INTEGRATED PHOTO FLASH CHARGER AND IGBT DRIVER

FEATURES

- **High Integrated Solution to Reduce Components**
- **Integrated Voltage Reference**
- **Integrated 50-V Power Switch**
- **Integrated Insulated Gated BiPolar Transistor (IGBT) Driver**
- **High Efficiency**
- **Programmable Peak Current, 0.9 A ~ 1.8 A**
- **Input Voltage of 1.6 V to 12 V**
- **Optimized Control Loop for Fast Charge Time**
- **Isolate IGBT VCC from VCC**
- **Output Voltage Feedback From Primary Side**
- **16-pin QFN Package**
- **Protection**
  - **MAX On Time**
  - **MAX Off Time**
  - **Over Current Shutdown to Monitor VDS at SW Pin (OVDS)**
  - **Thermal Disable**

APPLICATIONS

- **Digital Still Cameras**
- **Optical Film Cameras**

DESCRIPTION/ORDERING INFORMATION

This device offers a complete solution for charging photo flash capacitor from battery input, and subsequently discharging the capacitor to a xenon flash tube. This device has an integrated voltage reference, power switch, IGBT driver, and control logic blocks for charging applications and driving IGBT application. Compared with discreet solutions, this device reduces the component count, shrinks the solution size, and eases designs for xenon tube application. Additional advantages are a fast charging time and high efficiency from an optimized pulse width modulation (PWM) control algorithm.

Other provisions of the device includes sensing the output voltage from the primary side, programmable peak current, thermal shutdown, an output pin for charge completion, and input pins for charge enable and flash enable.
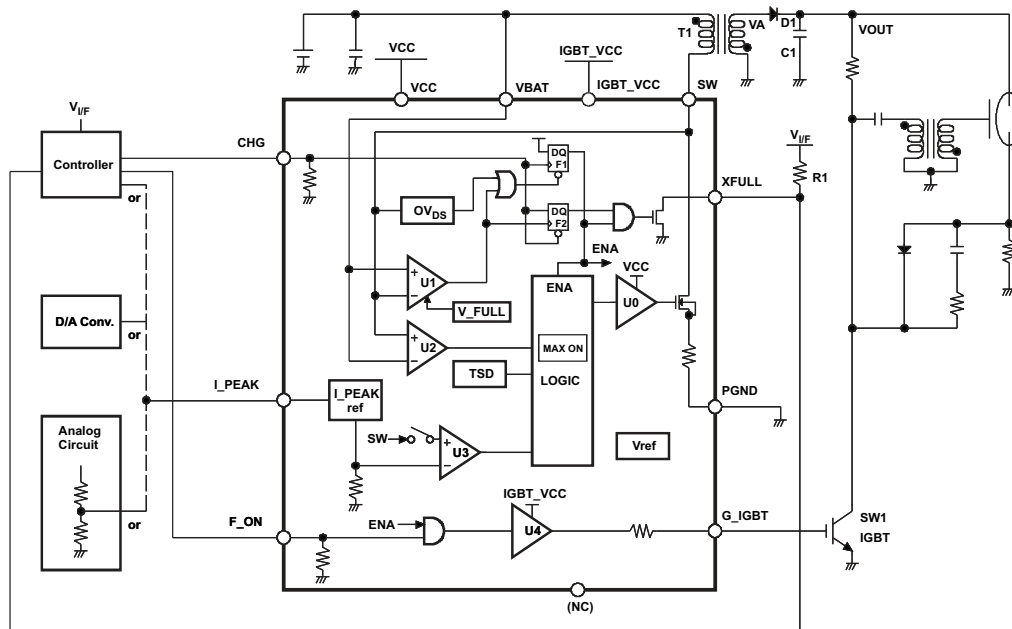

Figure 1. Typical Application Circuit

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

ORDERING INFORMATION

T_A	PACKAGE[1]	ORDERABLE PART NUMBER	TOP-SIDE MARKING
−35°C to 85°C	CAH	16-pin QFN	TPS65562RGT

(1) Package drawings, standard packing quantities, thermal data, symbolization, and PCB design guidelines are available at www.ti.com/sc/package.

ABSOLUTE MAXIMUM RATINGS[1]

over operating free-air temperature range (unless otherwise noted)

				UNIT
	Supply voltage	VCC, IGBT_VCC	−0.6 to 6	V
		VBAT	−0.6 to 13	
V_{SW}	Switch terminal voltage		−0.6 to 50	V
I_{SW}	Switch current between SW and PGND		3	A
V_I	Input voltage of CHG, I_PEAK, F_ON		−0.3 to V_{CC}	V
T_{stg}	Storage temperature		−40 to 150	°C
T_J	Maximum junction temperature		125	°C
	ESD rating	HBM (Human Body Model) JEDEC JES22-A114	1	kV

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

RECOMMENDED OPERATING CONDITIONS

			MIN	MAX	UNIT
	Supply voltage	VCC	2.7	4	V
		IGBT_VCC	2.7	5.5	
		VBAT	1.6	12	
V_{SW}	Switch terminal voltage		−0.3	45	V
I_{SW}	Switch current between SW and PGND			2	A
	Operating free-air temperature range		−35	85	°C
V_{IH}	High-level digital input voltage at CHG and F_ON		2		V
V_{IL}	Low-level digital input voltage at CHG and F_ON			0.5	V

DISSIPATION RATINGS

PACKAGE	$R_{\theta JA}$[1]	POWER RATINGS $T_A < 25°C$	POWER RATINGS RATE $T_A = 70°C$	POWER RATINGS RATE $T_A = 85°C$
QFN	47.4°C/W	2.11 W	1.16 W	844 mW

(1) The thermal resistance, $R_{\theta JA}$, is based on a soldered PowerPAD™ package on 2S2P JEDEC board using thermal vias.

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ELECTRICAL CHARACTERISTICS

T_A = 25°C, VBAT = 4.2 V, V_{CC} = 3 V, IGBT_VCC = 3 V, $V_{(SW)}$ = 4.2 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
R_{ONL}	ON resistance of XFULL	I_{XFULL} = −1 mA		1.5	3	kΩ
V_{PKH}[1]	Upper threshold voltage of I_PEAK	VCC = 3 V	2.4			V
V_{PKL}[1]	Lower threshold voltage of I_PEAK	VCC = 3 V			0.6	V
I_{CC1}	Supply current from VBAT	CHG = H, V_{SW} = 0 V (free run by t_{MAX})		17	50	µA
I_{CC2}	Supply current from VCC	CHG = H, V_{SW} = 0 V (free run by t_{MAX})		1.3	3	mA
I_{CC3}	Supply current from VCC and VBAT	CHG = L			1	µA
I_{LKG1}	Leakage current of SW terminal				2	µA
I_{LKG2}	Leakage current of XFULL terminal	V_{XFULL} = 5 V			1	µA
I_{sink}	Sink current at I_PEAK	V_{I_PEAK} = 3 V, CHG: High			2	µA
		V_{I_PEAK} = 3 V, CHG: Low			0.1	
R_{ONSW}	SW ON resistance between SW and PGND	I_{SW} = 1 A, V_{CC} = 3 V		0.4	0.9	Ω
R_{IGBT1}	G_IGBT pull up resistance	V_{G_IGBT} = 0 V, IGBT_VCC = 3 V	8	12	19.4	Ω
R_{IGBT2}	G_IGBT pull down resistance	V_{G_IGBT} = 3 V, IGBT_VCC = 3 V	36	53	70	Ω
I_{PEAK1}	Upper peak of ISW	V_{I_IPEAK} = 3 V	1.58	1.68	1.78	A
I_{PEAK2}	Lower peak of ISW	V_{I_IPEAK} = 0 V	0.7	0.8	0.9	A
V_{FULL}	Charge completion detect voltage at V(SW)	VBAT = 1.6 V, VCC = 3 V	28.0	28.7	29.4	V
		VCC = 3 V	28.6	29.0	29.4	V
V_{ZERO}	Zero current detection at VSW		1	20	60	mV
T_{SD}[1]	Thermal shutdown temperature		150	160	170	°C
O_{VDS}	Over current detection at VSW		0.95	1.2	1.45	V
T_{MIX}	MAX OFF time		25	50	80	µs
T_{MAX}	MAX ON time		50	100	160	µs
R_{INPD}	Pull down resistance of CHG, F_ON	V_{CHG} = V_{F_ON} = 4.2 V		100		kΩ

(1) Specified by design

SWITCHING CHARACTERISTICS

T_A =25°C, VBAT = 4.2 V, V_{CC} = 3 V, IGBT_VCC = 3 V, V_{SW} =4.2 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PD}[1]	Propagation delay	F_ON↑↓ G_IGBT↑↓		50		ns
		SW ON after VSW dips from V_{ZERO}		500		ns
		SW OFF after ISW exceeds I_{PEAK}		270		ns
		XFULL↓ after VSW exceeds V_{Full}		400		ns
		SW ON after CHG↑		12		µs
		SW OFF after CHG↓		20		ns

(1) Specified by design

**QFN PACKAGE
(TOP VIEW)**
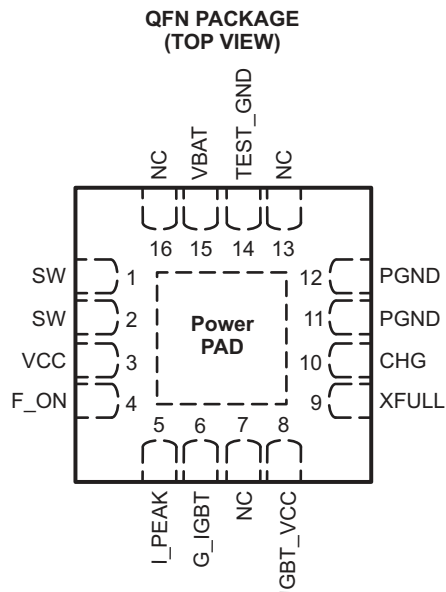
TERMINAL FUNCTIONS

| TERMINAL		I/O	DESCRIPTION
NAME	NO.		
SW	1, 2	O	Primary side switch. Connect SW to the switched side of the transformer
VCC	3	I	Power supply input. Connect VCC to an input supply from 2.7 V to 4.0 V. Bypass VCC to GND with a 1 μF ceramic capacitor as close as possible to the IC.
F_ON	4	I	G_IGBT control input. Drives F_ON with the flash discharge signal. A logic high on F_ON drives G_IGBT high when CHG is Low. See the *IGBT driver control* section for details.
I_PEAK	5	I	Primary side Peak current control input. The voltage at I_PEAK sets the peak current into SW. See the *Programming Peak Current* section for details on selecting V_{I_PEAK}.
G_IGBT	6	O	IGBT gate driver output. G_IGBT swings from PGND to VCC to drive external IGBT devices.
NC	7, 13, 16		No internal connection
IGBT_VCC	8	I	Power supply input for IGBT Driver output. Connect IGBT_VCC to an input supply from 2.7 V to 5.5 V.
XFULL	9	O	Charge completion indicator output. XFULL is an open-drain output that pulls low once the output is fully charged. XFULL is high impedance during charging and all fault conditions. XFULL is reset when CHG turns Low from High. See the *Indicating Charging status* section for details.
CHG	10	I	Charge control input. Drive CHG high to initiate charging of the output. Drive CHG low to terminate charging.
PGND	11, 12		Power ground. Connect to the ground plane.
TEST_GND	14		Used by TI, should be connected to PGND and ground plane.
VBAT	15	I	Battery voltage monitor input. Connect VBAT to an input supply from 1.6 V to 12 V. Bypass VBAT to GND with a 10 μF ceramic capacitor (C1 in Figure 1 as close as possible to the battery) and a 1 μF ceramic capacitor (C2 in Figure 1, as close as possible to the IC). It is no problem to input VBAT before VCC input.
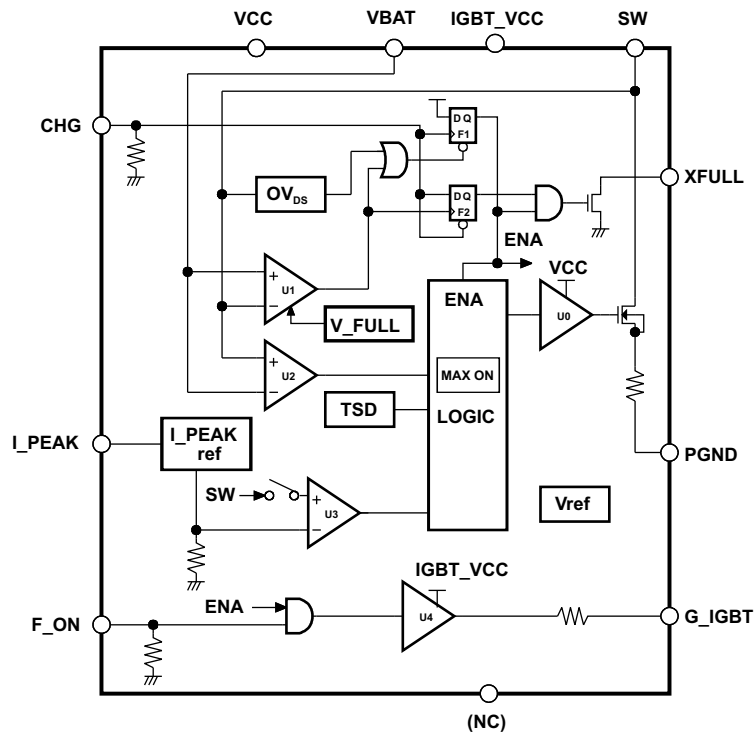
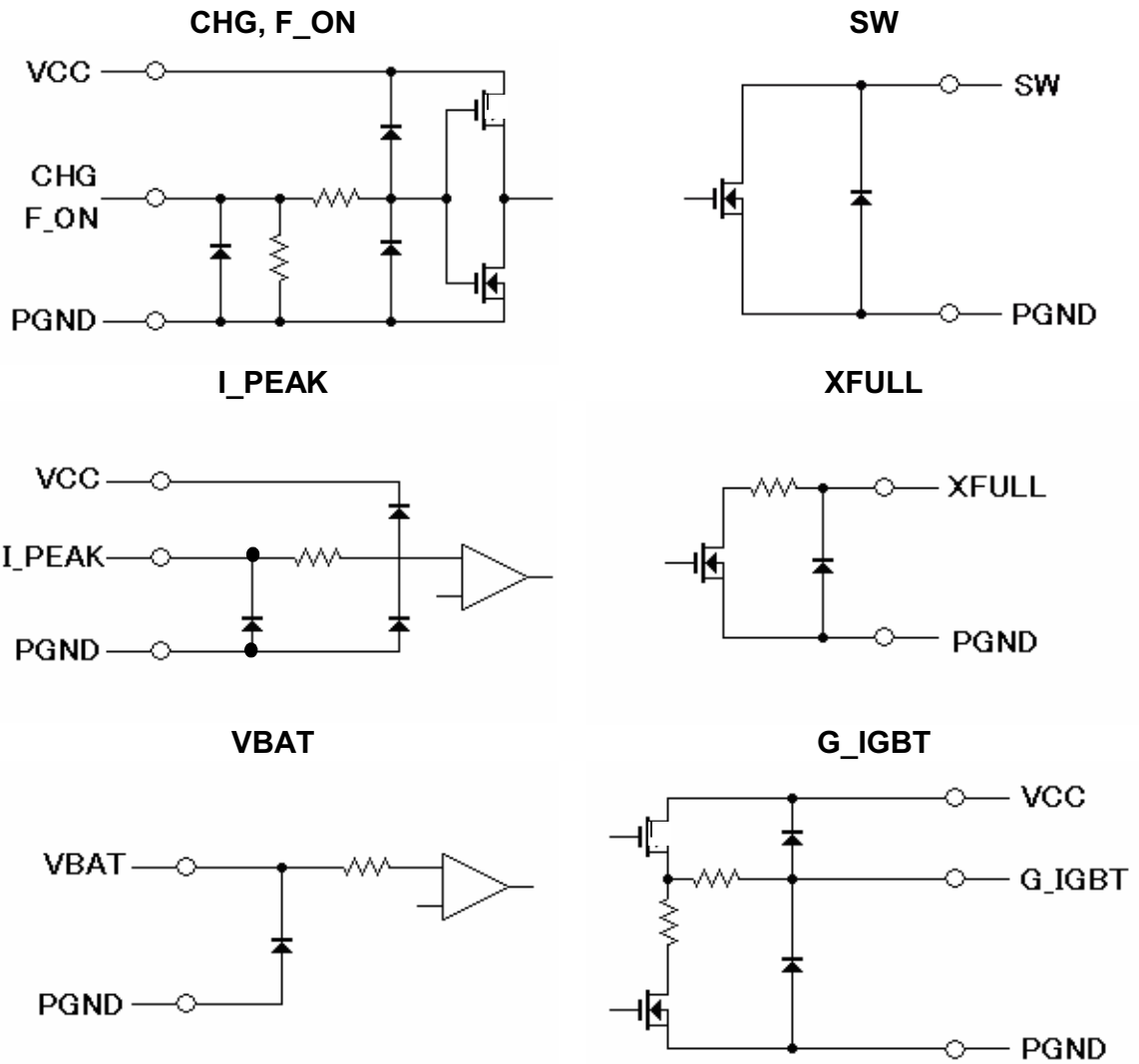**Figure 2. Block Diagram**

I/O Equivalent Circuits

CHG, F_ON   **SW**

I_PEAK   **XFULL**

VBAT   **G_IGBT**

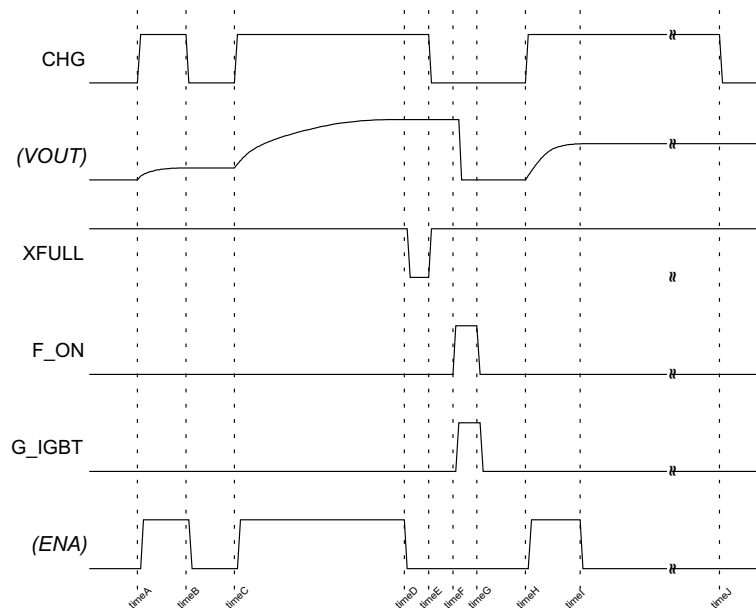**Figure 3. I/O Equivalent Circuits**

PRINCIPLES OF OPERATION

Figure 4. Whole Operation Sequence Chart

Start/Stop Charging

TPS65562 has one internal enable latch, F1, that holds the charge enable (ON/OFF status) of the device (see Figure 4).

The only way to start charging is to input CHG↑ (see time A/C/H in Figure 4). Each time CHG↑ is applied, the TPS65562 starts charging.

There are three trigger events to stop charging:
1. Forced stop by inputting CHG = L from the controller (see timeB in Figure 4).
2. Automatic stop by detecting a full charge. VOUT reaches the target value (see TimeD in Figure 4).
3. Protected stop by detecting an overcurrent function (OVDS) trigger at SW pin (see TimeI in Figure 4).

Indicate Charging Status

When the charging operation is complete, the TPS65562 drives the charge completion indicator pin, XFULL, to GND. A controller can detect the status of the device as a logic signal when connected through a pullup resister, R1 (see Figure 1).

The XFULL output enables the controller to detect the OVDS protection status. If OVDS protection occurs, XFULL never goes L during CHG = H. Therefore, the controller detects OVDS protection by measuring the time from CHG high to XFULL low. If the time to XFULL low is longer than the maximum designed charge time, then an OVDS protection occurred.

The device starts charging at timeH, and OVDS protection occurs at TimeI (see Figure 5). At TimeI, XFULL stays H. At TimeJ, the controller detects OVDS protection through the expiration of a timer ends and then sets CHG to low to terminate the operation.
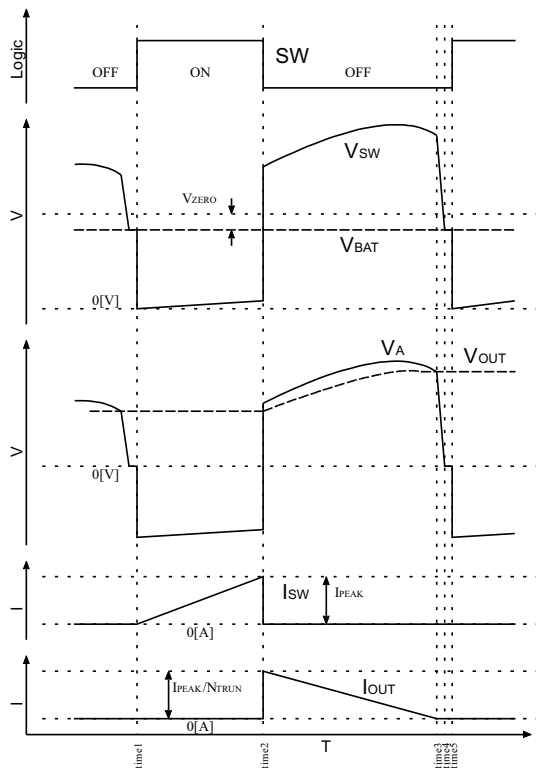
PRINCIPLES OF OPERATION (continued)
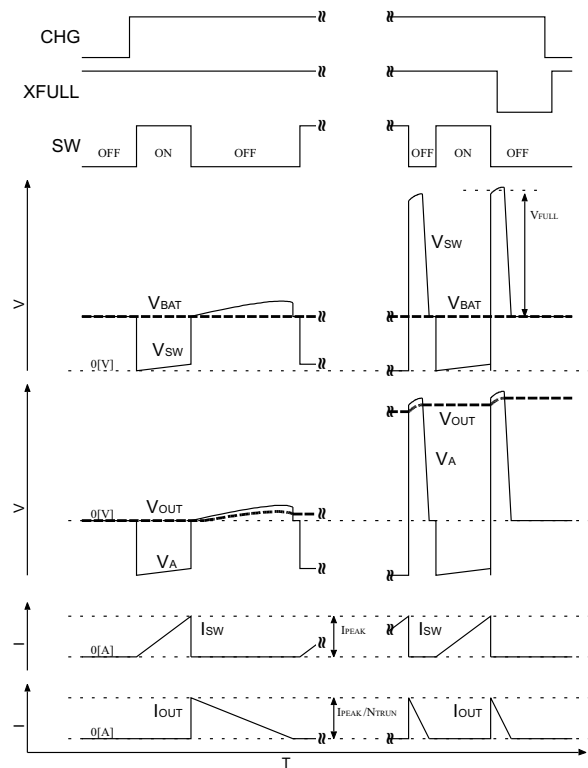

Figure 5. Timing Diagram at One Switching

Figure 6. Timing Diagram at Beginning/Ending

Control Charging

The TPS65562 provides three comparators to control charging. Figure 2 shows the block diagram of TPS65562, and Figure 5 shows a timing diagram of one switch cycle. Note that emphasis is placed on Time1 and Time3 of the waveform in Figure 5.

While SW is ON (Time1 to Time2 in Figure 5), U3 monitors current flow through the integrated power switch from SW pin to GND. When I(SW) exceeds I(PEAK), SW turns OFF (Time2 in Figure 5).

When SW turns OFF (Time2 in Figure 5), the magnetic energy in the transformer starts discharging. Meanwhile, U2 monitors the kickback voltage at the SW terminal. As the energy is discharging, the kickback voltage is increasing according to the increase of VO (Time2 to Time3 in Figure 5). When almost all energy is discharged, the system cannot continue rectification via the diode, and the charging current of IO goes to zero (Times3 in Figure 5). After rectification stops, the small amount of energy left in the transformer is released via parasitic paths, and the kickback voltage reaches zero (Time3 to Time4 in Figure 5). During this period, U2 makes SW turn ON when (V(SW) - VBAT) dips from V(ZERO) (Time5 in Figure 5). In the actual circuit, the period between Time4 and Time5 in Figure 5 is small or does not appear dependent on the delay time of the U2 detection to SW ON.

U1 also monitors the kickback voltage. When (V(SW) - VBAT) exceeds V(FULL), the TPS65562 stops charging (see Figure 6).

In Figure 5 and Figure 6, ON time is always the same period in every switch cycle. The ON time is calculated by Equation 1. L and I_{PEAK} are selected to ensure that t_{ON} does not exceed the MAX ON time (t_{MAX}).

$$T_{ON} = L \frac{I_{PEAK}}{V_{BAT}}$$

(1)

The OFF time is dependant on output voltage. As the output voltage gets higher, the OFF time gets shorter (see Equation 2).

PRINCIPLES OF OPERATION (continued)

$$T_{OFF} = N_{TURN} \times L \frac{I_{PEAK}}{V_{OUT}}$$

$$(2)$$

Programming Peak Current

The TPS65562 provides a method to program the peak primary current with a voltage applied to the I_PEAK pin. Figure 7 shows how to program I_{PEAK}.

The I_PEAK input is treated as a logic input below V_{PKL} (0.6 V) and above V_{PKH} (2.4 V). Between V_{PKL} and V_{PKH}, I_PEAK input is treated as an analog input. Using this characteristic, I_{PEAK} can be set by a logic signal or by an analog input.

Typical usages of this function are:
1. Setting the peak charging currents based on the battery voltage. Larger I_{PEAK} for a fully charged battery and lower I_{PEAK} for a discharged battery.
2. Reducing I_{PEAK} when powering a zooming lens motor. This avoids inadvertent shutdowns due to large current from the battery.

In Figure 1, three optional connections to I_PEAK are shown.
1. Use the controller to treat I_PEAK as the logic input pin. This option is the easiest.
2. Use a D/A converter to force I_{PEAK} to follow analog information, such as battery voltage.
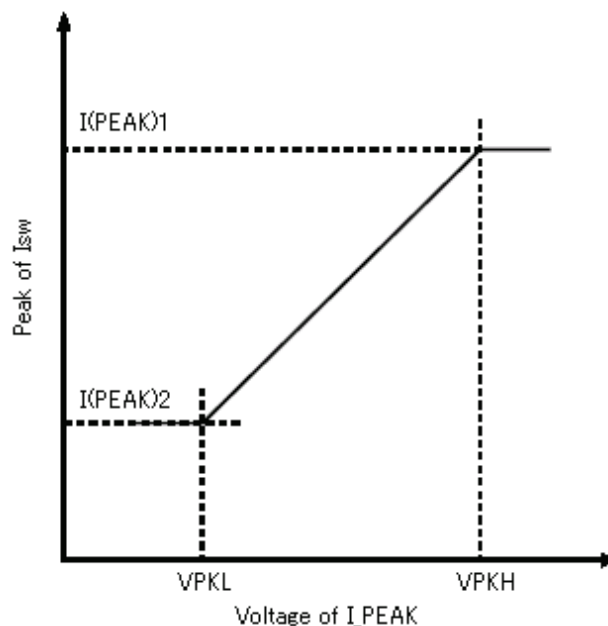3. Use an analog circuit to achieve the same results as the D/A converter.

Figure 7. I_PEAK vs I$_{SW}$

IGBT Driver Control

The IGBT driver is provided by the TPS65562. The driver voltage depends on VCC. TPS65562 has a mask filter as shown in Figure 8. The mask does not have hysteresis; therefore, there is no wait time from CHG forcing Low after FULL CHARGE to F_ON turning High.
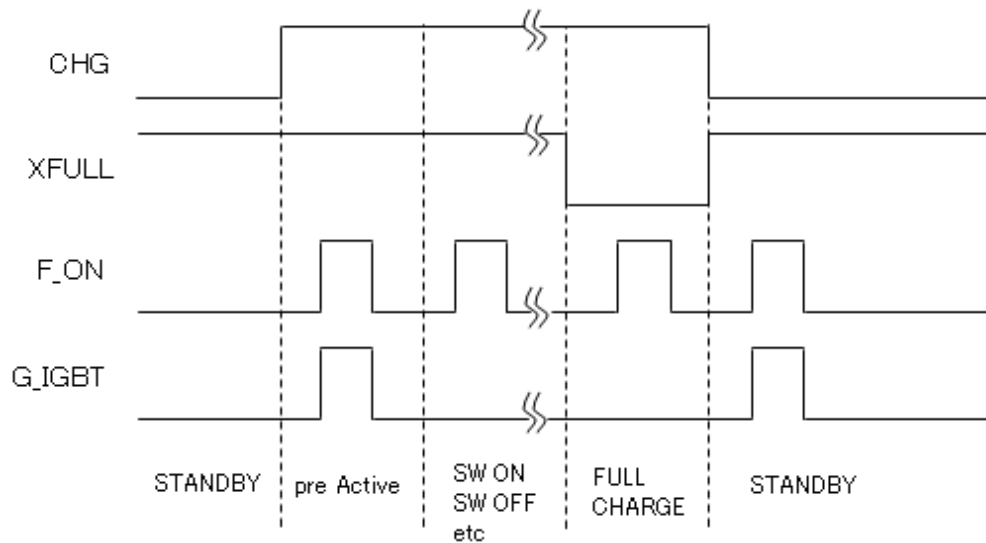
PRINCIPLES OF OPERATION (continued)

Figure 8. Relationships Between F_ON and CHG

Protections

TPS65562 provides four protection mechanisms: max on time, max off time, thermal disable, and over current shutdown.

MAX ON TIME

To prevent a condition such as pulling current from a poor power source (i.e., an almost empty battery), and never reaching peak current, the TPS65562 provides a maximum ON time function. If the ON time exceeds t_{MAX}, the TPS5562 is forced OFF regardless of I_{PEAK} detection.

MAX OFF TIME

To prevent a condition such as never increasing the voltage at the SW pin when the internal FET is OFF, the TPS65562 provides a maximum OFF time function. If the OFF time exceeds t_{MIN}, the TPS65562 is forced ON regardless of V_{ZERO} detection.

THERMAL DISABLE

Once the die temperature of the TPS65562, reaches 160°C, all functions stop. Once the die cools below 160°C, the TPS65562 restarts charging if CHG remains high during the entire over temperature condition.

OVER CURRENT SHUTDOWN

The TPS65562 provides an over voltage monitor function of the SW pin. The TPS65562 is latched off if the voltage on the SW pin is above OVDS during the switch ON time (see Figure 4 and its descriptions).

This function protects against short-circuits on the primary side of the transformer. A short-circuit of the primary side shorts the battery voltage to GND. SW pin can damage the device if not protected.

![Texas Instruments logo]

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPS65562RGTR	ACTIVE	QFN	RGT	16		TBD	Call TI	Call TI	-35 to 85	CAH	Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
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Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
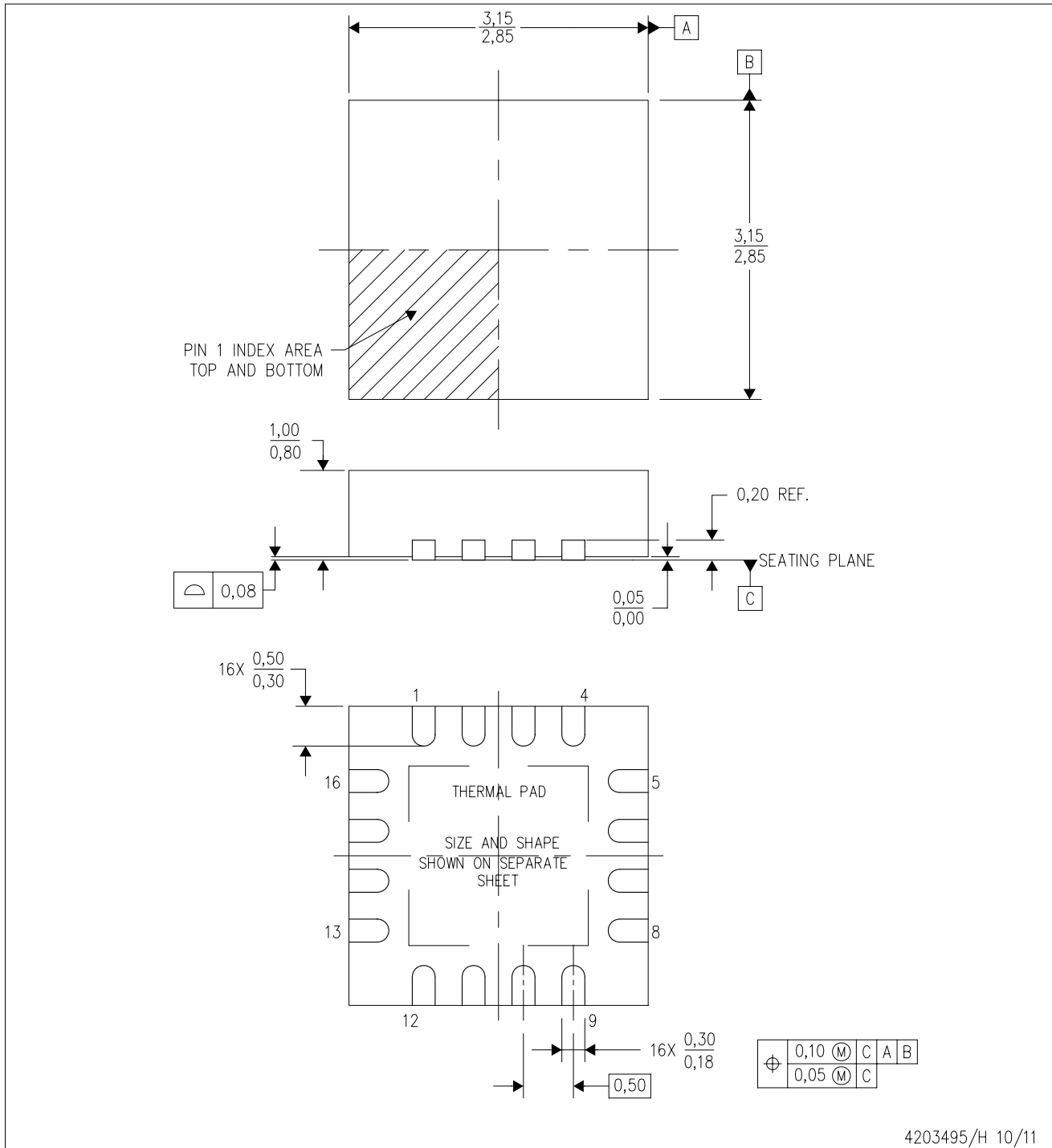
(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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RGT (S−PVQFN−N16)   PLASTIC QUAD FLATPACK NO−LEAD

PIN 1 INDEX AREA
TOP AND BOTTOM

$\frac{3,15}{2,85}$

$\frac{3,15}{2,85}$

$\frac{1,00}{0,80}$

0,20 REF.

SEATING PLANE

⌀ 0,08

$\frac{0,05}{0,00}$

C

16X $\frac{0,50}{0,30}$

1   4

16   5

THERMAL PAD

SIZE AND SHAPE
SHOWN ON SEPARATE
SHEET

13   8

12   9

16X $\frac{0,30}{0,18}$

0,50

⊕	0,10 Ⓜ	C	A	B
	0,05 Ⓜ	C		

4203495/H 10/11

NOTES: A.   All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M−1994.
             B.   This drawing is subject to change without notice.
             C.   Quad Flatpack, No−leads (QFN) package configuration.
             D.   The package thermal pad must be soldered to the board for thermal and mechanical performance.
             E.   See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
             F.   Falls within JEDEC MO−220.

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