

μPD48288209AF1 μPD48288218AF1 μPD48288236AF1

288M-BIT Low Latency DRAM Common I/O

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Description

The μ PD48288209AF1 is a 33,554,432-word by 9 bit, the μ PD48288218AF1 is a 16,777,216-word by 18 bit and the μ PD48288236AF1 is a 8,388,608-word by 36 bit synchronous double data rate Low Latency RAM fabricated with advanced CMOS technology using one-transistor memory cell.

The μ PD48288209AF1, μ PD48288218AF1 and μ PD48288236AF1 integrate unique synchronous peripheral circuitry and a burst counter. All input registers controlled by an input clock pair (CK and CK#) are latched on the positive edge of CK and CK#. These products are suitable for application which require synchronous operation, high speed, low voltage, high density and wide bit configuration.

Specification

• Density: 288M bit

• Organization

- Common I/O: 4M words x 9 bits x 8 banks

2M words x 18 bits x 8 banks

1M words x 36 bits x 8 banks

• Operating frequency: 533 / 400 / 300 MHz

• Interface: HSTL I/O

• Package: 144-pin FBGA

Package size: 18.5 x 11

- Lead free

Power supply

- 2.5 V VEXT

- 1.8 V V_{DD}

- 1.5 V or 1.8 V VDDQ

• Refresh command

Auto Refresh

- 8192 cycle / 32 ms for each bank

- 64K cycle / 32 ms for total

 \bullet Operating case temperature : Tc=0 to $95^{\circ}C$

Features

- SRAM-type interface
- Double-data-rate architecture
- PLL circuitry

• Cycle time: 1.875 ns @ trc = 15 ns

2.5 ns @ $t_{RC} = 15$ ns 2.5 ns @ $t_{RC} = 20$ ns 3.3 ns @ $t_{RC} = 20$ ns

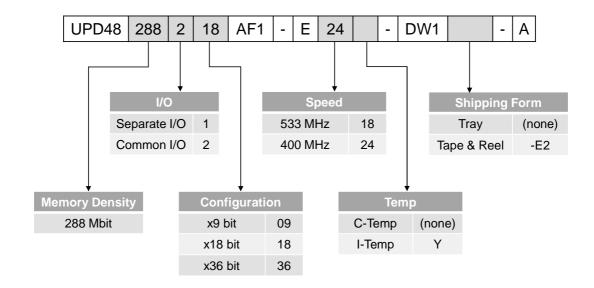
- Non-multiplexed addresses
- Multiplexing option is available.
- Data mask for WRITE commands
- Differential input clocks (CK and CK#)
- Differential input data clocks (DK and DK#)
- Data valid signal (QVLD)
- Programmable burst length: 2 / 4 / 8 (x9 / x18 / x36)
- User programmable impedance output (25 Ω 60 Ω)
- JTAG boundary scan

Ordering Information

Part number	Cycle Time	Clock Frequency		Organization (word x bit)	Core Supply Voltage (V _{EXT})	Core Supply Voltage (V _{DD})	Output Supply Voltage (VDDQ)	Package
	ns	MHz	Ns		V	V	V	
μPD48288209AF1-E24-DW1-A	2.5	400	15	32 M x 9	2.5 + 0.13	1.8 ± 0.1	1.5 ± 0.1	144-pin
μPD48288218AF1-E18-DW1-A	1.875	533	15	16 M x 18	2.5 – 0.12		or	FBGA
μPD48288218AF1-E24-DW1-A	2.5	400	15				1.8 ± 0.1	(18.5 x 11)
μPD48288236AF1-E18-DW1-A	1.875	533	15	8 M x 36				Lead-free
μPD48288236AF1-E24-DW1-A	2.5	400	15					

Part Number Definition

Example) UPD48288218AF1-E24-DW1-A



Pin Arrangement

144-pin FBGA (18.5 x 11) (Top View) [Common I/O x9]

	1	2	3	4	5	6	7	8	9	10	11	12
Α	VREF	Vss	VEXT	Vss					Vss	V EXT	TMS	тск
В	V _{DD}	Note 3 DNU	Note 3 DNU	VssQ					VssQ	DQ0	Note 3 DNU	V DD
С	V TT	Note 3 DNU	Note 3 DNU	V _{DD} Q					V _{DD} Q	DQ1	Note 3 DNU	V TT
D	Note 1 (A22)	Note 3 DNU	Note 3 DNU	VssQ					VssQ	QK0#	QK0	Vss
Ε	Note 1 (A21)	Note 3 DNU	Note 3 DNU	$V_{DD}Q$					V _{DD} Q	DQ2	Note 3 DNU	A20
F	A5	Note 3 DNU	Note 3 DNU	VssQ					VssQ	DQ3	Note 3 DNU	QVLD
G	A8	A6	A7	V DD					V DD	A2	A 1	A0
Н	BA2	A9	Vss	Vss					Vss	Vss	A4	А3
J	Note 2 NF	Note 2 NF	V DD	V _{DD}					V DD	V DD	BA0	СК
K	DK	DK#	V DD	V DD					V DD	V _{DD}	BA1	CK#
L	REF#	CS#	Vss	Vss					Vss	Vss	A14	A13
M	WE#	A16	A17	V DD					V DD	A12	A11	A10
N	A18	Note 3 DNU	Note 3 DNU	VssQ					VssQ	DQ4	Note 3 DNU	A19
P	A15	Note 3 DNU	Note 3 DNU	V _{DD} Q					V _{DD} Q	DQ5	Note 3 DNU	DM
R	Vss	Note 3 DNU	Note 3 DNU	VssQ					VssQ	DQ6	Note 3 DNU	Vss
Т	V TT	Note 3 DNU	Note 3 DNU	V _{DD} Q					V _{DD} Q	DQ7	Note 3 DNU	VTT
U	V DD	Note 3 DNU	Note 3 DNU	VssQ					VssQ	DQ8	Note 3 DNU	V DD
٧	VREF	ZQ	V EXT	Vss					Vss	V EXT	TDO	TDI

- **Notes 1.** Reserved for future use. This signal is internally connected and has parasitic characteristics of an address input signal. This may optionally be connected to Vss, or left open.
 - **2.** No function. This signal is internally connected and has parasitic characteristics of a clock input signal. This may optionally be connected to Vss, or left open.
 - **3.** Do not use. This signal is internally connected and has parasitic characteristics of a I/O. This may optionally be connected to Vss, or left open.

CK, CK#	: Input clock	TMS	: IEEE 1149.1 Test input
CS#	: Chip select	TDI	: IEEE 1149.1 Test input
WE#	: WRITE command	TCK	: IEEE 1149.1 Clock input
REF#	: Refresh command	TDO	: IEEE 1149.1 Test output
A0-A20	: Address inputs	V_{REF}	: HSTL input reference input
A21-A22	: Reserved for the future	V_{EXT}	: Power Supply
BA0-BA2	: Bank address input	V_{DD}	: Power Supply
DQ0-DQ8	: Data input/output	$V_{\mathrm{DD}}Q$: DQ Power Supply
DK, DK#	: Input data clock	V_{ss}	: Ground
DM	: Input data Mask	VssQ	: DQ Ground
QK0, QK0#	: Output data clock	V_{TT}	: Power Supply
QVLD	: Data Valid	NF	: No function

: Output impedance matching

indicates active LOW signal.

: Do not use

ZQ

DNU

Pin Arrangement

144-pin FBGA (18.5 x 11) (Top View) [Common I/O x18]

	1	2	3	4	5	6	7	8	9	10	11	12
Α	VREF	Vss	V EXT	Vss					Vss	V EXT	TMS	тск
В	V _{DD}	Note 3 DNU	DQ4	VssQ					VssQ	DQ0	Note 3 DNU	V DD
С	VTT	Note 3 DNU	DQ5	V _{DD} Q					$V_{DD}Q$	DQ1	Note 3 DNU	V TT
D	Note 1 (A22)	Note 3 DNU	DQ6	VssQ					VssQ	QK0#	QK0	Vss
E	Note 1 (A21)	Note 3 DNU	DQ7	V DD Q					$V_{DD}Q$	DQ2	Note 3 DNU	Note 1 (A20)
F	A5	Note 3 DNU	DQ8	VssQ					VssQ	DQ3	Note 3 DNU	QVLD
G	A8	A6	A7	V _{DD}					V DD	A2	A 1	A0
Н	BA2	Α9	Vss	Vss					Vss	Vss	A4	А3
J	Note 2 NF	Note 2 NF	V DD	V _{DD}					V DD	V _{DD}	BA0	СК
K	DK	DK#	V DD	V _{DD}					V DD	V _{DD}	BA1	CK#
L	REF#	CS#	Vss	Vss					Vss	Vss	A14	A13
M	WE#	A16	A17	V _{DD}					V DD	A12	A11	A10
N	A18	Note 3 DNU	DQ14	VssQ					VssQ	DQ9	Note 3 DNU	A19
Р	A15	Note 3 DNU	DQ15	V _{DD} Q					V _{DD} Q	DQ10	Note 3 DNU	DM
R	Vss	QK1	QK1#	VssQ					VssQ	DQ11	Note 3 DNU	Vss
Т	V TT	Note 3 DNU	DQ16	V _{DD} Q	-		_		V _{DD} Q	DQ12	Note 3 DNU	V TT
U	V DD	Note 3 DNU	DQ17	VssQ					VssQ	DQ13	Note 3 DNU	V DD
٧	VREF	ZQ	V EXT	Vss					Vss	V EXT	TDO	TDI

- **Notes 1.** Reserved for future use. This signal is internally connected and has parasitic characteristics of an address input signal. This may optionally be connected to Vss, or left open.
 - **2.** No function. This signal is internally connected and has parasitic characteristics of a clock input signal. This may optionally be connected to Vss, or left open.
 - **3.** Do not use. This signal is internally connected and has parasitic characteristics of a I/O. This may optionally be connected to Vss, or left open.

CK, CK#	: Input clock	TMS	: IEEE 1149.1 Test input
CS#	: Chip select	TDI	: IEEE 1149.1 Test input
WE#	: WRITE command	TCK	: IEEE 1149.1 Clock input
REF#	: Refresh command	TDO	: IEEE 1149.1 Test output
A0-A19	: Address inputs	V_{REF}	: HSTL input reference input
A20-A22	: Reserved for the future	V_{EXT}	: Power Supply
BA0-BA2	: Bank address input	$V_{\rm DD}$: Power Supply
DQ0-DQ17	: Data input/output	$V_{\mathrm{DD}}Q$: DQ Power Supply
DK, DK#	: Input data clock	$\mathbf{V}_{\mathbf{SS}}$: Ground
DM	: Input data Mask	VssQ	: DQ Ground
QK0-QK1, QK0#-QK1#	: Output data clock	V_{TT}	: Power Supply
QVLD	: Data Valid	NF	: No function
ZQ	: Output impedance matching	DNU	: Do not use

indicates active LOW signal.

Pin Arrangement

144-pin FBGA (18.5 x 11) (Top View) [Common I/O x36]

	1	2	3	4	5	6	7	8	9	10	11	12
Α	VREF	Vss	V EXT	Vss					Vss	V EXT	TMS	TCK
В	V _{DD}	DQ8	DQ9	VssQ					VssQ	DQ1	DQ0	V DD
С	V TT	DQ10	DQ11	V _{DD} Q					V _{DD} Q	DQ3	DQ2	V TT
D	Note (A22)	DQ12	DQ13	VssQ					VssQ	QK0#	QK0	Vss
E	Note (A21)	DQ14	DQ15	V _{DD} Q					V _{DD} Q	DQ5	DQ4	Note (A20)
F	A5	DQ16	DQ17	VssQ					VssQ	DQ7	DQ6	QVLD
G	A8	A6	A7	V _{DD}					V DD	A2	A 1	Α0
Н	BA2	A9	Vss	Vss					Vss	Vss	A4	А3
J	DK0	DK0#	V _{DD}	V _{DD}					V DD	V _{DD}	BA0	СК
K	DK1	DK1#	V DD	V _{DD}					V DD	V DD	BA1	CK#
L	REF#	CS#	Vss	Vss					Vss	Vss	A14	A13
M	WE#	A16	A17	V _{DD}					V DD	A12	A11	A10
N	A18	DQ24	DQ25	VssQ					VssQ	DQ35	DQ34	Note (A19)
Р	A15	DQ22	DQ23	V _{DD} Q					V _{DD} Q	DQ33	DQ32	DM
R	Vss	QK1	QK1#	VssQ					VssQ	DQ31	DQ30	Vss
Т	VTT	DQ20	DQ21	V _{DD} Q					VDDQ	DQ29	DQ28	VTT
U	V _{DD}	DQ18	DQ19	VssQ					VssQ	DQ27	DQ26	V _{DD}
٧	V REF	ZQ	V EXT	Vss					Vss	V EXT	TDO	TDI

Note Reserved for future use. This signal is internally connected and has parasitic characteristics of an address input signal. This may optionally be connected to Vss, or left open.

CK, CK#	: Input clock	TMS	: IEEE 1149.1 Test input
CS#	: Chip select	TDI	: IEEE 1149.1 Test input
WE#	: WRITE command	TCK	: IEEE 1149.1 Clock input
REF#	: Refresh command	TDO	: IEEE 1149.1 Test output
A0-A18	: Address inputs	$V_{ m REF}$: HSTL input reference input

QVLD : Data Valid

ZQ : Output impedance matching

indicates active LOW signal.

Pin Description

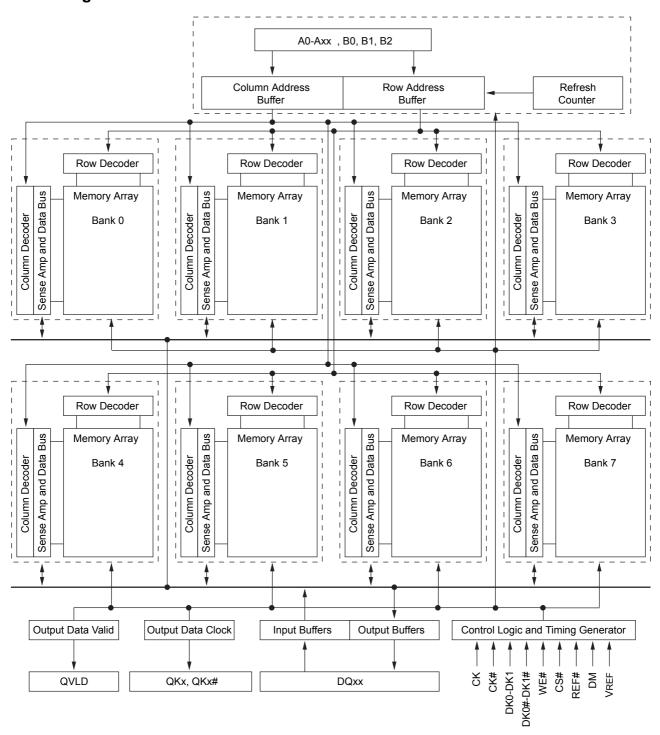
(1/2)

Symbol	Туре	Description
CK, CK#	Input	Clock inputs:
		CK and CK# are differential clock inputs. This input clock pair registers address and control inputs on the rising edge of CK. CK# is ideally 180 degrees out of phase with CK.
CS#	Input	Chip select
		CS# enables the commands when CS# is LOW and disables them when CS# is HIGH. When the command is disabled, new commands are ignored, but internal operations continue.
WE#, REF#	Input	WRITE command pin, Refresh command pin:
		WE#, REF# are sampled at the positive edge of CK, WE#, and REF# define (together with CS#) the command to be executed.
A0-A20	Input	Address inputs:
		A0–A20 define the row and column addresses for READ and WRITE operations. During a MODE REGISTER SET, the address inputs define the register settings. They are sampled at the rising edge of CK.
		In the x36 configuration, A19–A20 are reserved for address expansion; in the x18 configuration, A20 is reserved for address expansion. These expansion addresses can be treated as address inputs, but they do not affect the operation of the device.
A21-A22	Input	Reserved for future use:
		These signals should be tied to Vss or leave open.
BA0-BA2	Input	Bank address inputs;
		Select to which internal bank a command is being applied.
DQ0-DQxx	Input	Data input/output:
	/Output	The DQ signals form the 9/18/36 bit data bus. During READ commands, the data is referenced to both edges of QKx. During WRITE commands, the data is sampled at both edges of DKx.
		x 9 device uses DQ0 to DQ8.
		x18 device uses DQ0 to DQ17.
014 014 11	0	x36 device uses DQ0 to DQ35.
QKx, QKx#	Output	Output data clocks: QKx and QKx# are opposite polarity, output data clocks. They are always free running and edge-aligned with data output from the μ PD48288209/18/36AF1. QKx# is ideally 180 degrees out of phase with QKx.
		For the x36 device, QK0 and QK0# are aligned with DQ0–DQ17. QK1 and QK1# are aligned with DQ18–DQ35. For the x18 device, QK0 and QK0# are aligned with DQ0–DQ8. QK1 and QK1# are aligned with DQ9–DQ17. For the x9 device, QK0 and QK0# are aligned with DQ0–DQ8.
DKx, DKx#	Input	Input data clock;
		DKx and DKx# are the differential input data clocks. All input data is referenced to both edges of DK. DK# is ideally 180 degrees out of phase with DK.
		For the x36 device, DQ0–DQ17 are referenced to DK0 and DK0#, and DQ18–DQ35 are referenced to DK1 and DK1#. For the x9 and x18 devices, all DQs are referenced to DK and DK#.
DM	Input	Input data mask;
		The DM signal is the input mask signal for WRITE data. Input data is masked when DM is sampled HIGH along with the WRITE input data. DM is sampled on both edges of DK (DK1 for the x36 configuration). The signal should be Vss if not used.
QVLD	Output	Data valid; The QVLD indicates valid output data. QVLD is edge-aligned with QKx and QKx#.

(2/2)

Symbol	Туре	Description
ZQ	Input /Output	External impedance [$25 \Omega - 60 \Omega$]; This signal is used to tune the device outputs to the system data bus impedance. DQ output impedance is set to $0.2 \times RQ$, where RQ is a resistor from this signal to Vss. Connecting ZQ to Vss invokes the minimum impedance mode. Connecting ZQ to VpQ invokes the maximum impedance mode. Refer to Figure 2-5. Mode Register Bit Map to activate this function.
TMS, TDI	Input	JTAG function pins: IEEE 1149.1 test inputs: These balls may be left as no connects if the JTAG function is not used in the circuit
тск	Input	JTAG function pin; IEEE 1149.1 clock input: This ball must be tied to Vss if the JTAG function is not used in the circuit.
TDO	Output	JTAG function pin; IEEE 1149.1 test output: JTAG output. This ball may be left as no connect if JTAG function is not used.
VREF	Input	Input reference voltage; Nominally VDDQ/2. Provides a reference voltage for the input buffers.
Vext	Supply	Power supply; 2.5 V nominal. See Recommended DC Operating Conditions for range.
V _{DD}	Supply	Power supply; 1.8 V nominal. See Recommended DC Operating Conditions for range.
V _{DD} Q	Supply	DQ power supply; Nominally, 1.5 V or 1.8 V. Isolated on the device for improved noise immunity. See Recommended DC Operating Conditions for range.
Vss	Supply	Ground
VssQ	Supply	DQ ground; Isolated on the device for improved noise immunity.
Vπ	Supply	Power supply; Isolated termination supply. Nominally, VDDQ/2. See Recommended DC Operating Conditions for range.
NF		No function; These balls may be connected to Vss.
DNU		Do not use; These balls may be connected to Vss.

Block Diagram



μ PD48288209AF1, μ PD48288218AF1, μ PD48288236AF1

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1. Electrical Characteristics

Absolute Maximum Ratings

Parameter	Symbol	Conditions	Rating	Unit
Supply voltage	V _{EXT}		-0.3 to +2.8	V
Supply voltage	V_{DD}		-0.3 to +2.1	V
Output supply voltage,	$V_{DD}Q$		-0.3 to +2.1	V
Input voltage, Input / Output voltage				
Input / Output voltage	VIH / VIL		-0.3 to +2.1	V
Junction temperature	T _j MAX.		110	°C
Storage temperature	T _{stg}		-55 to +125	°C

Caution Exposing the device to stress above those listed in Absolute Maximum Ratings could cause permanent damage. The device is not meant to be operated under conditions outside the limits described in the operational section of this specification. Exposure to Absolute Maximum Rating conditions for extended periods may affect device reliability.

Recommended DC Operating Conditions

 $0^{\circ}C \le T_C \le 95^{\circ}C$; $1.7 \text{ V} \le V_{DD} \le 1.9 \text{ V}$, unless otherwise noted.

Parameter	Symbol	Conditions	MIN.	TYP.	MAX.	Unit	Note
Supply voltage	VEXT		2.38	2.5	2.63	V	1
Supply voltage	V _{DD}		1.7	1.8	1.9	V	1
Output supply voltage	V _{DD} Q		1.4		V _{DD}	V	1, 2, 3
Reference Voltage	VREF		0.49 x V _{DD} Q	0.5 x VddQ	0.51 x VddQ	V	1, 4, 5
Termination voltage	Vπ		0.95 x Vref	VREF	1.05 x Vref	V	1, 6
Input HIGH voltage	VIH (DC)		V _{REF} + 0.1			V	1
Input LOW voltage	VIL (DC)				V _{REF} – 0.1	V	1

Notes 1. All voltage referenced to Vss (GND).

- 2. During normal operation, VDDQ must not exceed VDD.
- **3.** V_{DD}Q can be set to a nominal 1.5 V \pm 0.1 V or 1.8 V \pm 0.1 V supply.
- **4.** Typically the value of V_{REF} is expect to be 0.5 x V_{DD}Q of the transmitting device. V_{REF} is expected to track variations in V_{DD}Q.
- **5.** Peak-to-peak AC noise on V_{REF} must not exceed $\pm 2\%$ V_{REF}(DC).
- 6. V_{TT} is expected to be set equal to V_{REF} and must track variations in the DC level of V_{REF} .

DC Characteristics

 $0^{\circ}C \le T_C \le 95^{\circ}C$; 1.7 V $\le V_{DD} \le 1.9$ V, unless otherwise noted

Parameter	Symbol	Test condition	MIN.	MAX.	Unit	Note
Input leakage current	lu		- 5	+5	μΑ	1,2
Output leakage current	Іьо		-5	+5	μΑ	1,2
Reference voltage current	IREF		-5	+5	μΑ	1,2
Output high current	Іон	$V_{OH} = V_{DD}Q/2$	(V _{DD} Q/2) / (1.15 x RQ/5)	(V _{DD} Q/2) / (0.85 x RQ/5)	mA	3,4
Output low current	Іоь	Vol = VddQ/2	(V _{DD} Q/2) / (1.15 x RQ/5)	(V _{DD} Q/2) / (0.85 x RQ/5)	mA	3,4

- **Notes 1.** Outputs are impedance-controlled. $| \text{Ioh } | = (V_{DD}Q/2)/(RQ/5)$ for values of $125 \Omega \le RQ \le 300 \Omega$.
 - 2. Outputs are impedance-controlled. IoL = $(V_{DD}Q/2)/(RQ/5)$ for values of 125 $\Omega \le RQ \le 300 \Omega$.
 - **3.** IoH and IoL are defined as absolute values and are measured at VDDQ/2. IoH flows from the device, IoL flows into the device.
 - **4.** If MRS bit A8 is 0, use $RQ = 250 \Omega$ in the equation in lieu of presence of an external impedance matched resistor.

Capacitance (TA = 25 °C, f = 1MHz)

Parameter	Symbol	Test conditions	MIN.	MAX.	Unit
Address / Control Input capacitance	Cin	Vin = 0 V	1.5	2.5	pF
I/O, Output, Other capacitance	C _{I/O}	V//O = 0 V	3.5	5.0	pF
(DQ, DM, QK, QVLD)					
Clock Input capacitance	Cclk	Vclk = 0 V	2.0	3.0	pF
JTAG pins	Cı	V _J = 0 V	2.0	5.0	pF

Remark These parameters are periodically sampled and not 100% tested. Capacitance is not tested on ZQ pin.

Recommended AC Operating Conditions

 $0^{\circ}C \le TC \le 95^{\circ}C$; 1.7 $V \le V_{DD} \le 1.9 V$, unless otherwise noted

Parameter	Symbol	Conditions	MIN.	MAX.	Unit	Note
Input HIGH voltage	VIH (AC)		V _{REF} + 0.2		V	1
Input LOW voltage	VIL (AC)			V _{REF} – 0.2	V	1

Note Overshoot: $V_{IH (AC)} \le V_{DD}Q + 0.7 \text{ V for } t \le t_{CK}/2$

Undershoot: $V_{\text{IL (AC)}} \ge -0.5 \text{ V for } t \le t_{\text{CK}}/2$

Control input signals may not have pulse widths less than tckh (MIN.) or operate at cycle rates less than tck

(MIN.).

DC Characteristics

I_{DD} / I_{SB} Operating Conditions

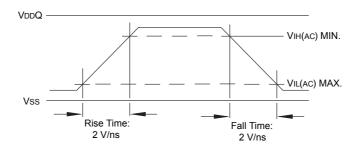
Parameter	Symbol	Test condition				MA	٩X.		Unit
					-E18		-E24		
					tRC=	15ns	tRC=	20ns	
								300MHz	
Standby current	I _{SB1}	tck = Idle	V _{DD}	x9/x18		55	55	55	mA
Standby current	1561	All banks idle, no inputs toggling	V DD	x36	55	55	55	55	111/
		All banks fule, no inputs togging	VEXT	X30	5	5	5	5	
Active standby	I _{SB2}	CS# = HIGH, No commands, half bank / address /	VEXI	x9/x18		215	215	190	mA
•	1582	,	V DD		250	215	215		IIIA
current		data change once every four clock cycles	VEXT	x36	5	5	5	190 5	
Operating aurrent	l	DI 2 cognestial bank cooper bank transitions		x9/x18		_			A
Operating current	DD1	BL=2, sequential bank access, bank transitions	V _{DD}			302	266	239	mA
		once every tRc, half address transitions once		x36	375	344	302	283	
		every t _{RC} , read followed by write sequence,	VEXT		10	10	10	10	
•		continuous data during WRITE commands.		0/ 40					
Operating current	I _{DD2}	BL=4, sequential bank access, bank transitions	V _{DD}	x9/x18		345	288	262	mA
		once every tRC, half address transitions once		x36	433	418	348	339	
		every tRC, read followed by write sequence,	VEXT		10	10	10	10	
	_	continuous data during WRITE commands.							_
Operating current	I _{DD3}	BL=8, sequential bank access, bank transitions	V _{DD}	x9/x18	377	357	299	276	mA
		once every tRc, half address transitions once							
		every tRC, read followed by write sequence,	VEXT		15	15	15	15	
		continuous data during WRITE commands.							
Burst refresh	REF1	Eight bank cyclic refresh, continuous	V _{DD}	x9/x18		464	464	362	mA
current		address/data, command bus remains in refresh		x36	563	432	432	338	
		for all banks	VEXT	1	45	30	30	25	
Disturbed	I _{REF2}	Single bank refresh, sequential bank access,	V _{DD}	x9/x18	260	219	205	173	mA
refresh current		half address transitions once every tRC,		x36	258	216	203	171	
		continuous data	VEXT	•	10	10	10	10	
Operating burst	I _{DD2W}	BL=2, cyclic bank access, half of address bits	V _{DD}	x9/x18	890	691	691	542	mA
write current		change every clock cycle, continuous data,		x36	998	792	792	641	
		measurement is taken during continuous WRITE	VEXT		40	35	35	30	
Operating burst	IDD4W	BL=4, cyclic bank access, half of address bits	V _{DD}	x9/x18	609	478	478	380	mA
write current		change every two clocks, continuous data,		x36	753	608	608	498	
		measurement is taken during continuous WRITE	VEXT		25	20	20	20	
Operating burst	IDD8W	BL=8, cyclic bank access, half of address bits	V _{DD}	x9/x18	478	378	378	303	mA
write current		change every four clocks, continuous data,							
		measurement is taken during continuous WRITE	VEXT	•	25	20	20	20	
Operating burst	I _{DD2R}	BL=2, cyclic bank access, half of address bits	V _{DD}	x9/x18	927	712	712	551	mA
read current		change every clock cycle, measurement is taken		x36	925	710	710	549	
		during continuous READ	VEXT	ı	40	35	35	30	
Operating burst	I _{DD4R}	BL=4, cyclic bank access, half of address bits	V _{DD}	x9/x18	619	477	477	377	mA
read current		change every two clocks, measurement is taken		x36	643	494	494	390	
		during continuous READ	VEXT		25	20	20	20	
Operating burst	I _{DD8R}	BL=8, cyclic bank access, half of address bits	VDD	x9/x18		368	368	292	mA
read current	.DDGK	change every four clocks, measurement is taken	• 55		710	300	300	202	, \
TOUG OUTTOIN			\/	1	25				
		during continuous READ	VEXT		25	20	20	20	<u> </u>

- **Remarks 1.** Idd specifications are tested after the device is properly initialized. $0^{\circ}\text{C} \le \text{Tc} \le 95^{\circ}\text{C}$; $1.7 \text{ V} \le \text{V}_{DD} \le 1.9 \text{ V}$, $2.38 \text{ V} \le \text{V}_{EXT} \le 2.63 \text{ V}$, $1.4 \text{ V} \le \text{V}_{DD}Q \le \text{V}_{DD}$, $\text{V}_{REF} = \text{V}_{DD}Q/2$
 - 2. $t_{CK} = t_{DK} = MIN.$, $t_{RC} = MIN.$
 - 3. Input slew rate is specified in **Recommended DC Operating Conditions** and **Recommended AC Operating Conditions**.
 - **4.** IDD parameters are specified with ODT disabled.
 - 5. Continuous data is defined as half the DQ signals changing between HIGH and LOW every half clock cycles (twice per clock).
 - **6.** Continuous address is defined as half the address signals between HIGH and LOW every clock cycles (once per clock).
 - 7. Sequential bank access is defined as the bank address incrementing by one ever trc.
 - **8.** Cyclic bank access is defined as the bank address incrementing by one for each command access. For BL=4 this is every other clock.
 - **9.** CS# is HIGH unless a READ, WRITE, AREF, or MRS command is registered. CS# never transitions more than per clock cycle.

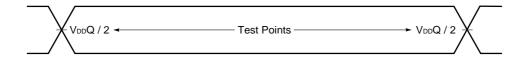
AC Characteristics

AC Test Conditions

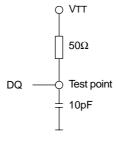
Input waveform



Output waveform



Output load condition



AC Characteristics <Read and Write Cycle>

Parameter	Symbol	-Е	18		-Е	24		Unit	Note
		(533	MHz)	(400	MHz)	(300	MHz)		
		MIN.	MAX.	MIN.	MAX.	MIN.	MAX.		
Clock									
Clock cycle time (CK,CK#,DK,DK#)	tck, tok	1.875	5.7	2.5	5.7	3.3	5.7	ns	
Clock frequency (CK,CK#,DK,DK#)	tck, tok	175	533	175	400	175	300	MHz	
Random Cycle time	t RC	15		15		20		ns	
Clock Jitter: period	t JIT PER	-100	100	-150	150	-200	200	ps	1, 2
Clock Jitter: cycle-to-cycle	t лт сс		200		300		400	ps	
Clock HIGH time (CK,CK#,DK,DK#)	t скн, t дкн	0.45	0.55	0.45	0.55	0.45	0.55	Cycle	
Clock LOW time (CK,CK#,DK,DK#)	tckl, tdkl	0.45	0.55	0.45	0.55	0.45	0.55	Cycle	
Clock to input data clock	tckdk	-0.3	0.3	-0.45	0.5	-0.45	1.0	ns	
Mode register set cycle time	tmrsc	6		6		6		Cycle	
to any command									
PLL Lock time	t CK Lock	15		15		15		μs	
Clock static to PLL reset	tCK Reset	30		30		30		ns	
Output Times									
Output data clock HIGH time	tqкн	0.9	1.1	0.9	1.1	0.9	1.1	t cкн	
Output data clock LOW time	t QKL	0.9	1.1	0.9	1.1	0.9	1.1	t ckl	
QK edge to clock edge skew	tскак	-0.2	0.2	-0.25	0.25	-0.3	0.3	ns	
QK edge to output data edge	t QKQ0, t QKQ1	-0.12	0.12	-0.2	0.2	-0.25	0.25	ns	3, 5
QK edge to any output data	tака	-0.22	0.22	-0.3	0.3	-0.35	0.35	ns	4, 5
QK edge to QVLD	takvld	-0.22	0.22	-0.3	0.3	-0.35	0.35	ns	
Setup Times									
Address/command and input	tas/tcs	0.3		0.4		0.5		ns	
Data-in and data mask to DK	tos	0.17		0.225		0.3		ns	
Hold Times									
Address/command and input	taн/tcн	0.3		0.4		0.5		ns	
Data-in and data mask to DK	tон	0.17		0.225		0.3		ns	

Notes 1. Clock phase jitter is the variance from clock rising edge to the next expected clock rising edge.

- 2. Frequency drift is not allowed.
- 3. t_{QKQ0} is referenced to DQ0–DQ17 in x36 and DQ0–DQ8 in x18. t_{QKQ1} is referenced to DQ18–DQ35 in x36 and DQ9–DQ17 in x18.
- 4. toko takes into account the skew between any QKx and any DQ.
- 5. tqkq, tqkqx are guaranteed by design.

Remark All timing parameters are measured relative to the crossing point of CK/CK#, DK/DK# and to the crossing point with V_{REF} of the command, address, and data signals.

CK# CK COMMAND, VALID VALID **ADDRESS** tas tан tckpk tckpk DKx# DKx tok tokh **t**DKL Don't care

Figure 1-1. Clock / Input Data Clock Command / Address Timings

Temperature and Thermal Impedance

Temperature Limits

Parameter	Symbol	MIN.	MAX.	Unit	Note
Reliability junction temperature	TJ	0	+110	°C	1
Operating junction temperature	TJ	0	+100	°C	2
Operating case temperature	Tc	0	+95	°C	3

- **Notes 1.** Temperatures greater than 110°C may cause permanent damage to the device. This is a stress rating only and functional operation of the device at or above this is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability of the part.
 - 2. Junction temperature depends upon cycle time, loading, ambient temperature, and airflow.
 - **3.** MAX operating case temperature; Tc is measured in the center of the package. Device functionality is not guaranteed if the device exceeds maximum Tc during operation.

Thermal Impedance

Substrate	Ball		θja (°C/W)		θjb	θјс
		Air Flow = 0 m/s	Air Flow = 1 m/s	Air Flow = 2 m/s	(°C/W)	(°C/W)
4 - Layer	Lead free	18.9	15.2	14.4	9.1	3.0

2. Operation

2.1 Command Operation

According to the functional signal description, the following command sequences are possible. All input states or sequences not shown are illegal or reserved. All command and address inputs must meet setup and hold times around the rising edge of CK.

Table 2-1. Address Widths at Different Burst Lengths

Burst Length	Configuration					
	х9	x18	x36			
BL=2	A0-A20	A0-A19	A0–A18			
BL=4	A0-A19	A0-A18	A0–A17			
BL=8	A0–A18	A0-A17	N/A			

Table 2-2. Command Table

Operation	Code	CS#	WE#	REF#	A0-An ^{Note1}	BA0-BA2	Note
Device DESELECT / No Operation	DESEL / NOP	Н	Х	Х	Х	Х	
MRS: Mode Register Set	MRS	L	L	L	OPCODE	Х	2
READ	READ	L	Н	Н	А	ВА	3
WRITE	WRITE	L	L	Н	А	ВА	3
AUTO REFRESH	AREF	L	Н	L	Х	ВА	

Notes 1. n = 20.

- 2. Only A0-A17 are used for the MRS command.
- 3. See Table 2-1.

Remark X = "Don't Care", H = logic HIGH, L = logic LOW, A = valid address, BA = valid bank address

2.2 Description of Commands

DESEL / NOP Note1

The NOP command is used to perform a no operation to the μ PD48288209/18/36AF1, which essentially deselects the chip. Use the NOP command to prevent unwanted commands from being registered during idle or wait states. Operations already in progress are not affected. Output values depend on command history.

MRS

The mode register is set via the address inputs A0–A17. See **Figure 2-5. Mode Register Bit Map** for further information. The MRS command can only be issued when all banks are idle and no bursts are in progress.

READ

The READ command is used to initiate a burst read access to a bank. The value on the BA0–BA2 inputs selects the bank, and the address provided on inputs A0–A20 selects the data location within the bank.

WRITE

The WRITE command is used to initiate a burst write access to a bank. The value on the BA0–BA2 inputs selects the bank, and the address provided on inputs A0–A20 selects the data location within the bank. Input data appearing on the DQ is written to the memory array subject to the DM input logic level appearing coincident with the data. If the DM signal is registered LOW, the corresponding data will be written to memory. If the DM signal is registered HIGH, the corresponding data inputs will be ignored (i.e., this part of the data word will not be written).



AREF

The AREF is used during normal operation of the μ PD48288209/18/36AF1 to refresh the memory content of a bank. The command is non-persistent, so it must be issued each time a refresh is required. The value on the BA0–BA2 inputs selects the bank. The refresh address is generated by an internal refresh controller, effectively making each address bit a "Don't Care" during the AREF command. The μ PD48288209/18/36AF1 requires 64K cycles at an average periodic interval of 0.49 μ s Note2 (MAX.). To improve efficiency, eight AREF commands (one for each bank) can be posted to μ PD48288209/18/36AF1 at periodic intervals of 3.9 μ s Note3.

Within a period of 32 ms, the entire memory must be refreshed. The delay between the AREF command and a subsequent command to same bank must be at least trace as continuous refresh. Other refresh strategies, such as burst refresh, are also possible.

Notes 1. When the chip is deselected, internal NOP commands are generated and no commands are accepted.

- **2.** Actual refresh is 32 ms / 8k / 8 = 0.488 μ s.
- 3. Actual refresh is 32 ms / 8k = 3.90 μ s.

2.3 Initialization

The μ PD48288209/18/36AF1 must be powered up and initialized in a predefined manner. Operational procedures other than those specified may result in undefined operations or permanent damage to the device. The following sequence is used for Power-Up:

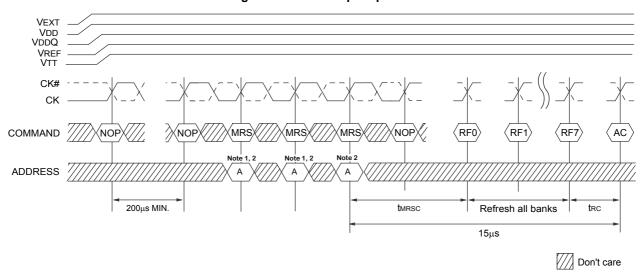
1. Apply power (Vext, Vdd, Vdd, Vref, Vtt) and start clock as soon as the supply voltages are stable. Apply Vdd and Vext before or at the same time as Vdd. Apply Vdd before or at the same time as Vref and Vtt. Although there is no timing relation between Vext and Vdd, the chip starts the power-up sequence only after both voltages are at their nominal levels. Vdd supply must not be applied before Vdd supply. CK/CK# must meet Vdd(DC) prior to being applied. Maintain all remaining balls in NOP conditions.

Note No rule of apply power sequence is the design target.

- 2. Maintain stable conditions for 200 μ s (MIN.).
- **3.** Issue at least three or more consecutive MRS commands: two dummies or more plus one valid MRS. It is recommended that all address pins are held LOW during the dummy MRS commands.
- **4.** t_{MRSC} after valid MRS, an AUTO REFRESH command to all 8 banks must be issued and wait for 15 μs with CK/CK# toggling in order to lock the PLL prior to normal operation.
- **5.** After tRC, the chip is ready for normal operation.

2.4 Power-On Sequence

Figure 2-1. Power-Up Sequence



Notes 1. Recommended all address pins held LOW during dummy MRS commands.

2. A10-A17 must be LOW.

Remark MRS: MRS command

RFp: REFRESH bank p

AC: Any command

2.5 Programmable Impedance Output Buffer

The μ PD48288209/18/36AF1 is equipped with programmable impedance output buffers. This allows a user to match the driver impedance to the system. To adjust the impedance, an external precision resistor (RQ) is connected between the ZQ ball and Vss. The value of the resistor must be five times the desired impedance. For example, a 300 Ω resistor is required for an output impedance of 60 Ω . To ensure that output impedance is one fifth the value of RQ (within 15 percent), the range of RQ is 125 Ω to 300 Ω . Output impedance updates may be required because, over time, variations may occur in supply voltage and temperature. The device samples the value of RQ. An impedance update is transparent to the system and does not affect device operation. All data sheet timing and current specifications are met during an update.

2.6 PLL Reset

The μ PD48288209/18/36AF1 utilizes internal Phase-locked loops for maximum output, data valid windows. It can be placed into a stopped-clock state to minimize power with a modest restart time of 15 μ s. The clock (CK/CK#) must be toggled for 15 μ s in order to stabilize PLL circuits for next READ operation.

2.7 Clock Input

Table 2-3. Clock Input Operation Conditions

Parameter	Symbol	Conditions	MIN.	MAX.	Unit	Note
Clock Input Voltage Level	VIN (DC)	CK and CK#	-0.3	V _{DD} Q + 0.3	٧	
Clock Input Differential Voltage Level	VID (DC)	CK and CK#	0.2	V _{DD} Q + 0.6	٧	8
Clock Input Differential Voltage Level	VID (AC)	CK and CK#	0.4	V _{DD} Q + 0.6	٧	8
Clock Input Crossing Point Voltage Level	VIX (AC)	CK and CK#	V _{DD} Q/2 - 0.15	V _{DD} Q/2 + 0.15	V	9

VIN(DC) MAX.

CK#

VDDQ/2 + 0.15

VDDQ/2 - 0.15

VIX(AC) MAX.

VID(DC)

VID(DC)

VID(DC)

VID(DC)

VIX(AC) MIN.

Minimum Clock Level

Figure 2-2. Clock Input

Notes 1. DKx and DKx# have the same requirements as CK and CK#.

- 2. All voltages referenced to Vss.
- **3.** Tests for AC timing, IDD and electrical AC and DC characteristics may be conducted at normal reference/supply voltage levels; but the related specifications and device operations are tested for the full voltage range specified.
- **4.** AC timing and IDD tests may use a V_{IL} to V_{IH} swing of up to 1.5 V in the test environment, but input timing is still referenced to V_{REF} (or the crossing point for CK/CK#), and parameters specifications are tested for the specified AC input levels under normal use conditions. The minimum slew rate for the input signals used to test the device is 2V/ns in the range between V_{IL(AC)} and V_{IH(AC)}.
- 5. The AC and DC input level specifications are as defined in the HSTL Standard (i.e. the receiver will effectively switch as a result of the signal crossing the AC input level, and will remain in that state as long as the signal does not ring back above[below] the DC input LOW[HIGH] level).
- **6.** The CK/CK# input reference level (for timing referenced to CK/CK#) is the point at which CK and CK# cross. The input reference level for signal other than CK/CK# is V_{REF}.
- 7. CK and CK# input slew rate must be $\geq 2V/ns$ ($\geq 4V/ns$ if measured differentially).
- 8. V_{ID} is the magnitude of the difference between the input level on CK and input level on CK#.
- 9. The value of V_{IX} is expected to equal V_{DD}Q/2 of the transmitting device and must track variations in the DC level of the same.

10.CK and CK# must cross within the region.

- 11.CK and CK# must meet at least VID(DC) (MIN.) when static and centered around VDDQ/2.
- 12. Minimum peak-to-peak swing.

2.8 Mode Register Set Command (MRS)

The mode register stores the data for controlling the operating modes of the memory. It programs the μ PD48288209/18/36AF1 configuration, burst length, and I/O options. During a MRS command, the address inputs A0–A17 are sampled and stored in the mode register. tmrsc must be met before any command can be issued to the μ PD48288209/18/36AF1. The mode register may be set at any time during device operation. However, any pending operations are not guaranteed to successfully complete, and all memory cell data are not guaranteed.

Since MRS is used for internal test mode entry, bits A10-A17 must be set to all "0" at the MRS setting.

CK# - KMRSC COMMAND CO

Figure 2-3. Mode Register Set Timing

Remark MRS: MRS command

AC: any command

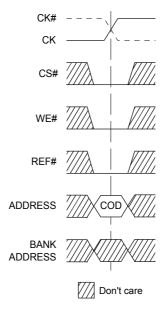


Figure 2-4. Mode Register Set

Remark COD: code to be loaded into the register.

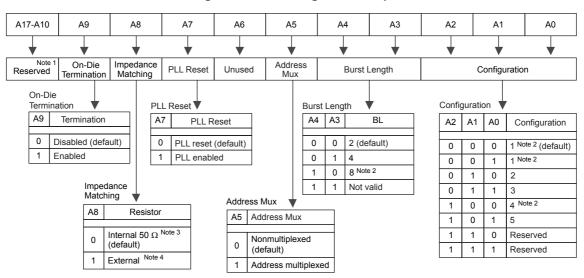


Figure 2-5. Mode Register Bit Map

Notes 1. Bits A10–A17 must be set to all '0'. A18-An are "Don't Care".

- **2.** BL=8 is not available for configuration 1 and 4.
- 3. $\pm 30\%$ temperature variation.
- 4. Within 15%.

2.9 Read & Write configuration (Non Multiplexed Address Mode)

Table 2-4 shows, for different operating frequencies, the different μ PD48288209/18/36AF1 configurations that can be programmed into the mode register. The READ and WRITE latency (trl and twl) values along with the row cycle times (trl) are shown in clock cycles as well as in nanoseconds.

Parameter		Configuration						
	1 Note1	2	3	4 Note1, 2	5			
trc	4	6	8	3	5	tск		
t RL	4	6	8	3	5	t cĸ		
twL	5	7	9	4	6	t cĸ		
Valid frequency range	266-175	400-175	533-175	200-175	333-175	MHz		

Table 2-4. Configuration Table

Notes 1. BL= 8 is not available.

2. The minimum trc is typically 3 cycles, except in the case of a WRITE followed by a READ to the same bank. In this instance the minimum trc is 4 cycles.

2.10 Write Operation (WRITE)

Write accesses are initiated with a WRITE command, as shown in **Figure 2-6**. Row and bank addresses are provided together with the WRITE command. During WRITE commands, data will be registered at both edges of DK according to the programmed burst length (BL). A WRITE latency (WL) one cycle longer than the programmed READ latency (RL + 1) is present, with the first valid data registered at the first rising DK edge WL cycles after the WRITE command.

Any WRITE burst may be followed by a subsequent READ command. Figure 2-10. WRITE Followed By READ: BL=2, RL=4, WL=5, Configuration 1 and Figure 2-11. WRITE Followed By READ: BL=4, RL=4, WL=5, Configuration 1 illustrate the timing requirements for a WRITE followed by a READ for bursts of two and four, respectively.

Setup and hold times for incoming input data relative to the DK edges are specified as t_{DS} and t_{DH}. The input data is masked if the corresponding DM signal is HIGH. The setup and hold times for data mask are also t_{DS} and t_{DH}.

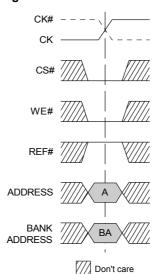


Figure 2-6. WRITE Command

Remark A: Address

BA: Bank address

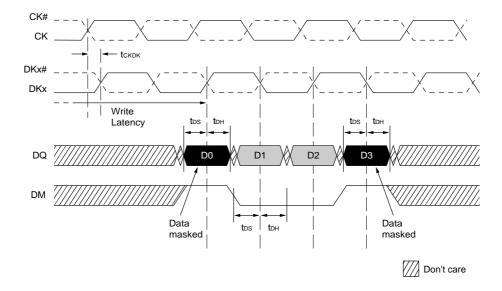


Figure 2-7. Basic WRITE Burst / DM Timing

Figure 2-8. WRITE Burst Basic Sequence: BL=2, RL=4, WL=5, Configuration 1

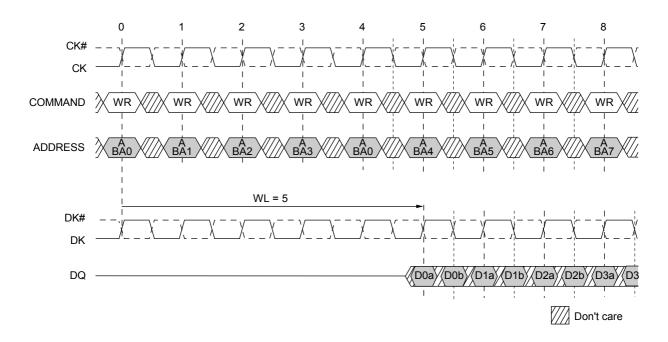
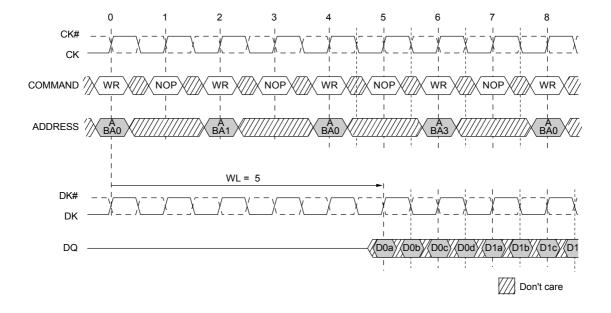


Figure 2-9. WRITE Burst Basic Sequence: BL=4, RL=4, WL=5, Configuration 1



Remarks 1. WR : WRITE command A/Bap : Address A of bank p WL : WRITE latency

: Data q to bank p

Dpq

2. Any free bank may be used in any given command. The sequence shown is only one example of a bank sequence.

COMMAND ADDRESS X A RL = 4 (D0a)(D0b) DQ QVLD QKx ₩ Undefined Don't care

Figure 2-10. WRITE Followed By READ: BL=2, RL=4, WL=5, Configuration 1

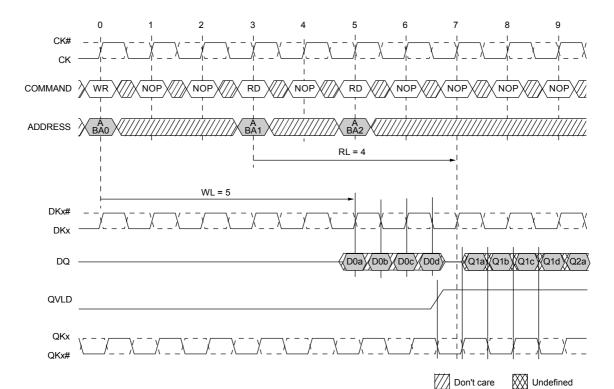


Figure 2-11. WRITE Followed By READ: BL=4, RL=4, WL=5, Configuration 1

Remark

WR

WL RL

Dpq

: WRITE command : READ command A/BAp: Address A of bank p : WRITE latency

: READ latency

: Data q to bank p Qpq : Data q from bank p

2.11 Read Operation (READ)

Read accesses are initiated with a READ command, as shown in **Figure 2-12**. Row and bank addresses are provided with the READ command.

During READ bursts, the memory device drives the read data edge-aligned with the QK signal. After a programmable READ latency, data is available at the outputs. The data valid signal indicates that valid data will be present in the next half clock cycle.

The skew between QK and the crossing point of CK is specified as tCKQK. tQKQO is the skew between QKO and the last valid data edge considered the data generated at the DQ0–DQ17 in x36 and DQ0–DQ8 in x18 data signals. tQKQI is the skew between QK1 and the last valid data edge considered the data generated at the DQ18–DQ35 in x36 and DQ9–DQ17 in x18 data signals. tQKQx is derived at each QKx clock edge and is not cumulative over time.

After completion of a burst, assuming no other commands have been initiated, DQ will go High-Z. Back-to-back READ commands are possible, producing a continuous flow of output data.

Minimum READ data valid window can be expressed as MIN.(tokh, tokh) – 2 x MAX.(tokox)

Any READ burst may be followed by a subsequent WRITE command. Figure 2-16. READ followed by WRITE, BL=2, RL=4, WL=5, Configuration 1 and Figure 2-17. READ followed by WRITE, BL=4, RL=4, WL=5, Configuration 1 illustrate the timing requirements for a READ followed by a WRITE.

CK# ---CK

CS#

WE#

ADDRESS

BANK
ADDRESS

Don't care

Figure 2-12. READ Command

Remark A : Address

BA: Bank address

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Figure 2-13. Basic READ Burst Timing

Note Minimum READ data valid window can be expressed as MIN.(t_{QKH} , t_{QKL}) – 2 x MAX.(t_{QKQx}) tckh and tckl are recommended to have 50% / 50% duty.

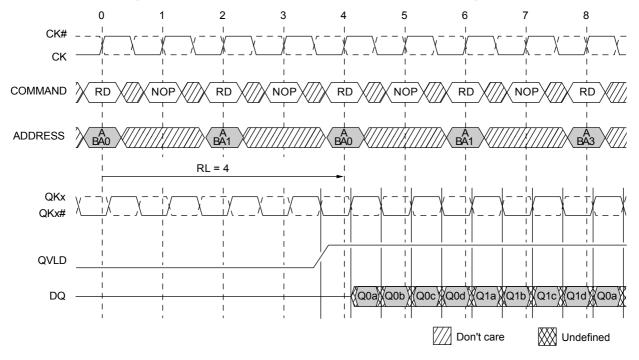
- **Remarks** 1. t_{QKQ0} is referenced to DQ0–DQ17 in x36 and DQ0–DQ8 in x18. t_{QKQ1} is referenced to DQ18–DQ35 in x36 and DQ9–DQ17 in x18.
 - 2. toko takes into account the skew between any QKx and any DQ.
 - 3. tckqk is specified as CK rising edge to QK rising edge.

Undefined

CK# CK COMMAND RD RD RD RD RD RD RD RD RD **RL = 4** QKx QKx# QVLD DQ Don't care Undefined

Figure 2-14. READ Burst Basic Sequence: BL=2, RL=4, Configuration 1





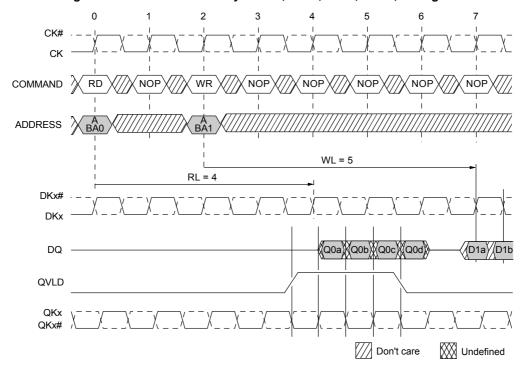
Remark RD : READ command

A/BAp : Address A of bank p
RL : READ latency
Qpq : Data q from bank p

COMMAND > WR NOP NOP NOP NOP ADDRESS X A WL = 5 DKx# DKx D1a D1b D2a D2b DQ QVLD QKx QKx# Don't care Undefined

Figure 2-16. READ followed by WRITE, BL=2, RL=4, WL=5, Configuration 1

Figure 2-17. READ followed by WRITE, BL=4, RL=4, WL=5, Configuration 1



Remark WR : WRITE command

RD : READ command

A/BAp : Address A of bank p

WL : WRITE latency

RL : READ latency

Dpq : Data q to bank p

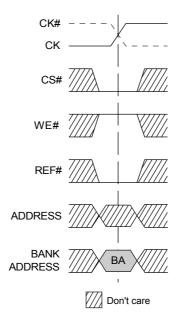
Qpq : Data q from bank p

2.12 Refresh Operation: AUTO REFRESH Command (AREF)

AREF is used to perform a REFRESH cycle on one row in a specific bank. The row addresses are generated by an internal refresh counter; external address balls are "Don't Care." The delay between the AREF command and a subsequent command to the same bank must be at least t_{RC} .

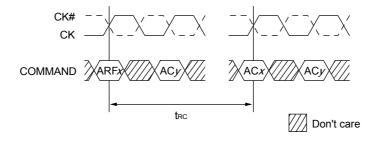
Within a period of 32 ms (treef), the entire memory must be refreshed. **Figure 2-19** illustrates an example of a continuous refresh sequence. Other refresh strategies, such as burst refresh, are also possible.

Figure 2-18. AUTO REFRESH Command



Remark BA: Bank address

Figure 2-19. AUTO REFRESH Cycle



Remarks 1. ACx : Any command on bank x

ARFx : Auto refresh bank x

ACy : Any command on different bank.

2. trc is configuration-dependent. Refer to Table 2-4. Configuration Table.

2.13 On-Die Termination

On-die termination (ODT) is enabled by setting A9 to "1" during an MRS command. With ODT on, all the DQs and DM are terminated to V_{TT} with a resistance R_{TT}. The command, address, and clock signals are not terminated. **Figure 2-20** below shows the equivalent circuit of a DQ receiver with ODT. ODTs are dynamically switched off during READ commands and are designed to be off prior to the μ PD48288209/18/36AF1 driving the bus. Similarly, ODTs are designed to switch on after the μ PD48288209/18/36AF1 has issued the last piece of data.

Table 2-5. On-Die Termination DC Parameters

Description	Symbol	MIN.	MAX.	Units	Note
Termination voltage	VTT	0.95 x VREF	1.05 x VREF	V	1, 2
On-Die termination	RTT	125	185	Ω	3

Notes 1. All voltages referenced to Vss (GND).

- 2. Vtt is expected to be set equal to Vref and must track variations in the DC level of Vref.
- 3. The R_{TT} value is measured at 95°C Tc.

Figure 2-20. On- Die Termination-Equivalent Circuit

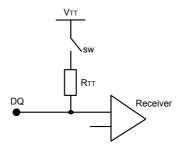
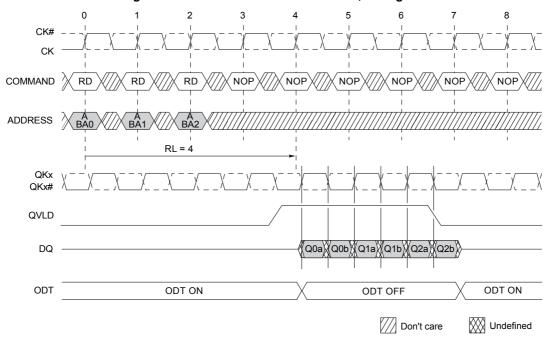


Figure 2-21. READ Burst with ODT: BL=2, Configuration 1



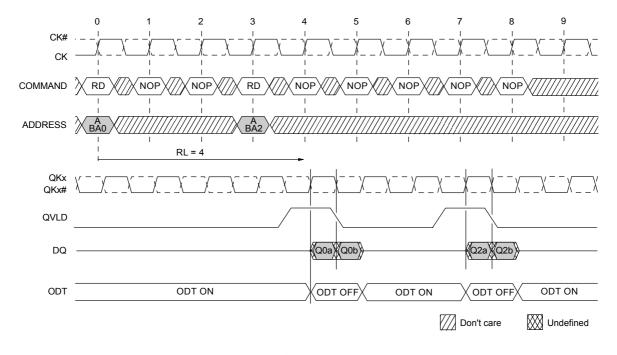
Remark RD : READ command

A/Bap : Address A of bank p
RL : READ latency
Qpq : Data q from bank p

CK# COMMAND / RD NOP NOP NOP NOP NOP ADDRESS **RL = 4** QKx QKx# QVLD DQ Q0a)(Q0b) (Q2a)(Q2b) ODT ODT ON ODT OFF ODT ON ODT OFF Don't care W Undefined

Figure 2-22. READ NOP READ with ODT: BL=2, Configuration 1

Figure 2-23. READ NOP NOP READ with ODT: BL=2, Configuration 1



Remark RD : READ command

A/BAp : Address A of bank p
RL : READ latency
Qpq : Data q from bank p

Figure 2-24. READ followed by WRITE with ODT: BL=2, Configuration 1

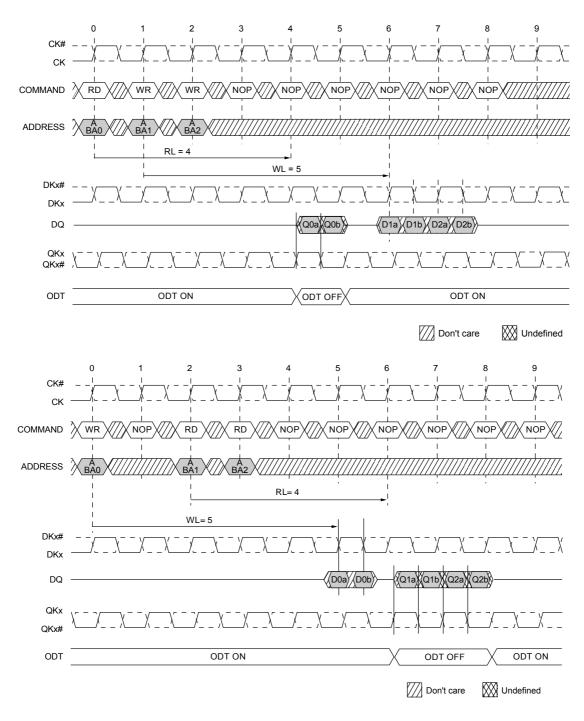


Figure 2-25. WRITE followed by READ with ODT: BL=2, Configuration 1

Remark RD : READ command

WR : WRITE command

A/BAp : Address A of bank p

RL : READ latency

WL : WRITE latency

Qpq : Data q from bank p

Dpq : Data q to bank p

2.14 Operation with Multiplexed Address

In multiplexed address mode, the address can be provided to the μ PD48288209/18/36AF1 in two parts that are latched into the memory with two consecutive rising clock edges. This provides the advantage that a maximum of 11 address balls are required to control the μ PD48288209/18/36AF1, reducing the number of balls on the controller side. The data bus efficiency in continuous burst mode is not affected for BL=4 and BL=8 since at least two clocks are required to read the data out of the memory. The bank addresses are delivered to the μ PD48288209/18/36AF1 at the same time as the WRITE command and the first address part, Ax.

This option is available by setting bit A5 to "1" in the mode register. Once this bit is set, the READ, WRITE, and MRS commands follow the format described in **Figure 2-26**. See **Figure 2-28**. **Power-Up Sequence in Multiplexed Address Mode** for the power-up sequence.

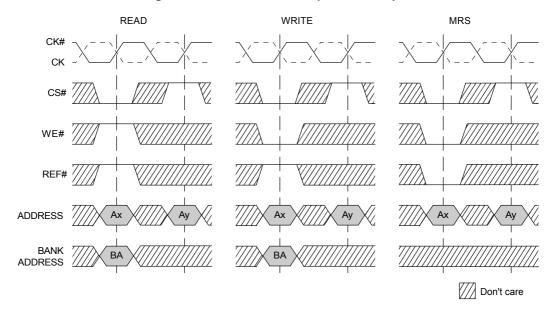


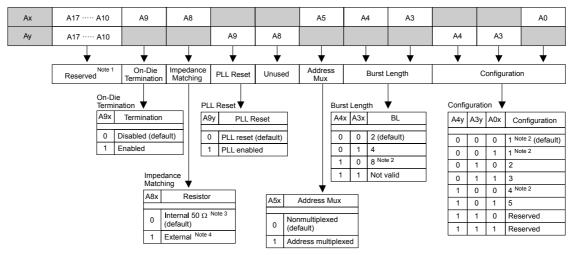
Figure 2-26. Command Description in Multiplexed

Remarks 1. Ax, Ay: Address

BA : Bank Address

2. The minimum setup and hold times of the two address parts are defined tas and tah.

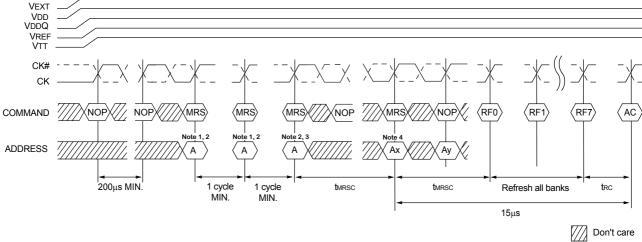
Figure 2-27. Mode Register Set Command in Multiplexed Address Mode



- Notes 1. Bits A10–A17 must be set to all '0'.
 - **2.** BL=8 is not available for configuration 1 and 4.
 - 3. $\pm 30\%$ temperature variation.
 - 4. Within 15%.

Remark The address A0, A3, A4, A5, A8, and A9 must be set as follows in order to activate the mode register in the multiplexed address mode.

Figure 2-28. Power-Up Sequence in Multiplexed Address Mode



- Notes 1. Recommended all address pins held LOW during dummy MRS command.
 - 2. A10-A17 must be LOW.
 - 3. Address A5 must be set HIGH (muxed address mode setting when μ PD48288209/18/36AF1 is in normal mode of operation).
 - **4.** Address A5 must be set HIGH (muxed address mode setting when μ PD48288209/18/36AF1 is already in muxed address mode).

Remark MRS: MRS command RFp: REFRESH Bank p AC: any command

2.15 Address Mapping in Multiplexed Mode

The address mapping is described in **Table 2-6** as a function of data width and burst length.

Table 2-6. Address Mapping in Multiplexed Address Mode

Data	Burst	Ball	Address										
Width	Length		Α0	А3	A4	A5	A8	A9	A10	A13	A14	A17	A18
x36	BL=2	Ax	A0	А3	A4	A5	A8	A9	A10	A13	A14	A17	A18
		Ау	Х	A1	A2	Χ	A6	A7	Χ	A11	A12	A16	A15
	BL=4	Ax	A0	А3	A4	A5	A8	A9	A10	A13	A14	A17	Х
		Ау	Χ	A1	A2	Χ	A6	A7	Χ	A11	A12	A16	A15
x18	BL=2	Ax	Α0	А3	A4	A5	A8	A9	A10	A13	A14	A17	A18
		Ау	Χ	A1	A2	Χ	A6	A7	A19	A11	A12	A16	A15
	BL=4	Ax	A0	А3	A4	A5	A8	A9	A10	A13	A14	A17	A18
		Ау	Χ	A1	A2	Χ	A6	A7	Х	A11	A12	A16	A15
	BL=8	Ax	A0	А3	A4	A5	A8	A9	A10	A13	A14	A17	Х
		Ау	Χ	A1	A2	Χ	A6	A7	Х	A11	A12	A16	A15
х9	BL=2	Ax	A0	А3	A4	A5	A8	A9	A10	A13	A14	A17	A18
		Ау	A20	A1	A2	Χ	A6	A7	A19	A11	A12	A16	A15
	BL=4	Ax	A0	А3	A4	A5	A8	A9	A10	A13	A14	A17	A18
		Ау	Χ	A1	A2	Χ	A6	A7	A19	A11	A12	A16	A15
	BL=8	Ax	A0	А3	A4	A5	A8	A9	A10	A13	A14	A17	A18
		Ау	Χ	A1	A2	X	A6	A7	Х	A11	A12	A16	A15

Remark X means "Don't care".

2.16 Read & Write configuration in Multiplexed Address Mode

In multiplexed address mode, the READ and WRITE latencies are increased by one clock cycle. The μ PD48288209/18/36AF1 cycle time remains the same, as described in **Table 2-7**.

400-175

Parameter Configuration Unit 4 Note1,2 Note1 5 1 2 3 4 6 8 3 5 tro tcĸ 5 7 9 4 6 **t**RL **t**cĸ 6 8 10 5 7 twL **t**cĸ

Table 2-7. Configuration in Multiplexed Address Mode

533-175

200-175

333-175

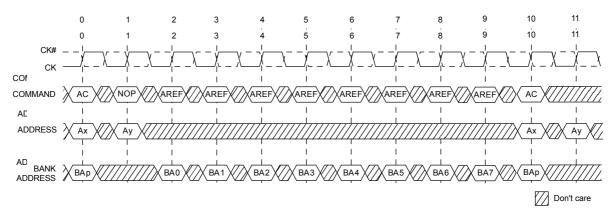
MHz

2.17 Refresh Command in Multiplexed Address Mode

266-175

Similar to other commands, the refresh command is executed on the next rising clock edge when in the multiplexed address mode. However, since only bank address is required for AREF, the next command can be applied on the following clock. The operation of the AREF command and any other command is represented in **Figure 2-29**.

Figure 2-29. Burst REFRESH Operation



Remark AREF : AUTO REFRESH

AC : Any command

Ax : First part Ax of address Ay : Second part Ay of address

BAp : Bank p is chosen so that tree is met.

Valid frequency range

Notes 1. BL = 8 is not available.

^{2.} The minimum trc is typically 3 cycles, except in the case of a WRITE followed by a READ to the same bank. In this instance the minimum trc is 4 cycles.

Figure 2-30. WRITE Burst Basic Sequence: BL=4, with Multiplexed Addresses, Configuration 1

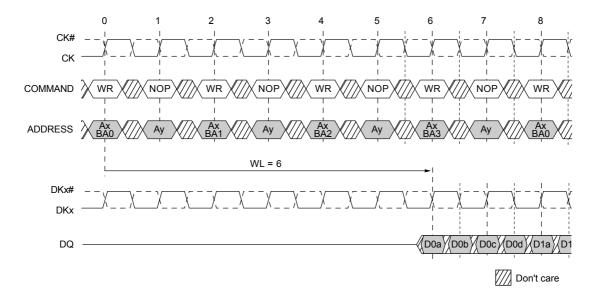
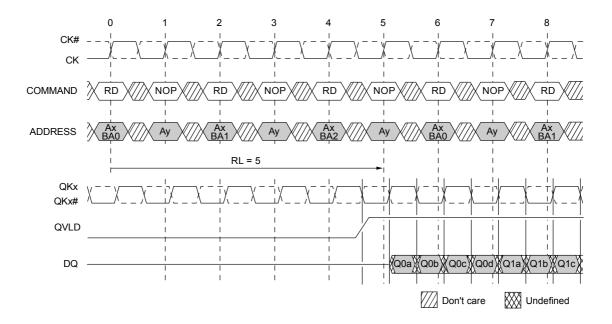


Figure 2-31. READ Burst Basic Sequence: BL=4, with Multiplexed Addresses, Configuration 1, RL=5



Remark WR : WRITE command

RD : READ command

Ax/BAp : Address Ax of bank p

Ay : Address Ay of bank p

Dpq : Data q to bank p

Qpq : Data q from bank p

WL : WRITE latency

RL : READ latency

2.18 Input Slew Rate Derating

Table 2-8 on page 40 and **Table 2-9** on page 41 define the address, command, and data setup and hold derating values. These values are added to the default tAS/tCS/tDS and tAH/tCH/tDH specifications when the slew rate of any of these input signals is less than the 2 V/ns the nominal setup and hold specifications are based upon.

To determine the setup and hold time needed for a given slew rate, add the tAS/tCS default specification to the "tAS/tCS VREF to CK/CK# Crossing" and the tAH/tCH default specification to the "tAH/tCH CK/CK# Crossing to VREF" derated values on **Table 2-8**. The derated data setup and hold values can be determined in a like manner using the "tDS VREF to CK/CK# Crossing" and "tDH to CK/CK# Crossing to VREF" values on **Table 2-9**.

The derating values on **Table 2-8** and **Table 2-9** apply to all speed grades.

The setup times on **Table 2-8** and **Table 2-9** represent a rising signal. In this case, the time from which the rising signal crosses V_{IH(AC)} MIN to the CK/CK# cross point is static and must be maintained across all slew rates. The derated setup timing represents the point at which the rising signal crosses V_{REF(DC)} to the CK/CK# cross point. This derated value is calculated by determining the time needed to maintain the given slew rate and the delta between V_{IH(AC)} MIN and the CK/CK# cross point. The setup values in **Table 2-8** and **Table 2-9** are also valid for falling signals (with respect to V_{IL[AC]} MAX and the CK/CK# cross point).

The hold times in **Table 2-8** and **Table 2-9** represent falling signals. In this case, the time from the CK/CK# cross point to when the signal crosses V_{IH(DC)} MIN is static and must be maintained across all slew rates. The derated hold timing represents the delta between the CK/CK# cross point to when the falling signal crosses V_{REF(DC)}. This derated value is calculated by determining the time needed to maintain the given slew rate and the delta between the CK/CK# cross point and V_{IH(DC)}. The hold values in **Table 2-8** and **Table 2-9** are also valid for rising signals (with respect to V_{IL[DC]} MAX and the CK and CK# cross point).

Note The above descriptions also pertain to data setup and hold derating when CK/CK# are replaced with DK/DK#.

Table 2-8. Address and Command Setup and Hold Derating Values

Command/ Address Slew Rate (V/ns)	tAS/tCS VREF to CK/CK# Crossing	tAS/tCS VIH(AC) MIN to CK/CK# Crossing	tAH/tCH CK/CK# Crossing to VREF	tAH/tCH CK/CK# Crossing to VIH(DC) MIN	Unit
	(CK, CK# Differential SI	lew Rate: 2.0 V/ns		
2.0	0	-100	0	-50	ps
1.9	5	-100	3	-50	ps
1.8	11	-100	6	-50	ps
1.7	18	-100	9	-50	ps
1.6	25	-100	13	-50	ps
1.5	33	-100	17	-50	ps
1.4	43	-100	22	-50	ps
1.3	54	-100	27	-50	ps
1.2	67	-100	34	-50	ps
1.1	82	-100	41	-50	ps
1.0	100	-100	50	-50	ps
	(CK, CK# Differential SI	lew Rate: 1.5 V/ns		
2.0	30	-70	30	-20	ps
1.9	35	-70	33	-20	ps
1.8	41	-70	36	-20	ps
1.7	48	-70	39	-20	ps
1.6	55	-70	43	-20	ps
1.5	63	-70	47	-20	ps
1.4	73	-70	52	-20	ps
1.3	84	-70	57	-20	ps
1.2	97	-70	64	-20	ps
1.1	112	-70	71	-20	ps
1.0	130	-70	80	-20	ps
	(CK, CK# Differential SI	lew Rate: 1.0 V/ns		
2.0	60	-40	60	10	ps
1.9	65	-40	63	10	ps
1.8	71	-40	66	10	ps
1.7	78	-40	69	10	ps
1.6	85	-40	73	10	ps
1.5	93	-40	77	10	ps
1.4	103	-40	82	10	ps
1.3	114	-40	87	10	ps
1.2	127	-40	94	10	ps
1.1	142	-40	101	10	ps
1.0	160	-40	110	10	ps

Table 2-9. Data Setup and Hold Derating Values

Data Slew Rate (V/ns)	tDS VREF to DK/DK# Crossing	tDS VIH(AC) MIN to DK/DK# Crossing	tDH DK/DK# Crossing to VREF	tDH DK/DK# Crossing to VIH(DC) MIN	Unit
		DK, DK# Differential	Slew Rate: 2.0 V/ns		
2.0	0	-100	0	-50	ps
1.9	5	-100	3	-50	ps
1.8	11	-100	6	-50	ps
1.7	18	-100	9	-50	ps
1.6	25	-100	13	-50	ps
1.5	33	-100	17	-50	ps
1.4	43	-100	22	-50	ps
1.3	54	-100	27	-50	ps
1.2	67	-100	34	-50	ps
1.1	82	-100	41	-50	ps
1.0	100	-100	50	-50	ps
		DK, DK# Differential	Slew Rate: 1.5 V/ns		
2.0	30	-70	30	-20	ps
1.9	35	-70	33	-20	ps
1.8	41	-70	36	-20	ps
1.7	48	-70	39	-20	ps
1.6	55	-70	43	-20	ps
1.5	63	-70	47	-20	ps
1.4	73	-70	52	-20	ps
1.3	84	-70	57	-20	ps
1.2	97	-70	64	-20	ps
1.1	112	-70	71	-20	ps
1.0	130	-70	80	-20	ps
		DK, DK# Differential	Slew Rate: 1.0 V/ns		
2.0	60	-40	60	10	ps
1.9	65	-40	63	10	ps
1.8	71	-40	66	10	ps
1.7	78	-40	69	10	ps
1.6	85	-40	73	10	ps
1.5	93	-40	77	10	ps
1.4	103	-40	82	10	ps
1.3	114	-40	87	10	ps
1.2	127	-40	94	10	ps
1.1	142	-40	101	10	ps
1.0	160	-40	110	10	ps

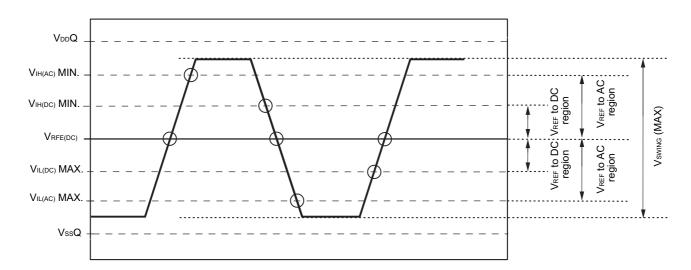


Figure 2-32. Nominal tAS/tCS/tDS and tAH/tCH/tDH Slew Rate

3. JTAG Specification

These products support a limited set of JTAG functions as in IEEE standard 1149.1.

Table 3-1. Test Access Port (TAP) Pins

Pin name	Pin assignments	Description
TCK	12A	Test Clock Input. All input are captured on the rising edge of TCK and all outputs propagate from the falling edge of TCK.
TMS	11A	Test Mode Select. This is the command input for the TAP controller state
TDI	12V	Test Data Input. This is the input side of the serial registers placed between TDI and TDO. The register placed between TDI and TDO is determined by the state of the TAP controller state machine and the instruction that is currently loaded in
TDO	11V	Test Data Output. This is the output side of the serial registers placed between TDI and TDO. Output changes in response to the falling edge of TCK.

Remark The device does not have TRST (TAP reset). The Test-Logic Reset state is entered while TMS is held HIGH for five rising edges of TCK. The TAP controller state is also reset on the POWER-UP.

Table 3-2. JTAG DC Characteristics (0°C ≤ Tc ≤ 95°C, 1.7 V ≤ VDD ≤ 1.9 V, unless otherwise noted)

Parameter	Symbol	Conditions	MIN.	MAX.	Unit	Notes
JTAG Input leakage current	lц	$0 \text{ V} \leq V_{IN} \leq V_{DD}$	-5.0	+5.0	μА	
JTAG I/O leakage current	I _{LO}	$0\ V \leq V_{IN} \leq V_{DD}\ Q\ ,$	-5.0	+5.0	μΑ	
		Outputs disabled				
JTAG input HIGH voltage	ViH		V _{REF} + 0.15	V _{DD} + 0.3	V	1, 2
JTAG input LOW voltage	VIL		VssQ - 0.3	V _{REF} – 0.15	V	1, 2
JTAG output HIGH voltage	V _{OH1}	I _{OHC} = 100 μA	V _{DD} Q - 0.2		V	
	V _{OH2}	I _{OHT} = 2 mA	$V_{DD}Q - 0.4$		V	
JTAG output LOW voltage	V _{OL1}	I _{OLC} = 100 μA		0.2	V	1
	V _{OL2}	I _{OLT} = 2 mA		0.4	V	1

Notes 1. All voltages referenced to Vss (GND).

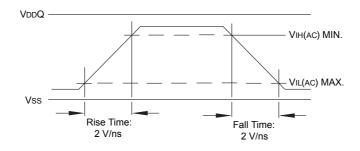
2. Overshoot: $V_{IH (AC)} \le V_{DD} + 0.7 \text{ V for } t \le t_{CK}/2.$

Undershoot: $V_{\text{IL (AC)}} \ge -0.5 \text{ V for } t \le t_{\text{CK}}/2.$

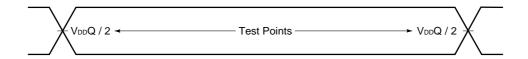
During normal operation, VDDQ must not exceed VDD.

JTAG AC Test Conditions

Input waveform



Output waveform



Output load condition

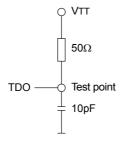


Table 3-3. JTAG AC Characteristics (0°C ≤ Tc ≤ 95°C)

Parameter	Symbol	Conditions	MIN.	MAX.	Unit	Note
Clock						
Clock cycle time	tтнтн		20		ns	
Clock frequency	f _{TF}			50	MHz	
Clock HIGH time	tthtl		10		ns	
Clock LOW time	tтьтн		10		ns	
Output time]					
TCK LOW to TDO unknown	t TLOX		0		ns	
TCK LOW to TDO valid	t _{TLOV}			10	ns	
	_					
Setup time						
TMS setup time	t _{MVTH}		5		ns	
TDI valid to TCK HIGH	tоvтн		5		ns	
Capture setup time	tcsJ		5		ns	1
	-					
Hold time						
TMS hold time	tтнмх		5		ns	
TCK HIGH to TDI invalid	t _{THDX}		5		ns	
Capture hold time	tchu		5		ns	1

Note tcsj and tchj refer to the setup and hold time requirements of latching data from the boundary scan register.

JTAG Timing Diagram

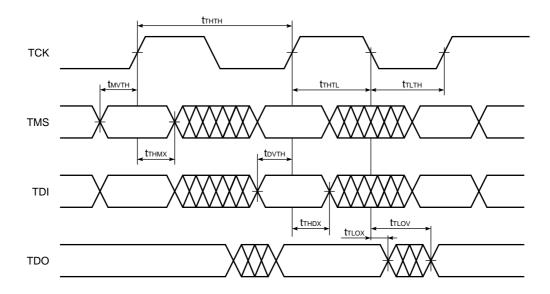


Table 3-4. Scan Register Definition (1)

Register name	Description
Instruction register	The 8 bit instruction registers hold the instructions that are executed by the TAP controller. The register can be loaded when it is placed between the TDI and TDO pins. The instruction register is automatically preloaded with the IDCODE instruction at power-up whenever the controller is placed in test-logic-reset state.
Bypass register	The bypass register is a single bit register that can be placed between TDI and TDO. It allows serial test data to be passed through the RAMs TAP to another device in the scan chain with as little delay as possible. The bypass register is set LOW (VSS) when the bypass instruction is executed.
ID register	The ID Register is a 32 bit register that is loaded with a device and vendor specific 32 bit code when the controller is put in capture-DR state with the IDCODE command loaded in the instruction register. The register is then placed between the TDI and TDO pins when the controller is moved into shift-DR state.
Boundary register	The boundary register, under the control of the TAP controller, is loaded with the contents of the RAMs I/O ring when the controller is in capture-DR state and then is placed between the TDI and TDO pins when the controller is moved to shift-DR state. Several TAP instructions can be used to activate the boundary register. The Scan Exit Order tables describe which device bump connects to each boundary register location. The first column defines the bit's position in the boundary register. The second column is the name of the input or I/O at the bump and the third column is the bump number.

Table 3-5. Scan Register Definition (2)

Register name	Bit size	Unit
Instruction register	8	bit
Bypass register	1	bit
ID register	32	bit
Boundary register	113	bit

Table 3-6. ID Register Definition

Part number	Part number Organization		ID [27:12] part no.	ID [11:1] vendor ID no.	ID [0] fix bit
μPD48288209AF1	32M x 9	0100	0001 0000 1010 0111	00000010000	1
μPD48288218AF1	16M x 18	0101	0001 0000 1010 0111	00000010000	1
μPD48288236AF1	8M x 36	0110	0001 0000 1010 0111	00000010000	1

Table 3-7. SCAN Exit Order

Bit	s	ignal nam	ne	Bump	Bit	s	ignal nam	ne	Bump	Bit	s	ignal nam	ne	Bump
no.	х9	x18	x36	ID	no.	х9	x18	x36	ID	no.	х9	x18	x36	ID
1	DK	DK	DK1	K1	39	DNU	DNU	DQ30	R11	77	DNU	DNU	DQ2	C11
2	DK#	DK#	DK1#	K2	40	DNU	DNU	DQ30	R11	78	DNU	DNU	DQ2	C11
3	CS#	CS#	CS#	L2	41	DNU	DNU	DQ32	P11	79	DQ1	DQ1	DQ3	C10
4	REF#	REF#	REF#	L1	42	DNU	DNU	DQ32	P11	80	DQ1	DQ1	DQ3	C10
5	WE#	WE#	WE#	M1	43	DQ5	DQ10	DQ33	P10	81	DNU	DNU	DQ0	B11
6	A17	A17	A17	М3	44	DQ5	DQ10	DQ33	P10	82	DNU	DNU	DQ0	B11
7	A16	A16	A16	M2	45	DNU	DNU	DQ34	N11	83	DQ0	DQ0	DQ1	B10
8	A18	A18	A18	N1	46	DNU	DNU	DQ34	N11	84	DQ0	DQ0	DQ1	B10
9	A15	A15	A15	P1	47	DQ4	DQ9	DQ35	N10	85	DNU	DQ4	DQ9	В3
10	DNU	DQ14	DQ25	N3	48	DQ4	DQ9	DQ35	N10	86	DNU	DQ4	DQ9	В3
11	DNU	DQ14	DQ25	N3	49	DM	DM	DM	P12	87	DNU	DNU	DQ8	B2
12	DNU	DNU	DQ24	N2	50	A19	A19	(A19)	N12	88	DNU	DNU	DQ8	B2
13	DNU	DNU	DQ24	N2	51	A11	A11	A11	M11	89	DNU	DQ5	DQ11	С3
14	DNU	DQ15	DQ23	P3	52	A12	A12	A12	M10	90	DNU	DQ5	DQ11	СЗ
15	DNU	DQ15	DQ23	P3	53	A10	A10	A10	M12	91	DNU	DNU	DQ10	C2
16	DNU	DNU	DQ22	P2	54	A13	A13	A13	L12	92	DNU	DNU	DQ10	C2
17	DNU	DNU	DQ22	P2	55	A14	A14	A14	L11	93	DNU	DQ6	DQ13	D3
18	DNU	QK1	QK1	R2	56	BA1	BA1	BA1	K11	94	DNU	DQ6	DQ13	D3
19	DNU	QK1#	QK1#	R3	57	CK#	CK#	CK#	K12	95	DNU	DNU	DQ12	D2
20	DNU	DNU	DQ20	T2	58	CK	CK	CK	J12	96	DNU	DNU	DQ12	D2
21	DNU	DNU	DQ20	T2	59	BA0	BA0	BA0	J11	97	DNU	DNU	DQ14	E2
22	DNU	DQ16	DQ21	Т3	60	A4	A4	A4	H11	98	DNU	DNU	DQ14	E2
23	DNU	DQ16	DQ21	Т3	61	А3	A3	A3	H12	99	DNU	DQ7	DQ15	E3
24	DNU	DNU	DQ18	U2	62	A0	A0	A0	G12	100	DNU	DQ7	DQ15	E3
25	DNU	DNU	DQ18	U2	63	A2	A2	A2	G10	101	DNU	DNU	DQ16	F2
26	DNU	DQ17	DQ19	U3	64	A1	A1	A1	G11	102	DNU	DNU	DQ16	F2
27	DNU	DQ17	DQ19	U3	65	A20	(A20)	(A20)	E12	103	DNU	DQ8	DQ17	F3
28	ZQ	ZQ	ZQ	V2	66	QVLD	QVLD	QVLD	F12	104	DNU	DQ8	DQ17	F3
29	DQ8	DQ13	DQ27	U10	67	DQ3	DQ3	DQ7	F10	105	(A21)	(A21)	(A21)	E1
30	DQ8	DQ13	DQ27	U10	68	DQ3	DQ3	DQ7	F10	106	A5	A5	A5	F1
31	DNU	DNU	DQ26	U11	69	DNU	DNU	DQ6	F11	107	A6	A6	A6	G2
32	DNU	DNU	DQ26	U11	70	DNU	DNU	DQ6	F11	108	A7	A7	A7	G3
33	DQ7	DQ12	DQ29	T10	71	DQ2	DQ2	DQ5	E10	109	A8	A8	A8	G1
34	DQ7	DQ12	DQ29	T10	72	DQ2	DQ2	DQ5	E10	110	BA2	BA2	BA2	H1
35	DNU	DNU	DQ28	T11	73	DNU	DNU	DQ4	E11	111	A9	A9	A9	H2
36	DNU	DNU	DQ28	T11	74	DNU	DNU	DQ4	E11	112	NF	NF	DK0#	J2
37	DQ6	DQ11	DQ31	R10	75	QK0	QK0	QK0	D11	113	NF	NF	DK0	J1
38	DQ6	DQ11	DQ31	R10	76	QK0#	QK0#	QK0#	D10					

Note Any unused balls that are in the order will read as a logic "0".

JTAG Instructions

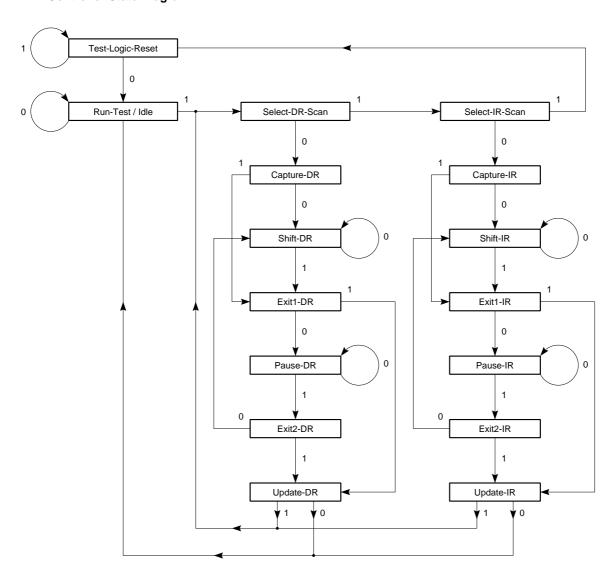
Many different instructions (2^8) are possible with the 8-bit instruction register. All used combinations are listed in **Table 3-8**, Instruction Codes. These six instructions are described in detail below. The remaining instructions are reserved and should not be used.

The TAP controller used in this RAM is fully compliant to the 1149.1 convention. Instructions are loaded into the TAP controller during the Shift-IR state when the instruction register is placed between TDI and TDO. During this state, instructions are shifted through the instruction register through the TDI and TDO balls. To execute the instruction once it is shifted in, the TAP controller needs to be moved into the Update-IR state.

Table 3-8

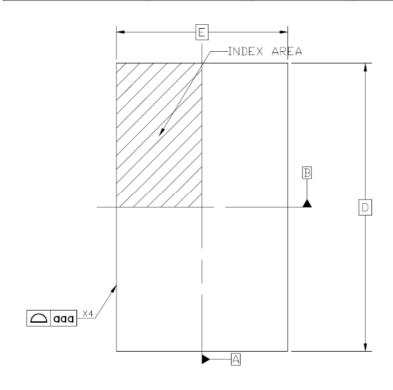
Instructions	Instruction Code [7:0]	Description
EXTEST	0000 0000	The EXTEST instruction allows circuitry external to the component package to be tested. Boundary-scan register cells at output pins are used to apply test vectors, while those at input pins capture test results. Typically, the first test vector to be applied using the EXTEST instruction will be shifted into the boundary scan register using the PRELOAD instruction. Thus, during the update-IR state of EXTEST, the output drive is turned on and the PRELOAD data is driven onto the output pins.
IDCODE	0010 0001	The IDCODE instruction causes the ID ROM to be loaded into the ID register when the controller is in capture-DR mode and places the ID register between the TDI and TDO pins in shift-DR mode. The IDCODE instruction is the default instruction loaded in at power up and any time the controller is placed in the test-logic-reset state.
SAMPLE / PRELOAD	0000 0101	SAMPLE / PRELOAD is a Standard 1149.1 mandatory public instruction. When the SAMPLE / PRELOAD instruction is loaded in the instruction register, moving the TAP controller into the capture-DR state loads the data in the RAMs input and DQ pins into the boundary scan register. Because the RAM clock(s) are independent from the TAP clock (TCK) it is possible for the TAP to attempt to capture the I/O ring contents while the input buffers are in transition (i.e., in a metastable state). Although allowing the TAP to sample metastable input will not harm the device, repeatable results cannot be expected. RAM input signals must be stabilized for long enough to meet the TAPs input data capture setup plus hold time (tcs plus tch). The RAMs clock inputs need not be paused for any other TAP operation except capturing the I/O ring contents into the boundary scan register. Moving the controller to shift-DR state then places the boundary scan register between the TDI and TDO pins.
CLAMP	0000 0111	When the CLAMP instruction is loaded into the instruction register, the data driven by the output balls are determined from the values held in the boundary scan register. Selects the bypass register to be connected between TDI and TDO. Data driven by output balls are determined from values held in the boundary scan register.
High-Z	0000 0011	The High-z instruction causes the boundary scan register to be connected between the TDI and TDO. This places all RAMs outputs into a High-Z state. Selects the bypass register to be connected between TDI and TDO. All outputs are forced into high impedance state.
BYPASS	1111 1111	When the BYPASS instruction is loaded in the instruction register, the bypass register is placed between TDI and TDO. This occurs when the TAP controller is moved to the shift-DR state. This allows the board level scan path to be shortened to facilitate testing of other devices in the scan path.
Reserved for Future Use	-	The remaining instructions are not implemented but are reserved for future use. Do not use these instructions.

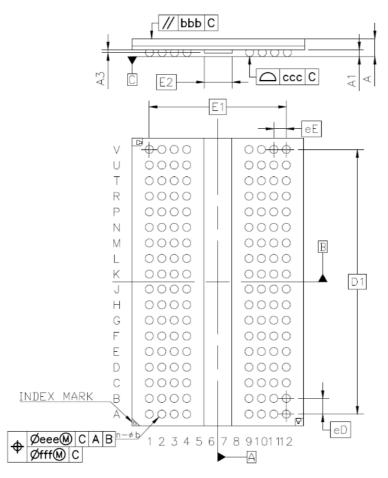
TAP Controller State Diagram



4. Package Dimensions

JEITA Package code	RENESAS code	Previous code	MASS(TYP.)[g]
P-TFBGA-11x18.5-0.80	PTBG0144GC-A	-	0.39





Reference	Dimension in Millimeters				
Symbol	Min.	Nom.	Max.		
D	18.40	18.50	18.60		
E	10.90	11.00	11.10		
D1	-	17.00	_		
E1))—(8.80	_		
Α	(1 -1)	3 - -	1.20		
A1	0.37	0.42	0.47 0.18 0.57		
А3	11-1	-			
b	0.47	0.52			
eD	7-	1.00	_		
еE	19—1	0.80	8-8		
aaa	-	-	0.10		
bbb	(* - *)	-	0.20		
ссс	(()	-	0.10		
eee	-	-	0.15		
fff	51 -4 5	_	0.08		
n	()):	144	-		
E2	1.75	1.80	1.85		

Top Marking Information

Due to space limitation of the package, top marking is different from part number. About marking code of I/O, configuration, speed grade and temperature range, refer to the Part Number Definition (see Page 2).



(1) D48288 : 288Mbit LLDRAM XXX : I/O, Configuration Code

AF1 : Fixed

(2) E : Fixed

XX : Speed Grade, Temp. Range Code

-DW1-A : Fixed

(3) YY : Year Code (last 2-digits)

WW : Week Code XXXXXX : Serial Code

(4) Laser dot : Lead-free identification dot TAIWAN : (Assembly) country of origin

(5) Laser dot : INDEX Mark

5. Recommended Soldering Condition

Please consult with our sales offices for soldering conditions of these products.

Types of Surface Mount Devices

 μ PD48288209AF1-DW1 : 144-pin FBGA (18.5 x 11) μ PD48288218AF1-DW1 : 144-pin FBGA (18.5 x 11) μ PD48288236AF1-DW1 : 144-pin FBGA (18.5 x 11)

Quality Grade

- A quality grade of the products is "Standard".
- Anti-radioactive design is not implemented in the products.
- Semiconductor devices have the possibility of unexpected defects by affection of cosmic ray that reach to the ground and so forth.

Revision History

μ PD48288209AF1, μ PD48288218AF1, μ PD48288236AF1

Rev.	Date	Description	
		Page	Summary
Rev. 1.00	'15.07.01	-	New Data Sheet
Rev. 1.01	'16.01.15	P. 2	Updated Ordering Information
			Added part number definition
		P. 50	Updated package bottom view
		P. 51	Added top marking information

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