

### General Description

The AOZ32034AQV is an integrated half-bridge solution with intelligent slew-rate control for wireless charger application. The device includes the high-side, low-side N-channel MOSFETs and its driver circuit. Typically, it's dedicated for the design of wireless charger transmitter circuit which is composed of full-bridge topology with resonant tank circuit to get best efficiency of power converter.

The AOZ32034AQV provides adjustable gate drive sink and source current control, by doing this control methodology, it's able to optimize EMI and driver losses to improve overall efficiency performance. Moreover, the features of AOZ32034AQV have multiple protection functions such as  $V_{CC}$  UVLO, over temperature protection to make the design more robust.

The AOZ32034AQV is available in a 4mm×4mm QFN-23L package and is rated over a -40°C to +125°C ambient temperature range.

### Features

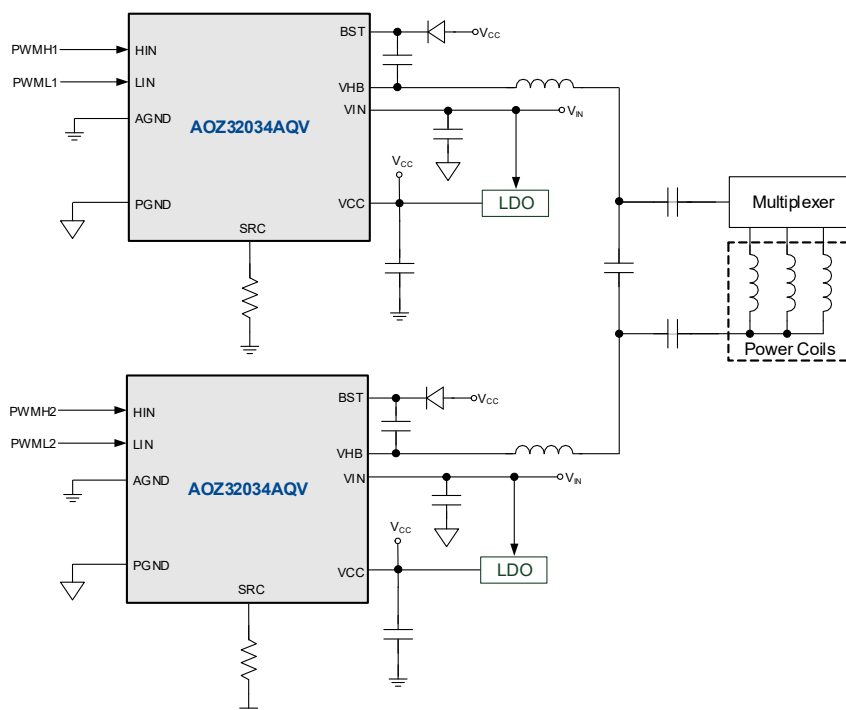
- Maximum input voltage 28V
  - Support 12V & 24V voltage rail system
- 15W~50W Coil Driver
  - For wireless charger transmitter circuit
- Slew-rate control to improve EMI performance
- Low  $R_{DS(ON)}$  internal NFETs
  - 7.5mΩ for Both HS/LS
- Integrated Bootstrap Diode
- Support protections
  - OTP, UVLO
- Thermally enhanced 23-pin 4×4 QFN

### Applications

- Wireless charger TX



### Typical Application (Wireless Charger TX)



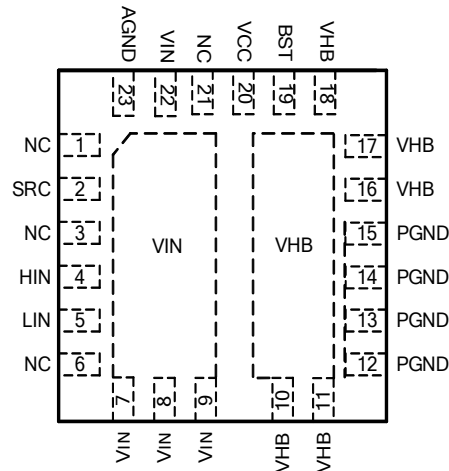
## Ordering Information

Part Number	Ambient Temperature Range	Package	Environmental
AOZ32034AQV	-40°C to +125°C	23-Pin 4x4 QFN	Green



All AOS products are offered in packages with Pb-free plating and compliant to RoHS standards. Please visit <http://www.aosmd.com/media/AOSGreenPolicy.pdf> for additional information.

## Pin Configuration



**AOZ32034AQV**  
23-Pin 4mm x 4mm QFN

## Pin Description

Pin Number	Pin Name	Pin Function
1	NC <sup>(1)</sup>	No connect
2	SRC	Slew-Rate Control to Adjust Driver Speed of Internal MOSFET
3	NC	No connect
4	HIN	PWM Input for High-Side MOSFET
5	LIN	PWM Input for Low-Side MOSFET
6	NC	No connect
7, 8, 9, 22	VIN	Supply Input. All IN pins must be connected together
10, 11, 16, 17, 18	VHB	Switching Node for Half-Bridge. All VHB must be connected together
12, 13, 14, 15	PGND	Power Ground
19	BST	Bootstrap Capacitor Connection. Connect an external capacitor between BST and VHB for supplying high-side MOSFET
20	VCC	Supply Input for Analog Functions. Bypass VCC to AGND with a 0.1uF~10uF ceramic capacitor and as close to VCC pin as possible
21	NC	No connect
23	AGND	Analog Ground

**Note:**

1. NC must be floating.

**Absolute Maximum Ratings<sup>(1)</sup>**

Parameter	Rating
V <sub>IN</sub> to AGND	-0.3V to 30V
V <sub>HB</sub> to AGND	-0.3V to 30V
B <sub>ST</sub> to AGND	-0.3V to 40V
B <sub>ST</sub> to V <sub>HB</sub>	-0.3 to 6V
S <sub>RC</sub> , V <sub>CC</sub> to AGND	-0.3V to 6V
P <sub>GND</sub> to AGND	-0.3V to +1V
Junction Temperature (T <sub>J</sub> )	+150°C
Storage Temperature (T <sub>S</sub> )	-65°C to +150°C
ESD Rating	±2kV

**Notes:**

1. Exceeding the Absolute Maximum ratings may damage the device.
2. The device is not guaranteed to operate beyond the Maximum Operating ratings.

**Recommend Operating Ratings<sup>(2)</sup>**

Parameter	Rating
Supply Voltage (V <sub>IN</sub> )	3.8V to 28V
Supply Voltage (V <sub>CC</sub> )	4.75V to 5.5V
Ambient Temperature (T <sub>A</sub> )	-40°C to +125°C
Package Thermal Resistance ( $\theta_{JA}$ ) ( $\theta_{JC}$ )	40°C/W 0.6°C/W

**Electrical Characteristics**

T<sub>A</sub> = -40°C to 125°C unless otherwise specified.

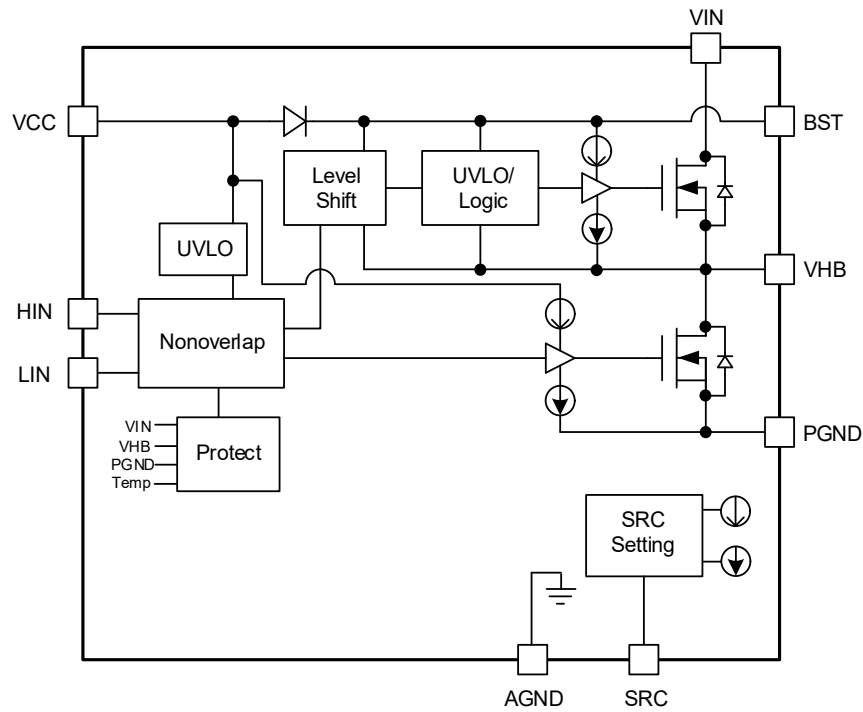
Symbol	Parameter	Conditions	Min	Typ.	Max.	Units
V <sub>UVLO_R</sub>	V <sub>CC</sub> UVLO Rising	V <sub>IN</sub> = 12V, V <sub>CC</sub> increase, Monitor S <sub>RC</sub> from low to high		4.3		V
V <sub>UVLO_F</sub>	V <sub>CC</sub> UVLO Falling	V <sub>IN</sub> = 12V, V <sub>CC</sub> decrease, Monitor S <sub>RC</sub> from high to low		4.2		V
V <sub>BST_UVLO_R</sub>	V <sub>BST</sub> -V <sub>HB</sub> UVLO Rising	V <sub>IN</sub> = 20V, (V <sub>BST</sub> -V <sub>HB</sub> ) increase, Monitor V <sub>HB</sub> from low to high		4.3		V
V <sub>BST_UVLO_F</sub>	V <sub>BST</sub> -V <sub>HB</sub> UVLO Falling	V <sub>IN</sub> = 20V, (V <sub>BST</sub> -V <sub>HB</sub> ) decrease, Monitor V <sub>HB</sub> from high to low		4.2		V
I <sub>VIN_QC</sub>	I <sub>VIN</sub> Quiescent Current	V <sub>IN</sub> = 12V, V <sub>CC</sub> = 5V, H <sub>IN</sub> = L <sub>IN</sub> = 0V, S <sub>RC</sub> = 100kΩ		20		μA
I <sub>VCC_QC</sub>	I <sub>VCC</sub> Quiescent Current	V <sub>IN</sub> = 12V, V <sub>CC</sub> = 5V, H <sub>IN</sub> = L <sub>IN</sub> = 0V, S <sub>RC</sub> = 100kΩ		180		μA
I <sub>BST-VHB_QC</sub>	I <sub>BST-VHB</sub> Quiescent Current	H <sub>IN</sub> /L <sub>IN</sub> = 0V, V <sub>HB</sub> = 1V, (V <sub>BST</sub> -V <sub>HB</sub> ) = 5V, Monitor (V <sub>BST</sub> -V <sub>HB</sub> ) Current			40	mA
V <sub>H<sub>IN</sub>_L</sub>	H <sub>IN</sub> /L <sub>IN</sub> Logic Low Voltage	V <sub>IN</sub> = 12V	0.7	0.9	1.1	V
V <sub>H<sub>IN</sub>_H</sub>	H <sub>IN</sub> /L <sub>IN</sub> Logic High Voltage	V <sub>IN</sub> = 12V	1.2	1.4	1.6	V
R <sub>H<sub>IN</sub>_IN</sub>	H <sub>IN</sub> /L <sub>IN</sub> Input Pull Low Impedance			280		kΩ
t <sub>H<sub>IN</sub>_RP</sub>	H <sub>IN</sub> Rising Propagation Delay	V <sub>IN</sub> = 10V, V <sub>CC</sub> = 5V, S <sub>RC</sub> = 20kΩ, V <sub>HB</sub> to GND = 100Ω, H <sub>IN</sub> = Low to High, Monitor V <sub>HB</sub> Low to High	62	87	110	ns
t <sub>H<sub>IN</sub>_FP</sub>	H <sub>IN</sub> Falling Propagation Delay	V <sub>IN</sub> = 10V, V <sub>CC</sub> = 5V, S <sub>RC</sub> = 20kΩ, V <sub>HB</sub> to GND = 100Ω, H <sub>IN</sub> = High to Low, Monitor V <sub>HB</sub> High to Low	60	86	110	ns
t <sub>L<sub>IN</sub>_RP</sub>	L <sub>IN</sub> Rising Propagation Delay	V <sub>IN</sub> = 10V, V <sub>CC</sub> = 5V, S <sub>RC</sub> = 20kΩ, V <sub>HB</sub> to GND = 100Ω, L <sub>IN</sub> = Low to High, Monitor V <sub>HB</sub> High to Low	72	91	105	ns
t <sub>L<sub>IN</sub>_FP</sub>	L <sub>IN</sub> Falling Propagation Delay	V <sub>IN</sub> = 10V, V <sub>CC</sub> = 5V, S <sub>RC</sub> = 20kΩ, V <sub>HB</sub> to GND = 100Ω, L <sub>IN</sub> = High to Low, Monitor V <sub>HB</sub> Low to High	50	75	95	ns

## Electrical Characteristics

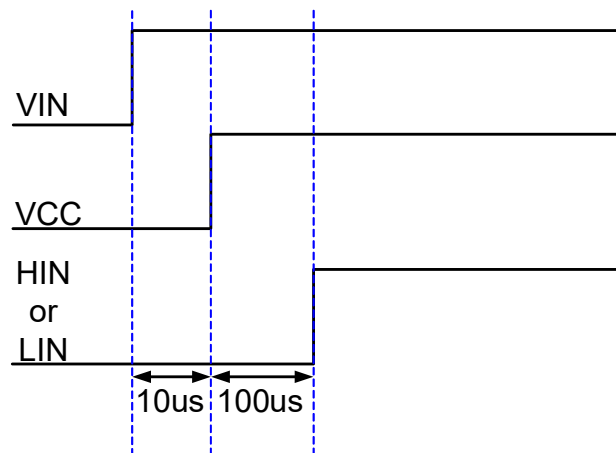
$T_A = -40^{\circ}\text{C}$  to  $125^{\circ}\text{C}$  unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ.	Max.	Units
$T_{DM\_R}$	Delay Matching Rising	Difference between $t_{HIN\_RP}$ and $t_{LIN\_RP}$		4		ns
$T_{DM\_F}$	Delay Matching Falling	Difference between $t_{HIN\_FP}$ and $t_{LIN\_FP}$		11		ns
$V_{SRC}$	SRC	$V_{IN}=12\text{V}$ , $V_{CC}=5\text{V}$ , $SRC=20\text{k}\Omega$	0.97	1	1.03	V
$I_{SRC\_MIN}$	SRC Min. Source Current	$V_{IN}=12\text{V}$ , $V_{CC}=5\text{V}$ , $SRC=4\text{V}$		0.5		$\mu\text{A}$
$I_{SRC\_MAX}$	SRC Max. Source Current	$V_{IN}=12\text{V}$ , $V_{CC}=5\text{V}$ , $SRC=0.8\text{V}$		140		$\mu\text{A}$
$SR_{HIN\_R}$	HIN Rising Slew Rate ( $SRC=20\text{k}\Omega$ )	$V_{IN}=10\text{V}$ , $V_{CC}=5\text{V}$ , $V_{HB}$ to GND= $100\Omega$ , HIN=Low to High, Monitor $V_{HB}$ Rising Slew Rate		0.4		V/ns
$SR_{HIN\_F}$	HIN Falling Slew Rate ( $SRC=20\text{k}\Omega$ )	$V_{IN}=10\text{V}$ , $V_{CC}=5\text{V}$ , $V_{HB}$ to GND= $100\Omega$ , HIN=High to Low, Monitor $V_{HB}$ Falling Slew Rate		0.5		V/ns
$SR_{LIN\_R}$	LIN Rising Slew Rate ( $SRC=20\text{k}\Omega$ )	$V_{IN}=10\text{V}$ , $V_{CC}=5\text{V}$ , $V_{HB}$ to $V_{IN}=100\Omega$ , LIN=High to Low, Monitor $V_{HB}$ Rising Slew Rate		0.5		V/ns
$SR_{LIN\_F}$	LIN Falling Slew Rate ( $SRC=20\text{k}\Omega$ )	$V_{IN}=10\text{V}$ , $V_{CC}=5\text{V}$ , $V_{HB}$ to $V_{IN}=100\Omega$ , LIN=Low to High Monitor $V_{HB}$ Falling Slew Rate		0.4		V/ns
$R_{H\_ON}$	$V_{IN}-V_{HB}$ RON	$V_{IN}=12\text{V}$ , $V_{CC}=5\text{V}$ , $HIN=5\text{V}$ , $(V_{BST}-V_{HB})=5\text{V}$ , $I_{V_{HB}}=1\text{A}$		7.5		$\text{m}\Omega$
$R_{L\_ON}$	$V_{HB}-\text{PGND}$ RON	$V_{IN}=12\text{V}$ , $V_{CC}=5\text{V}$ , $LIN=5\text{V}$ , $\text{PGND}=0$ , $I_{V_{HB}}=1\text{A}$		7.5		$\text{m}\Omega$
$V_{SD}$	Boost Diode Forward Voltage	Forward Current = 2mA		0.42		V
$T_{OTP}$	OTP	$V_{IN}=12\text{V}$ , $V_{CC}=5\text{V}$		150		$^{\circ}\text{C}$

### Functional Block Diagram



### Start-up Sequence



## Typical Performance Characteristics

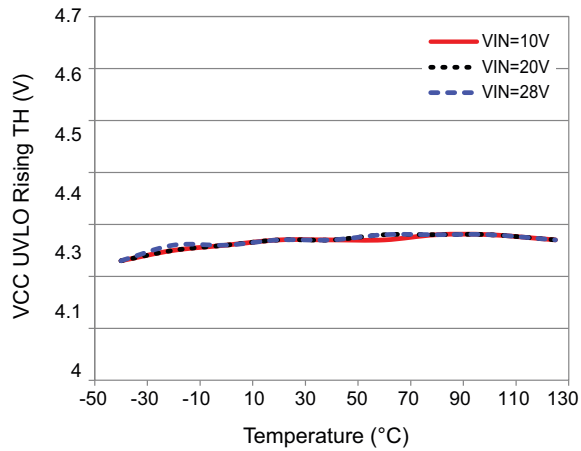


Figure 1. VCC UVLO Rising Threshold

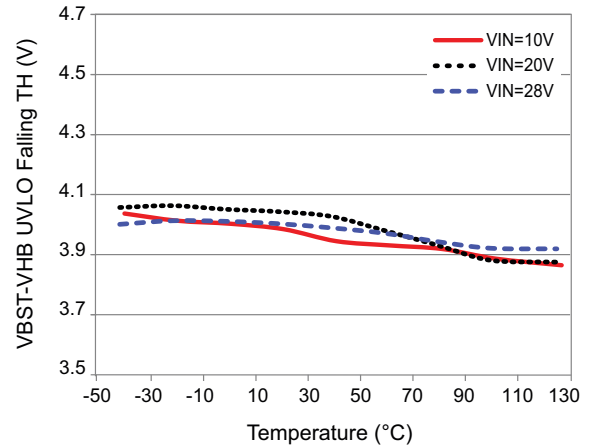


Figure 2.  $V_{BST-VHB}$  UVLO Falling Threshold

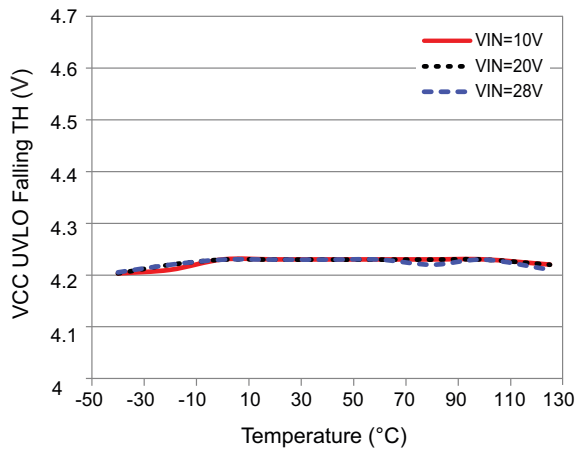


Figure 3. VCC UVLO Falling Threshold

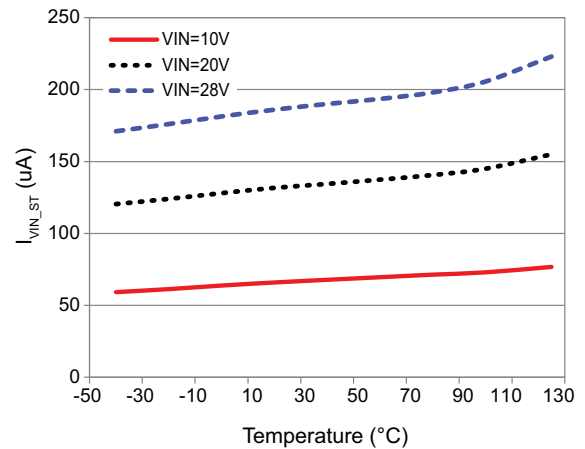


Figure 4. Input Standby Current

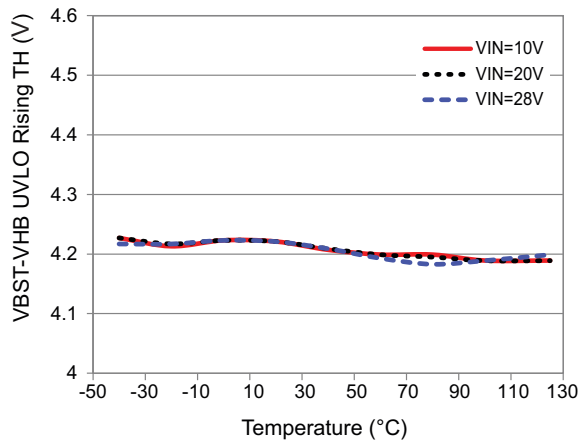


Figure 5.  $V_{BST-VHB}$  UVLO Rising Threshold

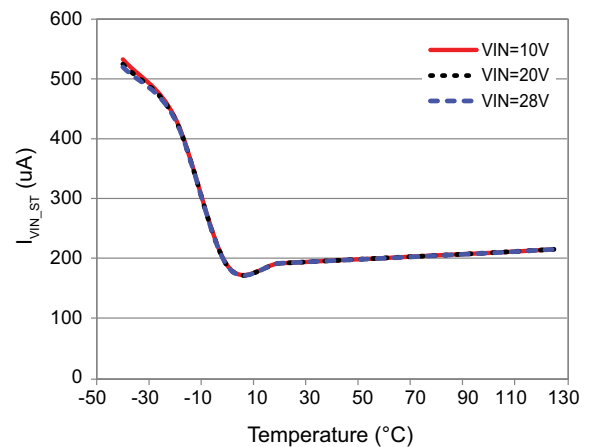


Figure 6.  $V_{CC}$  Standby Current

Typical Performance Characteristics (Continued)

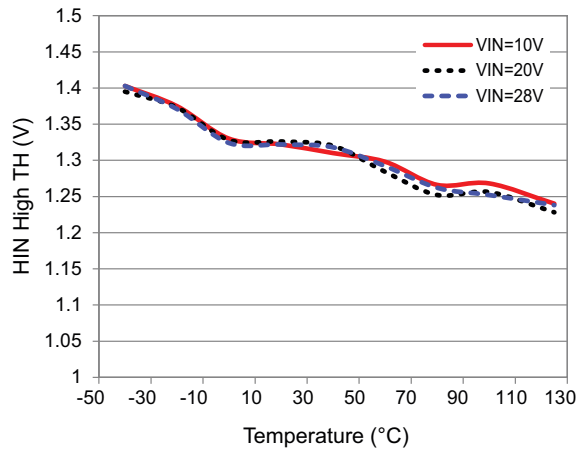


Figure 7. HIN High Threshold

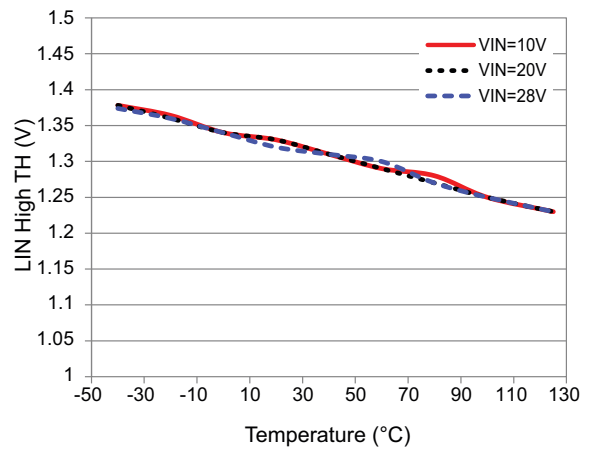


Figure 8. LIN High Threshold

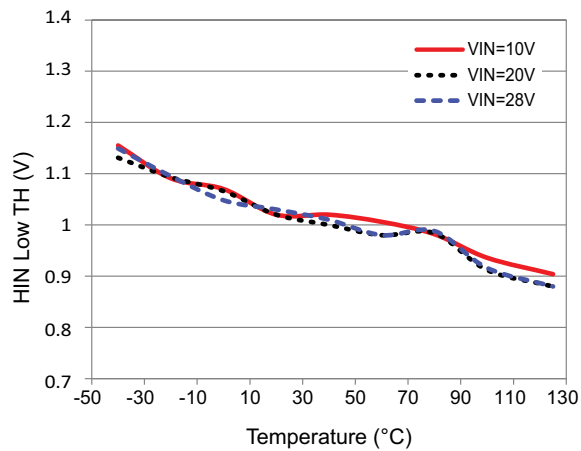


Figure 9. HIN Low Threshold

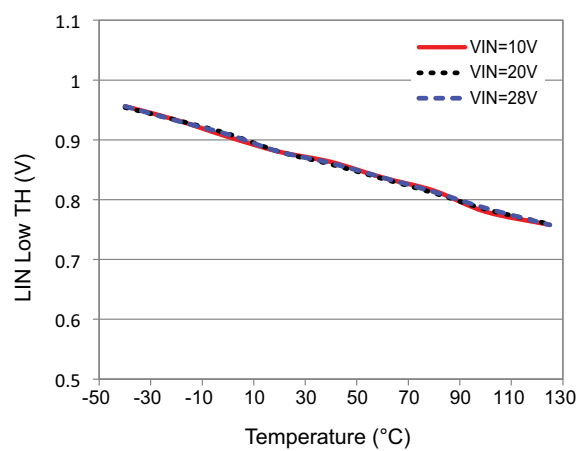


Figure 10. LIN Low Threshold

## Detailed Description

The AOZ32034AQV is an integrated half-bridge solution with intelligent slew-rate control for wireless charger application. The device includes the high-side, low-side N-channel MOSFETs and its driver circuit. Typically, it's dedicated for the design of wireless charger transmitter circuit which is composed of full-bridge topology with resonant tank circuit to get best efficiency of power converter.

The AOZ32034AQV provides adjustable gate drive sink and source current control, by doing this control methodology, it's able to optimize EMI and driver losses to improve overall efficiency performance.

In addition, the AOZ32034AQV provides several fault protections, such as UVLO, OTP and non-overlapping mechanism.

The AOZ32034AQV is available in 23-pin 4mm×4mm QFN package.

### Non-overlapping

For forbidding shoot-through, HIN or LIN is invalid when HIN or LIN goes high state before other one. For example, low-side gate state keeps low regardless of the state of LIN when HIN is high at first, and vice versa.

### Fault Protection

In order to protect power MOSFETs, over temperature protection (OTP) is implemented. AOZ32034AQV will be shutdown when OTP is occurred until VCC is reset. The threshold of OTP is 140°C.

### Adjustable Source/Sink Current

It's hard to meet all of EMI specifications in different applications. So, AOZ32034AQV provides external adjustable resistors for tuning gate drive source and sink current.

SRC is used to tune gate drive source and sink current, respectively. A resistor connects between SRC pin and GND to setting gate drive source / sink current by internal current mirror, as illustrated Figure 11. Source and sink current use maximum capability to drive when SRC pin is floating or the voltage on SRC pin is exceed 4V. The suggestion range of  $R_{SRC}$  is 10kΩ~100kΩ.

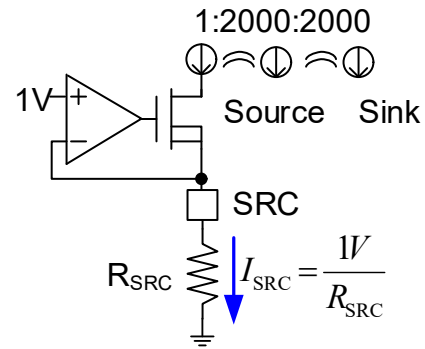


Fig. 11 Source/Sink Current Setting

In addition, source and sink current controls are implemented only during MOSFET Miller effect and  $V_{GS} > 1V$ , as illustrated Figure 12.

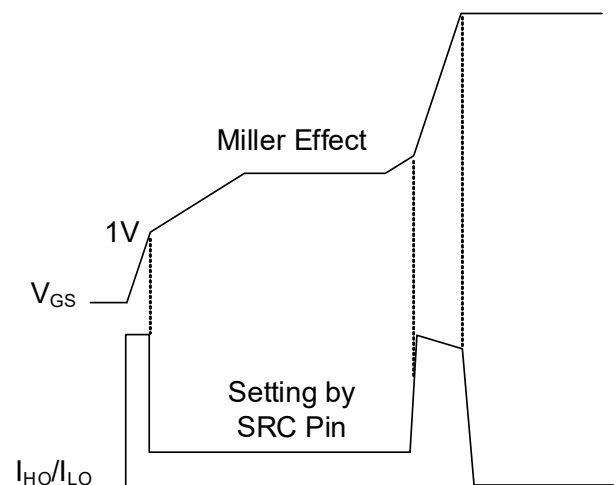


Fig. 12 Source /Sink Current Implement Waveform



## Layout Considerations

Several layout tips are listed below for the best electric and thermal performance.

1. The VIN pins and pad are connected to internal high side switch drain. They are also low resistance thermal conduction path. Connected a large copper plane to VIN pins to help thermal dissipation.
2. Input capacitors should be connected to the VIN pins and the PGND pins as close as possible to reduce the switching spikes.
3. The VHB pins and pad are connected to internal low side switch drain. They are low resistance thermal conduction path and most noisy switching node. Connected a large copper plane to VHB pins to help thermal dissipation.
4. Decoupling capacitor  $C_{VCC}$  should be connected to VCC and AGND as close as possible.
5. Bootstrap capacitor  $C_B$  should be connected to BST and VHB as close as possible.
6. A ground plane is preferred. PGND and AGND must be connected to the ground plane through vias.
7. Keep sensitive signal traces such as feedback trace and digital signals far away from the VHB pins.

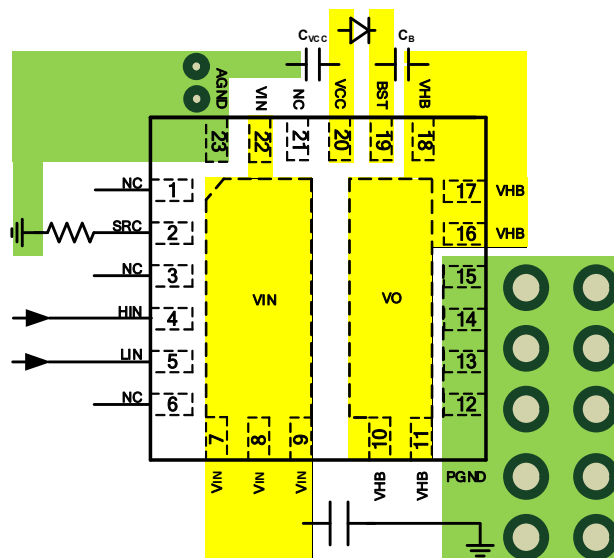
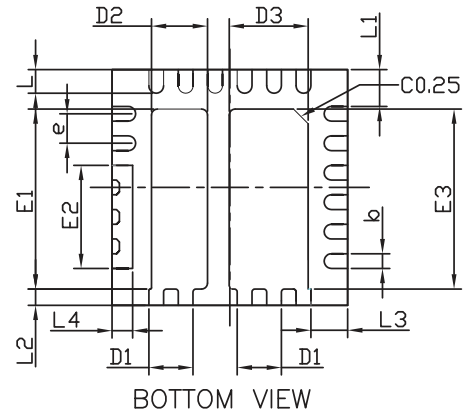
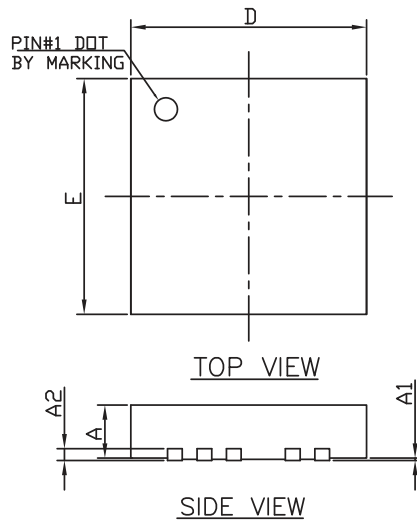


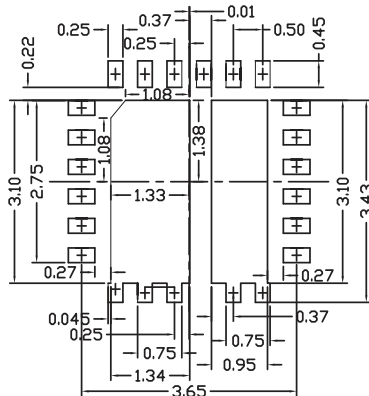
Figure 13. Layout Placement

Package Dimensions, QFN4x4-23L

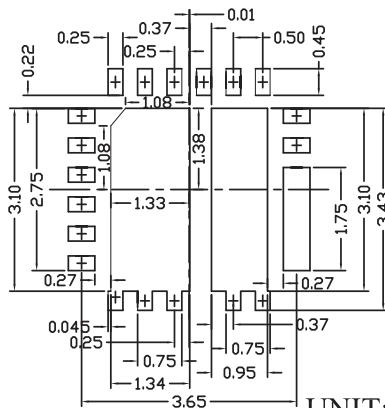


RECOMMENDED LAND PATTERN

Option 1



Option 2



UNIT: mm

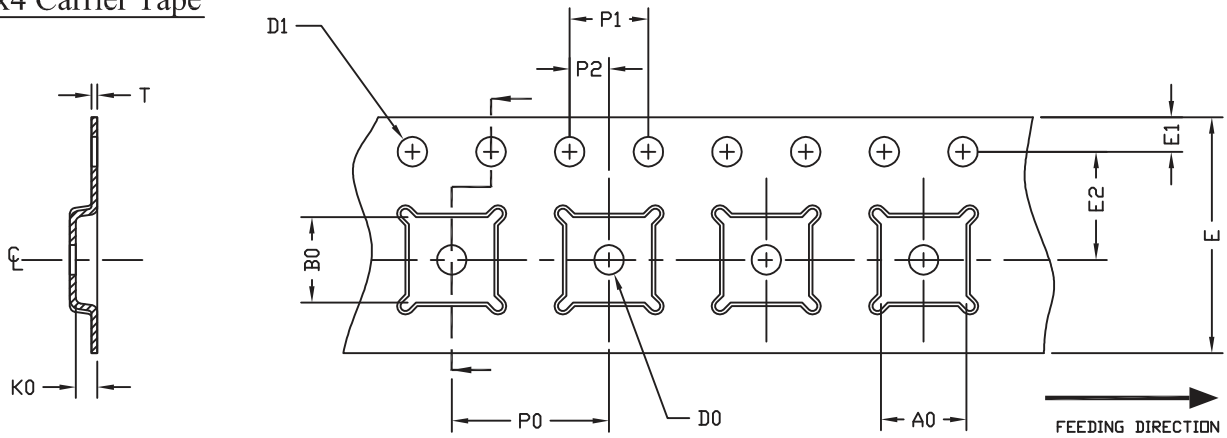
SYMBOLS	DIMENSIONS IN MILLIMETERS			DIMENSIONS IN INCHES		
	MIN	NOM	MAX	MIN	NOM	MAX
A	0.80	0.90	1.00	0.031	0.035	0.039
A1	0.00	—	0.05	0.000	—	0.002
A2	0.2 REF			0.008 REF		
E	3.90	4.00	4.10	0.153	0.157	0.161
E1	2.95	3.05	3.15	0.116	0.120	0.124
E2	1.65	1.75	1.85	0.065	0.069	0.073
E3	2.95	3.05	3.15	0.116	0.120	0.124
D	3.90	4.00	4.10	0.153	0.157	0.161
D1	0.65	0.75	0.85	0.026	0.030	0.034
D2	0.85	0.95	1.05	0.033	0.037	0.041
D3	1.24	1.34	1.44	0.049	0.053	0.057
L	0.35	0.40	0.45	0.014	0.016	0.018
L1	0.57	0.62	0.67	0.022	0.024	0.026
L2	0.23	0.28	0.33	0.009	0.011	0.013
L3	0.57	0.62	0.67	0.022	0.024	0.026
L4	0.30	0.35	0.40	0.012	0.014	0.016
b	0.20	0.25	0.30	0.008	0.010	0.012
e	0.50 BSC			0.020 BSC		

NOTE

1. CONTROLLING DIMENSION IS MILLIMETER.  
CONVERTED INCH DIMENSIONS ARE NOT NECESSARILY EXACT.
2. TOLERANCE :±0.05 UNLESS OTHERWISE SPECIFIED.
3. RADIUS ON ALL CORNER ARE 0.152 MAX., UNLESS OTHERWISE SPECIFIED.
4. PACKAGE WARPAGE: 0.012 MAX.
5. NO ANY PLASTIC FLASH ALLOWED ON THE TOP AND BOTTOM LEAD SURFACE.
6. PAD PLANARITY: ±0.102
7. CRACK BETWEEN PLASTIC BODY AND LEAD IS NOT ALLOWED.

### Tape and Reel Dimensions, QFN4x4-23L

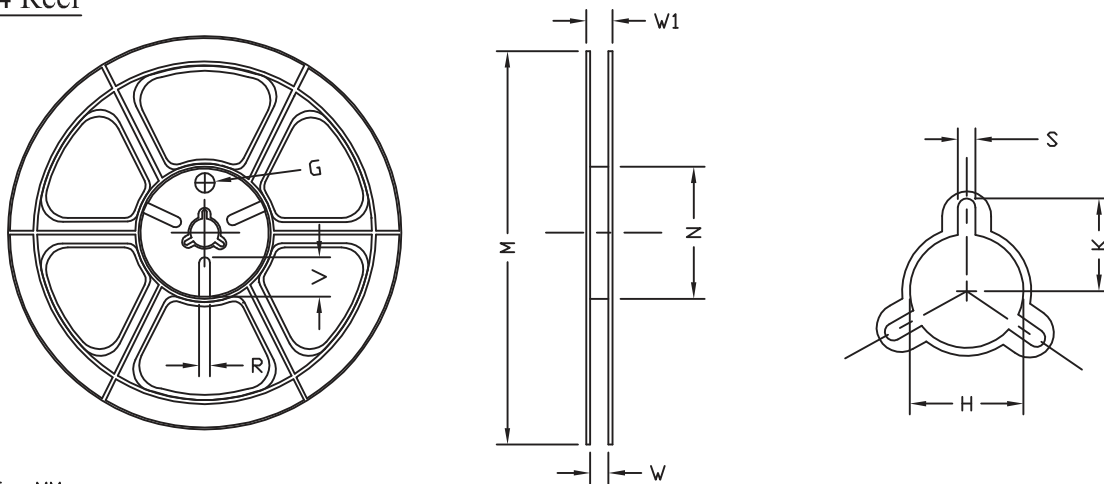
#### QFN4x4 Carrier Tape



UNIT: MM

PACKAGE	A0	B0	K0	D0	D1	E	E1	E2	P0	P1	P2	T
QFN4x4 (12 mm)	4.35 ±0.10	4.35 ±0.10	1.10 ±0.10	1.50 MIN.	1.50 <sup>+0.1</sup> -0.0	12.0 ±0.3	1.75 ±0.10	5.50 ±0.05	8.00 ±0.10	4.00 ±0.10	2.00 ±0.05	0.30 ±0.05

#### QFN4x4 Reel



UNIT: MM

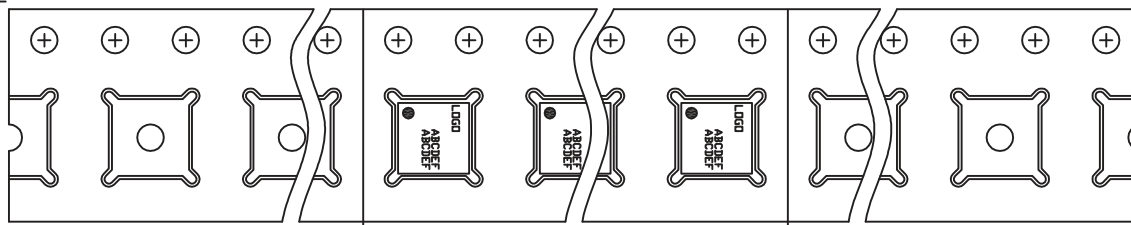
TAPE SIZE	REEL SIZE	M	N	W	W1	H	K	S	G	R	V
12 mm	ø330	ø330.0 ±2.0	ø79.0 ±1.0	12.4 <sup>+2.0</sup> -0.0	17.0 <sup>+2.6</sup> -1.2	ø13.0 ±0.5	10.5 ±0.2	2.0 ±0.5	---	---	---

#### QFN4x4 Tape

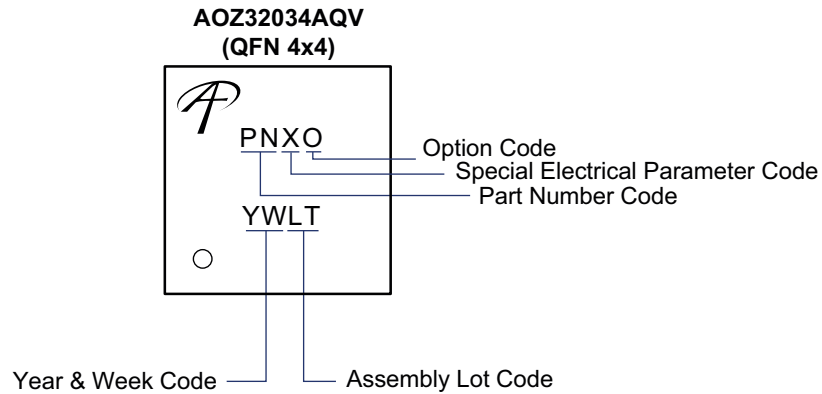
Leader / Trailer  
& Orientation

Normal

Unit Per Reel:



**Part Marking**



PART NO.	DESCRIPTION	CODE
AOZ32034AQV	Green Product	DU00

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2. A critical component in any component of a life support, device, or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.