

# Precision Coulomb Counter / Low Side Current Monitor for High-current Applications

# **BD7220FV-C**

#### **General Description**

BD7220FV-C is a coulomb counter IC for high-current applications. The product integrates a high-accuracy operational amp, a current accumulation logic circuit and a 16-bit  $\Delta\Sigma$ ADC to perform current accumulation with high precision. The current sense input supports shunt resistors and output current sensors, and uses the SPI communication interface. BD7220FV-C only requires SPI commands to carry out the calibration necessary for high-precision measurement, so it can also get current accumulation information for battery state-of-charge estimation.

# **Features**

- AEC-Q100 Qualified (Note 1)
- 16-bit ΔΣADC
- Flexible Noise Filter (4 settings)
- Automatic Calibration via SPI
- High-accuracy Op-amp with 3 Gain Settings (5 V/V, 25 V/V, 51 V/V)
- Supports Current Sensing Using Shunt Resistors
- Supports Current Sensing Using Current Output Type Current Sensors
- SPI I/F (Optional CRC)
- Coulomb Counter Function with SPI External Communication
- Accumulation Current Counter which Counts Charge and Discharge Independently
- Adjustable Current Detection Interruption (3 settings)
- 4 Operation Modes (NORMAL, SLEEP, SSHDN, OFF)
- Wake Up Current Detection Function
- UVLO

(Note 1) Grade 1

# **Key Specifications**

■ Input Voltage Range

VCC Input Voltage Range 4.5 V to 5.5 V VDD Input Voltage Range 2.5 V to 5.5 V

Operating Temperature

-40 °C to +125 °C

#### **Applications**

- Battery Current Sense for EV
- Electricity Storage Systems
- Automated Guided Vehicle (AGV)
- Robot

# **Package**

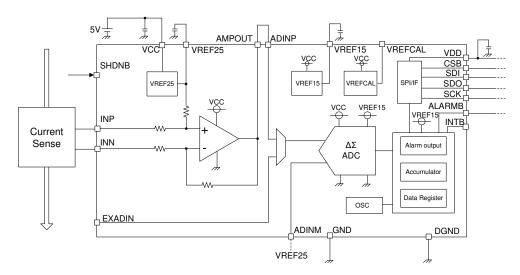
SSOP-B20

W (Typ) x D (Typ) x H (Max) 6.5 mm x 6.4 mm x 1.45 mm



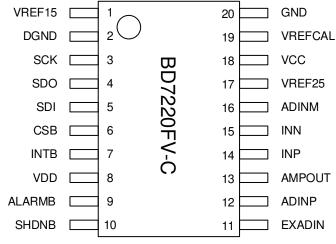
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# **Typical Application Circuit**



OProduct structure: Silicon integrated circuit OThis product has no designed protection against radioactive rays.

# **Pin Configuration**

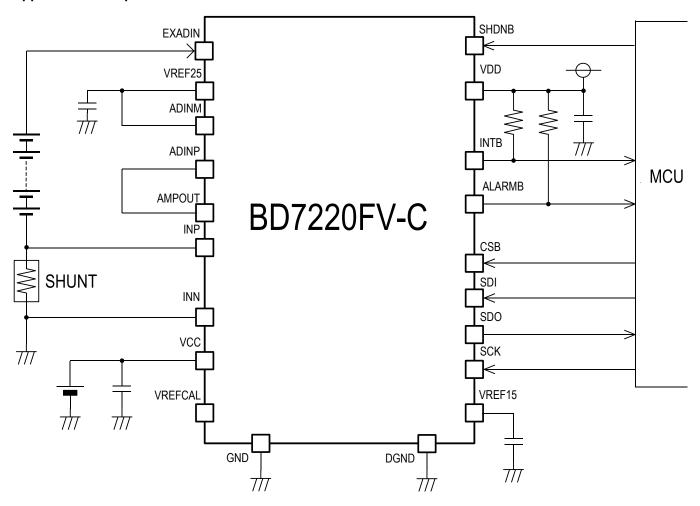


SSOP-B20 (TOP VIEW)

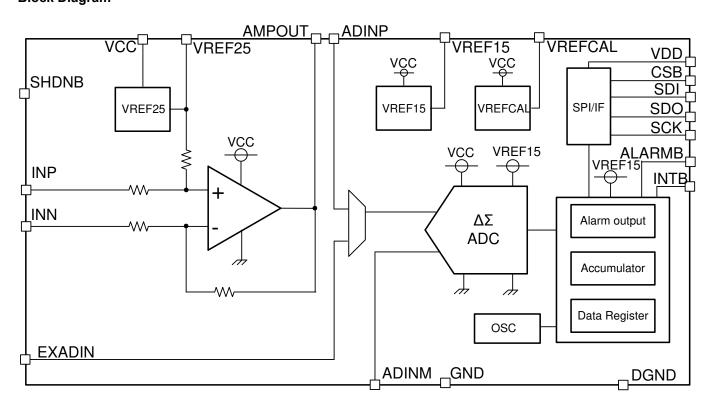
# **Pin Description**

Pin No.	Pin Name	I/O	Function
1	VREF15	0	LDO output for internal power
2	DGND	-	Digital ground
3	SCK	I	SPI clock input
4	SDO	0	SPI data output
5	SDI	I	SPI data input
6	CSB	I	SPI chip select input
7	INTB	0	Event interrupt output open drain
8	VDD	-	Power supply for SPI I/F
9	ALARMB	0	Alarm output open drain
10	SHDNB	I	Shutdown Input (H: operating, L: shutdown)
11	EXADIN	I	Delta sigma ADC select input
12	ADINP	I	Delta sigma ADC monitor input
13	AMPOUT	0	Internal amp output
14	INP	I	Internal amp non-inverting input
15	INN	I	Internal amp inverting input
16	ADINM	I	Delta sigma ADC reference input
17	VREF25	0	Internal reference output
18	VCC	-	Power supply
19	VREFCAL	0	Reference output for calibration in BD7220FV-C shipment process
20	GND	-	Analog ground

# **Application Example**



# **Block Diagram**



Absolute Maximum Ratings (Ta = 25 °C)

orate maximum reasings (re 20 s)			
Item	Symbol	Limit	Unit
Voltage Range1 (VCC, VDD)	V <sub>AMR_1</sub>	-0.3 to +7.0	V
Voltage Range2 (VREF15)	V <sub>AMR_2</sub>	-0.3 to +2.1	V
Voltage Range3 (VREF25, VREFCAL, INP, INN, EXADIN, SHDNB, INTB, ALARMB, AMPOUT, ADINP, ADINM, CSB, SDI, SDO, SCK)	V <sub>AMR_3</sub>	-0.3 to +7.0	V
Maximum Junction Temperature	Tjmax	150	°C
Storage Temperature Range	Tstg	-55 to +150	°C

Caution 1: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Caution 2: Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, design a PCB with thermal resistance taken into consideration by increasing board size and copper area so as not to exceed the maximum junction temperature rating.

#### Thermal Resistance (Note 2)

cima resistance				
Parameter	Symbol	Thermal Re	Unit	
Falameter	Symbol	1s (Note 4)	2s2p (Note 5)	Offic
SSOP-B20				
Junction to Ambient	θJA	115.4	57.3	°C/W
Junction to Top Characterization Parameter (Note 3)	$\Psi_{ m JT}$	10	8	°C/W

(Note 2) Based on JESD51-2A (Still-Air)

(Note 3) The thermal characterization parameter to report the difference between junction temperature and the temperature at the top center of the outside surface of the component package.

surface of the component package.
(Note 4) Using a PCB board based on JESD51-3
(Note 5) Using a PCB board based on JESD51-7

į	(Note 5) Using a PCB board based on JESD51-7.								
	Layer Number of Measurement Board	Material	Board Size						
	Single	FR-4	114.3 mm x 76.2 mm x 1.57 mmt						

Тор					
Copper Pattern	Thickness				
Footprints and Traces	70 µm				

Layer Number of Measurement Board	Material	Board Size		
4 Layers	FR-4	114.3 mm x 76.2 mm x 1.6 mmt		

Тор		2 Internal Lag	yers	Bottom		
Copper Pattern	Thickness	Copper Pattern	Thickness	Copper Pattern	Thickness	
Footprints and Traces	70 µm	74.2 mm x 74.2 mm	35 µm	74.2 mm x 74.2 mm	70 µm	

**Recommended Operating Condition** 

Itom	Cumbal		Limit		Unit	Condition
Item	Symbol	Min	Тур	Max		
Voltage Range1 (VCC)	Vopr_1	4.5	5.0	5.5	V	
Voltage Range2 (VDD)	Vopr_2	2.5	3.3	5.5	V	
Operating Temperature	Topr	-40	+25	+125	°C	

# **Electrical Characteristics**

(Unless otherwise specified, Ta = -40  $^{\circ}$ C to +125  $^{\circ}$ C, VCC = 5.0 V, VDD = 3.3 V)

Parameter	Symbol		Limit			Condition
r arameter	Symbol	Min	Тур	Max	Unit	Condition
VREF15						
Output Voltage1	V <sub>O15_1</sub>	1.470	1.500	1.530	٧	lo = 0 mA, Ta= +25 °C
Output Voltage2	V <sub>O15_2</sub>	1.450	1.500	1.550	٧	Io = 0 mA, Ta= -40 °C to +125 °C
Effective Output Capacitance	C <sub>VO15</sub>	0.4	T.	2.0	μF	Recommended Nominal Capacitor:1 µF
VREF25						
Output Voltage1	V <sub>O25_1</sub>	2.450	2.500	2.550	٧	lo = 0 mA, Ta= +25 °C
Output Voltage2	V <sub>O25_2</sub>	2.400	2.500	2.600	٧	Io = 0 mA, Ta= -40 °C to +125 °C
Effective Output Capacitance	C <sub>VO25</sub>	0.4	T.	2.0	μF	Recommended Nominal Capacitor:1 µF
osc						
Frequency1	f <sub>OSC_1</sub>	8110	8192	8274	kHz	Ta= +25 °C
Frequency2	f <sub>OSC_2</sub>	8028	8192	8356	kHz	Ta= -40 °C to +125 °C
Start up Time	t <sub>wakeosc</sub>	-	50	200	μs	

(Unless otherwise specified, Ta = -40  $^{\circ}$ C to +125  $^{\circ}$ C, VCC = 5.0 V, VDD = 3.3 V)

Doromotor	Cumbal	Limit			Linit	Condition
Parameter	Symbol	Min Typ Max		Max	Unit	Condition
Current Consumption						
Shutdown Current 1_1 (OFF) = HSHDN	I <sub>QVB1_1</sub>	-	0	5	μА	SHDNB = L, MODE_SEL [1:0] = 2'bXX VREF15 = OFF, VREF25 = OFF AMP = OFF, ΔΣADC = OFF OSC = OFF, Ta= +25 °C
Shutdown Current 1_2 (OFF) = HSHDN	I <sub>QVB1_2</sub>	-	0	10	μΑ	SHDNB = L, MODE_SEL [1:0] = 2'bXX VREF15 = OFF, VREF25 = OFF AMP = OFF, ΔΣADC = OFF OSC = OFF, Ta= -40 °C to +125 °C
Shutdown Current 2_1 (Soft Shutdown Mode) = SSHDN	I <sub>QVB2_1</sub>	-	20	40	μΑ	SHDNB = H, MODE_SEL [1:0] = 2'b11 VREF15 = ON, VREF25 = OFF AMP = OFF, ΔΣADC = OFF OSC = OFF, Ta= +25 °C
Shutdown Current 2_2 (Soft Shutdown Mode) = SSHDN	I <sub>QVB2_2</sub>	-	20	200	μA	SHDNB = H, MODE_SEL [1:0] = 2'b11 VREF15 = ON, VREF25 = OFF AMP = OFF, ΔΣADC = OFF OSC = OFF, Ta= -40 °C to +125 °C
Operating Current 1_1 (NORMAL Mode)	I <sub>QVB3_1</sub>	-	2.50	3.75	mA	SHDNB = H, MODE_SEL [1:0] = 2'b01 VREF15 = ON, VREF25 = ON AMP = ON, ΔΣADC = ON OSC = ON, Ta= +25 °C
Operating Current 1_2 (NORMAL Mode)	I <sub>QVB3_2</sub>	-	2.50	7.50	mA	SHDNB = H, MODE_SEL [1:0] = 2'b01 VREF15 = ON, VREF25 = ON AMP = ON, ΔΣΑDC = ON OSC = ON, Ta= -40 °C to +125 °C

# **Electrical Characteristics – continued**

(Unless otherwise specified, Ta = -40  $^{\circ}$ C to +125  $^{\circ}$ C, VCC = 5.0 V, VDD = 3.3 V)

Danaga atau	Complete al		Limit		l lada	Condition
Parameter	Symbol	Min	Тур	Max	Unit	Condition
Operating Current 2_1 (SLEEP Mode)	I <sub>QVB4_1_1</sub>	-	700	1050	μА	SHDNB = H, MODE_SEL [1:0] = 2'b10, SLEEP_INTERVAL [1:0] = 2'b00, SLEEP_SAMPLING_TIME [1:0] = 2'b00, VREF15 = ON, VREF25 = Intermittent Operation AMP = Intermittent Operation, ΔΣADC = Intermittent Operation, OSC = ON, Ta = +25 °C
Operating Current 2_2 (SLEEP Mode)	I <sub>QVB4_1_2</sub>	-	700	1250	μА	SHDNB = H, MODE_SEL [1:0] = 2'b10, SLEEP_INTERVAL [1:0] = 2'b00, SLEEP_SAMPLING_TIME [1:0] = 2'b00, VREF15 = ON, VREF25 = Intermittent Operation AMP = Intermittent Operation, $\Delta\Sigma$ ADC = Intermittent Operation, OSC = ON, Ta = -40 °C to +125 °C
Operating Current 3_1 (SLEEP Mode)	l <sub>OVB4_2_1</sub>	-	600	900	μА	SHDNB = H, MODE_SEL [1:0] = 2'b10, SLEEP_INTERVAL [1:0] = 2'b01, SLEEP_SAMPLING_TIME [1:0] = 2'b00, VREF15 = ON, VREF25 = Intermittent Operation AMP = Intermittent Operation, $\Delta\Sigma$ ADC = Intermittent Operation, OSC = ON, Ta= +25 °C
Operating Current 3_2 (SLEEP Mode)	I <sub>OVB4_2_2</sub>	-	600	1100	μА	SHDNB = H, MODE_SEL [1:0] = 2'b10, SLEEP_INTERVAL [1:0] = 2'b01, SLEEP_SAMPLING_TIME [1:0] = 2'b00, VREF15 = ON, VREF25 = Intermittent Operation AMP = Intermittent Operation, $\Delta\Sigma$ ADC = Intermittent Operation, OSC = ON, Ta= -40 °C to +125 °C
Operating Current 4_1 (SLEEP Mode)	I <sub>QVB4_3_1</sub>	-	540	810	μА	SHDNB = H, MODE_SEL [1:0] = 2'b10, SLEEP_INTERVAL [1:0] = 2'b10, SLEEP_SAMPLING_TIME [1:0] = 2'b00, VREF15 = ON, VREF25 = Intermittent Operation AMP = Intermittent Operation, ΔΣADC = Intermittent Operation, OSC = ON, Ta = +25 °C
Operating Current 4_2 (SLEEP Mode)	I <sub>QVB4_3_2</sub>	-	540	1010	μА	SHDNB = H, MODE_SEL [1:0] = 2'b10, SLEEP_INTERVAL [1:0] = 2'b10, SLEEP_SAMPLING_TIME [1:0] = 2'b00, VREF15 = ON, VREF25 = Intermittent Operation AMP = Intermittent Operation, $\Delta\Sigma$ ADC = Intermittent Operation, OSC = ON, Ta = -40 °C to +125 °C
Operating Current 5_1 (SLEEP Mode)	I <sub>OVB4_4_1</sub>	-	500	750	μА	SHDNB=H, MODE_SEL[1:0]=2'b10, SLEEP_INTERVAL[1:0]=2'b11, SLEEP_SAMPLING_TIME[1:0]=2'b00, VREF15=ON, VREF25=Intermittent Operation AMP=Intermittent Operation, ΔΣADC=Intermittent Operation, OSC = ON, Ta= +25 °C
Operating Current 5_2 (SLEEP Mode)	I <sub>OVB4_4_2</sub>	-	500	950	μА	SHDNB=H, MODE_SEL[1:0]=2'b10, SLEEP_INTERVAL[1:0]=2'b11, SLEEP_SAMPLING_TIME[1:0]=2'b00, VREF15=ON, VREF25=Intermittent Operation AMP=Intermittent Operation, ΔΣADC=Intermittent Operation, OSC = ON, Ta= -40 °C to +125 °C

# **Electrical Characteristics – continued**

(Unless otherwise specified, Ta = -40 °C to +125 °C, VCC = 5.0 V, VDD = 3.3 V)

Parameter	Symbol	Symbol		Unit	Condition	
Faiaillelei	Symbol	Min	Тур	Max	Offic	Condition
AMP Block						
Analog Input Valtage Range 1	V <sub>AIN1</sub>	-200	-	+400	mV	AMP_GAIN [1:0] = 2'b00(Gain = 5 V/V)
Analog Input Valtage Range 2	V <sub>AIN2</sub>	-80	-	+80	mV	AMP_GAIN [1:0] = 2'b01(Gain = 25 V/V) AMP_GAIN [1:0] = 2'b10(Gain = 25 V/V)
Analog Input Valtage Range 3	V <sub>AIN3</sub>	-40	-	+40	mV	AMP_GAIN [1:0] = 2'b11(Gain = 51 V/V)
Gain Setting Time	t <sub>GAINSET</sub>	-	-	1.5	ms	
ADC Block	•	•		_		
Resolution	-	-	-	16	bit	
ADC Conversion Time 1	t <sub>CONV1</sub>	0.20	0.25	0.30	ms	MCIC_R [1:0] = 2'b00 (Down sampling value = 32)
ADC Conversion Time 2	t <sub>CONV2</sub>	0.80	1.00	1.20	ms	MCIC_R [1:0] = 2'b01 (Down sampling value = 128)
ADC Conversion Time 3	t <sub>CONV3</sub>	1.60	2.00	2.40	ms	MCIC_R [1:0] = 2'b10 (Down sampling value = 256)
ADC Conversion Time 4	t <sub>CONV4</sub>	6.40	8.00	9.60	ms	MCIC_R [1:0] = 2'b11 (Down sampling value = 1024)
EXADIN Valtage Range	V <sub>EXADIN</sub>	0.5	-	4.5	V	

(Unless otherwise specified, Ta = -40 °C to +125 °C, VCC = 5.0 V, VDD = 3.3 V)

Parameter	Symbol	Limit			Unit	Condition	
Farameter	Syllibol	Min	Тур	Max	Ullit	Condition	
UVLO (VCC)							
VCC UVLO Detect Voltage	V <sub>CC_UVLOD</sub>	2.700	2.800	2.900	V	VCC = Sweep down	
VCC UVLO Release	V <sub>CC_UVLOR</sub>	2.717	3.000	3.283	V	VCC = Sweep up	
UVLO (VREF15)							
VREF15 UVLO Detect Voltage	V <sub>REF15_UVLOD</sub>	1.352	1.380	1.408	V	VREF15 = Sweep down	
UVLO (VDD)							
VDD UVLO Detect Voltage	$V_{DD\_UVLOD}$	1.550	1.700	1.850	V	VDD = Sweep down	
VDD UVLO Release	$V_{DD\_UVLOR}$	1.650	1.800	1.950	V	VDD = Sweep up	

# **Electrical Characteristics - continued**

(Unless otherwise specified, Ta = -40  $^{\circ}$ C to +125  $^{\circ}$ C, VCC = 5.0 V, VDD = 3.3 V)

Б .		Limit		11.2	Condition	
Parameter	Symbol	Min	Тур	Max	Unit	Condition
IO Interface						
SHDNB Input "H" Level	$V_{\text{IH\_SHDNB}}$	VCCx0.7	-	VCC+0.3	V	
SHDNB Input "L" Level	$V_{\text{IL\_SHDNB}}$	-0.3	-	+VCCx0.3	٧	
SHDNB Input Leak Current	I <sub>OFF_SHDNB</sub>	-1	-	+1	μΑ	
INTB Output "L" Level Voltage1	V <sub>OL_INTB1</sub>	0	-	0.4	٧	I <sub>O</sub> = 1 mA, Ta= +25 °C
INTB Output "L" Level Voltage2	V <sub>OL_INTB2</sub>	0	-	0.5	V	I <sub>O</sub> = 1 mA, Ta= -40 °C to +125 °C
INTB Output Off Leak Current	I <sub>OLK_INTB</sub>	-1	-	+1	μA	INTB = 5.5 V
ALARMB Output "L" Level Voltage1	V <sub>OL_ALARMB1</sub>	0	-	0.4	V	I <sub>O</sub> = 1 mA, Ta= +25 °C
ALARMB Output "L" Level Voltage2	V <sub>OL_ALARMB2</sub>	0	-	0.5	V	I <sub>O</sub> = 1 mA, Ta= -40 °C to +125 °C
ALARMB Output Off Leak Current	I <sub>OLK_ALARMB</sub>	-1	-	+1	μΑ	ALARMB = 5.5 V
SPI Bus Interface						
CSB, SDI, SCK Input "H" Level Voltage	$V_{\text{IH\_SPI}}$	VDDx0.7	-	VDD+0.3	٧	
CSB, SDI, SCK Input "L" Level Voltage	$V_{IL\_SPI}$	-0.3	-	+VDDx0.3	V	
SDO Output "L" Level Voltage	V <sub>OL_SDO</sub>	0	-	0.4	V	I <sub>OL</sub> = 1 mA
SDO Output "H" Level Voltage	V <sub>OH_SDO</sub>	VDD-0.2	-	VDD	٧	Ι <sub>ΟΗ</sub> = -100 μΑ
CSB, SDI, SCK Input Leak Current	I <sub>OLK_SPI</sub>	-1	0	+1	μА	
SDO Output Off Leak Current	I <sub>OLK_SDO</sub>	-1	0	+1	μΑ	
CSB-SCK Setup Time	t <sub>CSS</sub>	1000	-	-	ns	
SCK-CSB Hold Time	t <sub>CSH</sub>	1000	-	-	ns	
SCK "H" Pulse Width Time	t <sub>WH</sub>	1000	-	-	ns	
SCK "L" Pulse Width Time	t <sub>WL</sub>	1000	-	-	ns	
SCK-SDI Setup Time	t <sub>DIS</sub>	150	-	-	ns	
SCK-SDI Hold Time	t <sub>DIH</sub>	150	-	-	ns	
SCK-SDO Delay Time	t <sub>DOD</sub>	-	-	400	ns	
CSB "H" Pulse Width Time	t <sub>CS</sub>	500	-	-	ns	

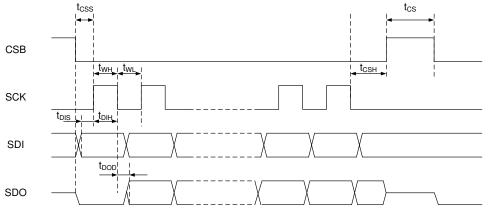


Figure 1. SPI Timing Chart

# **Typical Performance Curve (Reference Data)**

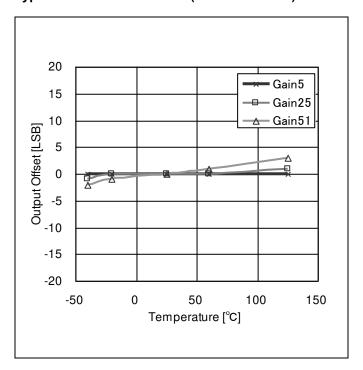


Figure 2. Output Offset vs Temperature (VCC = 5 V)

# **Description of Blocks**

#### 1. Current Accumulator

#### 1.1 Summary

This product performs accumulation current operation (Coulomb count) by monitoring the battery charge/discharge current. The Coulomb count value obtained is used for battery state-of-charge estimation. It can obtain both the current accumulation level and the current value itself.

## 1.1.1 Features

- ·Reading current converted to voltage by an external resistor as a digital value
- ·Accumulation current counters which monitors the charge/discharge current value
- ·Independent monitoring of charge and discharge values
- ·Current value monitoring via SPI
- ·Fixed-interval average current read function (interval can be adjusted in 4 stages)
- ·Overwrite function of accumulated current data via SPI I/F

#### 1.1.2 Composition

The Current Accumulator block is comprised of 3 sub-blocks. (See figure 1-1)

# Differential op-amp (AMP)

The AMP sub-block monitors the voltage converted by the external resistor  $R_{SNS}$  at the INP and INN pins. The voltage difference between INP and INN is then amplified to a voltage suitable for the  $\Delta\Sigma$ ADC input.

Address 00h (AMP\_GAIN [1:0]) can be used to adjust voltage amplification gain to 5 V/V, 25 V/V, or 51 V/V. Take note that the input voltage range changes with this gain setting.

#### **ΔΣADC**

The 16-bit ΔΣADC has an analog-to-digital conversion rate of 4 kHz using the standard setting.

A digital filter determines the ADC's output rate and frequency response. This filter has 4 settings, accessible through address 00h (MCIC\_R [1:0]). Please set an appropriate value considering the influence of the current's frequency response, agitation noise, and the like.

# **Accumulator**

The accumulator operates by using the current value in digital form from the  $\Delta\Sigma$ ADC output.

Chapter 1.3 details the calculation function of the Accumulator, while Chapter 3 describes the interrupt function.

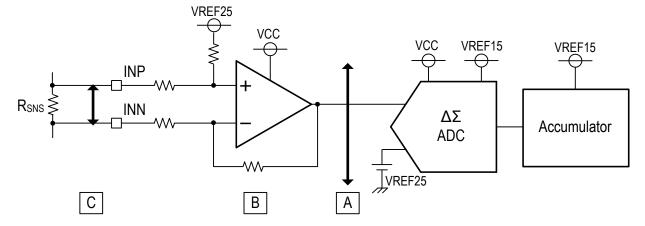


Figure 1-1. Accumulation Current Block Diagram

# 1.2 Register Structure

Regarding resister structure for current measurement, refer to Figure 1-2.

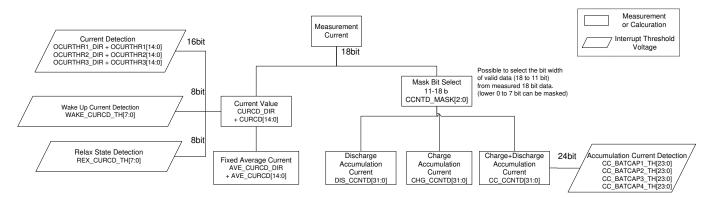


Figure 1-2. Register Structure for Current Measurement

# 1.3 Function Description

## 1.3.1 Current Measurement

BD7220FV-C measures current by monitoring the difference in voltage between the ends of the external current sense resistor ( $R_{SNS}$ ) connected to the differential op-amp inputs INP and INN. The op-amp amplifies the differential input biased on 2.5 V (VREF25) as a reference voltage. The amplified voltage is then used as input to the  $\Delta\Sigma$ ADC. For example, for a difference of 100 mV between the INP and INN pins and a 5 V/V gain setting:

$$2.5 \text{ [V]} + 100 \text{ [mV]} \times 5 = 3.0 \text{ [V]}$$

In this calculation, 3.0 V is input to the  $\Delta\Sigma$ ADC. (ADINP = 3.0 V, ADINM = 2.5 V,  $\Delta$ V = 0.5 V)

At shipment, the  $\Delta\Sigma$ ADC input voltage is defined as  $\Delta V = 4.5 \text{ V}$ . (ADINP=0.25 V to 4.75 V)

A: ΔΣ Input LSB Voltage = 
$$4.5 \div 2^{16} \approx \pm 68.66455$$
 [µV]

The CURCD register indicates the current value and is defined as [sign bit] + [15 bits] based on the computed LSB voltage.

Table 1-1 shows current unit information for each gain setting.

The LSB voltage input for the differential amplifier (Figure 1-1C) is the  $\Delta\Sigma$ ADC input LSB voltage divided by the amplifier gain setting (Figure 1-1B). The current flowing through the sense resistor R<sub>SNS</sub> is calculated using the 1LSB voltage and the sense resistor value.

The data in Table 1-1 is calculated with the assumption that  $R_{SNS} = 0.2 \text{ m}\Omega$ . If, for example,  $R_{SNS} = 0.1 \text{ m}\Omega$ , the current value (and, the measurement current range is same) in the table 1-1 is doubled.

Table 1-1. Current Monitor Unit

AMP_GAIN [1:0] register	Gain [times]	ΔΣΑDC Input 1LSB Voltage [μV]	OPamp Input 1LSB Voltage (INP-INN) [µV]	1LSB Current (@R <sub>SNS</sub> = 0.2 mΩ) [mA]	OPamp Input Voltage Range [mV]	Available Measurement Current Range (@R <sub>SNS</sub> = 0.2 mΩ) [A]
2'b00	5	68.66	13.73	68.66	-200 to +400	-1000 to +2000
2'b01, 2'b10	25	68.66	2.75	13.73	±80	±400
2'b11	51	68.66	1.35	6.73	±40	±200

Op-amp input LSB voltage

= ΔΣADC input LSB voltage ÷ gain setting value

LSB current

= Op-amp input LSB voltage ÷ external current sense resistance (R<sub>SNS</sub>)

Current measurement range = Op-amp input voltage range ÷ external current sense resistance (R<sub>SNS</sub>)

Figure 1-3 shows the structure of the CURCD register.

The MSB indicates the direction of the current and shows a current value using the remaining 15 bits.

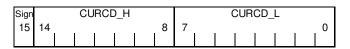


Figure 1-3. CURCD Register

When the current is 100 A, CURCD reads 1C71 [Hex] under the 25 V/V gain setting.

$$100$$
[A]  $\div 13.7329$ [mA]  $\approx 7281.78 \approx 7281$ [DEC] =  $1C71$ [HEX] =  $15'h1C71$ 

The lower bits that comprise the fractional value will be truncated.

#### Unit of Accumulation Current

BD7220FV-C can calculate and output accumulation current value based on the measured current. To enable this function, please set the CCNTEN register (address 00h) = 1. (Refer to 1.3.1 Current Measurement) The accumulation current value can be found in the 32-bit register CC\_CCNTD. The LSB and MSB values are computed as below. The accumulation current value is calculated with higher precision than the CC\_CCNTD register's LSB value.

LSB for internal accumulation current = current LSB  $\times$  (AD conversion term [s]  $\div$ 3600)

$$ex.$$
)  $GAIN = 5: 68.66$  ·· [mA] ×  $(250 \times 10^{-6} \div 3600) \approx 4.77$  [nAh]

 $LSB \ for \ CC_{CCNTD} register = \\ internal \ accumulation \ current \ LSB \ \times internal \ adjustment (6bits)$ ex.)  $GAIN = 5: 4.77 \text{ [nAh]} \times 2^6 \approx 0.3052 \text{ [µAh]}$ 

MSB value for  $CC_{CCNTD}$  register =  $CC_{CCNTD}$  register LSB  $\times$  2<sup>31</sup> = 655.36 [Ah]

Figure 1-4 shows structure of the CC CCNTD register and approximate accumulation current.

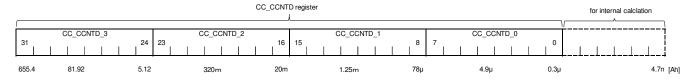


Figure 1-4. Structure of CC CCNTD Register (CC UNDIV=0)

To cancel the variation of current LSB in each amplifier gain setting, the current value can be scaled and accumulated in CC CCNTD register in the logic. This makes the LSB of accumulation current constant regardless of amplifier gain. For example, when current of 70 mA flows in R<sub>SNS</sub>, the CURCD register reading is 15'h0001 under a 5 V/V gain setting. Under a 25 V/V setting, CURCD reading is 15'h0005. At 25 V/V gain, not to accumulate 5 times larger from actual value, the logic accumulates the CURCD value divided by 5 to get the equivalent value at 5 V/V gain. At 51 V/V, the divisor is 10.2. The CC UNDIV register (address 01h) allows the user to enable or disable CURCD value scaling. Using scaling (CC UNDIV = "0") introduces a small amount of error but allows the user to change three amplifier gain settings. Disabling scaling (CC UNDIV = "1") requires that the amplifier setting be fixed, but ensures that errors are not introduced. However, the accumulation current capacity changes depending on the gain setting. Comparing to the value of 5 V/V setting, it is 1/5 when the setting is 25 V/V, and 1/10.2 when the setting is 51 V/V. It is possible to write the value of accumulation current into CC\_CCNTD resister. Please execute write operation after setting "0" to CCNTEN resister and current accumulating is

disabled. When CCNTEN resister is "0", update of accumulation current is discarded.

CC_UNDIV = 0 (Variable Gain)	2'b00(5 V/V) 2'b01(25 V/V) 2'b10(25 V/V) 2'b11(51 V/V)	_	_	UNIT	
CC_UNDIV = 1 (Fixed Gain)	2'b00(5 V/V)	2'b01(25 V/V) 2'b10(25 V/V)	2'b11(51 V/V)		
Internal Accumulation 1LSB	4.77	0.95	0.47	[nAh]	
CC_CCNTD_0 LSB	0.3052	0.0610	0.0299	[µAh]	
CC_CCNTD_1 LSB	78.125	15.625	7.659	[µAh]	
CC_CCNTD_2 LSB	20.00	4.00	1.96	[mAh]	
CC_CCNTD_3 LSB	5.12	1.02	0.50	[Ah]	
CC_CCNTD_3 MSB	655.36	131.07	64.25	[Ah]	
Maximum Accumulation Current	1310.41	262.08	128.47	[Ah]	

Table 1-2. Unit of CC CCNTD Register Accumulation ( $R_{SNS} = 0.2 \text{ m}\Omega$ )

Unit of Accumulation Current - continued

In addition, BD7220FV-C can accumulate charging accumulation current (via CHG\_CCNTD) and discharging accumulation current (via DIS\_CCNTD) separately. Please refer to Figure 1-5 and Figure 1-6 about the structure of CHG\_CCNTD and DIS\_CCNTD resister and their relationship with CC\_CCNTD.

CHG\_CCNTD is shifted 2 bits to the left from CC\_CCNTD. Accumulation scaling in internal logic for CHG\_CCNTD and DIS\_CCNTD is the same as in CC\_CCNTD, so the unit of accumulation current is the same regardless of the amplifier gain setting. Thus, the value of LSB is greater, but CHG\_CCNTD and DIS\_CCNTD capacities are 4 times that of CC\_CCNTD. The maximum battery capacity of CC\_CCNTD is around 1310[Ah], so CHG\_CCNTD and DIS\_CCNTD can accumulate up to 5240[Ah].

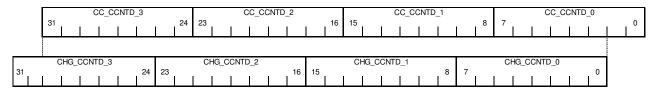


Figure 1-5. Relation between CC\_CCNTD and CHG\_CCNTD

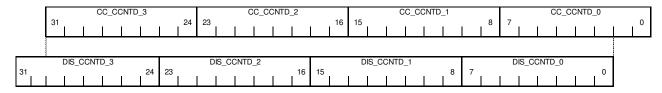


Figure 1-6. Relation between CC CCNTD and DIS CCNTD

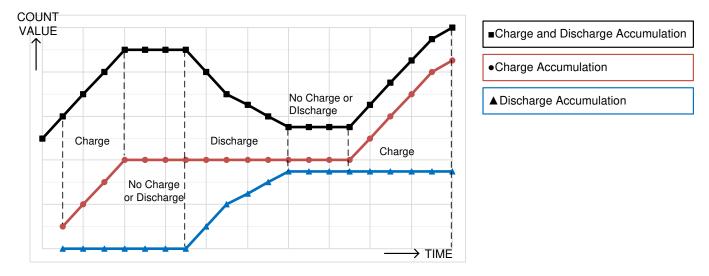


Figure 1-7. Difference between charge accumulation current and charge/discharge accumulation current

The values of CC\_CCNTD, CHG\_CCNTD and DIS\_CCNTD can be cleared by writing "1" to their reset registers. Resetting is valid regardless of the CCNTEN setting. The reset register is automatically cleared to "0" after writing "1".

Table 1-3. CCNTD Registers and Reset Registers

Register Name	Address	Bit	Bit Name	Function
CC_CCNTD	17h to 1Ah	[7] to [0]	CC_CCNTD [31:0]	Charge and discharge accumulation current
CHG_CCNTD	1Bh to 1Eh	[7] to [0]	CHG_CCNTD [31:0]	Charge accumulation current only
DIS_CCNTD	1Fh to 22h	[7] to [0]	DIS_CCNTD [31:0]	Discharge accumulation current only
CC_TRG_RST_CMD	02h	[0]	CCNTRST	Reset CC_CCNTD
CC_TRG_RST_CMD	02h	[1]	CHG_CCNTD_RST	Reset CHG_CCNTD
CC_TRG_RST_CMD	02h	[2]	DIS_CCNTD_RST	Reset DIS_CCNTD

# 1.3.3 Fixed-Interval Average Current

BD7220FV-C can output average current at fixed intervals set using registers. Moving average is not used.

The AVE\_CURCD\_DIR register denotes the current direction and the AVE\_CURCD [14:0] register represents the current value. The fixed interval is determined by the  $\Delta\Sigma$ ADC digital filter setting (address 00h: MCIC\_R [1:0]), OSR setting (address 01h: OSR [1:0]), and sampling time setting (address 01h: AVE\_CURCD\_COUNT [1:0]). Table 1-4 contains details regarding the fixed interval setting.

For MCIC\_R [1:0] = 2'b00, OSR [1:0] = 2'b00, and AVE\_CURCD\_COUNT [1:0] = 2'b10, ADC sampling time is 0.25 ms and the fixed interval is 0.25 ms x 64 = 16 ms.

Table 1-4. Fixed interval setting (listed by OSR [1:0] setting)

$\cap$ CD	[1·N]	2'b00(32)	or O'h1	1/221

MCIC_R [1:0]		2'b00(32)	2'b01(128)	2'b10(256)	2'b11(1024)
ADC Conversion Time [ms]		0.25	1	2	8
	Mearsurement Count	Averaging Time[ms]			
AVE_CURCD_COUNT [1:0] = 2'b00	4	1	4	8	32
AVE_CURCD_COUNT [1:0] = 2'b01	16	4	16	32	128
AVE_CURCD_COUNT [1:0] = 2'b10	64	16	64	128	512
AVE CURCD COUNT [1:0] = 2'b11	128	32	128	256	1024

(Note) "Averaging Time" = "ADC Conversion Time" x "Muasurement Count"

# OSR [1:0] = 2'b01(128)

0011[1:0] = 2 001(120)					
MCIC_R[1:0]		2'b00(32)	2'b01(128)	2'b10(256)	2'b11(1024)
ADC Conversion Time [ms]		0.25	0.25	0.5	2
	Mearsurement Count	Averaging Time[ms]			
AVE_CURCD_COUNT [1:0] = 2'b00	4	1	1	2	8
AVE_CURCD_COUNT [1:0] = 2'b01	16	4	4	8	32
AVE_CURCD_COUNT [1:0] = 2'b10	64	16	16	32	128
AVE_CURCD_COUNT [1:0] = 2'b11	128	32	32	64	256

(Note) "Averaging Time" = "ADC Conversion Time" x "Muasurement Count"

#### OSR [1:0] = 2'b10(512)

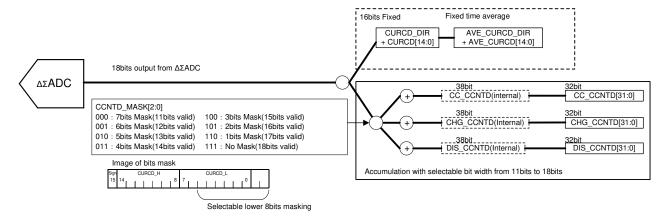
MCIC_R[1:0]		2'b00(32)	2'b01(128)	2'b10(256)	2'b11(1024)
ADC Conversion Time [ms]		0.25	0.25	0.25	0.5
	Mearsurement Count	Averaging Time[ms]			
AVE_CURCD_COUNT [1:0] = 2'b00	4	1	1	1	2
AVE_CURCD_COUNT [1:0] = 2'b01	16	4	4	4	8
AVE_CURCD_COUNT [1:0] = 2'b10	64	16	16	16	32
AVE_CURCD_COUNT [1:0] = 2'b11	128	32	32	32	64

(Note) "Averaging Time" = "ADC Conversion Time" x "Muasurement Count"

#### 1.3.4 Bit Mask Function of Accumulation Current

The current measured by BD7220FV-C is 18 bits valid data and it is stored in the 16 bits CURCD register (see 1.3.1. Current Measurement). In the default setting, 16 bits' width of current value same with CURCD register is accumulated to CC\_CCNTD resister, CHG\_CCNTD resister and DIS\_CCNTD resister, and it can select to mask lower bits of current (fixing lower bits to "0" in accumulation) by setting CCNTD\_MASK resister. With this bit mask function, it is possible to reduce the affection from noise in measuring minute current and accumulation.

Note that this function fixing lower bits to "0" in accumulation. This means current is rounded down when sign is positive, and current is rounded up when sign is negative since it is expressed by two's complement. And, complement by user's MCU is recommended, because accumulation current has constant error following to mask lower bits of current. (see next page) Note that this function affects only the accumulation from CURCD to CC\_CCNTD/CHG\_CCNTD/DIS\_CCNTD. This does not affect CURCD itself. CURCD does not have bit mask function and so always contains the full 16 bits of data.



#### Example of Internal 18bits valid data

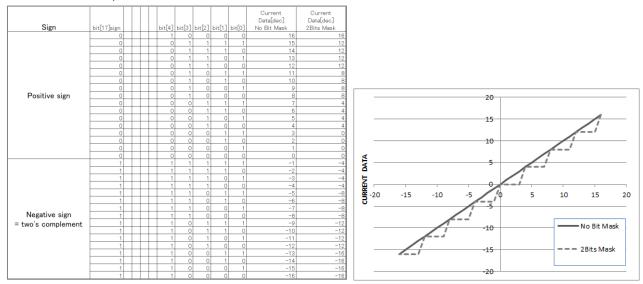


Figure 1-8. Bit Mask Function of Accumulation Current

#### 1.3.4 Bit Mask Function of Accumulation Current - continued

This is the example of CCNTD error complement with bit mask function.

# (1) Adding manual offset to CURCD

In the case that measuring minute current value is less than ±1LSB of bit mask, positive sign current data becomes zero and negative sign current data becomes –LSB with bit mask function. Adding manual offset (OFST10: address0Bh) to shift CURCD only positive sign data, current data becomes zero and it saves unintended current accumulation with measuring minute current. But it needs CCNTD error complement, because this offset is added constantly.

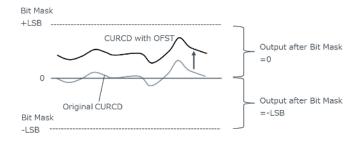


Figure 1-9. Image of Adding Manual Offset to CURCD

# (2) Adding error complement to CCNTD

Below is the formula of CCNTD error complement for (1) CURCD manual offset and for increasing lower bits of bit mask function.

Table 1-5. Formula of CCNTD Error Complement

item	Resistor	Formula of Error Complement
0	CC_CCNTD	$\{-0FST \times (TC / AD\_conversion\_term) / internal_bits_adjustment \} = \{-0FST \times 4000 / 2^6\}$
Complement of CURCD manual offset	CHG_CCNTD	$\{-0FST \times (TC / AD\_conversion\_term) / internal\_bits\_adjustment \} = \{-0FST \times 4000 / (2^6 / 2^2) \}$
	DIS_CCNTD	$\{-0FST \times (TC / AD\_conversion\_term) / internal_bits_adjustment \} = \{-0FST \times 4000 / (2^6 / 2^2) \}$
0l	CC_CCNTD	$\{2^{N-3} \times (TC / AD\_conversion\_term) / internal_bits\_adjustment \} = \{2^{N-3} \times 4000 / 2^6\}$
Complement of bit mask function	CHG_CCNTD	$\{2^{N-3} \times (TC / AD\_conversion\_term) / internal_bits\_adjustment \} = \{2^{N-3} \times 4000 / (2^6 / 2^2) \}$
Tunction	DIS_CCNTD	$\{2^{N-3} \times (TC / AD\_conversion\_term) / internal\_bits\_adjustment\} = \{2^{N-3} \times 4000 / (2^6 / 2^2)\}$

(Note) OFST: Manual Offset Value, TC: CCNTD Read Term, N: Bit Mask Width (Note) Condition: N > 2, CC UNDIV=1, CCNTD Read Term = 1 [s]

# 1.3.5 Setting of ΔΣADC Digital Filter

The  $\Delta\Sigma$ ADC digital filter can be set to one of four filter characteristics depending on the purpose. Using the wideband filter makes it possible to measure steep current changes. While this is not possible using the narrowband filter, using the narrowband filter helps suppress environmental noise.

The characteristic of the Digital Filter is changed using MCIC\_R register (address 00h MCIC\_R [1:0]). Because the conversion time depends on the MCIC\_R register setting, attention is necessary. (Refer to Table 1-6, Table 1-7)

# ADC sampling period (AD\_SAMP)

-Time to convert 1 sampling. The average current of this period is output to CURCD register, and accumulated to CC\_CCNTD register.

ADC conversion latency (AD\_LATE) consists of the following wait times.

- -Transaction time from OFF to ON in intermittent action
- -Reshuffle time for EXADIN pin input voltage measurement action
- -After calibration, time to convert initial data from  $\Delta\Sigma$ ADC startup condition

During this period, it cannot output measurement current.

Table 1-6. ADC Sampling Period [ms] (MCIC R [1:0] / OSR [1:0])

Address 00h	Address 01h OSR[1:0]				
MCIC_R[1:0]	2'b00(32) 2'b11(32)	2'b01(128)	2'b10(512)		
2'b11(1024)	8	2	0.5		
2'b10(256)	2	0.5	0.25		
2'b01(128)	1	0.25	0.25		
2'b00(32)	0.25	0.25	0.25		

Table 1-7. ADC Conversion Latency [ms] (MCIC\_R [1:0] / OSR [1:0])

		Address 01h				
Address 00h		OSR [1:0]				
MCIC_R [1:0]	2'b00(32)	0'501/100)	2'b10(512)			
	2'b11(32)	2'b01(128)				
2'b11(1024)	96.5	24.5	6.5			
2'b10(256)	24.5	6.5	2.5			
2'b01(128)	12.5	3.5	1.5			
2'b00(32)	3.5	1.5	1.5			

## 2. Operation Modes

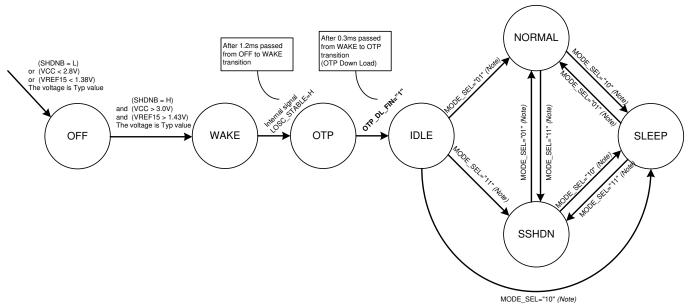
## 2.1 Features

- ·Automatic power on by supplying input voltage (VCC)
- ·Recall of OTP memory automatically in power on sequence
- ·High accuracy of calculation and accumulation of current in the Normal mode
- ·Low current consumption thanks to intermittent operation of AMP and ΔΣADC in SLEEP mode (The timing of the intermittent operation and ON time duration can be set by register)
- ·Retention of internal registers in SSHDN mode
- ·SHDNB pin which can turn all blocks off (OFF state, data in internal registers are cleared)

#### 2.2 Structure

BD7220FV-C starts to operate as soon as input power is supplied.

After the reference voltage turns on (WAKE), OTP settings (including calibration settings) are recalled. Then, BD7220FV-C becomes ready for SPI communication (IDLE). With finished recall from OTP, "1" is set to OTP\_DL\_FIN register. There are three other operating modes from the IDLE state. Using SPI commands, it is possible to freely change between these modes.



(Note) VDD > 2.5V is required to input SPI commands.

Figure 2-1. Operating Modes Transition Diagram

Operating Modes Transition Diagram and Modes Structure - continued

# NORMAL mode (MODE SEL [1:0] = "01")

AMP and  $\Delta\Sigma$ ADC are ON in this mode. The current is measured and the value is accumulated continuously with high accuracy.

## SLEEP mode (MODE SEL [1:0] = "10")

AMP,  $\Delta\Sigma$ ADC and VREF25 operate intermittently and the current consumption is reduced. The current cannot be measured during off time but the accumulation error can be reduced by accumulating a certain fixed value during off time. Using the SLEEP\_CC\_SEL register, the fixed accumulation value can be set to:

- Average current measured during ON time
- · Last measured current during ON time

This mode is valid for systems when current do not change frequently. Please refer to the Figure 2-3 for the detailed operation timing.

During intermittent operation, ON time to OFF time ratio in an intermittent period can be set through SLEEP\_INTERVAL (03h CC\_SET3 [3:2]), and the number of measurement times in ON time can be set through SLEEP\_SAMPLING\_TIME [1:0] (03h CC\_SET3 [5:4]). Each register has 4 settings. Note that as the number of measurement times increase by setting SLEEP\_SAMPLING\_TIME register, ON time is extended proportionally.

# SSHDN mode (MODE SEL [1:0] = "11")

Only VREF15, power supply for internal digital circuit, is ON. Internal register settings can be retained by VREF15. Current consumption of BD7220FV-C can be minimized in SSHDN as all blocks other than VREF15 are turned off.

Table 2-1 describes which blocks are turned ON and OFF in each operating mode.

Table 2-1. State of Blocks under each Operating Mode

MODE	MODE_SEL[1:0]	SHDNB (Pin)	VREF15	VREF25	AMP	ΔΣΑDC	OSC (8.192MHz)	SPI access
OFF ( VCC < 2.8V )	-	-	OFF	OFF	OFF	OFF	OFF	Invalid
OFF (SHDNB = L)	-	L	OFF	OFF	OFF	OFF	OFF	Invalid
WAKE (Reference Wake-up)	-	Н	ON	OFF	OFF	OFF	ON	Invalid
OTP (OTP Auto Loading)	-	Н	ON	OFF	OFF	OFF	ON	Invalid
IDLE	-	Н	ON	OFF	OFF	OFF	ON	Valid
NORMAL	2'b01	Н	ON	ON	ON	ON	ON	Valid
SLEEP	2'b10	Н	ON	Inter- mittent	Inter- mittent	Inter- mittent	ON	Valid
SSHDN (Soft Shutdown)	2'b11	Н	ON	OFF	OFF	OFF	OFF	Valid

#### 2.3 Function Description

## 2.3.1 NORMAL Mode

BD7220FV-C enters NORMAL mode when MODE\_SEL [1:0] (00h CC\_SET1 [1:0]) is set to "01". VREF25, AMP,  $\Delta\Sigma$ ADC and OSC turn on and start current measurement when in Normal mode. The actual time for current measurement from writing to the register depends on INI\_WAIT [1:0], MCIC\_R [1:0] and OSR [1:0] settings.

The INI\_WAIT register configures the time delay for VREF25 and AMP startup. The lead time for initial data conversion after  $\Delta\Sigma$ ADC turns on is set by the MCIC\_R and OSR registers. Please refer to Table 1-6.

When INI\_WAIT [1:0] = 2'b00 (1.5 ms), MCIC\_R [1:0] = 2'b00 (down sampling value = 32), and OSR [1:0] = 2'b00, the actual lead time to measure current is 5.0 ms.

To enable the function for accumulating current, "1" needs to be written to CCNTEN. The current can be measured and CURCD register is updated even when the CCNTEN register value is "0", but CC\_CCNTD, CHG\_CCNTD and DIS\_CCNTD registers are not updated.

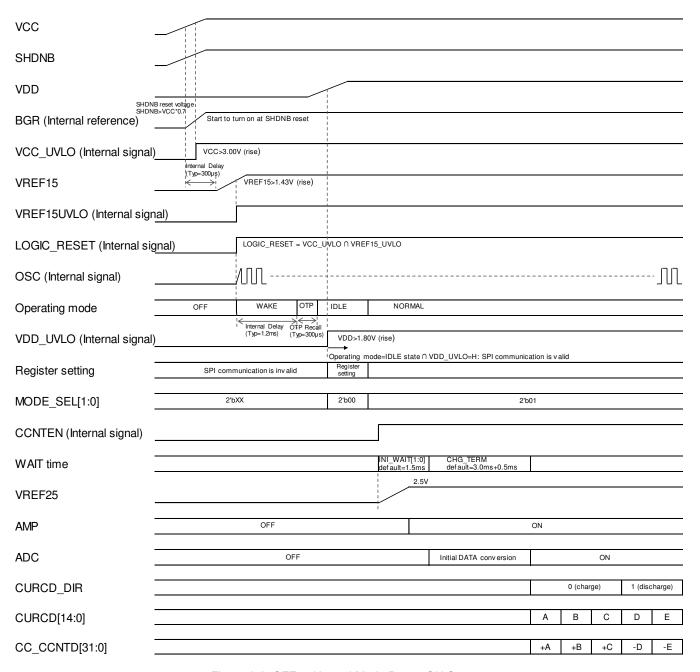


Figure 2-2. OFF to Normal Mode Power ON Sequence

#### 2.3.2 SLEEP Mode

The BD7220FV-C enters SLEEP mode when "10" is written to MODE\_SEL [1:0] (00h CC\_SET1 [1:0]). In SLEEP mode, the current consumption can be reduced since VREF25, AMP, and  $\Delta\Sigma$ ADC operate intermittently. The timing of intermittent operation can be configured using the SLEEP\_INTERVAL register. All available ratios of ON time to OFF time in intermittent operation are listed in Table 2-2.

Table 2-2. Ratio of ON Time to OFF Time in Intermittent Operation in SLEEP Mode

	ON Time	OFF Time
SLEEP_INTERVAL [1:0]=2'b00	1	7
SLEEP_INTERVAL [1:0]=2'b01	1	15
SLEEP_INTERVAL [1:0]=2'b10	1	31
SLEEP_INTERVAL [1:0]=2'b11	1	127

\*Ratio when ON time is taken as '1'

The ON time during intermittent operation is calculated by the following formula:

ON time = INI WAIT + AD LATE + (ADC SAMP 
$$\times$$
 SLEEP SAMPLING TIME)

# INI WAIT [1:0] (38h):

This is the wait time for VREF25 and AMP startup. The wait time is configurable from 1.5 ms (default) to 12 ms.

#### AD LATE (ADC conversion latency):

This is the wait time before outputting digital conversion value after a signal is input into the ADC. It is configured by the digital filter (MCIC\_R) and OSR settings. For details, please refer to Table 1-6.

## AD SAMP (ADC sampling period):

This is the time required for one AD conversion. It is configured by the digital filter (MCIC\_R) and OSR settings. For details, please refer to Table 1-5.

#### SLEEP SAMPLING TIME [1:0] (03h):

This register determines how many times current measurement is done during ON time of intermittent operation. The number is configured by the digital filter (MCIC R) and OSR settings. For details, please refer to Table 2-3.

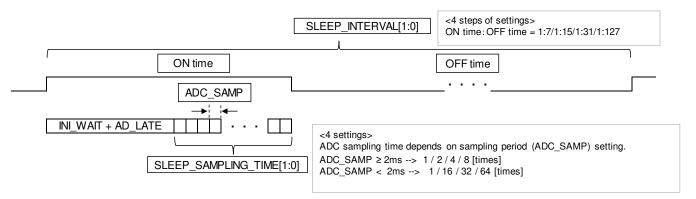


Figure 2-3. Intermittent Operation in SLEEP Mode

ON time and OFF time in default settings are calculated as following;

*ON time* = 1.5 [ms] + 3.5 [ms] + 
$$(250 \text{ [µs]} \times 1 \text{ } time) = 5.25 \text{ [ms]}$$
  
*OFF time* = 5.25 [ms] × 7 times = 36.75 [ms]

SLEEP Mode - continued

Table 2-3. Measurement Count during Intermittent Operation (ON Time) in SLEEP Mode

MCIC_R[1:0]	2'b00(32)	2'b01(128)	2'b10(256)	2'b11(1024)
ADC Conversion Time[ms]	0.25	1	2	8
		Measurement	Count [times]	
SLEEP_SAMPLING_TIME[1:0] = 2'b00	1	1	1	1
SLEEP_SAMPLING_TIME[1:0] = 2'b01	16	16	2	2
SLEEP_SAMPLING_TIME[1:0] = 2'b10	32	32	4	4
SLEEP_SAMPLING_TIME[1:0] = 2'b11	64	64	8	8

## OSR[1:0] = 2'b01(128)

MCIC_R[1:0]	2'b00(32)	2'b01(128)	2'b10(256)	2'b11(1024)
ADC Conversion Time[ms]	0.25 0.25		0.5	2
		Measurement	Count [times]	
SLEEP_SAMPLING_TIME[1:0] = 2'b00	1	1	1	1
SLEEP_SAMPLING_TIME[1:0] = 2'b01	16	16	16	2
SLEEP_SAMPLING_TIME[1:0] = 2'b10	32	32	32	4
SLEEP SAMPLING TIME[1:0] = 2'b11	64	64	64	8

#### OSR[1:0] = 2'b10(512)

MCIC_R[1:0]	2'b00(32)	2'b01(128)	2'b10(256)	2'b11(1024)
ADC Conversion Time[ms]	0.25	0.25	0.25	0.5
		Measurement	Count [times]	
SLEEP_SAMPLING_TIME[1:0] = 2'b00	1	1	1	1
SLEEP_SAMPLING_TIME[1:0] = 2'b01	16	16	16	16
SLEEP_SAMPLING_TIME[1:0] = 2'b10	32	32	32	32
SLEEP_SAMPLING_TIME[1:0] = 2'b11	64	64	64	64

In SLEEP mode, current measurement is enabled only during ON time. Current is never measured during OFF time (VREF25, APM, and  $\Delta\Sigma$ ADC are OFF) but a fixed current value configured by the SLEEP\_CC\_SEL register is accumulated. The values of related registers are updated as below.

#### CURCD:

The current value measured just before turning off is retained. The register value is not updated.

# CC\_CCNTD, CHG\_CCNTD, DIS\_CCNTD:

The current value measured in ON time is accumulated during OFF time.

Average value of current measured during ON time or the last value of ON time can be selected as the current value to be accumulated, using the SLEEP\_CC\_SEL register (03h CC\_SET3 [6]).

# SLEEP\_CC\_SEL (03h):

- 0: The last measured current during the previous ON time is accumulated during OFF time.
- 1: Average current measured during ON time is accumulated during OFF time.

Please don't change the configuration of the registers related to ON time / OFF time setting (INI\_WAIT, MCIC\_R, OSR, SLEEP\_SAMPLING\_TIME, SLEEP\_INTERVAL) and SLEEP\_CC\_SEL in SLEEP mode operating.

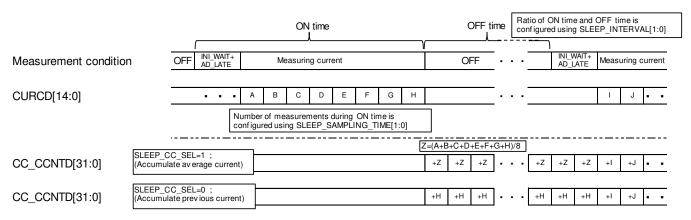


Figure 2-4. Accumulation Current during OFF Time

#### 2.3.3 SSHDN Mode

BD7220FV-C enters SSHDN mode when "11" is written to MODE\_SEL [1:0] (00h CC\_SET1 [1:0]).

In SSHDN mode, the minimum function blocks to retain register settings are turned on. For details, please refer to Table 2-1. The CURCD register which shows current value is reset to "0" but the CCNTD register which shows accumulation current keeps the value measured at the end.

# 2.3.4 OFF State

BD7220FV-C is turned off when the SHDNB pin is in "L" state.

In OFF state, all blocks are turned off and the current consumption is at minimum. Note that register settings are cleared and OTP settings are recalled at next power on. The register value of calibration needs to be written again. (Refer to section 4.2.1.)

Finished to recall from OTP at restart from OFF state, "1" is set to OTP\_DL\_FIN register. The data set to OTP\_DL\_FIN register is latched and cleared by writing "1" to it. So, to clear OTP\_DL\_FIN register after startup and monitoring the register value regularly enables to detect unexpected reset of BD7220FV-C.

#### 3. Interrupts

#### 3.1 Summary

6 types of interrupt can be generated through the open-drain INTB pin. All interrupt settings can be masked or unmasked through the register settings.

#### 3.1.1 Features

- 4 types of threshold can be configured for interrupt by the accumulation current [CC CCNTD]
- 3 types of threshold can be configured for interrupt by the measured current [CURCD]
- · Interrupt by the current the configured threshold or more
- Interrupt by detecting battery relaxation
- · Interrupt by SPI CRC error detection
- · Interrupt by completing the calibration

# 3.1.2 Structure

BD7220FV-C can generate interrupt signals from multiple sources by asserting the INTB pin low. Each interrupt source has associated status register and enable register. The status register indicates the status of each interrupt source and the enable register allows to output the interruption to the external open drain pin "INTB".

Regardless of the setting of its enable register, status register corresponding to the source of interrupt is set "1" when detecting each interrupt source event. The status register is latched and is not cleared automatically even if released from the interrupt source. To clear the enable register, write "1" to the register. Only the enabled interrupt event can assert the INTB pin low and announcing the interrupt occurred.

The INTB pin is kept asserted low until all enabled status registers are cleared when multiple interrupt sources occur. The INTB pin goes into Hi-z state when all enabled status registers are cleared. As "0" written in status registers is ignored, write "0" in other bits to clear only one interrupt source.

After power on or restart from OFF state, all status registers are "0" (non-detected) and all enable registers are "0" (interrupt assertion to the INTB pin is disabled). To enable interrupt assertion, configure the interrupt function by SPI.

Please refer to the Table 3-1 for more details about interrupt registers.

# 3.2 Register Descriptions

The list of interrupt registers are as follows.

Table 3-1. List of Interrupt Registers

Register			Register Map					
Name Bit Na	Bit Name	Function	Ena	Status/Clear				
			Address	bit	Address	bit		
NT_REQ1	CC_MON1_DET	Detection of Charging Current Accumulation Value1 (Threshold Setting)	39h	0	3Ch	0		
NT_REQ1	CC_MON1_RES	Detection of Discharging Current Accumulation Value1 (Threshold Setting)	39h	1	3Ch	1		
NT_REQ1	CC_MON2_DET	Detection of Charging Current Accumulation Value2 (Threshold Setting)	39h	2	3Ch	2		
NT_REQ1	CC_MON2_RES	Detection of Discharging Current Accumulation Value2 (Threshold Setting)	39h	3	3Ch	3		
NT_REQ1	CC_MON3_DET	Detection of Charging Current Accumulation Value3 (Threshold Setting)	39h	4	3Ch	4		
NT_REQ1	CC_MON3_RES	Detection of Discharging Current Accumulation Value3 (Threshold Setting)	39h	5	3Ch	5		
NT_REQ1	CC_MON4_DET	Detection of Charging Current Accumulation Value4 (Threshold Setting)	39h	6	3Ch	6		
NT_REQ1	CC_MON4_RES	Detection of Discharging Current Accumulation Value4 (Threshold Setting)	39h	7	3Ch	7		
NT_REQ2	ALARM_OCUR1_DET	Alarm Output for Detection of Charging Current (ALARMB Terminal Output for OCUR1_DET)	3Ah	0	3Dh	2		
NT_REQ2	ALARM_OCUR1_RES	Alarm Output for Detection of Discharging Current (ALARMB Terminal Output for OCUR1_RES)	3Ah	1	3Dh	3		
NT_REQ2	OCUR1_DET	Detection of Charging Current1 (Threshold/Number of Detection Setting)	3Ah	2	3Dh	2		
NT_REQ2	OCUR1_RES	Detection of Discharging Current1 (Threshold/Number of Detection Setting)	3Ah	3	3Dh	3		
NT_REQ2	OCUR2_DET	Detection of Charging Current2 (Threshold/Number of Detection Setting)	3Ah	4	3Dh	4		
NT_REQ2	OCUR2_RES	Detection of Discharging Current2 (Threshold/Number of Detection Setting)	3Ah	5	3Dh	5		
NT_REQ2	OCUR3_DET	Detection of Charging Current3 (Threshold/Number of Detection Setting)	3Ah	6	3Dh	6		
NT_REQ2	OCUR3_RES	Detection of Discharging Current3 (Threshold/Number of Detection Setting)	3Ah	7	3Dh	7		
NT_REQ3	WAKE_DET	Detection of Over Wake-Up Current (Threshold/Number of Detection Setting)	3Bh	0	3Eh	0		
NT_REQ3	WAKE_RES	Detection of Under Wake-Up Current (Threshold/Number of Detection Setting)	3Bh	1	3Eh	1		
NT_REQ3	REX_DET	Detection of Relax State (Threshold/Time of Detection Setting)	3Bh	2	3Eh	2		
NT_REQ3	CRCERR_DET	Detection of CRC Error	3Bh	4	3Eh	4		
NT_REQ3	CALIB_FIN	Detection of Calibration Finish	3Bh	6	3Eh	6		

(Note) Inside () indicate configurable parameters

## 3.3 Function Description

## 3.3.1 Interrupt by Accumulation Current

Interrupt can be generated by detecting the configured threshold crossing to the accumulation current measurement (CC\_CCNTD). 4 thresholds can be configured through CC\_BATCAP#\_TH (# = 1 to 4). CC\_BATCAP#\_TH are 24-bit registers those are compared to the upper 24 bits (out of 32) of CC\_CCNTD.

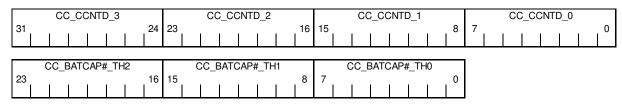


Figure 3-1. Relationship between CC\_CCNTD and CC\_BAT\_CAP\_TH

(When CC\_UNDIV = 0: scaling in every gain setting is enabled) LSB of CC\_BATCAP1\_TH =  $78.125 \mu$ Ah MSB of CC\_BATCAP1\_TH = 655.36 Ah

When CC\_UNDIV=1: scaling in every gain setting is disabled and in that case LSB / MSB varies. For example, the threshold of the accumulation current is configured to 1Ah, register value is as below.

$$1[Ah] / 78.125[\mu Ah] = 12800 => CC\_BATCAP1\_TH[23:0] = 24'h003200(24'd12800)$$

Please refer to section 1.3.2 for more details about CC\_CCNTD.

Table 3-2. The List of Accumulation Current and Threshold Registers

Address	Register Name	Description
17h to 1Ah	CC_CCNTD [31:0]	Accumulated Current Value
23h to 25h	CC BATCAP1 TH[23:0]	Interrupt Threshold for Current Accumulation 1
2311 (0 2311	CC_BATCAI 1_111[25:0]	It is compared with the upper 24 bits [31:8] in CC_CCNTD [31:0] register.
26h to 28h	CC BATCAP2 TH[23:0]	Interrupt Threshold for Current Accumulation 2
2611 (0 2611   CC_BATCAP2_1H [23		It is compared with the upper 24 bits [31:8] in CC_CCNTD [31:0] register.
29h to 2Bh	CC BATCAP3 TH[23:0]	Interrupt Threshold for Current Accumulation 3
2911 to 2611 CC_6ATCAP3_TH [23.0]		It is compared with the upper 24 bits [31:8] in CC_CCNTD [31:0] register.
2Ch to 2Eh	CC BATCARA THIOSO	Interrupt Threshold for Current Accumulation 4
ZCII (0 ZEII	CC_BATCAP4_TH [23:0]	It is compared with the upper 24 bits [31:8] in CC_CCNTD [31:0] register.

# 3.3.2 Interrupt for Current Measurement

Interrupt can be generated by detecting the configured threshold crossing for current measurement (CURCD). 3 thresholds can be configured through OCURTHR#\_DIR (# = 1 to 3) and OCURTHR# (# = 1 to 3). OCURTHR# is a 15-bit register that is compared to the 15 bits of CURCD. OCURDUR# (# = 1 to 3) configures the number of consecutive detections required to generate the interrupt. For example, if set to 4 times, an interrupt is generated only after 4 consecutive detections. The counter starts over from 0 if enough time passes without reaching the set number of consecutive detections.

Only specific to OCURTHR1, interrupts can also be notified through the dedicated ALARMB pin. Unlike interrupt notifications through INTB, the ALARMB interrupt source can be easily identified even without checking the status registers.

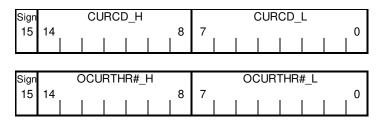


Figure 3-2. Relationship of CURCD and OCURTHR

Table 3-3. List of Current Detection Interrupt Setting Registers

Address	Register Name	Description
13h	CURCD_DIR	Current Direction (0:charging 1:discharging)
13h to 14h	CURCD [14:0]	Measured Current Value
2Fh	OCURTHR1_DIR	OCURTHR1 [14:0] Current Direction(0:charging 1:discharging)
2Fh to 30h	OCURTHR1 [14:0]	Interrupt Threshold for Current Measurment 1
31h	OCURTHR2_DIR	OCURTHR2 [14:0] Current Direction(0:charging 1:discharging)
31h to 32h	OCURTHR2 [14:0]	Interrupt Threshold for Current Measurment 2
33h	OCURTHR3_DIR	OCURTHR3 [14:0] Current Direction(0:charging 1:discharging)
33h to 34h	OCURTHR3 [14:0]	Interrupt Threshold for Current Measurment 3
35h	OCURDUR1 [1:0]	Crossing Detection Threshold for OCURTHR1 (00: 1 time, 01: 4 times, 10: 8 times, 11: 16 times)
35h	OCURDUR2 [1:0]	Crossing Detection Threshold for OCURTHR2 (00: 1 time, 01: 4 times, 10: 8 times, 11: 16 times)
35h	OCURDUR3 [1:0]	Crossing Detection Threshold for OCURTHR3 (00: 1 time, 01: 4 times, 10: 8 times, 11: 16 times)

#### 3.3.3 Interrupt by Battery Relaxation and Wakeup Current Detection

Relaxation state Interrupt can be generated when a relaxation state of the battery is detected. The detection threshold can be configured through REX\_CURCD\_TH. REX\_CURCD\_TH is an 8-bit register and set to the lower 8 bits (out of 15) of CURCD. REX\_CURCD\_TH cannot be set the CURCD\_DIR register indicate the sign of the current value. A Relax Timer starts counting up when REX\_CURCD\_TH is set to 100 mA and the value of CURCD is within -100 mA to +100 mA. The counter resets once the current increases higher than the threshold. REX\_EN should be set to "1" to enable the Relax Timer. Writing "0" to the REX\_EN stops the timer and resets the timer counter to "0". Once the Relax Timer is expired and the counter needs to be reset, writing REX\_EN =  $1 \rightarrow 0 \rightarrow 1$  is required. Relax timer detect time can be configured in 4 steps from 30 min to 120 min through the REX\_DUR register. Relaxation state interrupt is generated when relax timer is over the detect time.

Wakeup current interrupt is generated when an occurrence of charge or discharge current toward the battery is detected. The detection threshold can be configured through WAKE\_CURCD\_TH an 8-bit register which is set to the lower 8 bits (out of 15) of CURCD. WAKE\_CURCD\_TH cannot be set the CURCD\_DIR register indicate the sign of the current value. An interrupt is generated when WAKE\_CURCD\_TH is set to 100 mA and the value of CURCD is less than -100 mA or greater than +100 mA. The WAKE\_COUNT register configures the number of consecutive detections required to generate the interrupt. For example, if set to 4 times, an interrupt is generated only after 4 consecutive detections are detected. The counter starts over from 0 if enough time passes without reaching the set number of consecutive detections.

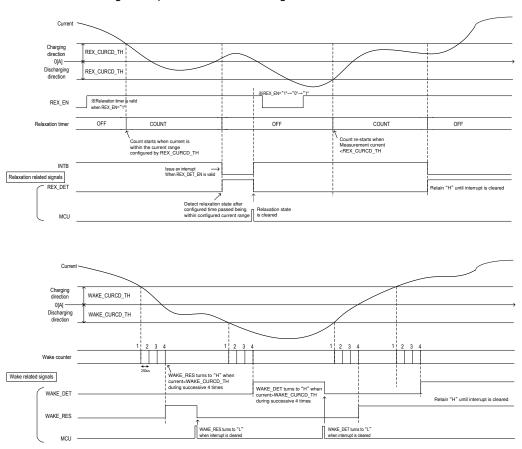


Figure 3-3. Relaxation State and Wake Up Current Detection Interrupt Timing Chart

Table 3-4. List of Relaxation State / Wake Up Interrupt Setting Registers

Address	Register Name	Description					
13h to 14h	CURCD [14:0]	Measured Current Value					
03h	REX_EN	Relax State Detection Timer Enable (0 : Timer OFF (0 Clear) 1 : Timer ON)					
35h	REX_DUR [1:0]	Relax State Detection Time (00 : 30 minutes 01 : 60 minutes 10 : 90 minutes 11 : 120 minutes)					
36h	REX_CURCD_TH [7:0]	Relaxation State Detection Threshold It is compared with the lower 8bits[7:0] of the CURCD [14:0] register.					
37h	WAKE_CURCD_TH [7:0] Wake up Detection Threshold It is compared with the lower 8bits[7:0] of the CURCD [14:0] register.						
38h	WAKE_COUNT [1:0]	Number of Detection times for Wake up (00 : 1 time 01 : 4 times 10 : 8 times 11 : 16 times)					

# 3.3.4 Interrupt for Detection of CRC Error

The BD7220FV-C incorporates SPI interface.

It is possible to add CRC code to the transmission and reception of SPI commands. Having CRC code or not is selectable by the leading EC bit of the data. For details, please refer to the section of 5. CRC error can trigger interrupt.

#### In case of Data write:

Please add a CRC bit to the command when it is transmitted from the MCU.

When CRC error is detected at the time of the command reception, a CRC error triggers interrupt.

Please write "1" to CRCERR\_DET\_EN register to produce an interrupt signal through INTB.

#### In case of Data read:

CRC bits cannot be added to a read command from the MCU.

Please judge the CRC error on the MCU side as CRC bit is added to all transmitted and received data as a result of BD7220FV-C sending "read data" and "read command".

#### 3.3.5 Interrupt for Completion of Calibration

The BD7220FV-C has two calibration modes. (For details, please refer to section 4.1.2.)

When calibration is completed with mode 2, interrupt is triggered. Please write "1" to CALIB\_FIN\_EN register to produce an interrupt signal through INTB.

## Customer's product shipment:

It is assumed that the calibration is executed with the condition where the BD7220FV-C and external components are mounted in the process. Dispersion of external components can be considered when the calibration is executed with a mounted external current detection device such as a shunt resistor or current sensor.

#### 4. Calibration

## 4.1 Summary

By communicating with SPI, BD7220FV-C can calibrate the variation of gain and offset. There are an auto calibration by SPI communication and a manual calibration by setting the calibration value into the registers. Because BD7220FV-C has a sequence circuit and a calculation circuit for calibration, it can output a calibration value automatically only by setting the input information and inputting a trigger signal. However, non-volatile memory is not built-in for this product. To be able to store values generated from automatic and manual calibration, it is necessary to use and maintain external memory. Figure 4-1 shows the flow chart of calibration.

#### 4.1.1 Features

·Calibration in BD7220FV-C shipment process:

The calibration values are kept in the built-in OTP and are downloaded automatically

·Calibration in customer's product shipment:

The calibration (gain and offset calibration) including the external current detection element in customer's process

·Manual calibration enabled to set the calibration value externally

## 4.1.2 Structure

BD7220FV-C can calibrate gain and offset error.

The three available calibration modes are explained below.

# 4.1.2.1 Calibration Mode 1 (in BD7220FV-C shipment process)

This calibration is carried out at the shipment of BD7220FV-C. Calibration is done with the internal regulator VREFCAL for calibration as a reference. An external capacitor at VREFCAL pin is not needed on customer's product, so please make the pin open. The external current detection element (shunt resistor or current sensor) is not calibrated. Only the gain and offset error of the built in AMP and  $\Delta\Sigma$ ADC are calibrated. The calibration values are kept in the built in OTP in BD7220FV-C, and are read back automatically at startup. Offset calibration is tuned to the 25 V/V AMP gain setting by default, so it is necessary to use Mode 2 and Mode 3 for 5 V/V and 51 V/V settings.

#### 4.1.2.2 Calibration Mode 2 (in customer's product process)

Calibration mode 2 is the calibration used in the customer's product shipment. This calibration includes the external current detection element.

It is possible to measure a highly precise current by carrying out the calibration when the external current detection element (shunt resistor or current sensor) is connected. To carry out this calibration, it is necessary to force as much current as will be used in the actual application. Mode 2 or Mode 3 offset calibration is necessary when using an AMP gain of 5 V/V or 51 V/V. If measurement will be performed using multiple AMP gain settings, Mode 2 calibration needs to be performed on each setting to be used.

BD7220FV-C outputs the calibration value after automatic operation. However, as BD7220FV-C does not have non-volatile memory, external memory is required to store the calibration values used. These values should be rewritten each time BD7220FV-C is reset.

# 4.1.2.3 Calibration Mode 3 (manual calibration)

This calibration is carried out manually by calculating the calibration value using MCUs or other methods.

Mode 3 can be used to compensate gain or offset variations caused by the measurement environment, such as the temperature. The value to be set for gain calibration is calculated by following the equation written later. The value to set for offset calibration is calculated referring to Table 1-1. Mode 2 or Mode 3 offset calibration is needed when using the 5 V/V and 51 V/V AMP gain settings, or for each gain setting used when measuring using multiple gain settings. However, as BD7220FV-C does not have non-volatile memory, external memory is required to store the calibration values used. These values should be rewritten each time BD7220FV-C is reset.

Calibration Mode 3 (manual calibration) - continued

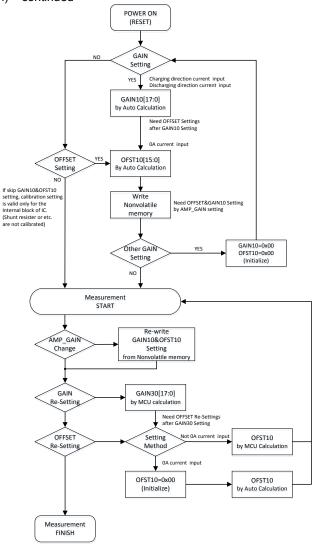


Figure 4-1. Flow Chart of Calibration

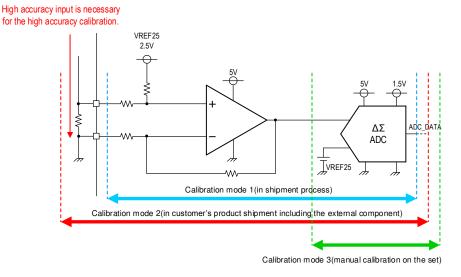


Figure 4-2. Calibration (Mode 1, Mode 2 and Mode 3)

#### 4.2 Function Description

## 4.2.1 Calibration Mode 2

At customer's shipment, gain and offset error can be calibrated using Calibration Mode 2.

The calibration is carried out while forcing a certain differential voltage across the INP and INN pins. Calibrating with an external current detection element (such as a shunt resistor) enables higher accuracy current measurement. Note that in Mode 2, Gain Calibration must be performed before Offset Calibration. The calibration values in the register are discarded when BD7220FV-C is reset, so please write these to external memory and rewrite when BD7220FV-C starts up.

#### 4.2.1.1 Calibration Mode2: Gain Calibration

Gain calibration is carried out by forcing voltage that correspond to 2 points of current value, one of charging current, PFS (current direction of INP voltage is positive); the other of discharging current, MFS (current direction of INP voltage is negative). Please set the data of differential voltage between PFS and MFS into the CALVIN\_DIFF register. Please refer to Figure 4-3 for the relationship between PFS, MFS and CALVIN\_DIFF registers. The current values in the figure are a 25 V/V AMP gain setting and  $R_{\text{SNS}}$  of 0.2 m $\Omega$ .

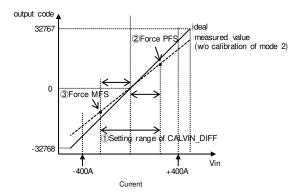


Figure 4-3. Force Setting for Gain Calibration

The flow of calibration is as follows.

- 1. Setting CALVIN\_DIFF register (Data of differential voltage between PFS and MFS)
- 2. Forcing charging current PFS and executing calibration (writing "1" to CALIB\_TRG register)
- 3. Forcing discharging current MFS and executing calibration (writing "1" to CALIB\_TRG register)

#### Notice 1

Please carry out the calibration for each gain if you will measure with changing amp gain settings. The value set to GAIN10 register is needed with each amp gain setting.

#### Notice 2

Calibrating discharge current before charge current results in a wrong calibration value, so the order of calibrating charge current before discharge current must be strictly followed. If the calibration order was not followed, please reset BD7220FV-C.

The data of differential input voltage is set in CALVIN\_DIFF [11:0] register.

The value to set is determined by the value of the external shunt resistor R<sub>SNS</sub> and the amp gain setting as in the following equation. The differential current between PFS and MFS is over the possible setting range when calculated result is a negative value.

$$\begin{array}{ll} \textit{CALVIN\_DIFF value} &= \left\{ 2^{13} \times \frac{(\Delta I \times R_{SNS} \times \textit{AMP\_GAIN})}{4.5} \right\} - \ 2^{12} \\ R_{SNS} : \textit{resistance of shunt } [\Omega] \end{array}$$

Please refer to Table 4-1 for the possible setting range of CALVIN\_DIFF for each amp gain setting. For an AMP gain of 25 V/V and  $R_{SNS}$  of 0.2  $m\Omega$ , the range of CALVIN\_DIFF is from 450 A to 800 A and the step is 109.86 mA. Please force 225 A for PFS and -225 A for MFS when CALVIN\_DIFF is 450 A.

Calibration Mode2: Gain Calibration - continued

The minimum setting of differential current  $\Delta I_{MIN}$  between PFS and MFS is changed by the value of R<sub>SNS</sub> and gain setting. It is calculated with the following equation.

$$\Delta I_{MIN} = 2.25 / (AMP\_GAIN \times R_{SNS})$$
  
 $R_{SNS}$ : resistance of shunt  $[\Omega]$ 

Table 4-1. Unit of CALVIN DIFF Register Setting Current

	AMP_GAIN = 5 times			AMP_GAIN = 25 times			AMP_GAIN = 51 times					
	AMP Input Differential Voltage	UNIT	AMP Input Differential Current $(R_{SNS} = 0.2 \text{ m}\Omega)$	UNIT	AMP Input Differential Voltage	UNIT	AMP Input Differential Current $(R_{SNS} = 0.2 \text{ m}\Omega)$	UNIT	AMP Input Differential Voltage	UNIT	AMP Input Differential Current $(R_{SNS} = 0.2 \text{ m}\Omega)$	UNIT
Minimum Setting	450.00	mV	2250.00	Α	90.00	mV	450.00	Α	44.12	mV	220.59	Α
Maximum Setting	599.96	mV	2999.82	Α	160.00	mV	800.02	Α	80.01	mV	400.03	Α
Step Width	109.86	μV	549.32	mA	21.97	μV	109.86	mA	10.77	μV	53.85	mA

For example, considering an amp gain setting is 25 V/V, PFS = +300 A and MFS = -300 A, the differential current between PFS and MFS is 600 A. The differential current is within the possible setting range from 450 A to 800 A in Table 4-1, so it is possible to carry out calibration. If the differential current is outside the range, adjustment of input current is needed. The value to set to CALVIN\_DIFF register is calculated as following.

$$CALVIN\_DIFF \ value = \left\{2^{13} \times \frac{(\Delta I \times R_{SNS} \times AMP\_GAIN)}{4.5}\right\} - 2^{12}$$

 $R_{SNS}$ : resistance of shunt[ $\Omega$ ]

$$\begin{array}{lll} \textit{CALVIN\_DIFF} &= \{2^{13} \times (600 \text{[A]} \times 0.2 \times 10^{-3} \times 25) \ / \ 4.5\} \ - \ 2^{12} &= \ 1365 \text{[DEC]} \\ &= \ 555 \text{[HEX]} \end{array}$$

Table 4-2 shows the flow of SPI communication in carrying out Gain Calibration.

After setting the register configurations and input voltage between INP and INN, you can carry out the calibration by writing "1" to CALIB\_TRG register. After writing "1" to CALIB\_TRG, the value will automatically clear to "0".

To complete gain calibration, 2 times trigger input for charge and discharge is needed.

Table 4-2. Calibration Mode 2: Flow of SPI Communication in Carrying Out Gain Calibration

				SI	PI		
ltem	Register	Description	R/W	REG ADDR	CONT	DATA	INP-INN Input
	CALVIN DIFF H = XXXX	Configuration of Diffrential Input Between	W	0Fh	1Eh	0Xh	
	CALVIN_DIFF_L = XXXX	INP and INN	W	10h	20h	XXh	
Initial Configuration	AMP_GAIN = XX MCIC_R = 11	Configure AMP_GAIN as Actual Use (For Higher Accurate Calibration, Configure MCIC_R Resister 2'b11 Regardless of Actual Use)	w	00h	00h	X0h	-
Charge Configuration	GAIN_CAL_FS = 1, CALIB_MODE = 001	Forcing Charge Current Configure CALIB_MODE	w	04h	08h	09h	Input Differential Voltage of Charge
Charge Calibration	CALIB_TRG = 1	Trigger Input to Start Calibration	W	02h	04h	20h	
Discharge Configuration	GAIN_CAL_FS = 0, CALIB_MODE = 001	Forcing Discharge Current Configure CALIB_MODE	w	04h	08h	01h	Input Differential Voltage of Discharge
Discharge Calibration	CALIB_TRG = 1	Trigger Input to Start Calibration	W	02h	04h	20h	
Read Calibration Output in GAIN10 Resister	read GAIN10_2		R	05h	0Bh	-	
	read GAIN10_1	Read Calibration Output		06h	0Dh	-	-
1 (63)3(6)	read GAIN10_0_GAIN30_2		R	07h	0Fh	-	

REG ADDR: Addresses in register map of BD7220FV-C CONT ADDR: The first 1 byte data in SPI communication

Ex.1) Write REG ADDR = 0Fh  $\rightarrow$  8'b0000 1111  $\rightarrow$  0(EC) + 6b'001111 + 0(Write) = 1Eh (EC Bit = 0, CRC is invalid) Ex.2) Read REG ADDR = 02h  $\rightarrow$  8'b0000 0010  $\rightarrow$  0(EC) + 6'b000010 + 1(Read) = 05h (EC Bit = 0, CRC is invalid)

\*Calibration finishes (Detection of CALIB FIN interrupt)

You will know when calibration is finished by using the CALIB\_FIN register for interrupt. CALIB\_FIN turns to "1" when Mode 2 Calibration is completed, so please clear the CALIB\_FIN register before carrying out calibration. To validate the interrupt of INTB pin, please write CALIB\_FIN\_EN register "1".

If you need to perform Mode 2 Gain Calibration again, please do so after resetting GAIN10, OFST10 and GAIN30 registers to "0".

#### 4.2.1.2 Calibration Mode2: Offset Calibration

Offset Calibration is carried out by input differential voltage without drawing current between the INP and INN pins. Write "1" to CALIB\_TRG under forcing differential voltage and Offset Calibration is automatically carried out. The flow of calibration is only one step:

1. After input differential voltage without drawing current between the INP and INN pins, and carry out calibration (Write "1" to CALIB\_TRG)

#### Notice 1

Please carry out the calibration for each gain if you will measure with multiple amp gain settings.

Table 4-3 shows the flow of SPI communication in carrying out Offset Calibration.

Table 4-3. Calibration Mode 2: Flow of SPI Communication in Carrying Out Offset Calibration

	Register	Description	SPI				
Item			R/W	REG	CONT	DATA	INP-INN Input
				ADDR	ADDR		
	AMP_GAIN = XX	Configure AMP_GAIN as Actual Use (For Higher Accurate Calibration, Configure MCIC_R Resister 2'b11 Regardless of Actual Use)	w	00h	00h	X0h	Input Differntial Voltage at No Current Flow (INP and INN Shorted)
	CALIB_MODE = 010	Configure CALIB_MODE	W	04h	08h	02h	
Offset Calibration	CALIB_TRG = 1	Trigger Input to Start Calibration	W	02h	04h	20h	
Read Calibration Output in OFST10	read OFST10_H	Read Caliration Output	R	0Ah	15h	-	-
	read OFST10 L		R	0Bh	17h	-	

<sup>\*</sup> The way of to check if calibration is finished is the same as in Gain Calibration.

If you need to perform Mode 2 Offset Calibration again, please do so after resetting OFST10 to "0".

#### 4.2.2 Calibration Mode 3: Manual Calibration

Under typical usage conditions, gain and offset error can be calibrated using Calibration Mode 3.

#### 4.2.2.1 Calibration Mode 3: Manual Calibration

Manual gain calibration uses the GAIN30 and GAIN31 registers. GAIN30 register is for configuration of the calibration value. Writing in the value calculated from an equation written later, gain is calibrated. GAIN31 register is a read-only register for reading the gain calibration configuration contained in the IC. When in Calibration Mode 2, the IC automatically calculates the needed gain adjustment ratio and sets the calibration value, but in Mode 3, you can calibrate by setting the configuration of gain adjustment ratio from 0.8x to 1.2x by 0.006% step. Gain adjustment ratio is calculated with the forced current value of 2 points, IP and IM, the corresponds output value of CURCD, PD and MD, the value of the external shunt resistor R<sub>SNS</sub> and the amp gain setting in the following equation. Any combinations of current polarity are possible to be used, and not restricted to the combination of charge and discharge.

$$Gain\ adjustment\ ratio\ = \frac{(IP-IM)\times R_{SNS}\times AMP\_GAIN\times (2^{15}-1)}{(PD-MD)\times 2.25}$$

 $R_{SNS}$ : resistance of shunt[ $\Omega$ ]

For example, considering an amp gain setting is x25 V/V, external shut resister value is 0.2 mΩ, forced current IP=+100 A and IM=-250 A, values of CURCD PD=7645 and MD=-19113, gain adjustment ratio is calculated as following.

Gain adjustment ratio = 
$$\frac{\{100 - (-250)\} \times 0.2 \times 10^{-3} \times 25 \times (2^{15} - 1)}{\{7654 - (-19113)\} \times 2.25} = 0.95244$$

The value to set in GAIN30 register is calculated with the value of GAIN31 and the ratio of gain adjustment in the following equation. If the adjustment ratio is less than 1x, set the negative value as a 2's complement representation.

$$GAIN30 \ value = \{GAIN31[DEC] \times (Gain \ adjustment \ ratio)\} - GAIN31[DEC]$$

For example, when GAIN31 register is 0xBD2C (48428[DEC]) and Gain Adjustment Ratio is 1.05x,

$$GAIN30 \ value = \{48428[DEC] \times 1.05\} - 48428 = 2421[DEC] = 975[HEX]$$

And in case of the adjustment ratio is less than 1x as above example, calculation is as following.

$$GAIN30 = \{48428 \times 0.95244\} - 48428 = -2303[DEC] = 3F701[HEX]$$

#### Notice 1

Please write registers for manual gain calibration in IDLE state or SSHDN mode, and do not write when measuring current.

# Notice 2

If you operate Mode 3 Gain Calibration again, please re-calculate the value to set after resetting GAIN30 register to "0" and reading the value of GAIN31 register.

#### Notice 3

The value of GAIN31 register changes depending on the configuration of amp gain. So, if you will measure with multiple AMP gain configurations, the value of GAIN30 register must be calculated for each gain configuration. Please read GAIN31 register again and re-calculate the value after resetting GAIN30 register when gain configuration is changed. Notice 4

The value of GAIN31 register changes depending on the calibration value obtained from Mode 2 Calibration (and stored in GAIN10 register.) If Mode 2 Calibration is to be performed again, please read GAIN31 register and re-calculate the value after resetting GAIN30 register.

#### 4.2.2.2 Calibration Mode 3: Manual Offset Calibration

Manual Offset Calibration uses the OFST10 register. Writing OFST10 with the value added or subtracted corresponding to the amount of current you want to adjust, will calibrate the offset. To convert from current to adjust to register value, please refer to LSB current in Table 1-1. The value to set to OFST10 register is calculated with the original value of OFST10 register and adjustment current as in the following equation.

In the case that the original value of OFST10 register is 16'h012C (=16'd300), the external current sense resistance is 0.2 m $\Omega$ , configuration of amp gain is 25 V/V, and adjust current is +5 A, LSB current is 13.73 mA. And the converted register value corresponding to the adjust current and setting value of OFST10 register is calculated as following.

adjust current[DEC] = 
$$5[A] \div 13.73[mA] \approx 364.17 \approx 364[DEC]$$
  
OFST10 value[DEC] =  $300[DEC] + 364[DEC] = 664[DEC] = 298[HEX]$ 

#### Notice 1

Please write registers for manual offset calibration in IDLE state or SSHDN mode, and do not write when measuring current.

Notice 2

If you operate Mode 3 Offset Calibration again, please overwrite OFST10 register with added/subtracted value corresponding to the additional adjust current.

Notice 3

When the configuration of amp gain is changed, please re-calculate OFST10 register value because of changing LSB current.

#### 5. SPI Interface

#### 5.1 Summary

BD7220FV-C is equipped with 500 kHz (Max) SPI interface. And also, it has the selectable (with/without) CRC code to improve the reliability of the communication (Mode 0 correspondence).

#### 5.1.1 Features

- · SPI Interface
- · "Data write" carried out for every 1byte unit
- \* "Data read" carried out for every consecutive byte unit (setting number of bytes).
- · Selectable with/without CRC cord (by the EC bit MSB of the control address)
- CRC polynomial: X<sup>8</sup>+X<sup>2</sup>+X+1

#### 5.1.2 Constitution

BD7220FV-C has SPI interface.

Power supply of the SPI interface is the VDD pin.

SPI interface is enabled by turning the CSB pin to "L" level. The IC takes in the MSB-first input data on the SDI pin synchronous to the rising edges of SCK clock. Output- data is supplied on the SDO pin in the MSB-first order synchronous to the falling edges of SCK clock. SPI interface is disabled with "H" level input on the CSB pin and returns to the initial state. The CSB pin should be fixed to "H" level every time after one data write/read operation is completed. (Regarding the data reading, consecutive read in byte is available.)

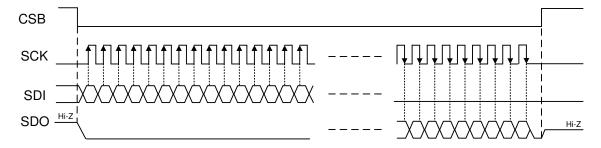


Figure 5-1. SPI Timing Chart

Configurations and controls can be done by reading/writing corresponding addresses in the control register. Write data is one-byte length, while read data length is specified in read commands. Set the RW bit to "0" for data write and "1" for data read. Also, set the EC bit to "1" if the CRC code for detecting a communication error is required or to "0" otherwise.

7	6	5	4	3	2	1	0
EC		Con	trol regi	ster add	Iress		RW

	"0"	"1"
EC	without CRC	with CRC
RW	write	read

Figure 5-2. Control Address Constitution

### 5.2 Function Description

The below figures show the communication format of the data read/write with/without CRC

## 5.2.1 Data Write Communication Format without CRC

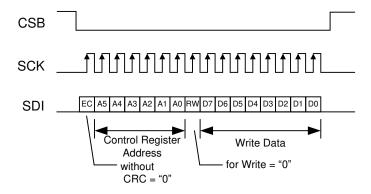


Figure 5-3. Data Write Communication Format without CRC

## 5.2.2 Data Read Communication Format without CRC

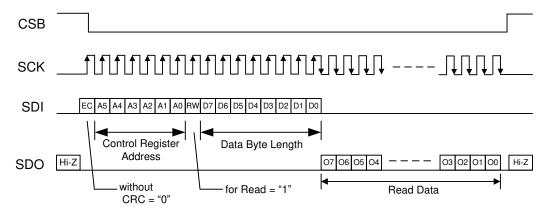


Figure 5-4. Data Read Communication Format without CRC

#### 5.2.3 Data Write Communication Format with CRC

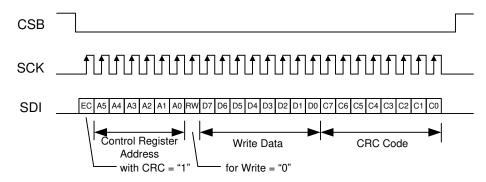


Figure 5-5. Data Write Communication Format with CRC

If a communication with CRC is selected (EC = "1"), 1-byte CRC (Cyclic Redundancy Code) is generated according to an  $X^8+X^2+X+1$  equation, and added at the end of each communication data. Set the CSB pin to "H" level to initialize CRC computation to the default FFh.

Data write is performed on the specified control register only if the result from CRC computation matches the received CRC. Otherwise, data write is not performed. CRC error flag is set when CRC error is detected, and an interrupt signal is output to MCU from the INTB pin. Refer to the chapter 3 about detail interrupt explanation.

#### 5.2.4 Data Read Communication Format with CRC

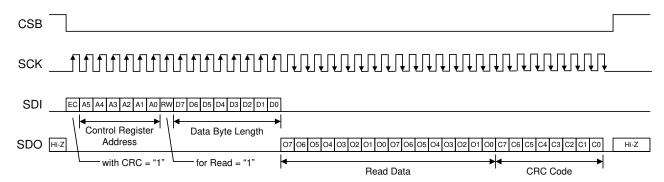


Figure 5-6. Data Read Communication Format with CRC

The CRC computation is also performed for each transmitted/received data during data read operation and the result is appended at the end of the read data. The external MCU can detect any communication errors by comparing the CRC computation result and the received CRC. The CRC code is not included in the data byte length.

The above figure is the example that the number of the read byte data is 2 bytes (data of read byte's number : 02h).

## 5.3 Multi Bytes Data Reading

The values of CURCD, AVE\_CURCD, CC\_CCNTD, CHG\_CCNTD, DIS\_CCNTD and EXADIN\_VALUE registers are updated on current measurement and current accumulation and when EXADIN TRG is triggered.

Please start reading from MSB address regarding these registers. By reading MSB address of multiple bytes' data, the value is retained in the internal buffer. This avoids updating the read data while reading over multiple address. Reading from LSB address may cause updating the data before finish reading all bytes.

If an access to another address is performed before reach to LSB address, read start from MSB address again. In this case, the read data is the value at the time of starting to read MSB again.

#### Ex.) CURCD Reading

Correct order : address 13h CURCD\_H  $\rightarrow$  address 14h CURCD\_L Incorrect order : address 14h CURCD\_L  $\rightarrow$  address 13h CURCD\_H

#### 6. EXADIN

#### 6.1 Summary

ΔΣADC in BD7220FV-C has an input channel selectable function. A current value is monitored through AMP in normal operation. It can also monitor the voltage of the outside EXADIN pin by a setting signal from SPI. It is possible to measure a value such as voltage of a battery or a thermistor by time sharing by this EXADIN function. (During the EXADIN pin measurement, it cannot measure a current value)

#### 6.1.1 Features

•The EXADIN pin which can A-D convert the analog data besides current is available (EXADIN Input voltage range: 0.5 V to 4.5 V)

#### 6.1.2 Structure

Figure 6-1 indicates the connection and structure of the EXADIN pin.

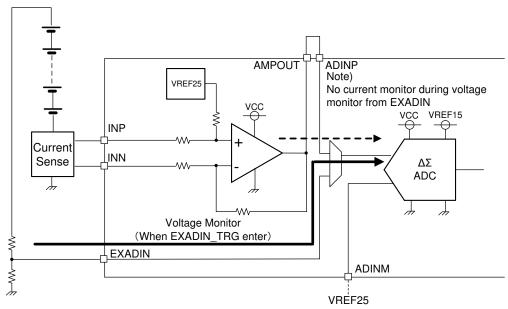


Figure 6-1. EXADIN Block Diagram

## 6.2 Function Description

To monitor of the EXADIN pin voltage is carried out by writing in "1" to EXADIN\_TRG register (address 02h CC\_TRG\_RST\_CMD [3]). After "1" was written in, EXADIN\_TRG becomes "0" automatically.

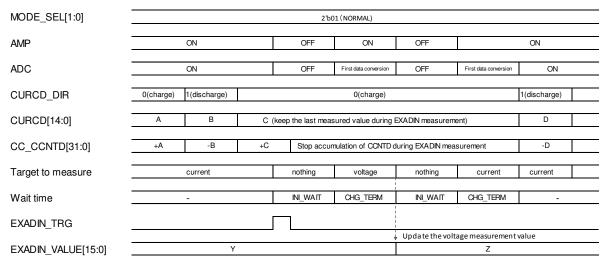


Figure 6-2. EXADIN Monitor Timing Chart

## 7. Register Map

Register address (dec)	Register address (Hex)	Register Name	D7	D6	D5	D4	D3	D2	D1	D0	Initial Value	Access (R, R/W)
0	00h	CC_SET1	AMP_G	AIN[1:0]	MCIC	_R[1:0]	-	CONTEN	MODE_	SEL[1:0]	80h	R/W
1	01h	CC_SET2	OSF	[1:0]	CC_UNDIV	AVE_CURC	D_COUNT[1:0]		CCNTD_MASK[2:0]		05h	R/W
2	02h	CC_TRG_RST_CMD	-	-	CALIB_TRG	-	EXADIN_TRG	DIS_CONTD_RST	CHG_CCNTD_RST	CCNTRST	00h	R/W
3	03h	CC_SET3	-	SLEEP_CC_SEL	SLEEP_SAMP	LING_TIME[1:0]	1	ERVAL[1:0]	-	REX_EN	01h	R/W
4	04h	ADC_CALIB	-	-	-	-	GAIN_CAL_FS		CALIB_MODE[2:0]		00h	R/W
5	05h	GAIN10_2	-	-	-	-		GAIN1	0[17:14]		00h	R/W
6	06h	GAIN10_1					10[13:6]				00h	R/W
7	07h	GAIN10_0_GAIN30_2			GAIN	10[5:0]			GAINS	0[17:16]	00h	R/W
8	08h	GAIN30_1					30[15:8]				00h	R/W
9	09h	GAIN30_0					130[7:0]				00h	R/W
10	0Ah	OFST10_H					10[15:8]				00h	R/W
11	0Bh	OFST10_L			,	OFST	[10[7:0]				00h	R/W
12	0Ch	GAIN31_2		-					GAINS	1[17:16]	00h	R
13	0Dh	GAIN31_1					31[15:8]				40h	R
14	0Eh	GAIN31_0				i e	131[7:0]				00h	R
15	0Fh	CALVIN_DIFF_H						CALVIN	_DIFF[11:8]		00h	R/W
16	10h	CALVIN_DIFF_L					_DIFF[7:0]				00h	R/W
17	11h	EXADIN_VALUE_H	1				/ALUE[15:8]				00h	R
18	12h	EXADIN_VALUE_L				EXADIN_1	VALUE[7:0]				00h	R
19	13h	CURCD_H	CURCD_DIR				CURCD[14:8]				00h	R
20	14h	CURCD_L				CUR	CD[7:0]				00h	R
21	15h	AVE_CURCD_H	AVE_CURCD_DIR				AVE_CURCD[14:8]				00h	R
22	16h	AVE_CURCD_L					URCD[7:0]				00h	R
23	17h	CC_CCNTD_3					ITD[31:24]				00h	R/W
24	18h	CC_CONTD_2					ITD[23:16]				00h	R/W
25	19h	CC_CCNTD_1					NTD[15:8]				00h	R/W
26	1Ah	CC_CCNTD_0					NTD[7:0]				00h	R/W
27	1Bh	CHG_CCNTD_3					NTD[31:24]				00h	R/W
28	1Ch	CHG_CCNTD_2					NTD[23:16]				00h	R/W
29	1Dh	CHG_CCNTD_1					XNTD[15:8]				00h	R/W
30	1Bh	CHG_CCNTD_0					CNTD[7:0]				00h	R/W
31	1Fh	DIS_CCNTD_3					VTD[31:24]				00h	R/W
32	20h	DIS_CONTD_2					VTD[23:16]				00h	R/W
33	21h	DIS_CONTD_1					NTD[15:8]				00h	R/W
34	22h	DIS_CCNTD_0					ONTD[7:0]				00h	R/W
35	23h	CC_BATCAP1_TH_2					P1_TH[23:16]				00h	R/W
36	24h	CC_BATCAP1_TH_1					AP1_TH[15:8]				00h	R/W
37	25h	CC_BATCAP1_TH_0					AP1_TH[7:0]				00h	R/W
38	26h	CC_BATCAP2_TH_2					P2_TH[23:16]				00h	R/W
39	27h	CC_BATCAP2_TH_1					AP2_TH[15:8]				00h	R/W
40	28h	CC_BATCAP2_TH_0					AP2_TH[7:0]				00h	R/W
41	29h	CC_BATCAP3_TH_2	1				P3_TH[23:16]				00h	R/W
42	2Ah	CC_BATCAP3_TH_1	1				AP3_TH[15:8]				00h	R/W
43	2Bh	CC_BATCAP3_TH_0					AP3_TH[7:0]				00h	R/W
44	2Ch	CC_BATCAP4_TH_2	1				P4_TH[23:16]				00h	R/W
45	2Dh	CC_BATCAP4_TH_1	1				AP4_TH[15:8]				00h	R/W
46	2Eh	CC_BATCAP4_TH_0				CC_BATG	AP4_TH[7:0]				00h	R/W
47	2Fh	OCURTHR1_H	OCURTHR1_DIR				OCURTHR1[14:8]				00h	R/W
48	30h	OCURTHR1_L	ļ	r		OCURT	THR1[7:0]				00h	R/W
49	31h	OCURTHR2_H	OCURTHR2_DIR				OCURTHR2[14:8]				00h	R/W
50	32h	OCURTHR2_L	ļ	T		OCURT	THR2[7:0]				00h	R/W
51	33h	OCURTHR3_H	OCURTHR3_DIR				OCURTHR3[14:8]				00h	R/W
52	34h	OCURTHR3_L			,		THR3[7:0]				00h	R/W
53	35h	CC_SET4	REX_D	UR[1:0]	OCURD	UR3[1:0]		UR2[1:0]	OCURE	DUR1[1:0]	00h	R/W
54	36h	REX_CURCD_TH	1				ICD_TH[7:0]				00h	R/W
55	37h	WAKE_CURCD_TH		r	,		RCD_TH[7:0]				00h	R/W
56	38h	CC_SET5	-	-		AIT[1:0]				XOUNT[1:0]	00h	R/W
57	39h	INT_EN1	CC_MON4_RES_EN	CC_MON4_DET_EN	CC_MON3_RES_EN	CC_MON3_DET_EN	CC_MON2_RES_EN	CC_MON2_DET_EN	CC_MON1_RES_EN	CC_MON1_DET_EN	00h	R/W
58	3Ah	INT_BN2	OCUR3_RES_EN	OCUR3_DET_EN	OCUR2_RES_EN	OCUR2_DET_EN	OCUR1_RES_EN	OCUR1_DET_EN		ALARM_OCUR1_DET_EN	00h	R/W
59	3Bh	INT_EN3	-	CALIB_FIN_EN	-	CRCERR_DET_EN		REX_DET_EN	WAKE_RES_EN	WAKE_DET_EN	00h	R/W
60	3Ch	INT_REQ1	CC_MON4_RES	CC_MON4_DET	CC_MON3_RES	CC_MON3_DET	CC_MON2_RES	CC_MON2_DET	CC_MON1_RES	CC_MON1_DET	00h	R/W
61	3Dh	INT_REQ2	OCUR3_RES	OCUR3_DET	OCUR2_RES	OCUR2_DET	OCUR1_RES	OCUR1_DET	-	-	00h	R/W
62	3Eh	INT_REQ3	-	CALIB_FIN	OTP_DL_FIN	CRCERR_DET		REX_DET	WAKE_RES	WAKE_DET	00h	R/W
63	3Fh	PAGE_SEL	HOSC_ON		1 .	1 .	1 .	l .	PAGE	SEL[1:0]	00h	R/W

#### Address 00h: CC SET1 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
00h	CC_SET1	CC_SET1 R/W		AMP_GAIN[1:0]		MCIC_R[1:0]		CCNTEN	MODE_	SEL[1:0]
oun	Initial Value	80h	1	0	0	0	0	0	0	0

Bit 7-6: AMP GAIN[1:0] 00: 5 V/V 01: 25 V/V 10: 25 V/V (default) 11: 51 V/V

Differential amplifier gain reshuffling register

Depending on the measurement current range, it is necessary to choose an appropriate amplifier gain setting. Reference measurement current range when attaching a shunt resistance of 0.2 m $\Omega$  is as follows. (Refer to Table 1-1)

-1000 A to +2000 A -> 5 V/V ±200 A -> 51 V/V

MCIC\_R[1:0] 00: 32 (default) 01: 128 10: 256 11: 1024

Built-in digital filter of  $\Delta$   $\Sigma$  ADC

(down sampling)

Changing the sampling time and filter response of built-in digital filter for ADC. The parameters that change under the influence of MCIC\_R are as follows.

CURCD data are updated every ADC sampling time, based on Table 1-5.

AVE CURCD[14:0]

AVE\_CURCD data are updated every Average time, based on Table 1-4.

CC\_CCNTD[31:0], CHG\_CCNTD[31:0], DIS\_CCNTD[31:0]

According to Table 2-3, the number of measurements during SLEEP mode CC\_CCNTD, CHG\_CCNTD, DIS\_CCNTD data update is constant (250µs). asurements during SLEEP mode is changed by MCIC\_R.

Bit 2: CCNTEN

0: Coulomb counter OFF (default) 1: Coulomb counter ON

Enable Coulomb counter

CCNTEN=1 enables current accumulation.
Based on CURCD data (refer to section 1.3.1), current accumulation and

output current accumulation data. (refer to section 1.3.2) Current accumulation data can read via SPI. There are 3 kinds of current accumulation data: CC\_CCNTD[31:0], CHG\_CCNTD[31:0], DIS\_CCNTD[31:0]

CCNTEN=0 to disables current accumulation.

When disabled, CC\_CCNTD[31:0], CHG\_CCNTD[31:0], DIS\_CCNTD[31:0] keep the last data written.  $When \ CCNTEN = 0, \ CC\_CCNTD[31:0], \ CHG\_CCNTD[31:0], \ DIS\_CCNTD[31:0] \ can \ overwrite$ 

by SPI command

Whether CCNTEN = 0 or 1, CC\_CCNTD[31:0] is reset to "0" by CCNTRST=1.

Whether CCNTEN = 0 or 1, CHG\_CCNTD[31:0] is reset to "0" by CHG\_CCNTD\_RST=1.

Whether CCNTEN = 0 or 1, DIS\_CCNTD[31:0] is reset to "0" by DIS\_CCNTD\_RST=1.

Bit 1-0: MODE\_SEL[1:0]

00: (default) which does not shift

01: NORMAL 10: SLEEP 11: SSHDN

Operation mode choice

Operation mode can be changed from IDEL to any of three mode in left, NORMAL, SLEEP and SSHDN.

In the NORMAL, SLEEP and SSHDN mode, transition is possible by SPI command freely Operation mode is not changed by writing "00". (Remaining current mode)

#### Address 01h: CC SET2 Register (R/W)

Addres (Index	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
01h	CC_SET2	R/W	OSR[1:0]		CC_UNDIV	AVE_CURCD	_COUNT[1:0]	(	CONTD_MASK[2:0	1]
UIN	Initial Value	05h	0	0	0	0	0	1	0	1

OSR[1:0] Bit 7-6:

00: 32(OSF=128kHz) (default) 01: 128(OSF=128kHz) 10: 512(OSF=2.048MHz) 11: 32(OSF=128kHz)

Over sampling rate setting **XOSF=Over Sampling Frequency**  Over sampling rate settings for built in  $\Delta$   $\Sigma$  ADC.

The parameters to change under the influence of OSR are as follows.

CURCD data are updated every ADC sampling time of table 1-5.

AVE CURCD data are updated every average time of table 1-4.

CC\_CCNTD[31:0], CHG\_CCNTD[31:0], DIS\_CCNTD[31:0]
According to setting of table 2-3, the Number of measurement during SLEEP mode is changed by MCIC\_R setting. CC\_CCNTD, CHG\_CCNTD, DIS\_CCNTD data update

is constant in every 250 μs.

CC UNDIV Bit 5: 0: Enable (default) Enable scaled accumulation

To cancel the variation of LSB current in each gain setting, built-in calculation logic

circuit can perform scaled calculation of current accumulation. So, LSB of current accumulation is constant regardless of gain setting.

CC\_UNDIV controls if this function is enabled

Under the configulation of CC\_UNDIV = 0, scaled calculation, gain setting can be changed with small error.

Under the configulation of CC\_UNDIV = 1, not scaled calculation, gain setting is

Note that, compare to the capacity at 5 V/V gain, the current accumulation capacity is reduced depending on setting of gain, 1/5 at 25 V/V and 1/10.2 at 51 V/V.

( LSB of current accumulation is changed)

Bit 4-3: AVE\_CURCD\_COUNT[1:0]

00: 4 times (default) 01: 16 times 10: 64 times 11: 128 times

Measurement count for average current calculation of a fixed time interval

The current measurement value of each fixed time interval can be calculated as an average

value. It is not a moving average.

AVE\_CURCD\_COUNT sets the number of measurements to be averaged.

Average is caluculated from accumulating the A-D converted data for the times

configured in AVE\_CURCD\_COUNT.
Fixed interval is configured with the setting of digital filter (address 00h:MCIC\_R[1:0]), the setting of OSR(address 00h:SR[1:0]), and the setting of measurement number (address 01h: AVE\_CURCD\_COUNT[1:0])

CCNTD\_MASK[2:0] 000: 7bits mask (effective 11bits) 001: 6bits mask (effective 12bits) 010: 5bits mask (effective 13bits) 011: 4bits mask (effective 14bits)

100: 3bits mask (effective 15bits)
101: 2bits mask (effective 16bits) (default)
110: 1bit mask (effective 17bits)
111: 0bit mask (effective 18bits)

Bit mask function for current accumulation

CCNTD\_MASK sets the lower bit masked from the measurement current value during current accumulation. When masked, these bits are set to "0".

#### Address 02h: CC TRG RST CMD Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
02h	CC_TRG_RST_CMD	R/W	-	-	CALIB_TRG	-	EXADIN_TRG	DIS_CCNTD_RST	CHG_CCNTD_RST	CCNTRST
UZII	Initial Value	00h	0	0	0	0	0	0	0	0

CALIB TRG Bit 5:

Calibration trigger

XAfter "1" is written, value returns to "0" automatically

No trigger for calibration
 Calibration start trigger

EXADIN TRG Bit 3:

Δ Σ ADC input signal change trigger

\*After "1" is written, value returns to "0" automatically

O: Current measurement

1: Switch Δ Σ ADC input to EXADIN (EXADIN pin voltage reading)

Bit 2:

Discharge current accumulation (DIS\_CCNTD) reset

\*After "1" is written, value returns to "0" automatically

DIS\_CCNTD\_RST
0: No Reset
1: Reset DIS\_CCNTD

Bit 1:

Bit 0:

Charge current accumulation (CHG\_CCNTD) reset

\*After "1" is written, value returns to "0" automatically

CCNTRST

0: No Reset 1: Reset CC\_CCNTD

CHG\_CCNTD\_RST 0: No Reset 1: Reset CHG\_CCNTD

Current accumulation

(CC\_CCNTD) reset

%After "1" is written, value returns to "0" automatically

Address 03h: CC SET3 Register (R/W)

_											
	Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
Ī	03h	CC_SET3	R/W	-	SLEEP_CC_SEL	SLEEP_SAMP	LING_TIME[1:0]	SLEEP_INT	ERVAL[1:0]	-	REX_EN
	USII	Initial Value	01h	0	0	0	0	0	0	0	1

Bit 6: SLEEP CC SEL Current accumulation method in SLEEP mode, during the OFF period

C: Accumulate the last current measurement taken during the ON period. (default)
 Accumulate the average of measurements taken during the ON period.

Bit 5-4: SLEEP\_SAMPLING\_TIME[1:0]

If ADC sampling period < 2 [ms]:

00: 1 (default) 01: 16 10:32 11:64

If ADC sampling period ≥2 [ms]:

00: 1 (default) 01: 2 10.4

Bit 3-2 : SLEEP\_INTERVAL[1:0] 00: ON/OFF=1:7 (default)

01: ONOFF=1:15 10: ONOFF=1:31

REX\_EN 0: RELAX timer OFF

Number of current accumulation when ON during SLEEP mode.

ON/OFF time ratio during SLEEP mode

11: ON/OFF=1:127

Relax state detection timer enable

When REX\_EN=1, if CURCD[14:0](13h+14h)  $\leq$  { 7'd0, REX\_CURCD\_TH[7:0](36h)},

the relax timer starts.

If REX\_EN = 1  $\rightarrow$  0 is set while the timer is counting, the timer counter is reset to 0. (Note that not keep the value but reset)

The parameters that change under the influence of SLEEP\_SAMPLING\_TIME are as follows. CC\_CCNTD[31:0], CHG\_CCNTD[31:0], DIS\_CCNTD[31:0]
The number of current accumulations in SLEEP mode are decided as in Table 2-3.

The current accumulation data update is constant for every 250 µs

If REX\_EN=  $0 \rightarrow 1$  again, start counting from value = 0.

Address 04h: ADC CALIB Register (R/W)

1: RELAX timer ON (default)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
04h	ADC_CALIB	R/W	-	-	-	-	GAIN_CAL_FS	(	CALIB_MODE[2:0]	
U411	Initial Value	00h	0	0	0	0	0	0	0	0

GAIN\_CAL\_FS Bit 3:

O: Negative full scale (MFS) direction
Positive full scale (PFS) direction

Input polarity for GAIN calibration

Bit 2-0 : CALIB\_MODE[2:0]

000: Normal operation (default) 001: Mode2 (GAIN)

010: Mode2 (OFFSET) 011: Reserved 100: Reserved

110: Reserved 111: Reserved Calibration mode setting

\*Do not use any of the "Reserved" modes. (011, 100, 101, 110, 111)

Calibration revision value may shift.

#### Address 05h: GAIN10 2 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
05h	GAIN10_2	R/W	-	-	-	-	GAIN10[17:14]			
0311	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 06h: GAIN10 1 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
06h	GAIN10_1	R/W				GAIN1	0[13:6]			
Ubn	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 07h: GAIN10 0 GAIN30 2 Register (R/W)

	dress ndex)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
	07h	GAIN10_0_GAIN30_2	R/W		GAIN10[5:0]						
'		Initial Value	00h	0	0	0	0	0	0	0	0

GAIN10[17:0]

Mode 2 gain calibration value

%Register data is cleared on reset (SHDNB=L or VCC UVLO). Please keep register data in nonvolatile memory on the system and rewrite after IC restarts.

## Address 08h: GAIN30 1 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
08h	GAIN30_1	R/W				GAIN3	0[15:8]			
UOII	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 09h: GAIN30 0 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
09h	GAIN30_0	R/W				GAINS	30[7:0]			
0911	Initial Value	00h	0	0	0	0	0	0	0	0

GAIN30[17:0]

Mode 3 gain calibration value

%:Register data is cleared on reset (SHDNB=L or VCC UVLO). Please keep register data in nonvolatile memory on the system and rewrite after IC restarts.

## Address 0Ah: OFST10\_H Register (R/W)

	ddress ndex)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
	0Ah	OFST10_H	R/W				OFST1	0[15:8]			
'	UAII	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 0Bh: OFST10\_L Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0Bh	OFST10_L	R/W		OFST10[7:0]						
UDII	Initial Value	00h	0	0	0	0	0	0	0	0

OFST10[15:0]

Mode 2 offset calibration value

\*\*Register data is cleared on reset (SHDNB=L or VCC UVLO).

Please keep register data in nonvolatile memory on the system and rewrite after IC restarts.

#### Address 0Ch: GAIN31 2 Register (R)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0Ch	GAIN31_2	R							GAIN31	[17:16]
UCh	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 0Dh: GAIN31 1 Register (R)

dress dex)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
0Dh	GAIN31_1	R		GAIN31[15:8]							
ווטי	Initial Value	40h	0	1	0	0	0	0	0	0	

#### Address 0Eh: GAIN31 0 Register (R)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0Eh	GAIN31_0	R				GAINS	31[7:0]			
UEN	Initial Value	00h	0	0	0	0	0	0	0	0

GAIN31[17:0]

Register for the GAIN30 operation (GAIN30 is register for the manual setting)

## Address 0Fh: CALVIN DIFF H Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0Fh	CALVIN_DIFF_H	R/W	-	-	-	-		CALVIN_[	DIFF[11:8]	
UFII	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 10h: CALVIN DIFF L Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
10h	CALVIN_DIFF_L	R/W				CALVIN_	DIFF[7:0]			
TON	Initial Value	00h	0	0	0	0	0	0	0	0

CALVIN\_DIFF[11:0]

Mode 2 calibration setting register

Input differential voltage information between INP and INN for GAIN calibration

## Address 11h: EXADIN VALUE H Register (R)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
11h	EXADIN_VALUE_H	R				EXADIN_V	ALUE[15:8]			
iin	Initial Value	00h	0	0	0	0	0	0	0	0

### Address 12h: EXADIN VALUE L Register (R)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
12h	EXADIN_VALUE_L	R				EXADIN_V	/ALUE[7:0]			
1211	Initial Value	00h	0	0	0	0	0	0	0	0

EXADIN\_VALUE[15:0]

Measurement value of EXADIN input

#### Address 13h: CURCD H Register (R)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
13h	CURCD_H	R	CURCD_DIR				CURCD[14:8]			
1311	Initial Value	00h	0	0	0	0	0	0	0	0

Bit 7: CURCD\_DIR Current direction bit

0: Charge (INP pin voltage > INN pin voltage)

1: Discharge (INP pin voltage < INN pin voltage)

#### Address 14h: CURCD L Register (R)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
14h	CURCD_L	R				CURC	D[7:0]			
1411	Initial Value	00h	0	0	0	0	0	0	0	0

CURCD[14:0]

Current value

LSB unit of CURCD depending on the gain setting is described in Chapter 1.3.1.

The reference level when attaching an external 0.2  $m\Omega$  resistance is as follows. (see Table 1-1)

5 V/V gain: LSB current ≈ 68.66 mA 25 V/V gain: LSB current ≈13.73 mA 51 V/V gain: LSB current ≈ 6.73 mA

ex.) When a 1,000 mA current flows, the register level is as follows. 5 V/V gain: 1000/LSB current = 1000/68.66=14 => CURCD=15'h000E 25 V/V gain: 1000/LSB current = 1000/13.73=72 => CURCD=15'h0048 51 V/V gain: 1000/LSB current = 1000/6.73=148 => CURCD=15'h0094

#### Address 15h: AVE CURCD H Register (R)

	dress dex)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
	5h	AVE_CURCD_H	R	AVE_CURCD_DIR			,	AVE_CURCD[14:8	[		
'	110	Initial Value	00h	0	0	0	0	0	0	0	0

AVE\_CURCD\_DIR Bit 7:

Average current direction bit

0: Charge

(INP pin voltage > INN pin voltage)

1: Discharge
(INP pin voltage <INN pin voltage)

#### Address 16h: AVE CURCD L Register (R)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
16h	AVE_CURCD_L	R				AVE_CU	RCD[7:0]			
1011	Initial Value	00h	0	0	0	0	0	0	0	0

AVE\_CURCD[14:0]

Average Current value

Number of averaging is set in AVE\_CURCD\_COUNT(01h)

Number of averaging is set in Ave\_cond\_coordinate. LSB unit of CURCD depending on the gain setting is described in Chapter 1.3.1. The reference level when attaching an external  $0.2~\text{m}\Omega$  resistance is as follows.

(see Table 1-1) 5 V/V gain: LSB current ≈ 68.66 mA 25 V/V gain: LSB current ≈13.73 mA 51 V/V gain: LSB current ≈ 6.73 mA

ex.) When a 1,000 mA current flows, the register level is as follows. 5 V/V gain: 1000/LSB current = 1000/68.66°14 => CURCD=15'h000E 25 V/V gain: 1000/LSB current = 1000/13.73°72 => CURCD=15'h0048 51 V/V gain: 1000/LSB current = 1000/6.73°148 => CURCD=15'h0094

#### Address 17h: CC CCNTD 3 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
17h	CC_CCNTD_3	R/W				CC_CCN	TD[31:24]			
1711	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 18h: CC CCNTD 2 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
18h	CC_CCNTD_2	R/W				CC_CCN	TD[23:16]			
180	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 19h: CC CCNTD 1 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
19h	CC_CCNTD_1	R/W				CC_CCN	ITD[15:8]			
1911	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 1Ah: CC CCNTD 0 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
1Ah	CC_CCNTD_0	R/W				CC_CCI	NTD[7:0]			
IAII	Initial Value	00h	0	0	0	0	0	0	0	0

CC\_CCNTD[31:0]

Current Accumulation register

LSB unit of CC\_CCNTD depending on the gain setting is described in Chapter 1.3.2. The reference level when attaching an external 0.2 m $\Omega$  resistance and CC\_UNDIV=0 is as follows. LSB level = 0.3052  $\mu Ah$  MSB level = 655.36 Ah When CC\_UNDIV=1, the current accumulation unit varies depending on setting of AMP\_GAIN. (see table 1-2)

#### Address 1Bh: CHG CCNTD 3 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
1Bh	CHG_CCNTD_3	R/W				CHG_CCN	ITD[31:24]			
IDII	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 1Ch: CHG CCNTD 2 Register (R/W)

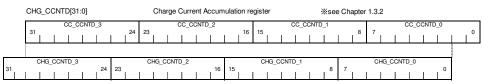
ddress Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
1Ch	CHG_CCNTD_2	R/W				CHG_CCN	ITD[23:16]			
1011	Initial Value	00h	0	0	0	0	0	0	0	0

### Address 1Dh: CHG CCNTD 1 Register (R/W)

		3,	(							
Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
1Dh	CHG_CCNTD_1	R/W		CHG_CCNTD[15:8]						
IDII	Initial Value	00h	0	0	0	0	0	0	0	0

### Address 1Eh: CHG\_CCNTD\_0 Register (R/W)

Addre (Inde		R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
1Eł	CHG_CCNTD_0	R/W				CHG_CC	:NTD[7:0]			
IEI	Initial Value	00h	0	0	0	0	0	0	0	0



LSB of CHG\_CCNTD is equivalent to bit[2] of CC\_CCNTD.

## Address 1Fh: DIS CCNTD 3 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
1Fh	DIS_CCNTD_3	R/W				DIS_CCN	TD[31:24]			
IFN	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 20h: DIS CCNTD 2 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
20h	DIS_CCNTD_2	R/W				DIS_CCN	TD[23:16]			
2011	Initial Value	00h	0	0	0	0	0	0	0	0

## Address 21h: DIS CCNTD 1 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
21h	DIS_CCNTD_1	R/W				DIS_CCN	ITD[15:8]			
2111	Initial Value	00h	0	0	0	0	0	0	0	0

## Address 22h: DIS CCNTD 0 Register (R/W)

Addre (Inde		R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
22h	DIS_CCNTD_0	R/W				DIS_CCI	NTD[7:0]			
221	Initial Value	00h	0	0	0	0	0	0	0	0

	D	IS_C	100	NTD	[31:0	0]					Di	sch	arge	Cur	rent A	CCL	ımulat	ion	regi	ster					e ch	apte	r 1.3	.2						
	Г			С	C_C	CNT	D_3						CC_	CCN	TD_2			Т			C	C_CC	OTN	_1						CC_C	CNT	D_0		
	3	1 	1			ĺ	ı	1	2	4 23	ı	ı	1	ı	1	1	16	6	15 	ĺ	١		ı	1	ı	8	7	ı	1	ī	i	ı	ı	0
	Г																																	
	_	DIS	S_C	CNTE	)_3			T			DIS_0	CCN	TD_2			T			DIS	_CCI	NTD_	1					С	IS_C	CNTI	0_0			i	
31		1					24	23	1		L	1			16	9	15		- [					8	7			L		1		0		

LSB of DIS\_CCNTD is equivalent to bit[2] of CC\_CCNTD.

### Address 23h: CC BATCAP1 TH 2 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
23h	CC_BATCAP1_TH_2	R/W				CC_BATCAF	P1_TH[23:16]			
2311	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 24h: CC BATCAP1 TH 1 Register (R/W)

Addre (Inde		R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
241	CC_BATCAP1_TH_1	R/W				CC_BATCA	P1_TH[15:8]			
241	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 25h: CC BATCAP1 TH 0 Register (R/W)

Add (Inc	lress dex)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
25	٦	CC_BATCAP1_TH_0	R/W				CC_BATCA	P1_TH[7:0]			
23	OI I	Initial Value	00h	0	0	0	0	0	0	0	0

CC CCNTD 3

CC\_BATCAP1\_TH[23:0]

Threshold of current accumulation detection1

31				24	23					16	15					8	7					0
	l	l	1	1			1	1	1	1		1	1	1	1	l		1	1	1	1	
		•														•						

	(	C_I	BATO	CAP	#_TH	2				CC_E	BATC	AP#	_TH	1			С	C_B	ATC	AP#	_TH	0	
23		ı	ı	ı	ı	l	16	15	ı	ì	ı	ı			8	7	1	ı		ì	i	l	0

CC\_CCNTD\_2

For higher 24bit of CC\_CCNTD, set the detection of current accumulation threshold. (The setting method is the same for CC\_BATCAP1 to 4\_TH)

CC CCNTD 1

CC CCNTD 0

LSB level of CC\_BATCAP1\_TH and the MSB levels are as follows (when CC\_UNDIV=0)

LSB level =78.125 µAh MSB level =655.36 Ah

The LSB value and MSB value of CC\_CCNTD are as follows. (Refer to section 1.3.2)

LSB level ≈0.3052 µAh MSB level =655.36 Ah

ex.) To set the detection threshold of current accumulation to 1Ah, the register values are as follows.

 $1 [Ah]/78.125 [\mu Ah] = 12800 \Rightarrow CC\_BATCAP1\_TH[23:0] = 24'h003200(24'd12800)$ 

The threshold range is as follows.

Thresholding range =78.125 µAh to 1310.72 Ah

When CC\_UNDIV=1, unit of current accumulation varies depending on setting of AMP\_GAIN. (see Table 1-2)

#### Address 26h: CC BATCAP2 TH 2 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
26h	CC_BATCAP2_TH_2	R/W				CC_BATCAF	P2_TH[23:16]			
260	Initial Value	00h	0	0	0	0	0	0	0	0

## Address 27h: CC BATCAP2 TH 1 Register (R/W)

 			1.0.0.							
Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
27h	CC_BATCAP2_TH_1	R/W				CC_BATCA	P2_TH[15:8]			
2/11	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 28h: CC\_BATCAP2\_TH\_0 Register (R/W)

ddress Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
28h	CC_BATCAP2_TH_0	R/W				CC_BATCA	AP2_TH[7:0]			
2011	Initial Value	00h	0	0	0	0	0	0	0	0

CC\_BATCAP2\_TH[23:0]:

Threshold of current accumulation detection2

\*\*Refer to CC\_BATCAP1\_TH for the setting

#### Address 29h: CC BATCAP3 TH 2 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
29h	CC_BATCAP3_TH_2	R/W				CC_BATCAF	P3_TH[23:16]			
29n	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 2Ah: CC BATCAP3 TH 1 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
2Ah	CC_BATCAP3_TH_1	R/W	CC_BATCAP3_TH[15:8]							
∠An	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 2Bh: CC BATCAP3 TH 0 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
2Bh	CC_BATCAP3_TH_0	R/W	CC_BATCAP3_TH[7:0]							
ZDII	Initial Value	00h	0	0	0	0	0	0	0	0

CC\_BATCAP3\_TH[23:0]

Threshold of current accumulation detection3

\*\*Refer to CC\_BATCAP1\_TH for the setting

#### Address 2Ch: CC BATCAP4 TH 2 Register (R/W)

Addres (Index		R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
2Ch	CC_BATCAP4_TH_2	R/W				CC_BATCAF	P4_TH[23:16]			
2011	Initial Value	00h	0	0	0	0	0	0	0	0

### Address 2Dh: CC BATCAP4 TH 1 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
2Dh	CC_BATCAP4_TH_1	R/W	CC_BATCAP4_TH[15:8]							
ZUII	Initial Value	00h	0	0	0	0	0	0	0	0

#### Address 2Eh: CC BATCAP4 TH 0 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
2Eh	CC_BATCAP4_TH_0	R/W				CC_BATCA	AP4_TH[7:0]			
2En	Initial Value	00h	0	0	0	0	0	0	0	0

CC\_BATCAP4\_TH[23:0]

Threshold of current accumulation detection4

\*\*Refer to CC\_BATCAP1\_TH for the setting

#### Address 2Fh: OCURTHR1 H Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
2Fh	OCURTHR1_H	R/W	OCURTHR1_DIR	HR1_DIR OCURTHR1[14:8]						
2111	Initial Value	00h	0	0	0	0	0	0	0	0

OCURTHR1\_DIR:

Current direction setting of OCURTHR1

0: Charge (INP pin voltage > INN pin voltage)

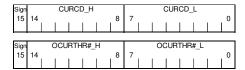
1: Discharge (INP pin voltage <INN pin voltage)

#### Address 30h: OCURTHR1 L Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
30h	OCURTHR1_L	R/W	OCURTHR1[7:0]							
3011	Initial Value	00h	0	0	0	0	0	0	0	0

OCURTHR1[14:0]:

Threshold of measurement current detection1



LSB unit of CURCD depending on the gain setting is described in Chapter 1.3.1. The reference level when attaching an external  $0.2m\Omega$  resistance is as follows.

(Refer to table 1-1) 5 V/V gain: LSB electric current level ≈ 68.66 mA 25 V/V gain: LSB electric current level ≈ 13.73 mA 51 V/V gain: LSB electric current level ≈ 6.73 mA

ex.) When a current value of 1,000mA flows, the register level is as follows. <code>OCCURTHR1\_DIR = 0</code> and

5 V/V gain: 1000/LSB current = 1000/68.66=14 => OCURTHR1=15'h000E 25 V/V gain: 1000/LSB current = 1000/13.73=72 => OCURTHR1=15'h0048 51 V/V gain: 1000/LSB current = 1000/6.73=148 => OCURTHR1=15'h0094

The thresholding ranges are as follows The thresholding range = -1000 A to +2000 A

25 V/V gain: Thresholding range = -1000 A to +2000 A

25 V/V gain: Thresholding range = -400A to +400 A

51 V/V gain: Thresholding range = -200 A to +200 A

Threshold range is different from measurement current range.

When OCURTHR1\_DIR = 0, detection of charging direction of measurement current interrupt be set. An interrupt occurs when current interrupt be set. An interrupt occurs when:  $OCURI\_DET\_EN (3Ah) = 1$  and CURCD (13h + 14h) > OCURTHR1 (2Fh + 30h) and over the consecutive detection setting by OCURDLR1[1:0](35h)  $OCURI\_RES\_EN (3Ah) = 1$  and  $CURCD (13h + 14h) \le OCURTHR1 (2Fh + 30h) \text{ and}$  over the consecutive detection setting by OCURDLR1[1:0](35h)

When OCURTHR1\_DIR = 1, detection of discharging direction of measurement current interrupt be set.An interrupt occurs when: OCUR1\_DET\_EN (3Ah) = 1 and CURCD (13h + 14h) > OCURTHR1 (2Fh + 30h) and over the consecutive detection setting by OCURDUR1[1:0](35h) OCUR1\_RES\_EN (3Ah) = 1 and CURCD (13h + 14h)  $\leq$  OCURTHR1 (2Fh + 30h) and

over the consecutive detection setting by OCURDUR1[1:0](35h)

#### Address 31h: OCURTHR2 H Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
31h	OCURTHR2_H	R/W	OCURTHR2_DIR				OCURTHR2[14:8]			
3111	Initial Value	00h	0	0	0	0	0	0	0	0

OCURTHR2\_DIR Bit 7:

Current direction setting of OCURTHR2

0: Charge
(INP pin voltage > INN pin voltage)
1: Discharge

(INP pin voltage <INN pin voltage)

#### Address 32h: OCURTHR2 L Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
32h	OCURTHR2_L	R/W		OCURTHR2[7:0]							
3211	Initial Value	00h	0	0	0	0	0	0	0	0	

OCURTHR2[14:0]

Threshold of measurement current detection2

Refer to OCURTHR1 for the setting

#### Address 33h: OCURTHR3 H Register (R/W)

	ddress ndex)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
Γ.	33h —	OCURTHR3_H	R/W	OCURTHR3_DIR	33_DIR OCURTHR3[14:8]						
`		Initial Value	00h	0	0 0 0 0 0					0	0

OCURTHR3\_DIR

0: Charge (INP pin voltage > INN pin voltage) 1: Discharge (INP pin voltage <INN pin voltage)

#### Current direction setting of OCURTHR3

#### Address 34h: OCURTHR3 L Register (R/W)

	Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
Ī	34h	OCURTHR3_L	R/W	OCURTHR3[7:0]							
	3411	Initial Value	00h	0	0	0	0	0	0	0	0

OCURTHR3[14:0]

Threshold of measurement current detection3

Refer to OCURTHR1 for the setting

#### Address 35h: CC SET4 Register (R/W)

	ress dex)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
25	35h -	CC_SET4	R/W	REX_D	UR[1:0]	OCURD	UR3[1:0]	OCURD	UR2[1:0]	OCURD	UR1[1:0]
30	)11	Initial Value	00h	0	0	0	0	0	0	0	0

REX\_DUR[1:0] Bit 7-6:

00:30 minutes (default) 01:60 minutes 10:90 minutes 11:120 minutes

Relaxation state detection time

Refer to REX\_CURCD\_TH for the relaxation state detection setting

OCURDUR3[1:0] 00:1 time(default) 01:4 times

10:8 times 11:16 times

Count for detection of measurement current

ex. If CURDUR3 = 4 times, when detect over or under current value 4times, interrupt occurs.
Refer to OCURTHR1, OCUR1\_DET\_EN, OCUR1\_RES\_EN

OCURDUR2[1:0] 00:1 time(default) 01:4 times Bit 3-2:

10:8 times 11:16 times Count for detection of measurement current

ex. If CURDUR2 = 4 times, when detect over or under current value 4times, interrupt occurs.

Refer to OCURTHR1, OCUR1\_DET\_EN, OCUR1\_RES\_EN

OCURDUR1[1:0] 00:1 time(default) 01:4 times Bit 1-0: 10:8 times 11:16 times

Count for detection of measurement current

ex. If CURDUR1 = 4 times, when detect over or under current value 4times, interrupt occurs.

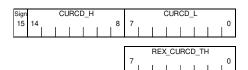
Refer to OCURTHR1, OCUR1\_DET\_EN, OCUR1\_RES\_EN

#### Address 36h: REX CURCD TH Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
36h	REX_CURCD_TH	R/W				REX_CUR	CD_TH[7:0]			
3011	Initial Value	00h	0	0	0	0	0	0	0	0

REX\_CURCD\_TH[7:0]

Threshold of the relaxation state detection measurement current



Refer to REX\_EN about relax timer setting

(Relaxation timer is enable when REX\_EN=1) REX\_CURCD\_TH can be set in the lower 8 bits of CURCD. (unsigned 8 bits expression)

LSB unit of CURCD depending on the gain setting is described in Chapter 1.3.1. The reference level when attaching an external  $0.2m\Omega$  resistance is as follows

(Refer to Table 1-1) 5 V/V gain: LSB current ≈ 68.66 mA 25 V/V gain: LSB current ≈ 13.73 mA 51 V/V gain: LSB current ≈ 6.73 mA

ex.) When a 1,000mA current flows, the register level is as follows 5 V/V gain: 1000/LSB current = 1000/68.66 = 14 => REX\_CURCD\_TH=15'h000E 25 V/V gain: 1000/LSB current = 1000/13.73 = 72 => REX\_CURCD\_TH=15'h0048 51 V/V gain: 1000/LSB current = 1000/6.73 = 148 => REX\_CURCD\_TH=15'h0094

The threshold ranges are as follows. 5 V/V gain: Threshold range  $\approx$  0 to 17.51 A 25 V/V gain: Threshold range  $\approx$  0 to 3.50 A 51 V/V gain: Threshold range ≈ 0 to 1.72 A

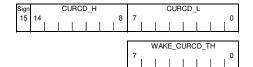
An interrupt occurs when:  $REX\_DET\_EN(3Bh)=1 \mbox{ and } \\ CURCD[14:0] \ (13h+14h) \ \le \{\ 7'd0,\ REX\_CURCD\_TH[7:0] \ (36h)\ \} \mbox{ and } \\$ Relaxation state detection time over the setting by REX\_DUR[1:0](35h)

#### Address 37h: WAKE CURCD TH Register (R/W)

Addre:		R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
37h	WAKE_CURCD_TH	R/W				WAKE_CUF	RCD_TH[7:0]			
3/11	Initial Value	00h	0	0	0	0	0	0	0	0

WAKE\_CURCD\_TH[7:0]

Threshold of wake-up current detection



WAKE\_CURCD\_TH can be set in the lower 8 bits of CURCD. (unsigned 8 bits expression)

LSB unit of CURCD depending on the gain setting is described in Chapter 1.3.1. The reference level when attaching an external 0.2 m $\Omega$  resistance is as follows. (Refer to Table 1-1)

5 V/V gain: LSB current ≈ 68.66 mA 25 V/V gain: LSB current ≈ 13.73 mA 51 V/V gain: LSB current ≈ 6.73 mA

ex.) When a 1,000 mA current flows, the register level is as follows. 5 V/V gain: 1000/LSB current = 1000/68.66=14 => WAKE\_CURCD\_TH=15'h000E 25 V/V gain: 1000/LSB current = 1000/13.73=72 => WAKE\_CURCD\_TH=15'h0048 51 V/V gain: 1000/LSB current = 1000/6.73=148 => WAKE\_CURCD\_TH=15'h0094

The threshold ranges are as follows. 5V/V gain: Threshold range ≈ 0 to 17.51A 25V/V gain: Threshold range ≈ 0 to 3.50A 51V/V gain: Threshold range ≈ 0 to 1.72A

The WAKE\_RES interrupt occurs when: WAKE\_RES\_EN(3Bh) = 1 and  $\label{eq:curcon} $$ CURCD[14:0] (13h + 14h) \le \{ 70d, WAKE\_CURCD\_TH[7:0] (37h) \}$ and $$ Wake up current detection over the setting by WAKE\_COUNT[1:0](38h) $$$ 

#### Address 38h: CC\_SET5 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
38h	CC_SET5	R/W	-	-	INI_W	AIT[1:0]		-	WAKE_C	OUNT[1:0]
3011	Initial Value	00h	0	0	0	0	0	0	0	0

Bit 5-4: INI WAIT[1:0]

00: 1.5ms(default) 01: 3.0ms 10: 6.0ms 11: 12.0ms

Initial wait time setting

The INI\_WAIT register sets initial wait time for start of VREF25 and AMP

Bit 1-0 : WAKE\_COUNT[1:0] 00: 1 time(default) 01: 4 times

10: 8times 11: 16times Wake up current detection count

Refer to WAKE\_CURCD\_TH for the wake up current detection setting

#### Address 39h: INT EN1 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
39h	INT_EN1	R/W	CC_MON4_RES _EN	CC_MON4_DET _EN	CC_MON3_RES _EN	CC_MON3_DET _EN	CC_MON2_RES _EN	CC_MON2_DET _EN	CC_MON1_RES _EN	CC_MON1_DET _EN
3911	Initial Value	00h	0	0	0	0	0	0	0	0

CC\_MON4\_RES\_EN Interrupt Enable : 1: Enable 0: Disable Detection of discharging direction current accumulation value4 1: Enable 0: Disable CC\_MON4\_DET\_EN=1: CC\_MON4\_DET=1 -> INTB=L, CC\_MON4\_DET=0 -> INTB=Hi-z CC\_MON4\_DET\_EN=0: INTB=Hi-z (regardless of CC\_MON4\_DET) Interrupt Enable : Detection of charging direction Bit 6: CC\_MON4\_DET\_EN current accumulation value4 CC\_MON3\_RES\_EN 1: Enable 0: Disable CC\_MON3\_RES\_EN=1: CC\_MON3\_RES=1 -> INTB=L,CC\_MON3\_RES=0 -> INTB=Hi-z CC\_MON3\_RES\_EN=0: INTB=Hi-z (regardless of CC\_MON3\_RES) Bit 5: Detection of discharging direction current accumulation value3 Interrupt Enable : Detection of charging direction current accumulation value3 1: Enable 0: Disable CC\_MON3\_DET\_EN=1: CC\_MON3\_DET=1 -> INTB=L,CC\_MON3\_DET=0 -> INTB=Hi-z CC\_MON3\_DET\_EN=0: INTB=Hi-z (regardless of CC\_MON3\_DET) Bit 4: CC\_MON3\_DET\_EN CC\_MON2\_RES\_EN 1: Enable 0: Disable CC\_MON2\_RES\_EN=1: CC\_MON2\_RES=1 -> INTB=L, CC\_MON2\_RES=0 -> INTB=Hi-z Detection of discharging direction current accumulation value2 CC\_MON2\_RES\_EN=0: INTB=Hi-z (regardless of CC\_MON2\_RES) Interrupt Enable : Detection of charging direction current accumulation value2 1: Enable 0: Disable CC\_MON2\_DET\_EN=1: CC\_MON2\_DET=1 -> INTB=L, CC\_MON2\_DET=0 -> INTB=Hi-z CC\_MON2\_DET\_EN=0: INTB=Hi-z (regardless of CC\_MON2\_DET) Bit 2: CC\_MON2\_DET\_EN Interrupt Enable : Detection of discharging direction 1: Enable 0: Disable CC\_MON1\_RES\_EN=1: CC\_MON1\_RES=1 -> INTB=L, CC\_MON1\_RES=0 -> INTB=Hi-z CC\_MON1\_RES\_EN=0: INTB=Hi-z (regardless of CC\_MON1\_RES) CC\_MON1\_RES\_EN current accumulation value1 1: Enable 0: Disable CC\_MON1\_DET\_EN=1: CC\_MON1\_DET=1 -> INTB=L, CC\_MON1\_DET=0 -> INTB=Hi-z CC\_MON1\_DET\_EN=0: INTB=Hi-z (regardless of CC\_MON1\_DET) Bit 0: CC MON1 DET EN Interrupt Enable: Detection of charging direction current accumulation value1

#### Address 3Ah: INT EN2 Register (R/W)

Addre (Inde		R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
3AI	INT_EN2	R/W	OCUR3_RES_EN	OCUR3_DET_EN	OCUR2_RES_EN	OCUR2_DET_EN	OCUR1_RES_EN	OCUR1_DET_EN	ALARM_OCUR1 _RES_EN	ALARM_OCUR1 _DET_EN
SAI	Initial Value	00h	0	0	0	0	0	0	0	0

	Initial Value	00h	0	0	0	0	0	0	0	0
Bit 7 :	OCUR3_RES_EN		Interrupt Enable : Detection of disch current measurer	narging direction		1: Enable 0: Dis OCUR3_RES_EN OCUR3_RES_EN	N=1: OCUR3_RES			INTB=Hi-z
Bit 6:	OCUR3_DET_EN		Interrupt Enable : Detection of charge current measurer	ging direction		1: Enable 0: Dis OCUR3_DET_EN OCUR3_DET_EN	I=1: OCUR3_DET			NTB=Hi-z
Bit 5 :	OCUR2_RES_EN		Interrupt Enable : Detection of disch current measurer	narging direction		1: Enable 0: Dis OCUR2_RES_EN OCUR2_RES_EN	l=1: OCUR2_RES			.NTB=Hi-z
Bit 4 :	OCUR2_DET_EN		Interrupt Enable : Detection of charge current measurer	ging direction		1: Enable 0: Dis OCUR2_DET_EN OCUR2_DET_EN	I=1: OCUR2_DET			NTB=Hi-z
Bit 3 :	OCUR1_RES_EN		Interrupt Enable : Detection of disch current measurer	narging direction		1: Enable 0: Dis OCUR1_RES_EN OCUR1_RES_EN	l=1: OCUR1_RES			:NTB=Hi-z
Bit 2 :	OCUR1_DET_EN		Interrupt Enable : Detection of charge current measurer	ging direction		1: Enable 0: Dis OCUR1_DET_EN OCUR1_DET_EN	I=1: OCUR1_DET			NTB=Hi-z
Bit 1 :	ALARM_OCUR1_RES_EN		Alarm output Ena Alarm detection o current measurer	f discharging direc	ction	1: Enable 0: Dis ALARM_OCUR1_ ALARM_OCUR1_	RES_EN=1: OCUR1_RES=1		CUR1_RES=0 -> A	
Bit 0:	ALARM_OCUR1_DET_EN		Alarm output Ena Alarm detection o current measurer	f charging direction	n	1: Enable 0: Dis ALARM_OCUR1_ ALARM_OCUR1_	DET_EN=1: OCUR1_DET=1		CUR1_DET=0 -> A	

#### Address 3Bh: INT EN3 Register (R/W)

	Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
Ī	3Bh	INT_EN3	R/W	-	CALIB_FIN_EN	-	CRCERR_DET_EN	-	REX_DET_EN	WAKE_RES_EN	WAKE_DET_EN
	SBII	Initial Value	00h	0	0	0	0	0	0	0	0

Bit 6: CALIB\_FIN\_EN Interrupt Enable : 1: Enable 0: Disable

 $\label{eq:calib_fine} $$ CALIB_FIN=1 -> INTB=L, CALIB_FIN=0 -> INTB=Hi-z CALIB_FIN_EN=0: INTB=Hi-z (regardless of CALIB_FIN) $$$ Detection of calibration finish

Bit 5:

When writing this register, "0" must be written

Interrupt Enable : Detection of CRC Error CRCERR\_DET\_EN

1: Enable 0: Disable CRCERR\_DET=1 -> INTB=L, CRCERR\_DET=0 -> INTB=Hi-z CRCERR\_DET\_EN=0: INTB=Hi-z (regardless of CRCERR\_DET)

REX\_DET\_EN Bit 2 : Interrupt Enable :

1: Enable 0: Disable REX\_DET\_EN=1: REX\_DET=1 -> INTB=L, REX\_DET=0 -> INTB=Hi-z REX\_DET\_EN=0: INTB=Hi-z (regardless of REX\_DET) Detection of Relax state

Interrupt Enable : Detection of under wake-up current measurement Bit 1: WAKE RES EN

1: Enable 0: Disable WAKE\_RES\_EN=1: WAKE\_RES=1 -> INTB=L, WAKE\_RES=0-> INTB=Hi-z

WAKE\_RES\_EN=0: INTB=Hi-z (regardless of WAKE\_RES)

WAKE DET EN Interrupt Enable : Bit 0: 1: Enable 0: Disable

WAKE\_DET\_EN=1: WAKE\_DET=1 -> INTB=L, WAKE\_DET=0 -> INTB=Hi-z
WAKE\_DET\_EN=0: INTB=Hi-z (regardless of WAKE\_DET) Detection of over wake-up current measurement

#### Address 3Ch: INT REQ1 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
3Ch	INT_REQ1	R/W	CC_MON4_RES	CC_MON4_DET	CC_MON3_RES	CC_MON3_DET	CC_MON2_RES	CC_MON2_DET	CC_MON1_RES	CC_MON1_DET
3011	Initial Value	00h	0	0	0	0	0	0	0	0

CC MON4 RES Bit 7: Read:

Interrupt Status for Detection of

discharging direction current accumulation value4

1: Event occurred 0: No event

1: Clear 0: Not Clear

Bit is set to "1" when detect below condition  $CC\_CCNTD[31:8](17h+18h+19h+1A) \leq CC\_BATCAP4\_TH[24:0](2Ch+2Dh+2Eh)$ 

Write

Interrupt status clear

Read:

1: Event occurred 0: No event

Bit is set to "1" when detect below condition CC\_CCNTD[31:8](17h+18h+19h+1A)  $\leq$  CC\_BATCAP4\_TH[24:0](2Ch+2Dh+2Eh) Interrupt Status for Detection of charging direction current accumulation value4

Interrupt status clear

CC\_MON3\_RES Bit 5:

CC MON4 DET

Bit 6:

Interrupt Status for Detection of discharging direction current accumulation value3

Bit is set to "1" when detect below condition  $CC\_CCNTD[31:8](17h+18h+19h+1A) \le CC\_BATCAP3\_TH[24:0](29h+2Ah+2Bh)$ 

Interrupt status clear

1: Event occurred 0: No event

1: Event occurred 0: No event

CC MON3 DET Bit 4:

Read: Interrupt Status for Detection of

charging direction current accumulation value3

Bit is set to "1" when detect below condition CC\_CCNTD[31:8](17h+18h+19h+1A) > CC\_BATCAP3\_TH[24:0](29h+2Ah+2Bh)

Interrupt status clear

1: Clear 0: Not Clear 1: Event occurred 0: No event

CC MON2 RES Bit 3:

Read:

Interrupt Status for Detection of discharging direction current accumulation value2

Bit is set to "1" when detect below condition  $CC\_CCNTD[31:8](17h+18h+19h+1A) \leq CC\_BATCAP2\_TH[24:0](26h+27h+28h)$ 

Interrupt status clear

1: Clear 0: Not Clear

Bit 2: CC\_MON2\_DET

Interrupt Status for Detection of

1: Event occurred 0: No event

Bit is set to "1" when detect below condition CC\_CCNTD[31:8](17h+18h+19h+1A) > CC\_BATCAP2\_TH[24:0](26h+27h+28h)

charging direction current accumulation value2

Write Interrupt status clear 1: Clear 0: Not Clear

CC MON1 RES Bit 1:

1: Event occurred 0: No event Bit is set to "1" when detect below condition

Interrupt Status for Detection of discharging direction current accumulation value1

 $CC\_CCNTD[31:8](17h+18h+19h+1A) \le CC\_BATCAP1\_TH[24:0](23h+24h+25h)$ 

Interrupt status clear

1: Clear 0: Not Clear

CC\_MON1\_DET Bit 0:

Interrupt status clear

1: Event occurred 0: No event

Interrupt Status for Detection of charging direction current accumulation value1

Bit is set to "1" when detect below condition CC\_CCNTD[31:8](17h+18h+19h+1A) > CC\_BATCAP1\_TH[24:0](23h+24h+25h)

1: Clear 0: Not Clear

#### Address 3Dh: INT REQ2 Register (R/W)

Address (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
aDI-	INT_REQ2	R/W	OCUR3_RES	OCUR3_DET	OCUR2_RES	OCUR2_DET	OCUR1_RES	OCUR1_DET	-	-
3Dh	Initial Value	00h	0	0	0	0	0	0	0	0

OCUR3\_RES Read:

1: Event occurred 0: No event Interrupt Status for Detection of discharging direction current measurement3

over the consecutive detection setting by OCURDUR3[1:0](35h)

Write

Interrupt status clear

1: Clear 0: Not Clear

Bit 6: OCUR3\_DET

Interrupt Status for Detection of charging direction current measurement3

1: Event occurred 0: No event
Bit is set to "1" when detect below condition
CURCD\_DIR+CURCD(13h+14h) > OCURTHR3\_DIR+OCURTHR3(33h+34h) and
over the consecutive detection setting by OCURDUR3[1:0](35h)

Write

Interrupt status clear

1: Clear 0: Not Clear

1: Clear 0: Not Clear

1: Clear 0: Not Clear

OCUR2\_RES

1: Event occurred 0: No event Interrupt Status for Detection of

Bit is set to "1" when detect below condition discharging direction current measurement2

CURCD\_DIR+CURCD(13h+14h) ≤ OCURTHR2\_DIR+OCURTHR2(31h+32h) and over the consecutive detection setting by OCURDUR2[1:0](35h)

Write:

Interrupt status clear

1: Event occurred 0: No event

Bit 4: OCUR2 DET Read:

Bit 3:

Bit 2:

OCUR1\_RES

OCUR1 DET

Interrupt Status for Detection of

Bit is set to "1" when detect below condition CURCD\_DIR+CURCD(13h+14h) > OCURTHR2\_DIR+OCURTHR2(31h+32h) and charging direction current measurement2

over the consecutive detection setting by OCURDUR2[1:0](35h)

Write

Interrupt status clear

1: Event occurred 0: No event Interrupt Status for Detection of Bit is set to "1" when detect below condition

discharging direction current measurement1

 $\label{eq:curcon} $\text{CURCD\_DIR+CURCD}(13h+14h) \le \text{OCURTHR1\_DIR+OCURTHR1}(2Fh+30h)$ over the consecutive detection setting by $\text{OCURDUR1}[1:0](35h)$ }$ 

Interrupt status clear

1: Event occurred 0: No event

Interrupt Status for Detection of charging direction current measurement1 Bit is set to "1" when detect below condition CURCD\_DIR+CURCD(13h+14h) > OCURTHR1\_DIR+OCURTHR1(2Fh+30h) and

over the consecutive detection setting by OCURDUR1[1:0](35h)

Write:

Interrupt status clear

1: Clear 0: Not Clear

1: Clear 0: Not Clear

#### Address 3Eh: INT REQ3 Register (R/W)

	dress dex)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
	Eh	INT_REQ3	R/W	-	CALIB_FIN	OTP_DL_FIN	CRCERR_DET	-	REX_DET	WAKE_RES	WAKE_DET
3	E11	Initial Value	00h	0	0	0	0	0	0	0	0

Reserved

Bit 5:

Bit 1:

OTP\_DL\_FIN

CRCERR\_DET

WAKE\_RES

CALIB\_FIN

Interrupt Status for calibration finish

Write: 1: Clear 0: Not Clear

Interrupt status clear

Status for OTP data download finish

Write: Interrupt status clear

Interrupt status for CRC Error

Write:

Interrupt status clear

Bit 2: REX\_DET

Read: Interrupt status for Relax state

1: Event occurred 0: No event Bit is set to "1" when detect below condition CURCD[14:0](13h+14h)  $\leq$  { 7'd0, REX\_CURCD\_TH[7:0](36h) } and after detection time setting by REX\_DUR[1:0](35h)

Interrupt status clear

Write

Interrupt status for Detection of

1: Event occurred 0: No event Bit is set to "1" when detect below condition

 $\label{eq:curcol} $$ CURCD[14:0](13h+14h) \le \{ \ 7'd0, \ WAKE\_CURCD\_TH[7:0](37h) \} $$ and over the consecutive detection setting by $WAKE\_COUNT[1:0](38h) $$$ 

1: Clear 0: Not Clear

1: Clear 0: Not Clear

1: Event occurred 0: No event

1: Clear 0: Not Clear

1: Clear 0: Not Clear

1: Clear 0: Not Clear

Write: Interrupt status clear

WAKE\_DET Bit 0:

Interrupt status for Detection of

over wake-up current

Bit is set to "1" when detect below condition CURCD[14:0](13h+14h) > { 7'd0, WAKE\_CURCD\_TH[7:0](37h) } and over the consecutive detection setting by WAKE\_COUNT[1:0](38h)

Interrupt status clear

## Address 3Fh: PAGE\_SEL Register (R/W)

ddress (Index)	Register Name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
3Fh	PAGE_SEL	R/W	HOSC_ON	-	-	-	-	-	Rese	erved
SFII	Initial Value	00h	0	0	0	0	0	0	0	0

HOSC\_ON
0: Built-in OSC normal operation (default) Bit 7: Built-in OSC force enable

1: Built-in OSC turns ON by force

Force Built-in OSC to turn on when HOSC\_ON=1. Before transitioning from SSHDN mode to another mode, set HOSC\_ON = 1 and the built-in OSC will turn ON. When HOSC\_ON=0, built-in OSC operate normally. The built-in OSC will turn on in WAKE, OTP, IDLE, NORMAL, SLEEP modes. (Refer to Table 2-1)

Bit 1-0: Reserved[1:0]

When writing this register, always write "00" to these bits

## I/O Equivalence Circuit

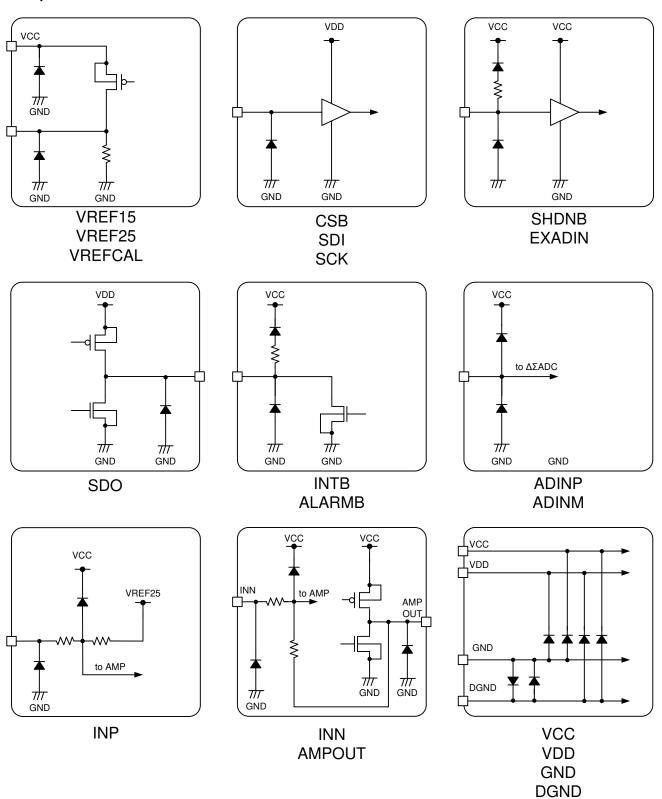
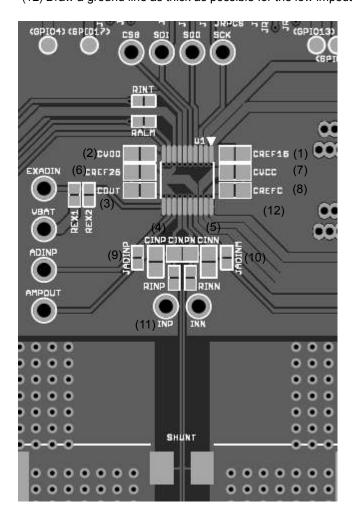


Figure 7. I/O Equivalence Circuit

## Layout

Optimum performance of this product cannot be achieved without taking the circuit board layout into consideration. The following points are important.

- (1) Place CREF15 as close as possible to the VREF15 pin and GND.
- (2) Place CVDD as close as possible to the VDD pin and GND.
- (3) Place COUT as close as possible to the AMPOUT pin and GND.
- (4) Place CINP as close as possible to the INP pin and GND.
- (5) Place CINN as close as possible to the INN pin and GND.
- (6) Place CREF25 as close as possible to the VREF25 pin and GND.
- (7) Place CVCC as close as possible to the VCC pin and GND.
- (8) Place CREFC as close as possible to the VREFCAL pin and GND.
- (9) Connect the ADINP pin and the AMPOUT pin as close as possible.
- (10) Connect the ADINM pin and the VREF25 pin as close as possible.
- (11) Draw a line the INP pin and the INN pin as equal as possible.
- (12) Draw a ground line as thick as possible for the low impedance.



## **Operational Notes**

#### 1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

#### 2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

#### 3. Ground Voltage

Except for pins the output and the input of which were designed to go below ground, ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

## 4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

## 5. Recommended Operating Conditions

The function and operation of the IC are guaranteed within the range specified by the recommended operating conditions. The characteristic values are guaranteed only under the conditions of each item specified by the electrical characteristics.

#### 6. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

#### 7. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

## 8. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

## 9. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

## **Operational Notes -continued**

## 10. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

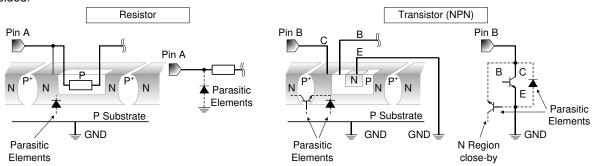
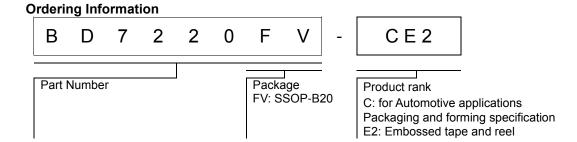


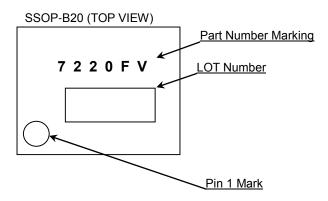
Figure 8. Example of Monolithic IC Structure

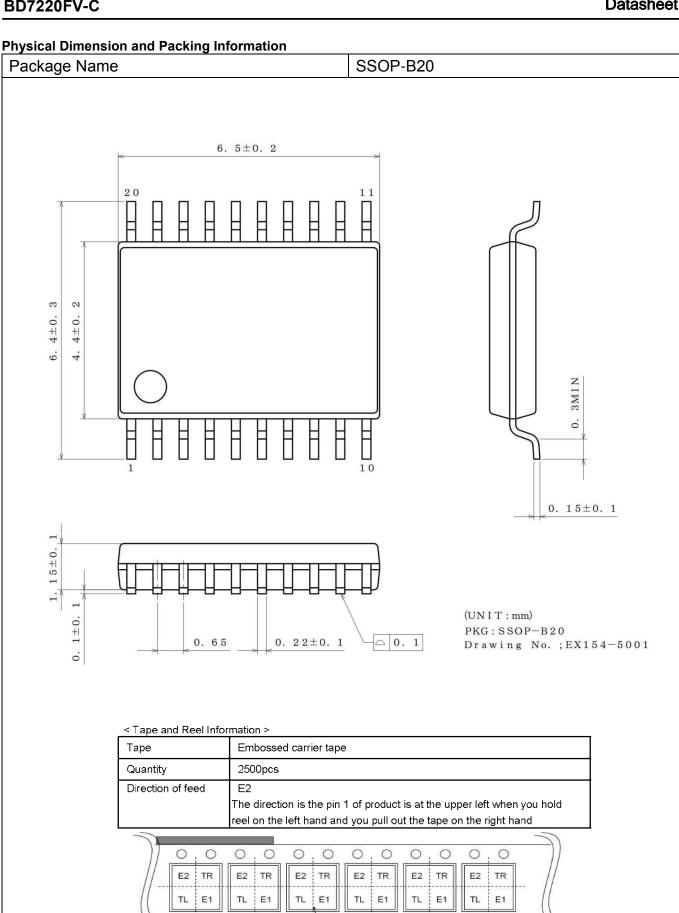
#### 11. Ceramic Capacitor

When using a ceramic capacitor, determine a capacitance value considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.



## **Marking Diagram**





Direction of feed

Pocket Quadrants

Reel

## **Revision History**

Date	Revision Number	Description
17. Dec. 2019	001	New Release
24. Sep. 2020	002	Page.16, Page.17 Bit Mask Function of Accumulation Current Corrected Figure 1-8 typo of "image of bits mask" and updated Figure 1-8. Add explanatory text. Add explanatory of example for CCNTD error complement.

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(Note1) Medical Equipment Classification of the Specific Applications

trotory medical Equipment diagonication of the opening approach			
JAPAN	USA	EU	CHINA
CLASSⅢ	CLASSⅢ	CLASS II b	CLASSIII
CLASSIV	CLASSIII	CLASSⅢ	

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  - [b] the temperature or humidity exceeds those recommended by ROHM
  - [c] the Products are exposed to direct sunshine or condensation
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