







SNAS818B - JULY 2021 - REVISED MAY 2022



### CDCDB800 DB800ZL-Compliant 8-Output Clock Buffer for PCle Gen 1 to Gen 6

### 1 Features

- 8 LP-HCSL outputs with programmable integrated 85-Ω (default) or 100-Ω differential output terminations
- 8 hardware output enable (OE#) controls
- Additive phase jitter after PCIE Gen 6 filter: 20 fs, RMS (maximum)
- Additive phase jitter after PCIE Gen 5 filter: 25 fs, RMS (maximum)
- Additive phase jitter after DB2000Q filter: 38 fs, RMS (maximum)
- Supports Common Clock (CC) and Individual Reference (IR) architectures
  - Spread spectrum-compatible
- Output-to-output skew: < 50 ps
- Input-to-output delay: < 3 ns
- Fail-safe input
- Programmable output slew rate control
- 3.3-V core and IO supply voltages
- Hardware-controlled low power mode (PD#)
- Current consumption: 72 mA maximum
- 6-mm × 6-mm, 48-pin VQFN package

### 2 Applications

- Microserver & tower server
- Storage area network & host bus adapter card
- Network attached storage
- Hardware accelerator
- Rack server

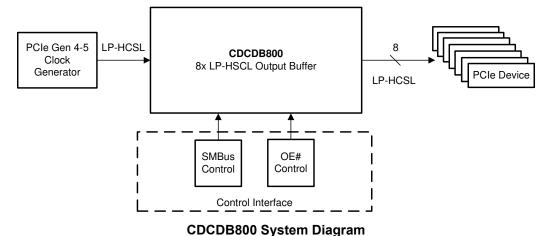
### 3 Description

The CDCDB800 is a 8-output LP-HCSL, DB800ZLcompliant, clock buffer capable of distributing the reference clock for PCle Gen 1-6, QuickPath Interconnect (QPI), UPI, SAS, and SATA interfaces. The SMBus interface and eight output enable pins allow the configuration and control of all eight outputs individually. The CDCDB800 is a DB800ZL derivative buffer and meets or exceeds the system parameters in the DB800ZL specification. It also meets or exceeds the parameters in the DB2000Q specification. The CDCDB800 is packaged in a 6-mm × 6-mm, 48-pin VQFN package.

#### **Device Information**

PART NUMBER	PACKAGE <sup>(1)</sup>	BODY SIZE (NOM)
CDCDB800	VQFN (48)	6.00 mm × 6.00 mm

For all available packages, see the orderable addendum at the end of the data sheet.





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### **4 Revision History**

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision A (September 2021) to Revision B (May 2022)	Page
Changed the data sheet title	1
Added PCle Gen 6 to the data sheet	
Changed the pin description for pin 46	3
Changes from Revision * (July 2021) to Revision A (September 2021)	Page
Changed pin descriptions for pins 47 and 48	3
Changed Absolute Maximum Ratings note 1	



### **5 Pin Configuration and Functions**

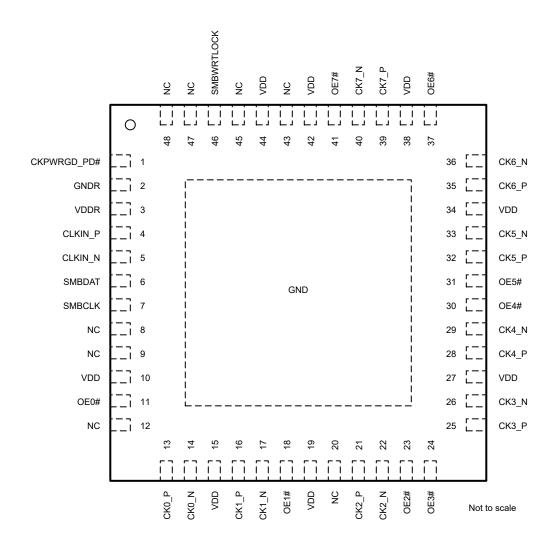


Figure 5-1. CDCDB800 RSL Package 48-Pin VQFN Top View

Table 5-1. Pin Functions CDCDB800

P	IN	TYPE <sup>(2)</sup>	DESCRIPTION		
NAME	NAME NO.		DESCRIPTION		
INPUT CLOCK					
CLKIN_P	4	I	LP-HCSL differential clock input. Typically connected directly to the differential		
CLKIN_N	5	I	output of clock source.		
OUTPUT CLOCKS					
CK0_P	13	0	LP-HCSL differential clock output of channel 0. Typically connected directly to PCle		
CK0_N	14	0	differential clock input. If unused, the pins can be left no connect.		
CK1_P	16	0	LP-HCSL differential clock output of channel 1. Typically connected directly to P		
CK1_N	17	0	differential clock input. If unused, the pins can be left no connect.		



### Table 5-1. Pin Functions CDCDB800 (continued)

Р	IN		Functions CDCDB800 (continued)		
NAME	IAME NO.		DESCRIPTION		
CK2_P	21	0	LP-HCSL differential clock output of channel 2. Typically connected directly to PCIe		
CK2_N	22	0	differential clock input. If unused, the pins can be left no connect.		
CK3_P	25	0	LP-HCSL differential clock output of channel 3. Typically connected directly to PCIe		
CK3_N	26	0	differential clock input. If unused, the pins can be left no connect.		
CK4_P	28	0	LP-HCSL differential clock output of channel 4. Typically connected directly to PCIe		
CK4_N	29	0	differential clock input. If unused, the pins can be left no connect.		
CK5_P	32	0	LP-HCSL differential clock output of channel 5. Typically connected directly to PCIe		
CK5_N	33	0	differential clock input. If unused, the pins can be left no connect.		
CK6_P	35	0	LP-HCSL differential clock output of channel 6. Typically connected directly to PCIe		
CK6_N	36	0	differential clock input. If unused, the pins can be left no connect.		
CK7_P	39	0	LP-HCSL differential clock output of channel 7. Typically connected directly to PCIe		
CK7_N	40	0	differential clock input. If unused, the pins can be left no connect.		
MANAGEMENT AN	D CONTROL (1)				
CKPWRGD_PD#	1	I, S, PD	Clock Power Good and Power Down multi-function input pin with internal $180\text{-}k\Omega$ pulldown. Typically connected to GPIO of microcontroller. If unused, the pin can be left no connect. After PWRGD has been asserted high for the first time, the pin becomes a PD# pin and it controls power-down mode: LOW: Power-down mode, all output channels tri-stated. HIGH: Normal operation mode.		
OE0#	E0# 11 I, S, PD		PD Output Enable for channel 0 with internal 180-k $\Omega$ pulldown, active low. Typical connected to GPIO of microcontroller. If unused, the pin can be left no connect LOW: enable output channel 0. HIGH: disable output channel 0.		
OE1#	18	I, S, PD	Output Enable for channel 1 with internal 180-k $\Omega$ pulldown, active low. Typically connected to GPIO of microcontroller. If unused, the pin can be left no connect. LOW: enable output channel 1. HIGH: disable output channel 1.		
OE2#	23	I, S, PD	Output Enable for channel 2 with internal 180-kΩ pulldown, active low. Typically connected to GPIO of microcontroller. If unused, the pin can be left no connect. LOW: enable output channel 2. HIGH: disable output channel 2.		
OE3#	24	I, S, PD	Output Enable for channel 3, with internal 180-k $\Omega$ pulldown, active low. Typically connected to GPIO of microcontroller. If unused, the pin can be left no connect. LOW: enable output channel 3. HIGH: disable output channel 3.		
OE4#	30	I, S, PD	Output Enable for channel 4, with internal 180-k $\Omega$ pulldown, active low. Typically connected to GPIO of microcontroller. If unused, the pin can be left no connect. LOW: enable output channel 4. HIGH: disable output channel 4.		
OE5#	31	I, S, PD	Output Enable for channel 5, with internal 180-kΩ pulldown, active low. Typically connected to GPIO of microcontroller. If unused, the pin can be left no connect. LOW: enable output channel 5. HIGH: disable output channel 5.		
OE6#	37	I, S, PD	Output Enable for channel 6 with internal 180-k $\Omega$ pulldown, active low. Typically connected to GPIO of microcontroller. If unused, the pin can be left no connect. LOW: enable output channel 6. HIGH: disable output channel 6.		
OE7#	41	I, S, PD	Output Enable for channel 7 with internal 180-k $\Omega$ pulldown, active low. Typically connected to GPIO of microcontroller. If unused, the pin can be left no connect. LOW: enable output channel 7. HIGH: disable output channel 7.		
SMBUS AND SMB	JS ADDRESS	-			
SMBDAT	6	1/0	Data pin of SMBus interface. Typically pulled up to 3.3-V VDD using external pullur resistor. The recommended pullup resistor value is > 8.5k.		

Table 5-1. Pin Functions CDCDB800 (continued)

	PIN	TYPE <sup>(2)</sup>	DESCRIPTION	
NAME	NO.	IIFE( /	DESCRIPTION	
SMBCLK	7	I	Clock pin of SMBus interface. Typically pulled up to 3.3-V VDD using external pullup resistor. The recommended pullup resistor value is > 8.5k.	
SMBWRTLOCK	46	I, PD	SMBWRTLOCK: Disables write commands on SMBus. All writes will be ignored when SMBWRTLOCK is asserted (reads are not affected). Internal 180-kΩ pulldown, active high.  0 = SMBus write enabled.  1 = SMBus write disabled.	
SUPPLY VOLTAGE	E AND GROUND			
GNDR	2	G	Ground.	
VDDR	3	Р	Power supply input for input clock receiver. Connect to 3.3-V power supply rail with decoupling capacitor to GND. Place a 0.1-µF capacitor close to each supply pin between power supply and ground.	
VDD	10, 15, 19, 27, 34, 38, 42, 44	Р	3.3-V power supply for output channels and core voltage.	
GND	DAP	G	Ground. Connect ground pad to system ground.	
NO CONNECT				
NC	8, 9, 12, 20, 43, 45	_	Do not connect to GND or VDD.	
NC	47, 48	_	No connect. Pins may be connected to GND, VDD, or otherwise tied to any potential within the Supply Voltage range stated in the Absolute Maximum Ratings.	

<sup>(1)</sup> The "#" symbol at the end of a pin name indicates that the active state occurs when the signal is at a low voltage level. When "#" is not present, the signal is active high.

The definitions below define the I/O type for each pin.

- - I = Input
  - O = Output
  - I / O = Input / Output
  - PU / PD = Internal 180-kΩ Pullup / Pulldown network biasing to VDD/2
  - PD = Internal 180-kΩ Pulldown
  - S = Hardware Configuration Pin
  - P = Power Supply
  - G = Ground



### **6 Specifications**

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1)

		MIN	MAX	UNIT
$V_{DD,}V_{DD\_R}$	Power supply voltage	-0.3	3.6	V
V <sub>IN</sub>	IO input voltage	-0.3	3.6	V
T <sub>J</sub>	Junction temperature		125	°C
T <sub>stg</sub>	Storage temperature	-65	150	°C

<sup>(1)</sup> Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

### 6.2 ESD Ratings

			VALUE	UNIT
V	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins <sup>(1)</sup>	±3500	V
V <sub>(ESD)</sub>	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins <sup>(2)</sup>	±1000	V

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### **6.3 Recommended Operating Conditions**

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
$V_{DD}$	IO, Core supply voltage	3	3.3	3.6	V
$V_{DD_R}$	Input supply voltage	3	3.3	3.6	V
T <sub>A</sub>	Ambient temperature	-40		105	°C

### 6.4 Thermal Information

		Device Package	
	THERMAL METRIC(1)	RSL (QFN)	UNIT
		48 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	32.2	°C/W
R <sub>0</sub> JC(top)	Junction-to-case (top) thermal resistance	22.3	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	14.3	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	0.5	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	14.2	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	6.1	°C/W

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

Product Folder Links: CDCDB800



### **6.5 Electrical Characteristics**

VDD, VDD\_R =  $3.3 \text{ V} \pm 5 \text{ \%}$ ,  $-40^{\circ}\text{C} \leq \text{T}_{A} \leq 105^{\circ}\text{C}$ . Typical values are at VDD = VDD\_R = 3.3 V,  $25^{\circ}\text{C}$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
CURRENT	CONSUMPTION					
		Active mode. CKPWRGD_PD# = 1			9	
$I_{DD\_R}$	Core supply current	Power-down mode. CKPWRGD PD# = 0			2.2	mA
		All outputs disabled			18	
I <sub>DD</sub>	IO supply current	All outputs active, 100MHz (Per output)			7.8	mA
		Power-down mode. CKPWRGD PD# = 0			1.5	
CLOCK INI	PUT					
f <sub>IN</sub>	Input frequency		50	100	250	MHz
V <sub>IN</sub>	Input voltage swing	Differential voltage between CLKIN_P and CLKIN_N(1)	200		2300	mV <sub>Diff-</sub>
dV/dt	Input voltage edge rate	20% - 80% of input swing	0.7			V/ns
DV <sub>CROSS</sub>	Total variation of V <sub>CROSS</sub>	Total variation across V <sub>CROSS</sub>		140		mV
DC <sub>IN</sub>	Input duty cycle		40		60	%
C <sub>IN</sub>	Input capacitance <sup>(2)</sup>	Differential capacitance between CLKIN_P and CLKIN_N pins		2.2		pF
CLOCK OL	JTPUT					
f <sub>OUT</sub>	Output frequency		50	100	250	MHz
C <sub>OUT</sub>	Output capacitance <sup>(1)</sup>	Differential capacitance between CKx_P and CKx_N pins		4		pF
V <sub>OH</sub>	Output high voltage	Simple and ad(2) (3)	225		270	
V <sub>OL</sub>	Output low voltage	Single-ended <sup>(2) (3)</sup>	10		150	
$V_{HIGH}$	Output high voltage	Measured into an AC load as defined in DB800ZL	660		850	
$V_{LOW}$	Output low voltage	Measured into an AC load as defined in DB800ZL	-150		150	
V <sub>MAX</sub>	Output Max voltage	Measured into an AC load as defined in DB800ZL			1150	
V <sub>CROSS</sub>	Crossing point voltage	(3) (4)	130		200	mV
V <sub>CROSSAC</sub>	Crossing point voltage (AC load)	Measured into an AC load as defined in DB800ZL	250		550	
DV <sub>CROSS</sub>	Total variation of V <sub>CROSS</sub>	Variation of V <sub>CROSS</sub> (3) (4)		35	140	
V <sub>ovs</sub>	Overshoot voltage	(3)			V <sub>OH</sub> +75	
V <sub>ovs(AC)</sub>	Overshoot voltage (AC load)	Measured into an AC load as defined in DB800ZL		V	<sub>HIGH</sub> +30 0	
V <sub>uds</sub>	Undershoot voltage	(3)			V <sub>OL</sub> -75	
V <sub>uds(AC)</sub>	Undershoot voltage	Measured into an AC load as defined in DB800ZL			V <sub>LOW</sub> - 300	mV
$V_{rb}$	Ringback Voltage	Measured into an AC load as defined in DB800ZL and taken from Single Ended waveform (relative to V <sub>HIGH</sub> and V <sub>LOW</sub> )	-0.2		0.2	V
7	Differential impedance (Default setting, 85 $\Omega$ )	Measured at V <sub>OL</sub> /V <sub>OH</sub>	81	85	89	
Z <sub>DIFF</sub>	Differential impedance (Output impedance selection bit =1, 100 $\Omega$ )	Measured at V <sub>OL</sub> /V <sub>OH</sub>	95	100	105	0
Z <sub>DIFF_CROS</sub>	Differential impedance (Default setting, 85 Ω)	Measured at V <sub>CROSS</sub>	68	85	102	Ω
S	Differential impedance (Output impedance selection bit = 1, 100 $\Omega$ )	Measured at V <sub>CROSS</sub>	80	100	120	



VDD, VDD\_R =  $3.3 \text{ V} \pm 5 \text{ \%}$ ,  $-40 ^{\circ}\text{C} \leq \text{T}_{A} \leq 105 ^{\circ}\text{C}$ . Typical values are at VDD = VDD\_R = 3.3 V,  $25 ^{\circ}\text{C}$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	S	MIN	TYP	MAX	UNIT
t <sub>EDGE</sub>	Differential edge rate	Measured (+-150 mV) around V	CROSS (7)	2		4	V/ns
Dt <sub>EDGE</sub>	Edge rate matching	Measured (+-75 mV) V <sub>CROSS</sub> (7	)			20	%
t <sub>STABLE</sub>	Power good assertion to stable clock output	CKPWRGD_PD# pin transitions from 0 to 1, f <sub>IN</sub> = 100 MHz	Measured when positive output reaches 0.2V			1.8	ms
t <sub>DRIVE_PD</sub> #	Power good assertion to outputs driven high	CKPWRGD_PD# pin transitions from 0 to 1, f <sub>IN</sub> = 100 MHz	Measured when positive output reaches 0.2V			300	μs
t <sub>OE</sub>	Output enable assertion to stable clock output	OEx# pin transitions from 1 to 0	)			10	
t <sub>OD</sub>	Output enable de-assertion to no clock output	OEx# pin transitions from 0 to 1				10	CLKIN Periods
t <sub>PD</sub>	Power-down assertion to no clock output	CKPWRGD_PD# pin transitions	s from 1 to 0			3	
t <sub>DCD</sub>	Duty cycle distortion	Differential; f <sub>IN</sub> = 100MHz, f <sub>IN_DC</sub> = 50%		-1		1	%
t <sub>DLY</sub>	Propagation delay	(5)		0.5		3	ns
t <sub>SKEW</sub>	Skew between outputs	(6)	(6)			50	ps
t <sub>DELAY(IN-</sub>	Input to output delay variation	Input-to-output delay variation a across voltage and temperature	-250		250	ps	
J <sub>CKx_DB2000</sub>	Additive jitter for DB2000Q	DB2000Q filter, for input of 200 differential swing @ 1.5 V/ns	mV			0.038	ps, RMS
	Additive jitter for PCIe6.0	PLL BW: 0.5 - 1 MHz; CDR = 10 MHz	Input clock slew rate = 2 V/ns			0.02	
	Additive jitter for PCle5.0	PCle5.0 filter				0.025	
J <sub>CKx</sub> PCIE	Additive jitter for PCle4.0	PLL BW = 2 - 5 MHz; CDR =	Input clock slew rate ≥ 1.8 V/ns			0.06	ps, RMS
	Additive jitter for PCIe3.0	10 MHz	Input clock slew rate ≥ 0.6 V/ns			0.1	
J <sub>CKx</sub>	Additive jitter	f <sub>IN</sub> = 100 MHz; slew rate ≥ 3 V/r 20 MHz integration bandwidth.	ns; 12 kHz to		100	160	fs, RMS
NF	Noise floor	f <sub>IN</sub> = 100 MHz; f <sub>Offset</sub> ≥ 10 MHz	Input clock slew rate ≥ 3 V/ns		-160	-155	dBc/Hz
SMBUS INT	ERFACE, OEx#, CKPWRGD_PD#				-		
V <sub>IH</sub>	High level input voltage			2.0			V
V <sub>IL</sub>	Low level input voltage					8.0	V
I <sub>IH</sub>	Input leakage current	With internal pull-up/pull-down	GND ≤ V <sub>IN</sub> ≤ V <sub>DD</sub>	-30		30	μA
I <sub>IL</sub>	Input leakage current	With internal pull-up/pull-down	GND ≤ V <sub>IN</sub> ≤ V <sub>DD</sub>	-30		30	μΑ
I <sub>IH</sub>	Input leakage current	Without internal pull-up/pull- down	GND ≤ V <sub>IN</sub> ≤ V <sub>DD</sub>	-5		5	μA
I <sub>IL</sub>	Input leakage current	Without internal pull-up/pull- down	GND ≤ V <sub>IN</sub> ≤ V <sub>DD</sub>	-5		5	μΑ

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## VDD, VDD\_R = $3.3 \text{ V} \pm 5 \text{ \%}$ , $-40^{\circ}\text{C} \leq \text{T}_{A} \leq 105^{\circ}\text{C}$ . Typical values are at VDD = VDD\_R = 3.3 V, $25^{\circ}\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
C <sub>IN</sub>	Input capacitance				4.5		pF
C <sub>OUT</sub>	Output capacitance				4.5		pF
I <sub>IH</sub>	Input leakage current	With internal pull-up/pull-down	$V_{IN} = V_{DD}$	-30		30	μΑ
C <sub>IN</sub>	Input capacitance <sup>(1)</sup>				4.5		pF

- (1) Voltage swing includes overshoot.
- (2) Not tested in production. Ensured by design and characterization.
- (3) Measured into DC test load.
- (4) V<sub>CROSS</sub> is single-ended voltage when CKx\_P = CKx\_N with respect to system ground. Only valid on rising edge of CKx, when CKx\_P is rising.
- (5) Measured from rising edge of CLK\_IN to any CKx output.
- (6) Measured from rising edge of any CKx output to any other CKx output.
- (7) Measured into AC test load.

### 6.6 Timing Requirements

VDD, VDD\_R =  $3.3 \text{ V} \pm 5 \text{ %}$ ,  $-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le 105^{\circ}\text{C}$ . Typical values are at VDD = VDD\_A = 3.3 V,  $25^{\circ}\text{C}$  (unless otherwise noted)

			MIN	NOM	MAX	UNIT
SMBUS-0	COMPATIBLE INTERFACE TIMING					
f <sub>SMB</sub>	SMBus operating frequency		10		400	kHz
t <sub>BUF</sub>	Bus free time between STOP and START		4.7			
t <sub>HD_STA</sub>	START condition hold time	SMBCLK low after SMBDAT low	4			
t <sub>SU_STA</sub>	START condition setup time	SMBCLK high before SMBDAT low	4.7			μs
t <sub>SU_STO</sub>	STOP condition setup time		4			
t <sub>HD_DAT</sub>	SMBDAT hold time		300			
t <sub>SU_DAT</sub>	SMBDAT setup time		250			ns
t <sub>TIMEOUT</sub>	Detect SMBCLK low timeout	In terms of device input clock frequency	1e6			cycles
t <sub>LOW</sub>	SMBCLK low period		4.7			
t <sub>HIGH</sub>	SMBCLK high period		4		50	μs
t <sub>F</sub>	SMBCLK/SMBDAT fall time <sup>(1)</sup>				300	
t <sub>R</sub>	SMBCLK/SMBDAT rise time <sup>(2)</sup>			,	1000	ns

- (1) TF = (VIHMIN + 0.15) to (VILMAX 0.15)
- (2) TR = (VILMAX 0.15) to (VIHMIN + 0.15)



### **6.7 Typical Characteristics**

Figure 6-1 shows both the phase noise of the source as well as the output of the DUT (CDCDB800). It can be seen from the phase noise plot that the DUT has a very low phase noise profile with total jitter of 71 fs, rms. If we rms subtract the clock reference noise, the additive jitter of CDCDB800 under typical conditions would be lower than 71 fs, rms.

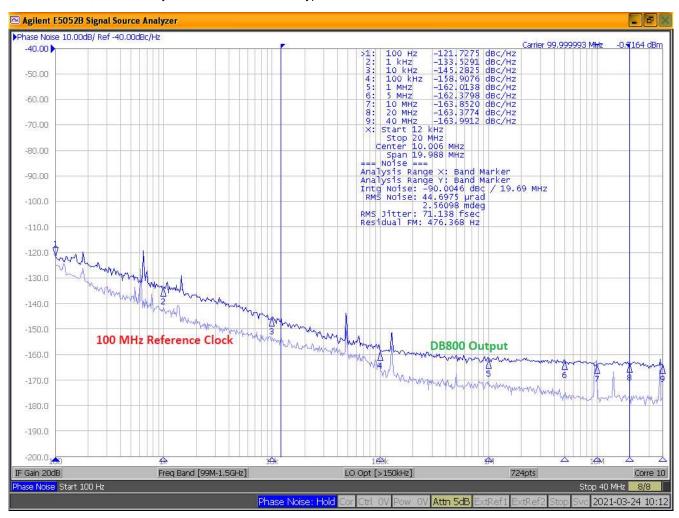


Figure 6-1. CDCDB800 Clock Out (CK0:8) Phase Noise



### 7 Parameter Measurement Information

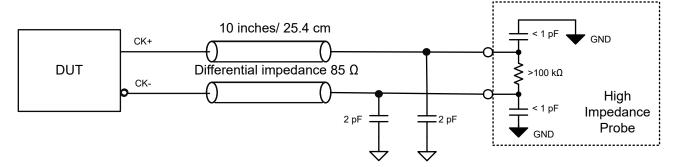
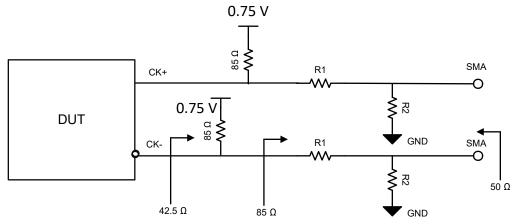


Figure 7-1. AC Test Load (Referencing Intel DB2000QL Document)



R1 = 47  $\Omega$  and R2 = 147  $\Omega$ .

Figure 7-2. DC Simulation Load (Referencing Intel DB2000QL Document)

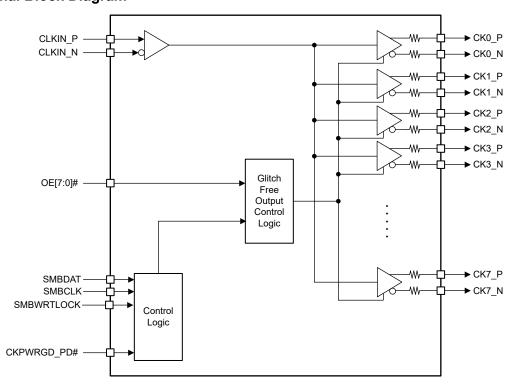


### 8 Detailed Description

### 8.1 Overview

The CDCDB800 is a low additive-jitter, low propagation delay clock buffer designed to meet the strict performance requirements for PCIe Gen 1-6, QPI, UPI, SAS, and SATA reference clocks. The CDCDB800 allows buffering and replication of a single clock source to up to eight individual outputs in the LP-HCSL format. The CDCDB800 also includes status and control registers accessible by an SMBus version 2.0 compliant interface. The device integrates a large amount of external passive components to reduce overall system cost.

### 8.2 Functional Block Diagram



### 8.3 Feature Description

#### 8.3.1 Fail-Safe Input

The CDCDB800 is designed to support fail-safe input operation feature. This feature allows the user to drive the device inputs before  $V_{DD}$  is applied without damaging the device. Refer to the *Absolute Maximum Ratings* table for more information on the maximum input supported by the device.

### 8.3.2 Output Enable Control

The CDCDB800 uses SMBus and OE# to control the state of the output channels. The OE# pins control the state of the output with the same number. For example, the OE5# pin controls the state of the CK5 output driver. The SMBus registers may enable or disable the output when the corresponding OE# pin is held low.

### 8.3.3 SMBus

The CDCDB800 has an SMBus interface that is active only when CKPWRGD\_PD# = 1. The SMBus allows individual enable/disable of each output.

When CKPWRGD\_PD# = 0, the SMBus pins are placed in a Hi-Z state, but all register settings are retained. The SMBus register values are only retained while  $V_{DD}$  remains inside of the recommended operating voltage.

#### 8.3.3.1 SMBus Address Assignment

The CDCDB800 responds to SMBus address of 0xD8 for write operations (Read/Write = 0), and 0xD9 for read operations (Read/Write = 1).

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### 8.4 Device Functional Modes

### 8.4.1 CKPWRGD\_PD# Function

The CKPWRGD\_PD# pin is used to set two state variables inside of the device: PWRGD and PD#. The PWRGD and PD# variables control which functions of the device are active at any time, as well as the state of the input and output pins.

The PWRGD and PD# states are multiplexed on the CKPWRGD\_PD# pin. CKPWRGD\_PD# must remain below  $V_{OL}$  and not exceed  $V_{DDR}$  + 0.3 V until  $V_{DD}$  and  $V_{DDR}$  are present and within the recommended operating conditions. After CKPWRGD\_PD# is set high, a valid CLKIN must be present to use PD#.

The first rising edge of the CKPWRGD\_PD# pin sets PWRGD = 1. After PWRGD is set to 1, the CKPWRGD\_PD# pin is used to assert PD# mode only. PWRGD variable will only be cleared to 0 with the removal of  $V_{DD}$  and  $V_{DDR}$ .

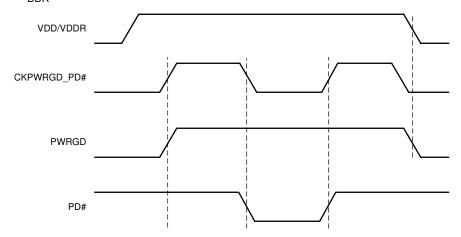


Figure 8-1. PWRGD and PD# State Changes



### 8.4.2 OE[7:0]# and SMBus Output Enables

Each output channel, 0 to 7, can be individually enabled or disabled by a SMBus control register bit, called SMB enable bits. Additionally, each output channel has a dedicated, corresponding, OE[7:0]# hardware pin. The OE[7:0]# pins are asynchronously asserted-low signals that may enable or disable the output.

Refer to Table 8-1 for enabling and disabling outputs through the hardware and software. Note that both the SMB enable bit must be a 1 and the OEx# pin must be an input low voltage 0 for the output channel to be active.

Table 8-1. OE[7:0]# Functionali
---------------------------------

Control Inputs		e Variables rnal)	OE[7:0]# HARDWARE PINS AND SMBus CONTROL REGISTER BITS				CK[7:0]_P/											
CKPWRGD_P D#	PWRGD	PD#	CERIN	OE[7:0]#	OUT_EN_CLK[ 7:0]	DRIVE_OP_ST ATE_CTRL	CK[7:0]_N											
0	0	0	Х	Х	Х	Х	LOW/LOW											
				X	0	0	LOW/LOW											
			X <sup>(1)</sup>			1	TRI-STATE											
1		1			Х	0	LOW/LOW											
	1															'	^	1
			Running <sup>(1)</sup>	0	1	Х	Running											
0		0	X <sup>(2)</sup>	×	Х	0	LOW/LOW											
			A(2)			1	TRI-STATE											

<sup>(1)</sup> To enter the power-down state, CLKIN must remain active for at least 3 clock cycles after CKPWRGD PD# transitions from 1 to 0.

### 8.4.3 Output Slew Rate Control

The CDCDB800 provides output slew rate control feature which customer can use to compensate for increased output trace length based on their board design. The slew rate of a bank of 4 outputs 0 to 3 and 4 to 7, can be changed within a given range by a SMBus control register called CAPTRIM. Refer to Table 8-15 for more information.

### 8.4.4 Output Impedance Control

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The integrated termination on the CDCDB800 can be programmed either for 85  $\Omega$  or 100  $\Omega$ . This flexibility ensures that the customer can use the same device across various applications irrespective of the characteristic board impedance which is typically either 85  $\Omega$  or 100  $\Omega$ . This termination resistor can be changed for all the outputs as whole using bit 5 of a register called OUTSET. Refer to Table 8-13 for more information.

### 8.5 Programming

The CDCDB800 uses SMBus to program the states of its eight output drivers. See *SMBus* for more information on the SMBus programming, and *Register Maps* for information on the registers.

Table 8-2. Command Code Definition

BIT	DESCRIPTION			
7	0 = Block Read or Block Write operation 1 = Byte Read or Byte Write operation			
(6:0)	legister address for <i>Byte</i> operations, or starting register address for <i>Block</i> , operations			

Product Folder Links: CDCDB800

<sup>(2)</sup> To enter the powered-up state with active clock outputs, CLKIN must be active before CKPWRGD\_PD# transitions from 0 to 1.



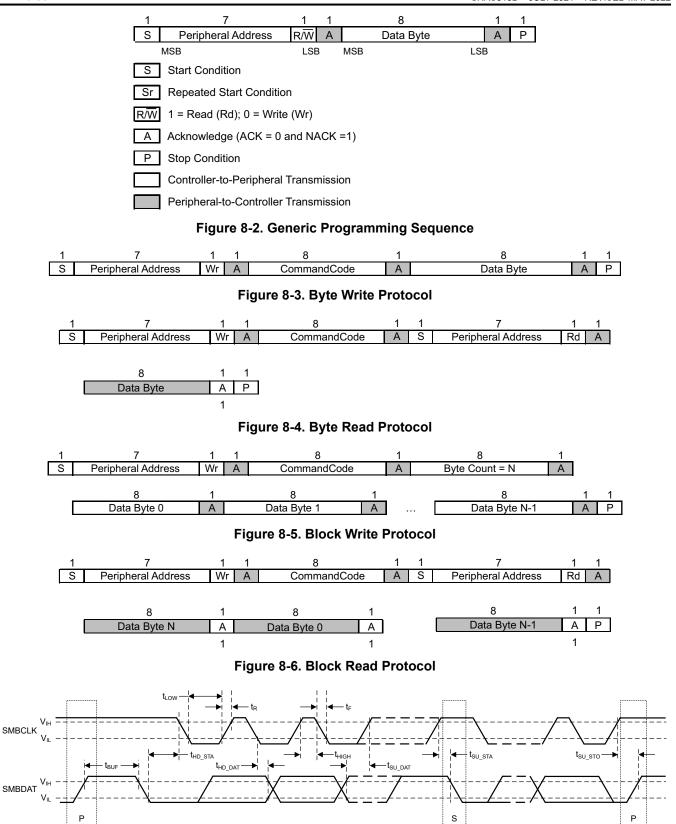


Figure 8-7. SMBus Timing Diagram



### 8.6 Register Maps

### 8.6.1 CDCDB800 Registers

Table 8-3 lists the CDCDB800 registers. All register locations not listed in Table 8-3 should be considered as reserved locations and the register contents should not be modified.

Table 8-3. CDCDB800 Registers

Address	Acronym	Register Name	Section
0h	RCR1	Reserved Control Register 1	Go
1h	OECR1	Output Enable Control 1	Go
2h	OECR2	Output Enable Control 2	Go
3h	OERDBK	Output Enable# Pin Read Back	Go
4h	RCR2	Reserved Control Register 2	Go
5h	VDRREVID	Vendor/Revision Identification	Go
6h	DEVID	Device Identification	Go
7h	BTRDCNT	Byte Read Count Control	Go
8h	OUTSET	Output Setting Control	Go
4Ch	CAPTRIM	Slew Rate Capacitor Cluster 1 & 2	Go

Complex bit access types are encoded to fit into small table cells. Table 8-4 shows the codes that are used for access types in this section.

Table 8-4. CDCDB800 Access Type Codes

Access Type	Code	Description			
Read Type					
R	R	Read			
Write Type					
W	W	Write			
Reset or Default	Reset or Default Value				
-n		Value after reset or the default value			

### 8.6.1.1 RCR1 Register (Address = 0h) [reset = 47h]

RCR1 is shown in Table 8-5.

Return to the Summary Table.

The RCR1 register contains reserved bits.

Table 8-5. RCR1 Register Field Descriptions

Bit	Field	Туре	Reset	Description
7-4	Reserved	R	4h	Reserved.
3-0	Reserved	R/W	7h	Writing to these bits will not affect the functionality of the device.

### 8.6.1.2 OECR1 Register (Address = 1h) [reset = FFh]

OECR1 is shown in OECR1 Register Field Descriptions.

Return to the Summary Table.

The OECR1 register contains bits that enable or disable individual output clock channels [5:0].

Table 8-6. OECR1 Register Field Descriptions

Table 8-6. OECRT Register Field Descriptions								
Bit	Field	Type	Reset	Description				
7	OUT_EN_CLK5	R/W	1h	This bit controls the output enable signal for output channel CK5_P/CK5_N.  0h = Output Disabled  1h = Output Enabled				
6	OUT_EN_CLK4	R/W	1h	This bit controls the output enable signal for output channel CK4_P/CK4_N.  0h = Output Disabled  1h = Output Enabled				
5	OUT_EN_CLK3	R/W	1h	This bit controls the output enable signal for output channel CK3_P/CK3_N.  0h = Output Disabled  1h = Output Enabled				
4	OUT_EN_CLK2	R/W	1h	This bit controls the output enable signal for output channel CK2_P/CK2_N.  0h = Output Disabled  1h = Output Enabled				
3	Reserved	R/W	1h	Writing to this bit will not affect the functionality of the device.				
2	OUT_EN_CLK1	R/W	1h	This bit controls the output enable signal for output channel CK1_P/CK1_N.  0h = Output Disabled  1h = Output Enabled				
1	OUT_EN_CLK0	R/W	1h	This bit controls the output enable signal for output channel CK0_P/CK0_N.  0h = Output Disabled  1h = Output Enabled				
0	Reserved	R/W	1h	Writing to this bit will not affect the functionality of the device.				

### 8.6.1.3 OECR2 Register (Address = 2h) [reset = 0Fh]

OECR2 is shown in OECR2 Register Field Descriptions.

Return to the Summary Table.

The OECR2 register contains bits that enable or disable individual output clock channels [7:6].

**Table 8-7. OECR2 Register Field Descriptions** 

Bit	Field	Туре	Reset	Description
7-3	Reserved	R/W	1h	Writing to these bits will not affect the functionality of the device.
2	OUT_EN_CLK7	R/W	1h	This bit controls the output enable signal for output channel CK7_P/CK7_N.  0h = Output Disabled 1h = Output Enabled
1	Reserved	R/W	1h	Writing to this bit will not affect the functionality of the device.
0	OUT_EN_CLK6	R/W	1h	This bit controls the output enable signal for output channel CK6_P/CK6_N.  0h = Output Disabled  1h = Output Enabled

### 8.6.1.4 OERDBK Register (Address = 3h) [reset = 0h]

OERDBK is shown in Table 8-8.

Return to the Summary Table.



The OERDBK register contains bits that report the current state of the OE[7:0]# input pins.

### Table 8-8. OERDBK Register Field Descriptions

Bit	Field	Туре	Reset	Description
7	RB_OE7	R	0h	This bit reports the logic level present on the OE7# pin.
6	RB_OE6	R	0h	This bit reports the logic level present on the OE6# pin.
5	RB_OE5	R	0h	This bit reports the logic level present on the OE5# pin.
4	RB_OE4	R	0h	This bit reports the logic level present on the OE4# pin.
3	RB_OE3	R	0h	This bit reports the logic level present on the OE3# pin.
2	RB_OE2	R	0h	This bit reports the logic level present on the OE2# pin.
1	RB_OE1	R	0h	This bit reports the logic level present on the OE1# pin.
0	RB_OE0	R	0h	This bit reports the logic level present on the OE0# pin.

### 8.6.1.5 RCR2 Register (Address = 4h) [reset = 0h]

RCR2 is shown in RCR2 Register Field Descriptions.

Return to the Summary Table.

The RCR2 register contains reserved bits.

### Table 8-9. RCR2 Register Field Descriptions

Bit	Field	Туре	Reset	Description
7-0	Reserved	R	0h	Reserved.

### 8.6.1.6 VDRREVID Register (Address = 5h) [reset = 0Ah]

VDRREVID is shown in Table 8-10.

Return to the Summary Table.

The VDRREVID register contains a vendor identification code and silicon revision code.

### Table 8-10. VDRREVID Register Field Descriptions

Bit	Field	Туре	Reset	Description				
7-4	REV_ID	R	0h	Silicon revision code. Silicon revision code bits [3:0] map to register bits [7:4] directly.				
3-0	VENDOR_ID	R	Ah	Vendor identification code. Vendor ID bits [3:0] map to register bits [3:0] directly.				

### 8.6.1.7 DEVID Register (Address = 6h) [reset = E7h]

DEVID is shown in Table 8-11.

Return to the Summary Table.

The DEVID register contains a device identification code.

### Table 8-11. DEVID Register Field Descriptions

Bit	Field	Туре	Reset	Description
7-0	DEV_ID	R	E7h	Device ID code. Device ID bits[7:0] map to register bits[7:0] directly.

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### 8.6.1.8 BTRDCNT Register (Address = 7h) [reset = 8h]

BTRDCNT is shown in Table 8-12.

Return to the Summary Table.

The BTRDCNT register contains bits [4:0] which configure the number of bytes which will be read back.

**Table 8-12. BTRDCNT Register Field Descriptions** 

_					
	Bit	Field	Туре	Reset	Description
	7-5	Reserved	R/W	0h	Writing to these bits will not affect the functionality of the device.
	4	BYTE_COUNTER	R/W	0h	Writing to this register configures how many bytes will be read back.
	3-0	BYTE_COUNTER	R/W	8h	withing to this register configures flow many bytes will be read back.

### 8.6.1.9 OUTSET Register (Address = 8h) [reset = 0h]

OUTSET is shown in Table 8-13.

Return to the Summary Table.

Bit5 of the OUTSET register sets the termination for all the outputs while bit4 can be used to set the power-down state for all outputs. The remaining bits for this register are reserved.

**Table 8-13. OUTSET Register Field Descriptions** 

Bit	Field	Туре	Description					
7-6	Reserved	R	0h	Reserved.				
5	CH_ZOUT_SEL	R/W	0h	Select between 85 $\Omega$ (0) and 100 $\Omega$ (1) Output impedance				
4	d_DRIVE_OP_STATE_CTRL	R/W	Oh	Power-down state of all output clocks. 0: LOW/LOW 1: TRI_STATE				
3-0	Reserved	R/W	0h	Register bits can be written to 0. Writing a different value than 0 will affect device functionality.				

### 8.6.1.10 CAPTRIM Register (Address = 4Ch) [reset = 66h]

CAPTRIM is shown in Table 8-15.

Return to the Summary Table.

Bits [7:4] of the CAPTRIM register is used to control the slew rate for output channel cluster 2. Bits [3:0] control the slew rate for output channel cluster 1. Refer below for cluster identification.

**Table 8-14. Cluster Identification** 

Cluster	Outputs
1	CK3, CK2, CK1, CK0
2	CK7, CK6, CK5, CK4

### Table 8-15. CAPTRIM Register Field Descriptions

Bit	Field	Туре	Reset	Description						
7-4	CLUSTER2_CAP_TRIM	R/W	6h	Slew Rate Reduction Cap Trim for Cluster 2. Default value of 6h.  0: minimum						
				F: maximum						
3-0	CLUSTER1 CAP TRIM	R/W	6h	Slew Rate Reduction Cap Trim for Cluster 1. Default value of 6h.						
3-0	OLOGILITI_OAI _ITTIW	1000	OII	0: minimum						
				F: maximum						

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### 9 Application and Implementation

### **Note**

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

### 9.1 Application Information

The CDCDB800 is a fanout buffer that supports PCle generation 4 and PCle generation 5 REFCLK distribution. The device is used to distribute up to eight copies of a typically 100-MHz clock.

### 9.2 Typical Application

Figure 9-1 shows a CDCDB800 typical application. In this application, a clock generator provides a 100-MHz reference to the CDCDB800 which then distributes that clock to PCIe endpoints. The clock generator may be a discrete clock generator like the CDCI6214 or it may be integrated in a larger component such as a Platform Controller Hub (PCH) or application processor.

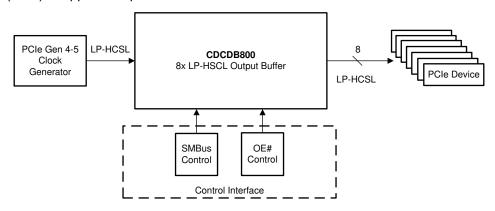


Figure 9-1. Typical Application

#### 9.2.1 Design Requirements

Consider a typical server motherboard application which must distribute a 100-MHz PCIe reference clock from the PCH of a processor chipset to multiple endpoints. An example of clock input and output requirements is:

- · Clock Input:
  - 100-MHz LP-HCSL
- Clock Output:
  - 2x 100-MHz to processors, LP-HCSL
  - 3x 100-MHz to riser/retimer, LP-HCSL
  - 3x 100-MHz to DDR memory controller, LP-HCSL

### 9.2.2 Detailed Design Procedure

The following items must be determined before starting design of a CDCDB800 socket:

· Output Enable Control Method

### 9.2.2.1 Output Enable Control Method

The device provides an option to either use SMBus programmed registers (software) to control the outputs or by using the hardware OE# pins. In case of using software to control the outputs, the hardware OE# pins can be left floating as each of these pins have a pulldown to ground. Refer to Table 8-1 and the *Register Maps* section for more information on programming the register.

When the user wants to control the outputs with the hardware OE# pins, they can do so for example by connecting these pins to a GPIO controller and follow the *Pin Configuration and Functions* section to set the outputs to HIGH/LOW. The bits OUT\_EN\_CLK7 to OUT\_EN\_CLK0 used to control the outputs are shown in registers OECR1 (Table 8-6) and OECR2 (Table 8-7). These register bits are set to 1 by default to ensure that the outputs are "software enabled" and their state is therefore set by hardware OE# pins.

### 9.2.3 Application Curves

Figure 6-1 in the *Typical Characteristics* section can be used as both an application curve and a typical characteristics plot in this example.

The Figure 9-2 and Figure 9-3 show characterization data for the Output slew rate for various CAPTRIM codes and across temperature. Customers can use these plots as reference for choosing the appropriate output slew rate based on their system requirement.

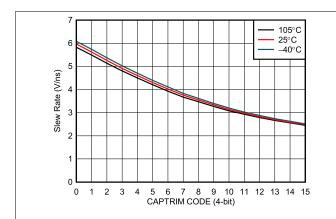


Figure 9-2. Output Slew Rate vs. CAPTRIM Code

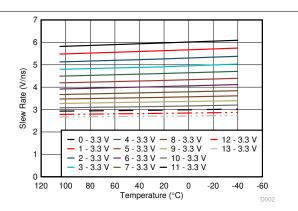


Figure 9-3. Slew Rate Variation Across Temperature for Different CAPTRIM Code



### 10 Power Supply Recommendations

High-performance clock buffers are sensitive to noise on the power supply, which can dramatically increase the additive jitter of the buffer. Thus, it is essential to reduce noise from the system power supply, especially when the jitter and phase noise is critical to applications.

Filter capacitors are used to eliminate the low-frequency noise from the power supply, where the bypass capacitors provide the very low impedance path for high-frequency noise and guards the power-supply system. against induced fluctuations. These bypass capacitors also provide instantaneous current surges as required by the device and should have low equivalent series resistance (ESR). To properly use the bypass capacitors, place the capacitors very close to the power-supply terminals and lay out with short loops to minimize inductance. TI recommends to insert a ferrite bead between the board power supply and the chip power supply that isolates the high-frequency switching noises generated by the clock buffer. These beads prevent the switching noise from leaking into the board supply. It is imperative to choose an appropriate ferrite bead with very low DC resistance to provide adequate isolation between the board supply and the chip supply, as well as to maintain a voltage at the supply terminals that is greater than the minimum voltage required for proper operation.

Figure 10-1 shows the recommended power supply filtering and decoupling method.

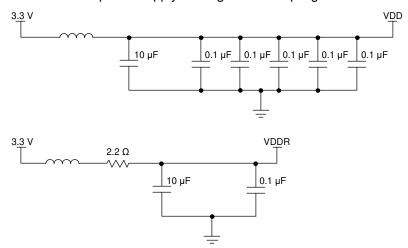


Figure 10-1. Power Supply Decoupling

Product Folder Links: CDCDB800

### 11 Layout

### 11.1 Layout Guidelines

The following section provides the layout guidelines to ensure good thermal performance and power supply connections for the CDCDB800.

In *Layout Examples*, the CDCDB800 has  $85-\Omega$  differential output impedance LP-HCSL format drivers as per register default settings. All transmission lines connected to CKx pins should be  $85-\Omega$  differential impedance,  $42.5-\Omega$  single-ended impedance to avoid reflections and increased radiated emissions. If  $100-\Omega$  output impedance is enabled, the transmission lines connected to CKx pins should be  $100-\Omega$  differential impedance,  $50-\Omega$  single-ended impedance. Take care to eliminate or reduce stubs on the transmission lines.

### 11.2 Layout Examples

Figure 11-1 through Figure 11-3 are printed circuit board (PCB) layout examples that show the application of thermal design practices and a low-inductance ground connection between the device DAP and the PCB.

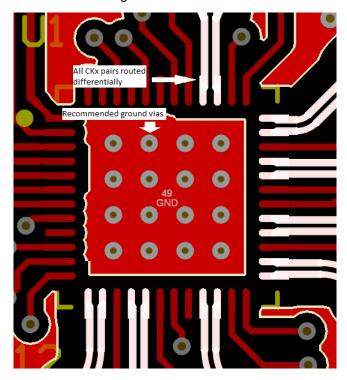


Figure 11-1. PCB Layout Example for CDCDB800, Top layer



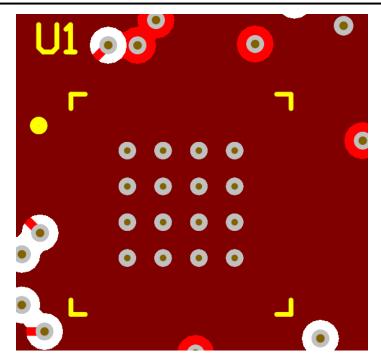


Figure 11-2. PCB Layout Example for CDCDB800, GND Layer

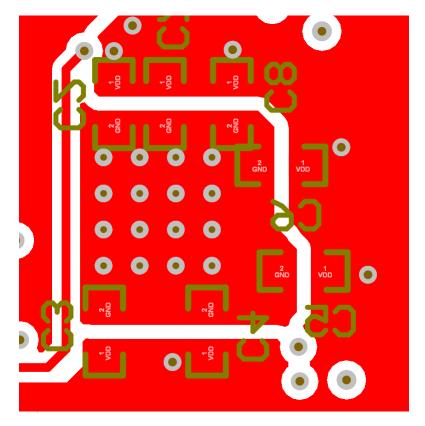


Figure 11-3. PCB Layout Example for CDCDB800, Bottom Layer

### 12 Device and Documentation Support

### 12.1 Device Support

### 12.1.1 TICS Pro

TICS Pro is an offline software tool for EVM programming and also for register map generation to program a device configuration for a specific application. For TICS Pro, go to https://www.ti.com/tool/TICSPRO-SW.

### 12.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Subscribe to updates* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 12.3 Support Resources

TI E2E<sup>™</sup> support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

#### 12.4 Trademarks

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### 12.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 12.6 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

### 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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#### PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
CDCDB800RSLR	ACTIVE	VQFN	RSL	48	4000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 105	CDCB800	Samples
CDCDB800RSLT	ACTIVE	VQFN	RSL	48	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 105	CDCB800	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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## **PACKAGE OPTION ADDENDUM**

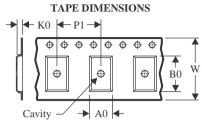
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### **PACKAGE MATERIALS INFORMATION**

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### TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
CDCDB800RSLR	VQFN	RSL	48	4000	330.0	16.4	6.3	6.3	1.1	12.0	16.0	Q2
CDCDB800RSLT	VQFN	RSL	48	250	180.0	16.4	6.3	6.3	1.1	12.0	16.0	Q2

## **PACKAGE MATERIALS INFORMATION**

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### \*All dimensions are nominal

	Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)	
ı	CDCDB800RSLR	VQFN	RSL	48	4000	367.0	367.0	35.0	
	CDCDB800RSLT	VQFN	RSL	48	250	210.0	185.0	35.0	

4207548/B 06/11

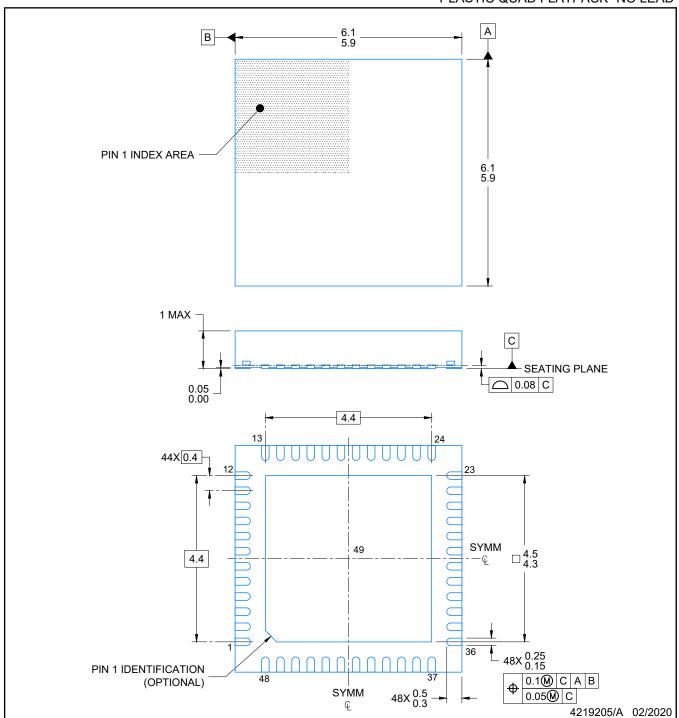
# RSL (S-PVQFN-N48) PLASTIC QUAD FLATPACK NO-LEAD 6,15 5,85 6,15 5,85 PIN 1 INDEX AREA TOP AND BOTTOM 1,00 0,80 0,20 REF. SEATING PLANE 0,08 0,05 0,00 0,40 48 THERMAL PAD SIZE AND SHAPE SHOWN ON SEPARATE SHEET 37 36 $48 \times \frac{0.26}{0.14}$ 4,40

NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.

- B. This drawing is subject to change without notice.
- C. Quad Flatpack, No-leads (QFN) package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.



PLASTIC QUAD FLATPACK- NO LEAD

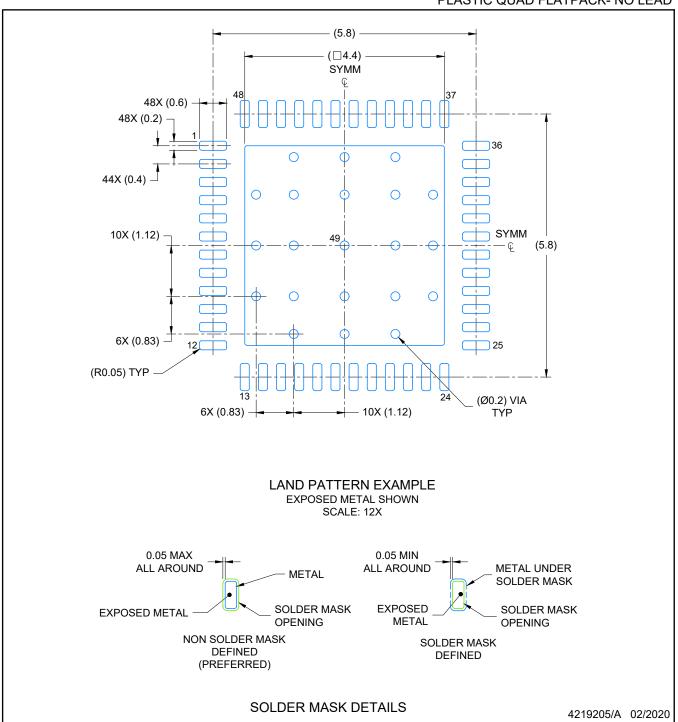


### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.



PLASTIC QUAD FLATPACK- NO LEAD

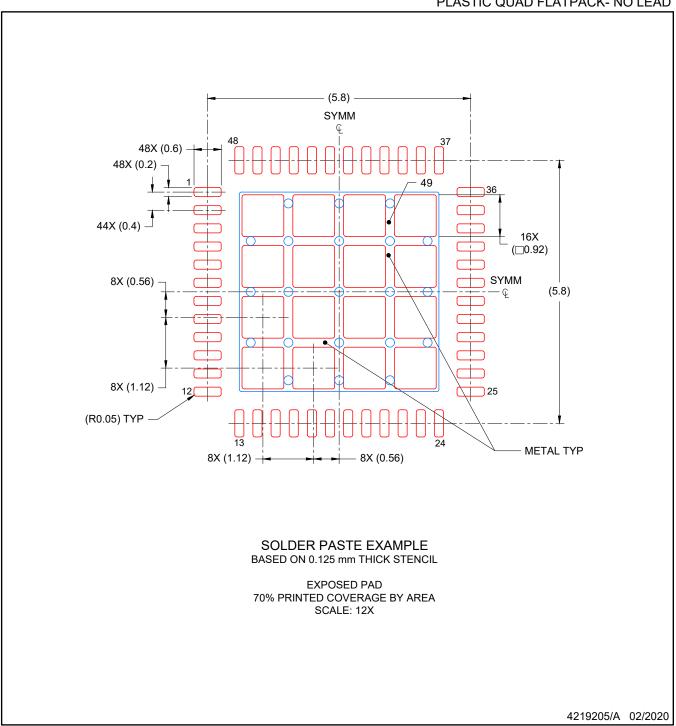


NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- 5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



PLASTIC QUAD FLATPACK- NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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