

DESCRIPTION

The MP1917A is a high-frequency, half-bridge, N-channel power MOSFET driver. Its low-side and high-side driver channels are controlled independently and matched with less than 5ns of time delay. Under-voltage lockout (UVLO) on both the high-side and low-side supplies forces the outputs low in the event that the supply is insufficient. The integrated bootstrap diode reduces the external component count.

The MP1917A is available in a QFN-10 (4mmx4mm) package.

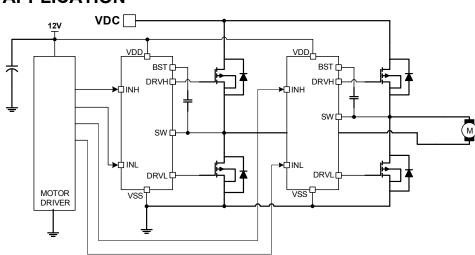
FEATURES

- Drives an N-Channel MOSFET Half-Bridge
- 115V Bootstrap Voltage Range
- On-Chip Bootstrap Diode
- Typical Propagation Delay of 20ns
- Gate Driver Matching of Less than 5ns
- Drives a 2.2nF Load with 15ns of Rise Time and 12ns of Fall Time at 12V VDD
- TTL-Compatible Input
- Quiescent Current of Less than 150µA
- UVLO for Both High-Side and Low-Side Gate Drivers
- QFN-10 (4mmx4mm) Package

APPLICATIONS

- Isolated Brick Power
- Telecom Half-Bridge Power Supplies
- Avionics DC/DC Converters
- Two-Switch Forward Converters
- Active Clamp Forward Converters

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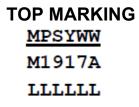
TYPICAL APPLICATION



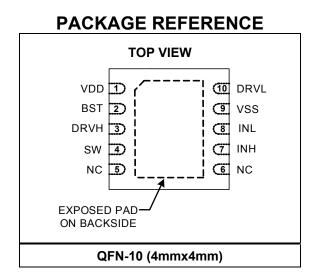
ORDERING INFORMATION

Part Number	Package	Top Marking
MP1917AGR*	QFN-10 (4mmx4mm)	See Below

* For Tape & Reel, add suffix –Z (e.g. MP1917AGR–Z).



MPS: MPS prefix Y: Year code WW: Week code M1917A: First six digits of the part number LLL: Lot number





ABSOLUTE MAXIMUM RATINGS (1)

Supply voltage (V _{DD})	0.3V to 18V
SW voltage (V _{SW})	
1V (-18	V for < 100nS) to 105V
BST voltage (V _{BST})	0.3V to 115V
BST to SW	-0.3V to 18V
DRVH to SW	
0.3V (-2V for < 100nS	
DRVL to VSS	
0.3V (-2V for <	100nS) to (V _{DD} + 0.3V)
All other pins	0.3V to (V _{DD} + 0.3V)
Continuous power dissipa	
QFN-10 (4mmx4mm)	
Junction temperature	150°C
Lead temperature	260°C
Storage temperature	

Recommended Operating Conditions ⁽³⁾

Supply voltage (V _{DD})	
•••••••	1.0V to 98V
SW slew rate	<50V/ns
Operating junction te	emp (T _J)40°C to 125°C

Thermal Resistance $^{(4)}$ θ_{JA} θ_{JC}

QFN-10 (4mmx4mm)......47......7....°C/W

NOTES:

1) Exceeding these ratings may damage the device.

- 2) The maximum allowable power dissipation is a function of the maximum junction temperature T_J (MAX), the junction-toambient thermal resistance θ_{JA} , and the ambient temperature T_A. The maximum allowable continuous power dissipation at any ambient temperature is calculated by P_D (MAX) = (T_J (MAX) - T_A) / θ_{JA} . Exceeding the maximum allowable power dissipation produces an excessive die temperature, causing the regulator to go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- 3) The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB.



ELECTRICAL CHARACTERISTICS

 V_{DD} = V_{BST} - V_{SW} = 12V, V_{SS} = V_{SW} = 0V, no load at DRVH and DRVL, T_A = +25°C, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Supply Currents						
VDD quiescent current	IDDQ	INL = INH = 0		100	150	μA
VDD operating current	Iddo	fsw = 500kHz		9		mA
Floating driver quiescent current	IBSTQ	INL = INH = 0		60	90	μA
Floating driver operating current	IBSTO	fsw = 500kHz		7.5		mA
Leakage current	Ilk	BST = SW = 100V		0.05	1	μA
Inputs			·			•
INL/INH high				2	2.4	V
INL/INH low			1	1.4		V
INL/INH internal pull-down	D			185		kΩ
resistance	RIN			COL		K12
Under-Voltage Protection						
VDD rising threshold	V_{DDR}		6.4	6.8	7.2	V
VDD hysteresis	Vddh			0.5		V
BST-SW rising threshold	VBSTR		6.1	6.5	6.9	V
BST-SW hysteresis	VBSTH			0.5		V
Bootstrap Diode						
Bootstrap diode VF at 100µA	V _{F1}			0.5		V
Bootstrap diode VF at 100mA	V _{F2}			0.95		V
Bootstrap diode dynamic R	R _D	At 100mA		2.5		Ω
Low-Side Gate Driver						
Low-level output voltage	Voll	I ₀ = 100mA		0.1		V
High-level output voltage to rail	VOHL	I ₀ = -100mA		0.36		V
Que en en en et (5)	IOHL	$V_{DRVL} = 0V, V_{DD} = 12V$		3		Α
Source current ⁽⁵⁾		$V_{DRVL} = 0V, V_{DD} = 16V$		4.7		Α
Ciple current (5)	IOLL	$V_{DRVL} = V_{DD} = 12V$		4.5		Α
Sink current ⁽⁵⁾		$V_{DRVL} = V_{DD} = 16V$		6		А
Floating Gate Driver			·			•
Low-level output voltage	V _{OLH}	I ₀ = 100mA		0.1		V
High-level output voltage to rail	Vонн	I _O = -100mA		0.32	1	V
	I _{ОНН}	$V_{DRVH} = 0V, V_{DD} = 12V$		2.6		А
Source current ⁽⁵⁾		$V_{DRVH} = 0V, V_{DD} = 16V$		4		Α
Olight groups of (5)	Iolh	$V_{DRVH} = V_{DD} = 12V$		4.5		Α
Sink current ⁽⁵⁾		$V_{DRVH} = V_{DD} = 16V$		5.9		Α



ELECTRICAL CHARACTERISTICS (continued)

 $V_{DD} = V_{BST} - V_{SW} = 12V$, $V_{SS} = V_{SW} = 0V$, no load at DRVH and DRVL, $T_A = +25^{\circ}C$, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Switching Specification – Low-S	Switching Specification – Low-Side Gate Driver					
Turn-off propagation delay INL falling to DRVL falling	TDLFF			20		ns
Turn-on propagation delay INL rising to DRVL rising	T _{DLRR}			20		
DRVL rise time		C _L = 2.2nF		15		ns
DRVL fall time		C∟ = 2.2nF		15		ns
Switching Specification – Floatin	ng Gate D	river			-	
Turn-off propagation delay INH falling to DRVH falling	TDHFF			20		ns
Turn-on propagation delay INH rising to DRVH rising	TDHRR			20		ns
DRVH rise time		C∟ = 2.2nF		15		ns
DRVH fall time		C∟ = 2.2nF		15		ns
Switching Specification – Match	ing					
Floating driver turn-off to low-side driver turn-on ⁽⁵⁾	T _{MON}			1	5	ns
Low-side driver turn-off to floating driver turn-on ⁽⁵⁾	TMOFF			1	5	ns
Minimum input pulse width that changes the output ⁽⁵⁾	T _{PW}				50	ns
Bootstrap diode turn-on or turn-off time $^{(5)}$	T _{BS}			10		ns
Thermal shutdown				150		С°
Thermal shutdown hysteresis				25		С°

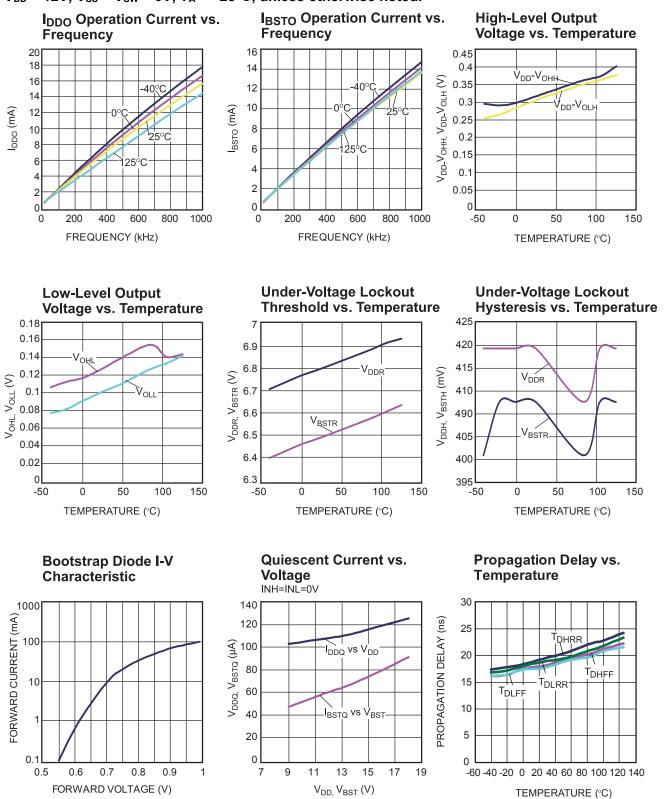
NOTE:

5) Guaranteed by design.



TYPICAL PERFORMANCE CHARACTERISTICS

 V_{DD} = 12V, V_{SS} = V_{SW} = 0V, T_A = +25°C, unless otherwise noted.



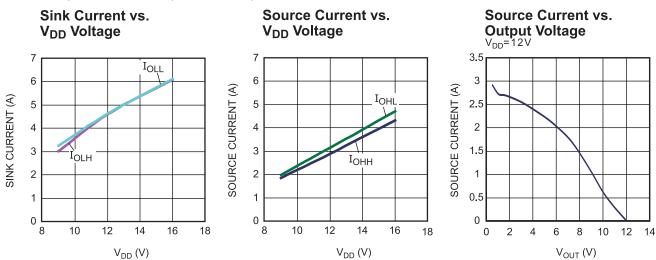
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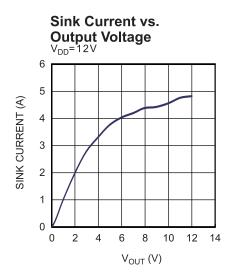
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TYPICAL PERFORMANCE CHARACTERISTICS (continued)

 V_{DD} = 12V, V_{SS} = V_{SW} = 0V, T_A = +25°C, unless otherwise noted.







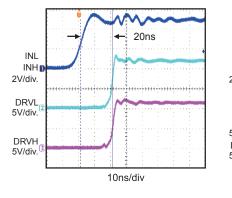
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

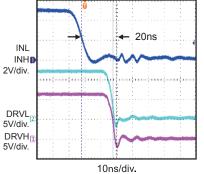
 V_{DD} = 12V, V_{SS} = V_{SW} = 0V, T_A = +25°C, unless otherwise noted.

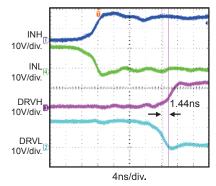
Turn-On Propagation Delay

Turn-Off Propagation Delay

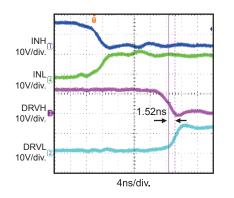
Gate Drive Matching T_{MOFF}





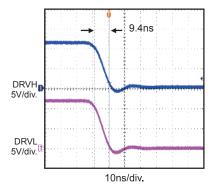


Gate Drive Matching T_{MON}



DRVH DRVL 5V/div.

Drive Fall Time 2.2nF Load



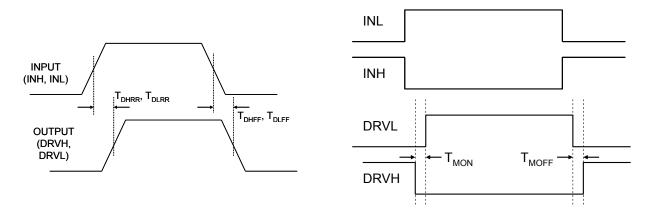


PIN FUNCTIONS

Pin #	Name	Description			
1	VDD	Supply input. VDD supplies power to the internal circuitry. Place a decouplin capacitor on ground close to VDD to ensure a stable and clean supply.			
2	BST	Bootstrap. Positive power supply for the internal floating high-side MOSFET driver. Connect a bypass capacitor between BST and SW.			
3	DRVH	Floating driver output.			
4	SW	Switching node.			
5, 6	NC	No connection.			
7	INH	Control signal input for the floating driver.			
8	INL	Control signal input for the low-side driver.			
9	VSS, exposed pad	Chip ground. Connect the exposed pad to VSS for proper thermal operation.			
10	DRVL	Low-side driver output.			



TIMING DIAGRAM







FUNCTIONAL BLOCK DIAGRAM

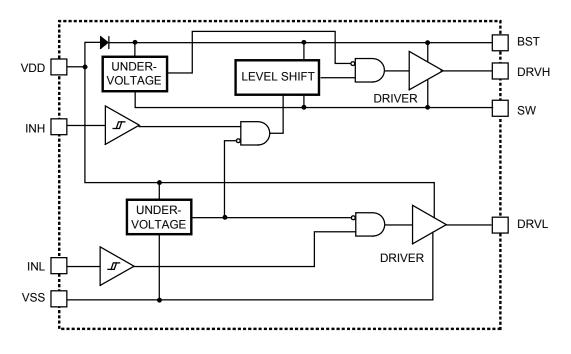


Figure 2: Functional Block Diagram



APPLICATION INFORMATION

The INH and INL input signals can be controlled independently. If both INH and INL control the high-side and low-side MOSFETs of the same bridge, set a sufficient dead time between INH and INL low, and vice versa, to avoid shootthrough (see Figure 3). Dead time is defined as the time interval between INH low and INL low.

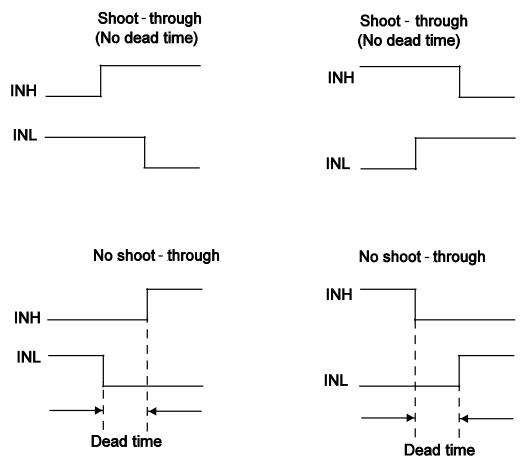


Figure 3: Shoot-Through Timing Diagram



REFERENCE DESIGN CIRCUITS

Half-Bridge Converter

The MP1917A drives the MOSFETS with alternating signals with dead time in half-bridge converter topology. The input voltage can rise up

to 100V with the alternating signals INT and INL coming from the PWM controller (see Figure 4).

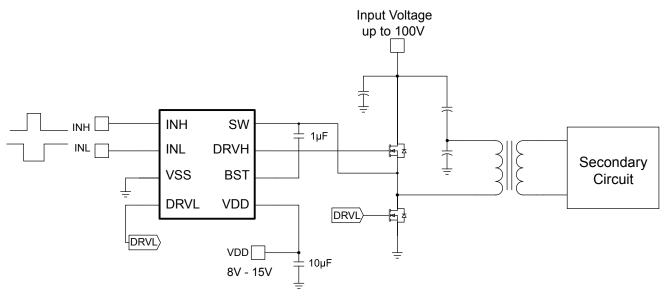
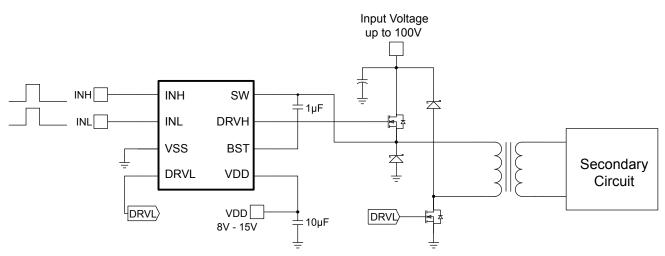


Figure 4: Half-Bridge Converter

Two-Switch Forward Converter

In two-switch forward converter topology, both MOSFETs are turned on and off simultaneously. The INH and INL input signals come from a PWM controller that senses the output voltage and output current during current-mode control.

The Schottky diodes clamp the reverse swing of the power transformer, and must be rated for the input voltage. The input voltage can rise up to 100V (see Figure 5).







Active Clamp Forward Converter

In active clamp forward converter topology, the MP1917A drives the MOSFETs with alternating signals. The high-side MOSFET, in conjunction with C_{reset}, is used to reset the power transformer with minimal power loss.

This topology is optimal for running at duty cycles exceeding 50%. The device may not be able to run at 100V under this topology (see Figure 6).

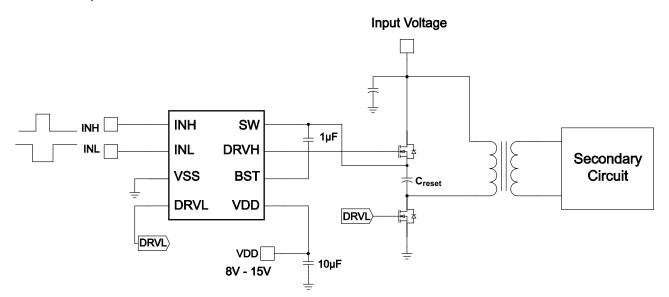
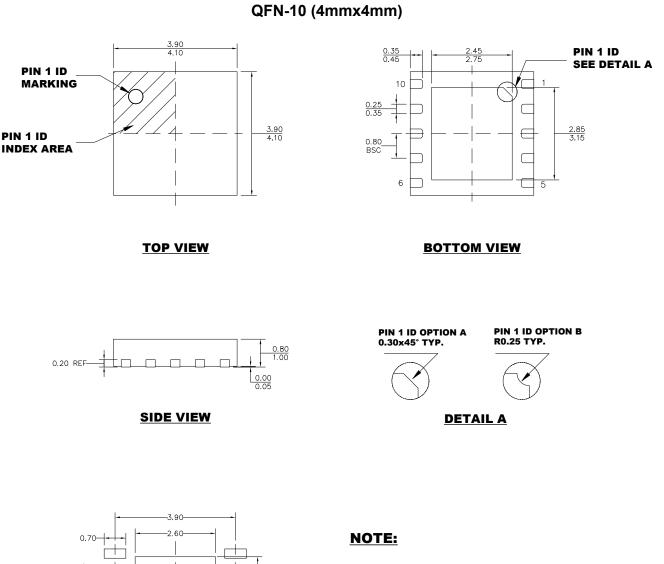


Figure 6: Active Clamp Forward Converter



PACKAGE INFORMATION



 ALL DIMENSIONS ARE IN MILLIMETERS.
EXPOSED PADDLE SIZE DOES NOT INCLUDE MOLD FLASH.
LEAD COPLANARITY SHALL BE 0.10 MILLIMETERS MAX.
JEDEC REFERENCE IS MO-220.
DRAWING IS NOT TO SCALE.

RECOMMENDED LAND PATTERN

0.80

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