General Description

The MAX4475–MAX4478/MAX4488/MAX4489 wideband, low-noise, low-distortion operational amplifiers offer rail-to-rail outputs and single-supply operation down to 2.7V. They draw 2.2mA of quiescent supply current per amplifier while featuring ultra-low distortion (0.0002% THD+N), as well as low input voltage-noise density (4.5nV/ \sqrt{Hz}) and low input current-noise density (0.5fA/ \sqrt{Hz}). These features make the devices an ideal choice for applications that require low distortion and/or low noise.

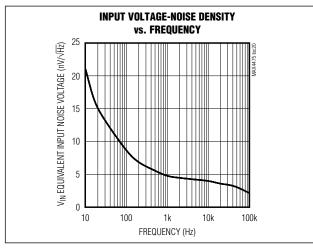
For power conservation, the MAX4475/MAX4488 offer a low-power shutdown mode that reduces supply current to 0.01µA and places the amplifiers' outputs into a high-impedance state. These amplifiers have outputs which swing rail-to-rail and their input common-mode voltage range includes ground. The MAX4475–MAX4478 are unity-gain stable with a gain-bandwidth product of 10MHz. The MAX4488/4489 are internally compensated for gains of +5V/V or greater with a gain-bandwidth product of 42MHz. The single MAX4475/MAX4476/MAX4488 are available in space-saving, 6-pin SOT23 and TDFN packages.

Applications

ADC Buffers DAC Output Amplifiers Low-Noise Microphone/Preamplifiers Digital Scales Strain Gauges/Sensor Amplifiers Medical Instrumentation

µMAX is a registered trademark of Maxim Integrated Products, Inc.

Typical Operating Characteristic



_Features

◆ Low Input Voltage-Noise Density: 4.5nV/√Hz

- ◆ Low Input Current-Noise Density: 0.5fA/√Hz
- Low Distortion: 0.0002% THD+N (1kΩ load)
- ♦ Single-Supply Operation from +2.7V to +5.5V
- Input Common-Mode Voltage Range Includes Ground
- Rail-to-Rail Output Swings with a 1kΩ Load
- 10MHz GBW Product, Unity-Gain Stable (MAX4475–MAX4478)
- 42MHz GBW Product, Stable with A_V ≥ +5V/V (MAX4488/MAX4489)
- Excellent DC Characteristics

 $V_{OS} = 70 \mu V$

IBIAS = 1pA

Large-Signal Voltage Gain = 120dB

- Low-Power Shutdown Mode: Reduces Supply Current to 0.01µA Places Output in High-Impedance State
- Available in Space-Saving SOT23, TDFN, μMAX[®], and TSSOP Packages

__Ordering Information

PART	RT TEMP RANGE PIN- PACKAG		TOP MARK
MAX4475AUT+T	-40°C to +125°C	6 SOT23	AAZV
MAX4475AUA+	-40°C to +125°C	8 µMAX	—
MAX4475ASA+	-40°C to +125°C	8 SO	_
MAX4475ATT+T	-40°C to +125°C	6 TDFN-EP*	+ADD
MAX4475AUT/V+T	-40°C to +125°C	6 SOT23	+ACQQ

+Denotes a lead(Pb)-free/RoHS-compliant package. *EP = Exposed pad (connect to V_{SS}).

N denotes an automotive qualified part.

T = Tape and reel.

Ordering Information continued at end of data sheet.

Pin Configurations and Typical Operating Circuit appear at end of data sheet.

Selector Guide

PART	GAIN BW (MHz)	STABLE GAIN (V/V)	NO. OF AMPS	SHDN
MAX4475	10	1	1	Yes
MAX4476	10	1	1	_
MAX4477	10	1	2	—
MAX4478	10	1	4	—
MAX4488	42	5	1	Yes
MAX4489	42	5	2	

_ Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

ABSOLUTE MAXIMUM RATINGS

Power-Supply Voltage (V_{DD} to V_{SS})-0.3V to +6.0V <u>Analog</u> Input Voltage (IN_+, IN_-)....(V_{SS} - 0.3V) to (V_{DD} + 0.3V) SHDN Input Voltage(V_{SS} - 0.3V) to +6.0V Output Short-Circuit Duration to Either SupplyContinuous Continuous Input Current (IN+, IN-)±10mA Continuous Power Dissipation (T_A = +70°C)

6-Pin SOT23 (derate 9.1mW/°C above +70°C)	727mW
6-Pin TDFN (derate 18.2mW/°C above 70°C)	
8-Pin µMAX (derate 4.5mW/°C above +70°C)	362mW
8-Pin SO (derate 5.88mW/°C above +70°C)	471mW

14-Pin SO (derate 8.33mW/°C above +70°C)667mW 14-Pin TSSOP (derate 9.1mW/°C above +70°C)727mW	
Operating Temperature Range40°C to +125°C	
Junction Temperature+150°C	
Storage Temperature Range65°C to +150°C	
Lead Temperature (soldering, 10s)+300°C	
Soldering Temperature (reflow)	
SOT23, µMAX, TSSOP, TDFN, and 8-Pin SO+260°C	
14-Pin SO+240°C	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

 $(V_{DD} = +5V, V_{SS} = 0V, V_{CM} = 0V, V_{OUT} = V_{DD}/2, R_L \text{ tied to } V_{DD}/2, \overline{SHDN} = V_{DD}, T_A = -40^{\circ}C \text{ to } +125^{\circ}C, \text{ unless otherwise noted.}$ Typical values are at $T_A = +25^{\circ}C.$ (Notes 1, 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	МАХ	UNITS		
Supply Voltage Range	V _{DD}	(Note 3)			2.7		5.5	V	
		Normal mode		$V_{DD} = 3V$		2.2		mA	
Quiescent Supply Current Per Amplifier	ID	Normal mode		$V_{DD} = 5V$		2.5	4.4	IIIA	
		Shutdown mode	(SHDN =	V _{SS}) (Note 2)		0.01	1.0	μA	
Input Offset Voltage	Vos	$T_A = +25^{\circ}C$				±70	±350	μV	
Input Onset Voltage	VOS	$T_A = -40^{\circ}C \text{ to } +1$	25°C				±750	μv	
Input Offset Voltage Tempco	TCVOS					±0.3	±6	µV/°C	
Input Bias Current	Ι _Β	(Note 4)				±1	±150	pА	
Input Offset Current	los	(Note 4)				±1	±150	рА	
Differential Input Resistance	R _{IN}					1000		GΩ	
Input Common-Mode Voltage	Vсм	Guaranteed by	TA = +	25°C	-0.2		V _{DD} - 1.6	i v	
Range	V CM	CMRR Test	TA = -4	40°C to +125°C	-0.1		V _{DD} - 1.7	v	
Occurrent Maria Deitartian Detie	CMRR	(V _{SS} - 0.2V) ≤ V _{CM} ≤ (V _{DD} - 1.6V)	T _A = +	25°C	90	115		-10	
Common-Mode Rejection Ratio		$(V_{SS} - 0.1V) \le V_{CM} \le (V_{DD} - 1.7V)$	T _A = -4	40°C to +125°C	90			dB	
Power-Supply Rejection Ratio	PSRR	$V_{DD} = 2.7 \text{ to } 5.5 \text{V}$	/		90	120		dB	
		$R_L = 10k\Omega$ to V _{DD} /2; V _{OUT} = 100mV to (V _{DD} - 125mV)		90	120				
Large-Signal Voltage Gain	Avol	$R_L = 1k\Omega$ to $V_{DD}/2$; $V_{OUT} = 200mV$ to ($V_{DD} - 250mV$)		85	110		dB		
		$\label{eq:RL} \begin{array}{l} R_{L} = 500\Omega \text{ to V}_{DD}/2; \\ V_{OUT} = 350 \text{mV to (V}_{DD} \text{ - } 500 \text{mV}) \end{array}$		85	110				

DC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = +5V, V_{SS} = 0V, V_{CM} = 0V, V_{OUT} = V_{DD}/2, R_L \text{ tied to } V_{DD}/2, \overline{SHDN} = V_{DD}, T_A = -40^{\circ}C \text{ to } +125^{\circ}C, \text{ unless otherwise noted.}$ Typical values are at $T_A = +25^{\circ}C.$ (Notes 1, 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	МАХ	UNITS
		$ V_{IN+} - V_{IN-} \ge 10 \text{mV},$	V _{DD} - V _{OH}		10	45	
		$R_L = 10 k\Omega$ to $V_{DD}/2$	V _{OL} - V _{SS}		10	40	
Output Voltage Swing	Vour	$ V_{IN+} - V_{IN-} \ge 10 \text{mV},$	V _{DD} - V _{OH}		80	200	mV
Output voltage Swing	Vout	$R_L = 1k\Omega$ to $V_{DD}/2$	V _{OL} - V _{SS}		50	150	IIIV
		$ V_{IN+} - V_{IN-} \ge 10 \text{mV},$	V _{DD} - V _{OH}		100	300	
		$R_L = 500\Omega$ to $V_{DD}/2$	V _{OL} - V _{SS}		80	250	
Output Short-Circuit Current	Isc				48		mA
Output Leakage Current	ILEAK	Shutdown mode (\overline{SHDN} V _{OUT} = V _{SS} to V _{DD}	$= V_{SS}),$		±0.001	±1.0	μA
SHDN Logic Low	VIL					$0.3 \times V_{DD}$	V
SHDN Logic High	VIH			$0.7 \times V_{DD}$			V
SHDN Input Current		$\overline{\text{SHDN}} = V_{SS} \text{ to } V_{DD}$			0.01	1	μA
Input Capacitance	CIN				10		рF

AC ELECTRICAL CHARACTERISTICS

 $(V_{DD} = +5V, V_{SS} = 0V, V_{CM} = 0V, V_{OUT} = V_{DD}/2, R_L \text{ tied to } V_{DD}/2, \overline{SHDN} = V_{DD}, T_A = +25^{\circ}C.)$

PARAMETER	SYMBOL	CONDITI	ONS	MIN	TYP	MAX	UNITS
		MAX4475-MAX4478 A _V = +1V/V			10		N 41 1-
Gain-Bandwidth Product	GBWP	MAX4488/MAX4489	$A_V = +5V/V$		42		MHz
Slew Rate	SR	MAX4475-MAX4478	$A_V = +1V/V$		3		Mue
Siew Rate	эп	MAX4488/MAX4489	$A_V = +5V/V$		10		V/µs
Full-Power Bandwidth (Note 5)		MAX4475-MAX4478	$A_V = +1V/V$		0.4		MHz
Full-Power Bandwidth (Note 5)		MAX4488/MAX4489	$A_V = +5V/V$		1.25		IVIHZ
Peak-to-Peak Input Noise Voltage	e _{n(P-P)}	f = 0.1Hz to 10Hz			260		nV _{P-P}
		f = 10Hz		21			
Input Voltage-Noise Density	en	f = 1kHz			4.5		nV/√Hz
		f = 30kHz			3.5		
Input Current-Noise Density	in	f = 1kHz			0.5		fA/√Hz
		$V_{OUT} = 2V_{P-P},$ $A_V = +1V/V$	f = 1kHz		0.0002		
		$(MAX4475-MAX4478), R_L = 10k\Omega \text{ to GND}$	f = 20kHz		0.0007		_
Total Harmonic Distortion Plus Noise (Note 6)		$V_{OUT} = 2V_{P-P},$ $A_V = +1V/V$	f = 1kHz		0.0002		
	THD + N	(MAX4475–MAX4478), R _L = 1k Ω to GND	f = 20kHz		0.001		%
		$V_{OUT} = 2V_{P-P},$ $A_V = +5V/V$	f = 1kHz		0.0004		
		(MAX4488/MAX4489), R _L = 10k Ω to GND	f = 20kHz		0.0006		

AC ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = +5V, V_{SS} = 0V, V_{CM} = 0V, V_{OUT} = V_{DD}/2, R_L \text{ tied to } V_{DD}/2, \overline{SHDN} = V_{DD}, T_A = +25^{\circ}C.)$

PARAMETER	SYMBOL	CONDITIONS		MIN	ΤΥΡ Ν	ЛАХ	UNITS
Total Harmonic Distortion Plus Noise (Note 6)		$V_{OUT} = 2V_{P-P},$ $A_V = +5V/V$	f = 1kHz	0.	.0005		0/
	THD + N	(MAX4488/MAX4489), R _L = 1k Ω to GND	f = 20kHz	0).008		%
Capacitive-Load Stability		No sustained oscillations			200		pF
Gain Margin	GM				12		dB
	ΦM	MAX4475–MAX4478, A _V = +1V/V			70		degrees
Phase Margin	ΨΙνι	MAX4488/MAX4489, A _V = +5V/V			80		uegrees
Settling Time		To 0.01%, V _{OUT} = 2V st	To 0.01%, V _{OUT} = 2V step		2		μs
Delay Time to Shutdown	t _{SH}				1.5		μs
Enable Delay Time from Shutdown	t _{EN}	$V_{OUT} = 2.5V, V_{OUT}$ settles to 0.1%			10		μs
Power-Up Delay Time		$V_{DD} = 0$ to 5V step, V_{OL}	T stable to 0.1%		13		μs

Note 1: All devices are 100% tested at $T_A = +25^{\circ}C$. Limits over temperature are guaranteed by design.

Note 2: SHDN is available on the MAX4475/MAX4488 only.

Note 3: Guaranteed by the PSRR test.

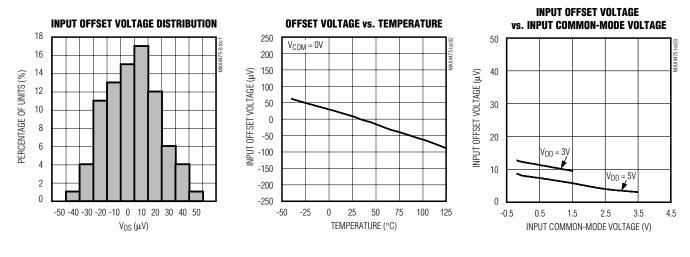
Note 4: Guaranteed by design.

Note 5: Full-power bandwidth for unity-gain stable devices (MAX4475–MAX4478) is measured in a closed-loop gain of +2V/V to accommodate the input voltage range, $V_{OUT} = 4V_{P-P}$.

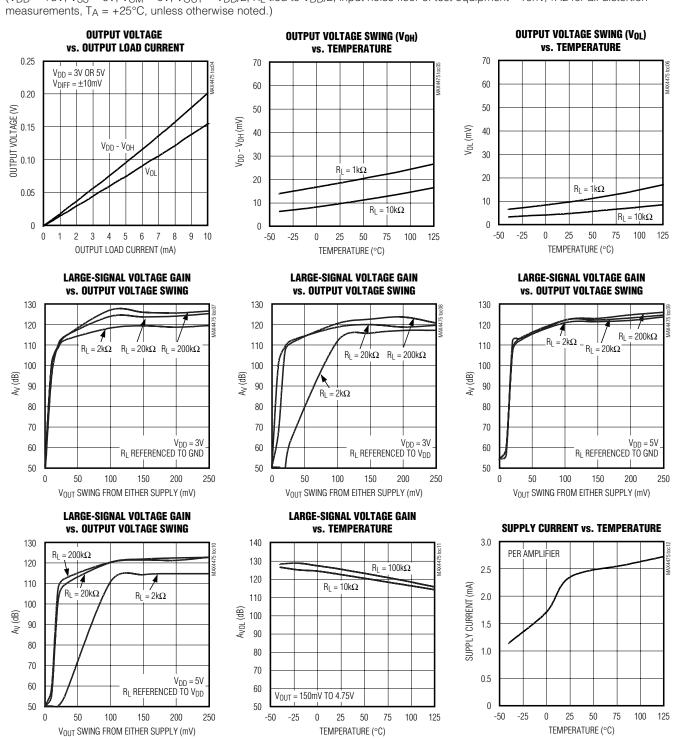
Note 6: Lowpass-filter bandwidth is 22kHz for f = 1kHz and 80kHz for f = 20kHz. Noise floor of test equipment = $10nV/\sqrt{Hz}$.

Typical Operating Characteristics

 $(V_{DD} = +5V, V_{SS} = 0V, V_{CM} = 0V, V_{OUT} = V_{DD}/2, R_L \text{ tied to } V_{DD}/2, \text{ input noise floor of test equipment } = 10nV/\sqrt{Hz} \text{ for all distortion measurements}, T_A = +25^{\circ}C, \text{ unless otherwise noted.})$



///XI/M



Typical Operating Characteristics (continued)

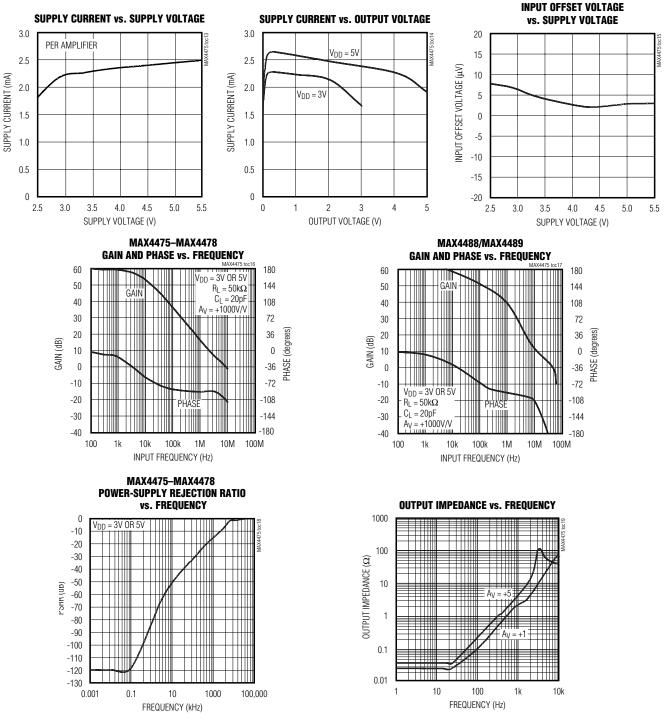
(V_{DD} = +5V, V_{SS} = 0V, V_{CM} = 0V, V_{OUT} = V_{DD}/2, R_L tied to V_{DD}/2, input noise floor of test equipment =10nV/√Hz for all distortion

MIXIM

MAX4475-MAX4478/MAX4488/MAX4489

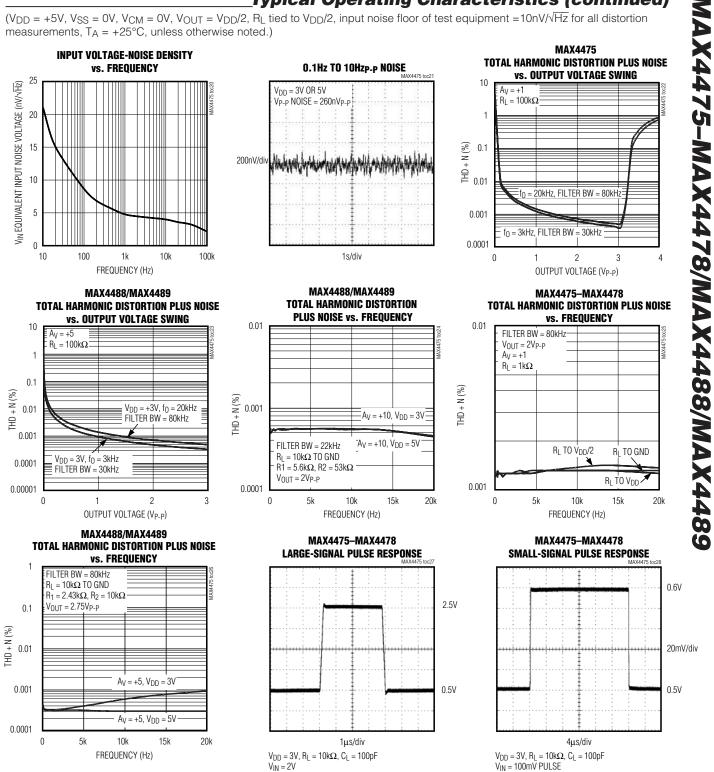
Typical Operating Characteristics (continued)

 $(V_{DD} = +5V, V_{SS} = 0V, V_{CM} = 0V, V_{OUT} = V_{DD}/2, R_L \text{ tied to } V_{DD}/2, \text{ input noise floor of test equipment } = 10 \text{nV}/\sqrt{\text{Hz}} \text{ for all distortion measurements}, T_A = +25^{\circ}\text{C}, \text{ unless otherwise noted.}$



Typical Operating Characteristics (continued)

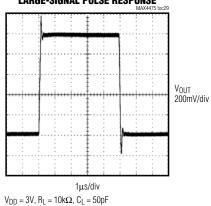
7



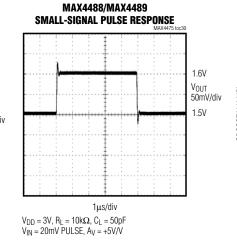
Typical Operating Characteristics (continued)

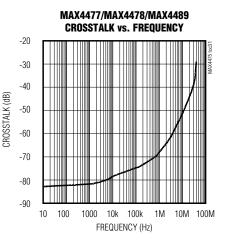
 $(V_{DD} = +5V, V_{SS} = 0V, V_{CM} = 0V, V_{OUT} = V_{DD}/2, R_L$ tied to $V_{DD}/2$, input noise floor of test equipment = $10nV/\sqrt{Hz}$ for all distortion measurements, $T_A = +25^{\circ}C$, unless otherwise noted.)

MAX4488/MAX4489 LARGE-SIGNAL PULSE RESPONSE



$$[\]label{eq:VDD} \begin{split} V_{DD} &= 3V, \, R_L = 10 k \Omega, \, C_L = 50 p F \\ V_{IN} &= 20 m V \, PULSE, \, A_V = +5 V / V \end{split}$$





Pin Description

		PIN				
MAX4475/ MAX4488	MAX4475/ MAX4488	MAX4476	MAX4477/ MAX4489 MAX4478		NAME	FUNCTION
SOT23/TDFN	SO/µMAX	SOT23/TDFN	SO/µMAX	SO/TSSOP		
1	6	1	1, 7	1, 7, 8, 14	OUT, OUTA, OUTB, OUTC, OUTD	Amplifier Output
2	4	2	4	11	V _{SS}	Negative Supply. Connect to ground for single- supply operation
3	3	3	3, 5	3, 5, 10, 12	IN+, INA+, INB+, INC+, IND+	Noninverting Amplifier Input
4	2	4	2, 6	2, 6, 9, 13	IN-, INA-, INB-, INC-, IND-	Inverting Amplifier Input
6	7	6	8	4	V _{DD}	Positive Supply
5	8	_	_	_	SHDN	Shutdown Input. Connect to V_{DD} for normal operation (amplifier(s) enabled).
_	1, 5	5	_	—	N.C.	No Connection. Not internally connected.
EP (TDFN only)	_	EP (TDFN only)	—	_	EP	Exposed Paddle. Connect to V _{SS} .



Detailed Description

The MAX4475–MAX4478/MAX4488/MAX4489 singlesupply operational amplifiers feature ultra-low noise and distortion. Their low distortion and low noise make them ideal for use as preamplifiers in wide dynamicrange applications, such as 16-bit analog-to-digital converters (see *Typical Operating Circuit*). Their highinput impedance and low noise are also useful for signal conditioning of high-impedance sources, such as piezoelectric transducers.

These devices have true rail-to-rail ouput operation, drive loads as low as $1 k \Omega$ while maintining DC accuracy, and can drive capactive loads up to 200pF without oscillation. The input common-mode voltage range extends from (V_DD - 1.6V) to 200mV below the negative rail. The push-pull output stage maintains excellent DC characteristics, while delivering up to $\pm 5 mA$ of current.

The MAX4475–MAX4478 are unity-gain stable, while the MAX4488/MAX4489 have a higher slew rate and are stable for gains \geq 5V/V. The MAX4475/MAX4488 feature a low-power shutdown mode, which reduces the supply current to 0.01µA and disables the outputs.

Low Distortion

Many factors can affect the noise and distortion that the device contributes to the input signal. The following guidelines offer valuable information on the impact of design choices on Total Harmonic Distortion (THD).

Choosing proper feedback and gain resistor values for a particular application can be a very important factor in reducing THD. In general, the smaller the closedloop gain, the smaller the THD generated, especially when driving heavy resistive loads. The THD of the part normally increases at approximately 20dB per decade, as a function of frequency. Operating the device near or above the full-power bandwidth significantly degrades distortion.

Referencing the load to either supply also improves the part's distortion performance, because only one of the MOSFETs of the push-pull output stage drives the output. Referencing the load to midsupply increases the part's distortion for a given load and feedback setting. (See the Total Harmonic Distortion vs. Frequency graph in the *Typical Operating Characteristics*.)

For gains \geq 5V/V, the decompensated devices MAX4488/MAX4489 deliver the best distortion performance, since they have a higher slew rate and provide a higher amount of loop gain for a given closed-loop gain setting. Capacitive loads below 100pF do not significantly affect distortion results. Distortion performance is relatively constant over supply voltages.

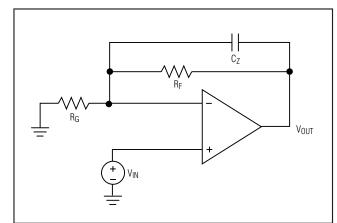


Figure 1. Adding Feed-Forward Compensation

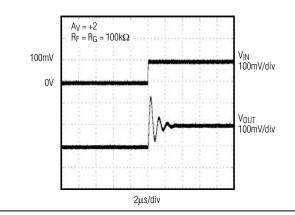


Figure 2a. Pulse Response with No Feed-Forward Compensation

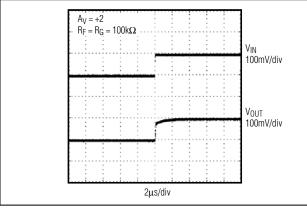


Figure 2b. Pulse Response with 10pF Feed-Forward Compensation

Low Noise

The amplifier's input-referred noise-voltage density is dominated by flicker noise at lower frequencies, and by thermal noise at higher frequencies. Because the thermal noise contribution is affected by the parallel combination of the feedback resistive network (R_F II R_G, Figure 1), these resistors should be reduced in cases where the system bandwidth is large and thermal noise is dominant. This noise contribution factor decreases, however, with increasing gain settings.

For example, the input noise-voltage density of the circuit with R_F = 100k Ω , R_G = 11k Ω (Av = +5V/V) is $e_{\rm I}$ = 14nV/ $\sqrt{\rm Hz}$, $e_{\rm I}$ can be reduced to 6nV/ $\sqrt{\rm Hz}$ by choosing R_F = 10k Ω , R_G = 1.1k Ω (Av = +5V/V), at the expense of greater current consumption and potentially higher distortion. For a gain of 100V/V with R_F = 100k Ω , R_G = 1.1k Ω , the $e_{\rm I}$ is still a low 6nV/ $\sqrt{\rm Hz}$.

Using a Feed-Forward Compensation Capacitor, Cz

The amplifier's input capacitance is 10pF. If the resistance seen by the inverting input is large (feedback network), this can introduce a pole within the amplifier's bandwidth resulting in reduced phase margin. Compensate the reduced phase margin by introducing a feed-forward capacitor (Cz) between the inverting input and the output (Figure 1). This effectively cancels the pole from the inverting input of the amplifier. Choose the value of Cz as follows:

 $C_{Z} = 10 \times (R_{F} / R_{G}) [pF]$

In the unity-gain stable MAX4475–MAX4478, the use of a proper Cz is most important for $A_V = +2V/V$, and $A_V = -1V/V$. In the decompensated MAX4488/MAX4489, Cz is most important for $A_V = +10V/V$. Figures 2a and 2b show transient response both with and without Cz.

Using a slightly smaller Cz than suggested by the formula above achieves a higher bandwidth at the expense of reduced phase and gain margin. As a general guideline, consider using Cz for cases where RG II RF is greater than 20k Ω (MAX4475–MAX4478) or greater than 5k Ω (MAX4488/MAX4489).

Applications Information

The MAX4475–MAX4478/MAX4488/MAX4489 combine good driving capability with ground-sensing input and rail-to-rail output operation. With their low distortion and low noise, they are ideal for use in ADC buffers, medical instrumentation systems and other noise-sensitive applications.

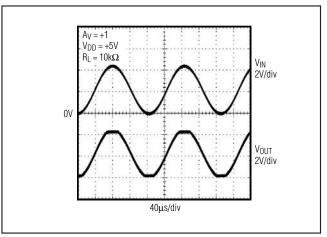


Figure 3. Overdriven Input Showing No Phase Reversal

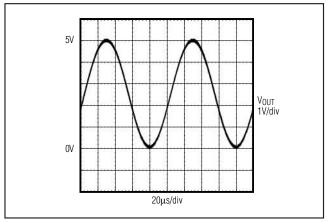


Figure 4. Rail-to-Rail Output Operation

Ground-Sensing and Rail-to-Rail Outputs

The common-mode input range of these devices extends below ground, and offers excellent commonmode rejection. These devices are guaranteed not to undergo phase reversal when the input is overdriven (Figure 3).

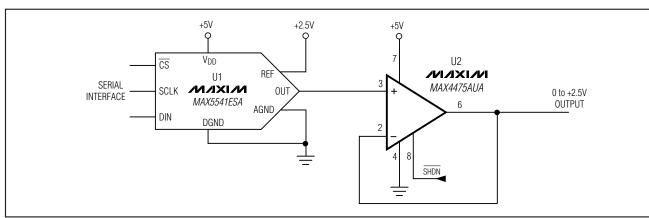
Figure 4 showcases the true rail-to-rail output operation of the amplifier, configured with $A_V = 5V/V$. The output swings to within 8mV of the supplies with a 10k Ω load, making the devices ideal in low-supply voltage applications.

Power Supplies and Layout

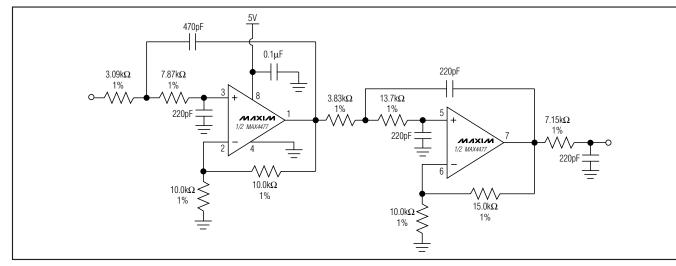
The MAX4475–MAX4478/MAX4488/MAX4489 operate from a single +2.7V to +5.5V power supply or from dual supplies of $\pm 1.35V$ to $\pm 2.75V$. For single-supply operation, bypass the power supply with a 0.1µF ceramic



Typical Application Circuit



_Typical Operating Circuit



capacitor placed close to the V_{DD} pin. If operating from dual supplies, bypass each supply to ground.

Good layout improves performance by decreasing the amount of stray capacitance and noise at the op amp's inputs and output. To decrease stray capacitance, minimize PC board trace lengths and resistor leads, and place external components close to the op amp's pins.

Typical Application Circuit

The *Typical Application Circuit* shows the single MAX4475 configured as an output buffer for the MAX5541 16-bit DAC. Because the MAX5541 has an unbuffered voltage output, the input bias current of the op amp used must be less than 6nA to maintain 16-bit accuracy. The MAX4475 has an input bias current of only 150pA (max), virtually eliminating this as a source

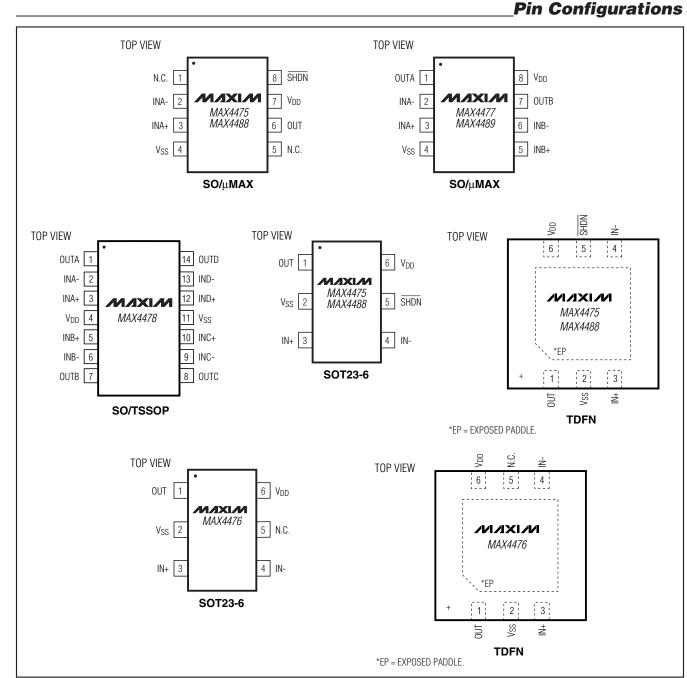
of error. In addition, the MAX4475 has excellent openloop gain and common-mode rejection, making this an excellent ouput buffer amplifier.

DC-Accurate Lowpass Filter

The MAX4475–MAX4478/MAX4488/MAX4489 offer a unique combination of low noise, wide bandwidth, and high gain, making them an excellent choice for active filters up to 1MHz. The *Typical Operating Circuit* shows the dual MAX4477 configured as a 5th order Chebyschev filter with a cutoff frequency of 100kHz. The circuit is implemented in the Sallen-Key topology, making this a DC-accurate filter.

MAX4475-MAX4478/MAX4488/MAX4489





MAX4475-MAX4478/MAX4488/MAX4489

Ordering Information (continued)

PART	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX4476AUT+T	-40°C to +125°C	6 SOT23	AAZX
MAX4476ATT+T	-40°C to +125°C	6 TDFN-EP*	+ADF
MAX4477AUA+	-40°C to +125°C	8 µMAX	_
MAX4477AUA+	-40°C to +125°C	8 µMAX	—
MAX4477ASA+	-40°C to +125°C	8 SO	
MAX4478AUD+	-40°C to +125°C	14 TSSOP	—
MAX4478AUD/V+	-40°C to +125°C	14 TSSOP	
MAX4478ASD+	-40°C to +125°C	14 SO	_
MAX4488AUT+T	-40°C to +125°C	6 SOT23	AAZW
MAX4488AUA+	-40°C to +125°C	8 µMAX	—
MAX4488ASA+	-40°C to +125°C	8 SO	
MAX4488ATT+T	-40°C to +125°C	6 TDFN-EP*	+ADE
MAX4489AUA+	-40°C to +125°C	8 µMAX	_

+Denotes a lead(Pb)-free/RoHS-compliant package.

*EP = Exposed pad (connect to V_{SS}).

N denotes an automotive qualified part.

T = Tape and reel.

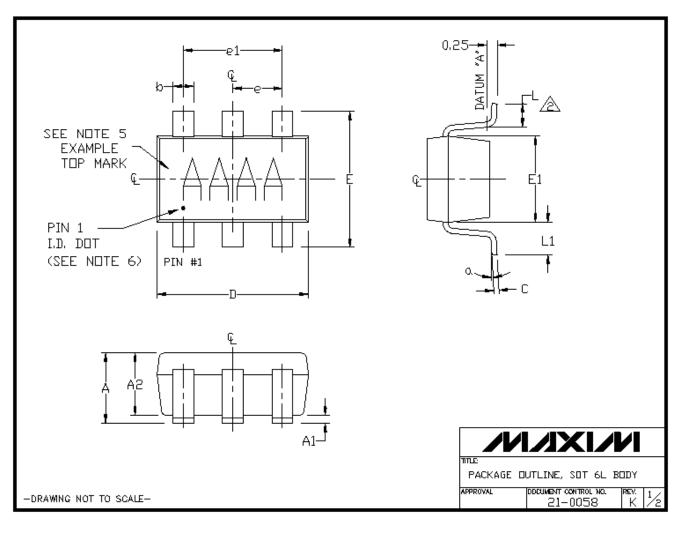
Chip Information

PROCESS: BICMOS

Package Information

For the latest package outline information and land patterns, go to <u>www.maxim-ic.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
6 SOT23	U6F-6	<u>21-0058</u>	<u>90-0175</u>
8 µMAX	U8-1	<u>21-0036</u>	<u>90-0092</u>
14 TSSOP	U14-2	<u>21-0066</u>	<u>90-0117</u>
8 SO	S8-4	<u>21-0041</u>	
14 SO	S14-4	<u>21-0041</u>	
6 TDFN-EP	T633-2	<u>21-0137</u>	<u>90-0058</u>



M/IXI/M

Package Information (continued)

U6-4, U6CN-2,

MAX

1,45

0.15

1.30

0.50

0,20

3.00

3.00 1.75

0,60

°01

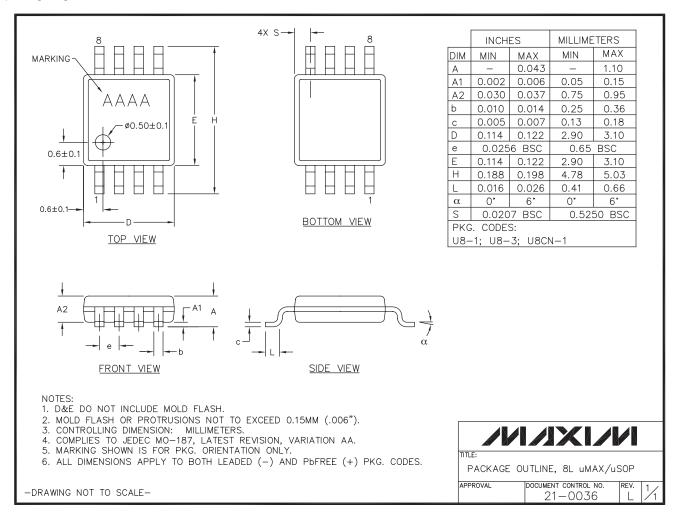
For the latest package outline information and land patterns, go to www.maxim-ic.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

-DRA	WNG NOT TO SCALE-	APPROVAL	DC	кимент сонтво 21-005
	ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND P&FREE (+) PKG. CODES.	TITLE PACKA	GE OUT	
	MARKING IS FOR PACKAGE DRIENTATION REFERENCE DNLY.			
	LEAD TO BE COPLANAR WITHIN 0.1mm. NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY.			
8.	SOLDER THICKNESS MEASURED AT FLAT SECTION OF LEAD BETWEEN 0.08mm AND 0.15mm FROM LEADTIP.			-
7,	MEETS JEDEC MO178, VARIATION AB.	U6-1, U	6-2, U	6—4, UGCN- 3, UGFH-6
6.	PIN 1 I.D. DOT IS 0.3mm ∅ MIN. LOCATED ABO∨E PIN 1.	A PKG CD	0*	2.5*
	RIGHT. (SEE EXAMPLE TOP MARK)	e		D.95 BSC
5,	PIN 1 IS LOWER LEFT PIN WHEN READING TOP MARK FROM LEFT TO	L1 e1		0.60 REF.
4.	PACKAGE DUTLINE INCLUSIVE OF SOLDER PLATING.		0.35	0.45
З,	PACKAGE DUTLINE EXCLUSIVE OF MOLD FLASH & METAL BURR. MOLD FLASH, PROTRUSION OR METAL BURR SHOULD NOT EXCEED 0.25mm.	 Е Е1	2.60	2.60
	LEAD SURFACE.	ь С D	0.35 0.08 2.80	0.40 0.15 2.90
A	FOOT LENGTH MEASURED AT INTERCEPT POINT BETWEEN DATUM A &	A2	0.90	1.10
1.	ALL DIMENSIONS ARE IN MILLIMETERS.	A1	0.00	0.05
ND	TES	SYMBOL	<u>MIN</u> 0,90	1.25



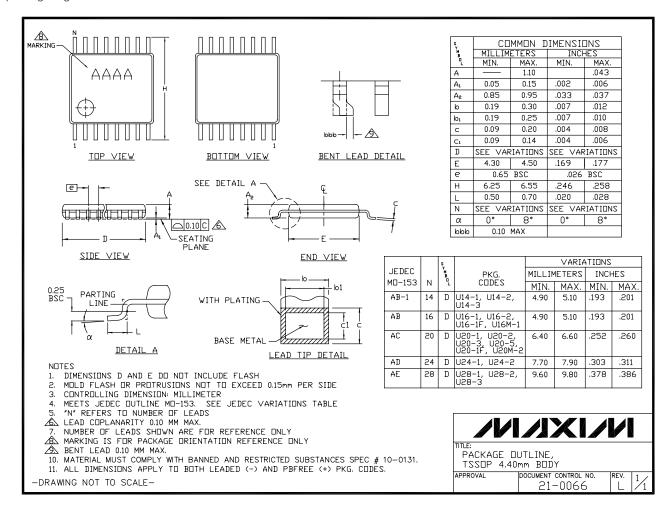
Package Information (continued)

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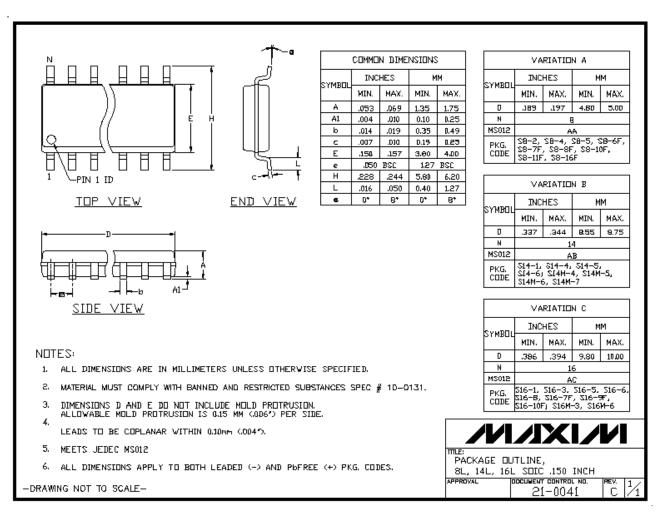
Package Information (continued)

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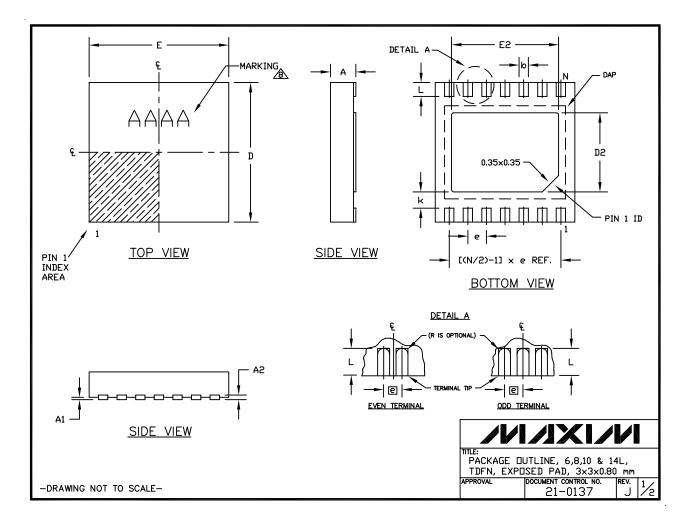
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Package Information (continued)

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COMMON DIMENSIONS			PACKAGE VARIATIONS								
SYMBOL	MIN.	MAX.		PKG. CODE	Ν	D2	E2	е	JEDEC SPEC	b	[(N/2)-1] x e
А	0.70	0.80		T633-2	6	1.50±0.10	2.30±0.10	0.95 BSC	MO229 / WEEA	0.40±0.05	1.90 REF
D	2.90	3.10		T833-2	8	1.50±0.10	2.30±0.10	0.65 BSC	MO229 / WEEC	0.30±0.05	1.95 REF
E	2.90	3.10		T833-3	8	1.50±0.10	2.30±0.10	0.65 BSC	MO229 / WEEC	0.30±0.05	1.95 REF
A1	0.00	0.05		T1033-1	10	1.50±0.10	2.30±0.10	0.50 BSC	MO229 / WEED-3	0.25±0.05	2.00 REF
L	0.20	0.40		T1033MK-1	10	1.50±0.10	2.30±0.10	0.50 BSC	MO229 / WEED-3	0.25±0.05	2.00 REF
k	0.25	MIN.		T1033-2	10	1.50±0.10	2.30±0.10	0.50 BSC	MO229 / WEED-3	0.25±0.05	2.00 REF
A2	0.20	REF.		T1433-1	14	1.70±0.10	2.30±0.10	0.40 BSC		0.20±0.05	2.40 REF
				T1433-2	14	1.70±0.10	2.30±0.10	0.40 BSC		0.20±0.05	2.40 REF
				T1433-3F	14	1.70±0.10	2.30±0.10	0.40 BSC		0.20±0.05	2.40 REF
NOTES.											
2. COPL 3. WARP 4. PACK 5. DRAW 6. "N" I 7. NUME 8. MARK	ANARITY AGE SH AGE LEI ING CO S THE SER OF ING IS	SHALL NGTH/P/ NFORMS TOTAL N LEADS FOR PA	NOT EXC EXCEED CKAGE V TO JED UMBER (SHOWN A CKAGE (I. ANGLES IN CEED 0.08 m O 0.10 mm. WIDTH ARE CO EC M0229, E DF LEADS. ARE FOR REF RIENTATION R BOTH LEADED	m. DNSID XCEP EREN	ERED AS S T DIMENSIO CE ONLY. ENCE ONLY	NS "D2" AN	ID "E2", AN	С(S). ND T1433-1 & T [.]	1433–2.	

Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
4	12/09	Added lead-free designations and an automotive part to the Ordering Information and added input current spec in Absolute Maximum Ratings section	1, 2, 13
5	7/10	Added /V designation to the MAX4475 product and soldering temperature	1, 2

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