## ANALOG DEVICES

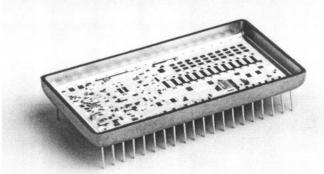
## Ultra High-Speed 12-Bit A/D Converter

HAS-1204

### FEATURES

12-Bit Resolution 500kHz Word Rates Internal Track-and-Hold Single 40-Pin DIP

APPLICATIONS Medical Instrumentation Radar Systems Test Systems Waveform Analysis Fast Fourier Transforms

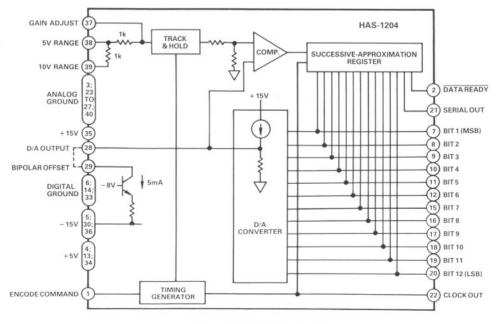


### GENERAL DESCRIPTION

The HAS-1204 A/D Converter is complete 12 bit hybrid A/D converter in a single 40-pin metal DIP. In this context, "complete" means the unit includes a track-and-hold (T/H) amplifier, encoder, and all the necessary timing circuits. It is a remarkable, self-contained device ready to perform the conversion function without the need for external circuits.

The maximum conversion time of the HAS-1204 is 2.0 microseconds, including the acquisition time of the internal T/H. The large-signal bandwidth of the T/H is 4MHz and the small-signal bandwidth is 7MHz. This combination of characteristics assures that the HAS-1204 will operate at word rates from dc through 600KHz, digitizing analog signals containing frequency components to 250kHz with minimum atternuation or distortion.

Integrating the T/H, encoder, and timing circuits into a single package allows optimum matching of T/H-encoder parameters to obtain the best possible performance. It also lowers the overall power dissipation to a maximum 2/2 watts, making the HAS 1204 an ideal choice for designers who face space and/or power restrictions for their designs.





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# **SPECIFICATIONS** (typical @ + 25°C with nominal power supplies unless otherwise noted)

Model RESOLUTION (FS = Full Scale)	Units Bits (% FS)	HAS-1204BM	HAS-1204SM	- OUTLINE DIMENSIONS
	Bits (%FS)	12(0.024)	*	Dimensions shown in inches and (mm).
LSB WEIGHT		1.00		
5V Input Range 10V Input Range	mV	1.22	*	0.19 (4.83) MAX L
	mV	2.44	*	
ACCURACY				
Linearity@dc	%FS $\pm 1/2$ LSB	0.0125	*	0.24 (6.09)
Monotonicity		Guaranteed	*	MIN
Nonlinearity vs. Temperature	ppm/°C	3	*	4
Gain Error	%FS(max)	0.1(0.7)	*	0.125 (3.175)
Gain vs. Temperature	ppm/°C	35	*	±0.010 (2.54) ID BEAD (ON BOTTOM) AND ±0.002 DOT (ON TOP) DENOTE PIN 1
DYNAMIC CHARACTERISTICS				IS PLACES
In-Band Harmonics <sup>1</sup>				
(dc to 60kHz)	dB below FS	75	*	
(60kHz to 120kHz)	dB below FS	75	*	± 0.020
(120kHz to 200kHz)	dB below FS	70	*	Σ # (c)
Conversion Rate	kHz	500	*	(25.063)
Conversion Time	μs, max	2.0	*	145 (29.083) MAX 900 (22.860) ± 0.0
Aperture Uncertainty (Jitter)	ps	60	*	F1-1-
Aperture Time (Delay)	ns (min/max)	10(4/18)	*	
Signal to Noise Ratio (SNR) <sup>2</sup>	dB	69	*	
Transier Response <sup>2</sup> Overvoltage <del>Reco</del> very <sup>4</sup>	ns	400	*	
Overvoltage Recovery* Input Bandwidth	ns	900	*	0.125 (3.175) 1.900 (48.260) ± 0.006
	MILE	-	~	
Small Signal, - 3dH <sup>5</sup>	MHz	7	*	
Large Signal, - 3dB <sup>6</sup> Two-Tone Linearity @ Input Frequ <del>encies</del> )	7) MHz	4		HAS-1204 PIN DESIGNATIONS
(17.5kHz; 52.5kHz)		7		
	dB below FS		*	(As viewed from bottom)
ANALOGINPUT	11/	-/-	1~	
Voltage Ranges	V,FS	0 to - 5; 0 to - 10	$ \times $	
	111	$\pm$ ; $\pm$ 2.5	1 1 / /	PIN EUNCTION PIN FUNCTION
Overvoltage	V, max	2×FS	*_/_/_/	40 ANALOG SBOUND 1 ENCODE COMMAND
Impedance				39 10VRANGE 2 DATA READY
5V Ranges	$\Omega$ (max)	$1,000(\pm 10)$	/*/	38 5V RANGE 3 ANALOG GROUND 37 GAIN ADJUST 4 +5V
10V Ranges	$\Omega(\max)$	$2,000(\pm 20)$	1	36 -15V 57 -15V
Offset <sup>8</sup>	100000 00		/ / /	35 + 15V & DIGITAL GROUND
Initial-10V Input	mV (max)	10(60)	* /	34 35 DIGITAL GROUND 32 EACTORY USE ONLY 9 BIT 3
vs. Temperature (Unipolar)	FS ppm/°C	15	*	38 DIGITAL GROUND 8 BIT 2 32 EACTORY USE ONLY 9 BIT 3
vs. Temperature (Bipolar)	FS ppm/°C	50	*	B1 FACTORY USE ONLY 10 BIT 4
ENCODE COMMAND INPUT <sup>9</sup>				30 -15V 11 BIT5
Logic Levels, TTL-Compatible	V	"0" = $0$ to $+0.4$	*	29 BIPOLAR OFFSET 12 BIT 6 28 D/A OUTPUT 13 +5V
		" $1" = +2.4 \text{ to } +5$	*	28 D/A OUTPUT 13 +5V 27 ANALOG GROUND 14 DIGITAL GROUND
Impedance	LS TTL Loads	2	*	26 ANALOG GROUND 15 BIT 7
Rise and Fall Times	ns, max	10	*	25 ANALOG GROUND 16 BIT 8
Width				24 ANALOG GROUND 17 BIT 9   23 ANALOG GROUND 18 BIT 10
Min	ns	90	*	22 CLOCK OUT 19 BIT 11
Max	ns	160	*	21 SERIALOUT 20 BIT 12 (LSB)
Frequency	kHz	dc to 500	*	
DIGITAL OUTPUT <sup>10</sup>				
Format	Data Bits	12 Parallel; NRZ	*	NOTES
	Data Ready	1;RZ	*	NOTES *Specification same as HAS-1204BM
Logic Levels, TTL-Compatible	V	"0" = 0 to $+0.4$	*	*Specification same as HAS-1204BM <sup>1</sup> In-band harmonics expressed in terms of spurious in-band signals
5.5 Prod 6.0 million		"1" = +2.4  to  +5	*	generated at 500kHz encode rate at analog input frequencies
Drive	TTL Loads	1 Standard	*	shown in ( ).
Coding				<sup>2</sup> RMS signal to rms noise ratio with 50kHz analog input and encode rate of 500kHz; input signal at -1.0dP
Unipolar Mode		Complementary	*	500kHz; input signal at -1.0dB. <sup>3</sup> For full-scale step input, 12-bit accuracy attained in specified time.
		Binary (CBN)		*Recovers to specified performance in specified time after 2×FS
Bipolar Mode		Complementary	*	input voltage.
		Offset Binary		<sup>5</sup> With analog input 40dB below FS.
		(COB)		<sup>6</sup> With FS analog input. (Large-signal bandwidth flat within 0.5dB, dc to 1MHz).
OWER REQUIREMENTS				<sup>7</sup> Each input frequency applied at a level 7dB below full scale.
$+15V \pm 0.5V$	mA, max	54	*	<sup>8</sup> Externally adjustable to zero.
$-15V \pm 0.5V$	mA, max	40	*	<sup>9</sup> Transition from digital "0" to digital "1" initiates encoding.
$+5V \pm 0.5V$	mA, max	160	*	<sup>10</sup> Use trailing edge of Data Ready pulse to strobe digital outputs into external circuits (See Figure 2).
Power Dissipation <sup>11</sup>	W, max	2.2	*	<sup>11</sup> Power dissipation shown is at zero input. Power dissipation is 2.3 Watts
EMPERATURE RANGE <sup>12</sup>	++ , 111aA	£ . £		maximum for -10V input.
Operating	°C			$^{12}$ T = Case temperature.
Storage	°C	-25  to  + 85	-55 to $+100$	<sup>13</sup> Maximum junction temperature = 150°C. Operating unit requires 500 cubic feet per minute (CEPM) moving air.
	°C	-65 to +150	*	500 cubic feet per minute (CFPM) moving air.
HERMAL RESISTANCE <sup>13</sup>				Specifications subject to change without notice.
Junction to Air, 0ja (Free Air)				
Junction to Case, $\theta$ jc	°C/W	25	*	
	°C/W	16	*	
RICES				
(1-24)	\$	378	435	
(100's)				
For applications assistance, call Compute	\$	294	338	

### THEORY OF OPERATION/TIMING

Refer to the block diagram of the HAS-1204.

Analog input signals to be digitized are applied to either Pin 38 (5V RANGE) or Pin 39 (10V RANGE), depending upon their amplitude. These signals are inputs to the internal track-and-hold (T/H) which is normally operating in the "track" mode as a buffer amplifier, following all changes in analog as they occur.

An external strap, shown between Pin 28 and Pin 29, is used if operating the converter in the bipolar mode; it is important to keep this strap as short as possible. For unipolar operation, connect Pin 29 to ground.

The user determines the point at which the digitizing process is to be initiated by controlling the application of the TTL-compatible Encode Command pulse. Its positive-going leading edge switches the T/H to the "hold" mode of operation, "freezing" the analog input signal and beginning the digitizing process. As shown in the block diagram, the Encode Command applied to Pin 1 generates the required timing signals within the HAS-1204 A/D, making it unnecessary to add external circuitz.

The held value of analog input is part of the input to a high-speed comparator within the converter. The other isput is the analog output of the internal high-speed, high-accuracy D/A converter. The resulting output of the comparator is applied to the successive approximation register (SAR), also controlled by timing signals initiated by the encode command.

Digital outputs are available in both serial and parallel formats, as shown in Figure 1, HAS-1204 Timing.

Times shown in the timing diagram are typical times, unless noted otherwise. In the illustration, the Track/Hold signal is internal, not available to the user; it is included to help understand the operation of the converter. Timing intervals are measured from the leading edge of the Encode Command supplied by the user; this makes it easier to establish appropriate system timing.

Note the trailing edge of each clock pulse occurs after its corresponding serial output information has changed. If the serial output of the HAS-1204 converter is the desired signal, the trailing edges of clock pulses should be used as the stobes.

To assure the serial output data are fully established, the user is urged to incorporate a delay of approximately 30 nanoseconds between the trailing clock edge and the latch. This compensates for latch setup time, and slight variations in timing between the clock pulses and their associated data.

The portion of Figure 1 pertaining to Data Ready timing shows it returns to the digital "0" state 10 nanoseconds before the track-and-hold switches from "hold" to "track". The trailing edge of clock pulse #12 and the "track" transition are time-coincident, so this change in Data Ready occurs 10 nanoseconds before the trailing edge of the last clock pulse; and at the same time as the Bit 12 data change.

Time coincidence between the change of the Data Ready pulse and the arrival of Bit 12 (LSB) data might seem to preclude using the Data Ready pulse as a strobe. Despite that initial impression, the trailing (falling) edge of the Data Ready is reconmended for strobing the parallel outputs into external circuits. This can be accomplished by using an inverter with a time delay  $(t_D)$  of the appropriate amount for the latch which is being used, as illustrated in Figure 2.

The timing elationships discussed above are generated internally because the clock pulses' rising edges control the switching. The 30-nanosecond width of each clock signal helps assure that its serial output data are firmly established by the time the clock's trailing edge arrives.

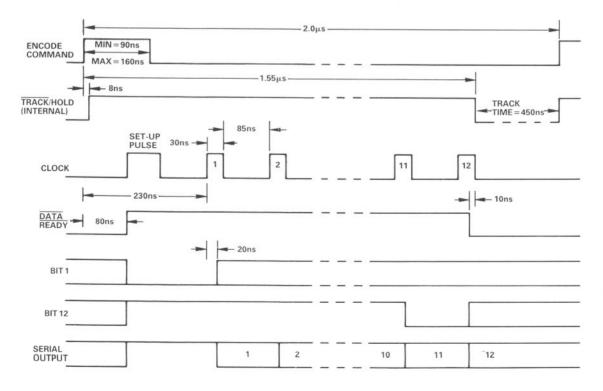


Figure 1. HAS-1204 Timing

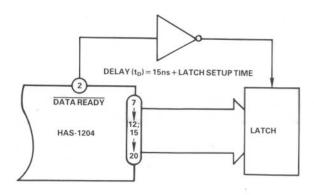


Figure 2. Output Strobe

### **APPLICATIONS INFORMATION**

Figures 3 and 4 provide needed details on the adjustment of controls for setting the amount of offset and gain.

As noted in both illustrations, the OFFSET control must be set first for proper performance of the converter. Since the HAS-1204 is capable of operating in either a unipolar or bipolar mode, OFFSET ADJUST and GAIN ADJUST include information

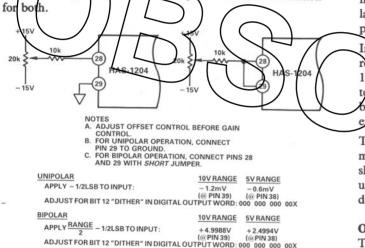
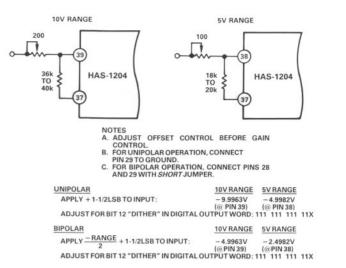


Figure 3. Offset Adjust



#### Figure 4. Gain Adust

However, careful adjustment of available controls is not the only way to help assure optimum performance. Like all high-speed, high-resolution components, the HAS-1204 is also sensitive to layout constraints. The use of a large, low-impedance ground plane is imperative.

In addition, bypass capacitors on the power supply leads are recommended. For most applications, electrolytic capacitors of 10-22 microfarads in parallel with ceramic capacitors of  $0.01\mu$ F to  $0.1\mu$ F will enhance the converter's effectiveness. These should be connected as closely as possible to the power supply pins entering the hybrid.

To prevent cross soupling of analog and digital signals which may "mask" lower order bits, analog and digital signal paths should be physically separated as much as possible. The user is urged to pay careful attention to both electrical and mechanical design to obtain best results.

### **ORDERING INFORMATION**

Two versions of HAS-1204 A/D Converters are available as standard products; both are housed in 40-pin hermetically-sealed metal packages. With the exception of operating temperatures, the specifications are the same for both units. For a temperature range of  $-25^{\circ}$ C to  $+85^{\circ}$ C, specify the model HAS-1204BM; for a range of  $-55^{\circ}$ C to  $+100^{\circ}$ C, order model number HAS-1204SM. Units screened to military requirements are also available; contact the factory for details.

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