INTEGRATED CIRCUITS

Product specification Supersedes data of 1997 May 27 IC18 Data Handbook

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DESCRIPTION

The Serial Triple Gauge Driver (STGD), is a single chip air core driver providing drive to one major gauge, and two minor gauges, for automotive applications such as Speedometer, Fuel, Temperature, Tachometer, Volts, and Oil pressure information display. The STGD operates in conjunction with a microcontroller receiving serial data inputs, and can provide status back to the microcontroller either serially or via a status line. The protocol is compatible with the Philips Single Gauge Driver (SGD) and Dual Gauge Driver (DGD). The STGD also includes a protected battery supply for external single Serial Gauge Drivers or Dual Gauge Drivers.

PIN CONFIGURATION

 Figure 1. Pin Configuration

S_{CLK}O MINOR GAUGE 2 MINOR GAUGE 1 MAJOR GAUGE 10-BIT SR $\begin{array}{|c|c|c|c|c|}\n\hline\n\end{array}$ 10-BIT SR $\begin{array}{|c|c|c|c|}\n\hline\n\end{array}$ DATA_{OUT} $DATA_{IN}$ \circ CSO GOE O 9-BIT DATA 9-BIT DATA LATCH 10-BIT DATA LATCH 4-BIT STATUS
LATCH LATCH RUN ENABLE 7–BIT, SINE 7-BIT 7-BIT BIAS, TSD /COSINE Tan Tan DAC VBATT^O SwBATT, DAC DAC COMMON **REFERENCE** MUX MUX MUX Ġ. SwControl O-SwBATT1 O SwBATT2 \circ $C2-$ COM $\frac{1}{\mathcal{O}}$ \vec{c} GND COS– COS+ $\frac{1}{5}$ \vec{c} $\frac{1}{5}$ SR01117

BLOCK DIAGRAM

FEATURES

- Major Gauge 10-bit resolution Drive provides 0.35° resolution
	- **–** Sine/Cosine outputs for 360° operation
	- **–** 0.2° accuracy typical throughout entire range
- Minor gauge drivers provide 0.35° resolution
	- **–** 112° operation
	- **–** 0.5° accuracy typical throughout entire range
- Serial Data Input
	- **–** Supports interface from microcontrollers
	- **–** Compatible with Philips SGD SA5775A and DGD SA5777A
- Serial Data Output
	- **–** Permits the STGD to be wired in series using a common chip select to additional STGDs, SGDs, and DGDs
	- **–** Permits fault status information to be returned to the microcontroller
- Over Voltage Protection, Over Temperature Protection and Low Standby Current Operation
	- **–** Gauge drivers disabled when supply voltage exceeds specified operating voltage, protection to 40V.
	- **–** Gauge drivers disabled when die temperature exceeds operating range
	- **–** External switch may supply overvoltage protected battery supply to other devices operating off battery
- Thermally Enhanced SO-28 surface mount package

Figure 2. STGD Internal Block Diagram

ORDERING INFORMATION

PIN DESCRIPTION

ABSOLUTE MAXIMUM RATINGS

1. 160°C junction temperature is permitted during high battery (>16V) fault operation

DC ELECTRICAL CHARACTERISTICS

 V_{BATT} = 8.0 to 16V; T_{amb} = -40 to +105 °C

AC ELECTRICAL CHARACTERISTICS

 $V_{\text{BAT}} = 7.5$ to 16V; $T_{\text{amb}} = -40$ to +105°C

Figure 3. System Connections for the STGD

FUNCTIONAL DESCRIPTION

Figure 1 shows the pin-out of the STGD, which is packaged in an SO-28 pin package, enhanced for improved thermal management. Four pins on each side of the package serve as a heat spreader to remove heat from the die, and also function as the ground connection. The recommended mounting includes an area of copper on the PC board to aid in thermal management.

Figure 2 is a block diagram of the STGD. A serial interface connects the STGD to the microcontroller. A data output pin is provided to permit the STGD to be wired in series with other Philips air core gauge drivers such as the Serial Gauge Driver, SA5775, and the Dual Gauge Driver, SA5777 or additional STGDs. Status information may be passed back to the microcontroller via a status output, or via the serial interface.

Figure 3 shows the connection of the STGD in a typical application.

APPLICATION INFORMATION

Figure 4 demonstrates the connections between the STGD, the microcontroller, and optionally additional gauge drivers such as the SGD and DGD. With an active high on the chip select input (CS), data is shifted into the STGD through DATA_{IN} on the rising edge of S_{CLK}. Several gauge drivers may be wired in series using a common chip select and clock line, when more than three gauges are needed. The DATA $_{\text{OUT}}$ pins are cascaded to the DATA_{IN} pins of the following gauge drivers. Status information can be returned to the microcontroller via the ST pins of each gauge driver. These are open-drain, active low outputs, which may be wire OR'ed together to signal that a fault, such as a thermal shut down, has occurred within one of the gauge drivers. This pin may be connected to a microcontroller port pin for polling in software, or may be connected to an external interrupt input to cause entry into an interrupt service routine. The STGD, may also pass status information back to the microcontroller serially. The rising edge of chip select loads status information into the shift register for the first four bits that will be shifted out of the STGD by the shift clock. Figure 11 shows the data bits within the shift register. A low on the ST pin signals that one or more status bits have been set in the status register. A high indicates all status bits are reset. The status output bits include minor gauge over current, major gauge over current, thermal shutdown and RUN. Gauge data is captured in latches by the falling edge of the chip select.

Figure 4. Serial Communications Between STGD, Microcontroller and Other Gauge Drivers

Figure 5 shows the gauge connections to the STGD. The major gauge, G1, supports full 360° operation with two coils driven. The seven least significant bits of the gauge information are converted to an analog level by digital-to-analog converter. The display range is divided into eight sections, two sections per quadrant. The coils are driven with a Sine/Cosine approximation. The three most significant bits of gauge display information control the multiplexer to select which coil is fed by the DAC and which coil receives a fixed bias. The multiplexer also determines the polarity of the voltages supplied to the coils.

The minor gauges, G2 and G3, each have one coil driven by a DAC. The other coils of each gauge are wired in series with the switched battery supply to supply the bias. The switched battery supply is turned off during over voltage conditions. Only 9-bits of information are required for the minor gauges, however, 10-bits are shifted through the part to maintain compatibility with the SGD and DGD. Hence, all gauges, both major and minor, are supplied with 10-bit data for consistency.

Figure 5. Gauge Connections to the STGD

Figure 6. Gauge Enable/Standby Circuit and Over Voltage Protection Circuit

Figure 6 shows the protection and gauge enable logic for the STGD. The battery supply voltage V_{BAT} is monitored, and if the supply exceeds the specified operating range, the STGD is put in a shutdown mode in which the output buffers are disabled. The STGD will also enter the shutdown mode by excessive die temperature, and will return to normal operation when the die temperature decreases to within specified limits. Thermal shutdown may occur at V_{BATT} supply voltages over 16V at high ambient temperatures near 105°C. Internal logic will continue to function and status may be read out to determine the source of the shutdown. The STGD may be

placed in a standby mode with a low on both the GOE and RUN input pin. In this mode, battery current drain is minimized.

The SwBATT1 and SwBATT2 inputs are the supply for the DACs, and the output buffers driving the coils including the COM output which stabilizes the voltages applied to the bias coils of the minor gauges. Both SwBATT1 and SwBATT2 should be connected to the collector of the control transistor as these inputs are not connected internally and supply different portions of the circuit. This switched battery supply is protected from voltages exceeding the specified operating range and is controlled by the SwCONTROL output. This supply may optionally be used to supply additional circuits which operate from unregulated battery supplies but which need protection from over voltage transients. Typical devices which may benefit from this protection include the Serial Gauge Driver, SA5775A and Dual Gauge Driver, SA5777A, which are often used in conjunction with the STGD in 4 and 5 gauge applications.

This switched battery supply is turned off when the STGD enters the standby mode in response to the RUN and GOE inputs both being low, or a V_{BATT} supply exceeding the specified operating range. The switched battery supply depends on the RUN signal to prevent undesired needle movement on the minor gauges when going from standby to active mode. This movement would otherwise occur if the voltage to the fixed bias coils of the minor gauges was switched on before the coil voltages provided by the DACs within the STGD were defined. The start up jump is prevented as follows. In the sleep mode the switched battery supply is off, and the gauge drive outputs of the STGD are in a high impedance state. The gauges are in their zero position from the previous power-down sequence. When the RUN input goes high, but the GOE is kept low, the STGD enters the start up mode in which the minor gauges are driven to zero, the internal 5V regulator for the logic is turned on, and the switched battery supply is turned on to supply the bias coil and STGD output buffers. However, the output buffers for the major gauge remain in the high impedance output state. The microcontroller may load values into the STGD via the serial interface while GOE is low. When the microcontroller applies a high to GOE, the major gauge output buffers are enabled. When the RUN signal is removed the STGD continues to operate in the normal mode, however, the controlling microcontroller should also monitor RUN and, when it goes low, send a series of values to the STGD to move the needles to their zero positions before taking GOE low to put the part in the standby mode.

Table 1 describes the operation and control of the SwBATT supply, the output buffers, and the operations normally performed by the microcontroller. Normal operation of a vehicle will follow the sequence of the truth table from top to bottom. The RUN input is typically connected to the switched ignition voltage, while GOE is controlled by the microcontroller.

Table 1. Truth Table

THERMAL MANAGEMENT AND POWER DISSIPATION

The power dissipated by the STGD has three components. The first term in the equation below represents the power dissipated in the STGD from current through the coil resistance. This component of the power dissipation is a function of both the battery voltage and the coil resistance. Most of the external loads such as the coils are resistive, so the current drawn by the output buffers is proportional to the supply voltage, resulting in power dissipation that is proportional to the square of the supply voltage for these circuits.

The highest power dissipation for a given coil driver will occur when the coil voltage is being driven to 50% of V_{BAT} . Thus the power dissipated by each coil driver is $(V_{BATT}/2)* (V_{BATT}/2Rc)$ or V_{BATT}(V_{BATT}/4Rc). If the coil resistance of the two minor gauge coils and the two coils of the major gauge all have the same resistance, then the maximum total power dissipation of the drivers becomes 4*V_{BATT}(V_{BATT}/4Rc) or simply V_{BATT}(V_{BATT}/Rc). Much of the internal analog circuits appears to the supply pins as a current sink and is represented by the second term. The current drawn by these circuits is relatively constant despite changes in supply voltage, resulting in power dissipation that is proportional to the supply voltage. Finally some power is dissipated in driving the external PNP transistor used to control the switched battery supply. The total power dissipation is a combination of these components and may be calculated from the formula:

 $P_D=V_{BATT}(V_{BATT}/R_C)+V_{BATT}(0.012) +$ VOL2(VBATT–VOL2–VBE(PNP))/R^B

Where:

 P_D = Power dissipation in watts

 V_{BATT} = Battery supply voltage in volts

- R_C = Coil resistance in ohms at ambient temperature including any self heating effects
- V_{OL2} = Output low voltage of the SwCONTROL pin as specified in the DC Characteristics

 $V_{BE(PNP)}$ = The V_{BE} drop of the external PNP transistor

 R_B = Resistor is series with base of external PNP transistor.

The minimum value of $R_B = V_{BATTMAX}I_{OL} = 16/0.050=320 \Omega$

The actual value used is dependent on the current needed to keep the PNP in saturation.

All gauges at 45° to a quadrant axis, as this is the highest internal power dissipation position.

If only the nominal coil resistance is known at a given nominal ambient temperature such as 25°C, the coil operating resistance at a high temperature ambient may be calculated using the following formula:

$$
R_{CA} = R_{CN} (1 + (0.4\% / \degree C)^* ((T_{SH} + T_{amb}) - 25\degree C))
$$

Where:

 R_{CA} = Resistance of Coil at Ambient temperature, including self heating

 R_{CN} = Nominal Resistance of Coil at 25 $°C$, without self heating T_{amb} = Ambient temperature, $°C$

 T_{SH} = Self heating of coil, $°C$

0.4%/°C = Resistance increase coefficient for copper

Figure 7 shows power dissipation plotted as a function of coil resistance and voltage. Since coil resistance is a function of temperature, the maximum power dissipation plotted will only occur at the lowest specified operating temperature. The power dissipation is lowest at the highest ambient temperature because of the increase in coil resistance with temperature.

This maximum power dissipation will only occur during a fault condition in which the system voltage rises to 18V, generally because of a failed voltage regulator controlling the vehicles battery voltage. Power dissipation will be lower when air core meter movements with higher nominal coil resistance are used.

Figure 7. Power Dissipation of the STGD as a Function of Coil Resistance and Operating Voltage

The STGD is specified to operate up to V_{BATT} max. The over voltage shutdown circuit will turn off the output buffers and the switched battery supply when the battery voltage reaches V_{OVSD} Over temperature conditions will also cause the output buffers to be disabled.

The STGD employs a thermally enhanced SO-28 package. The center four pins on each side are fused to the die pad to create a path for removal of heat from the package to the copper foil on the PC board. An area of copper foil is required on the PC board for heat dissipation at higher power dissipation levels.

In order to determine the size of the copper foil required, both thermal testing and thermal modeling were used. The effective ΘJA (thermal resistance, junction to ambient) was determined using both single and double sided PCBs with heat-sinking copper foil areas. Figures 8 and 9 show the effect of PCB copper foil area on the effective thermal resistance of the STGD part/PCB system.

Figure 8 shows the thermal resistance of the STGD mounted on a PC board with heat-sinking copper on the component side only. Figure 9 is a similar plot for a two sided PC board (same size copper areas on each side). Both plots assume a 60 x 60 x 1.57 mm FR4 board with varying square-shaped sizes of 2 oz. copper. The two sided board also assumes 8 thermal bias with 0.36 mm2 cross section.

It is important to note that at such a high ambient temperature (worst case of 105°C assumed), radiation is just as significant as convection in the dissipation of heat. Good radiation is highly dependent on the emissivity of the heated surface, so the thermal radiation properties of the copper foil should be considered. Bare, clean copper is a good thermal conductor, but it has a low emissivity, and is therefore a bad radiator. It is recommended that the copper areas intended for heat dissipation be left covered with solder mask or otherwise blackened to increase the emissivity, thereby improving the heat radiating ability of the board.

Figure 8. θJA **for SO28 with 8 Fused Pins One-sided PCB (2 oz. Copper), e = 0.9, Tamb = 105**°**C, P = 1.4–1.8W**

Figure 9. θJA **for SO28 with 8 Fused Pins Two-sided PCB (2 oz. Copper), e = 0.9, Tamb = 105**°**C, P = 1.4–1.8W**

Sample Calculations for Power Dissipation and Thermal Management

Worst Case Example

The worst case example will occur when the STGD is operating at V_{BATTMAX} (16V, in the highest specified ambient temperature (105°C), and with the lowest specified coil resistance (171 ohms at 25°C). Typical coil self heating of 15°C is assumed.

Calculation of Coil resistance operating at 105°C ambient.

 $R_{\text{CA}} = R_{\text{CN}} (1 + (0.4\% \text{W}/^{\circ}\text{C})^*(T_{\text{SH}} + T_{\text{amb}}) - 25^{\circ}\text{C}))$

 $= 171 x(1+(0.4%((15+105)-25)))$

= 236 Ohms at T_{amb} =105°C, with 15°C of self heating.

Calculation of STGD power dissipation at 16 volt operation.

 $P_D = V_{BATT} (V_{BATT}/R_C) + V_{BATT} (0.012)$

 $+V_{OL2}$ (V_{BATT} – V_{OL2} – $V_{BE(PNP)}$) / R_L

 $= 16(16/236) + 16(0.012) + 1.5(16-1.5-0.5)/320$

= 1.085+0.192+0.066 Watts

 $= 1.34$ Watts

Required board area and Junction Temperature calculation

The maximum junction temperature desired is 150°C. The permissible temperature rise and required Θ_{JA} may be calculated as:

 $\Delta T = T_i - T_{amb}$ $\Theta_{JA} = \Delta T/P_D$

Where; ∆T = Temperature rise in °C

 P_D = Power dissipation

 ${\sf T}_{\sf j}$ = Junction Temperature

Tamb = Ambient Temperature

 $\Delta T = T_J - T_{amb} = 150 - 105 = 45^{\circ}C$

$$
\Theta_{JA} = \Delta T/P_D = 55^{\circ}C/1.34 \text{ watts} = 33^{\circ}C/W.
$$

From Figure 8, the copper area required, using a single sided board, to keep the junction temperature within limits is approximately 2200 mm2. Figure 9 shows 1200 mm2 is required on each side of a double-sided board.

The above example illustrates the worst case situation of the STGD operating in at a maximum battery voltage, with the lowest nominal coil resistance (171Ω at room temperature), and at the highest ambient temperature. This will produce the highest junction temperature. At lower ambient temperatures the power dissipation may be higher because the coil resistance is decreased, however the junction temperature will be lower.

Serial Interface

Figure 10 demonstrates the serial interface timing referenced in the AC specifications. Figure 11 shows the order of information transfer through the serial interface. On a low to high transition of the CS pin, status information replaces the four most significant bits of data in the shift register and are the first bits shifted out. Output data is changed on the falling edge of S_{CLK} , while input data is captured on the rising edge of S_{CLK} . Major gauge data is loaded first, starting with the most significant bit, followed by minor gauge 1 data then minor gauge 2 data.

Figure 10. Serial Interface Timing

Figure 11. Internal Shift Register

Figure 12. Major Gauge Output Voltages (VSWBATT = 14V)

Figure 13. Typical Minor Gauge Output Voltage vs. Input Code (VSWBATT = 14V)

Figure 14. Minor Gauge Total Span

Figure 15. Meter Position (degrees) vs. Input Code for Minor Gauges

NOTES

Data sheet status

[1] Please consult the most recently issued datasheet before initiating or completing a design.

Definitions

Short-form specification — The data in a short-form specification is extracted from a full data sheet with the same type number and title. For detailed information see the relevant data sheet or data handbook.

Limiting values definition — Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.

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