

Not recommended for new designs—see TPH3206PSB

## 600V GaN FET in TO-220 (drain tab)

### Description

The TPH3206PD 600V,  $150m\Omega$  Gallium Nitride (GaN) FET is a normally-off device. It combines state-of-the-art high voltage GaN HEMT and low voltage silicon MOSFET technologies—offering superior reliability and performance.

Transphorm GaN offers improved efficiency over silicon, through lower gate charge, lower crossover loss, and smaller reverse recovery charge.

#### Related Literature

- ANOOOO: Recommended External Circuitry for GaN FETs
- ANOOO3: Printed Circuit Board Layout and Probing
- ANOO10: Paralleling GaN FETs

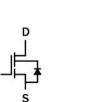
### **Ordering Information**

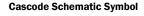
Part Number	Package	Package Configuration
TPH3206PD	3 lead TO-220	Drain

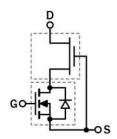
#### TPH3206PD TO-220 (top view)











**Cascode Device Structure** 

#### **Features**

- JEDEC qualified GaN technology
- Dynamic R<sub>DS(on)eff</sub> production tested
- · Robust design, defined by
  - Intrinsic lifetime tests
  - Wide gate safety margin
  - Transient over-voltage capability
- Very low Q<sub>RR</sub>
- Reduced crossover loss
- · RoHS compliant and Halogen-free packaging

#### **Benefits**

- Enables AC-DC bridgeless totem-pole PFC designs
  - Increased power density
  - Reduced system size and weight
  - Overall lower system cost
- · Achieves increased efficiency in both hard- and softswitched circuits
- Easy to drive with commonly-used gate drivers
- · GSD pin layout improves high speed design

### **Applications**

- Datacom
- · Broad industrial
- PV inverter
- Servo motor

Key Specifications		
V <sub>DS</sub> (V) min	600	
V <sub>(TR)DSS</sub> (V) max	750	
$R_{DS(on)eff}(m\Omega)$ max*	180	
Q <sub>RR</sub> (nC) typ	52	
Q <sub>G</sub> (nC) typ	6	

<sup>\*</sup> Dynamic on-resistance; see Figures 19 and 20

Common Topology Power Recommendations				
CCM bridgeless totem-pole* 1519W max				
Hard-switched inverter**	1717W max			

Conditions: F<sub>SW</sub>=45kHz; T<sub>J</sub>=115°C; T<sub>HEATSINK</sub>=90°C; insulator between device and heatsink (6 mil Sil-Pad® K-10); power de-rates at lower voltages with constant current

- VIN=230VAC: VOUT=390VDC
- $V_{IN}$ =380 $V_{DC}$ ;  $V_{OUT}$ =240 $V_{AC}$

## **Absolute Maximum Ratings** (T<sub>c</sub>=25 °C unless otherwise stated.)

Symbol	Parameter		Limit Value	Unit
V <sub>DSS</sub>	Drain to source voltage (T <sub>J</sub> = -5	55°C to 150°C)	600	
V <sub>(TR)DSS</sub>	Transient drain to source volta	nge a	750	V
V <sub>GSS</sub>	Gate to source voltage		±18	
P <sub>D</sub>	Maximum power dissipation @	T <sub>C</sub> =25°C	96	W
1	Continuous drain current @T <sub>C</sub> =25°C b		17	А
I <sub>D</sub>	Continuous drain current @T <sub>C</sub> =100°C b		12	А
I <sub>DM</sub>	Pulsed drain current (pulse width: 10µs)		60	А
(di/dt) <sub>RDMC</sub>	Reverse diode di/dt, repetitive °		1200	A/µs
(di/dt) <sub>RDMT</sub>	Reverse diode di/dt, transient	Reverse diode di/dt, transient d		A/µs
Tc	Operating temperature	Case	-55 to +150	°C
Tı	Operating temperature	Junction	-55 to +175	°C
Ts	Storage temperature	Storage temperature		°C
T <sub>SOLD</sub>	Soldering peak temperature e		260	°C

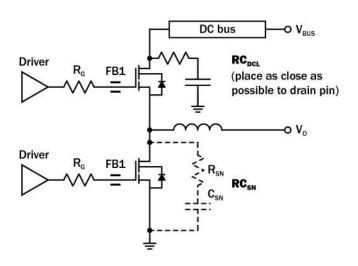
#### Notes:

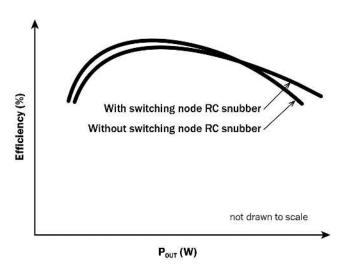
- a. In off-state, spike duty cycle D<0.01, spike duration  $<1\mu$ s
- b. For increased stability at high current operation, see Circuit Implementation on page 3
- c. Continuous switching operation
- d. ≤300 pulses per second for a total duration ≤20 minutes
- e. For 10 sec., 1.6mm from the case

### **Thermal Resistance**

Symbol	Parameter	Typical	Unit
Rojc	Junction-to-case	0.7	°C/W
Roja	Junction-to-ambient	40	°C/W

### **Circuit Implementation**





**Simplified Half-bridge Schematic** 

**Efficiency vs Output Power** 

Recommended gate drive: (0V, 8-10V) with  $R_{G(tot)} = 25\Omega$ , where  $R_{G(tot)} = R_G + R_{DRIVER}$ 

Gate Ferrite Bead (FB1)	Required DC Link RC Snubber (RC <sub>DCL</sub> ) <sup>a</sup>	Recommended Switching Node RC Snubber (RC <sub>SN</sub> ) b, c
MMZ1608Q121BTA00	10nF + $8\Omega$	22pF + 15Ω

#### Notes:

- a.  $RC_{DCL}$  should be placed as close as possible to the drain pin
- b. A switching node RC snubber (C, R) is recommended for high switching currents (>70% of I\_RDMC1 or I\_RDMC2; see page 5 for I\_RDMC1 and I\_RDMC2)
- c.  $\mbox{$I_{RDM}$ values can be increased by increasing $R_{G}$ and $C_{SN}$}$

### **Electrical Parameters** (T<sub>J</sub>=25 °C unless otherwise stated)

Symbol	Parameter	Min	Тур	Max	Unit	Test Conditions
Forward Device Characteristics						
$V_{(BL)DSS}$	Maximum drain-source voltage	600	_	_	V	V <sub>GS</sub> =0V
V <sub>GS(th)</sub>	Gate threshold voltage	1.65	2.1	2.6	V	V <sub>DS</sub> =V <sub>GS</sub> , I <sub>D</sub> =500μA
В	Drain-source on-resistance a	_	150	180	mΩ	V <sub>GS</sub> =8V, I <sub>D</sub> =11A,
R <sub>DS(on)eff</sub>	Drain-source on-resistance -	_	340	_		V <sub>GS</sub> =8V, I <sub>D</sub> =11A, T <sub>J</sub> =175°C
l	Drain to course leakage current	_	2.5	90		V <sub>DS</sub> =600V, V <sub>GS</sub> =0V
I <sub>DSS</sub>	Drain-to-source leakage current	_	8	_	ŀμA	V <sub>DS</sub> =600V, V <sub>GS</sub> =0V, T <sub>J</sub> =150°C
1	Gate-to-source forward leakage current	_	_	100	nΛ	V <sub>GS</sub> =18V
I <sub>GSS</sub>	Gate-to-source reverse leakage current	_	_	-100	· nA	V <sub>GS</sub> =-18V
C <sub>ISS</sub>	Input capacitance	_	760	_		
Coss	Output capacitance	_	44	_	pF	V <sub>GS</sub> =0V, V <sub>DS</sub> =480V, <i>f</i> =1MHz
C <sub>RSS</sub>	Reverse transfer capacitance	_	5	_		
C <sub>O(er)</sub>	Output capacitance, energy related <sup>b</sup>	_	64	_	nE	V <sub>GS</sub> =0V, V <sub>DS</sub> =0V to 480V
$C_{O(tr)}$	Output capacitance, time related °	_	105	_	pF	
Q <sub>G</sub>	Total gate charge	_	6.2	9.3		
Q <sub>GS</sub>	Gate-source charge	_	2.1	_	nC	$V_{DS}$ =100V, $V_{GS}$ =0V to 4.5V, $I_{D}$ =11A
$Q_{GD}$	Gate-drain charge	_	2.2	_		
Qoss	Output charge	_	44.4	_	nC	V <sub>GS</sub> =0V, V <sub>DS</sub> =0V to 400V
t <sub>D(on)</sub>	Turn-on delay	_	6	_		
t <sub>R</sub>	Rise time	_	4.5	_	no	V <sub>DS</sub> =480V, V <sub>GS</sub> =0V to 10V,
t <sub>D(off)</sub>	Turn-off delay	_	9.7	_	ns	$I_D=11A$ , $R_G=22\Omega$
t <sub>F</sub>	Fall time	_	4	_		
Reverse	Device Characteristics					
ls	Reverse current	_	_	12	А	V <sub>GS</sub> =0V, T <sub>C</sub> =100 °C, ≤25% duty cycle
V <sub>SD</sub> Rev		_	2.6	_		V <sub>GS</sub> =0V, I <sub>S</sub> =12A
	Reverse voltage <sup>a</sup>	_	4.6	_	V	V <sub>GS</sub> =0V, I <sub>S</sub> =12A, T <sub>J</sub> =175°C
		_	1.8	_		V <sub>GS</sub> =0V, I <sub>S</sub> =6A
t <sub>RR</sub>	Reverse recovery time	_	17	_	ns	I <sub>S</sub> =11A, V <sub>DD</sub> =400V,
Q <sub>RR</sub>	Reverse recovery charge	_	52	_	nC	di/dt=2000A/μs

#### Notes:

a. Dynamic on-resistance; see Figures 19 and 20 for test circuit and conditions

b. Equivalent capacitance to give same stored energy as  $V_{DS}$  rises from OV to 400V

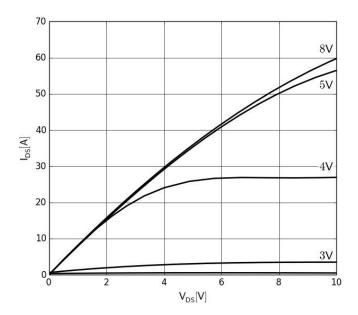
c. Equivalent capacitance to give same charging time as  $V_{DS}$  rises from 0V to 400V

### **Electrical Parameters** (T<sub>J</sub>=25 °C unless otherwise stated)

Symbol	Parameter		Тур	Max	Unit	Test Conditions	
Reverse Dev	teverse Device Characteristics						
Is	Reverse current	_	_	12	А	V <sub>GS</sub> =0V, T <sub>C</sub> =100°C, ≤25% duty cycle	
		_	2.6	_		V <sub>GS</sub> =0V, I <sub>S</sub> =12A	
$V_{\text{SD}}$	Reverse voltage a	_	4.6	_	V	V <sub>GS</sub> =0V, I <sub>S</sub> =12A, T <sub>J</sub> =175°C	
		_	1.8	_		V <sub>GS</sub> =0V, I <sub>S</sub> =6A	
t <sub>RR</sub>	Reverse recovery time	_	17	_	ns	I <sub>S</sub> =11A, V <sub>DD</sub> =400V, di/dt=2000A/μs	
$Q_{RR}$	Reverse recovery charge	_	52	_	nC		
(di/dt) <sub>RDMC</sub>	Reverse diode di/dt, repetitive b	_	_	1200	A/µs		
I <sub>RDMC1</sub>	Reverse diode switching current, repetitive (dc) c, e	_	_	11	А	Circuit implementation and parameters on page 3	
I <sub>RDMC2</sub>	Reverse diode switching current, repetitive (ac) c, e	_	_	14	А	Circuit implementation and parameters on page 3	
(di/dt) <sub>RDMT</sub>	Reverse diode di/dt, transient d	_	_	2400	A/µs		
I <sub>RDMT</sub>	Reverse diode switching current, transient d,e	_	_	18	А	Circuit implementation and parameters on page 3	

#### Notes:

- a. Includes dynamic  $R_{\text{DS(on)}}$  effect
- b. Continuous switching operation
- c. Definitions: dc = dc-to-dc converter topologies; ac = inverter and PFC topologies, 50-60Hz line frequency
- d. ≤300 pulses per second for a total duration ≤20 minutes
- e.  $I_{RDM}$  values can be increased by increasing  $R_{G}$  and  $C_{SN}$  on page 3



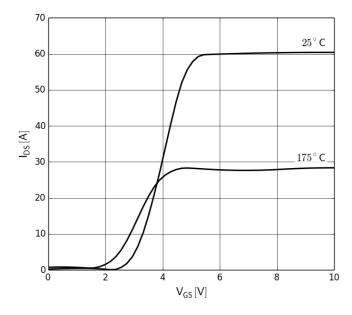
30 25 20 20 15 10 5 0 0 2 4 6 8 10 V<sub>DS</sub>[V]

Figure 1. Typical Output Characteristics T<sub>J</sub>=25 °C

Parameter: V<sub>GS</sub>

Figure 2. Typical Output Characteristics T<sub>J</sub>=175°C

Parameter: V<sub>GS</sub>



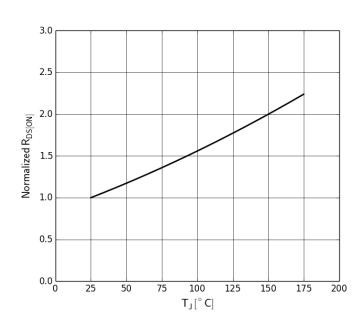
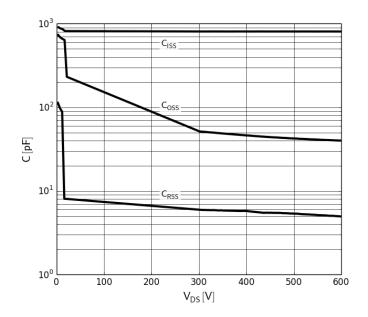


Figure 3. Typical Transfer Characteristics  $V_{DS}$ =10V, parameter:  $T_J$ 

Figure 4. Normalized On-resistance  $I_D=12A, V_{GS}=8V$ 



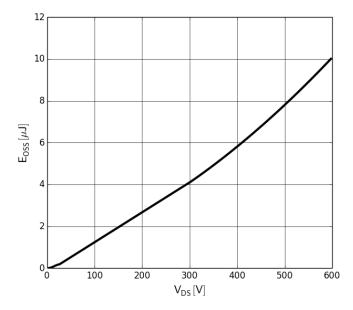
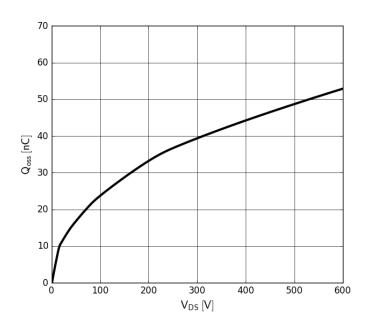


Figure 5. Typical Capacitance  $V_{GS}$ =0V, f=1MHz

Figure 6. Typical Coss Stored Energy



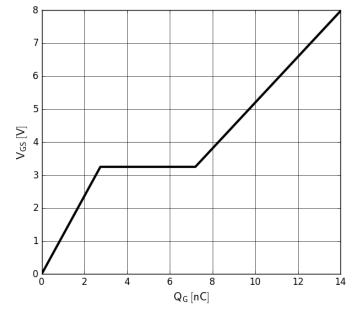


Figure 7. Typical Qoss

Figure 8. Typical Gate Charge IDS=12A, VDS=400V

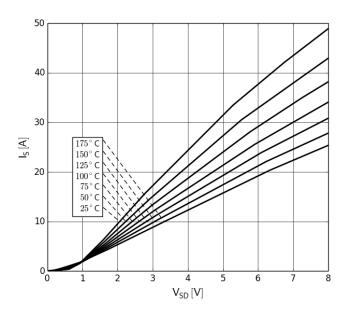


Figure 9. Forward Characteristics of Rev. Diode  $I_S {=} f(V_{SD}), \ Parameter \ T_J$ 

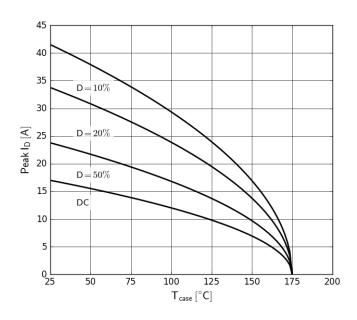


Figure 10. Current Derating Pulse width  $\leq 100 \mu s$ 

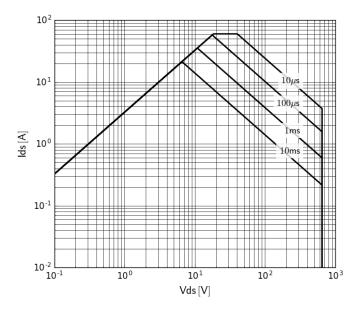


Figure 11. Safe Operating Area  $T_c$ =25 ° C (calculated based on thermal limit)

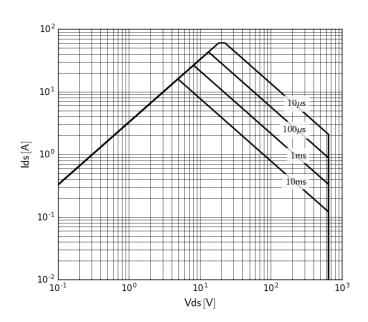
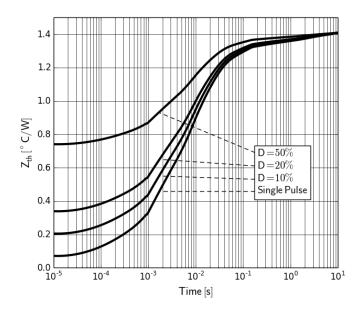


Figure 12. Safe Operating Area T<sub>c</sub>=80 °C (calculated based on thermal limit)



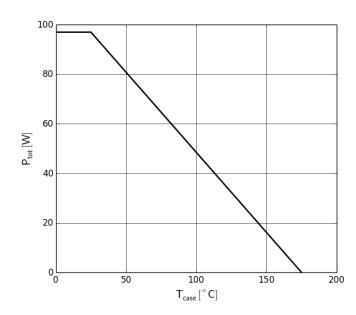
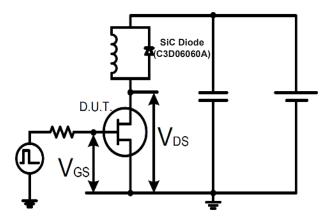


Figure 13. Transient Thermal Resistance

Figure 14. Power Dissipation

### **Test Circuits and Waveforms**



**Figure 15. Switching Time Test Circuit** (see circuit implementation on page 3 for methods to ensure clean switching)

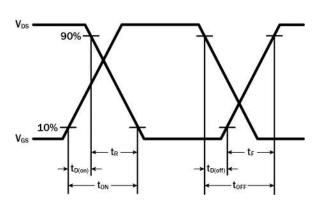


Figure 16. Switching Time Waveform

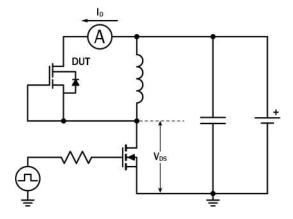


Figure 17. Diode Characteristics Test Circuit

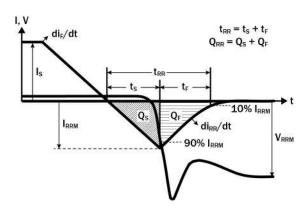


Figure 18. Diode Recovery Waveform

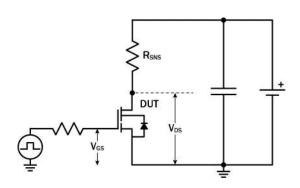


Figure 19. Dynamic R<sub>DS(on)eff</sub> Test Circuit

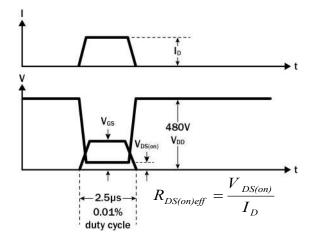


Figure 20. Dynamic R<sub>DS(on)eff</sub> Waveform

### **Design Considerations**

The fast switching of GaN devices reduces current-voltage crossover losses and enables high frequency operation while simultaneously achieving high efficiency. However, taking full advantage of the fast switching characteristics of GaN switches requires adherence to specific PCB layout guidelines and probing techniques.

Before evaluating Transphorm GaN devices, see application note <u>Printed Circuit Board Layout and Probing for GaN Power Switches</u>. The table below provides some practical rules that should be followed during the evaluation.

#### When Evaluating Transphorm GaN Devices:

DO	DO NOT
Minimize circuit inductance by keeping traces short, both in the drive and power loop	Twist the pins of TO-220 or TO-247 to accommodate GDS board layout
Minimize lead length of TO-220 and TO-247 package when mounting to the PCB	Use long traces in drive circuit, long lead length of the devices
Use shortest sense loop for probing; attach the probe and its ground connection directly to the test points	Use differential mode probe or probe ground clip with long wire
See ANOOO3: Printed Circuit Board Layout and Probing	

### **GaN Design Resources**

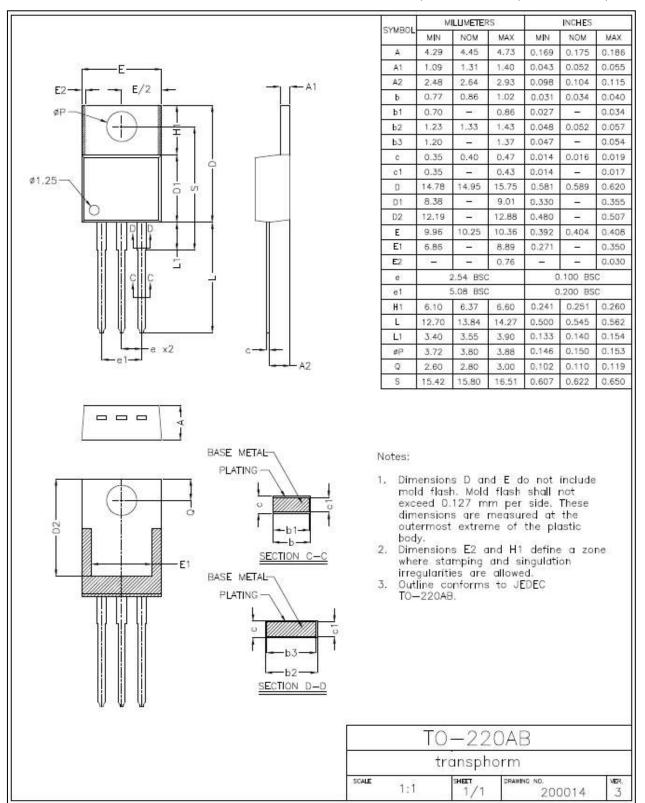
The complete technical library of GaN design tools can be found at <a href="mailto:transphormusa.com/design">transphormusa.com/design</a>:

- Reference designs
- Evaluation kits
- Application notes
- · Design guides
- Simulation models
- Technical papers and presentations

#### Mechanical

### 3 Lead TO-220 (PD) Package

Pin 1: Gate; Pin 2: Source; Pin 3: Drain, Tab: Drain



# **Revision History**

Version	Date	Change(s)	
4	11/14/2016	Added app note AN0009	
5	12/12/2016	Updated dynamic measurement verbiage	
6	11/2/2017	Updated package drawing, Figures 11 & 12 (pg 7), effective on-resistance symbol to R <sub>DS(on)eff</sub> to adhere to new JEDEC standards; Added app note AN0010, common topology max power recommendations (pg 1), switching current values (pg 2), Circuit Implementation (pg 3), Qoss value (pg 4), Figures 7 & 8 (pg 6)	
7	3/27/2018	No longer recommended for new designs	