*Click **here** for production status of specific part numbers.*

MAX40242

20V, Low Input Bias-Current, Low-Noise, Dual Op Amp

General Description

The MAX40242 provides a combination of high voltage, low noise, low input bias current in a dual channel and features rail-to-rail at the output.

This dual amplifier operates over a wide supply voltage range from a single 2.7V to 20V supply or split ±1.35V to ±10V supplies and consumes only 1.2mA quiescent supply current per channel.

The MAX40242 is a unity-gain stable amplifier with a gain-bandwidth product of 10MHz. The device outputs drive up to 200pF load capacitor without any external isolation resistor compensation.

The MAX40242 is available in 8-thin wafer-level packages (WLPs) and is rated for operation over the -40°C to +125°C automotive temperature range.
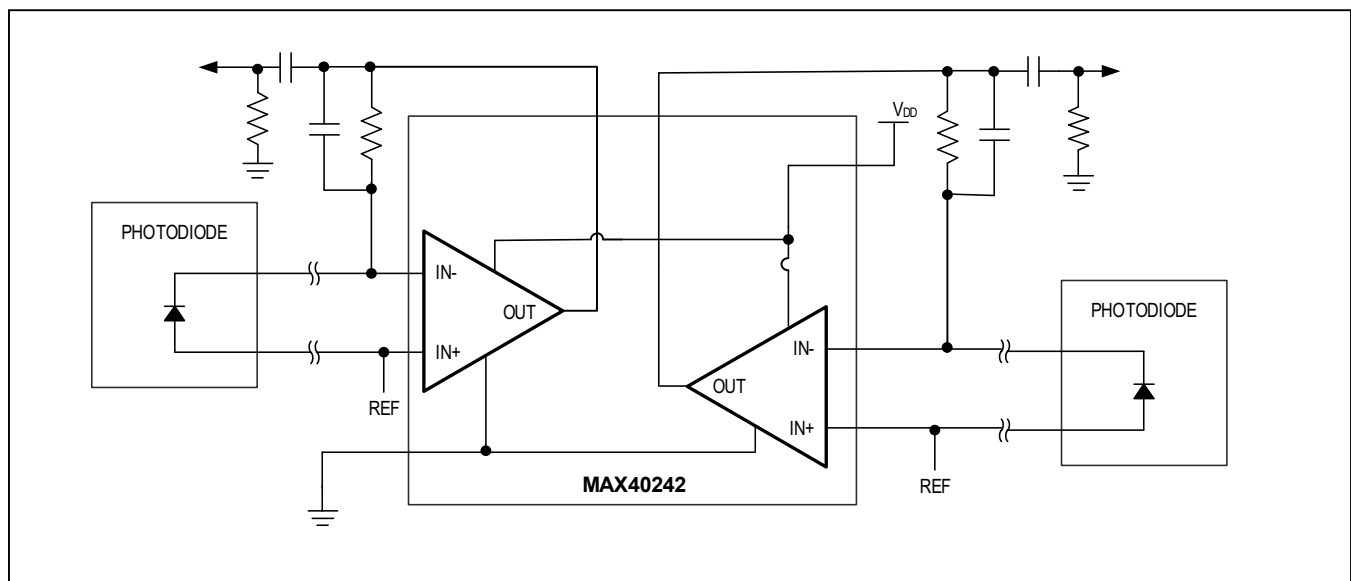
Applications

- Chemical Sensor Interface
- Photodiode Sensor Interface
- Medical Pulse Oximetry
- Industrial: Process and Control
- Precision Instrumentation

Benefits and Features

- 2.7V to 20V Single Supply or ±1.35V to ±10V Dual Supplies
- 2pA (Max) Input Bias Current
- 5nV/$\sqrt{Hz}$ Input Voltage Noise
- 10MHz Bandwidth
- 8V/µs Slew Rate
- Rail-to-Rail Output
- Integrated EMI Filters
- 1.2mA Supply Current per Amplifier
- Tiny, 0.85mm x 1.65mm 8-WLP

Ordering Information **appears at end of data sheet.**

Typical Application Circuit

maxim integrated™

Absolute Maximum Ratings

Supply Voltage (V_{DD} to V_{SS})................................-0.3V to +22V
All Other Pins ....(V_{SS} - 0.3V) to (V_{DD} + 0.3V)
Short-Circuit Duration to V_{DD} or V_{SS}.....................................1s
Continuous Input Current (Any Pins) ....±20mA
Differential Input Voltage ....±6V
Continuous Power Dissipation (T_A = +70°C)
  8-THIN WLP (derate 11.4mW/°C above +70°C) 912mW

Operating Temperature Range......................... -40°C to +125°C
Junction Temperature...+150°C
Storage Temperature Range........................... -65°C to +150°C
Lead Temperature (soldering, 10s) ....+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Package Thermal Characteristics (Note 1)

8-THIN WLP
  Junction-to-Ambient Thermal Resistance (θ_{JA}) .....87.71°C/W
  Junction-to-Case Thermal Resistance (θ_{JC})..............NA°C/W

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to **www.maximintegrated.com/thermal-tutorial**.

Electrical Characteristics

(V_{DD} = 10V, V_{SS} = 0V, V_{IN+} = V_{IN-} = $V_{DD}/2$, R_L = 10kΩ to $V_{DD}/2$, T_A = -40°C to +125°C, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
POWER SUPPLY							
Supply Voltage Range	V_{DD}	Guaranteed by PSRR		2.7		20	V
Power-Supply Rejection Ratio	PSRR	V_{DD} = 2.7V to 20V, V_{CM} = 0V	T_A = +25°C	106	130		dB
			-40°C ≤ T_A ≤ +125°C	100			
Quiescent Current Per Amplifier	I_{DD}	R_{LOAD} = infinity	T_A = +25°C		1.2	1.6	mA
			-40°C ≤ T_A ≤ +125°C			1.8	
Power-Up Time	t_{ON}	R_{LOAD} = 10kΩ to $V_{DD}/2$, C_{LOAD} = 20pF, V_{OUT} reaches $V_{DD}/2$ to 1%			20		µs
DC CHARACTERISTICS							
Input Common-Mode Range	V_{CM}	Guaranteed by CMRR test		V_{SS} - 0.05		V_{DD} - 1.5	V
Common-Mode Rejection Ratio	CMRR	V_{CM} = V_{SS} - 0.05V to V_{DD} - 1.5V	T_A = +25°C	94	111		dB
			-40°C ≤ T_A ≤ +125°C	90			
Input Offset Voltage	V_{OS}	T_A = +25°C			50	600	µV
		-40°C ≤ T_A ≤ +125°C				800	
Input Offset Voltage Drift (Note 3)	TC V_{OS}				0.25	2.5	µV/°C
Input Bias Current (Note 3)	I_B	T_A = +25°C			0.02	2	pA
		-40°C ≤ T_A ≤ +85°C				15	
		-40°C ≤ T_A ≤ +125°C				75	

Electrical Characteristics (continued)

(V_{DD} = 10V, V_{SS} = 0V, V_{IN+} = V_{IN-} = V_{DD}/2, R_L = 10kΩ to V_{DD}/2, T_A = -40°C to +125°C, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Input Offset Current (Note 3)	I_{OS}	T_A = +25°C			0.04	1	pA
		-40°C ≤ T_A ≤ +85°C				10	
		-40°C ≤ T_A ≤ +125°C				25	
Open Loop Gain	A_{VOL}	250mV ≤ V_{OUT} ≤ V_{DD} - 250mV	T_A = +25°C	134	145		dB
			-40°C ≤ T_A ≤ +125°C	129			
Input Resistance	R_{IN}	Differential			50		MΩ
		Common mode			200		
Output Short-Circuit Current (Note 3)		To V_{DD} or V_{SS}	Noncontinuous (1s)		95		mA
Output Voltage Low	V_{OL}	V_{OUT} - V_{SS}, R_{LOAD} = 10KΩ to V_{DD}/2, T_A = +25°C			11	15	mV
		V_{OUT} - V_{SS}, R_{LOAD} = 10KΩ to V_{DD}/2, -40°C < T_A < 125°C				25	
		V_{OUT} - V_{SS}, R_{LOAD} = 2KΩ to V_{DD}/2, T_A = +25°C			47	60	
		V_{OUT} - V_{SS}, R_{LOAD} = 2KΩ to V_{DD}/2, -40°C < T_A < 125°C				85	
Output Voltage High	V_{OH}	V_{OUT} - V_{SS}, R_{LOAD} = 10KΩ to V_{DD}/2, T_A = +25°C			20	26	mV
		V_{OUT} - V_{SS}, R_{LOAD} = 10KΩ to V_{DD}/2, -40°C < T_A < 125°C				37	
		V_{OUT} - V_{SS}, R_{LOAD} = 2KΩ to V_{DD}/2, T_A = +25°C			80	100	
		V_{OUT} - V_{SS}, R_{LOAD} = 2KΩ to V_{DD}/2, -40°C < T_A < 125°C				135	
AC CHARACTERISTICS							
Input Voltage-Noise Density	e_n	f = 1kHz			5		nV/√Hz
Input Voltage Noise		0.1Hz ≤ f ≤ 10Hz			1.6		µV_{P-P}
Input Current-Noise Density	I_N	f = 1kHz			0.3		pA/√Hz
Input Capacitance	C_{IN}				4		pF
Gain-Bandwidth Product	GBW				10		MHz
Phase Margin	PM	C_{LOAD} = 20pF			60		°
Slew Rate	SR	A_V = 1V/V, V_{OUT} = 2V_{P-P}, 10% to 90%			8		V/µs
Large-Signal Bandwidth	BW	R_{LOAD} = 10KΩ to V_{DD}/2, C_{LOAD} = 20pF, A_V = 1V/V			1		MHz
Capacitive Loading	C_{LOAD}	No sustained oscillation, A_V = 1V/V			200		pF
Crosstalk	X_T	R_{LOAD} = 2KΩ to V_{DD}/2, C_{LOAD} = 20pF, V_{OUT} = 5V_{P-P}, f = 100kHz			-98		dB

Electrical Characteristics (continued)

(V_{DD} = 10V, V_{SS} = 0V, V_{IN+} = V_{IN-} = V_{DD}/2, R_L = 10kΩ to V_{DD}/2, T_A = -40°C to +125°C, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)

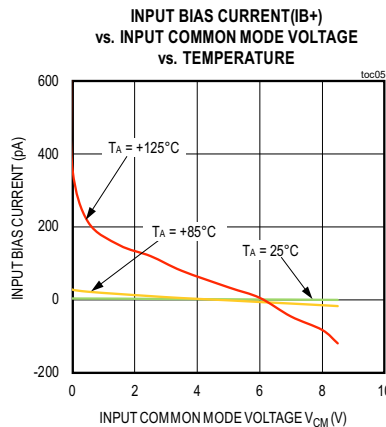| PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Total Harmonic Distortion Plus Noise	THD+N	V_{OUT} = 2V_{P-P}, A_V = +1V/V	f = 1kHz		-124		dB
			f = 20kHz		-100		
EMI Rejection Ratio	EMIRR	V_{RF_PEAK} = 100mV	f = 400MHz		35		dB
			f = 800MHz		40		
			f = 1800MHz		50		
			f = 2400MHz		57		
Settling Time		To 0.1%, V_{OUT} = 2V step, A_V = -1V/V			2		µs

Note 2: All devices are production tested at T_A = +25°C. Specifications over temperature are guaranteed by design.
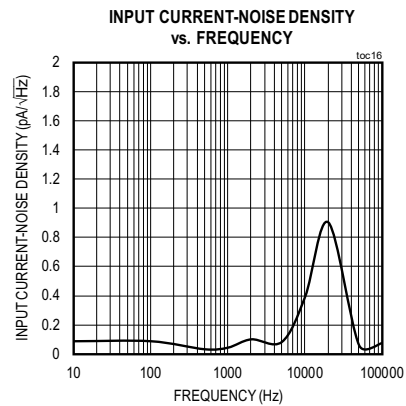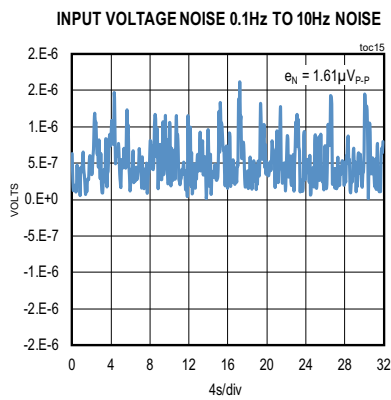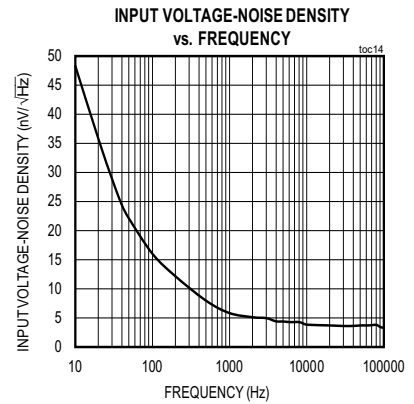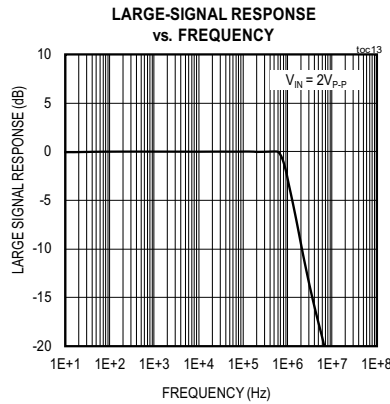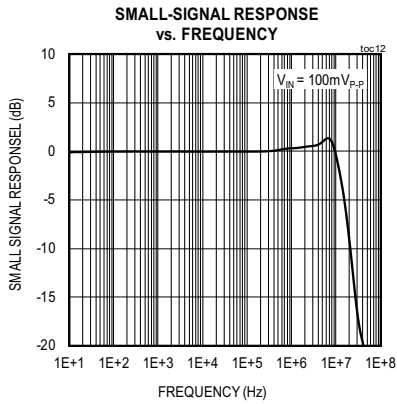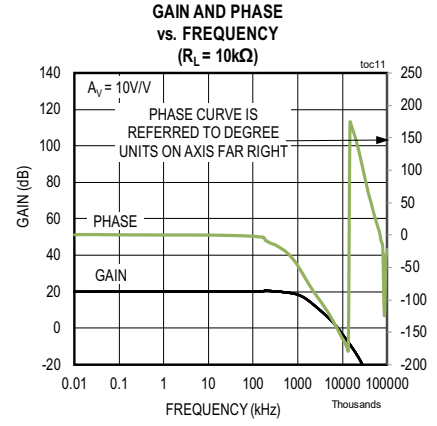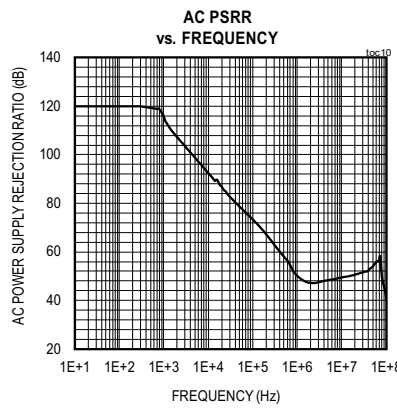**Note 3:** Guaranteed by design.

Typical Operating Characteristics

(V_{DD} = 10V, V_{SS} = 0V, outputs have R_L = 10kΩ to V_{DD}/2. T_A = +25°C, unless otherwise specified.)

INPUT OFFSET VOLTAGE HISTOGRAM

INPUT OFFSET VOLTAGE DRIFT HISTOGRAM

SUPPLY CURRENT PER AMPLIFIER vs. TEMPERATURE

INPUT OFFSET VOLTAGE vs. INPUT COMMON-MODE VOLTAGE vs. TEMPERATURE

INPUT BIAS CURRENT(IB+) vs. INPUT COMMON MODE VOLTAGE vs. TEMPERATURE

INPUT BIAS CURRENT(IB-) vs. INPUT COMMON MODE VOLTAGE vs. TEMPERATURE

COMMON-MODE REJECTION RATIO vs. TEMPERATURE
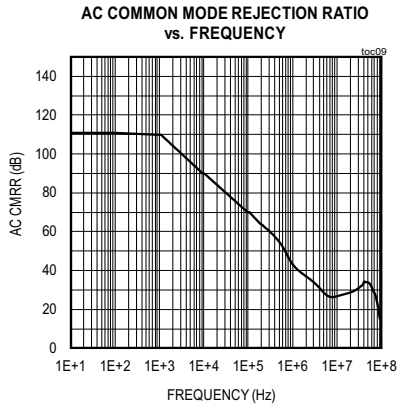
POWER-SUPPLY REJECTION RATIO vs. TEMPERATURE

Typical Operating Characteristics (continued)

(V_{DD} = 10V, V_{SS} = 0V, outputs have R_L = 10kΩ to V_{DD}/2. T_A = +25°C, unless otherwise specified.)

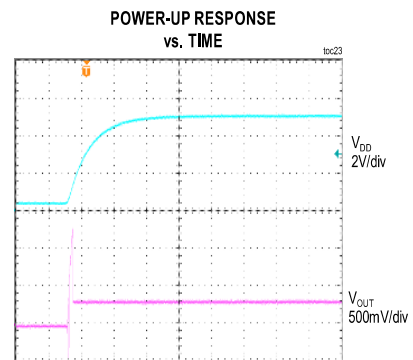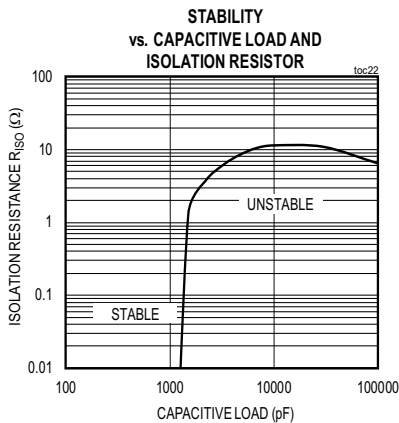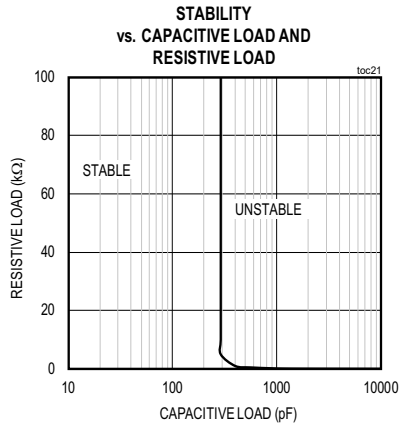**AC COMMON MODE REJECTION RATIO vs. FREQUENCY**

toc09

AC PSRR vs. FREQUENCY

toc10

GAIN AND PHASE vs. FREQUENCY (R_L = 10kΩ)

toc11

A_V = 10V/V

PHASE CURVE IS REFERRED TO DEGREE UNITS ON AXIS FAR RIGHT

PHASE

GAIN

Thousands

SMALL-SIGNAL RESPONSE vs. FREQUENCY

toc12

V_{IN} = 100mV$_{P-P}$

LARGE-SIGNAL RESPONSE vs. FREQUENCY

toc13

V_{IN} = 2V$_{P-P}$

INPUT VOLTAGE-NOISE DENSITY vs. FREQUENCY

toc14

INPUT VOLTAGE NOISE 0.1Hz TO 10Hz NOISE

toc15

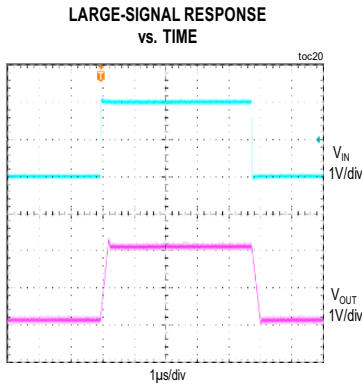e_N = 1.61µV$_{P-P}$

INPUT CURRENT-NOISE DENSITY vs. FREQUENCY

toc16

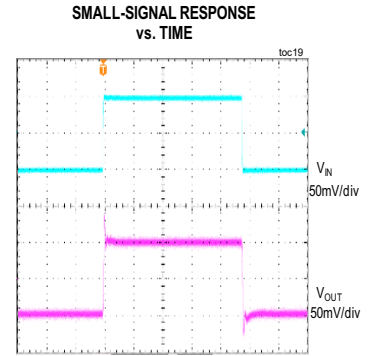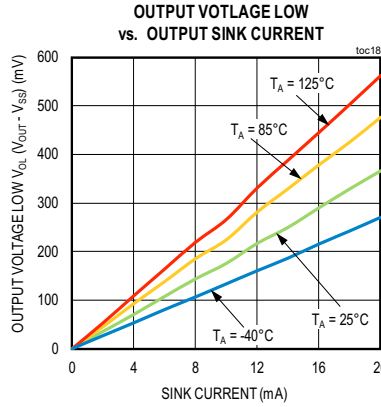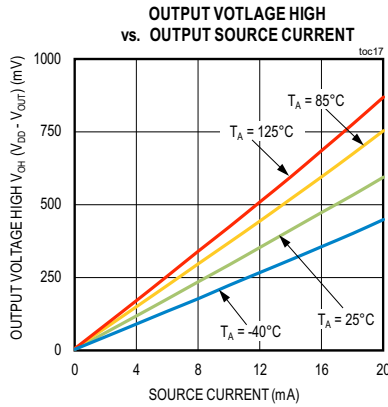## Typical Operating Characteristics (continued)

(V_{DD} = 10V, V_{SS} = 0V, outputs have R_L = 10kΩ to V_{DD}/2. T_A = +25°C, unless otherwise specified.)

OUTPUT VOLTAGE HIGH vs. OUTPUT SOURCE CURRENT

toc17

T_A = 85°C
T_A = 125°C
T_A = -40°C
T_A = 25°C

OUTPUT VOLTAGE LOW vs. OUTPUT SINK CURRENT

toc18

T_A = 125°C
T_A = 85°C
T_A = -40°C
T_A = 25°C

SMALL-SIGNAL RESPONSE vs. TIME

toc19

V_{IN} 50mV/div
V_{OUT} 50mV/div

LARGE-SIGNAL RESPONSE vs. TIME

toc20

V_{IN} 1V/div
V_{OUT} 1V/div
1μs/div

STABILITY vs. CAPACITIVE LOAD AND RESISTIVE LOAD

toc21

STABLE
UNSTABLE

STABILITY vs. CAPACITIVE LOAD AND ISOLATION RESISTOR

toc22

UNSTABLE
STABLE

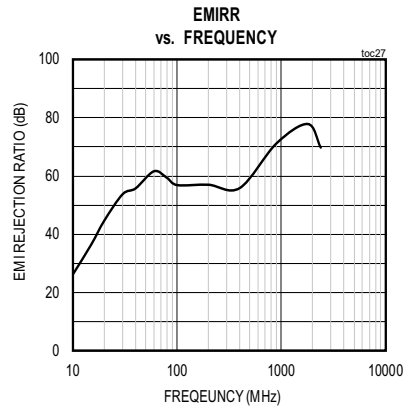**POWER-UP RESPONSE vs. TIME**

toc23

V_{DD} 2V/div
V_{OUT} 500mV/div

Typical Operating Characteristics (continued)

(V_{DD} = 10V, V_{SS} = 0V, outputs have R_L = 10kΩ to V_{DD}/2. T_A = +25°C, unless otherwise specified.)

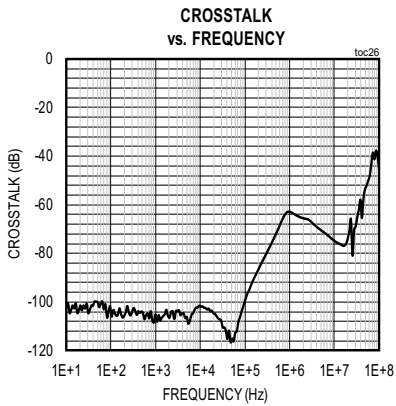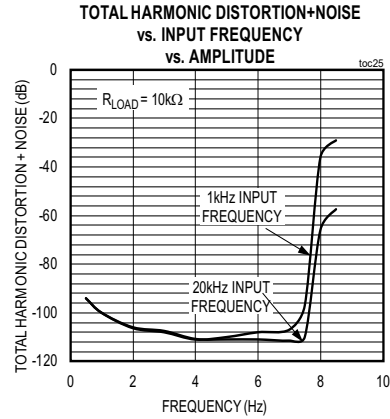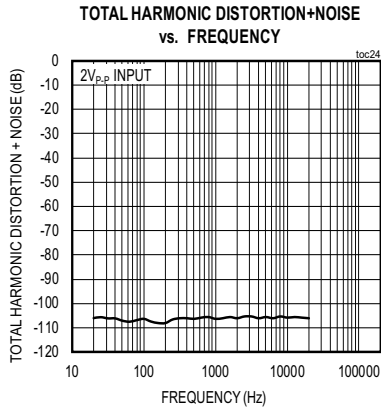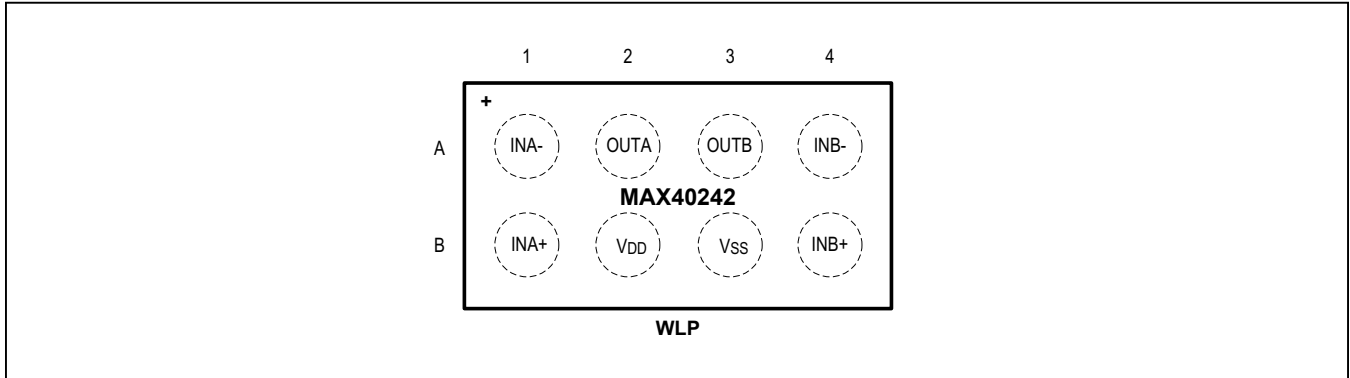**TOTAL HARMONIC DISTORTION+NOISE vs. FREQUENCY**

toc24

2V$_{P-P}$ INPUT

TOTAL HARMONIC DISTORTION+NOISE vs. INPUT FREQUENCY vs. AMPLITUDE

toc25

R_{LOAD} = 10kΩ

1kHz INPUT FREQUENCY

20kHz INPUT FREQUENCY

CROSSTALK vs. FREQUENCY

toc26

EMIRR vs. FREQUENCY

toc27

Pin Configuration

Bump Description

BUMP (WLP)	NAME	FUNCTION
A1	INA-	Channel A Negative Input
A2	OUTA	Channel A Output
A3	OUTB	Channel B Output
A4	INB-	Channel B Negative Input
B1	INA+	Channel A Positive Input
B2	V_{DD}	Positive Supply Voltage
B3	V_{SS}	Negative Supply Voltage. Connect V_{SS} to ground if single supply is used.
B4	INB+	Channel B Positive Input

Detailed Description

Combining high input impedance, low input bias current, wide bandwidth, and fast settling time, the MAX40242 is an ideal amplifier for driving precision analog-to-digital inputs and buffering digital-to-analog converter outputs.

Input Bias Current

The MAX40242 features a high-impedance CMOS input stage and a special ESD structure that allows low input bias current operation at low-input, common-mode voltages. Low input bias current is useful when interfacing with high-ohmic or capacitive sensors and is beneficial for designing transimpedance amplifiers for photodiode sensors. This makes the device ideal for ground-referenced medical and industrial sensor applications.

Integrated EMI Filter

Electromagnetic interference (EMI) noise occurs at higher frequency that results in malfunction or degradation of electrical equipment.

The MAX40242 has an input EMI filter to avoid the output from getting affected by radio frequency interference. The EMI filter, composed of passive devices, presents significant higher impedance to higher frequencies.

High Supply Voltage Range

The device features 1.2mA current consumption per channel and a voltage supply range from either 2.7V to 20V single supply or ±1.35V to ±10V split supply.

Typical Application Circuit

High-Impedance Sensor Application

High impedance sources like pH sensor, photodiodes in applications require negligible input leakage currents to the input transimpedance/buffer structure. The MAX40242 benefits with clean and precise signal conditioning due to its input structure.

The device interfaces to both current-output sensors (photodiodes) (Figure 1), and high-impedance voltage sources (piezoelectric sensors). For current output sensors, a transimpedance amplifier is the most noise-efficient method for converting the input signal to a voltage. High-value feedback resistors are commonly chosen to create large gains, while feedback capacitors help stabilize the amplifier by cancelling any poles introduced in the feedback loop by the highly capacitive sensor or cabling. A combination of low-current noise and low-voltage noise is important for these applications. Take care to calibrate out photodiode dark current if DC accuracy is important. The high bandwidth and slew rate also allow AC signal processing in certain medical photodiode sensor applications such as pulse-oximetry. For voltage-output sensors, a noninverting amplifier is typically used to buffer and/or apply a small gain to the input voltage signal. Due to the extremely high impedance of the sensor output, a low input bias current with minimal temperature variation is very important for these applications.

Transimpedance Amplifier

As shown in Figure 1, the noninverting pin is biased at 2V with C2 added to bypass high-frequency noise. This bias voltage to reverse biases the photodiode D1 at 2V which is often enough to minimize the capacitance across the junction. Hence, the reverse current (IR) produced by the photodiode as light photons are incident on it, a proportional voltage is produced at the output of the amplifier by the given relation:

$$V_{OUT} = I_R \times R1$$

The addition of C1 is to compensate for the instability caused due to the additional capacitance at the input (junction capacitance Cj and input capacitance of the op amp C_{IN}), which results in loss of phase margin. More information about stabilizing the transimpedance amplifier can be found in **Application Note 5129: Stabilize Your Transimpedance Amplifier**.
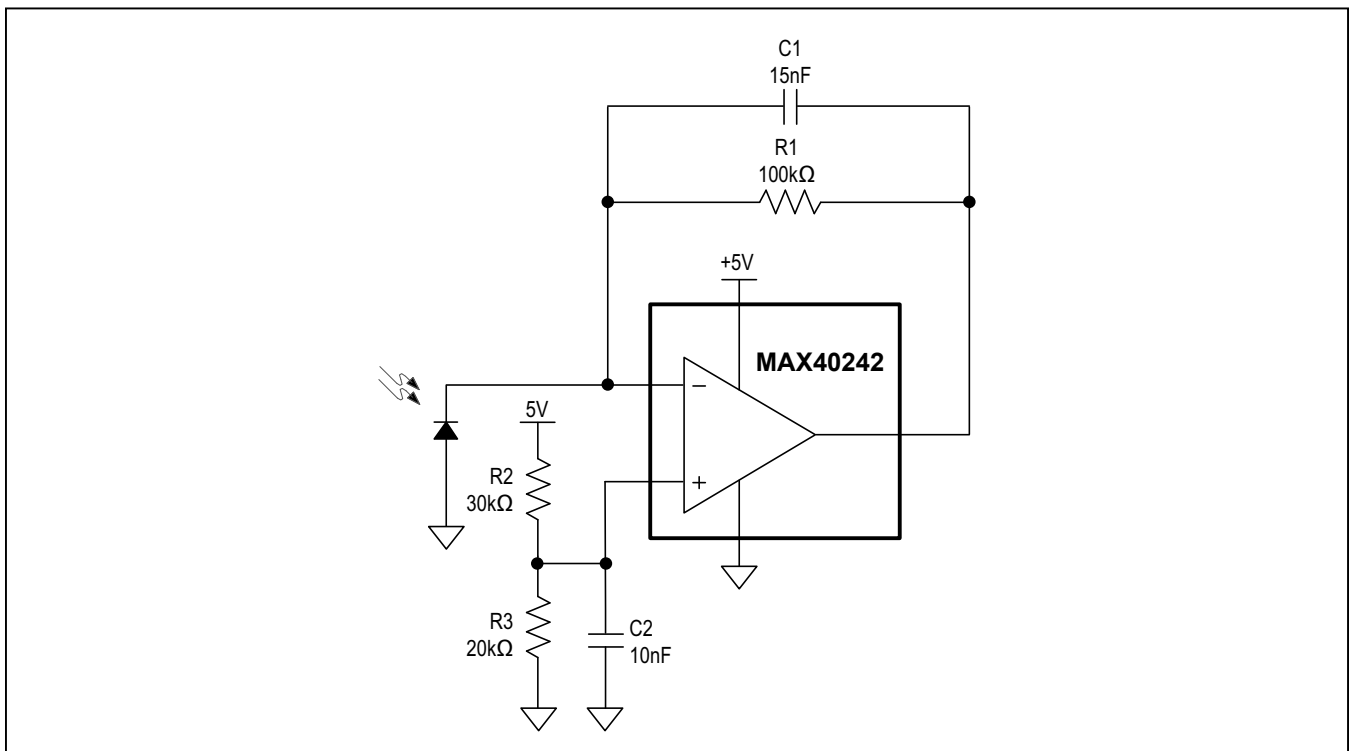

Figure 1. High-Impedance Source/Sensor Preamp Application

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE	TOP MARK
MAX40242ANA+	-40ºC to +125ºC	8-THIN WLP	+AAN

+Denotes lead(Pb)-free/RoHS-compliant package.

Chip Information

PROCESS: BiCMOS

Package Information

For the latest package outline information and land patterns (footprints), go to **www.maximintegrated.com/packages**. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
8-THIN WLP	N80D1+1	**21-100280**	**Apps Note 1891**

Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	6/18	Initial release	—