

MAX11209/MAX11211

18-Bit, Single-Channel, Ultra-Low-Power, Delta-Sigma ADCs with Programmable Gain and GPIO

General Description

The MAX11209/MAX11211 are ultra-low-power (< 300µA active current), high-resolution, serial-output ADCs. These devices provide the highest resolution per unit power in the industry, and are optimized for applications that require very high dynamic range with low power, such as sensors on a 4mA to 20mA industrial control loop. Optional input buffers provide isolation of the signal inputs from the switched capacitor sampling network allowing these converters to be used with high-impedance sources without compromising available dynamic range or linearity. The devices provide a high-accuracy internal oscillator that requires no external components. When used with the specified data rates, the internal digital filter provides more than 100dB rejection of 50Hz or 60Hz line noise. The devices are configurable using the SPI™ interface and include four GPIOs that can be used for external mux control. The MAX11209 includes digital programmable gain of 1 to 128.

The MAX11209/MAX11211 operate over the -40°C to +85°C temperature range, and are available in a 16-pin QSOP package.

Applications

Sensor Measurement (Temperature and Pressure)
 Portable Instrumentation
 Battery Applications
 Weigh Scales

SPI and QSPI are trademarks of Motorola, Inc.

MICROWIRE is a trademark of National Semiconductor Corp.

Features

- ◆ 18-Bit Noise-Free Resolution
- ◆ 570nVRMS Noise at 10sps, ±3.6VFS Input
- ◆ 3ppm INL (typ), 15ppm (max)
- ◆ No Missing Codes
- ◆ Ultra-Low Power Dissipation
 Operating-Mode Current Drain < 300µA (max)
 Sleep-Mode Current Drain < 0.4µA
- ◆ Programmable Gain (1 to 128) (MAX11209)
- ◆ Four SPI-Controlled GPIOs for External Mux Control
- ◆ 2.7V to 3.6V Analog Supply Voltage Range
- ◆ 1.7V to 3.6V Digital and I/O Supply Voltage Range
- ◆ Fully Differential Signal and Reference Inputs
- ◆ High-Impedance Inputs
 Optional Input Buffers on Both Signal and Reference Inputs
- ◆ > 100dB (min) 50Hz/60Hz Rejection
- ◆ SPI-, QSPI™-, MICROWIRE™-Compatible Serial Interface
- ◆ On-Demand Offset and Gain Self-Calibration and System Calibration
- ◆ User-Programmable Offset and Gain Registers
- ◆ -40°C to +85°C Operating Temperature Range
- ◆ ±2kV ESD Protection
- ◆ Lead(Pb)-Free and RoHS-Compliant QSOP Package

Ordering Information

| PART | TEMP RANGE | PIN-PACKAGE |
|--------------|----------------|-------------|
| MAX11209EEE+ | -40°C to +85°C | 16 QSOP |
| MAX11211EEE+ | -40°C to +85°C | 16 QSOP |

+Denotes a lead(Pb)-free/RoHS-compliant package.

Selector Guide

| RESOLUTION (BITS) | 4-WIRE SPI, 16-PIN QSOP, PROGRAMMABLE GAIN | 4-WIRE SPI, 16-PIN QSOP | 2-WIRE SERIAL, 10-PIN µMAX |
|-------------------|--|-------------------------|---|
| 24 | MAX11210 | MAX11200 | MAX11201 (with buffers) MAX11202 (without buffers) |
| 20 | MAX11206 | MAX11207 | MAX11208 |
| 18 | MAX11209 | MAX11211 | MAX11212 |
| 16 | MAX11213 | MAX11203 | MAX11205 |

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maximintegrated.com.

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ABSOLUTE MAXIMUM RATINGS

| | | | |
|--|-------------------------------------|---|-----------------|
| Any Pin to GND | -0.3V to +3.9V | Continuous Power Dissipation (T _A = +70°C) | |
| AVDD to GND | -0.3V to +3.9V | 16-Pin QSOP (derate 8.3mW/°C above +70°C)..... | 667mW |
| DVDD to GND | -0.3V to +3.9V | Operating Temperature Range | -40°C to +85°C |
| Analog Inputs (AINP, AINN, REFP, REFN) | | Junction Temperature | +150°C |
| to GND | -0.3V to (V _{AVDD} + 0.3V) | Storage Temperature Range..... | -55°C to +150°C |
| Digital Inputs and Digital Outputs | | Lead Temperature (soldering, 10s) | +300°C |
| to GND | -0.3V to (V _{DVDD} + 0.3V) | Soldering Temperature (reflow) | +260°C |
| ESD _{HB} (AVDD, AINP, AINN, REFP, REFN, DVDD, CLK, \overline{CS} , SCLK, DIN, $\overline{RDY/DOUT}$, GND, GPIO_) | ±2kV (Note 1) | | |

Note 1: Human Body Model to specification MIL-STD-883 Method 3015.7.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{AVDD} = +3.6V, V_{DVDD} = +1.7V, V_{REFP} - V_{REFN} = V_{AVDD}; internal clock, single-cycle mode (SCYCLE = 1), T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C under normal conditions, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|---------------------------------------|-------------------|--|------------------|------|-------------------|-------------------|
| STATIC PERFORMANCE | | | | | | |
| Noise-Free Resolution (Notes 2, 3) | NFR | 120sps | | 18 | | Bits |
| | | 10sps | | 18 | | |
| Noise (Notes 2, 3) | V _N | 120sps | | 2.1 | | μV _{RMS} |
| | | 10sps | | 0.55 | | |
| Integral Nonlinearity | INL | At 10sps (Note 4) | -15 | | +15 | ppmFSR |
| Zero Error | | After self and system calibration, V _{REFP} - V _{REFN} = 2.5V | -15 | | +15 | ppmFSR |
| Zero Drift | | | | 50 | | nV/°C |
| Full-Scale Error | | After self and system calibration, V _{REFP} - V _{REFN} = 2.5V (Note 5) | -20 | | +20 | ppmFSR |
| Full-Scale Error Drift | | | | 0.05 | | ppmFSR/°C |
| Power-Supply Rejection | | AVDD DC rejection | 70 | 80 | | dB |
| | | DVDD DC rejection | 90 | 100 | | |
| ANALOG INPUTS/REFERENCE INPUTS | | | | | | |
| Common-Mode Rejection | CMR | DC rejection | 90 | 123 | | dB |
| | | 50Hz/60Hz rejection at 120sps | 90 | | | |
| | | 50Hz/60Hz rejection at 1sps to 15sps | 144 | | | |
| Normal-Mode 50Hz Rejection | NMR ₅₀ | LINEF = 1, for 1sps to 15sps (Notes 6, 7) | 100 | 144 | | dB |
| Normal-Mode 60Hz Rejection | NMR ₆₀ | LINEF = 0, for 1sps to 15sps (Notes 6, 7) | 100 | 144 | | dB |
| Common-Mode Voltage Range | | AIN buffers disabled | V _{GND} | | V _{AVDD} | V |

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ELECTRICAL CHARACTERISTICS (continued)

($V_{AVDD} = +3.6V$, $V_{DVDD} = +1.7V$, $V_{REFP} - V_{REFN} = V_{AVDD}$; internal clock, single-cycle mode (SCYCLE = 1), $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^\circ C$ under normal conditions, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|--|-----------|--|-----------------------|--------------------|-----------------------|-----------|
| Absolute Input Voltage | | Low input voltage | Buffers disabled | $V_{GND} - 30mV$ | | V |
| | | | Buffers enabled | $V_{GND} + 100mV$ | | |
| | | High input voltage | Buffers disabled | $V_{AVDD} + 30mV$ | | |
| | | | Buffers enabled | $V_{AVDD} - 100mV$ | | |
| DC Input Leakage | | Sleep mode | | ± 1 | | μA |
| AIN Dynamic Input Current | | Buffer disabled | | ± 1.4 | | $\mu A/V$ |
| | | Buffer enabled | | ± 20 | | nA |
| REF Dynamic Input Current | | Buffer disabled | | ± 2.1 | | $\mu A/V$ |
| | | Buffer enabled | | ± 30 | | nA |
| AIN Input Capacitance | | Buffer disabled | | 5 | | pF |
| REF Input Capacitance | | Buffer disabled | | 7.5 | | pF |
| AIN Voltage Range | | Unipolar | 0 | | V_{REF} | V |
| | | Bipolar | $-V_{REF}$ | | $+V_{REF}$ | |
| Input Sampling Rate | fs | LINEF = 0 | | 246 | | kHz |
| | | LINEF = 1 | | 204.8 | | |
| REF Voltage Range | | Buffers disabled | 0 | | V_{AVDD} | V |
| | | Buffers enabled | 0.1 | | $V_{AVDD} - 0.1$ | |
| REF Sampling Rate | | LINEF = 0 | | 246 | | kHz |
| | | LINEF = 1 | | 204.8 | | |
| LOGIC INPUTS (SCLK, CLK, DIN, GPIO1–GPIO4) | | | | | | |
| Input Current | | Input leakage current | | ± 1 | | μA |
| Input Low Voltage | V_{IL} | | | | $0.3 \times V_{DVDD}$ | V |
| Input High Voltage | V_{IH} | | $0.7 \times V_{DVDD}$ | | | V |
| Input Hysteresis | V_{HYS} | | | 200 | | mV |
| External Clock | | 60Hz line frequency | | 2.4576 | | MHz |
| | | 55Hz line frequency | | 2.25275 | | |
| | | 50Hz line frequency | | 2.048 | | |
| LOGIC OUTPUTS ($\overline{RDY}/DOUT$, GPIO1–GPIO4) | | | | | | |
| Output Low Level | V_{OL} | $I_{OL} = 1mA$; also tested for $V_{DVDD} = 3.6V$ | | | 0.4 | V |
| Output High Level | V_{OH} | $I_{OH} = 1mA$; also tested for $V_{DVDD} = 3.6V$ | $0.9 \times V_{DVDD}$ | | | V |
| Leakage Current | | High-impedance state | | ± 500 | | nA |
| Output Capacitance | | High-impedance state | | 9 | | pF |

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ELECTRICAL CHARACTERISTICS (continued)

($V_{AVDD} = +3.6V$, $V_{DVDD} = +1.7V$, $V_{REFP} - V_{REFN} = V_{AVDD}$; internal clock, single-cycle mode (SCYCLE = 1), $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^\circ C$ under normal conditions, unless otherwise noted.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|------------|---|-----|------|-----------------|---------|
| POWER REQUIREMENTS | | | | | | |
| Analog Supply | V_{AVDD} | | 2.7 | | 3.6 | V |
| Digital Supply | V_{DVDD} | | 1.7 | | 3.6 | V |
| Total Operating Current | | AVDD + DVDD | | 235 | 300 | μA |
| | | | | 255 | | |
| AVDD Sleep Current | | | | 0.15 | 2 | μA |
| AVDD Operating Current | | Buffers disabled | | 185 | 235 | μA |
| | | Buffers enabled | | 205 | | |
| DVDD Sleep Current | | | | 0.25 | 2 | μA |
| DVDD Operating Current | | | | 50 | 65 | μA |
| SPI TIMING CHARACTERISTICS | | | | | | |
| SCLK Frequency | fSCLK | | | | 5 | MHz |
| SCLK Clock Period | tCP | | 200 | | | ns |
| SCLK Pulse-Width High | tCH | | 80 | | | ns |
| SCLK Pulse-Width Low | tCL | 60% duty cycle at 5MHz | 80 | | | ns |
| \overline{CS} Low to 1st SCLK Rise Setup | tCSS0 | | 40 | | | ns |
| \overline{CS} High to 17th SCLK Setup | tCSS1 | | 40 | | | ns |
| \overline{CS} High After 16th SCLK Falling Edge Hold | tCSH1 | | 3 | | | ns |
| \overline{CS} Pulse-Width High | tCSW | | 40 | | | ns |
| DIN to SCLK Setup | tDS | | 40 | | | ns |
| DIN Hold After SCLK | tDH | | 0 | | | ns |
| $\overline{RDY}/\overline{DOUT}$ Transition Valid After SCLK Fall | tDOT | Output transition time, data changes on falling edge of SCLK | | | 40 | ns |
| $\overline{RDY}/\overline{DOUT}$ Remains Valid After SCLK Fall | tDOH | Output hold time allows for negative edge data read | 3 | | | ns |
| $\overline{RDY}/\overline{DOUT}$ Valid Before SCLK Rise | tDOL | $t_{DOL} = t_{CL} - t_{DOT}$ | 40 | | | ns |
| \overline{CS} Rise to $\overline{RDY}/\overline{DOUT}$ Disable | tDOD | $C_{LOAD} = 20pF$ | | | 25 | ns |
| \overline{CS} Fall to $\overline{RDY}/\overline{DOUT}$ Valid | tDOE | Default value of \overline{RDY} is 1 for minimum specification; maximum specification for valid 0 on $\overline{RDY}/\overline{DOUT}$ | 0 | | 40 | ns |
| DATA Fetch | tDF | Maximum time after \overline{RDY} asserts to read DATA register; tCNV is the time for one conversion | 0 | | tCNV - 60 x tCP | |

Note 2: These specifications are not fully tested and are guaranteed by design and/or characterization.

Note 3: $V_{AINP} = V_{AINN}$.

Note 4: ppmFSR is parts per million of full scale.

Note 5: Positive full-scale error includes zero-scale errors (unipolar offset error or bipolar zero error) and applies to both unipolar and bipolar input ranges.

Note 6: For data rates (1, 2.5, 5, 10, 15)sps and (0.83, 2.08, 4.17, 8.33, 12.5)sps.

Note 7: Normal-mode rejection of power line frequencies of 60Hz/50Hz apply only for single-cycle data rates at 15sps/10sps and lower or continuous data rate of 60sps/50sps.

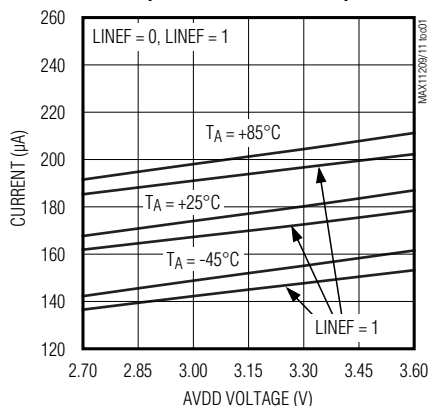
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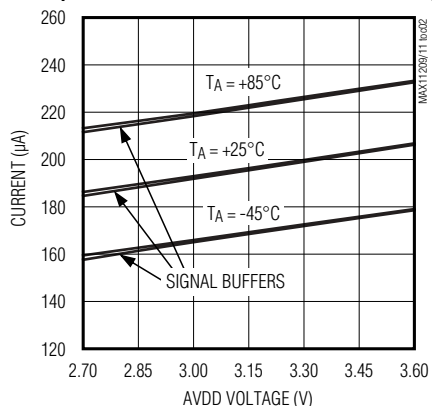
Typical Operating Characteristics

($V_{AVDD} = 3.6V$, $V_{DVDD} = 1.8V$, $V_{REFP} - V_{REFN} = 2.5V$; internal clock; $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^\circ C$.)

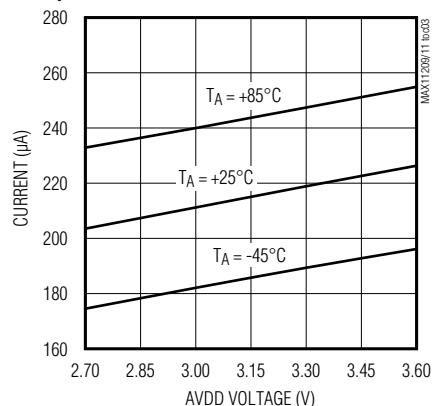
ANALOG ACTIVE CURRENT vs. AVDD VOLTAGE (NO BUFFERS ENABLED)



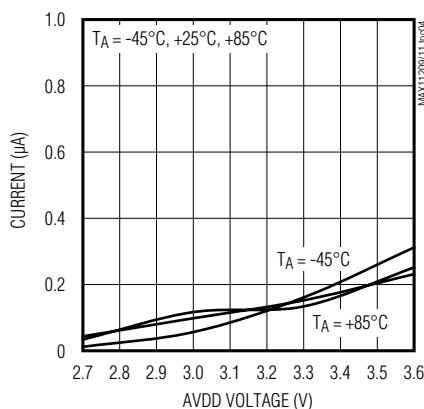
ANALOG ACTIVE CURRENT vs. AVDD VOLTAGE (SIGNAL OR REFERENCE BUFFERS ENABLED)



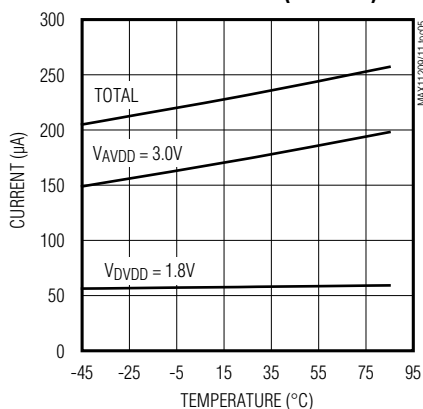
ANALOG ACTIVE CURRENT vs. AVDD VOLTAGE (SIGNAL AND REFERENCE BUFFERS ENABLED)



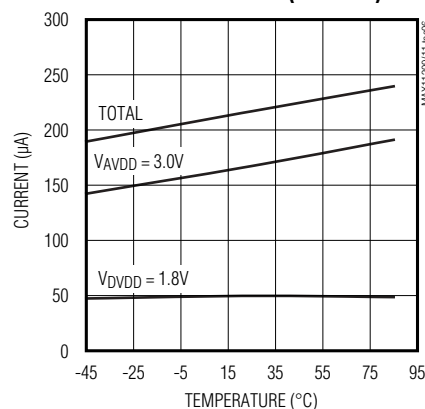
ANALOG SLEEP CURRENT vs. AVDD VOLTAGE



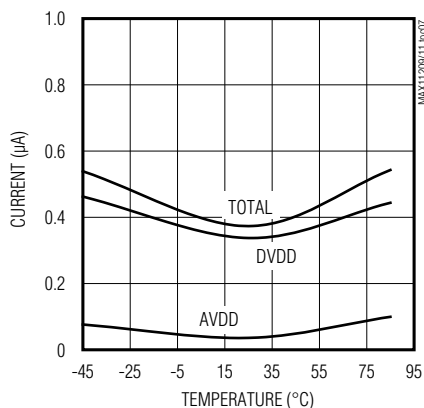
ACTIVE SUPPLY CURRENT vs. TEMPERATURE (LINEF = 0)



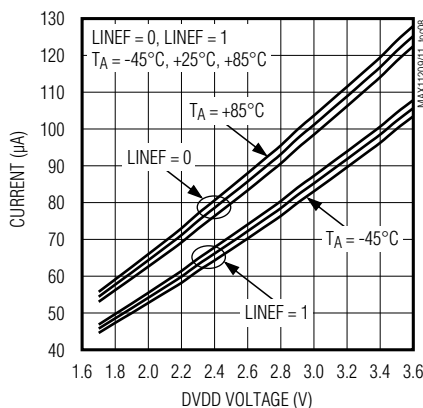
ACTIVE SUPPLY CURRENT vs. TEMPERATURE (LINEF = 1)



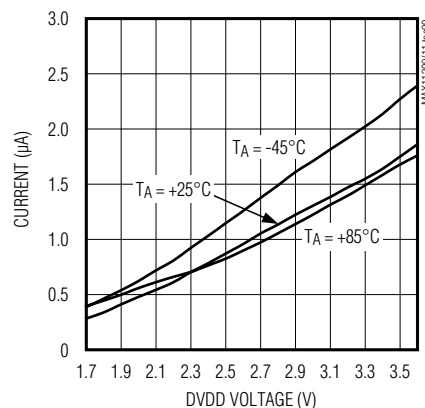
SLEEP CURRENT vs. TEMPERATURE



DIGITAL ACTIVE CURRENT vs. DVDD VOLTAGE



DIGITAL SLEEP CURRENT vs. DVDD VOLTAGE



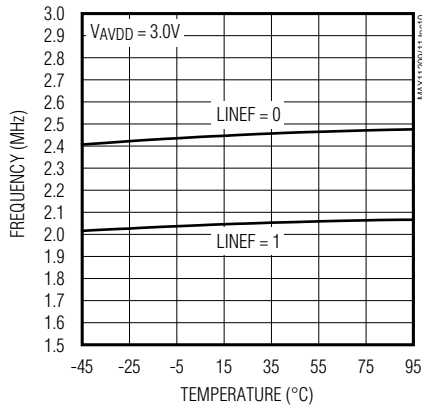
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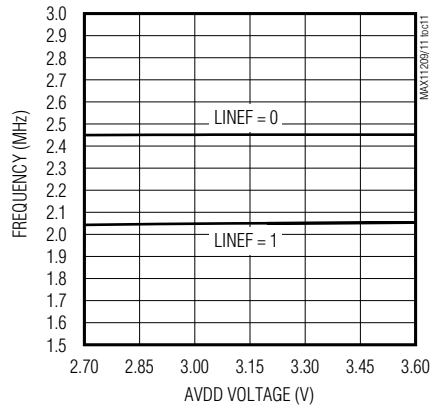
Typical Operating Characteristics (continued)

($V_{AVDD} = 3.6V$, $V_{DVDD} = 1.8V$, $V_{REFP} - V_{REFN} = 2.5V$; internal clock; $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^\circ C$.)

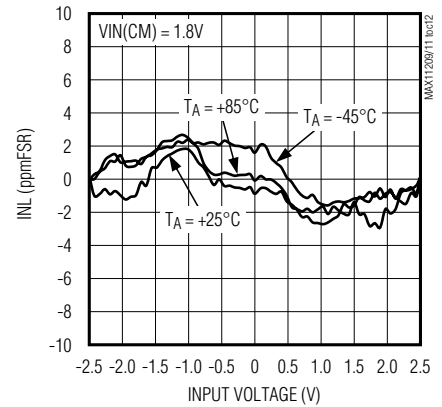
INTERNAL OSCILLATOR FREQUENCY vs. TEMPERATURE



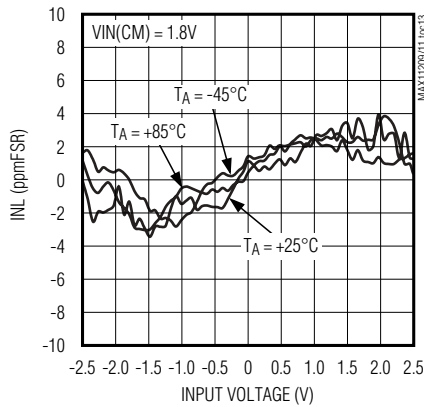
INTERNAL OSCILLATOR FREQUENCY vs. AVDD VOLTAGE



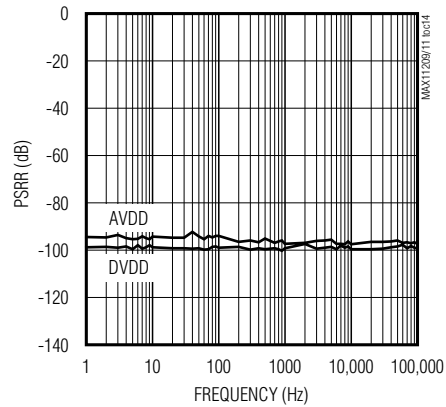
INTEGRAL NONLINEARITY vs. INPUT VOLTAGE



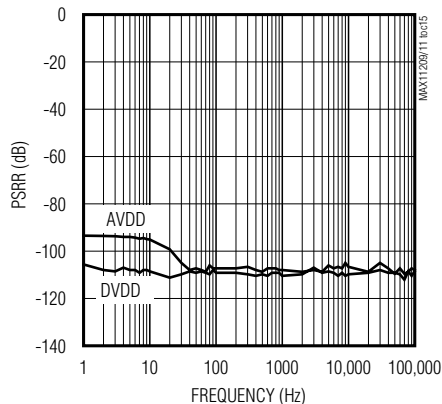
TUE vs. INPUT VOLTAGE



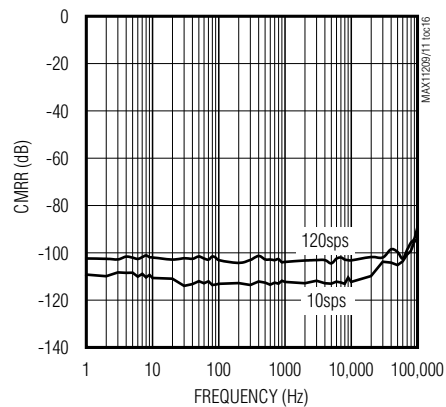
PSRR vs. FREQUENCY (DATA RATE 120sps)



PSRR vs. FREQUENCY (DATA RATE 10sps)



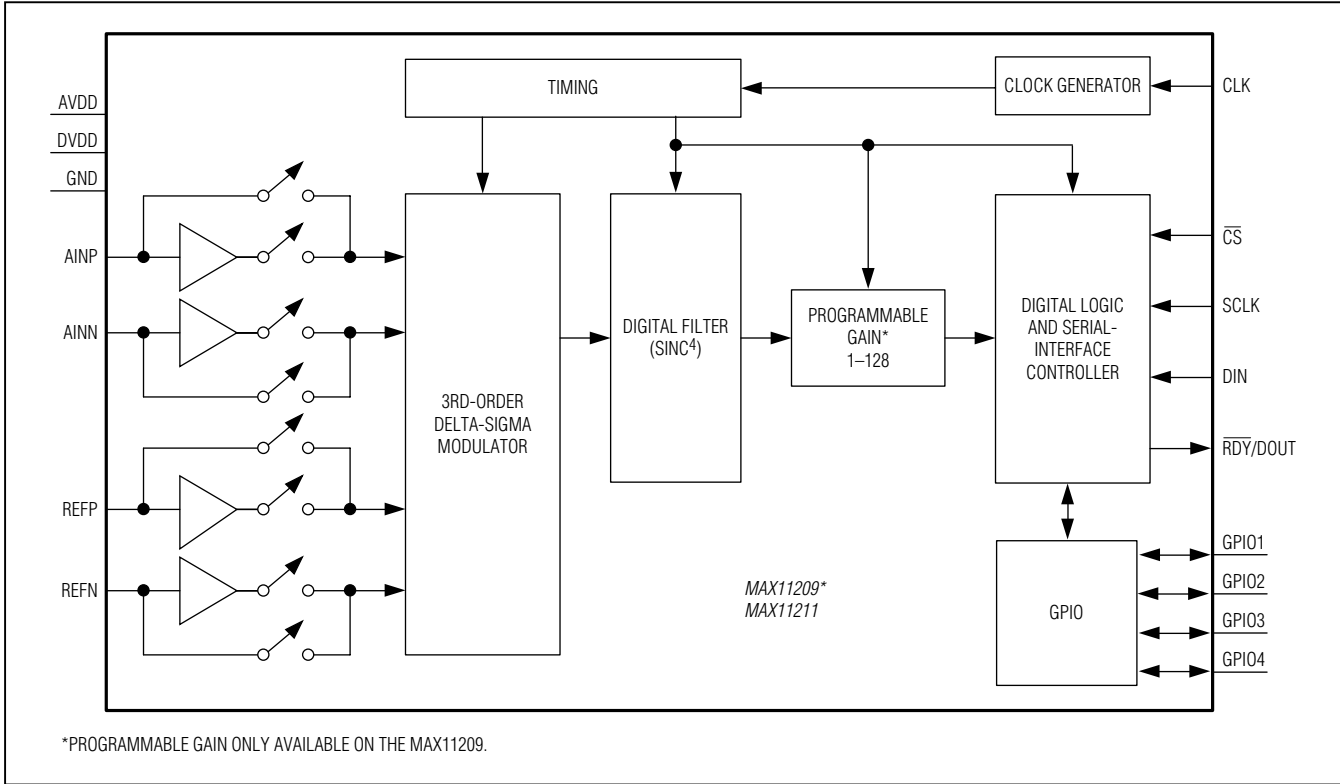
CMRR vs. FREQUENCY



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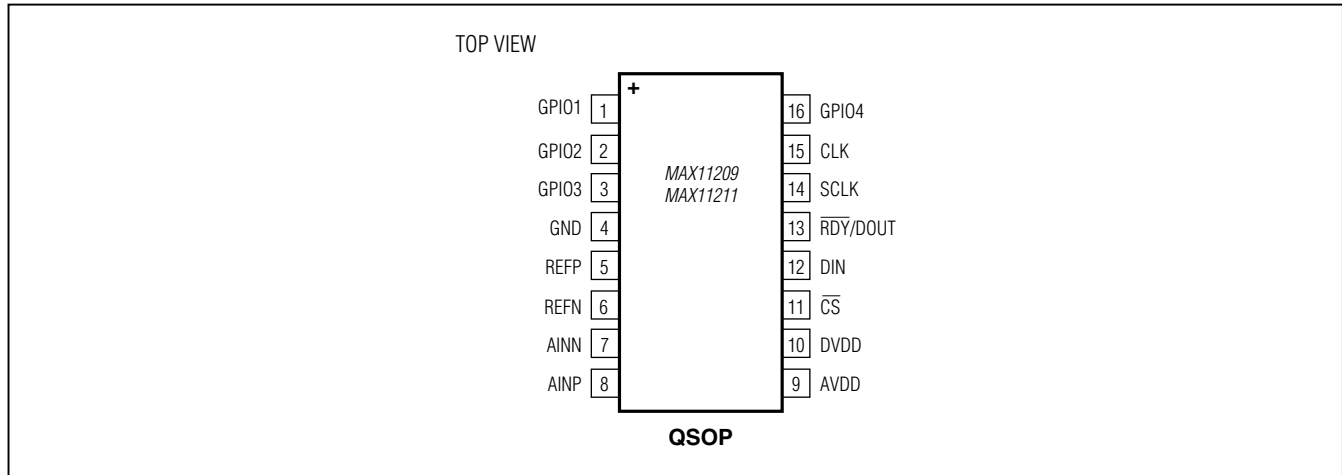
Functional Diagram



MAX11209/MAX11211

18-Bit, Single-Channel, Ultra-Low-Power, Delta-Sigma ADCs with Programmable Gain and GPIO

Pin Configuration



Pin Description

| PIN | NAME | FUNCTION |
|-----|-------------------------------------|---|
| 1 | GPIO1 | General-Purpose I/O 1. Register controllable using SPI. |
| 2 | GPIO2 | General-Purpose I/O 2. Register controllable using SPI. |
| 3 | GPIO3 | General-Purpose I/O 3. Register controllable using SPI. |
| 4 | GND | Ground. Ground reference for analog and digital circuitry. |
| 5 | REFP | Differential Reference Positive Input. REFP must be more positive than REFN. Connect REFP to a voltage between AVDD and GND. |
| 6 | REFN | Differential Reference Negative Input. REFN must be more negative than REFP. Connect REFN to a voltage between AVDD and GND. |
| 7 | AINN | Negative Fully Differential Analog Input |
| 8 | AINP | Positive Fully Differential Analog Input |
| 9 | AVDD | Analog Supply Voltage. Connect a supply voltage between +2.7V and +3.6V with respect to GND. |
| 10 | DVDD | Digital Supply Voltage. Connect a digital supply voltage between +1.7V and +3.6V with respect to GND. |
| 11 | $\overline{\text{CS}}$ | Active-Low, Chip-Select Logic Input |
| 12 | DIN | Serial-Data Input. Data present at DIN is shifted to the device's internal registers at the rising edge of the serial clock at SCLK, when the device is accessed for an internal register write or for a command operation. |
| 13 | $\overline{\text{RDY}}/\text{DOUT}$ | Data Ready Output/Serial-Data Output. This output serves a dual function. In addition to the serial-data output function, the $\overline{\text{RDY}}/\text{DOUT}$ also indicates that the data is ready when the $\overline{\text{RDY}}$ is logic-low. $\overline{\text{RDY}}/\text{DOUT}$ changes on the falling edge of SCLK. |
| 14 | SCLK | Serial-Clock Input. Apply an external serial clock to SCLK. |
| 15 | CLK | External Clock Signal Input. When external clock mode is selected (EXTCLK = 1), provide a 2.4576MHz or 2.048MHz clock signal at CLK. Other frequencies can be used, but the data rate and digital filter notch frequencies scale accordingly. |
| 16 | GPIO4 | General-Purpose I/O 4. Register controllable using SPI. |

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Detailed Description

The MAX11209/MAX11211 are ultra-low-power (< 300µA active), high-resolution, low-speed, serial-output ADCs. These ADCs provide the highest resolution per unit power in the industry, and are optimized for applications that require very high dynamic range with low power such as sensors on a 4mA to 20mA industrial control loop. Optional input buffers provide isolation of the signal inputs from the switched capacitor sampling network, allowing the devices to be used with very high impedance sources without compromising available dynamic range. The devices provide a high-accuracy internal oscillator, which requires no external components. When used with the specified data rates, the internal digital filter provides more than 144dB rejection of 50Hz or 60Hz line noise. The devices are highly configurable using the SPI interface and include four GPIOs for external mux control.

Analog Inputs

The devices accept two analog inputs (AINP, AINN) in buffered or unbuffered mode. The input buffer isolates

the inputs from the capacitive load presented by the modulator, allowing for high source-impedance analog transducers. The value of the SIGBUF bit in the CTRL1 register determines whether the input buffer is enabled or disabled. See Table 12.

Input Voltage Range

The modulator input range is programmable for bipolar (-VREF to +VREF) or unipolar (0 to VREF) ranges. The U/B bit in the CTRL1 register configures the MAX11209/MAX11211 for unipolar or bipolar transfer functions. See Table 12.

System Clock

The devices incorporate a highly stable internal oscillator that provides the system clock. The system clock runs the internal state machine and is trimmed to 2.4576MHz or 2.048MHz. The internal oscillator clock is divided down to run the digital and analog timing. The LINEF bit in the CTRL1 register determines the internal oscillator frequency. See Tables 10 and 12. Set LINEF = 0 to select the 2.4576MHz oscillator and LINEF = 1 to select the

Table 1. Continuous Conversion with SCYCLE Bit = 0

| RATE[2:0] | DATA RATE* (sps) | | BIPOLAR NFR (BITS) | BIPOLAR ENOB (BITS) | UNIPOLAR NFR (BITS) | UNIPOLAR ENOB (BITS) | OUTPUT NOISE (µVRMS) |
|-----------|------------------|-----------|--------------------|---------------------|---------------------|----------------------|----------------------|
| | LINEF = 0 | LINEF = 1 | | | | | |
| 100 | 60 | 50 | 18.0 | 18.0 | 18.0 | 18.0 | 0.74 |
| 101 | 120 | 100 | 18.0 | 18.0 | 18.0 | 18.0 | 1.03 |
| 110 | 240 | 200 | 18.0 | 18.0 | 18.0 | 18.0 | 1.45 |
| 111 | 480 | 400 | 18.0 | 18.0 | 18.0 | 18.0 | 2.21 |

*LINEF = 0 sets the clock frequency to 2.4576MHz and the input sampling frequency to 245.76kHz. LINEF bit = 1 sets the clock frequency to 2.048MHz and the input sampling frequency to 204.8kHz.

Table 2. Single-Cycle Conversion with SCYCLE Bit = 1

| RATE[2:0] | SINGLE-CYCLE DATA RATE* (sps) | | BIPOLAR NFR (BITS) | BIPOLAR ENOB (BITS) | UNIPOLAR NFR (BITS) | UNIPOLAR ENOB (BITS) | OUTPUT NOISE (µVRMS) |
|-----------|-------------------------------|-----------|--------------------|---------------------|---------------------|----------------------|----------------------|
| | LINEF = 0 | LINEF = 1 | | | | | |
| 000 | 1 | 0.833 | 18.0 | 18.0 | 18.0 | 18.0 | 0.21 |
| 001 | 2.5 | 2.08 | 18.0 | 18.0 | 18.0 | 18.0 | 0.27 |
| 010 | 5 | 4.17 | 18.0 | 18.0 | 18.0 | 18.0 | 0.39 |
| 011 | 10 | 8.33 | 18.0 | 18.0 | 18.0 | 18.0 | 0.55 |
| 100 | 15 | 12.5 | 18.0 | 18.0 | 18.0 | 18.0 | 0.74 |
| 101 | 30 | 25 | 18.0 | 18.0 | 18.0 | 18.0 | 1.03 |
| 110 | 60 | 50 | 18.0 | 18.0 | 18.0 | 18.0 | 1.45 |
| 111 | 120 | 100 | 18.0 | 18.0 | 18.0 | 18.0 | 2.21 |

*LINEF = 0 sets the clock frequency to 2.4576MHz and the input sampling frequency to 245.76kHz. LINEF bit = 1 sets the clock frequency to 2.048MHz and the input sampling frequency to 204.8kHz.

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2.048MHz oscillator. The 2.4576MHz oscillator provides maximum 60Hz rejection, and the 2.048MHz oscillator provides maximum 50Hz rejection. See Figures 1 and 2. For optimal simultaneous 50Hz and 60Hz rejection, apply a 2.25275MHz external clock at CLK.

Reference

The devices provide differential inputs REFP and REFN for an external reference voltage. Connect the external reference directly across the REFP and REFN to obtain the differential reference voltage. The common-mode voltage range for VREFP and VREFN is between 0 and VAVDD.

The devices accept reference inputs in buffered or unbuffered mode. The value of the REFBUF bit in the CTRL1 register determines whether the reference buffer is enabled or disabled. See Table 12.

Buffers

The devices include reference and signal input buffers capable of reducing the average input current from 2.1 μ A/V on the reference inputs and from 1.4 μ A/V on the analog inputs to a constant 30nA current on the reference inputs and 20nA current on the analog inputs. The reference and signal input buffers can be selected individually by programming the CTRL1 register bits REFBUF and SIGBUF. When enabled, the reference and input signal buffers require an additional 20 μ A from the AVDD supply pin.

Power-On Reset (POR)

The devices utilize power-on reset (POR) supply-monitoring circuitry on both the digital supply (DVDD) and the analog supply (AVDD). The POR circuitry ensures proper device default conditions after either a digital or analog power sequencing event. The digital POR trigger threshold is approximately 1.2V and has 100mV of hysteresis. The analog POR trigger threshold is approximately 1.25V and has 100mV of hysteresis. Both POR circuits have lowpass filters that prevent high-frequency supply glitches from triggering the POR.

Calibration

The devices provide two sets of calibration registers which offer the user several options for calibrating their system. The calibration register value defaults are all zero, which require a user to either perform a calibration or program the register through the SPI interface to use them. The on-chip calibration registers are enabled or disabled by programming the NOSYSG, NOSYSO, NOSCG, and NOSCO bits in the

CTRL3 register. The default values for these calibration control bits are 1, which disables the use of the internal calibration registers.

The devices power up with the internal calibration registers disabled, and therefore a full-scale input produces a result of 60% of the full-scale digital range. To use the full-scale digital range a calibration must be performed.

The first level of calibration is the self-calibration where the part performs the required connections to zero and full-scale internally. This level of calibration is typically sufficient for 1 μ V of offset accuracy and 2ppm of full-scale accuracy. The self-calibration routine does not include the source resistance effects from the signal source driving the input pins, which can change the offset and gain of the system.

A second level of calibration is available where the user can calibrate a system zero scale and system full scale by presenting a zero-scale signal or a full-scale signal to the input pins and initiating a system zero scale or system gain calibration command.

A third level of calibration allows for the user to write to the internal calibration registers through the SPI interface to achieve any digital offset or scaling the user requires with the following restrictions. The range of digital offset correction is $\pm V_{REF}/4$. The range of digital gain correction is from 0.5 to 1.5. The resolution of offset correction is 0.5 LSB.

The calibration operations are controlled with the CAL1 and CAL0 bits in the command byte. The user requests a self-calibration by setting the CAL1 bit to 0 and CAL0 bit to 1. A self-calibration requires 200ms to complete, and both the SCOC and SCGC registers contain the values that correct the chip output for zero scale and full scale. The user requests a system zero-scale calibration by setting the CAL1 bit to 1 and the CAL0 bit to 0 and presents a system zero-level signal to the input pins. The SOC register contains the values that correct the chip zero scale. The system zero calibration requires 100ms to complete, and the SOC register contains values that correct the chip zero scale. The user requests a system full-scale calibration by setting the CAL1 bit to 1 and the CAL0 bit to 1 and presents a system full-scale signal level to the input pins. The system full-scale calibration requires 100ms to complete, and the SGC register contains values that correct for the chip full-scale value. See Tables 3a and 3b for an example of a self-calibration sequence and a system calibration sequence.

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Table 3a. Example of Self-Calibration

| STEP | DESCRIPTION | REGISTER | | | | BIT | | | |
|------|-----------------------------------|----------|----------|----------|----------|--------|--------|-------|-------|
| | | SCOC | SCGC | SOC | SGC | NOSYSG | NOSYSO | NOSCG | NOSCO |
| 1 | Initial power-up | 0x000000 | 0x000000 | 0x000000 | 0x000000 | 1 | 1 | 1 | 1 |
| 2 | Enable self-calibration registers | 0x000000 | 0x000000 | 0x000000 | 0x000000 | 1 | 1 | 0 | 0 |
| 3 | Self-calibration, DIN = 10010000 | 0x00007E | 0xBFD345 | 0x000000 | 0x000000 | 1 | 1 | 0 | 0 |

Table 3b. Example of System Calibration

| STEP | DESCRIPTION | REGISTER | | | | BIT | | | |
|------|--|----------|----------|----------|----------|--------|--------|-------|-------|
| | | SCOC | SCGC | SOC | SGC | NOSYSG | NOSYSO | NOSCG | NOSCO |
| 1 | Initial power-up | 0x000000 | 0x000000 | 0x000000 | 0x000000 | 1 | 1 | 1 | 1 |
| 2 | Enable self-calibration registers | 0x000000 | 0x000000 | 0x000000 | 0x000000 | 1 | 1 | 0 | 0 |
| 3 | Self-calibration, DIN = 10010000 | 0x00007E | 0xBFD345 | 0x000000 | 0x000000 | 1 | 1 | 0 | 0 |
| 4 | Enable system offset register | 0x00007E | 0xBFD345 | 0x000000 | 0x000000 | 1 | 0 | 0 | 0 |
| 5 | System-calibration offset, DIN = 1010000 | 0x00007E | 0xBFD345 | 0xFFEE1D | 0x000000 | 1 | 0 | 0 | 0 |
| 6 | Enable system gain register | 0x00007E | 0xBFD345 | 0xFFEE1D | 0x000000 | 0 | 0 | 0 | 0 |
| 7 | System-calibration gain, DIN = 1011000 | 0x00007E | 0xBFD345 | 0xFFEE1D | 0x81CB5B | 0 | 0 | 0 | 0 |

Noise vs. Data Rate

The devices offer software-selectable internal oscillator frequencies as well as software-selectable output data rates. The LINEF bit in the CTRL1 register (Table 12) determines the internal oscillator frequency. The RATE bits in the command byte (Table 8) determine the ADC's output data rate. The devices also offer the option of running in zero latency single-cycle conversion mode (Table 2) or continuous conversion mode (Table 1). Set SCYCLE = 0 in the CTRL1 register (Table 12) to run in continuous conversion mode and SCYCLE = 1 for single-cycle conversion mode.

Single-cycle conversion mode gives an output result with no data latency. The devices output data up to 100sps (2.048MHz internal oscillator) or 120sps (2.4576MHz internal oscillator) with no data latency. In continuous conversion mode, the output data rate is four times the single-cycle conversion mode, for sample rates up to 400sps or 480sps. In continuous conversion mode, the output data requires three additional 24-bit cycles to settle from an input step.

Digital Filter

The devices include a SINC⁴ digital filter that produces spectral nulls at the multiples of the data rate. For all data rates less than 30sps, a spectral null occurs at the

line frequency of 60Hz and is guaranteed to attenuate 60Hz normal-mode components by more than 100dB. Simultaneous 50Hz and 60Hz attenuation can be accomplished by using an external clock with a frequency of 2.25275MHz. This guarantees a minimum of 80dB rejection at 50Hz and 85dB rejection at 60Hz. The SINC⁴ filter has a -3dB frequency equal to 24% of the data rate. See Figures 1 and 2.

GPIOs

The devices provide four GPIO ports. When set as outputs, these digital I/Os can be used to drive the digital inputs to a multiplexer or multichannel switch. Figure 3 details an example where four single-ended signals are multiplexed in a break-before-make switching sequence, using the MAX313, a quad SPST analog switch.

The devices' GPIO ports are configurable through the CTRL2 register. See Table 13. To select AIN1, write the command to CTRL2 according to Table 4a. This selects all GPIOs as outputs, as well as setting all logic signals to 0 except the selected channel AIN1.

To select channel AIN3 next, it is a good idea to set all switches to a high-impedance state first (see Table 4b).

Then select channel AIN3 by driving IN3 high (see Table 4c).

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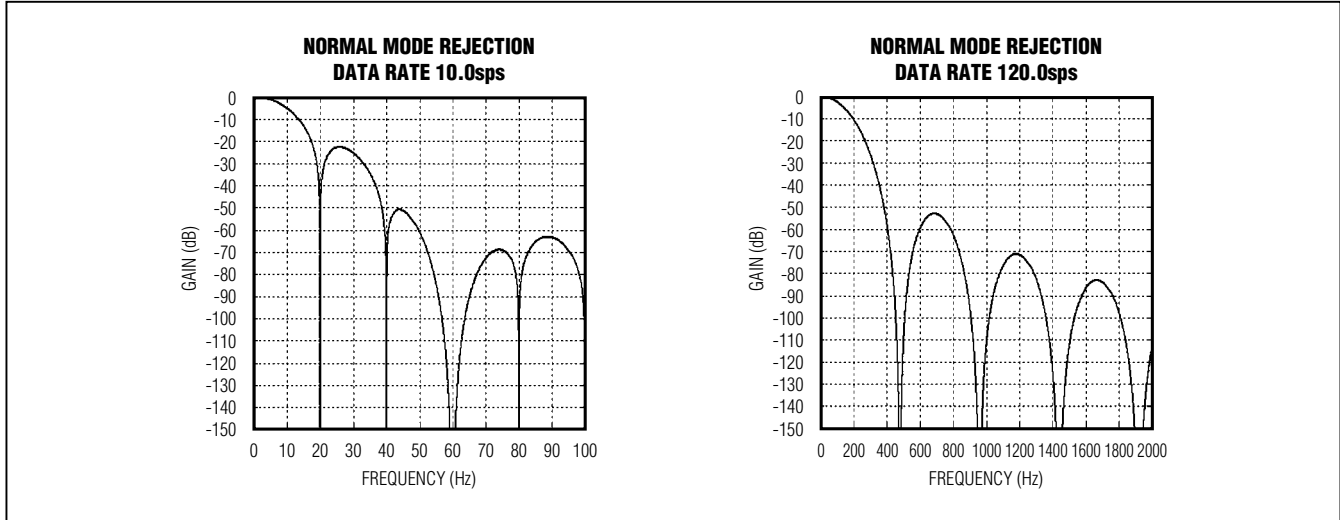


Figure 1. Normal-Mode Frequency Response (2.4576MHz Oscillator, LINEF = 0)

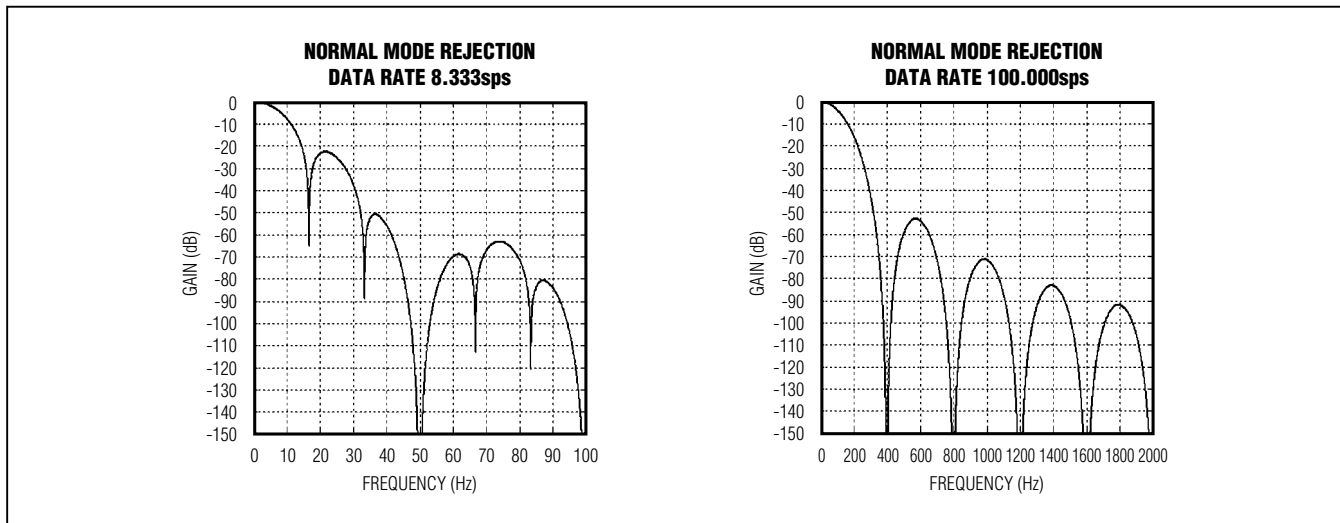


Figure 2. Normal-Mode Frequency Response (2.048MHz Oscillator, LINEF = 1)

Table 4a. Data Command to Select Channel AIN1 in Figure 3

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|----------|------|------|------|------|------|------|------|------|
| BIT NAME | DIR4 | DIR3 | DIR2 | DIR1 | DIO4 | DIO3 | DIO2 | DIO1 |
| VALUE | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 1 |

Table 4b. Set All Channels High Impedance in Figure 3

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|----------|------|------|------|------|------|------|------|------|
| BIT NAME | DIR4 | DIR3 | DIR2 | DIR1 | DIO4 | DIO3 | DIO2 | DIO1 |
| VALUE | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 |

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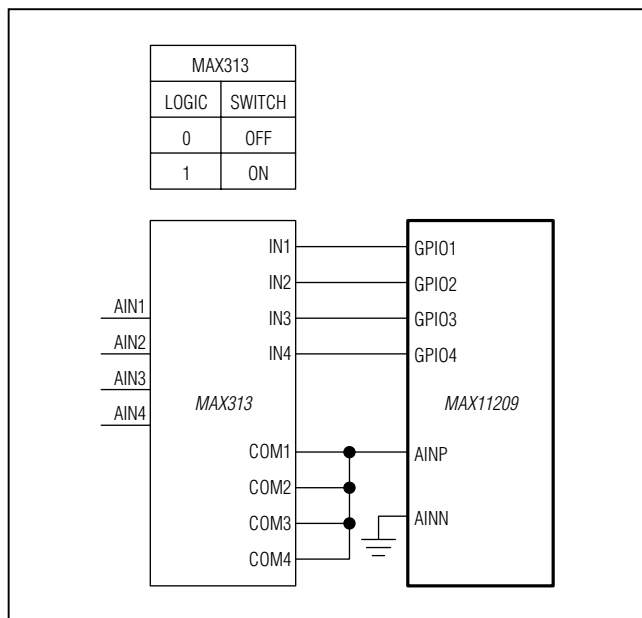


Figure 3. MAX11209 GPIOs Drive an External 4-Channel Switch (MAX313)

It is not always necessary to transition to a high-impedance state between channel selections, but depends on the source analog signals as well as the control structure of the multiplexed switches.

Digital Programmable Gain (MAX11209)

The MAX11209 offers programmable gain settings that can be digitally set to 1, 2, 4, 8, 16, 32, 64, or 128. The DGAIN_ bits in the CTRL3 register (see Table 14) configure the digital gain setting and control the input referred gain. The MAX11209's input range is 0V to VREF/gain (unipolar) or $\pm VREF/gain$ (bipolar). The MAX11209 always outputs 18 bits of data. But as this is a digital programmable gain, the noise floor remains constant, depending on the output rate setting. At an output rate of 10sps, as shown in Figure 4, the noise floor is such that all gain settings from 1 to 32 provide 18 bits of noise-free resolution. A gain setting of 128 at 10sps means the LSB is below the noise floor. The MAX11209 digital gain is beneficial for low-voltage applications that only require a small portion of the 0V to VREF or $\pm VREF$ ranges.

Table 4c. Data Command to Select Channel AIN3 in Figure 3

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|----------|------|------|------|------|------|------|------|------|
| BIT NAME | DIR4 | DIR3 | DIR2 | DIR1 | DIO4 | DIO3 | DIO2 | DIO1 |
| VALUE | 1 | 1 | 1 | 1 | 0 | 1 | 0 | 0 |

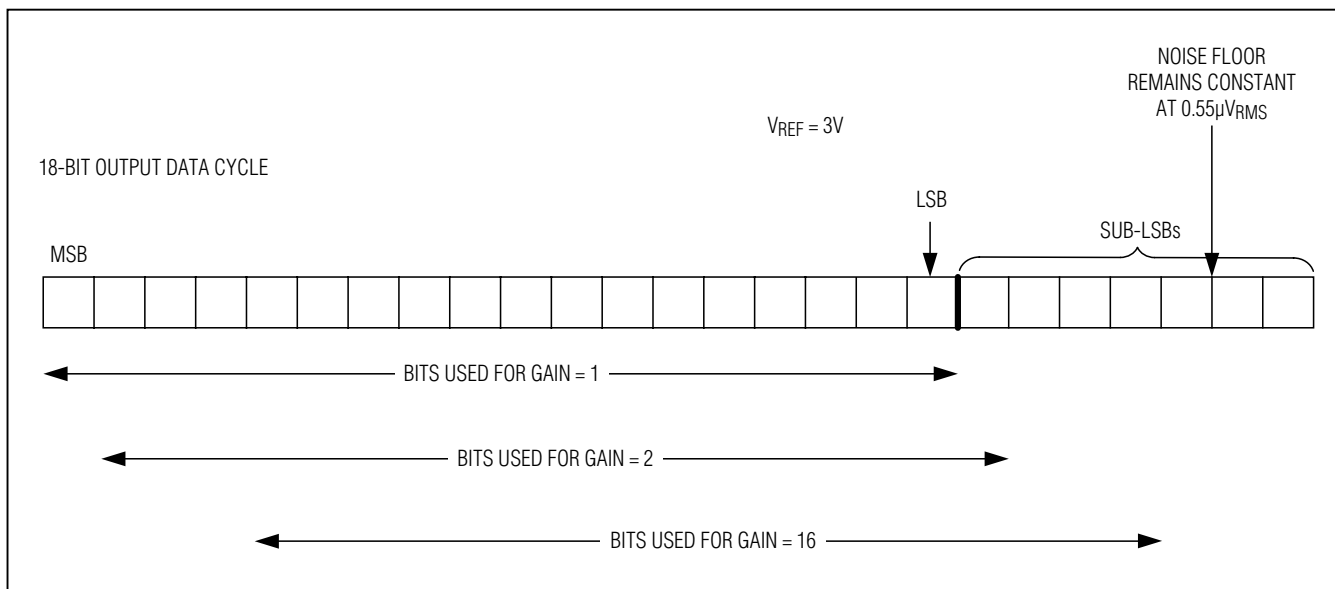


Figure 4. MAX11209 Digital Programmable Gain Example (10sps Output Rate)

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Serial-Digital Interface

The MAX11209/MAX11211 interface is fully compatible with SPI-, QSPI-, and MICROWIRE-standard serial interfaces. The SPI interface provides access to nine on-chip registers that are 8 or 24 bits wide.

Drive \overline{CS} low to transfer data in and out of the devices. Clock in data at DIN on the rising edge of SCLK. The $\overline{RDY}/DOUT$ output serves two functions: conversion status and data read. To find the conversion status, assert \overline{CS} low and read the $\overline{RDY}/DOUT$ output; the conversion

is in progress if the $\overline{RDY}/DOUT$ output reads logic-high and the conversion is complete if the $\overline{RDY}/DOUT$ output reads logic-low. Data at $\overline{RDY}/DOUT$ changes on the falling edge of SCLK and is valid on the rising edge of SCLK. DIN and DOUT are transferred MSB first. Drive \overline{CS} high to force DOUT high impedance and cause the devices to ignore any signals on SCLK and DIN. Connect \overline{CS} low for 3-wire operation. Figures 5, 6, and 7 show the SPI timing diagrams.

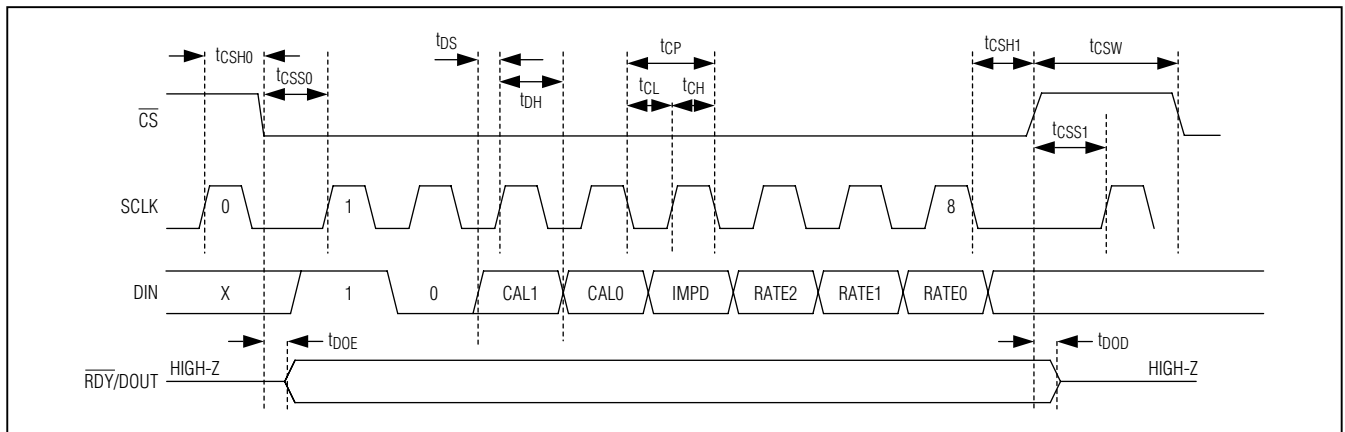


Figure 5. SPI Command Byte

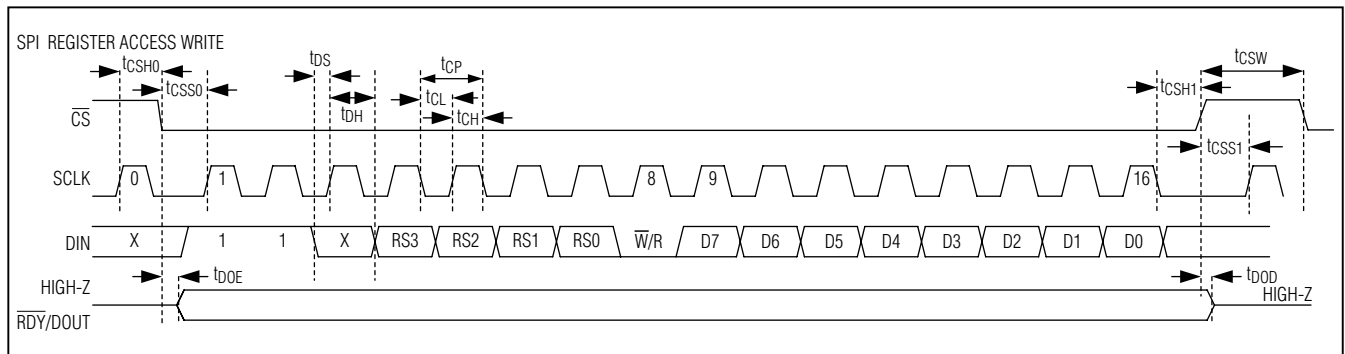


Figure 6. SPI Register Access Write

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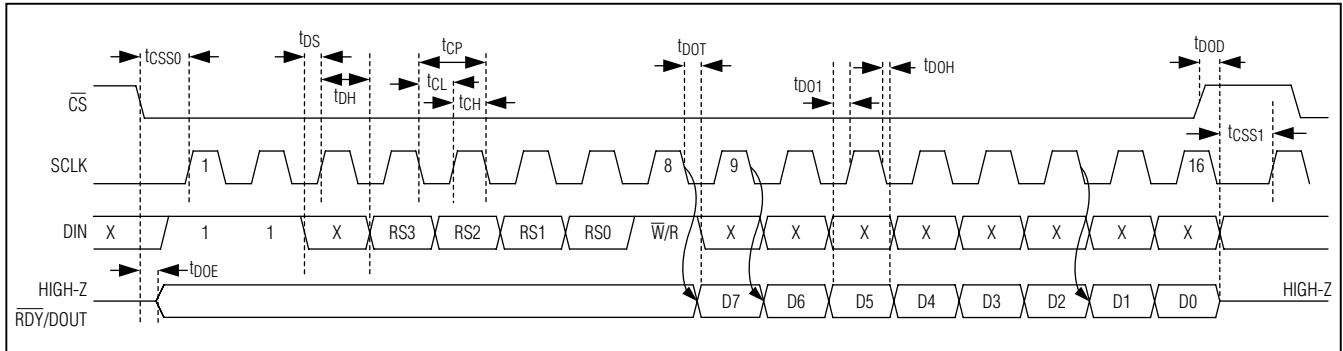


Figure 7. SPI Register Access Read

Command Byte

Communication between the user and the device is conducted through SPI using a command byte. The command byte consists of two modes differentiated as command modes and data modes. Command modes and data modes are further differentiated by decoding the remaining bits in the command byte. The mode selected is determined by the MODE bit. If the MODE bit is 0, then the user is requesting either a conversion, calibration, or power-down; see Table 5. If the MODE bit is 1, then the user is selecting a data command and can either read from or write to a register; see Table 6.

The Status register (STAT1) is a read-only register and provides general chip operational status to the user. If the user attempts to calibrate the system and overranges the internal signal scaling, then a gain overrange condition is flagged with the SYSOR bit. The last data rate programmed for the ADC is available in the RATE bits. If the input signal has exceeded positive or negative full scale, this condition is flagged with the OR and UR bits. If the modulator is busy converting, then the MSTAT bit is set. If a conversion result is ready for readout, the RDY bit is set; see Table 11.

The Control 1 register (CTRL1) is a read/write register, and the bits determine the internal oscillator frequency, unipolar or bipolar input range, selection of an internal or external clock, enabling or disabling the reference and input signal buffers, the output data format (offset binary

or two's complement), and single-cycle or continuous conversion mode. See Table 12.

The Control 2 register (CTRL2) is a read/write register, and the bits configure the GPIOs as inputs or outputs and their values. See Table 13.

The Control 3 register (CTRL3) is a read/write register, and the bits determine the MAX11209 programmable gain setting and the calibration register settings for both the MAX11209 and MAX11211. See Table 14.

The Data register (DATA) is a read-only register. Data is output from RDY/DOUT on the next 24 SCLK cycles once CS is forced low. The data bits transition on the falling edge of SCLK. Data is output MSB first, and is offset binary or two's complement, depending on the setting of the FORMAT bit in the CTRL1 register. See Table 15.

The System Offset Calibration register (SOC) is a read/write register, and the bits contain the digital value that corrects the data for system zero scale. See Table 17.

The System Gain Calibration register (SGC) is a read/write register, and the bits contain the digital value that corrects the data for system full scale. See Table 18.

The Self-Cal Offset Calibration register (SCOC) is a read/write register, and the bits contain the value that corrects the data for chip zero scale. See Table 19.

The Self-Cal Gain Calibration register (SCGC) is a read/write register, and the bits contain the value that corrects the data for chip full scale. See Table 20.

Table 5. Command Byte (MODE = 0)

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|----------|-----------|----------|------|------|------|-------|-------|-------|
| BIT NAME | START = 1 | MODE = 0 | CAL1 | CAL0 | IMPD | RATE2 | RATE1 | RATE0 |

Table 6. Command Byte (MODE = 1)

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|----------|-----------|----------|----|-----|-----|-----|-----|-----|
| BIT NAME | START = 1 | MODE = 1 | 0 | RS3 | RS2 | RS1 | RS0 | W/R |

Note: The START bit is used to synchronize the data from the host device. The START bit is always 1.

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Table 7. Operating Mode (MODE Bit)

| MODE BIT SETTING | OPERATING MODE |
|------------------|---|
| 0 | The command byte initiates a conversion or an immediate power-down. See Tables 5 and 8. |
| 1 | The device interprets the command byte as a register access byte, which is decoded as per Tables 6 and 9. |

Table 8. Command Byte (MODE = 0, LINEF = 0)

| COMMAND | START | MODE | CAL1 | CAL0 | IMPD | RATE2 | RATE1 | RATE0 |
|---------------------------------|-------|------|------|------|------|-------|-------|-------|
| Self-calibration cycle | 1 | 0 | 0 | 1 | 0 | 0 | 0 | 0 |
| System offset calibration cycle | 1 | 0 | 1 | 0 | 0 | 0 | 0 | 0 |
| System gain calibration | 1 | 0 | 1 | 1 | 0 | 0 | 0 | 0 |
| Immediate power-down | 1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |
| Convert 1sps | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Convert 2.5sps | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |
| Convert 5sps | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 |
| Convert 10sps | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 1 |
| Convert 15sps | 1 | 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| Convert 30sps | 1 | 0 | 0 | 0 | 0 | 1 | 0 | 1 |
| Convert 60sps | 1 | 0 | 0 | 0 | 0 | 1 | 1 | 0 |
| Convert 120sps | 1 | 0 | 0 | 0 | 0 | 1 | 1 | 1 |

Table 9. Register Selection (MODE = 1)

| RS3 | RS2 | RS1 | RS0 | REGISTER ACCESS | POWER-ON RESET STATUS | REGISTER SIZE (BITS) |
|-----|-----|-----|-----|-----------------|-----------------------|----------------------|
| 0 | 0 | 0 | 0 | STAT1 | 0x00 | 8 |
| 0 | 0 | 0 | 1 | CTRL1 | 0x02 | 8 |
| 0 | 0 | 1 | 0 | CTRL2 | 0x0F | 8 |
| 0 | 0 | 1 | 1 | CTRL3 | 0x1E | 8 |
| 0 | 1 | 0 | 0 | DATA | 0x000000 | 24 |
| 0 | 1 | 0 | 1 | SOC | 0x000000 | 24 |
| 0 | 1 | 1 | 0 | SGC | 0x000000 | 24 |
| 0 | 1 | 1 | 1 | SCOC | 0x000000 | 24 |
| 1 | 0 | 0 | 0 | SCGC | 0x000000 | 24 |

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Table 10. Register Address Map

| REGISTER NAME | R/W | ADDRESS SEL (RS[3:0]) | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|---------------|-----|-----------------------|----------|---------|---------|--------|--------|--------|--------|----------|
| STAT1 | R | 0x0 | SYSOR | RATE2 | RATE1 | RATE0 | OR | UR | MSTAT | RDY |
| CTRL1 | R/W | 0x1 | LINEF | U/B | EXTCLK | REFBUF | SIGBUF | FORMAT | SCYCLE | RESERVED |
| CTRL2 | R/W | 0x2 | DIR4 | DIR3 | DIR2 | DIR1 | DIO4 | DIO3 | DIO2 | DIO1 |
| CTRL3 | R/W | 0x3 | DGAIN2* | DGAIN1* | DGAIN0* | NOSYSG | NOSYSO | NOSCG | NOSCO | RESERVED |
| DATA | R | 0x4 | D[17:16] | | | | | | | |
| | | | D[15:8] | | | | | | | |
| | | | D[7:0] | | | | | | | |
| SOC | R/W | 0x5 | B[23:16] | | | | | | | |
| | | | B[15:8] | | | | | | | |
| | | | B[7:0] | | | | | | | |
| SGC | R/W | 0x6 | B[23:16] | | | | | | | |
| | | | B[15:8] | | | | | | | |
| | | | B[7:0] | | | | | | | |
| SCOC | R/W | 0x7 | B[23:16] | | | | | | | |
| | | | B[15:8] | | | | | | | |
| | | | B[7:0] | | | | | | | |
| SCGC | R/W | 0x8 | B[23:16] | | | | | | | |
| | | | B[15:8] | | | | | | | |
| | | | B[7:0] | | | | | | | |

*These DGAIN_ bits set the digital gain for the MAX11209. These bits are don't-care bits for the MAX11211.

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STAT1: Status Register

Table 11. STAT1 Register (Read Only)

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|----------|-------|-------|-------|-------|----|----|-------|-----|
| BIT NAME | SYSOR | RATE2 | RATE1 | RATE0 | OR | UR | MSTAT | RDY |
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

SYSOR: The system gain overrange bit, when set to 1, indicates that a system gain calibration was over range. The SCGC calibration coefficient is maximum value of 1.9999999. This bit, when set to 1, indicates that the full-scale value out of the converter is likely not available.

RATE[2:0]: The data rate bits indicate the conversion rate that corresponds to the result in the DATA register or the rate that was used for calibration coefficient calculation. If the previous conversions were done at a different rate, the RATE[2:0] bits indicate a rate different than the rate of the conversion in progress.

OR: The overrange bit, OR, is set to 1 to indicate the conversion result has exceeded the maximum value of the converter and that the result has been clipped or limited to the maximum value. The OR bit is set to 0 to indicate the conversion result is within the full-scale range of the device.

UR: The underrange bit, UR, is set to 1 to indicate the conversion result has exceeded the minimum value of the converter and that the result has been clipped or limited to the minimum value. The UR bit is set to 0 to indicate the conversion result is within the full-scale range of the device.

MSTAT: The measurement status bit, MSTAT is set to 1 when a signal measurement is in progress. When MSTAT = 1, a conversion, self-calibration, or system calibration is in progress and indicates that the modulator is busy. When the modulator is not converting, the MSTAT bit is set to 0.

RDY: The RDY ready bit is set to 1 to indicate that a conversion result is available. Reading the DATA register resets the RDY bit to 0 only after another conversion has been initiated. If the DATA has not been read before another conversion is initiated, the RDY bit remains 1; if the DATA is read before another conversion is initiated, the RDY bit resets to 0. If the DATA for the previous conversion is read during a following conversion, the RDY bit is reset immediately after the DATA read operation has completed.

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CTRL1: Control 1 Register

The byte-wide CTRL1 register is a bidirectional read/write register. The byte written to the CTRL1 register indicates if the part converts continuously or single cycle, if an external or internal clock is used, if the reference and signal buffers are activated, the format of the data when in bipolar mode, and if the analog signal input range is unipolar or bipolar.

Table 12. CTRL1 Register (Read/Write)

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|----------|-------|--------------|--------|--------|--------|--------|--------|--------|
| BIT NAME | LINEF | U/ \bar{B} | EXTCLK | REFBUF | SIGBUF | FORMAT | SCYCLE | UNUSED |
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 |

LINEF: Use the line frequency bit, LINEF, to select if the data rate is centered for 50Hz power mains or 60Hz power mains. To select data rates for 50Hz power mains, write 1 to LINEF and to select data rates for 60Hz power mains, write 0 to LINEF.

U/ \bar{B} : The unipolar/bipolar bit, U/ \bar{B} , selects if the input range is unipolar or bipolar. A 1 in this bit location selects a unipolar input range and a 0 selects a bipolar input range.

EXTCLK: The external clock bit, EXTCLK, controls the selection of the system clock. A 1 enables an external clock as system clock, whereas as a 0 enables the internal clock.

REFBUF: The reference buffer bit, REFBUF, enables/disables the reference buffers. A 1 enables the reference buffers. A 0 powers down the reference buffers and the reference inputs bypass the reference buffers when driving the ADC.

SIGBUF: The signal buffer, SIGBUF, enables/disables the signal buffers. A 1 enables the signal buffer. A 0 powers down the signal buffers and the analog signal inputs bypass the signal buffers when driving the ADC.

FORMAT: The format bit, FORMAT, controls the digital format of the data. Unipolar data is always in offset binary format. The bipolar format is two's complement if the FORMAT bit is set to 0 or offset binary if the FORMAT bit is set to 1.

SCYCLE: The single-cycle bit, SCYCLE, determines if the device runs in "no-latency" single-conversion mode (SCYCLE = 1) or if the device runs in "latent" continuous-conversion mode (SCYCLE = 0). When in single-cycle conversion mode, the device completes one no-latency conversion and then powers down into a leakage-only state. When in continuous-conversion mode, the part is continuously converting and the first three data from the part are incorrect due to the SINC⁴ filter latency.

Important Note: When operating in continuous-conversion mode (SCYCLE = 0), it is recommended to keep \overline{CS} low to properly detect the end of conversion. The end of conversion is signaled by $\overline{RDY}/DOUT$ changing from 0 to 1. The transition of $\overline{RDY}/DOUT$ from 0 to 1 must be used to synchronize the DATA register read back. If the $\overline{RDY}/DOUT$ output is not used to synchronize the DATA read back, a timing hazard exists where the DATA register is updated internally after a conversion has completed simultaneously with the DATA register being read out, causing an incorrect read of DATA.

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CTRL2: Control 2 Register

The byte-wide CTRL2 register is a bidirectional read/write register. The byte written to the CTRL2 register controls the direction and values of the digital I/O ports.

Table 13. CTRL2 Register (Read/Write)

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|----------|------|------|------|------|------|------|------|------|
| BIT NAME | DIR4 | DIR3 | DIR2 | DIR1 | DIO4 | DIO3 | DIO2 | DIO1 |
| DEFAULT | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |

DIR[4:1]: The direction bits configure the direction of the DIO bit. When a DIR bit is set to 0, the associated DIO bit is configured as an input and the value returned by a read of the DIO bit is the value being driven on the associated GPIO. When a DIR bit is set to 1, the associated DIO bit is configured as an output and the GPIO port is driven to a logic value of the associated DIO bit.

DIO[4:1]: The data input/output bits are bits associated with the GPIO ports. When a DIO is configured as an input, the value read from the DIO bit is the logic value being driven at the GPIO port. When the direction is configured as an output, the GPIO port is driven to a logic value associated with the DIO bit.

CTRL3: Control 3 Register

The byte-wide CTRL3 register is a bidirectional read/write register. The CTRL3 register controls the operation and calibration of the device.

Table 14. CTRL3 Register (Read/Write)

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|----------|---------|---------|---------|--------|--------|-------|-------|----------|
| BIT NAME | DGAIN2* | DGAIN1* | DGAIN0* | NOSYSG | NOSYSO | NOSCG | NOSCO | RESERVED |
| DEFAULT | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 0 |

*These DGAIN_ bits are don't-care bits for the MAX11209.

DGAIN[2:0] (MAX11209 only): The digital gain bits control the input referred gain. With a gain of 1, the input range is 0 to VREF (unipolar) or $\pm VREF$ (bipolar). As the gain is increased by 2x, the input range is reduced to 0 to VREF/gain or $\pm VREF/gain$. Digital gain is applied to the final offset and gain-calibrated digital data. The DGAIN[2:0] bits decode to digital gains as follows:

| | |
|---------|-----------|
| 000 = 1 | 100 = 16 |
| 001 = 2 | 101 = 32 |
| 010 = 4 | 110 = 64 |
| 011 = 8 | 111 = 128 |

NOSYSG: The no-system gain bit, NOSYSG, controls the system gain calibration coefficient. A 1 in this bit location disables the use of the system gain value when computing the final offset and gain corrected data value. A 0 in this location enables the use of the system gain value when computing the final offset and gain corrected data value.

NOSYSO: The no system offset bit, NOSYSO, controls the system offset calibration coefficient. A 1 in this location disables the use of the system offset value when computing the final offset and gain corrected data value. A 0 in this location enables the use of the system offset value when computing the final offset and gain corrected data value.

NOSCG: The no self-calibration gain bit, NOSCG, controls the self-calibration gain calibration coefficient. A 1 in this location disables the use of the self-calibration gain value when computing the final offset and gain corrected data value. A 0 in this location enables the use of the self-calibration gain value when computing the final offset and gain corrected data value.

NOSCO: The no self-calibration offset bit, NOSCO, controls the use of the self-calibration offset calibration coefficient. A 1 in this location disables the use of the self-calibration offset value when computing the final offset and gain corrected data value. A 0 in this location enables the use of the self-calibration offset value when computing the final offset and gain corrected data value.

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DATA: Data Register

The data register is a 24-bit read-only register. Any attempt to write data to the data register has no effect. The data read from this register is clocked out MSB first. The data register holds the conversion result. D17 is the MSB, and D0 is the LSB. The result is stored in a format according to the FORMAT bit in the CTRL1 register.

The data format while in unipolar mode is always straight binary. In straight binary format, the most negative value is 0x00000 (VAINP - VAINN = 0V), the midscale value is 0x20000 (VAINP - VAINN = VREF/2), and the most positive value is 0x3FFFF (VAINP - VAINN = VREF).

In bipolar mode, if the FORMAT bit = 1, then the data format is offset binary. If the FORMAT bit = 0, then the data format is two's complement. In two's complement the negative full-scale value is 0x20000 (VAINP - VAINN = -VREF), the midscale is 0x00000 (VAINP - VAINN = 0V), and the positive full scale is 0x1FFFF (VAINP - VAINN = VREF). Any input exceeding the available input range is limited to the minimum or maximum data value.

Table 15. DATA Register (Read Only)

| BIT | D17 | D16 | D15 | D14 | D13 | D12 | D11 | D10 |
|---------|-----|-----|-----|-----|-----|-----|-----|-----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

| BIT | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 |
|---------|----|----|----|----|----|----|----|----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

| BIT | D1 | D0 | | | | | | |
|---------|----|----|---|---|---|---|---|---|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Table 16a. Output Data Format for the Unipolar Input Range

| INPUT VOLTAGE VAINP - VAINN | DIGITAL OUTPUT CODE FOR UNIPOLAR RANGE |
|--|--|
| | STRAIGHT BINARY FORMAT |
| $\geq V_{REF}$ | 0x3FFFF |
| $V_{REF} \times \left(1 - \frac{1}{2^{18} - 1}\right)$ | 0x3FFFF |
| $\frac{V_{REF}}{2^{18} - 1}$ | 0x00001 |
| 0 | 0x00000 |

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Table 16b. Output Data Formats for the Bipolar Input Range

| INPUT VOLTAGE $V_{AINP} - V_{AINN}$ | DIGITAL OUTPUT CODE FOR BIPOLAR RANGES | |
|---|--|-------------------------|
| | OFFSET BINARY FORMAT | TWO'S COMPLEMENT FORMAT |
| $\geq V_{REF}$ | 0x3FFFF | 0x1FFFF |
| $V_{REF} \times \left(1 - \frac{1}{2^{17} - 1}\right)$ | 0x3FFFE | 0x1FFFE |
| $\frac{V_{REF}}{2^{17} - 1}$ | 0x2000 | 0x00001 |
| 0 | 0x20000 | 0x00000 |
| $\frac{-V_{REF}}{2^{17} - 1}$ | 0x1FFFF | 0x3FFFF |
| $-V_{REF} \times \left(1 - \frac{1}{2^{17} - 1}\right)$ | 0x00001 | 0x20001 |
| $\leq -V_{REF}$ | 0x00000 | 0x20000 |

SOC: System Offset Calibration Register

The system offset calibration register is a 24-bit read/write register. The data written/read to/from this register is clocked in/out MSB (most significant bit) first. This register holds the system offset calibration value. The format is always in two's complement binary format. A write to the system-calibration register is allowed. The value written remains valid until it is either rewritten or until an on-demand system-calibration operation is performed, which overwrites the user-supplied value.

The system offset calibration value is subtracted from each conversion result provided the NOSYSO bit in the CTRL3 register is set to 0. The system offset calibration value is subtracted from the conversion result after self-calibration but before system gain correction. The system offset calibration value is also applied prior to the 1x or 2x scale factor associated with bipolar and unipolar modes.

Table 17. SOC Register (Read/Write)

| BIT | B23 | B22 | B21 | B20 | B19 | B18 | B17 | B16 |
|---------|-----|-----|-----|-----|-----|-----|-----|-----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

| BIT | B15 | B14 | B13 | B12 | B11 | B10 | B9 | B8 |
|---------|-----|-----|-----|-----|-----|-----|----|----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|---------|----|----|----|----|----|----|----|----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

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SGC: System Gain Calibration Register

The system gain calibration register is a 24-bit read/write register. The data written/read to/from this register is clocked in/out MSB first. This register holds the system gain calibration value. The format is always in two's complement binary format. A write to the system-calibration register is allowed. The written value remains valid until it is either rewritten or until an on-demand system-calibration operation is performed, which overwrites the user-supplied value.

The system gain calibration value is used to scale the offset corrected conversion result, provided the NOSYSG bit in the CTRL3 register is set to 0. The system gain calibration value scales the offset-corrected result by up to 2x or corrects a gain error of approximately -50%. The amount of positive gain error that can be corrected is determined by modulator overload characteristics, which can be as much as +25%. The gain is corrected to within 2 LSB.

Table 18. SGC Register (Read/Write)

| BIT | B23 | B22 | B21 | B20 | B19 | B18 | B17 | B16 |
|---------|-----|-----|-----|-----|-----|-----|-----|-----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

| BIT | B15 | B14 | B13 | B12 | B11 | B10 | B9 | B8 |
|---------|-----|-----|-----|-----|-----|-----|----|----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|---------|----|----|----|----|----|----|----|----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

SCOC: Self-Calibration Offset Register

The self-calibration offset register is a 24-bit read/write register. The data written/read to/from this register is clocked in/out MSB first. This register holds the self-calibration offset value. The format is always in two's complement binary format. A write to the self-calibration offset register is allowed. The written value remains valid until it is either rewritten or until an on-demand self-calibration operation is performed, which overwrites the user-supplied value.

The self-calibration offset value is subtracted from each conversion result provided the NOSCO bit in the CTRL3 register is set to 0. The self-calibration offset value is subtracted from the conversion result before the self-calibration gain correction and before the system offset and gain correction. The self-calibration offset value is also applied prior to the 2x scale factor associated with unipolar mode.

Table 19. SCOC Register (Read/Write)

| BIT | B23 | B22 | B21 | B20 | B19 | B18 | B17 | B16 |
|---------|-----|-----|-----|-----|-----|-----|-----|-----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

| BIT | B15 | B14 | B13 | B12 | B11 | B10 | B9 | B8 |
|---------|-----|-----|-----|-----|-----|-----|----|----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|---------|----|----|----|----|----|----|----|----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

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SCGC: Self-Calibration Gain Register

The self-calibration gain register is a 24-bit read/write register. The data written/read to/from this register is clocked in/out MSB first. This register holds the self-calibration gain calibration value. The format is always in two's complement binary format. A write to the self-calibration register is allowed. The written value remains valid until it is either rewritten or until an on-demand self-calibration operation is performed, which overwrites the user-supplied value. Any attempt to write to this register during an active calibration operation is ignored.

The self-calibration gain value is used to scale the self-calibration offset corrected conversion result before the system offset and gain calibration values have been applied, provided the NOSCG bit in the CTRL3 register is set to 0. The self-calibration gain value scales the self-calibration offset corrected conversion result by up to 2x or can correct a gain error of approximately -50%. The gain is corrected to within 2 LSB.

Table 20. SCGC Register (Read/Write)

| BIT | B23 | B22 | B21 | B20 | B19 | B18 | B17 | B16 |
|---------|-----|-----|-----|-----|-----|-----|-----|-----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

| BIT | B15 | B14 | B13 | B12 | B11 | B10 | B9 | B8 |
|---------|-----|-----|-----|-----|-----|-----|----|----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

| BIT | B7 | B6 | B5 | B4 | B3 | B2 | B1 | B0 |
|---------|----|----|----|----|----|----|----|----|
| DEFAULT | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Table 21. Data Rates for All Combinations of RATE[2:0] (LINEF = 0)

| RATE[2:0] | SINGLE-CYCLE DATA RATE (sps) | CONTINUOUS DATA RATE (sps) |
|-----------|------------------------------|----------------------------|
| 000 | 1 | — |
| 001 | 2.5 | — |
| 010 | 5 | — |
| 011 | 10 | — |
| 100 | 15 | 60 |
| 101 | 30 | 120 |
| 110 | 60 | 240 |
| 111 | 120 | 480 |

Table 22. Data Rates for All Combinations of RATE[2:0] (LINEF = 1)

| RATE[2:0] | SINGLE-CYCLE DATA RATE (sps) | CONTINUOUS DATA RATE (sps) |
|-----------|------------------------------|----------------------------|
| 000 | 0.833 | — |
| 001 | 2.08 | — |
| 010 | 4.17 | — |
| 011 | 8.33 | — |
| 100 | 12.5 | 50 |
| 101 | 25 | 100 |
| 110 | 50 | 200 |
| 111 | 100 | 400 |

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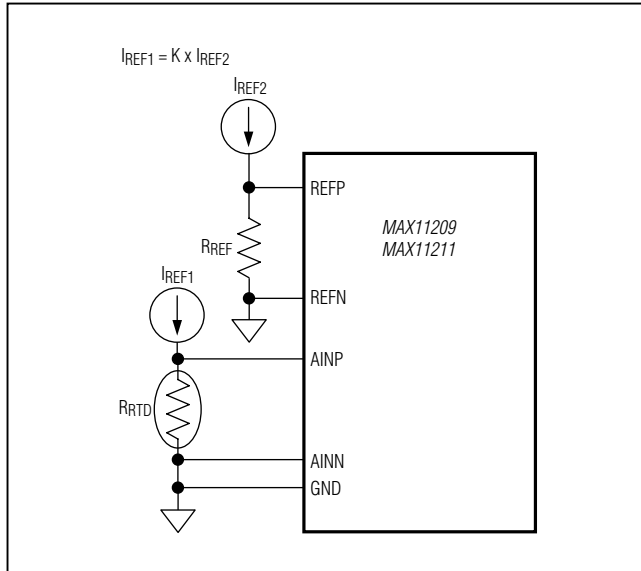


Figure 8. RTD Temperature Measurement Circuit

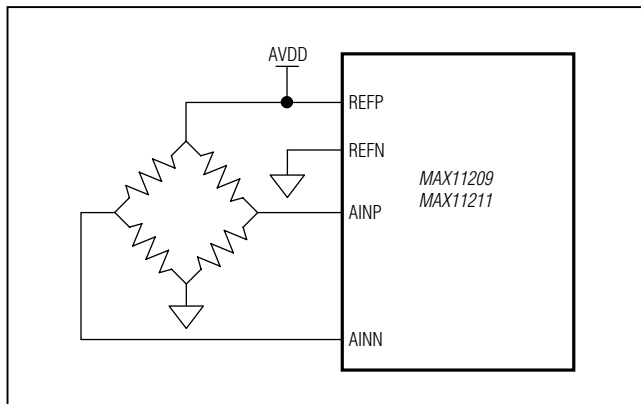


Figure 9. Resistive Bridge Measurement Circuit

Applications Information

See Figure 8 for the RTD temperature measurement circuit and Figure 9 for a resistive bridge measurement circuit.

Adding more active circuitry to the analog input signal path is not always the best solution to a small-signal problem. Sometimes, increasing the dynamic range of an active device can lead to a simpler solution that also helps power consumption and linearity.

Often, circuit designers immediately look for an external op amp or programmable gain amplifier (PGA) when confronted with coupling low-amplitude signals to sampled digital systems. In many cases, choosing an ADC with more dynamic range and better low-noise performance yields a solution that works better, simpler, and with less power.

One such example is measurements from a strain gauge in a Wheatstone bridge configuration. Assuming a differential output signal from the bridge in Figure 10, the bridge's output voltage varies from 5mV to 105mV, while the noise of the bridge itself limits the sensitivity to approximately 1 μ V. This gives approximately 100,000 discrete levels that are available for quantization, a feat accomplished quite well with any ADC having 17 bits or more of usable resolution. However, as it is not likely that a 19-bit ADC will have an input range of 105mV, a gain stage is needed to boost the signal to span the input range of the ADC (typically between 3V and 5V).

This solution requires finding an amplifier and associated passives that meet the overall system noise and linearity needs. Also, the power consumed in the gain stage may equal or surpass that of the ADC itself, a fact that is significant in systems where power consumption is severely constrained, such as portable sensors or 4–20mA loops.

The low-noise floor of the MAX11209 family of 16-/18-/20-bit devices gives the designer the ability to use simple binary shifting (digital gain) of the data word to align the sample range with the available code space. Digital gain is internally available in the MAX11209.

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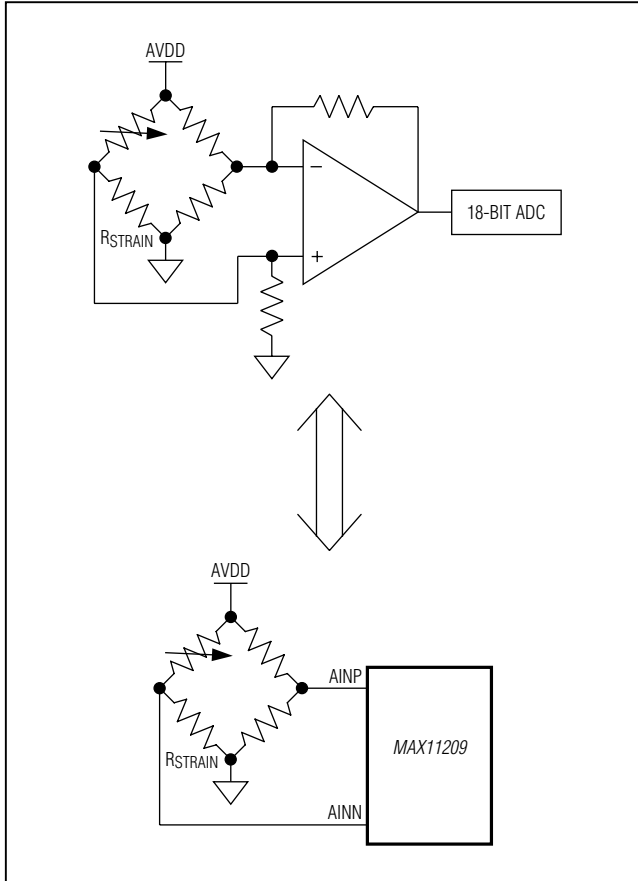


Figure 10. The MAX11209 ADC Eliminates an External Gain Stage.

Chip Information

PROCESS: BiCMOS

Package Information

For the latest package outline information and land patterns, go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

| PACKAGE TYPE | PACKAGE CODE | OUTLINE NO. | LAND PATTERN NO. |
|--------------|--------------|-------------------------|-------------------------|
| 16 QSOP | E16+4 | 21-0055 | 90-0167 |

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Revision History

| REVISION NUMBER | REVISION DATE | DESCRIPTION | PAGES CHANGED |
|-----------------|---------------|-----------------|---------------|
| 0 | 8/10 | Initial release | — |



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