T _A = 25°C	2	4		IIIA	
		$V_{IC} = 0 V$	T _A = Full range	1				
	Output about to CND	\/ 1\/	T _A = 25°C		20	35	5	
I _{OS,GND}	Output short to GND	$V_{ID} = 1 V$	T _A = Full range			40	mA	
1	Output short to V _{CC} ⁽⁴⁾	V _{ID} = -1 V	$T_A = 25^{\circ}C$		15	30	mA	
I _{OS,VCC}	Outhor short to ACC / /	v _{ID} = -1 v	T _A = Full range			45	IIIA	
αV_{IO}	Input offset voltage drift		T _A = 25°C		10		μV/°C	
αl _{IO}	Input offset current drift		T _A = 25°C		10		pA/°C	

⁽¹⁾ For full-range temperature limits: V_{CC} = 3 V to 32 V, V_{ICR} = 0 V to V_{CC} – 1.5 V (unless otherwise noted) (2) Full range is –40°C to +85°C for LP2904.

6.7 Operating Characteristics

over operating free-air temperature range (unless otherwise noted)

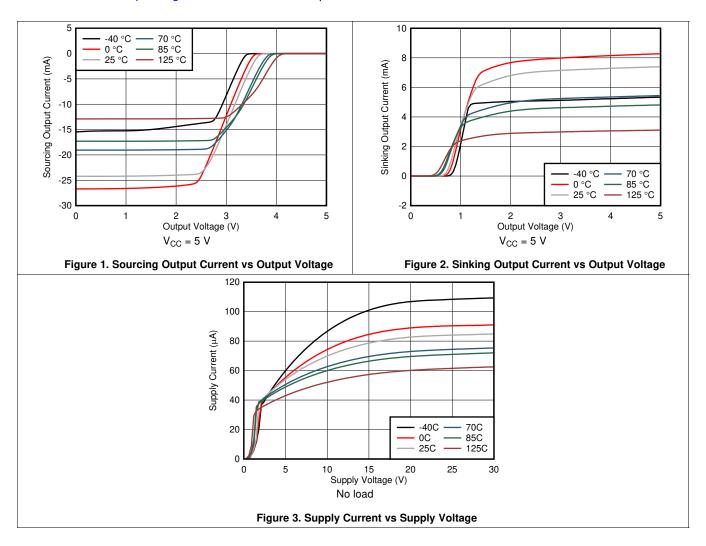
		MIN	NOM	MAX	UNIT
GBW	Gain bandwidth product		100		kHz
SR	Slew rate		50		V/ms

All typical values are at $T_A = 25^{\circ}C$. Short circuits from outputs to V_{CC} can cause excessive heating and eventual destruction.



7 Typical Characteristics

See Recommended Operating Conditions for device temperature limits.



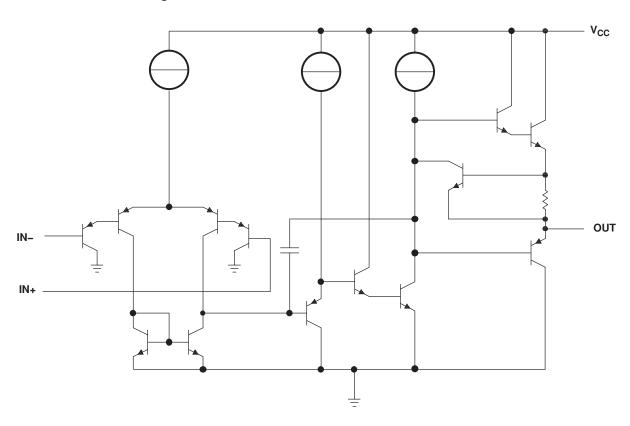


8 Detailed Description

8.1 Overview

The LP358 and LP2904 devices consist of two independent, low-power, unity-gain, stable operational amplifiers designed to operate from a single supply over a wide range of voltages. Operation from split supplies also is possible. The input voltage range includes ground and extends up to V_{CC} – 1.5 V. The output cannot drive to either rail, however, loads terminated to ground can support V_{OL} as low as ground. Loads to V_{CC} can support V_{OH} as high as V_{CC} .

8.2 Functional Block Diagram



8.3 Feature Description

8.3.1 Power Efficient Output Stage

Darlington source driver and emitter follower sink driver will pass bias current through the load to keep device quiescent current independent of load current.

8.3.2 Input Common-Mode Range

The valid common-mode range is from device ground to $V_{CC} - 1.5$ V. Inputs may exceed V_{CC} up to the maximum V_{CC} without device damage. At least one input must be in the valid input common-mode range for output to be correct phase. If both inputs exceed valid range then output phase is undefined. If either input is less than -0.3 V then input current must be limited to 1 mA and output phase is undefined.

8.4 Device Functional Modes

These devices are powered on when the supply is connected. This device can be operated as a single-supply operational amplifier or dual-supply amplifier depending on the application.

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9 Application and Implementation

NOTE

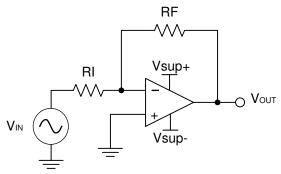
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The LP358 and LP2904 operational amplifiers are useful in a wide range of signal conditioning applications due to the wide $V_{\rm CC}$ range. Inputs can be powered before $V_{\rm CC}$ for flexibility in multiple supply circuits.

9.2 Typical Application

A typical application for an operational amplifier in an inverting amplifier. This amplifier takes a positive voltage on the input, and makes the voltage a negative voltage. In the same manner, the amplifier also makes negative voltages positive.



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Figure 4. Application Schematic

9.2.1 Design Requirements

The supply voltage must be selected such that it is larger than the input voltage range and output range. For instance, this application scales a signal of ± 0.5 V to ± 1.8 V. Setting the supply at ± 12 V is sufficient to accommodate this application

9.2.2 Detailed Design Procedure

Determine the gain required by the inverting amplifier using Equation 1 and Equation 2.

$$A_{V} = \frac{VOUT}{VIN}$$

$$A_{V} = \frac{1.8}{-0.5} = -3.6$$
(2)

Once the desired gain is determined, select a value for RI or RF. Selecting a value in the $k\Omega$ range is desirable because the amplifier circuit uses currents in the milliamp range. This ensures the part does not draw too much current. This example uses 10 $k\Omega$ for RI, so 36 $k\Omega$ is used for RF; this is determined by Equation 3.

$$A_{V} = -\frac{RF}{RI} \tag{3}$$



Typical Application (continued)

9.2.3 Application Curves

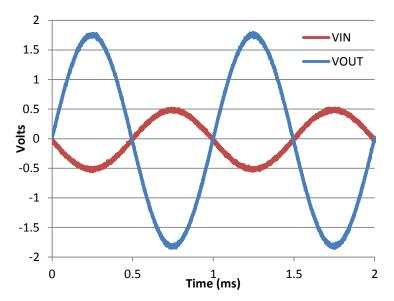


Figure 5. Input and Output Voltages of the Inverting Amplifier



10 Power Supply Recommendations

CAUTION

Supply voltages larger than 32 V can permanently damage the device (see *Absolute Maximum Ratings*).

Place $0.1-\mu F$ bypass capacitors close to the power-supply pins to reduce errors coupling in from noisy or high impedance power supplies. For more detailed information on bypass capacitor placement, see *Layout*.



11 Layout

11.1 Layout Guidelines

For best operational performance of the device, use good PCB layout practices, including:

- Noise can propagate into analog circuitry through the power pins of the circuit as a whole, as well as the
 operational amplifier. Bypass capacitors are used to reduce the coupled noise by providing low impedance
 power sources local to the analog circuitry.
 - Connect low-ESR, 0.1-μF ceramic bypass capacitors between each supply pin and ground, placed as close to the device as possible. A single bypass capacitor from V+ to ground is applicable for singlesupply applications.
- Separate grounding for analog and digital portions of circuitry is one of the simplest and most-effective
 methods of noise suppression. One or more layers on multilayer PCBs are usually devoted to ground planes.
 A ground plane helps distribute heat and reduces EMI noise pickup. Make sure to physically separate digital
 and analog grounds, paying attention to the flow of the ground current. For more detailed information, refer to
 Circuit Board Layout Techniques, (SLOA089).
- To reduce parasitic coupling, run the input traces as far away from the supply or output traces as possible. If it is not possible to keep them separate, it is much better to cross the sensitive trace perpendicular as opposed to in parallel with the noisy trace.
- Place the external components as close to the device as possible. Keeping RF and RG close to the inverting
 input minimizes parasitic capacitance, as shown in Layout Example.
- Keep the length of input traces as short as possible. Always remember that the input traces are the most sensitive part of the circuit.
- Consider a driven, low-impedance guard ring around the critical traces. A guard ring can significantly reduce leakage currents from nearby traces that are at different potentials.

11.2 Layout Example

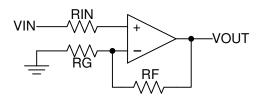


Figure 6. Operational Amplifier Schematic for Noninverting Configuration

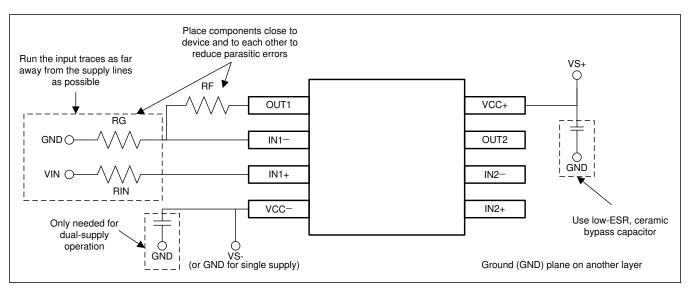


Figure 7. Operational Amplifier Board Layout for Noninverting Configuration

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12 Device and Documentation Support

12.1 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

Table 1. Related Links

PARTS	PRODUCT FOLDER	ORDER NOW	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
LP358	Click here	Click here	Click here	Click here	Click here
LP2904	Click here	Click here	Click here	Click here	Click here

12.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

12.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community TI's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

12.4 Trademarks

E2E is a trademark of Texas Instruments.

12.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

12.6 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.



13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material			Device Marking (4/5)	Samples
							(6)				
LP2904D	LIFEBUY	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	LP2904	
LP2904DR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	LP2904	Samples
LP358D	LIFEBUY	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	LP358	
LP358DR	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	LP358	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

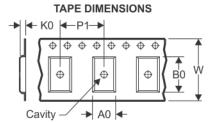
www.ti.com

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

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TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

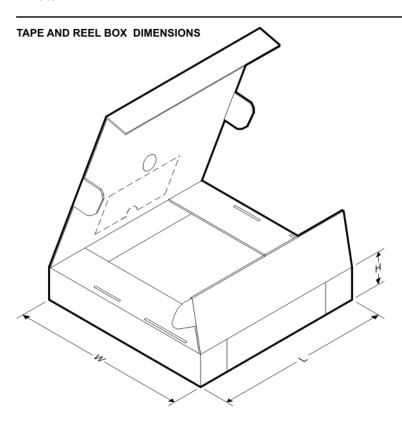
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

7 til dimonolorio aro nomina												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LP2904DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
LP358DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins SPQ		Length (mm)	Width (mm)	Height (mm)	
LP2904DR	SOIC	D	8	2500	340.5	336.1	25.0	
LP358DR	SOIC	D	8	2500	340.5	336.1	25.0	

PACKAGE MATERIALS INFORMATION

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TUBE



*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
LP2904D	D	SOIC	8	75	507	8	3940	4.32
LP358D	D	SOIC	8	75	507	8	3940	4.32

PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.5. Reference JEDEC registration MS-012, variation AA.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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