

1A, Low Voltage, Low Dropout Regulator with Reversed-Battery Protection

Features

- Fixed and Adjustable Output Voltages to 1.24V
- 410 mV Typical Dropout at 1A Load
	- Best Recommended for 3.0V to 2.5V Conversion
	- Best Recommended for 2.5V to 1.8V Conversion
- 1A Minimum Guaranteed Output Current
- 1% Initial Accuracy
- Low Ground Current
- Current-Limiting and Thermal-Shutdown Protection
- Reversed-Battery and Reversed-Leakage Protection
- Fast Transient Response
- Low Profile SOT-223 Package
- Power SO-8 Package

Applications

- LDO Linear Regulator for PC Add-In Cards
- High-Efficiency Linear Power Supplies
- SMPS Post Regulator
- Multimedia and PC Processor Supplies
- Battery Chargers
- Low Voltage Microcontrollers and Digital Logic

General Description

The MIC39100, MIC39101, and MIC39102 are 1A low dropout linear voltage regulators that provide low voltage, high current output from an extremely small package. The MIC39100/1/2 offers extremely low dropout (typically 410 mV at 1A) and low ground current (typically 11 mA at 1A).

The MIC39100 is a fixed output regulator offered in the SOT-223 package. The MIC39101 and MIC39102 are fixed and adjustable regulators, respectively, in a thermally enhanced 8-lead SOIC package.

The MIC39100/1/2 is ideal for PC add-in cards that need to convert from standard 5V to 3.3V, 3.3V to 2.5V, or 2.5V to 1.8V. A guaranteed maximum dropout voltage of 630 mV over all operating conditions allows the MIC39100/1/2 to provide 2.5V from a supply as low as 3.13V and 1.8V from a supply as low as 2.43V.

The MIC39100/1/2 is fully protected with overcurrent limiting, thermal-shutdown, and reverse-battery protection. Fixed voltages of 5.0V, 3.3V, 2.5V, and 1.8V are available on MIC39100/1 with adjustable output voltages to 1.24V on MIC39102.

Typical Application Circuits

Functional Block Diagrams

1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

Operating Ratings ‡

 \dagger Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational sections of this specification is not intended. Exposure to maximum rating conditions for extended periods may affect device reliability.

‡ Notice: The device is not guaranteed to function outside its operating ratings.

- **Note 1:** Devices are ESD sensitive. Handling precautions are recommended. Human body model, 1.5 kΩ in series with 100 pF.
	- **2:** $P_{D(MAX)} = (T_{J(MAX)} T_A) \div \theta_{JA}$, where θ_{JA} depends upon the printed circuit layout (see [Application Informa](#page-11-0)[tion](#page-11-0)).

TABLE 1-1: ELECTRICAL CHARACTERISTICS

Electrical Characteristics: V_{IN} = V_{OUT} + 1V; V_{EN} = 2.25V; T_J = +25°C, bold values indicate –40°C ≤ T_J ≤ +125°C, unless noted. [Note 1](#page-4-0)

TABLE 1-1: ELECTRICAL CHARACTERISTICS (CONTINUED)

Electrical Characteristics: V_{IN} = V_{OUT} + 1V; V_{EN} = 2.25V; T_J = +25°C, bold values indicate -40° C ≤ T_J ≤ +125°C, unless noted. Note 1

Note 1: Specification for packaged product only.

2: Output voltage temperature coefficient is $\Delta V_{\text{OUT(WORST CASE)}} \div (T_{\text{J(MAX)}} - T_{\text{J(MIN)}})$, where $T_{\text{J(MAX)}} =$ +125°C and $T_{J(MIN)} = -40$ °C.

- **3:** $V_{DO} = V_{IN} - V_{OUT}$ when V_{OUT} decreases to 99% of its nominal output voltage with $V_{IN} = V_{OUT} + 1V$. For output voltages below 2.25V, dropout voltage is the input-to-output voltage differential with the minimum input voltage being 2.25V. Minimum input operating voltage is 2.25V.
- **4:** I_{GND} is the quiescent current $(I_{IN} = I_{GND} + I_{OUT})$.
- **5:** For a 2.5V device, $V_{1N} = 2.250V$ (device is in dropout).
- **6:** V_{REF} ≤ V_{OUT} ≤ (V_{IN} – 1V), 2.25V ≤ V_{IN} ≤ 16V, 10 mA ≤ I_L ≤ 1A, T_J = T_{MAX}.

TEMPERATURE SPECIFICATIONS [\(Note 1](#page-5-0))

Note 1: The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air (i.e., T_A, T_J, θ_{JA}). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum +125°C rating. Sustained junction temperatures above +125°C can impact the device reliability.

2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

Ratio.

10 100 1K 10K 100K 1M

10K

FIGURE 2-2: Power Supply Rejection Ratio. **FREQUENCY (Hz)**

0

FIGURE 2-3: Power Supply Rejection Ratio.

Ratio.

FIGURE 2-4: Power Supply Rejection

Current.

FIGURE 2-5: Dropout Voltage vs. Output

Temperature.

FIGURE 2-6: Dropout Voltage vs.

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FIGURE 2-7: Dropout Characteristics (2.5V).

FIGURE 2-8: Dropout Characteristics (3.3V).

FIGURE 2-9: Ground Current vs. Output Current.

FIGURE 2-10: Ground Current vs. Supply Voltage (2.5V).

FIGURE 2-11: Ground Current vs. Supply Voltage (2.5V).

FIGURE 2-12: Ground Current vs. Supply Voltage (3.3V).

FIGURE 2-13: Ground Current vs. Supply Voltage (3.3V).

FIGURE 2-14: Ground Current vs. Temperature.

FIGURE 2-15: Ground Current vs. Temperature.

FIGURE 2-16: Ground Current vs. Temperature.

FIGURE 2-17: Output Voltage vs. Temperature.

FIGURE 2-18: Short-Circuit vs.

FIGURE 2-19: Error Flag Voltage vs. Pull-Up Resistor Value.

Temperature.

FIGURE 2-21: Flag-Low Voltage vs. Temperature.

FIGURE 2-22: Load Transient Response.

FIGURE 2-24: Line Transient Response.

3.0 PIN DESCRIPTIONS

The descriptions of the pins are listed in [Table 3-1.](#page-10-0)

TABLE 3-1: PIN FUNCTION TABLE

4.0 APPLICATION INFORMATION

The MIC39100/1/2 is a high performance, low dropout voltage regulator suitable for moderate to high current voltage regulator applications. Its 630 mV dropout voltage at full load and over temperature makes it especially valuable in battery-powered systems and as high efficiency noise filters in post-regulator applications. Unlike older NPN-pass transistor designs, where the minimum dropout voltage is limited by the base-to-emitter voltage drop and collector-to-emitter saturation voltage, dropout performance of the PNP output of these devices is limited only by the low V_{CF} saturation voltage.

A trade-off for the low dropout voltage is a varying base drive requirement that reduces the drive requirement to only 2% of the load current.

The MIC39100/1/2 regulator is fully protected from damage due to fault conditions. Linear current limiting is provided. Output current during overload conditions is constant. Thermal shutdown disables the device when the die temperature exceeds the maximum safe operating temperature. Transient protection allows device (and load) survival even when the input voltage spikes above and below nominal. The output structure of these regulators allows voltages in excess of the desired output voltage to be applied without reverse current flow.

FIGURE 4-1: Capacitor Requirements.

4.1 Output Capacitor

The MIC39100/1/2 requires an output capacitor to maintain stability and improve transient response. Proper capacitor selection is important to ensure proper operation. The MIC39100/1/2 output capacitor selection is dependent upon the equivalent series resistance (ESR) of the output capacitor to maintain stability. When the output capacitor is 10 µF or greater, the output capacitor should have an ESR less than 2Ω. This will improve transient response as well as promote stability. Ultra-low ESR capacitors (<100 m Ω), such as ceramic-chip capacitors, may promote instability. These very low ESR levels may cause an oscillation and/or underdamped transient response. A low-ESR solid tantalum capacitor works extremely well and provides good transient response and stability over temperature. Aluminum electrolytics can also be used, as long as the ESR of the capacitor is <2Ω.

The value of the output capacitor can be increased without limit. Higher capacitance values help to improve transient response and ripple rejection and reduce output noise.

4.2 Input Capacitor

An input capacitor of 1 μ F or greater is recommended when the device is more than four inches away from the bulk ac supply capacitance or when the supply is a battery. Small, surface mount, ceramic chip capacitors can be used for bypassing. Larger values will help to improve ripple rejection by bypassing the input to the regulator, further improving the integrity of the output voltage.

4.3 Error Flag

The MIC39101 features an error flag (FLG) that monitors the output voltage and signals an error condition when this voltage drops 5% below its expected value. The error flag is an open-collector output that pulls low under fault conditions and may sink up to 10 mA. Low output voltage signifies a number of possible problems, including an overcurrent fault (the device is in current-limit) or low input voltage. The flag output is inoperative during overtemperature conditions. A pull-up resistor from FLG to either V_{IN} or V_{OUT} is required for proper operation. For information regarding the minimum and maximum values of pull-up resistance, refer to [Figure 2-19.](#page-9-0)

4.4 Enable Input

The MIC39101 and MIC39102 feature an active-HIGH enable input (EN) that allows on/off control of the regulator. Current drain reduces to zero when the device is shutdown, with only microamperes (µA) of leakage current. The EN input has TTL/CMOS-comparable thresholds for simple logic interfacing. EN can be directly tied to V_{IN} and pulled-up to the maximum supply voltage.

4.5 Transient Response and 3.3V to 2.5V or 2.5V to 1.8V Conversion

The MIC39100/1/2 has excellent transient response to variations in input voltage and load current. The device has been designed to respond quickly to load current variations and input voltage variations. Large output capacitors are not required to obtain this performance. A standard 10 µF output capacitor, preferably tantalum, is all that is required. Larger values help to improve performance even further.

By virtue of its low dropout voltage, this device does not saturate into dropout as readily as similar NPN-based designs. When converting from 3.3V to 2.5V or 2.5V to 1.8V, the NPN-based regulators are already operating in dropout, with typical dropout requirements of 1.2V or

greater. To convert down to 2.5V or 1.8V without operating in dropout, NPN-based regulators require an input voltage of 3.7V at the very least.

The MIC39100 regulator will provide excellent performance with an input as low as 3.0V or 2.5V respectively. This gives the PNP-based regulators a distinct advantage over older, NPN-based linear regulators.

4.6 Minimum Load Current

The MIC39100/1/2 regulator is specified between finite loads. If the output current is too small, leakage currents dominate and the output voltage rises. A 10 mA minimum load current is necessary for proper regulation.

4.7 Adjustable Regulator Design

The MIC39102 allows programming the output voltage anywhere between 1.24V and the 16V maximum operating rating of the family. Two resistors are used. Resistors can be quite large, up to 1 MΩ, because of the very high input impedance and low bias current of the sense comparator: The resistor values are calculated by [Equation 4-1:](#page-12-0)

EQUATION 4-1:

$$
R1 = R2 \left(\frac{V_{OUT}}{1.240} - 1 \right)
$$

Where:

 V_{OUT} Desired output voltage.

Applications with widely varying load currents may scale the resistors to draw the minimum load current required for proper operation ([Figure 4-2\)](#page-12-1).

FIGURE 4-2: Adjustable Regulator with Resistors.

EQUATION 4-2:

$$
V_{OUT} = 1.240V \left(1 + \frac{R1}{R2}\right)
$$

4.8 Power SOIC-8 Thermal Characteristics

One of the secrets of the MIC39101/2's performance is its power SO-8 package. Lower thermal resistance means more output current or higher input voltage for a given package size.

Lower thermal resistance is achieved by joining the four ground leads with the die attach paddle to create a single-piece electrical and thermal conductor. This concept has been used by MOSFET manufacturers for years, proving very reliable and cost effective for the user.

Thermal resistance consists of two main elements, θ_{JC} (junction-to-case thermal resistance) and θ_{CA} (case-to-ambient thermal resistance, see [Figure 4-3\)](#page-12-2). θ_{JC} is the resistance from the die to the leads of the package. θ_{CA} is the resistance from the leads to the ambient air and it includes θ_{CS} (case-to-sink thermal resistance) and θ_{SA} (sink-to-ambient thermal resistance).

FIGURE 4-3: Thermal Resistance.

Using the power SOIC-8 reduces the θ_{JC} dramatically and allows the user to reduce θ_{CA} . The total thermal resistance, θ_{JA} , (junction-to-ambient thermal (junction-to-ambient resistance) is the limiting factor in calculating the maximum power dissipation capability of the device. Typically, the power SOIC-8 has a θ_{JC} of 20°C/W, which is significantly lower than the standard SOIC-8 (typically 75°C/W). θ_{CA} is reduced due to the capability of soldering Pins 5 through 8 directly to a ground plane.

This significantly reduces the case-to-sink thermal resistance as well as the sink-to-ambient thermal resistance.

Low dropout linear regulators from Microchip are rated to a maximum junction temperature of +125°C. It is important not to exceed this maximum junction temperature during operation of the device. To prevent this maximum junction temperature from being exceeded, the appropriate ground plane heat sink must be used.

[Figure 4-4](#page-13-0) shows copper area versus power dissipation with each trace corresponding to a different temperature rise above ambient.

From these curves, the minimum area of copper necessary for the part to operate safely can be determined. The maximum allowable temperature rise must be calculated to determine operation along which curve.

For example, the maximum ambient temperature is 50°C, the ∆T is determined as in [Equation 4-3:](#page-13-1)

EQUATION 4-3:

$$
\Delta T = 125^{\circ}C - 50^{\circ}C = 75^{\circ}C
$$

Where:

 ΔT $T_{J(MAX)} - T_{A(MAX)}$ $T_{J(MAX)}$ +125°C $T_{A(MAX)}$ Max. ambient operating temperature Using [Figure 4-4,](#page-13-0) the minimum amount of required copper can be determined based on the required power dissipation. Power dissipation in a linear regulator is calculated as in [Equation 4-4](#page-13-2):

EQUATION 4-4:

$$
P_D = (V_{IN} - V_{OUT})I_{OUT} + V_{IN} \times I_{GND}
$$

Using a 2.5V output device and a 3.3V input at an output current of 1A, the power dissipation is calculated as in [Equation 4-5](#page-13-3):

EQUATION 4-5:

$$
P_D = (3.3V - 2.5V)1A + 3.3V \times 11 mA
$$

= 800 mW + 36 mW = 836 mW

From [Figure 4-4](#page-13-0), the minimum amount of copper required to operate this application at a ∆T of 75°C is 160 mm².

4.9 Quick Method

Determine the power dissipation requirements for the design along with the maximum ambient temperature at which the device will be operated. Refer to [Figure 4-5,](#page-14-0) which shows safe operating curves for three different ambient temperatures: 25°C, 50°C, and 85°C. From these curves, the minimum amount of copper can be determined by knowing the maximum power dissipation required. If the maximum ambient temperature is 50°C and the power dissipation is as above, 836 mW, the curve in [Figure 4-5](#page-14-0) shows that the required area of copper is 160 mm².

The θ_{JA} of this package is ideally 63°C/W, but it will vary depending upon the availability of copper ground plane to which it is attached.

FIGURE 4-5: Copper Area vs. Power SOIC Power Dissipation (TA).

5.0 PACKAGING INFORMATION

5.1 Package Marking Information

3-Lead SOT-223 Package Outline and Recommended Land Pattern

8-Lead SOIC Package Outline and Recommended Land Pattern

APPENDIX A: REVISION HISTORY

Revision A (August 2017)

- Converted Micrel document MIC39100/1/2 to Microchip data sheet DS20005834A.
- Minor text changes throughout.

PRODUCT IDENTIFICATION SYSTEM

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