# 6A Austin MicroLynx II<sup>TM</sup>: SIP Non-Isolated DC-DC Power Module

2.4Vdc -5.5Vdc input; 0.75Vdc to 3.63Vdc output; 6A Output Current



**EZ-SEQUENCE**<sup>TM</sup>

### **Applications**

- Distributed power architectures
- Intermediate bus voltage applications
- Telecommunications equipment
- Servers and storage applications
- Networking equipment
- Enterprise Networks
- Latest generation IC's (DSP, FPGA, ASIC) and Microprocessor powered applications

#### **Features**

- Compliant to RoHS EU Directive 2011/65/EU and amended Directive (EU) 2015/863
- Compliant to REACH Directive (EC) No 1907/2006
- Flexible output voltage sequencing EZ-SEQUENCE™
- Delivers up to 6A output current
- High efficiency 96% at 3.3V full load (V<sub>IN</sub> = 5.0V)
- Small size and low profile:

25.4 mm x 12.7mm x 6.68 mm (1.00 in x 0.5 in x 0.263 in)

- Low output ripple and noise
- High Reliability:

Calculated MTBF = 12.8M hours at 25°C Full-load

- Programmable Output voltage
- Line Regulation: 0.3% (typical)
- Load Regulation: 0.4% (typical)
- Temperature Regulation: 0.4 % (typical)
- Remote On/Off
- Output overcurrent protection (non-latching)
- Wide operating temperature range (-40°C to 85°C)
- ANSI/UL\* 62368-1 and CAN/ CSA<sup>†</sup> C22.2 No. 62368-1 Recognized, DIN VDE<sup>‡</sup> 0868-1/A11:2017 (EN62368-1:2014/A11:2017)
- ISO\*\* 9001 and ISO 14001 certified manufacturing facilities

### **Description**

Austin MicroLynx<sup>TM</sup> II SIP power modules are non-isolated dc-dc converters that can deliver up to 6A of output current with full load efficiency of 96.0% at 3.3V output. These modules provide a precisely regulated output voltage programmable via an external resistor from 0.75Vdc to 3.63Vdc over a wide range of input voltage ( $V_{IN} = 2.4 - 5.5$ Vdc). Austin MicroLynx<sup>TM</sup> II has a sequencing feature, EZ-SEQUENCE<sup>TM</sup> that enable designers to implement various types of output voltage sequencing when powering multiple modules on board. Their open-frame construction and small footprint enable designers to develop costand space-efficient solutions. In addition to sequencing, standard features include remote On/Off, programmable output voltage and over current protection.

- $^{st}$  UL is a registered trademark of Underwriters Laboratories, Inc.
- <sup>†</sup> CSA is a registered trademark of Canadian Standards Association.
- <sup>‡</sup> VDE is a trademark of Verband Deutscher Elektrotechniker e.V.
- $\ensuremath{^{**}}$  ISO is a registered trademark of the International Organization of Standards



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### **Absolute Maximum Ratings**

Stresses in excess of the absolute maximum ratings can cause permanent damage to the device. These are absolute stress ratings only, functional operation of the device is not implied at these or any other conditions in excess of those given in the operations sections of the data sheet. Exposure to absolute maximum ratings for extended periods can adversely affect the device reliability.

Parameter	Device	Symbol	Min	Max	Unit
Input Voltage	All	V <sub>IN</sub>	-0.3	5.8	Vdc
Continuous					
Sequencing voltage	All	Vseq	-0.3	$V_{IN,max}$	Vdc
Operating Ambient Temperature	All	T <sub>A</sub>	-40	85	°C
(see Thermal Considerations section)					
Storage Temperature	All	$T_{stg}$	-55	125	°C

### **Electrical Specifications**

Unless otherwise indicated, specifications apply over all operating input voltage, resistive load, and temperature conditions.

Parameter	Device	Symbol	Min	Тур	Max	Unit
Operating Input Voltage	Vo,set ≤ 3.63	V <sub>IN</sub>	2.4	_	5.5	Vdc
Maximum Input Current	All	I <sub>IN,max</sub>				Adc
$(V_{IN}=V_{IN, min} \text{ to } V_{IN, max}, I_O=I_{O, max})$					6.0	
Input No Load Current	V <sub>O,set</sub> = 0.75 Vdc	I <sub>IN,No</sub> load		20		mA
$(V_{IN} = V_{IN, nom}, Io = 0, module enabled)$	V <sub>O,set</sub> = 3.3Vdc	I <sub>IN,No</sub> load		45		mA
Input Stand-by Current	All	I <sub>IN,stand-by</sub>		0.6		mA
$(V_{IN} = V_{IN, nom}, module disabled)$						
Inrush Transient	All	l²t			0.04	A <sup>2</sup> s
Input Reflected Ripple Current, peak-to-peak (5Hz to 20MHz, $1\mu$ H source impedance; $V_{IN, min}$ to $V_{IN, max}$ , $I_{O}$ = $I_{Omax}$ ; See Test configuration section)	All			35		mАр-р
Input Ripple Rejection (120Hz)	All			30		dB

### CAUTION: This power module is not internally fused. An input line fuse must always be used.

This power module can be used in a wide variety of applications, ranging from simple standalone operation to being part of a complex power architecture. To preserve maximum flexibility, internal fusing is not included, however, to achieve maximum safety and system protection, always use an input line fuse. The safety agencies require a fast-acting fuse with a maximum rating of 6 A (see Safety Considerations section). Based on the information provided in this data sheet on inrush energy and maximum dc input current, the same type of fuse with a lower rating can be used. Refer to the fuse manufacturer's data sheet for further information.

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### **Electrical Specifications** (continued)

Parameter	Device	Symbol	Min	Тур	Max	Unit
Output Voltage Set-point	All	V <sub>O</sub> , set	-2.0	_	+2.0	% <b>V</b> <sub>O, set</sub>
(V <sub>IN</sub> = <sub>IN, min</sub> , I <sub>O</sub> =I <sub>O, max</sub> , T <sub>A</sub> =25°C)						
Output Voltage	All	V <sub>O, set</sub>	-3.0	_	+3.0	% V <sub>O, set</sub>
(Over all operating input voltage, resistive load, and temperature conditions until end of life)						
Adjustment Range Selected by an external resistor	All	Vo	0.7525		3.63	Vdc
Output Regulation						
Line ( $V_{IN}=V_{IN, min}$ to $V_{IN, max}$ )	All		_	0.3	_	% V <sub>O, set</sub>
Load ( $I_O=I_{O, min}$ to $I_{O, max}$ )	All		_	0.4	_	$\% V_{O, set}$
Temperature (T <sub>ref</sub> =T <sub>A, min</sub> to T <sub>A, max</sub> )	All		_	0.4	_	% <b>V</b> o, set
Output Ripple and Noise on nominal output						
( $V_{IN}$ = $V_{IN, nom}$ and $I_{O}$ = $I_{O, min}$ to $I_{O, max}$						
Cout = 1μF ceramic//10μFtantalum capacitors)						
RMS (5Hz to 20MHz bandwidth)	All		_	10	15	$mV_{rms}$
Peak-to-Peak (5Hz to 20MHz bandwidth)	All		_	40	50	$mV_{pk-pk}$
External Capacitance						
$ESR \ge 1 \ m\Omega$	All	C <sub>O, max</sub>	_	_	1000	μF
ESR ≥ 10 mΩ	All	C <sub>O, max</sub>	_	_	3000	μF
Output Current	All	I <sub>o</sub>	0		6	Adc
Output Current Limit Inception (Hiccup Mode )	All	I <sub>O, lim</sub>	_	220	_	% I。
$(V_O=90\% \text{ of } V_{O,set})$						
Output Short-Circuit Current	All	I <sub>O, s/c</sub>	_	2	_	Adc
(V <sub>0</sub> ≤250mV) ( Hiccup Mode )						
Efficiency	V <sub>O, set</sub> =	η		81.0		%
$V_{IN} = V_{IN. nom}$ , $T_A = 25$ °C	0 75Vdc V <sub>0, set</sub> = 1.2Vdc	η		87.0		%
Io=I <sub>O. max</sub> , V <sub>O</sub> = V <sub>O.set</sub>	V <sub>o,set</sub> = 1.5Vdc	η		89.0		%
	V <sub>o,set</sub> = 1.8Vdc	η		90.0		%
	V <sub>0.set</sub> = 2.5Vdc	η		93.0		%
	V <sub>0,set</sub> = 3.3Vdc	η		96.0		%
Switching Frequency	All	f <sub>sw</sub>	_	300	_	kHz
Dynamic Load Response						
$(dIo/dt=2.5A/\mu s; V_{IN} = V_{IN, nom}; T_A=25^{\circ}C)$	All	$V_{pk}$	_	130	_	mV
Load Change from Io= 50% to 100% of Io,max; 1μF ceramic// 10 μF tantalum						
Peak Deviation						
Settling Time (Vo<10% peak deviation)	All	t <sub>s</sub>	_	25	_	μs
(dIo/dt=2.5A/ $\mu$ s; V <sub>IN</sub> = V <sub>IN, nom</sub> ; T <sub>A</sub> =25°C)	All	$V_{pk}$	_	130	_	mV
Load Change from Io= 100% to 50%of Io,max: $1\mu F$ ceramic// $10~\mu F$ tantalum						
Peak Deviation						
Settling Time (Vo<10% peak deviation)	All	ts		25	_	μs

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### **Electrical Specifications** (continued)

Parameter	Device	Symbol	Min	Тур	Max	Unit
Dynamic Load Response						
(dlo/dt=2.5A/ $\mu$ s; V V <sub>IN</sub> = V <sub>IN, nom</sub> ; T <sub>A</sub> =25°C) Load Change from lo= 50% to 100% of lo,max; Co = 2x150 $\mu$ F polymer capacitors Peak Deviation	All	$V_{pk}$	_	50	_	mV
Settling Time (Vo<10% peak deviation)	All	ts	_	50	_	μs
(dIo/dt=2.5A/ $\mu$ s; V <sub>IN</sub> = V <sub>IN, nom</sub> ; T <sub>A</sub> =25°C) Load Change from Io= 100% to 50% of Io,max: Co = 2x150 $\mu$ F polymer capacitors	All	$V_{pk}$	_	50	_	mV
Peak Deviation Settling Time (Vo<10% peak deviation)	All	t <sub>s</sub>	_	50	_	μs

### **General Specifications**

Parameter	Min	Тур	Max	Unit
Calculated MTBF (I <sub>O</sub> =I <sub>O, max</sub> , T <sub>A</sub> =25°C)		Hours		
per Telecordia SR-332 Issue 1: Method 1 Case 3				
Weight	_	2.8 (0.1)	_	g (oz.)

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### **Feature Specifications**

Unless otherwise indicated, specifications apply over all operating input voltage, resistive load, and temperature conditions. See Feature Descriptions for additional information.

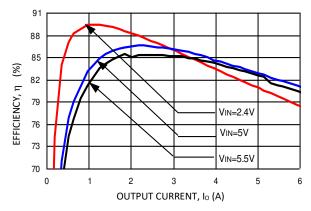
Parameter	Device	Symbol	Min	Тур	Max	Unit
On/Off Signal interface						
Device code with Suffix "4" – Positive logic						
(On/Off is open collector/drain logic input;						
Signal referenced to GND - See feature description section)						
Input High Voltage (Module ON)	All	ViH	_	_	$V_{\text{IN, max}}$	V
Input High Current	All	Іін	_	_	10	μΑ
Input Low Voltage (Module OFF)	All	VIL	-0.2	_	0.3	V
Input Low Current	All	lıL	_	0.2	1	mA
Device Code with no suffix – Negative Logic						
(On/OFF pin is open collector/drain logic input with						
external pull-up resistor; signal referenced to GND)						
Input High Voltage (Module OFF)	All	ViH	1.5	_	V <sub>IN,max</sub>	Vdc
Input High Current	All	Іін		0.2	1	mA
Input Low Voltage (Module ON)	All	VIL	-0.2	_	0.3	Vdc
Input low Current	All	lıL		_	10	μΑ
Turn-On Delay and Rise Times						
$(I_0=I_{0, max}, V_{IN}=V_{IN, nom}, T_A=25  {}^{\circ}C, )$						
Case 1: On/Off input is set to Logic Low (Module	All	Tdelay	_	3.9	_	msec
ON) and then input power is applied (delay from instant at which $V_{IN} = V_{IN, min}$ until $Vo = 10\%$ of $Vo, set$ )						
Case 2: Input power is applied for at least one second	All	Tdelay	_	3.9	_	msec
and then the On/Off input is set to logic Low (delay from						
instant at which Von/Off=0.3V until Vo=10% of Vo, set) Output voltage Rise time (time for Vo to rise from 10%	All	Trise	_	4.2	8.5	msec
of Vo, set to 90% of Vo, set)	All	11130		4.2	0.5	111366
Output voltage overshoot – Startup				_	1	% V <sub>O, set</sub>
I <sub>O</sub> = I <sub>O, max</sub> ; V <sub>IN</sub> = 2.4 to 5.5Vdc, T <sub>A</sub> = 25 °C						
Sequencing Delay time						
Delay from $V_{\text{IN},\text{min}}$ to application of voltage on SEQ pin	All	TSEQ-delay	10			msec
Tracking Accuracy (Power-Up: 2V/ms)	All	VSEQ -		100	200	mV
(Power-Down: 1V/ms)	All	VSEQ -		200	400	mV
( $V_{IN,  min}$ to $V_{IN,  max}$ ; $I_{O,  min}$ to $I_{O,  max}$ VSEQ < $V_{O}$ )						
Overtemperature Protection	All	$T_{ref}$	_	150	_	°C
(See Thermal Consideration section)						
Input Undervoltage Lockout						
Turn-on Threshold	All			2.2		V
Turn-off Threshold	All			2.0		V

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### **Characteristic Curves**

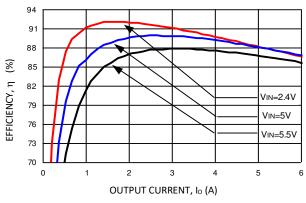
The following figures provide typical characteristics for the Austin MicroLynx™ II SIP modules at 25°C.



98 95 92 88 89 Vin=2.4V Vin=5.5V 77 74 0 1 2 3 4 5 6 OUTPUT CURRENT, Io (A)

Figure 1. Converter Efficiency versus Output Current (Vout = 0.75Vdc).

Figure 4. Converter Efficiency versus Output Current (Vout = 1.8Vdc).



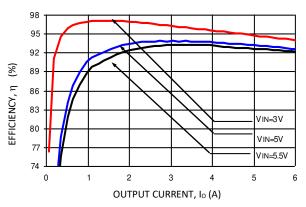
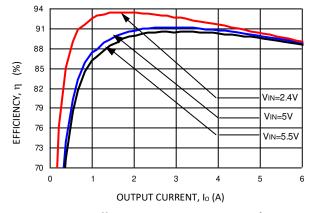


Figure 2. Converter Efficiency versus Output Current (Vout = 1.2Vdc).

Figure 5. Converter Efficiency versus Output Current (Vout = 2.5Vdc).



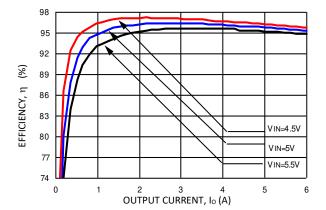


Figure 3. Converter Efficiency versus Output Current (Vout = 1.5Vdc).

Figure 6. Converter Efficiency versus Output Current (Vout = 3.3Vdc).

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### Characteristic Curves (continued)

The following figures provide typical characteristics for the MicroLynx™ II SIP modules at 25°C.

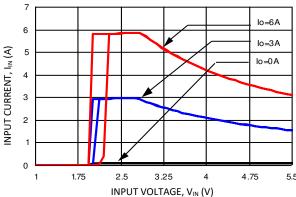


Figure 7. Input voltage vs. Input Current (Vout =2.5Vdc).

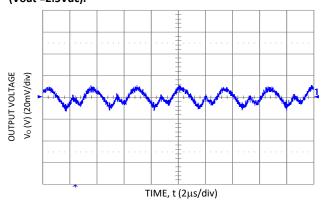


Figure 8. Typical Output Ripple and Noise (Vin = 5.0V dc, Vo = 0.75 Vdc, Io=6A).

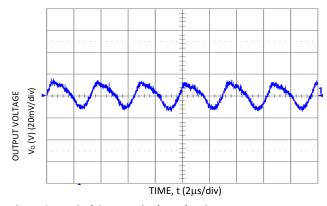


Figure 9. Typical Output Ripple and Noise (Vin = 5.0V dc, Vo = 3.3 Vdc, Io=6A).

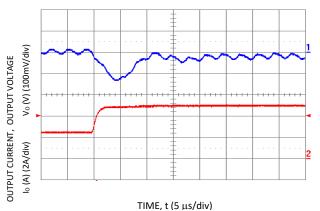


Figure 10. Transient Response to Dynamic Load Change from 50% to 100% of full load (Vo = 3.3Vdc).

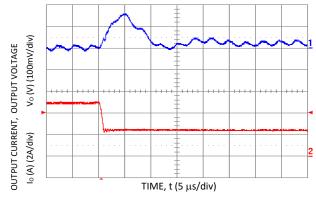


Figure 11. Transient Response to Dynamic Load Change from 100% to 50% of full load (Vo = 3.3 Vdc).

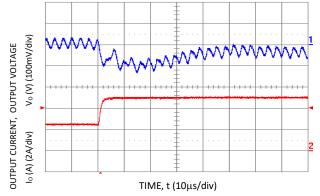


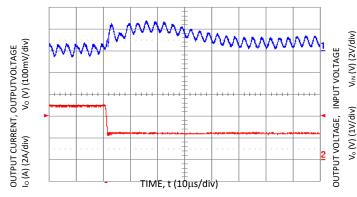
Figure 12. Transient Response to Dynamic Load Change from 50% to 100% of full load (Vo = 3.3 Vdc, Cext =  $2x150 \mu F$  Polymer Capacitors).

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### **Characteristic Curves** (continued)

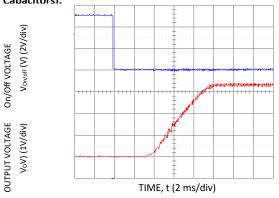
The following figures provide typical characteristics for the Austin MicroLynx™ II SIP modules at 25°C.



TIME, t (2 ms/div)

Figure 13. Transient Response to Dynamic Load Change from 100% of 50% full load (Vo = 3.3Vdc, Cext =  $2x150 \mu F$  Polymer

Figure 16. Typical Start-Up with application of Vin (Vin = 5.0Vdc, Vo = 3.3Vdc, Io = 6A).



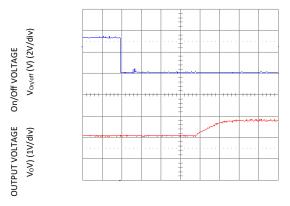
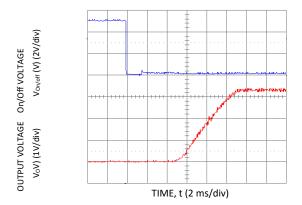


Figure 14. Typical Start-Up Using Remote On/Off (Vin = 5.0Vdc, Vo = 3.3Vdc, Io = 6A).

Figure 17 Typical Start-Up Using Remote On/Off with Prebias (Vin = 3.3Vdc, Vo = 1.8Vdc, Io = 1.0A, Vbias =1.0Vdc).



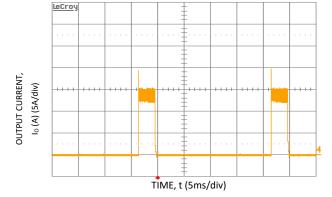


Figure 15. Typical Start-Up Using Remote On/Off with Low-ESR external capacitors (7x150uF Polymer) (Vin = 5.0Vdc, Vo = 3.3Vdc, Io = 6A, Co =  $1050\mu$ F).

Figure 18. Output short circuit Current (Vin = 5.0Vdc, Vo = 0.75Vdc).

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### Characteristic Curves (continued)

The following figures provide thermal derating curves for the Austin MicroLynx™ II SIP modules.

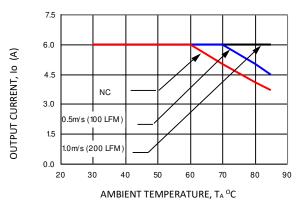


Figure 19. Derating Output Current versus Local Ambient Temperature and Airflow (Vin = 5.0, Vo=3.3Vdc).

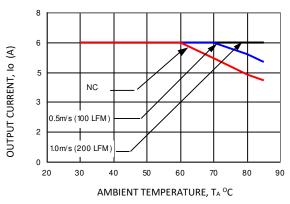


Figure 22. Derating Output Current versus Local Ambient Temperature and Airflow (Vin = 3.3dc, Vo=0.75 Vdc).

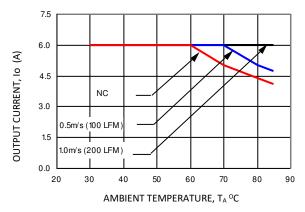


Figure 20. Derating Output Current versus Local Ambient Temperature and Airflow (Vin = 5.0Vdc, Vo=0.75 Vdc).

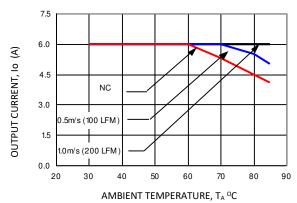
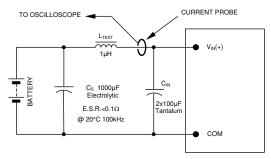


Figure 21. Derating Output Current versus Local Ambient Temperature and Airflow (Vin = 3.3Vdc, Vo=2.5 Vdc).

### **Test Configurations**

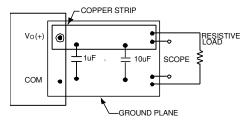
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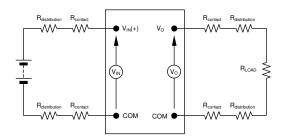
NOTE: Measure input reflected ripple current with a simulated source inductance ( $L_{TEST}$ ) of 1 $\mu$ H. Capacitor  $C_S$  offsets possible battery impedance. Measure current as shown above.

Figure 23. Input Reflected Ripple Current Test Setup.



NOTE: All voltage measurements to be taken at the module terminals, as shown above. If sockets are used then Kelvin connections are required at the module terminals to avoid measurement errors due to socket contact resistance.

Figure 24. Output Ripple and Noise Test Setup.



NOTE: All voltage measurements to be taken at the module terminals, as shown above. If sockets are used then Kelvin connections are required at the module terminals to avoid measurement errors due to socket contact resistance.

Figure 25. Output Voltage and Efficiency Test Setup.

Efficiency 
$$\eta = \frac{V_0. I_0}{V_{IN, IRM}} \times 100 \%$$

### **Design Considerations**

#### **Input Filtering**

The Austin MicroLynx<sup>TM</sup> II SIP module should be connected to a low-impedance source. A highly inductive source can affect the stability of the module. An input capacitance must be placed directly adjacent to the input pin of the module, to minimize input ripple voltage and ensure module stability.

To minimize input voltage ripple, low-ESR polymer and ceramic capacitors are recommended at the input of the module. Figure 26 shows the input ripple voltage (mVp-p) for various outputs with 1x150  $\mu\text{F}$  polymer capacitors (Panasonic p/n: EEFUE0J151R, Sanyo p/n: 6TPE150M) in parallel with 1 x 47  $\mu\text{F}$  ceramic capacitor (Panasonic p/n: ECJ-5YB0J476M, Taiyo- Yuden p/n: CEJMK432BJ476MMT) at full load. Figure 27 shows the input ripple with 2x150  $\mu\text{F}$  polymer capacitors in parallel with 2 x 47  $\mu\text{F}$  ceramic capacitor at full load.

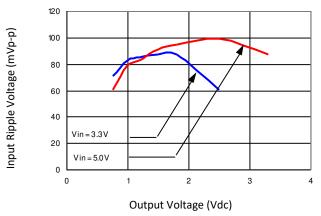


Figure 26. Input ripple voltage for various output with 1x150  $\mu$ F polymer and 1x47  $\mu$ F ceramic capacitors at the input (80% of lo,max).

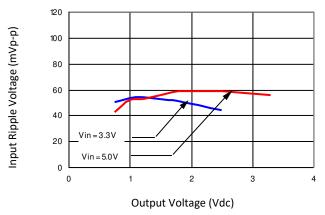


Figure 27. Input ripple voltage for various output with 2x150  $\mu$ F polymer and 2x47  $\mu$ F ceramic capacitors at the input (80% of Io,max).

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### **Design Considerations** (continued)

#### **Output Filtering**

The Austin MicroLynx<sup>TM</sup> II SIP module is designed for low output ripple voltage and will meet the maximum output ripple specification with 1  $\mu\text{F}$  ceramic and 10  $\mu\text{F}$  tantalum capacitors at the output of the module. However, additional output filtering may be required by the system designer for a number of reasons. First, there may be a need to further reduce the output ripple and noise of the module. Second, the dynamic response characteristics may need to be customized to a particular load step change.

To reduce the output ripple and improve the dynamic response to a step load change, additional capacitance at the output can be used. Low ESR polymer and ceramic capacitors are recommended to improve the dynamic response of the module. For stable operation of the module, limit the capacitance to less than the maximum output capacitance as specified in the electrical specification table.

### **Safety Considerations**

For safety agency approval the power module must be installed in compliance with the spacing and separation requirements of the end-use safety agency standards, i.e., UL ANSI/UL\* 62368-1 and CAN/CSA+ C22.2 No. 62368-1 Recognized, DIN VDE 0868-1/A11:2017 (EN62368-1:2014/A11:2017).

For the converter output to be considered meeting the Requirements of safety extra-low voltage (SELV) or ES1, the input must meet SELV/ES1 requirements. The power module has extra-low voltage (ELV) outputs when all inputs are ELV.

The input to these units is to be provided with a fast-acting fuse with a maximum rating of 6A in the positive input lead.

## 6A Austin MicroLynx<sup>TM</sup>II: SIP Non-Isolated DC-DC Power Modules

2.4Vdc -5.5Vdc input; 0.75Vdc to 3.63Vdc output; 6A Output Current

### **Feature Description**

#### **Remote On/Off**

The Austin Lynx<sup>™</sup> II SIP power modules feature an On/Off pin for remote On/Off operation. Two On/Off logic options are available in the Austin Lynx<sup>™</sup> II series modules. Positive Logic On/Off signal, device code suffix "4", turns the module ON during a logic High on the On/Off pin and turns the module OFF during a logic Low. Negative logic On/Off signal, no device code suffix, turns the module OFF during logic High on the On/Off pin and turns the module ON during logic Low.

For positive logic modules, the circuit configuration for using the On/Off pin is shown in Figure 28. The On/Off pin is an open collector/drain logic input signal (Von/Off) that is referenced to ground. During a logic-high (On/Off pin is pulled high internal to the module) when the transistor Q1 is in the Off state, the power module is ON. Maximum allowable leakage current of the transistor when Von/off =  $V_{IN,max}$  is  $10\mu A$ . Applying a logic-low when the transistor Q1 is turned-On, the power module is OFF. During this state VOn/Off must be less than 0.3V. When not using positive logic On/off pin, leave the pin unconnected or tie to  $V_{IN}$ .

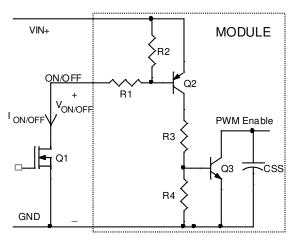


Figure 28. Circuit configuration for using positive logic On/OFF.

For negative logic On/Off devices, the circuit configuration is shown is Figure 29. The On/Off pin is pulled high with an external pull-up resistor (typical  $R_{\text{pull-up}} = 5k$ , +/- 5%). When transistor Q1 is in the Off state, logic High is applied to the On/Off pin and the power module is Off. The minimum On/off voltage for logic High on the On/Off pin is 1.5Vdc. To turn the module ON, logic Low is applied to the On/Off pin by turning ON Q1. When not using the negative logic On/Off, leave the pin unconnected or tie to GND.

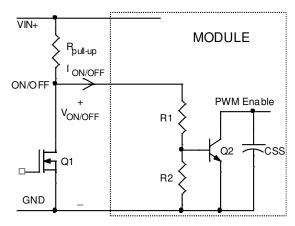


Figure 29. Circuit configuration for using negative logic On/OFF.

#### **Overcurrent Protection**

To provide protection in a fault (output overload) condition, the unit is equipped with internal current-limiting circuitry and can endure current limiting continuously. At the point of current-limit inception, the unit enters hiccup mode. The unit operates normally once the output current is brought back into its specified range. The typical average output current during hiccup is 2A.

#### **Input Undervoltage Lockout**

At input voltages below the input undervoltage lockout limit, module operation is disabled. The module will begin to operate at an input voltage above the undervoltage lockout turn-on threshold.

#### **Overtemperature Protection**

To provide over temperature protection in a fault condition, the unit relies upon the thermal protection feature of the controller IC. The unit will shutdown if the thermal reference point  $T_{\rm ref}$ , exceeds 150°C (typical), but the thermal shutdown is not intended as a guarantee that the unit will survive temperatures beyond its rating. The module will automatically restart after it cools down.

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### Feature Descriptions (continued)

#### **Output Voltage Programming**

The output voltage of the Austin MicroLynx™ II SIP can be programmed to any voltage from 0.75 Vdc to 3.3 Vdc by connecting a single resistor (shown as Rtrim in Figure 30) between the TRIM and GND pins of the module. Without an external resistor between TRIM pin and the ground, the output voltage of the module is 0.7525 Vdc. To calculate the value of the resistor *Rtrim* for a particular output voltage Vo, use the following equation:

$$Rtrim = \left[ \frac{21070}{Vo - 0.7525} - 5110 \right] \Omega$$

For example, to program the output voltage of the Austin MicroLynx<sup>TM</sup> II module to 1.8 Vdc, *Rtrim* is calculated is follows:

$$Rtrim = \left[ \frac{21070}{1.8 - 0.7525} - 5110 \right]$$

 $Rtrim = 15.004k\Omega$ 

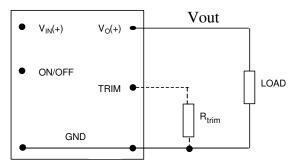


Figure 30. Circuit configuration to program output voltage using an external resistor.

Table 1 provides *Rtrim* values required for some common output voltages.

Table 1

V <sub>O, set</sub> (V)	Rtrim (KΩ)
0.7525	Open
1.2	41.973
1.5	23.077
1.8	15.004
2.5	6.947
3.3	3.160

By using a 1% tolerance trim resistor, set point tolerance of ±2% is achieved as specified in the electrical specification. The POL Programming Tool, available at <a href="https://www.gecriticalpower.com">www.gecriticalpower.com</a> under the Design Tools section, helps determine the required external trim resistor needed for a specific output voltage.

#### **Voltage Margining**

Output voltage margining can be implemented in the Austin MicroLynx  $^{TM}$  II modules by connecting a resistor,  $R_{margin-up}$ , from the Trim pin to the ground pin for margining-up the output voltage and by connecting a resistor,  $R_{margin-down}$ , from the Trim pin to the Output pin for margining-down. Figure 31 shows the circuit configuration for output voltage margining. The POL Programming Tool, available at www.gecriticalpower.com under the Design Tools section, also calculates the values of  $R_{margin-up}$  and  $R_{margin-down}$  for a specific output voltage and % margin. Please consult your local GE technical representative for additional details.

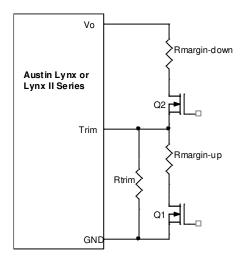


Figure 31. Circuit Configuration for margining Output voltage.

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### Feature Descriptions (continued)

#### **Voltage Sequencing**

Austin MicroLynx<sup>™</sup> II series of modules include a sequencing feature, EZ-SEQUENCE<sup>™</sup> that enables users to implement various types of output voltage sequencing in their applications. This is accomplished via an additional sequencing pin. When not using the sequencing feature, either tie the SEQ pin to V<sub>IN</sub> or leave it unconnected.

When an analog voltage is applied to the SEQ pin, the output voltage tracks this voltage until the output reaches the set-point voltage. The SEQ voltage must be set higher than the set-point voltage of the module. The output voltage follows the voltage on the SEQ pin on a one-to-one volt basis. By connecting multiple modules together, customers can get multiple modules to track their output voltages to the voltage applied on the SEQ pin.

For proper voltage sequencing, first, input voltage is applied to the module. The On/Off pin of the module is left unconnected (or tied to GND for negative logic modules or tied to VIN for positive logic modules) so that the module is ON by default. After applying input voltage to the module, a minimum of 10msec delay is required before applying voltage on the SEQ pin. During this time, potential of 50mV (± 10 mV) is maintained on the SEQ pin. After 10msec delay, an analog voltage is applied to the SEQ pin and the output voltage of the module will track this voltage on a one-to-one volt bases until output reaches the setpoint voltage. To initiate simultaneous shutdown of the modules, the SEQ pin voltage is lowered in a controlled manner. Output voltage of the modules tracks the voltages below their set-point voltages on a one-to-one basis. A valid input voltage must be maintained until the tracking and output voltages reach ground potential to ensure a controlled shutdown of the modules.

When using the EZ-SEQUENCE™ feature to control start-up of the module, pre-bias immunity feature during start-up is disabled. The pre-bias immunity feature of the module relies on the module being in the diode-mode during start-up. When using the EZ-SEQUENCE™ feature, modules goes through an internal set-up time of 10msec, and will be in synchronous rectification mode when voltage at the SEQ pin is applied. This will result in sinking current in the module if pre-bias voltage is present at the output of the module. When pre-bias immunity during start-up is required, the EZ-SEQUENCE™ feature must be disabled. For additional guidelines on using EZ-SEQUENCE™ feature of Austin MicroLynx™ II , contact GE technical representative for preliminary application note on output voltage sequencing using Austin Lynx II series.

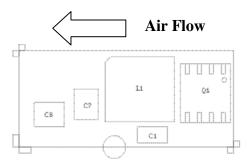
## 6A Austin MicroLynx<sup>TM</sup>II: SIP Non-Isolated DC-DC Power Modules

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#### **Thermal Considerations**

Power modules operate in a variety of thermal environments; however, sufficient cooling should be provided to help ensure reliable operation.

Considerations include ambient temperature, airflow, module power dissipation, and the need for increased reliability. A reduction in the operating temperature of the module will result in an increase in reliability. The thermal data presented here is based on physical measurements taken in a wind tunnel. The test set-up is shown in Figure 33. Note that the airflow is parallel to the long axis of the module as shown in figure 32. The derating data applies to airflow in either direction of the module's long axis.



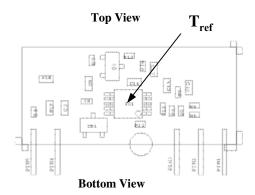


Figure 32. Tref Temperature measurement location.

The thermal reference point,  $T_{\text{ref 1}}$  used in the specifications of thermal derating curves is shown in Figure 32. For reliable operation this temperature should not exceed 125°C.

The output power of the module should not exceed the rated power of the module (Vo,set x Io,max).

Please refer to the Application Note "Thermal Characterization Process For Open-Frame Board-Mounted Power Modules" for a detailed discussion of thermal aspects including maximum device temperatures.

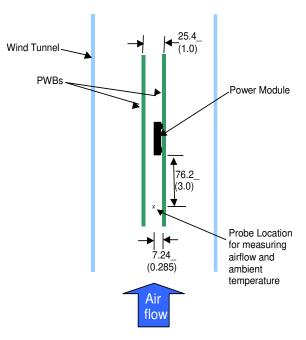


Figure 33. Thermal Test Set-up.

#### **Heat Transfer via Convection**

Increased airflow over the module enhances the heat transfer via convection. Thermal derating curves showing the maximum output current that can be delivered by various module versus local ambient temperature ( $T_A$ ) for natural convection and up to 1m/s (200 ft./min) are shown in the Characteristics Curves section.

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### **Post solder Cleaning and Drying Considerations**

Post solder cleaning is usually the final circuit-board assembly process prior to electrical board testing. The result of inadequate cleaning and drying can affect both the reliability of a power module and the testability of the finished circuit-board assembly. For guidance on appropriate soldering, cleaning and drying procedures, refer to Board Mounted Power Modules: Soldering and Cleaning Application Note.

### **Through-Hole Lead-Free Soldering Information**

The RoHS-compliant through-hole products use the SAC (Sn/Ag/Cu) Pb-free solder and RoHS-compliant components. They are designed to be processed through single or dual wave

soldering machines. The pins have an RoHS-compliant finish that is compatible with both Pb and Pb-free wave soldering processes. A maximum preheat rate of 3°C/s is suggested. The wave preheat process should be such that the temperature of the power module board is kept below 210°C. For Pb solder, the recommended pot temperature is 260°C, while the Pb-free solder pot is 270°C max. Not all RoHS-compliant through-hole products can be processed with paste-through-hole Pb or Pb-free reflow process. If additional information is needed, please consult with your GE technical representative for more details.

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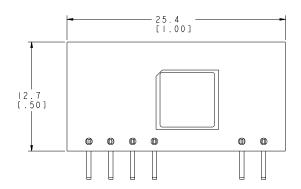
2.4Vdc -5.5Vdc input; 0.75Vdc to 3.63Vdc output; 6A Output Current

### **Mechanical Outline**

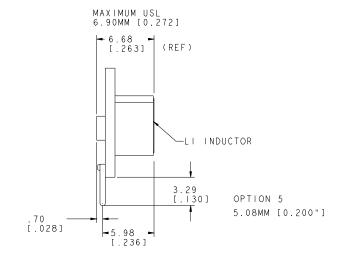
Dimensions are in millimeters and (inches).

Tolerances: x.x mm  $\pm$  0.5 mm (x.xx in.  $\pm$  0.02 in.) [unless otherwise indicated] x.xx mm  $\pm$  0.25 mm (x.xxx in  $\pm$  0.010 in.)

Top View

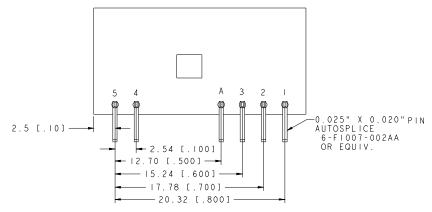


**Side View** 



### **Bottom View**

PIN	FUNCTION
1	Vo
2	Trim
3	GND
Α	SEQ
4	Vin
5	On/Off



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### **Recommended Pad Layout**

Dimensions are in millimeters and (inches).

Tolerances: x.x mm  $\pm$  0.5 mm (x.xx in.  $\pm$  0.02 in.) [unless otherwise indicated] x.xx mm  $\pm$  0.25 mm (x.xxx in  $\pm$  0.010 in.)

PIN FUNCTION

1 Vo

2 Trim

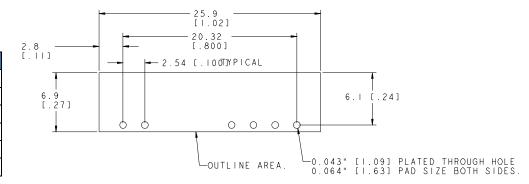
3 GND

A SEQ

4 Vin

On/Off

5



Through-Hole Pad Layout - Back view

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### **Ordering Information**

Please contact your GE Sales Representative for pricing, availability and optional features.

**Table 2. Device Codes** 

Device Code	Input Voltage	Output Voltage	Output Current	Efficiency 3.3V@ 6A	Connector Type	Comcodes
ATH006A0XZ	2.4 – 5.5Vdc	0.75 – 3.63Vdc	6 A	96.0%	SIP	CC109104717
ATH006A0X4Z	2.4 – 5.5Vdc	0.75 – 3.63Vdc	6 A	96.0%	SIP	CC109104725

<sup>-</sup>Z refers to RoHS-compliant versions.

# **Contact Us**

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