

EL2170C, EL2270C, EL2470C 70MHz/ImA Current Mode Feedback Amplifiers

Features

- Single (EL2170C), dual (EL2270C) and quad (EL2470C) topologies
- 1mA supply current (per amplifier)
- 70MHz -3dB bandwidth
- Tiny SOT23-5 Package (EL2170C)
- · Low cost
- Single- and dual-supply operation down to ±1.5V
- $0.15\%/0.15^{\circ}$ diff. gain/diff. phase into 150Ω
- 800V/µs slew rate
- Large output drive current:
 - 100mA (EL2170C)
 - 55mA (EL2270C)
 - 55mA (EL2470C)
- Also available with disable in single (EL2176C) and dual (EL2276C)
- Higher speed EL2180C/EL2186C family also available (3mA/ 250MHz) in single, dual and quad

Applications

- Low power/battery applications
- HDSL amplifiers
- Video amplifiers
- Cable drivers
- RGB amplifiers
- · Test equipment amplifiers
- Current to voltage converters

Ordering Information

Part No.	Package	Tape & Reel	Outline #
EL2170CN	PDIP-8	-	MDP0031
EL2170CS	SO-8		MDP0027
EL2170CS-T7	SO-8	7 in	MDP0027
EL2170CS-T13	SO-8	13 in	MDP0027
EL2170CW	SOT-23	- 1	MDP0038
EL2170CW-T7	SOT-23	7 in	MDP0038
EL2170CW-T13	SOT-23	13 in	MDP0038
EL2270CN	PDIP-8	-	MDP0031
EL2270CS	SO-8	-	MDP0027
EL2270CS-T7	SO-8	7 in	MDP0027
EL2270CS-T13	SO-8	13 in	MDP0027
EL2470CN	PDIP-14	-	MDP0031
EL2470CS	SO-14	-	MDP0027
EL2470CS-T7	SO-14	7 in	MDP0027
EL2470CS-T13	SO-14	13 in	MDP0027

General Description

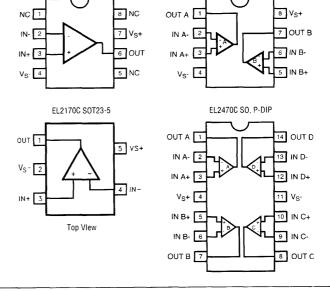
The EL2170C/EL2270C/EL2470C are single/dual/quad current-feed-back operational amplifiers which achieve a -3dB bandwidth of 70MHz at a gain of +1 while consuming only 1mA of supply current per amplifier. They will operate with dual supplies ranging from $\pm 1.5 \rm V$ to $\pm 6 \rm V$, or from single supplies ranging from $\pm 3 \rm V$ to $\pm 12 \rm V$. In spite of their low supply current, the EL2270C and the EL2470C can output 55mA while swinging to $\pm 4 \rm V$ on $\pm 5 \rm V$ supplies. The EL2170C can output 100mA with similar output swings. These attributes make the EL2170C/EL2270C/EL2470C excellent choices for low power and/or low voltage cable-driver, HDSL, or RGB applications.

For applications where board space is extremely critical, the EL2170C is available in the tiny 5-Lead SOT23 package, which has a footprint 28% the size of an 8-Lead SO. The EL2170C/EL2270C/EL2470C are each also available in industry standard pinouts in PDIP and SO packages.

For Single and Dual applications with disable, consider the EL2176C (8-Pin Single) or EL2276C (14-Pin Dual). For higher speed applications where power is still a concern, consider the EL2180C/El2186C family which also comes in similar Single, Dual, Triple and Quad configurations. The EL2180C/EL2186C family provides a -3dB bandwidth of 250MHz while consuming 3mA of supply current per amplifier.

EL2270C SO. P-DIP

Connection Diagrams EL2170C SO, P-DIP



70MHz/1mA Current Mode Feedback Amplifiers

Absolute Maximum Ratings (T_A = 25 °C)

Voltage between VS+ and VS-+12.6V Operating Junction Temperature

Common-Mode Input Voltage Plastic Packages V_{S-} to V_{S+} 150°C Differential Input Voltage ±6V Output Current (EL2170C) ±120mA Current into +IN or -IN ±7.5mA Output Current (EL2270C) ±60mA Internal Power Dissipation Output Current (EL2470C) See Curves ±60mA

Operating Ambient Temperature Range

Storage Temperature Range

-65°C to +150°C

Important Note:

All parameters having Min/Max specifications are guaranteed. Typ values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_1 = T_C = T_A$

-40°C to +85°C

DC Electrical Characteristics

 V_S = ±5V, R_L = 150 $\!\Omega_{\!\scriptscriptstyle 1}$, T_A = 25 $^{\circ}C$ unless otherwise specified

Parameter	Description	Conditions	Min	Тур	Max	Units
V _{OS}	Input Offset Voltage			2.5	15	mV
TCV _{OS}	Average Input Offset Voltage Drift	Measured from T _{MIN} to T _{MAX}		5		μV/°C
dV _{OS}	V _{OS} Matching	EL2270C, EL2470C only		0.5		mV
+I _{IN}	+ Input Current			0.5	5	μA
d+I _{IN}	+I _{IN} Matching	EL2270C, EL2470C only		20		nA
-I _{IN}	- Input Current			4	15	μА
d-I _{IN}	-I _{IN} Matching	EL2270C, EL2470C only		1.5		μА
CMRR	Common Mode Rejection Ratio	$V_{CM} = \pm 3.5 V$	45	50		dB
-ICMR	- Input Current Common Mode Rejection	$V_{CM} = \pm 3.5 V$		4	10	μA/V
PSRR	Power Supply Rejection Ratio	V _S is moved from ±4V to ±6V	60	70		dB
-IPSR	- Input Current Power Supply Rejection	V _S is moved from ±4V to ±6V		0.5	5	μA/V
R _{OL}	Transimpedance	$V_{OUT} = \pm 2.5V$	150	400		kΩ
+R _{IN}	+ Input Resistance	$V_{CM} = \pm 3.5 V$	1	4		MΩ
+C _{IN}	+ Input Capacitance			1.2		pF
CMIR	Common Mode Input Range		±3.5	±4.0		ν
$\overline{V_{O}}$	Output Voltage Swing	$V_S = \pm 5$	±3.5	±4.0		· v
		$V_S = +5$ Single-Supply, High		4.0		V
		$V_S = +5$ Single-Supply, Low		0.3		v
Io	Output Current	EL2170C only	80	100		mA
		EL2270C only, per Amplifier	50	55		mA
		EL2470C only, per Amplifier	50	55		mA
I _S	Supply Current	Per Amplifier		ı	2	mA

AC Electrical Characteristics

 $V_S = \pm 5V$, $R_F = R_G = 1k\Omega$, $R_L = 150\Omega$, $T_A = 25^{\circ}C$ unless otherwise specified

Parameter	Description	Conditions	Min	Тур	Max	Units
-3dB BW	-3dB Bandwidth	$A_V = +1$		70		MHz
-3dB BW	-3dB Bandwidth	$A_V = +2$		60		MHz
SR	Slew Rate	$V_{OUT} = \pm 2.5 \text{V}, A_{V} = +2$	400	800		V/µs
t _r , t _f	Rise and Fall Time	$V_{OUT} = \pm 500 \text{mV}$		4.5		ns
t _{pd}	Propagation Delay	$V_{OUT} = \pm 500 \text{mV}$		4.5		ns

70MHz/1mA Current Mode Feedback Amplifiers

AC Electrical Characteristics

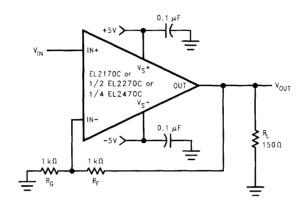
 V_S = ±5V, R_F = R_G = 1k Ω , R_L = 150 Ω , T_A = 25 $^{\circ}C$ unless otherwise specified

Parameter	Description	Conditions	Min	Тур	Max	Units
os	Overshoot	$V_{OUT} = \pm 500 \text{mV}$		3.0		%
t _s	0.1% Settling	$V_{OUT} = \pm 2.5V, A_V = -1$		40		ns
dG	Differential Gain	$A_V = +2, R_L = 150\Omega^{[1]}$		0.15		%
dP	Differential Phase	$A_V = +2, R_L = 150\Omega^{[1]}$		0.15		۰
dG	Differential Gain	$A_V = +1, R_L = 500\Omega^{[1]}$		0.02		%
dP	Differential Phase	$A_V = +1, R_L = 500\Omega^{[1]}$		0.01		۰
C _S	Channel Separation	EL2270C, EL2470C only, f = 5MHz		85		dB

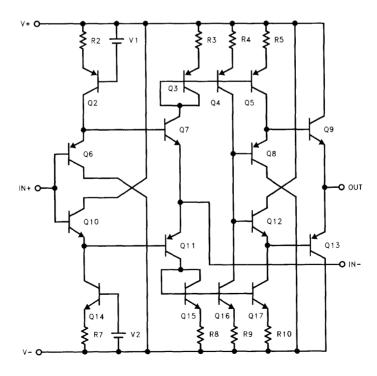
^{1.} DC offset from 0V to 0.714V, AC amplitude $286mV_{P-P}$, f = 3.58MHz.

70MHz/1mA Current Mode Feedback Amplifiers

Test Circuit (per Amplifier)

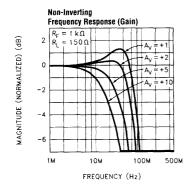


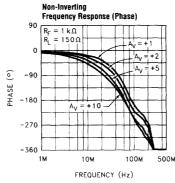
Simplified Schematic (per Amplifier)

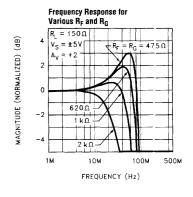


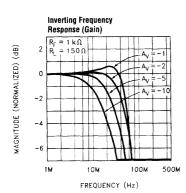
70MHz/1mA Current Mode Feedback Amplifiers

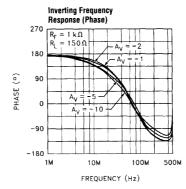
Typical Performance Curves

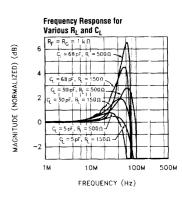


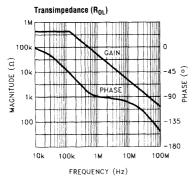


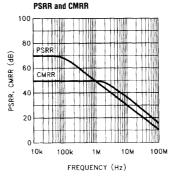


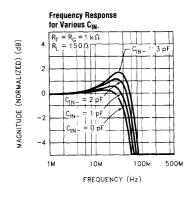




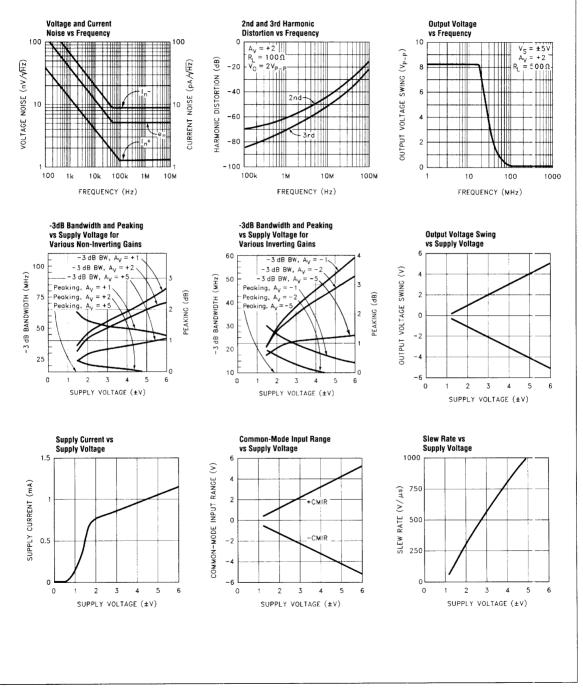




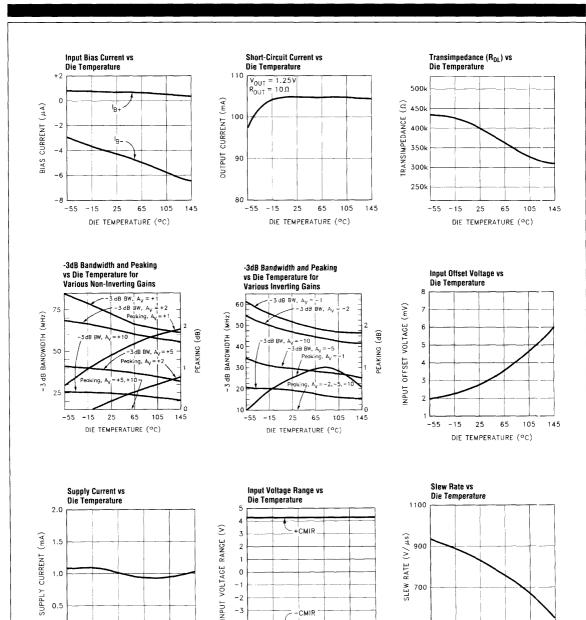




70MHz/1mA Current Mode Feedback Amplifiers



70MHz/1mA Current Mode Feedback Amplifiers



-CMIR

DIE TEMPERATURE (°C)

65

105 145 -55

25

DIE TEMPERATURE (°C)

105 145

- 1 -2 -3

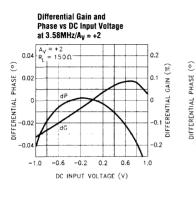
> -55 -15

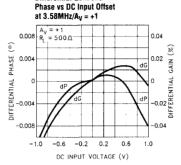
-55

25 65 105 145

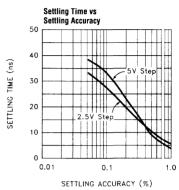
DIE TEMPERATURE (°C)

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Differential Gain and

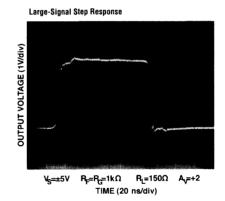


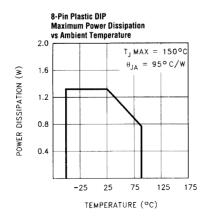
Small-Signal Step Response

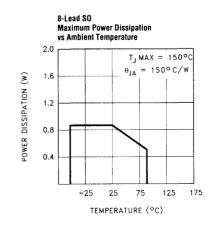
(λp)

V_S=±5V R_F=R_C=1kΩ R_L=150Ω A_F=+2

TIME (10 ne/div)

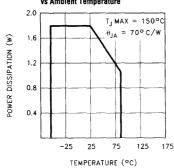




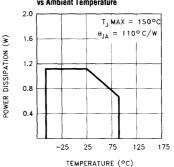


70MHz/1mA Current Mode Feedback Amplifiers

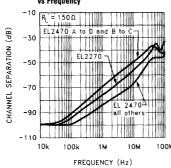




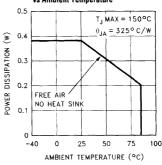
14-Lead SO Maximum Power Dissipation vs Ambient Temperature



Channel Separation vs Frequency



5-Lead Plastic SOT23 Maximum Power Dissipation vs Ambient Temperature



70MHz/1mA Current Mode Feedback Amplifiers

Applications Information

Product Description

The EL2170C/EL2270C/EL2470C are current-feedback operational amplifiers that offer a wide -3dB bandwidth of 70MHz and a low supply current of 1mA per amplifier. All of these products also feature high output current drive. The EL2170C can output 100mA, while the EL2270C and the EL2470C can output 55mA per amplifier. The EL2170C/EL2270C/EL2470C work with supply voltages ranging from a single 3V to ±6V, and they are also capable of swinging to within 1V of either supply on the input and the output. Because of their current-feedback topology, the EL2170C/EL2270C/ EL2470C do not have the normal gain-bandwidth product associated with voltage-feedback operational amplifiers. This allows their -3dB bandwidth to remain relatively constant as closed-loop gain is increased. This combination of high bandwidth and low power, together with aggressive pricing make the EL2170C/EL2270C/ wEL2470C the ideal choice for many low-power/highbandwidth applications such as portable computing, HDSL, and video processing.

For applications where board space is extremely critical, the EL2170C is available in the tiny 5-Lead SOT23 package, which has a footprint 28% the size of an 8-Lead SO. The EL2170C/EL2270C/EL2470C are each also available in industry standard pinouts in PDIP and SO packages.

For Single and Dual applications with disable, consider the EL2176C (8-Pin Single) and EL2276C (14-Pin Dual). If higher speed is required, refer to the EL2180C/EL2186C family which provides Singles, Duals, Triples, and Quads with 250MHz of bandwidth while consuming 3mA of supply current per amplifier.

Power Supply Bypassing and Printed Circuit Board Layout

As with any high-frequency device, good printed circuit board layout is necessary for optimum performance. Ground plane construction is highly recommended. Lead lengths should be as short as possible. The power supply pins must be well bypassed to reduce the risk of oscillation. The combination of a 4.7µF tantalum capac-

itor in parallel with a 0.1µF capacitor has been shown to work well when placed at each supply pin.

For good AC performance, parasitic capacitance should be kept to a minimum especially at the inverting input (see the Capacitance at the Inverting Input section). Ground plane construction should be used, but it should be removed from the area near the inverting input to minimize any stray capacitance at that node. Carbon or Metal-Film resistors are acceptable with the Metal-Film resistors giving slightly less peaking and bandwidth because of their additional series inductance. Use of sockets, particularly for the SO package should be avoided if possible. Sockets add parasitic inductance and capacitance which will result in some additional peaking and overshoot.

Capacitance at the Inverting Input

Any manufacturer's high-speed voltage- or current-feed-back amplifier can be affected by stray capacitance at the inverting input. For inverting gains this parasitic capacitance has little effect because the inverting input is a virtual ground, but for non-inverting gains this capacitance (in conjunction with the feedback and gain resistors) creates a pole in the feedback path of the amplifier. This pole, if low enough in frequency, has the same destabilizing effect as a zero in the forward openloop response. The use of large value feedback and gain resistors further exacerbates the problem by further lowering the pole frequency.

The experienced user with a large amount of PC board layout experience may find in rare cases that the EL2170C/EL2270C/EL2470C have less bandwidth than expected. The reduction of feedback resistor values (or the addition of a very small amount of external capacitance at the inverting input, e.g.0.5pF) will increase bandwidth as desired. Please see the curves for Frequency Response for Various $R_{\rm F}$ and $R_{\rm G}$, and Frequency Response for Various $C_{\rm IN}$.

Feedback Resistor Values

The EL2170C/EL2270C/EL22470C have been designed and specified at gains of +1 and +2 with $R_F=1 \mathrm{k}\Omega$. This value of feedback resistor gives 70MHz of -3dB band-

70MHz/1mA Current Mode Feedback Amplifiers

width at $A_V = +1$ with about 1.5dB of peaking, and 60MHz of -3dB bandwidth at $A_V = +2$ with about 0.5dB of peaking. Since the EL2170C/EL2270C/EL2470C are current-feedback amplifiers, it is also possible to change the value of R_F to get more bandwidth. As seen in the curve of Frequency Response For Various R_F and R_G , bandwidth and peaking can be easily modified by varying the value of the feedback resistor.

Because the EL2170C/EL2270C/EL2470C are current-feedback amplifiers, their gain-bandwidth product is not a constant for different closed-loop gains. This feature actually allows the EL2170C/EL2270C/EL2470C to maintain about the same -3dB bandwidth, regardless of closed-loop gain. However, as closed-loop gain is increased, bandwidth decreases slightly while stability increases. Since the loop stability is improving with higher closed-loop gains, it becomes possible to reduce the value of $R_{\rm F}$ below the specified $1 {\rm k} \Omega$ and still retain stability, resulting in only a slight loss of bandwidth with increased closed-loop gain.

Supply Voltage Range and Single-Supply Operation

The EL2170C/EL2270C/EL2470C have been designed to operate with supply voltages having a span of greater than 3V, and less than 12V. In practical terms, this means that the EL2170C/EL2270C/EL2470C will operate on dual supplies ranging from $\pm 1.5V$ to $\pm 6V$. With a single-supply, the EL2170C/EL2270C/EL2470C will operate from $\pm 3V$ to $\pm 12V$.

As supply voltages continue to decrease, it becomes necessary to provide input and output voltage ranges that can get as close as possible to the supply voltages. The EL2170C/EL2270C/EL2470C have an input voltage range that extends to within 1V of either supply. So, for example, on a single +5V supply, the EL2170C/EL2270C/EL2470C have an input range which spans from 1V to 4V. The output range of the EL2170C/EL2270C/EL2470C is also quite large, extending to within 1V of the supply rail. On a ±5V supply, the output is therefore capable of swinging from -4V to +4V. Single-supply output range is even larger because of the increased negative swing due to the external pull-down resistor to ground. On a single +5V supply, output voltage range is about 0.3V to 4V.

Video Performance

For good video performance, an amplifier is required to maintain the same output impedance and the same frequency response as DC levels are changed at the output. This is especially difficult when driving a standard video load of 150Ω , because of the change in output current with DC level. Until the EL2170C/EL2270C/EL2470C. good Differential Gain could only be achieved by running high idle currents through the output transistors (to reduce variations in output impedance). These currents were typically more than the entire 1mA supply current of each EL2170C/EL2270C/EL2470C amplifier! Special circuitry has been incorporated in the EL2170C/EL2270C/EL2470C to reduce the variation of output impedance with current output. This results in dG and dP specifications of 0.15% and 0.15° while driving 150Ω at a gain of +2.

Video Performance has also been measured with a 500Ω load at a gain of +1. Under these conditions, the EL2170C/EL2270C/EL2470C have dG and dP specifications of 0.01% and 0.02° respectively while driving 500Ω at A_V = +1.

Output Drive Capability

In spite of its low ImA of supply current, the EL2170C is capable of providing a minimum of $\pm 80\text{mA}$ of output current. Similarly, each amplifier of the EL2270C and the EL2470C is capable of providing a minimum of $\pm 50\text{mA}$. These output drive levels are unprecedented in amplifiers running at these supply currents. With a minimum $\pm 80\text{mA}$ of output drive, the EL2170C is capable of driving 50Ω loads to $\pm 4\text{V}$, making it an excellent choice for driving isolation transformers in telecommunications applications. Similarly, the $\pm 50\text{mA}$ minimum output drive of each EL2270C and EL2470C amplifier allows swings of $\pm 2.5\text{V}$ into 50Ω loads.

Driving Cables and Capacitive Loads

When used as a cable driver, double termination is always recommended for reflection-free performance. For those applications, the back-termination series resistor will decouple the EL2170C/EL2270C/EL2470C from the cable and allow extensive capacitive drive. However, other applications may have high capacitive loads without a back-termination resistor. In these appli-

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cations, a small series resistor (usually between 5Ω and 50Ω) can be placed in series with the output to eliminate most peaking. The gain resistor (R_G) can then be chosen to make up for any gain loss which may be created by this additional resistor at the output. In many cases it is also possible to simply increase the value of the feedback resistor (R_F) to reduce the peaking.

Current Limiting

The EL2170C/EL2270C/EL2470C have no internal current-limiting circuitry. If any output is shorted, it is possible to exceed the Absolute Maximum Ratings for output current or power dissipation, potentially resulting in the destruction of the device.

Power Dissipation

With the high output drive capability of the EL2170C/EL2270C/EL2470C, it is possible to exceed the 150°C Absolute Maximum junction temperature under certain very high load current conditions. Generally speaking, when R_L falls below about 25 Ω , it is

important to calculate the maximum junction temperature ($T_{\rm JMAX}$) for the application to determine if power-supply voltages, load conditions, or package type need to be modified for the EL2170C/EL2270C/EL2470C to remain in the safe operating area. These parameters are calculated as follows:

$$T_{JMAX} = T_{MAX} + (\theta_{JA} * n * PD_{MAX}) [1]$$

Where:

- T_{MAX}=Maximum Ambient Temperature
- θ_{JA} =Thermal Resistance of the Package
- n=Number of Amplifiers in the Package
- PD_{MAX}=Maximum Power Dissipation of Each Amplifier in the Package.

PD_{MAX} for each amplifier can be calculated as follows:

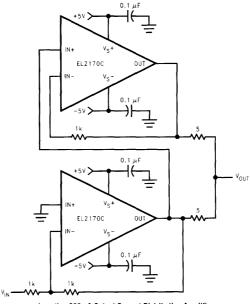
$$PD_{MAX} = (2 * V_S * I_{SMAX}) + (V_S - V_{OUTMAX}) * (V_{OUTMAX}/R_{\odot}))$$
 [2]

Where:

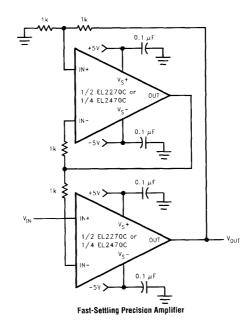
- V_S=Supply Voltage
- I_{SMAX}=Maximum Supply Current of 1 Amplifier
- V_{OUTMAX}=Max. Output Voltage of the Application
- R₁=Load Resistance

70MHz/1mA Current Mode Feedback Amplifiers

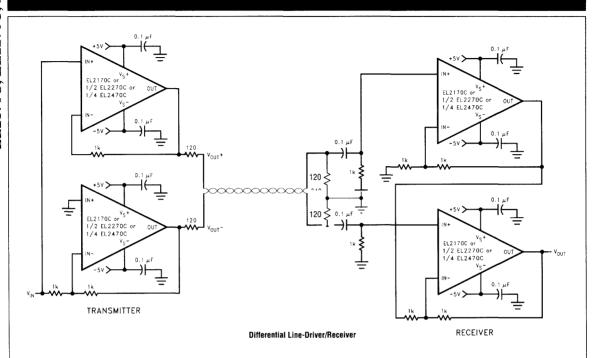
Typical Application Circuits



Inverting 200mA Output Current Distribution Amplifier



70MHz/1mA Current Mode Feedback Amplifiers



70MHz/1mA Current Mode Feedback Amplifiers

EL2170C/EL2270C/EL2470C Macromodel

-Vsupply output

```
* Revision A, March 1995
* AC characteristics used Rf=Rg=1K\Omega, RL=150\Omega
* Connections:
                 +input
                         -input
                              +Vsupply
.subckt EL2170/el 3
* Input Stage
e1 10 0 3 0 1.0
vis 10 9 0 V
h2 9 12 vxx 1.0
r1 2 11 165
11 11 12 25 nH*
iinp 3 0 0.5 uA
iinm 2 0 4 uA*
r12 3 0 4 Meg
* Slew Rate Limiting
h1 13 0 vis 600
r2 13 14 1 K
d1 14 0 dclamp
d2 0 14 dclamp
* High Frequency Pole
e2 30 0 14 0 0.00166666666
13 30 17 0.5 uH
c5 17 0 0.69 pF
r5 17 0 300
* Transimpedance Stage
g1 0 18 17 0 1.0
rol 18 0 400 K
cdp 18 0 1.9 pF
* Output Stage
q1 4 18 19 qp
q2 7 18 20 qn
q3 7 19 21 qn
q4 4 20 22 qp
r7 21 6 4
r8 22 6 4
ios1 7 19 0.4 mA
ios2 20 4 0.4 mA
* Supply Current
ips 7 4 1 nA
* Error Terms
ivos 0 23 2 mA
vxx 23 0 0 V
e4 24 0 3 0 1.0
e5 25 0 7 0 1.0
e6 26 0 4 0 -1.0
```

r9 24 23 0.316 K

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```
r10 25 23 3.2 K
r11 26 23 3.2 K

*

* Models

*
.model qn npn(is=5e-15 bf=200 tf=0.01nS)
.model qp pnp(is=5e-15 bf=200 tf=0.01nS)
.model dclamp d(is=1e-30 ibv=0.266
+ bv=1.3v n=4)
```

EL2170C/EL2270C/EL2470C Macromodel

